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## **Second Order RCL Circuit**

Applying KVL to an RCL series circuit we get

$$v_L(t) + v_R(t) + v_C(t) = v(t)$$

where

$$v_L(t) = L\frac{d}{dt}i(t), \quad v_R(t) = Ri(t), \quad v_C(t) = \frac{Q(t)}{C}$$

Substituting

$$i(t) = \frac{dQ(t)}{dt} = C\frac{d}{dt}v_C(t)$$

into the equation above we get

$$LC\frac{d^2}{dt^2}v_C(t) + RC\frac{d}{dt}v_C(t) + v_C(t) = v(t)$$

or

$$\frac{d^{2}}{dt^{2}}v_{C}(t) + \frac{R}{L}\frac{d}{dt}v_{C}(t) + \frac{1}{LC}v_{C}(t) = \ddot{v}_{C}(t) + 2\zeta\omega_{n}\dot{v}_{C}(t) + \omega_{n}^{2}v_{C}(t) = \omega_{n}^{2}v(t)$$

where

$$\omega_n = \frac{1}{LC}, \qquad \zeta = \frac{R}{2} \sqrt{\frac{C}{L}}$$

are respectively the damping coefficient and natural frequency of the second order system. To find the homogeneous solution we substitute  $v_C(t) = e^{st}$  into the homogeneous DE and get

$$(s^2 + 2\zeta\omega_n s + \omega_n^2)e^{st} = 0$$

Solving the algebraic equation  $\,s^2+2\zeta\omega_n s+\omega_n^2=0\,$  we get the two roots

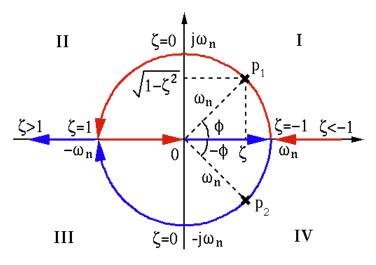
$$s_{1,2} = (-\zeta \pm \sqrt{\zeta^2 - 1})\omega_n = (-\zeta \pm j\sqrt{1 - \zeta^2})\omega_n = -\omega_n e^{\mp j\phi}$$

where

$$\phi = \tan^{-1} \left( \frac{\sqrt{1 - \zeta^2}}{\zeta} \right)$$

The two roots can be represented as the root locus on the complex plane, for a fixed value of  $\omega_n$ .

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The homogeneous solution takes the following form:

$$v_C(t) = Ae^{s_1t} + Be^{s_2t}$$

As in the RCL circuit  $\zeta \geq 0$  and  $\omega_n \geq 0$ , we only need to consider the root locus in the II and III quadrants in the figure above. Specifically we consider the following two cases:

•  $0 \le \zeta < 1$ . The system is underdampled.

$$s_{1,2} = -\zeta \omega_n \pm j \sqrt{1 - \zeta^2} \omega_n = -\zeta \omega_n \pm j \omega_d$$

where  $\omega_d = \sqrt{1-\zeta^2}\omega_n$  is the damped nature frequency. Crrespondingly

$$e^{st} = e^{-\zeta \omega_n t} e^{\pm j\omega_d t} = e^{-\zeta \omega_n t} (\cos \omega_d t \pm \sin \omega_d t)$$

is an exponentially decaying sinusoidal ossillation. In particular, when  $\zeta=0$ ,  $e^{st}=\cos\omega_n t\pm\sin\omega_n t$ , the system is undamped.

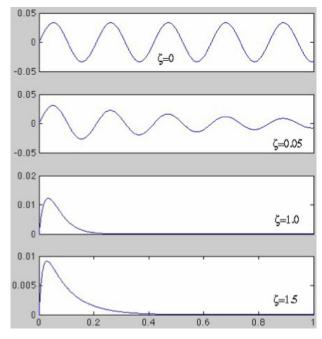
•  $\zeta \geq 1$ . The system is over-damped.

$$s_{1,2} = -(\zeta \pm \sqrt{\zeta^2 - 1})\omega_n < 0$$

and the crresponding solution

$$e^{st} = e^{-(\zeta \pm \sqrt{\zeta^2 - 1})\omega_n t}$$

decays exponentially. In particular, when  $\zeta=1$ , the system is critically damped.

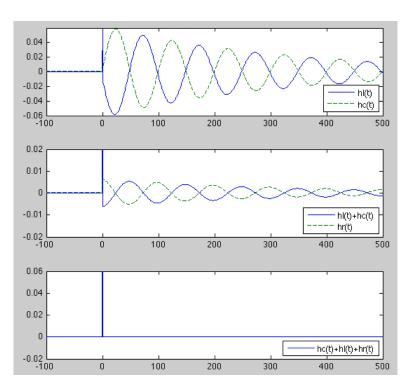


It can be shown that the impulse response of the system is

$$h_C(t) = \frac{\omega_n e^{-\zeta \omega_n t}}{\sqrt{1 - \zeta^2}} \sin(\omega_d t) u(t)$$

In particular, when R=0 and  $\zeta=0$ ,

$$h_C(t) = \omega_n \sin(\omega_n t) u(t)$$



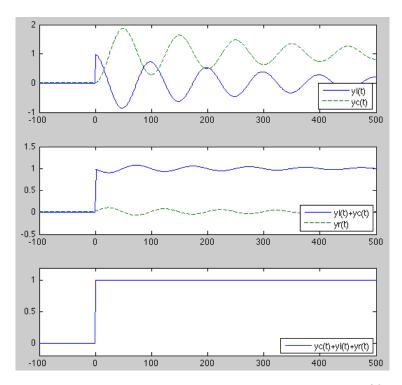
Impulse responses by R, L, and C of an RCL system. Top: impulse responses  $h_L(t)$  (solid curve) and  $h_C(t)$  (dashed curve); middle: their sum  $h_L(t) + h_C(t)$  (solid curve) and impulse response  $h_R(t)$  (dashed curve); bottom: the sum of all three:  $\delta(t) = h_L(t) + h_R(t) + h_C(t)$ .

