

BLG354E / CRN: 21560 6th Week Lecture (cont.)

1

Implementation of Discrete Time Systems

Transfer function of DT-LTI system is given as,

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_0 + b_1 z^{-1} + \dots + b_m z^{-m}}{a_0 + a_1 z^{-1} + \dots + a_n z^{-n}}$$

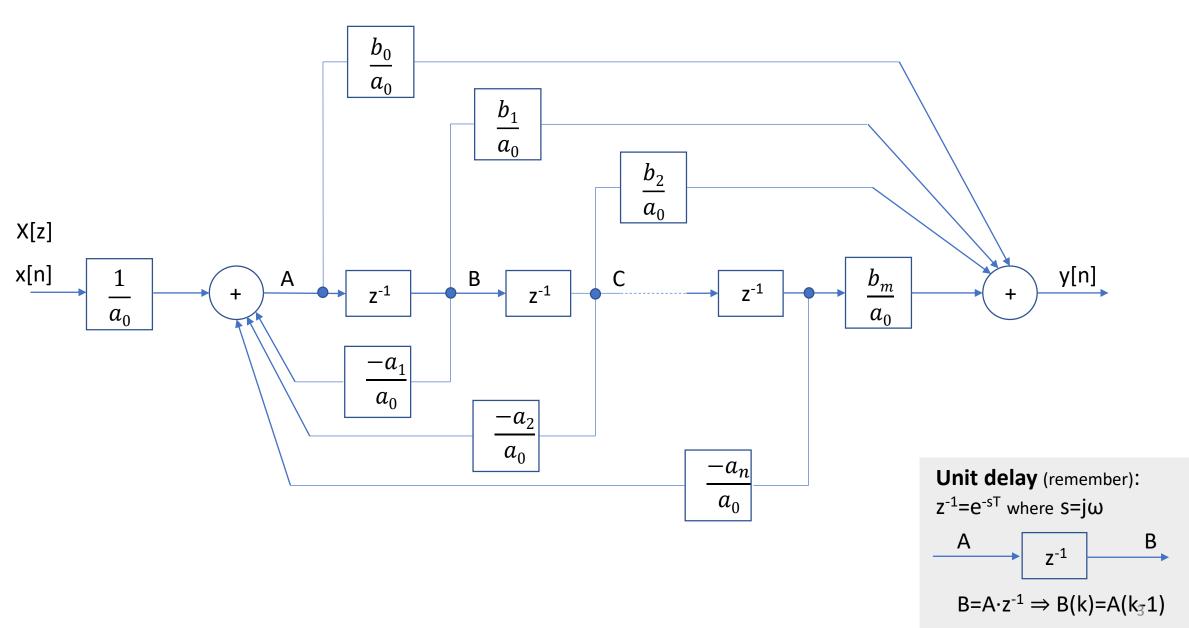
$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_m z^{-m}}{a_0 + a_1 z^{-1} + \dots + a_n z^{-n}} \cdot \frac{A(z)}{A(z)} = \frac{Y(z)}{X(z)}$$

$$Y(z) = \frac{1}{a_0} (b_0 + b_1 z^{-1} \dots + b_m z^{-m}) A(z)$$

$$A(z) = \frac{1}{a_0} X(z) - \frac{1}{a_0} (a_1 z^{-1} + a_2 z^{-2} + \dots + a_n z^{-n}) A(z)$$

Direct Programming

(canonical form)



Example:

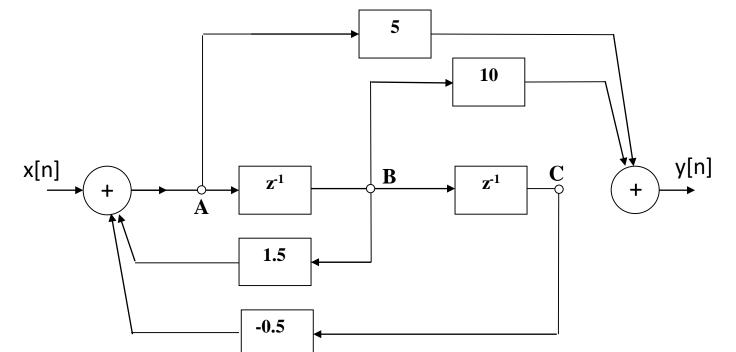
$$H(z) = \frac{Y(z)}{X(z)} = \frac{5(1+2z^{-1})}{(1-z^{-1})(1-0.5z^{-1})}$$

