Cheat Sheet for EE463

Performance Parameters

$$FormFactor = \frac{V_{rms}}{V_{avg}}$$

$$CrestFactor = \frac{V_{peak}}{V_{rms}}$$

$$DistortionFactor = \frac{I_{1rms}}{I_{rms}}$$

 ϕ : phase difference between fundamentals of current and voltage $DisplacementPowerFactor = \cos(\phi)$

$$TruePowerFactor = \frac{P}{S} = DPF \frac{I_{1,RMS}}{I_{RMS}}$$

$$THD = \sqrt{(\frac{I_{rms}}{I_{1rms}})^2 - 1}$$

Single Phase Diode Rectifier

$$V_{av} = \frac{2\sqrt{2}V_s}{\pi}$$

u: commutation period

$$\cos(u) = 1 - \frac{2\omega L_s I_d}{\sqrt{2}V_s}$$

$$I_{d,avg} = \frac{\int_b^f i(\theta)d\theta}{\pi}$$

$$I_{d,shortcircuit} = \frac{V_s}{\omega L_s}$$

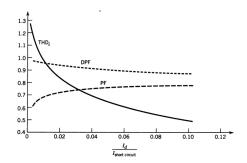


Figure 1: Characteristics of source current wrt w*Ls (battery on load side)

Three Phase Rectifier

• Half Wave

$$V_{av} = \frac{3\sqrt{6}V_s}{2\pi}$$

Crossing points (integration) on the waves are from $\pi/6$ to $5\pi/6$

• Full Wave

Full Bridge Rectifier Average Output V_s :rms value of source voltage

$$V_{av} = \frac{3\sqrt{6}V_s}{\pi} - \frac{3wL_sI_d}{\pi}$$

Single Phase Controlled Rectifiers-Thyristors

 \bullet Idealized Circuit α : firing angle

$$V_{av}(\alpha) = \frac{2\sqrt{2}V_d}{\pi} \cdot \cos \alpha$$

• Effect of Ls

$$\cos(\alpha + u) = \cos(\alpha) - \frac{2\omega L_s I_d}{\sqrt{2}V_s}$$
$$V_d = 0.9V_s \cos(\alpha) - \frac{2\omega L_s I_d}{\pi}$$

• Inverter Mode

$$ExtinctionAngle = \gamma = 180 - (\alpha + u)$$

Three Phase Controlled Rectifiers-Thyristors

• Idealized Circuit α : firing angle

$$V_{av}(\alpha) = \frac{3\sqrt{2}V_{LL}}{\pi} \cdot \cos \alpha$$

• Effect of Ls

$$\cos(\alpha + u) = \cos(\alpha) - \frac{2\omega L_s I_d}{\sqrt{2}V_{LL}}$$
$$V_d = \frac{3\sqrt{2}}{\pi}V_{LL}\cos(\alpha) - \frac{3\omega L_s I_d}{\pi}$$

• Output Waveforms

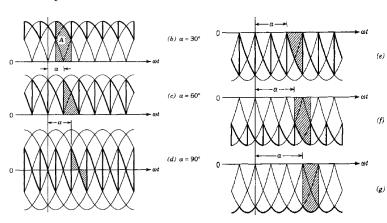


Figure 2: Output Voltage Waveforms

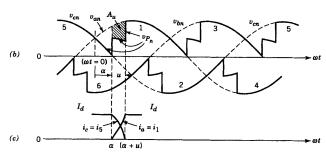


Figure 6-25 Commutation in the presence of L_s .

Figure 3: Effect of Commutation



Figure 6-30 Waveforms in a discontinuous-current-conduction mode.

Figure 4: Discontinuous Current Conduction Mode

Trigonometric

$$\sin A \cos B = \frac{1}{2} \left[\sin(A - B) + \sin(A + B) \right]$$

$$\sin A \sin B = \frac{1}{2} \left[\cos(A - B) - \cos(A + B) \right]$$

$$\cos A \cos B = \frac{1}{2} \left[\cos(A - B) + \cos(A + B) \right]$$

$$\cos A + \cos B = 2 \cos(\frac{A + B}{2}) \cos(\frac{A - B}{2})$$

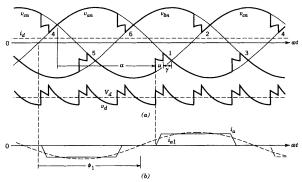


Figure 6-32 Waveforms in the inverter of Fig. 6-31

Figure 5: Inverter Waveforms

$$\cos A - \cos B = -2\sin(\frac{A+B}{2})\sin(\frac{A-B}{2})$$
$$\sin^2(A) = \frac{1}{2} - \frac{\cos(2A)}{2}$$

