# TCET Formula Sheet

# Decibel Formulas Relative Power Gain

$$A_P = \frac{P_O}{P_I}$$

where  $P_O$  and  $P_I$  are defined as the following:

$$P_I = \frac{V_I^2}{R_I}$$

$$P_O = \frac{V_O^2}{R_O}$$

### Relative Voltage Gain

$$A_V = \frac{V_O}{V_I}$$

#### Relative Power Gain in dB

$$A_P(db) = 10 \log_{10} A_P$$
  
Given that  $R_O = R_I$ 

If  $R_O \neq R_I$  then the general form is given by the following:

$$A_P(db) = 10 \log_{10} \left( \frac{\frac{V_O^2}{R_O}}{\frac{V_I^2}{R_I}} \right)$$

## Relative Voltage Gain in dB

$$A_V(db) = 20 \log_{10} \left(\frac{V_O}{V_I}\right) = 20 \log_{10} A_V$$

If  $R_O \neq R_I$  then the general form is given by the following:

$$A_V(db) = 20\log_{10}\left(\frac{V_O}{V_I}\right) - 10\log_{10}\left(\frac{R_O}{R_I}\right)$$

# Absolute Power Gain dBm

$$A_{P(dBm)} = 10 \log_{10} \left( \frac{P}{1 \text{ mW}} \right), \text{ dBm}$$

### Absolute Power Gain dBw

$$A_{P(dBw)} = 10 \log_{10} \left(\frac{P}{1 \text{ W}}\right), \text{ dBw}$$

# Signal-to-Noise Ratio

$$\begin{split} \text{SNR} &= 10 \log_{10} \left( \frac{\text{Signal Power}}{\text{Noise Power}} \right) \\ \text{And given that } R_O &= R_I, \\ \text{SNR} &= 10 \log_{10} \left( \frac{V_S^2}{V_N^2} \right) \\ \text{SNR} &= 20 \log_{10} \left( \frac{V_S}{V_N} \right) \text{dB} \end{split}$$

### Impulse Noise

$$dB_S = 20 \log_{10} \left( \frac{P}{0.0002\bar{\mu}} \right)$$
, where P is sound pressure in  $\bar{\mu}$ 

 $\bar{\mu} = 1 \frac{\text{dyne}}{\text{cm}^2} = 10^{-6}$  of atmospheric pressure at sea level

# Impedance Matching

$$R_S = R_L$$
, for DC sources.  
 $Z_S = Z_L$ , for AC sources.

Source Resistance = Load Resistance

$$P_I = P_O$$

### Key Points L-Pad Networks

- 1. The primary applications of L-networks involve impedance matching in RF circuits, transmitters, and receivers.
- 2. L-networks are useful in matching one amplifier output to the input of a following stage.
- 3. Any RF circuit application covering a narrow frequency range is a candidate for an L-network.
- There are four basic versions of the L-network, with two low-pass versions and two high-pass versions.
- 5. Most widely used since they attenuate harmonics, noise, and other undesired signals.
- 6. The impedances that are being matched determine the Q (quality factor) of the circuit, which cannot be specified or controlled.\*\*
- 7. There are limits to the range of impedances that it can match.\*\*

### Key Points Pi-Pad Networks

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- 4. There are four basic versions of the L-network, with two low-pass versions and two high-pass versions.
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## **Key Points T-Pad Networks**

- 1. The primary applications of L-networks involve impedance matching in RF circuits, transmitters, and receivers.
- 2. L-networks are useful in matching one amplifier

output to the input of a following stage.

- Any RF circuit application covering a narrow frequency range is a candidate for an L-network.
- 4. There are four basic versions of the L-network, with two low-pass versions and two high-pass versions.
- 5. Most widely used since they attenuate harmonics, noise, and other undesired signals.
- 6. The impedances that are being matched determine the Q (quality factor) of the circuit, which cannot be specified or controlled.\*\*
- 7. There are limits to the range of impedances that it can match.\*\*

### L-Pad Formulas (Low Pass 1)

$$Q = \sqrt{\frac{R_L}{R_S}} - 1$$
 
$$X_L = QR_S$$
 
$$L = \frac{X_L}{2\pi f}$$
 
$$X_C = \frac{R_L}{Q}$$
 
$$C = \frac{1}{2\pi f X_C}$$
 
$$BW = \frac{f}{Q}$$

### L-Pad Formulas (Low Pass 2)

$$Q = \sqrt{\frac{R_L}{R_S} - 1}$$
 
$$X_L = QR_S$$
 
$$L = \frac{X_L}{2\pi f}$$
 
$$X_C = \frac{R_L}{Q}$$
 
$$C = \frac{1}{2\pi f X_C}$$
 
$$BW = \frac{f}{Q}$$

## Resonance

Etc.

# Telecommunications (Chapter 11)

Etc.