Write the Psuedo code H(z) for the signal sampled at f_s

$$H(z) = \frac{Y(z)}{X(z)} = \frac{(5+10z^{-1})A(z)}{(1-1.5z^{-1}+0.5z^{-2})A(z)}$$

$$Y(z) = (5 + 10 z^{-1}) A(z)$$

$$A(z) = X(z) + 1.5 z^{-1} A(z) - 0.5z^{-2} A(z)$$



Pseudo code: Timer Interrupt @Ts=1/f_s

X = READ (ADC)

A = X + 1.5 B - 0.5 C

Y = 5 A + 10 B

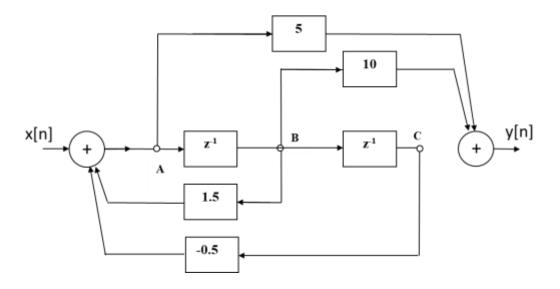
Output (Y)

C=B

B=A

Return

Proof:



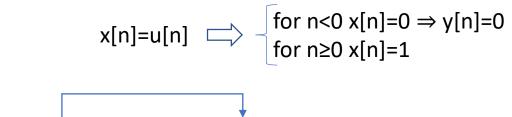
$$Y=5A+10B=$$

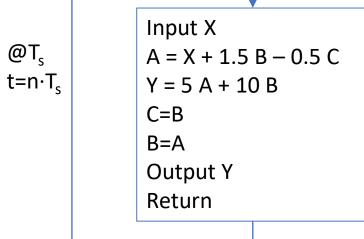
=5A+10Az⁻¹=(5+10z⁻¹)A

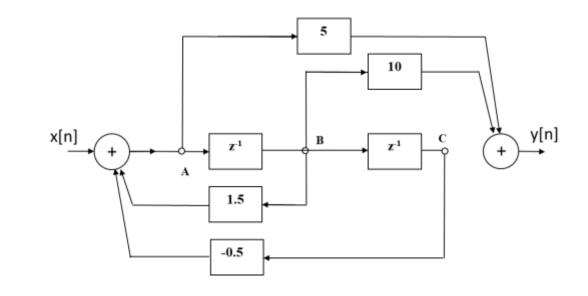
A=X+1.5B-0.5C
A=X+1.5Az⁻¹-0.5Az⁻²
$$\Rightarrow$$
 $A = \frac{X}{1-1.5z^{-1}+0.5z^{-2}}$

$$Y = \frac{\left(5 + 10z^{-1}\right)X}{1 - 1.5z^{-1} + 0.5z^{-2}} \implies T(z) = \frac{Y(z)}{X(z)} = \frac{\left(5 + 10z^{-1}\right)}{\left(1 - 1.5z^{-1} + 0.5z^{-2}\right)}$$

Step Response:







	_				
n	x[n]	Α	B /	C	y[n]
0	1	1	0	0	5
1	1	2.5	1	0	22.5
2	1	4.25	2.5	1	46.25
3	1	6.125	4.25	2.5	73.125

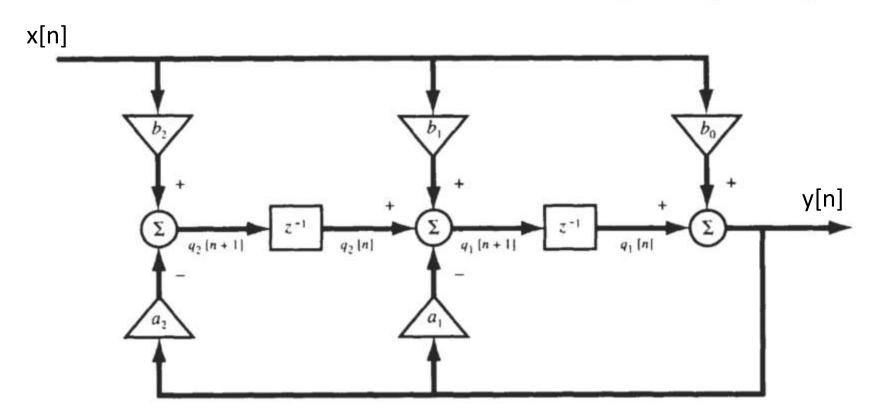
Initial condition

Canonical Simulation (First form)