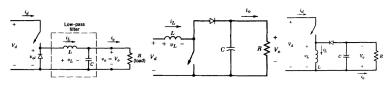
Symmetry	Condition Required	a_h and b_h
Even	f(-t) = f(t)	$b_h = 0$ $a_h = \frac{2}{\pi} \int_0^{\pi} f(t) \cos(h\omega t) d(\omega t)$
Odd	f(-t) = -f(t)	$a_h = 0$ $b_h = \frac{2}{\pi} \int_0^{\pi} f(t) \sin(h\omega t) d(\omega t)$
Half-wave	$f(t) = -f(t + \frac{1}{2}T)$	$a_h = b_h = 0$ for even h $a_h = \frac{2}{\pi} \int_0^{\pi} f(t) \cos(h\omega t) d(\omega t)$ for odd h $b_h = \frac{2}{\pi} \int_0^{\pi} f(t) \sin(h\omega t) d(\omega t)$ for odd h

Figure 6: Fourier Transform Table

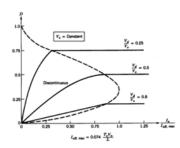
Power Flow

$$P = V_s I_{s1} cos(\phi) = V_d I_d$$

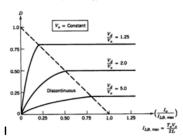
$$S = V_s I_s$$

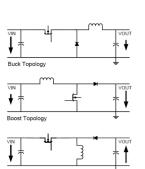


In order to keep Vo constant



With Constant Vo (Vd can vary)



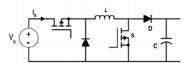


Convection Thermal Resistance Conduction Heat Loss Fluid Temperature Ris

Ah
$$R$$
 Radiation Heat Loss (Black body radiation) $\Delta T = rac{P}{Q.d.C}$

 q_R : nouronnext took (Wm2) $q_R =
ho \epsilon F(T_1^4 - T_2^4)$ $\zeta = rac{2F}{2F}$

Positive Output Buck-Boost Converter



BUCK

$$\frac{I_o}{I_d} = \frac{V_d}{V_o} = \frac{1}{D} \qquad I_o = i_{L,peak} \frac{D + \Delta_1}{2}$$

$$I_{LB} = \frac{T_s V_d}{2L} D(1 - D) : \frac{V_o}{V_d} = \frac{D}{D + \Delta_1} \Delta_1 = \frac{I_o}{4I_{LB,max}D} \frac{V_o}{V_d} = \frac{D^2}{D^2 + \frac{1}{4}(I_o I_{LB,max})}$$

$$I_{LB} = \frac{1}{2} i_{L,peak} = \frac{t_{on}}{2L} (V_d - V_o) = \frac{DT_s}{2L} (V_d - V_o) = I_{oB}$$

$$I_{LB} = \frac{T_s V_o}{2L} (1 - D) \quad D = \frac{V_o}{V_c} \left(\frac{I_o / I_{LB,max}}{1 - V_o / V_c}\right)^{1/2}$$

BOOST

$$\frac{V_o}{V_d} = \frac{T_s}{t_{\text{off}}} = \frac{1}{1 - D} \qquad I_{oB} = \frac{T_s V_o}{2L} D(1 - D)^2 \qquad I_{LB, \text{max}} = \frac{T_s V_o}{8L}$$

$$I_{oB, \text{max}} = \frac{2}{27} \frac{T_s V_o}{L} \qquad I_{oB} = \frac{27}{4} D(1 - D)^2 I_{oB, \text{max}} \qquad \therefore \frac{V_o}{V_d} = \frac{\Delta_1 + D}{\Delta_1}$$

$$I_d = \frac{V_d}{2L} D T_s (D + \Delta_1) \qquad I_o = \left(\frac{T_s V_d}{2L}\right) D \Delta_1 \qquad D = \left[\frac{4}{27} \frac{V_o}{V_d} \left(\frac{V_o}{V_d} - 1\right) \frac{I_o}{I_{oB, \text{max}}}\right]^{1/2}$$

$$\therefore \frac{\Delta V_o}{V_o} = \frac{DT_s}{RC}$$

$$\begin{split} V_{ripple,buck} &= V_o (1-D) (\frac{f_c}{f_s})^2 \frac{\pi^2}{2} \\ I_{ob} &= \frac{T_s V_o D (1-D)^2}{2L} \\ V_{ripple,boost} &= \frac{V_o D T s}{RC} \end{split}$$

Snubber Circuit

$$fr = \frac{1}{2\pi\sqrt{LpCp}}$$

$$R = \sqrt{\frac{Li}{Ci}}$$

$$\frac{LpI^2}{V^2} < Cs < \frac{ton}{10R}$$

$$Pr = CsV^2fs$$