CHAPTER 5

Transform Analysis of LTI Systems

Basic Problems

23. (a) Solution:

The frequency response is

$$H(e^{j\omega}) = \frac{b}{1 - 0.8e^{-j\omega} - 0.81e^{-2j\omega}}$$

(b) Solution:

b = 0.1702.

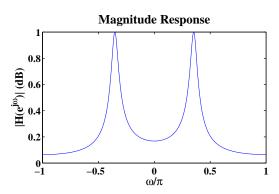


FIGURE 5.1: Magnitude response.

- (c) See plot below.
- (d) Solution:

$$y[n] = 2 \times 0.0577 \cos(\frac{\pi n}{3} + \frac{\pi}{4}) - 2 \times 0.0809 \sin(\frac{\pi n}{3} + \frac{\pi}{4})$$

(e) See plot below.

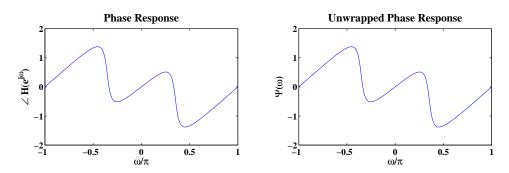


FIGURE 5.2: Wrapped and the unwrapped phase responses.

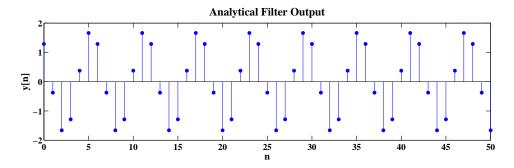


FIGURE 5.3: Analytical response y[n] to the input $x[n] = 2\cos(\pi n/3 + 45^{\circ})$.

The magnitude response is:

$$|H(e^{j\omega})| = \frac{|b|}{\sqrt{1+a^2+2a\cos 2\omega}}$$

The phase response is:

$$\angle H(e^{j\omega}) = \tan^{-1} \frac{a \sin 2\omega}{1 + a \cos 2\omega}$$

In order to constrain the maximum magnitude response equal to 1, we have

$$|b| = 1 - a$$

Hence, for a = 0.8, we choose b = 0.2.

(b) Solution:

(i) The output is:

$$y[n] = 3\cos\left(\frac{\pi}{2}n\right)$$

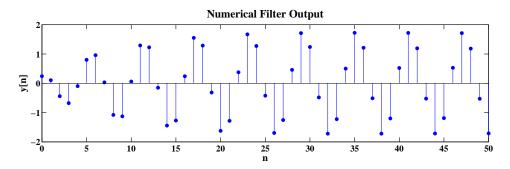


FIGURE 5.4: Numerical response y[n] to the input $x[n] = 2\cos(\pi n/3 + 45^{\circ})$.

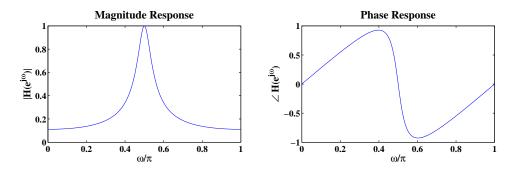


FIGURE 5.5: Magnitude and phase responses for a = 0.8.

(ii) The output is:

$$y[n] = \frac{1-a}{1+a^2} 3\sin\frac{\pi}{4}n + \frac{a(1-a)}{1+a^2} 3\cos\frac{\pi}{4}n$$

(c) See plots below.

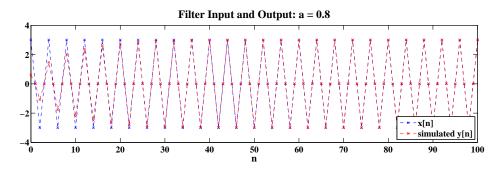


FIGURE 5.6: Plot for input x[n] and output y[n] when $x[n] = 3\cos(\pi n/2)$.

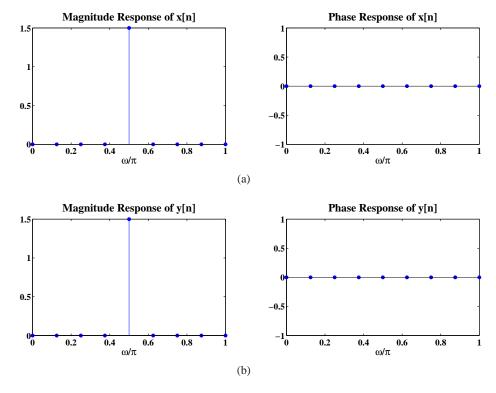


FIGURE 5.7: Magnitude and phase responses of (a) x[n]. (b) y[n].