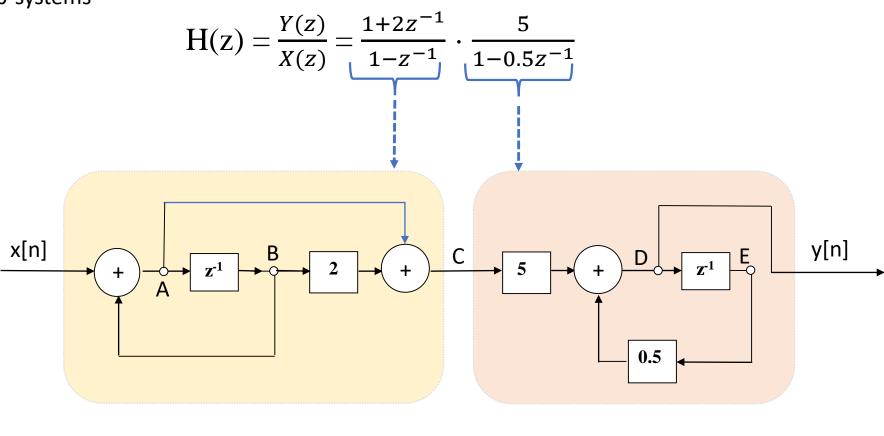
$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$

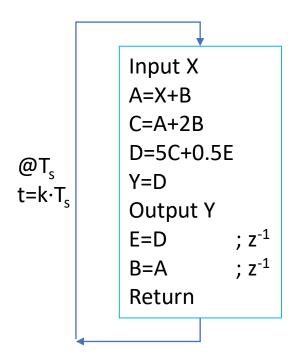
$$(1+a_1z^{-1}+a_2z^{-2})Y(z)=(b_0+b_1z^{-1}+b_2z^{-2})X(z)$$

$$Y(z) = -a_1 z^{-1} Y(z) - a_2 z^{-2} Y(z) + b_0 X(z) + b_1 z^{-1} X(z) + b_2 z^{-2} X(z)$$

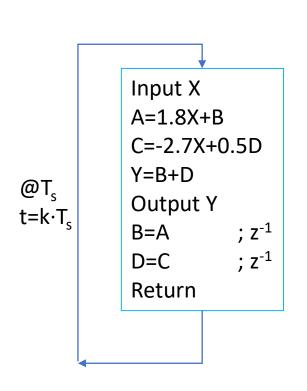


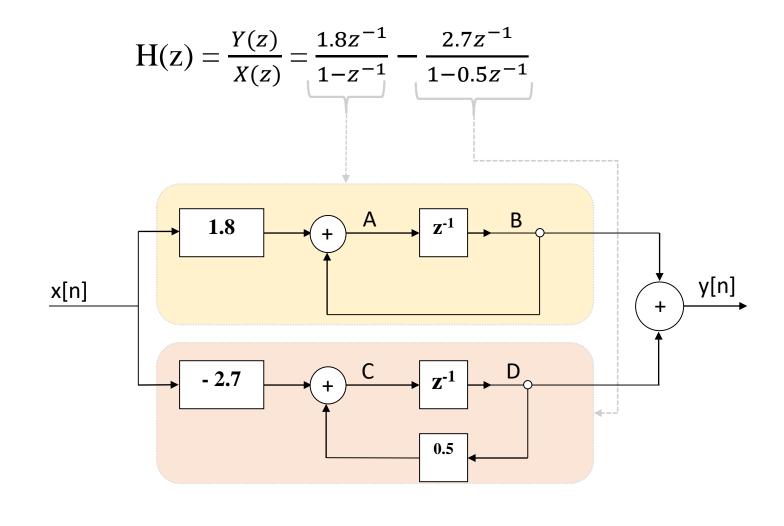
Programming of Cascaded sub-systems





Programming of Parallel sub-systems

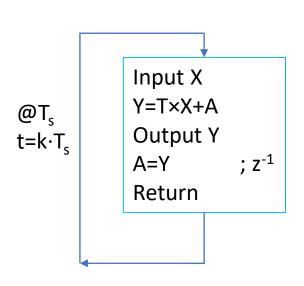


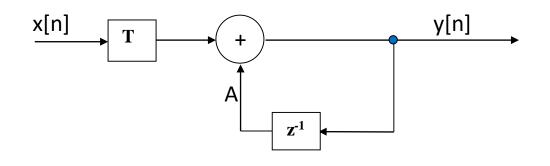


Example:

