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## 1 Week 4 Module 3: Part 2 - Advanced Fourier Analysis

### 1.1 Discrete Fourier Series DFS



#### Discrete Fourier Series

⌘ DFS = DFT with periodicity explicit  $\tilde{X}[k] = DFS\{x[n]\}$

- The DFS maps an N-Periodic signal onto an N-Periodic sequence of Fourier coefficients
- The inverse DFS maps n<sub>periodic</sub> sequence of Fourier coefficients a set onto an N-periodic signal
- DFS is an extension of the DFT for periodic sequences
- A circular time-shift is a natural extension of a shift fo finite length signals.

#### 1.1.1 Finite-length time shifts revisited

- The DFS helps us understand how to define time shifts for finite-lenght signals.

test

**For an N-periodic sequence  $\tilde{x}[n]$**

$\tilde{x}[n - M]$  is well-defined for all  $M \in \mathbb{N}$

$DFS\{\tilde{x}[n - M]\} = e^{-j\frac{2\pi}{N}Mk} \tilde{X}[k]$  delay factor

$IDFS\left\{e^{-j\frac{2\pi}{N}Mk} \tilde{X}[k]\right\} = \tilde{x}[n - M]$  delay factor

**For an N-length signal  $x[n]$**

$\tilde{x}[n - M]$  not well-defined for all  $M \in \mathbb{N}$

build  $\tilde{x}[n] = x[n \bmod N] \Rightarrow \tilde{X}[k] = X[k]$

$IDFT\left\{e^{-j\frac{2\pi}{N}Mk} X[k]\right\} = IDFS\left\{e^{-j\frac{2\pi}{N}Mk} \tilde{X}[k]\right\} = \tilde{x}[n - M] = x[(n - M) \bmod N]$



### Periodicity

↗ Shifts for finite-length signals are "naturally" circular

**Analysis Formula for a N-Periodic Signal in the frequency domain**

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j\frac{2\pi}{N}nk}, k \in \mathbb{Z} \quad (1)$$

$X[k]$  Signal vector in the frequency domain

$x[n]$  Signal vector in the (discrete) time domain

**Reminder** This is the inner Product in explicite form

**Synthesis Formula for a N-Periodic Signal in the time domain**

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}nk}, k \in \mathbb{Z} \quad (2)$$

$x[n]$  Signal vector in the (discrete) time domain

$X[k]$  Signal vector in the frequency domain

$\frac{1}{N}$  Normalisation coefficient

**Reminder** This is the inner Product in explicite fashion

## 1.2 The Discret-Time Fourier Transform (DTFT)

### 1.2.1 Overview Fourier Transform

- N-Point finite-length signals: DFT
- N-Point periodic signals: DFS
- Infinite length (non periodic) signals: DTFT

### 1.2.2 Karplus Strong revisited and the DTFT

#### Plotting the DTFT

- Frequencies go from  $-\pi$  to  $\pi$
- Positive frequencies are on the right hand side of the x-axis
- Negative frequencies are on the left hand side of the x-axis
- Low frequencies are centered around 0
- High frequencies will be on the extreme of the bound

### 1.2.3 Formal Definition of the DTFT

- $x[n] \in \ell_2(\mathbb{Z})$ , the space of square summable infinity sequences
- define the function of  $\omega \in \mathbb{R}$

$$F(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}, \text{ with } \omega = \frac{2\pi}{N} \text{ and } N \rightarrow \infty$$

- inversion (when  $F(\omega)$  exists):

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} F(e^{j\omega}) e^{j\omega n} d\omega, \text{ with } n \in \mathbb{Z}$$

### 1.2.4 Properties of the DTFT

<b>linearity</b>	$DTFT\{\alpha x[n] + \beta y[n]\} = \alpha X(e^{j\omega}) + \beta Y(e^{j\omega})$
<b>timeshift</b>	$DTFT\{x[n - M]\} = e^{-j\omega M} X(e^{j\omega})$
<b>modulation</b>	$DTFT\{e^{-j\omega_0 M} x[n]\} = X(e^{j(\omega - \omega_0)})$
<b>time reversal</b>	$DTFT\{x[-n]\} = X(e^{-j\omega})$
<b>conjugation</b>	$DTFT\{x^*[n]\} = X^* X(e^{-j\omega})$

### 1.2.5 Some particular cases

- if  $x[n]$  is symmetric, the DTFT is symmetric:  $x[n] = x[-n] \iff X(e^{j\omega}) = X(e^{-j\omega})$
- if  $x[n]$  is real, the DTFT is Hermitian-symmetric:  $x[n] = x^*[n] \iff X(e^{j\omega}) = X^*(e^{-j\omega})$
- if  $x[n]$  is real, the magnitude of the DTFT is symmetric  $x[n] \in \mathbb{R} \implies |X(e^{j\omega})| = |X(e^{-j\omega})|$
- if  $x[n]$  is real and symmetric,  $X(e^{j\omega})$  is also real and symmetric

## 1.3 TODO Sinusoidal Modulation