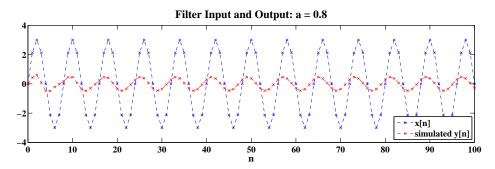


FIGURE 5.8: Plot for input x[n] and output y[n] when $x[n] = 3\sin(\pi n/4)$.

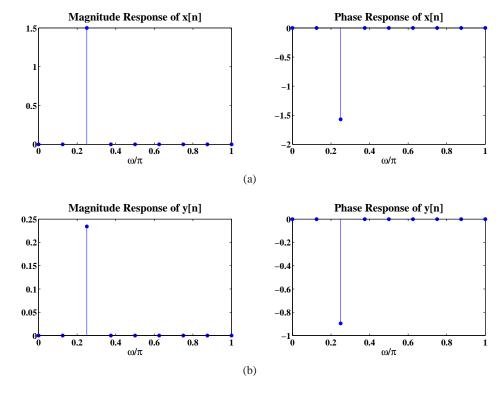


FIGURE 5.9: Magnitude and phase responses of (a) x[n]. (b) y[n].

The frequency response is:

$$H(e^{j\omega}) = \frac{e^{j\frac{\pi}{6}}}{1 - 0.8e^{-j(\omega - \frac{\pi}{4})}} + \frac{e^{-j\frac{\pi}{6}}}{1 - 0.8e^{-j(\omega + \frac{\pi}{4})}} + \frac{5}{1 + 0.9e^{-j\omega}}$$

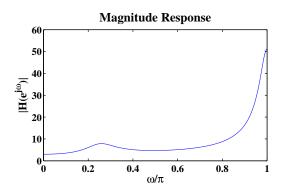


FIGURE 5.10: Magnitude response of the system.

(b) See plot below.

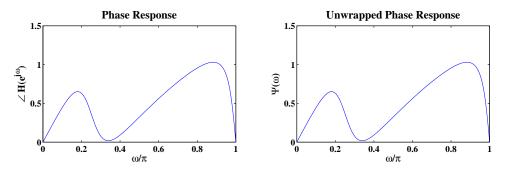


FIGURE 5.11: Wrapped and the unwrapped phase responses of the system.

(c) Solution:

The output is:

$$y[n] = H(e^{j0}) + \frac{3}{2}e^{j\frac{\pi}{6}}e^{j\frac{\pi}{4}n}H(e^{j\frac{\pi}{4}}) + \frac{3}{2}e^{-j\frac{\pi}{6}}e^{-j\frac{\pi}{4}n}H(e^{-j\frac{\pi}{4}}) + 5e^{-j\pi n}H(e^{-j\pi})$$

(d) See plot below.

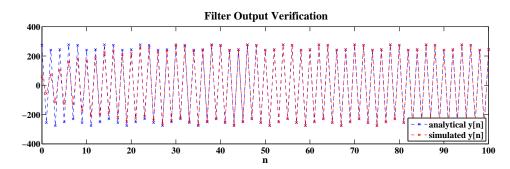


FIGURE 5.12: MATLAB verification of the steady-state response to x[n].

$$c_k^{(x)} = \frac{1}{10} \cdot \frac{1 - 0.8^{10}}{1 - 0.8e^{-j\frac{2\pi}{10}k}}$$

(b) Solution:

$$c_k^{(y)} = c_k^{(x)} H(e^{j\frac{2\pi}{10}k})$$

(c) Solution:

$$y_{\rm ss}[n] = \sum_{n=0}^{9} c_k^{(x)} e^{-j\frac{2\pi}{10}kn} H(e^{j\frac{2\pi}{10}k})$$

(d) See plot below.