Digital integrator without delay:

$$H(z) = \frac{Y(z)}{X(z)} = T\left(\frac{1}{1-z^{-1}}\right)$$

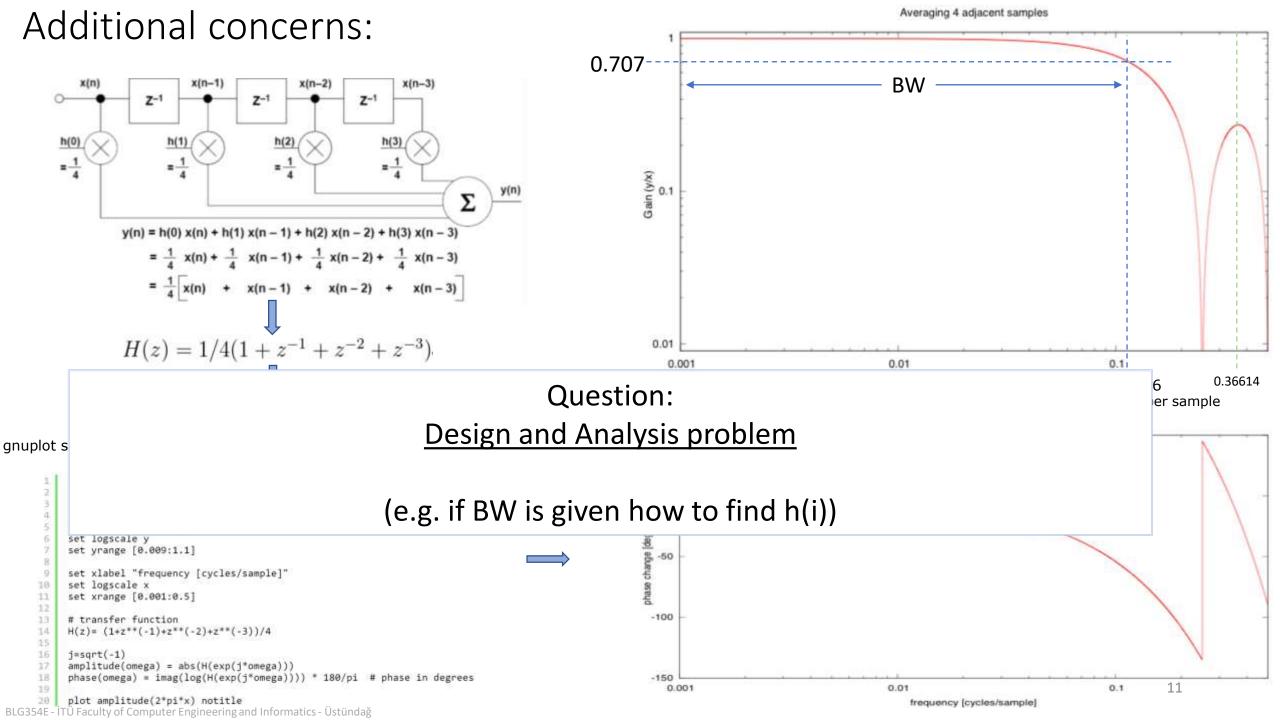




Initial condition: If y(kT)=0 for k<0 then A=0

$$Y[n]=y(kT)=T[x(0)+x(T)+x(2T)+...+x(kT)]$$

- 1) Simulate the system for $x[n]=u[n-2]-u[n-4]+\delta[n-3]$
- 2) What are the differences between integrator with and without Delay?
- 3) Write the pseudo code for digital integrator with delay



Programming with the assistance of ChatGPT3:

Question to ChatGPT3:



could you please write the transfer function of a second order butterworth type low pass digital filter in z-domain where the cut off frequency is 100Hz.



 $H(z) = (b0 + b1z^{-1} + b2z^{-2}) / (1 + a1z^{-1} + a2z^{-2})$

with a cutoff frequency of 100Hz can be written as:

where the coefficients b0, b1, b2, a1, and a2 are determined as follows:

$$b0 = (w0^2) / (4 + 2w0 + 2w0^2)$$

$$b1 = (2w0^2) / (4 + 2w0 + 2w0^2)$$

$$b2 = (w0^2) / (4 + 2w0 + 2w0^2)$$

$$a1 = (-8 + 2w0^2) / (4 + 2w0 + 2w0^2)$$

$$a2 = (4 - 2w0 + 2w0^2) / (4 + 2w0 + 2w0^2)$$

where w0 = tan(pi*fc/fs), where fc is the cutoff frequency (100Hz in this case) and fs is the sampling frequency.