The step response is

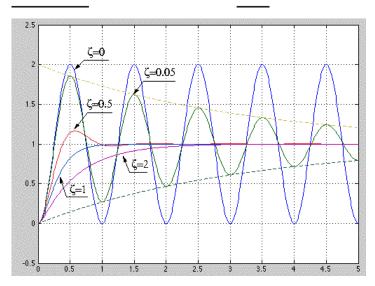
$$v_C(t) = \left[1 - \frac{e^{-\zeta \omega_n t}}{\sqrt{1 - \zeta^2}} \sin(\omega_d t + \phi)\right] u(t)$$

In particular, when R=0 ,  $\zeta=0$  and  $\phi=\pi/2$  , then the general form of the step response becomes

$$y(t) = \left[1 - \frac{e^{-\zeta \omega_n t}}{\sqrt{1 - \zeta^2}}\sin(\omega_d t + \phi)\right]u(t) = \left[1 - \sin(\omega_n t + \pi/2)\right]u(t) = \left[1 - \cos(\omega_n t)\right]u(t)$$



Step responses by R, L as well as C of an RCL system. Top: step responses  $v_L(t)$  (solid curve) and  $v_C(t)$  (dashed curve); middle: their sum  $v_L(t) + v_C(t)$  (solid curve) and step responses  $y_R(t)$  (dashed curve); bottom: the sum of all three:  $v_C(t) + v_L(t) + v_R(t) = u(t)$ .



Step response of second-order system for different  $\zeta$ . Step responses corresponding to five different values of  $\zeta$ : 0, 0.05, 0.5, 1, and 2. The envelope of the step response for  $\zeta = 0.05$  is also plotted to show the exponential decay of the sinusoid.

**Example** It is often desirable for a second order system to reach a steady state value, e.g.,  $\underline{v=2}$ , without overshoot. This can be achieved by applying first

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a step input  $\underline{u(t)}$  and then another one  $\underline{u(t-T/2)}$ , where  $\underline{T=2\pi/\omega_n}$ . The response of the first step input is  $\underline{[1-\cos(\omega_n t)]u(t)}$  and that to the second is

$$[1-\cos(\omega_n(t-T/2))]u(t-T/2) = [1-\cos(\omega_nt-\pi)]u(t-T/2) = [1+\cos(\omega_nt)]u(t-T/2)$$

As the sum of the two individual responses the total response for t > T/2 is 2, i.e.,

$$v_C(t) = \left\{ \begin{array}{ll} [1-\cos\omega_n t] u(t) & 0 < t < T/2 = \pi/\omega_n \\ 2 & t > T/2 = \pi/\omega_n \end{array} \right.$$

**Example** Consider the response  $y(t) = v_C(t)$  of the RCL system when R = 0.

• Find the step response y(t) of the RLC circuit (with zero initial conditions  $v_C(0) = \dot{v}_C(0) = 0$ ).

The general solution is the sum of the homogeneous solution  $A \sin \omega_n t + B \cos \omega_n t$  and the particular solution V = 1:

$$y(t) = 1 + A\sin\omega_n t + B\cos\omega_n t$$

Based on the zero initial conditions

$$y(0) = 1 + B = 0,$$
  $\dot{y}(0) = A\omega_n = 0$ 

we get A=0 and B=-1 and

$$y(t) = 1 - \cos \omega_n t$$

• Find the system's response to a square impulse

$$x(t) = \begin{cases} 1 & 0 \le t < t_0 \\ 0 & \text{else} \end{cases}$$

As  $\underline{x(t) = u(t) - u(t - t_0)}$ , the response is

$$y(t) = [[1 - \cos(\omega_n t)]u(t) - [1 - \cos(\omega_n (t - t_0))]u(t - t_0)]$$

i.e.,

$$y(t) = [-\cos(\omega_n t) + \cos(\omega_n (t - t_0))], \qquad t > t_0$$

• Find the impulse response h(t). The impulse input can be written as

$$\delta(t) = \lim_{t_0 \to 0} \begin{cases} 1/t_0 & 0 \le t < t_0 \\ 0 & \text{else} \end{cases}$$

When  $t_0 \to 0$ , we have first order approximations  $\cos(\omega_n t_0) \approx 1$ ,  $\sin(\omega_n t_0) \approx \omega_n t_0$ , and get

$$\cos(\omega_n(t-t_0)) = \cos(\omega_n t) \cos(\omega_n t_0) + \sin(\omega_n t) \sin(\omega_n t_0) \approx \cos(\omega_n t) + \sin(\omega_n t) \omega_n t_0$$

Substituting this into the equation above we get the impulse response

$$y(t) = \omega_n \sin(\omega_n t)$$

• When  $t_0 = T/2 = \pi/\omega_n$ , we have

$$cos(\omega_n(t-t_0)) = cos(\omega_n t - \pi) = -cos(\omega_n t)$$

and the response for  $\,t>T/2\,\,{\rm is}\,$ 

$$y(t) = -2\cos(\omega_n t)$$

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