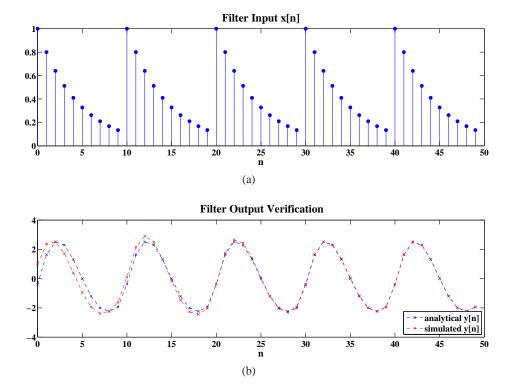


FIGURE 5.13: Signal plots of (a) x[n]. (b) y[n].

$$H(e^{j\omega}) = \frac{1}{2} \frac{1 - 0.45e^{-j\omega} + 0.05e^{-2j\omega}}{1 - 0.325e^{-j\omega} + 0.0225e^{-2j\omega}}$$

The frequency response exists and is unique.

(b) Solution:

The frequency response does NOT exist since the frequency changes.

(c) Solution:

The frequency response exists but is NOT unique.

(d) Solution:

$$y[n] = x[n] - x[n-1] + \delta[n]$$

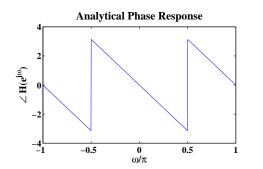
which is NOT LTI system.

28. (a) Solution:

The frequency response is:

$$H(e^{j\omega}) = \frac{1}{2}(1 - 2\cos\omega + 4\cos 2\omega - 2\cos 3\omega + \cos 4\omega) + \frac{1}{2}j(2\sin\omega - 4\sin 2\omega + 2\sin 3\omega - \sin 4\omega)$$

(b) See plot below.



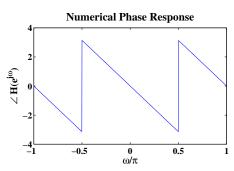


FIGURE 5.14: Phase response plot.

29. (a) Solution:

The frequency response is:

$$H(e^{j\omega}) = (1 - 2\cos\omega + 3\cos2\omega - 4\cos3\omega + 4\cos5\omega - 3\cos6\omega + 2\cos7\omega - \cos8\omega) + j(2\sin\omega - 3\sin2\omega + 4\sin3\omega - 4\sin5\omega + 3\sin6\omega - 2\sin7\omega + \sin8\omega)$$

The analytical phase response is:

$$\angle H(e^{j\omega}) = \tan^{-1} \frac{2\sin\omega - 3\sin2\omega + 4\sin3\omega - 4\sin5\omega + 3\sin6\omega - 2\sin7\omega + \sin8\omega}{1 - 2\cos\omega + 3\cos2\omega - 4\cos3\omega + 4\cos5\omega - 3\cos6\omega + 2\cos7\omega - \cos8\omega}$$

- (b) See plot below.
- 30. (a) See plot below.
 - (b) See plot below.
 - (c) See plot below.
 - (d) See plot below.
 - (e) See plot below.

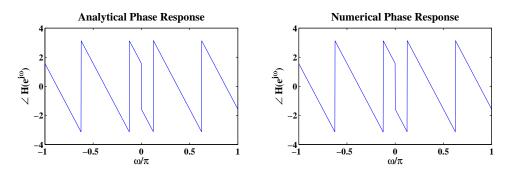


FIGURE 5.15: Phase response plot.

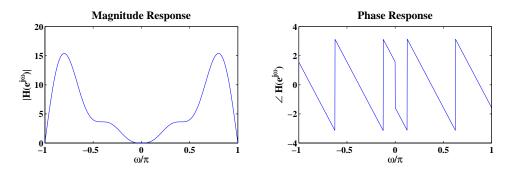


FIGURE 5.16: Magnitude and phase responses of the system.

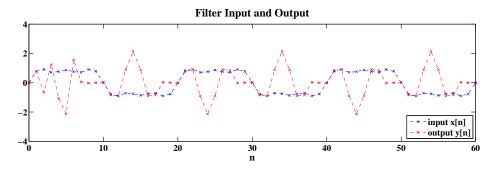


FIGURE 5.17: Plot of the input and steady state response.

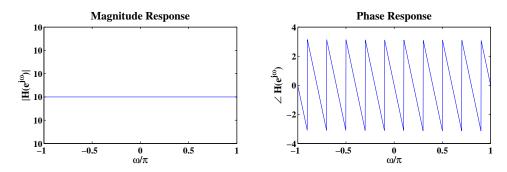


FIGURE 5.18: Magnitude and phase responses of the system.