So, if we assume a sampling frequency of 1000Hz, we can calculate the coefficients as:

$$w0 = tan(pi*100/1000) = 0.157$$

$$b0 = 0.002484$$

$$b1 = 0.004968$$

$$b2 = 0.002484$$

$$a1 = -1.6907$$

$$a2 = 0.7328$$

And the transfer function becomes:

$$H(z) = (0.002484 + 0.004968z^{-1} + 0.002484z^{-2}) / (1 - 1.6907z^{-1} + 0.7328z^{-2})$$

I hope this helps! Let me know if you have any further questions.



could you please write me a python code that draws the bode plot for the transfer function you provided above.



Sure! Here's a Python code snippet that uses the `scipy` library to generate the bode plot for the transfer function of a second order Butterworth low pass digital filter with a cutoff frequency of 100Hz:

```
Copy code
python
import numpy as np
import matplotlib.pyplot as plt
from scipy.signal import freqz
# Define the coefficients of the transfer function
b = [0.002484, 0.004968, 0.002484]
a = [1.0, -1.6907, 0.7328]
# Calculate the frequency response of the filter
w, h = freqz(b, a)
# Convert the angular frequency to Hz
f = w / (2*np.pi)
# Calculate the magnitude and phase of the frequency response
mag = 20*np.log10(np.abs(h))
phase = np.unwrap(np.angle(h)) * 180 / np.pi
# Plot the magnitude and phase responses
fig, ax1 = plt.subplots()
ax2 = ax1.twinx()
ax1.semilogx(f, mag, 'b')
ax2.semilogx(f, phase, 'r')
ax1.set_xlabel('Frequency (Hz)')
ax1.set_ylabel('Magnitude (dB)', color='b')
ax2.set_ylabel('Phase (degrees)', color='r')
plt.title('Bode Plot')
plt.show()
```

import numpy as np import matplotlib.pyplot as plt from scipy.signal import freqz

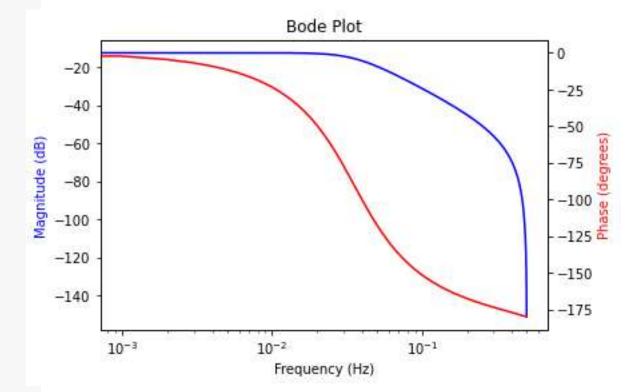
Define the coefficients of the transfer function b = [0.002484, 0.004968, 0.002484] a = [1.0, -1.6907, 0.7328]

Calculate the frequency response of the filter w, h = freqz(b, a)

Convert the angular frequency to Hz f = w / (2*np.pi)

Calculate the magnitude and phase of the frequency response mag = 20*np.log10(np.abs(h)) phase = np.unwrap(np.angle(h)) * 180 / np.pi

Plot the magnitude and phase responses fig, ax1 = plt.subplots() ax2 = ax1.twinx() ax1.semilogx(f, mag, 'b') ax2.semilogx(f, phase, 'r') ax1.set_xlabel('Frequency (Hz)') ax1.set_ylabel('Magnitude (dB)', color='b') ax2.set_ylabel('Phase (degrees)', color='r') plt.title('Bode Plot') plt.show()



scipy.signal.freqz

scipy.signal.freqz(b, a=1, worN=512, whole=False, plot=None, fs=6.283185307179586, include_nyquist=False)

Compute the frequency response of a digital filter.