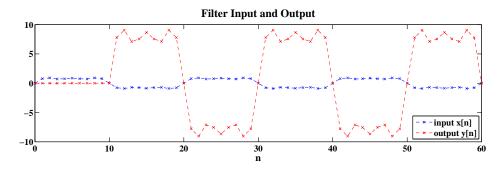


FIGURE 5.19: Plot of the input and steady state response.

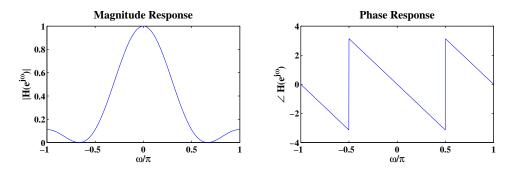


FIGURE 5.20: Magnitude and phase responses of the system.

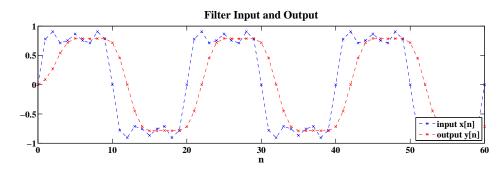


FIGURE 5.21: Plot of the input and steady state response.

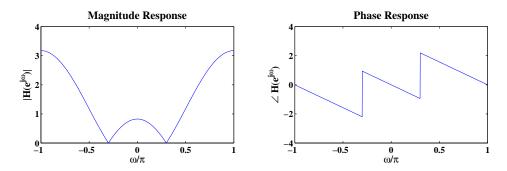


FIGURE 5.22: Magnitude and phase responses of the system.

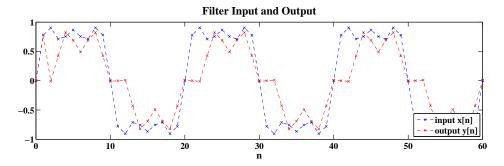


FIGURE 5.23: Plot of the input and steady state response.

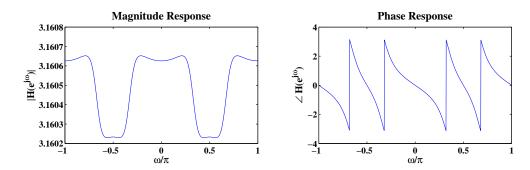


FIGURE 5.24: Magnitude and phase responses of the system.

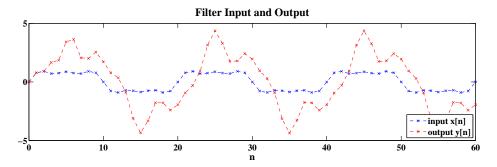


FIGURE 5.25: Plot of the input and steady state response.

The impulse response of the filter is:

$$h[n] = \frac{1}{2\pi} \frac{2}{n - n_d} \left[\sin \frac{2\pi}{8} (n - n_d) - \sin \frac{\pi}{8} (n - n_d) \right] + \frac{1}{2\pi} \frac{1}{n - n_d} \left[\sin \frac{7\pi}{8} (n - n_d) - \sin \frac{5\pi}{8} (n - n_d) \right]$$

(b) See plot below.

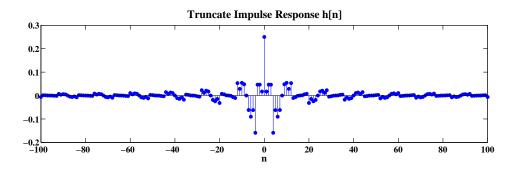


FIGURE 5.26: Impulse response for $n_d=0$ for $-100 \le n \le 100$.

(c) See plot below.

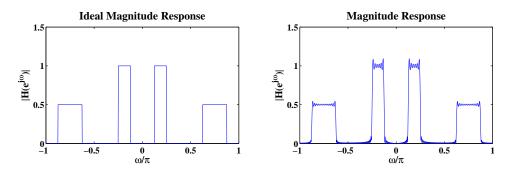


FIGURE 5.27: Magnitude response of the filter using MATLAB and the ideal filter magnitude response.

32. See plot below.

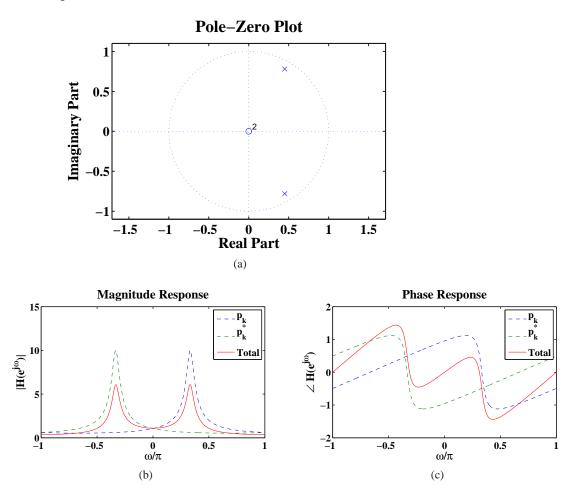


FIGURE 5.28: Figure 5.19 reproduction.

- 33. (a) See plot below.
 - (b) See plot below.
 - (c) See plot below.

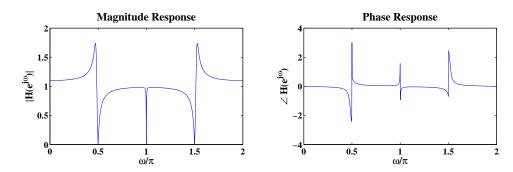


FIGURE 5.29: Magnitude and phase responses of the system using MATLAB function freqz.

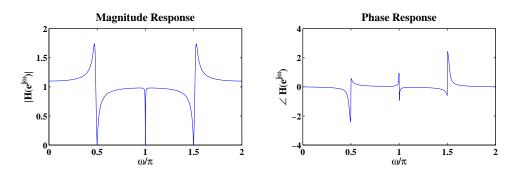


FIGURE 5.30: Magnitude and phase responses of the system using MATLAB function freqz0.

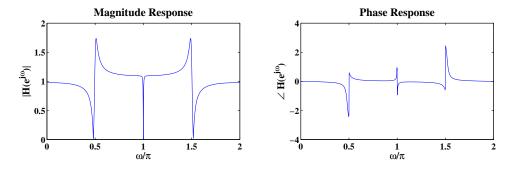


FIGURE 5.31: Magnitude and phase responses of the system using MATLAB function myfreqz.

34. (a) See plot below.

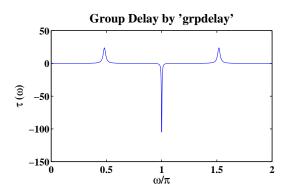


FIGURE 5.32: Group delay of the system using MATLAB function grpdelay.

(b) See plot below.

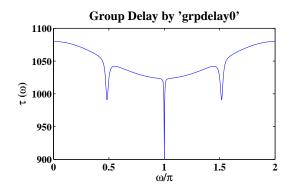


FIGURE 5.33: Group delay of the system using MATLAB function grpdelay0.

(c) See plot below.

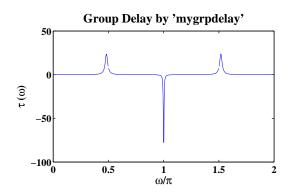


FIGURE 5.34: Group delay of the system using MATLAB function mygrpdelay.

$$\tau_{\rm gd}(\omega) = \frac{\alpha^2 - \alpha \cos \omega}{1 + \alpha^2 - 2\alpha \cos \omega}$$

(b) Solution:

$$\tau_{\rm gd}(\omega) = \frac{-\alpha^2 + \alpha \cos \omega}{1 + \alpha^2 - 2\alpha \cos \omega}$$

(c) Solution:

$$=\frac{\tau_{\rm gd}(\omega)}{4\alpha^3\cos\phi\cos\omega-2\alpha^4-4\alpha^2\cos^2\phi+2\alpha^3\sin\omega\sin2\omega+2\alpha\cos\phi\cos\omega+2\alpha^2\cos\phi\cos\omega\cos2\omega-2\alpha^2\cos2\omega}{4\alpha^2\cos^2\phi+1+\alpha^4-4\alpha^3\cos\phi\cos\omega-4\alpha\cos\phi\cos\omega+2\alpha^2\cos2\omega}$$

The system function is:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{2}(1+z^{-1})$$

The magnitude response is:

$$|H(e^{j\omega})| = |\cos\frac{\omega}{2}|$$

The phase response is:

$$\angle H(e^{j\omega}) = -\frac{\omega}{2}$$

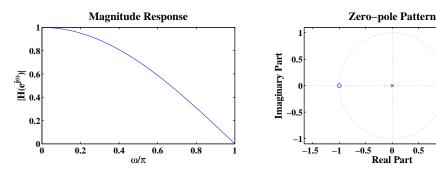


FIGURE 5.35: Pole-zero pattern and magnitude response of the system.