Given the M-order numerator b and N-order denominator a of a digital filter, compute its frequency response:

$$jw$$
 $-jw$ $-jwM$
 jw $B(e)$ $b[0] + b[1]e$ $+ ... + b[M]e$
 $H(e) = \frac{-jw}{-jw}$ $-jwN$
 $A(e)$ $a[0] + a[1]e$ $+ ... + a[N]e$

Parameters: **b** : array_like

Numerator of a linear filter. If b has dimension greater than 1, it is assumed that the coefficients are stored in the first dimension, and b.shape[1:], a.shape[1:], and the shape of the frequencies array must be compatible for broadcasting.

a : array_like

Denominator of a linear filter. If b has dimension greater than 1, it is assumed that the coefficients are stored in the first dimension, and b.shape[1:], a.shape[1:], and the shape of the frequencies array must be compatible for broadcasting.

worN: {None, int, array_like}, optional

If a single integer, then compute at that many frequencies (default is N=512). This is a convenient alternative to:

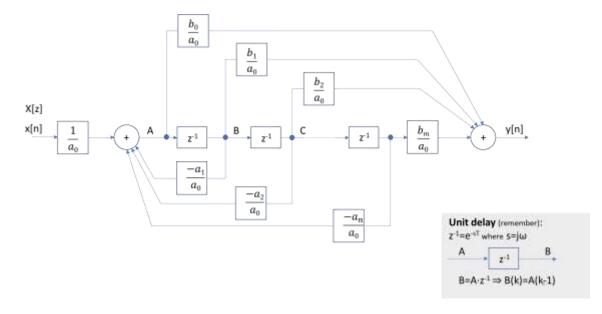
```
np.linspace(0, fs if whole else fs/2, N, endpoint=include_nyquist)
```

Using a number that is fast for FFT computations can result in faster computations (see Notes).

If an array_like, compute the response at the frequencies given. These are in the same units as fs.



Write the Pseudo code for real-time operation on an embedded system where the sampling period is 1000Hz





could you please write the pseudo code that performs the transfer function H(z) you provided above



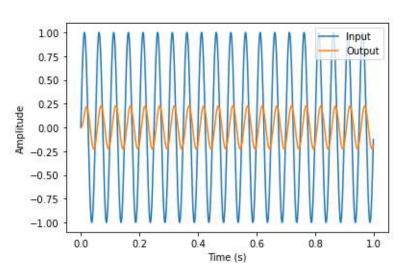
Sure, here's a pseudo code that performs the transfer function of a second order Butterworth low pass digital filter in z-domain with a cutoff frequency of 100Hz using the transfer function:

```
Copy code
// Initialize the filter coefficients
b0 = 0.002484
b1 = 0.004968
b2 = 0.002484
a1 = -1.6907
a2 = 0.7328
// Initialize the filter states
z1 = 0.0
z2 = 0.0
// Filter input samples
function butterworth_lpf(x):
    // Calculate output sample y
   y = b0*x + b1*z1 + b2*z2
    // Update filter states
    z2 = z1
    z1 = x - a1*z1 - a2*z2
    // Return output sample y
    return y
// Test the filter with a sine wave input
fs = 1000 // Sample rate
f = 50 // Input signal frequency
N = 1000 // Number of samples
t = linspace(0, 1, N)
x = \sin(2*pi*f*t)
// Filter the input signal
y = zeros(N)
for i = 0 to N-1:
    y[i] = butterworth_lpf(x[i])
```

Python code



What is the inconsistency?



How can it be corrected?

```
import numpy as np
# Define the filter coefficients
b = [0.002484, 0.004968, 0.002484]
a = [1.0, -1.6907, 0.7328]
# Initialize the filter states
z1 = 0.0
z2 = 0.0
v1 = 0.0
y2 = 0.0
# Filter input samples
def butterworth lpf(x):
  global z1, z2, y1, y2
  y = b[0]*x + b[1]*z1 + b[2]*z2 - a[1]*y1 - a[2]*y2
  z^2 = z^1
  z1 = x
  y2 = y1
  y1 = y
  return y
# Test the filter with a sine wave input
fs = 1000 # Sample rate
f = 20 # Input signal frequency
t = np.arange(0, 1, 1/fs)
x = np.sin(2*np.pi*f*t)
# Filter the input signal
y = np.zeros like(x)
for i in range(len(x)):
  y[i] = butterworth_lpf(x[i])
# Plot the input and output signals
import matplotlib.pyplot as plt
plt.plot(t, x, label='Input')
plt.plot(t, y, label='Output')
plt.xlabel('Time (s)')
plt.ylabel('Amplitude')
plt.legend()
plt.show()
```