(b) Solution:

The system function is:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{2}(1 + z^{-2})$$

The magnitude response is:

$$|H(e^{j\omega})| = |\cos\omega|$$

The phase response is:

$$\angle H(e^{j\omega}) = -\omega$$

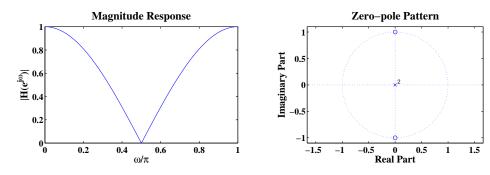


FIGURE 5.36: Pole-zero pattern and magnitude response of the system.

(c) Solution:

The system function is:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{4}(1 - z^{-1} + z^{-2} - z^{-3})$$

The magnitude response is:

$$|H(e^{j\omega})| = \frac{1}{4}\sqrt{4 - 6\cos\omega + 4\cos2\omega - 2\cos3\omega}$$

The phase response is:

$$\angle H(e^{j\omega}) = -\tan^{-1} \frac{\sin 2\omega (2\cos \omega - 1)}{1 + \cos 2\omega (1 - 2\cos \omega)}$$

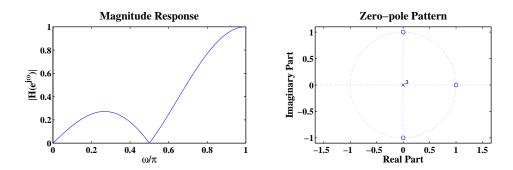


FIGURE 5.37: Pole-zero pattern and magnitude response of the system.

(d) Solution:

The system function is:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{4}(1 - z^{-1} + z^{-3} - z^{-4})$$

The magnitude response is:

$$|H(e^{j\omega})| = \frac{1}{4}\sqrt{4 - 4\cos\omega - 2\cos 2\omega + 4\cos 3\omega - 2\cos 4\omega}$$

The phase response is:

$$\angle H(e^{j\omega}) = \tan^{-1} \frac{\sin \omega - \sin 3\omega + \sin 4\omega}{1 - \cos \omega + \cos 3\omega - \cos 4\omega}$$

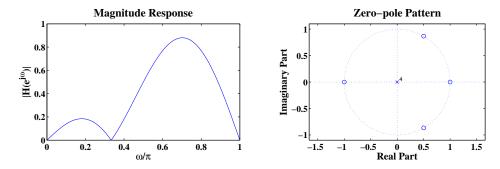


FIGURE 5.38: Pole-zero pattern and magnitude response of the system.

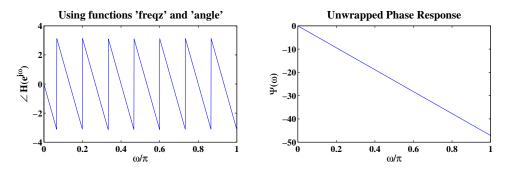


FIGURE 5.39: Phase response of pure delay y[n] = x[n-15] using the functions freqz, angle and unwrap.

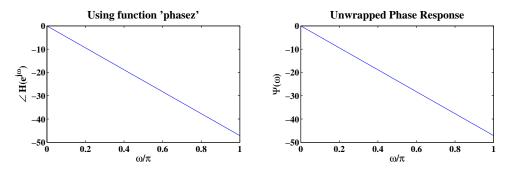


FIGURE 5.40: Phase response of pure delay y[n] = x[n-15] using the function phasez.

(b) Solution:

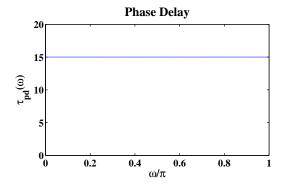


FIGURE 5.41: Phase delay of pure delay y[n] = x[n-15] using the function phasedelay.

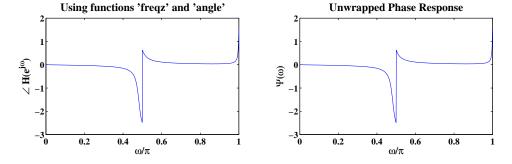


FIGURE 5.42: Phase response of H(z) using the functions freqz, angle and unwrap.

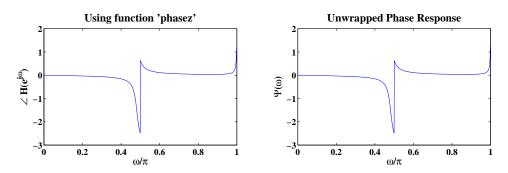


FIGURE 5.43: Phase response of H(z) using the function phasez.

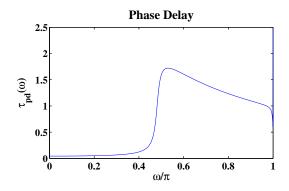


FIGURE 5.44: Phase delay of H(z) using the function phasedelay.

zeros:
$$z_1 = e^{j0} = 1$$
, $z_2 = e^{j\pi} = -1$
poles: $p_1 = re^{j\frac{\pi}{4}}$, $p_2 = re^{-j\frac{\pi}{4}}$, $r \in (0,1)$

The system function is:

$$H(z) = b_0 \frac{(1 - z^{-1})(1 + z^{-1})}{(1 - re^{j\frac{\pi}{4}}z^{-1})(1 - re^{-j\frac{\pi}{4}}z^{-1})}$$

Choose r = 0.95.

- (b) See plot below.
- (c) See plot below.

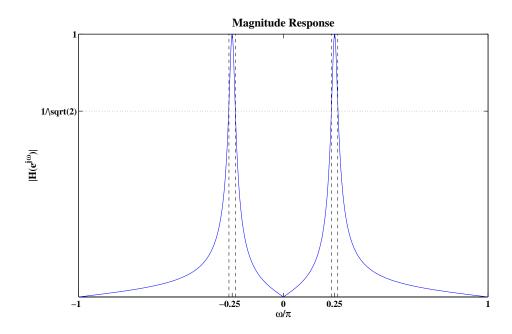


FIGURE 5.45: Magnitude response of the filter.

$$|b_0| = \frac{1}{\sqrt{1 + 4r^2 \cos^2 \phi + r^4 + 4r^2 \cos \phi + 4r \cos \phi + 2r^2}}$$

- (b) See plot below.
- (c) Solution:

$$|b_0| = \frac{1}{\prod_{k=-1}^{1} 2(1 + \cos \phi_k)}$$

(d) Solution:

$$|b_0| = \frac{1}{\prod_{k=-2}^2 2(1 + \cos \phi_k)}$$

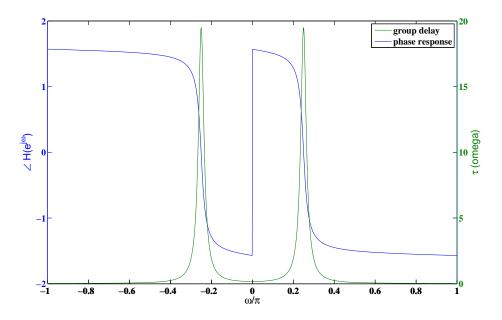


FIGURE 5.46: Phase and group-delay responses of the filter.

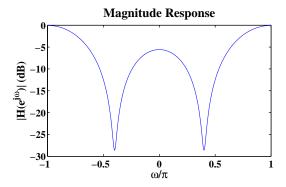


FIGURE 5.47: Magnitude response of the notch filter for r=0.95.

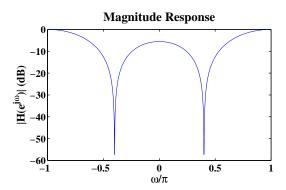


FIGURE 5.48: Magnitude response of the notch filter for r = 1.

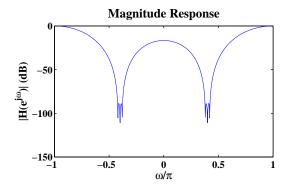


FIGURE 5.49: Magnitude response of cascade of three FIR notch filters for r=0.95.

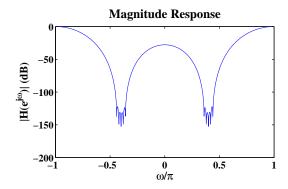


FIGURE 5.50: Magnitude response of cascade of five FIR notch filters for r=0.95.

zeros:
$$z_1 = e^{j\frac{2\pi}{3}}, \quad z_2 = e^{-j\frac{2\pi}{3}}$$
poles: $p_1 = re^{j(\frac{2\pi}{3} + \phi)}, \quad p_2 = re^{-j(\frac{2\pi}{3} + \phi)}, \quad r \in (0, 1)$

The system function is:

$$H(z) = b_0 \frac{(1 - e^{j\frac{2\pi}{3}}z^{-1})(1 - e^{-j\frac{2\pi}{3}}z^{-1})}{(1 - re^{j(\frac{2\pi}{3} + \phi)}z^{-1})(1 - re^{-j(\frac{2\pi}{3} + \phi)}z^{-1})}$$

Choose $r = 0.9, \phi = 0.01$.

(b) See plot below.

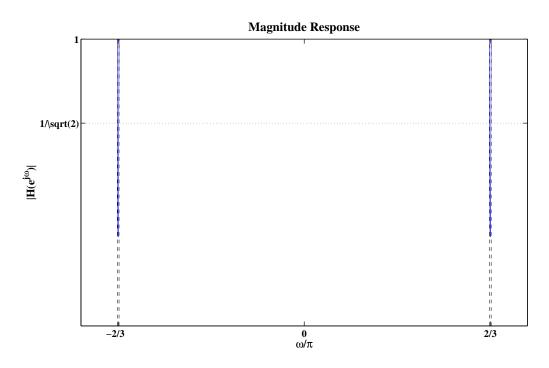


FIGURE 5.51: Magnitude response of the filter.

(c) See plot below.

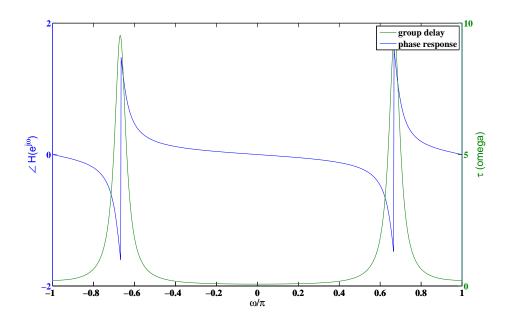


FIGURE 5.52: Phase and group-delay responses of the filter.

41. tba

$$H(z) = H_{\min}(z) \cdot H_{\rm ap}(z)$$

$$H_{\min}(z) = \frac{1 + 5.6569z^{1} + 16z^{2}}{1 - 0.8z^{-1} + 0.64z^{-2}}$$

$$H_{\rm ap}(z) = \frac{1 + 5.6569z^{-1} + 16z^{-2}}{1 + 5.6569z^{1} + 16z^{2}}$$

(b) See plot below

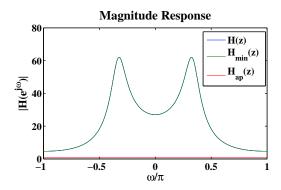


FIGURE 5.53: Magnitude responses of H(z) and its minimum-phase and all-pass components .

(c) See plot below.

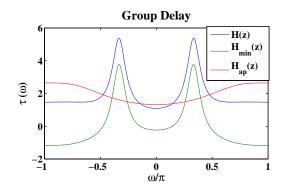


FIGURE 5.54: Group-delays of H(z) and its minimum-phase and all-pass components .

The system function is:

$$H(s) = \frac{10s}{s^2 + 2s + 101}$$

(b) Solution:

$$y(t) = 5H(0) - 2e^{-j\frac{2\pi}{3}}e^{j10t}H(10) - 2e^{j\frac{2\pi}{3}}e^{-j10t}H(-10)$$
$$+ \frac{3}{2j}e^{j20t}H(20) - \frac{3}{2j}e^{-j20t}H(-20) + 2e^{-j100t}H(-100)$$

where

$$H(0) = 0, H(-10) = -0.5525, H(10) = 0.4525, H(-20) = -0.4338,$$

 $H(20) = 0.3697, H(-100) = -0.1010$

$$H_{\text{max}}(j\Omega) = \frac{\Omega^2 - 3.4641\Omega + 4}{\Omega^2 - 4\Omega + 5}$$
$$H_{\text{min}}(j\Omega) = \frac{\Omega^2 + 3.4641\Omega + 4}{\Omega^2 + 4\Omega + 5}$$

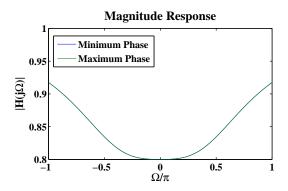


FIGURE 5.55: Magnitude responses of he minimum-phase and maximum-phase system components.

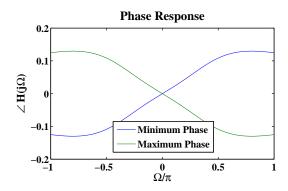


FIGURE 5.56: Phase responses of he minimum-phase and maximum-phase system components.

(b)