

Periodic Complex Exponential and Sinusoidal Signals

$$x(t) = e^{j\omega_0 t} \quad (1.21)$$

$$e^{j\omega_0 t} = e^{j\omega_0(t+T)} = e^{j\omega_0 t} \cdot \frac{e^{j\omega_0 T}}{1} \quad (1.22)$$

which implies that $\omega_0 T_0$ is a multiple of 2π , i.e.,

$$\omega_0 T_0 = 2\pi k, \quad k = 0, \pm 1, \pm 2, \dots \quad (1.23)$$

Thus,

$$\omega_0 = \frac{2\pi}{T_0} \quad (1.25)$$

$$T_0 = \frac{2\pi}{|\omega_0|} \quad (1.26)$$

Periodicity Properties of Disc-Time Comp. Exp.

$$e^{j(\omega_0 + 2\pi)n} = e^{j2\pi n} e^{j\omega_0 n} = e^{j\omega_0 n}$$

always 1

* Therefore, in considering disc-time comp. exp., we need only consider a frequency interval of length 2π in which to choose ω_0

* as ω_0 increases from 0, signals that oscillate more and more rapidly until $\omega_0 = \pi$. After $\omega_0 = \pi$, decrease the rate of oscillation until $\omega_0 = -\pi$

* In order for the signal $e^{j\omega_0 n}$ to be periodic with period $N > 0$,

$$e^{j\omega_0(n+N)} = e^{j\omega_0 n} \quad (1.53)$$

①

$$e^{jw_0 N} = 1 \quad (1.54)$$

$$w_0 N = 2\pi m \quad (1.55)$$

$$\frac{w_0}{2\pi} = \frac{m}{N} \quad (1.56)$$

↓

If rational number the signal is periodic

The fundamental freq. of periodic signal $e^{jw_0 n}$ is

$$\frac{2\pi}{N} = \frac{w_0}{m} \quad (1.57)$$

The fundamental period of periodic signal $e^{jw_0 n}$ is

$$N = m \left(\frac{2\pi}{w_0} \right) \quad (1.58)$$

$e^{jw_0 n}$	$e^{jw_0 n}$
* Distinct signals for distinct values of w_0	* Identical signals for values of w_0 separated by multiples of 2π
* Periodic for any choice of w_0	* Periodic only if $w_0 = 2\pi m/N$ for some integers $N > 0$ and m
* Fundamental freq. w_0	* Fundamental freq * w_0/m
* Fundamental period	* Fundamental period
$w_0 = 0$: undefined	$w_0 = 0$: undefined
$w_0 \neq 0$: $\frac{2\pi}{w_0}$	$w_0 \neq 0$: $m \left(\frac{2\pi}{w_0} \right)$

* Assumes that m and N do not have any factors in common

(2)

Periodicity Ex 1 (1.25): Determine whether or not each of the following as-time signals is periodic. If the signal is periodic, determine its fundamental period.

a) $x(t) = 3 \cos(4t + \frac{\pi}{3})$

If $x(t)$ is periodic

$$x(t) = x(t+T)$$

$$x(t) \stackrel{?}{=} x(t+T) \Rightarrow x(t+T) = 3 \cos\left(4(t+T) + \frac{\pi}{3}\right)$$

$$\begin{aligned} 3 \cos\left(4t + \frac{\pi}{3}\right) &= 3 \cos\left(4(t+T) + \frac{\pi}{3}\right) \\ &= 3 \cos\left(4t + 4T + \frac{\pi}{3}\right) \end{aligned}$$

$$4T = 2\pi m \quad m \in \mathbb{Z}$$

$$T = \frac{2\pi m}{4}$$

Fundamental period is $T_0 = \underline{\frac{2\pi}{4}}$

b) $x(t) = e^{j(\pi t - 1)}$

$$x(t) \stackrel{?}{=} x(t+T)$$

$$\begin{aligned} e^{j(\pi t - 1)} &= e^{j(\pi(t+T) - 1)} = e^{j(\pi t + \pi T - 1)} \\ &= e^{j(\pi t + 1 + \pi T)} = e^{j(\pi t - 1) + j\pi T} = e^{j(\pi t - 1)} \cdot \frac{e^{j\pi T}}{1} \end{aligned}$$

$$e^{j\pi T} = \underbrace{\cos(\pi T)}_{-1} + j\underbrace{\sin(\pi T)}_0, \quad T \text{ must be 2 at least}, \quad \text{so } T_0 = 2$$

③

$$c) \quad x(t) = [\cos(2t - \frac{\pi}{3})]^2$$

From half-angle formula

$$\cos^2(2t - \frac{\pi}{3}) = \frac{\cos(4t - \frac{2\pi}{3}) + 1}{2} \quad \text{does not affect on periodicity}$$

$$\cos(4t - \frac{2\pi}{3}) = \cos(4t + 4T - \frac{2\pi}{3})$$

$$4T = 2\pi m$$

$$T_0 = \frac{2\pi}{4}$$

$$d) \quad x(t) = E_v \left\{ \cos(4\pi t) u(t) \right\}$$

For $f(t)$,

$$\text{Odd}\{f(t)\} = \frac{1}{2} \left\{ f(t) - f(-t) \right\}$$

$$E_v \{ f(t) \} = \frac{1}{2} \left\{ f(t) + f(-t) \right\}$$

$$E_v \left\{ \cos(4\pi t) u(t) \right\} = \frac{1}{2} \left\{ \cos(4\pi t) u(t) + \cos(-4\pi t) u(-t) \right\}$$

$$= \frac{1}{2} \left\{ \underbrace{\cos(4\pi t) u(t)}_{t < 0 \Rightarrow 0} + \underbrace{\cos(4\pi t) u(-t)}_{t > 0 \Rightarrow 0} \right\}$$

$$= \frac{1}{2} \left\{ \cos(4\pi t) u(t) \right\}$$

$$\cos(4\pi t) = \cos(4\pi t + 4\pi T) \Rightarrow 4\pi T = 2\pi m$$

$$T_0 = \frac{1}{2}$$

(4)

$$\begin{aligned}
 e) \quad x(t) &= \mathcal{E}v \left\{ \sin(4\pi t) u(t) \right\} \\
 &= \frac{1}{2} \left\{ \sin(4\pi t) u(t) + \sin(-4\pi t) u(-t) \right\} \\
 &\stackrel{u(-t)=0}{=} \frac{1}{2} \left\{ \sin(4\pi t) u(t) - \sin(4\pi t) u(-t) \right\} \\
 \xrightarrow{t>0} \quad x(t) &= \frac{1}{2} \left\{ \sin(4\pi t) \right\} \quad t=0 \quad u(0)=\frac{1}{2} \\
 \sin(4\pi t) &= \sin(4\pi(t+T))
 \end{aligned}$$

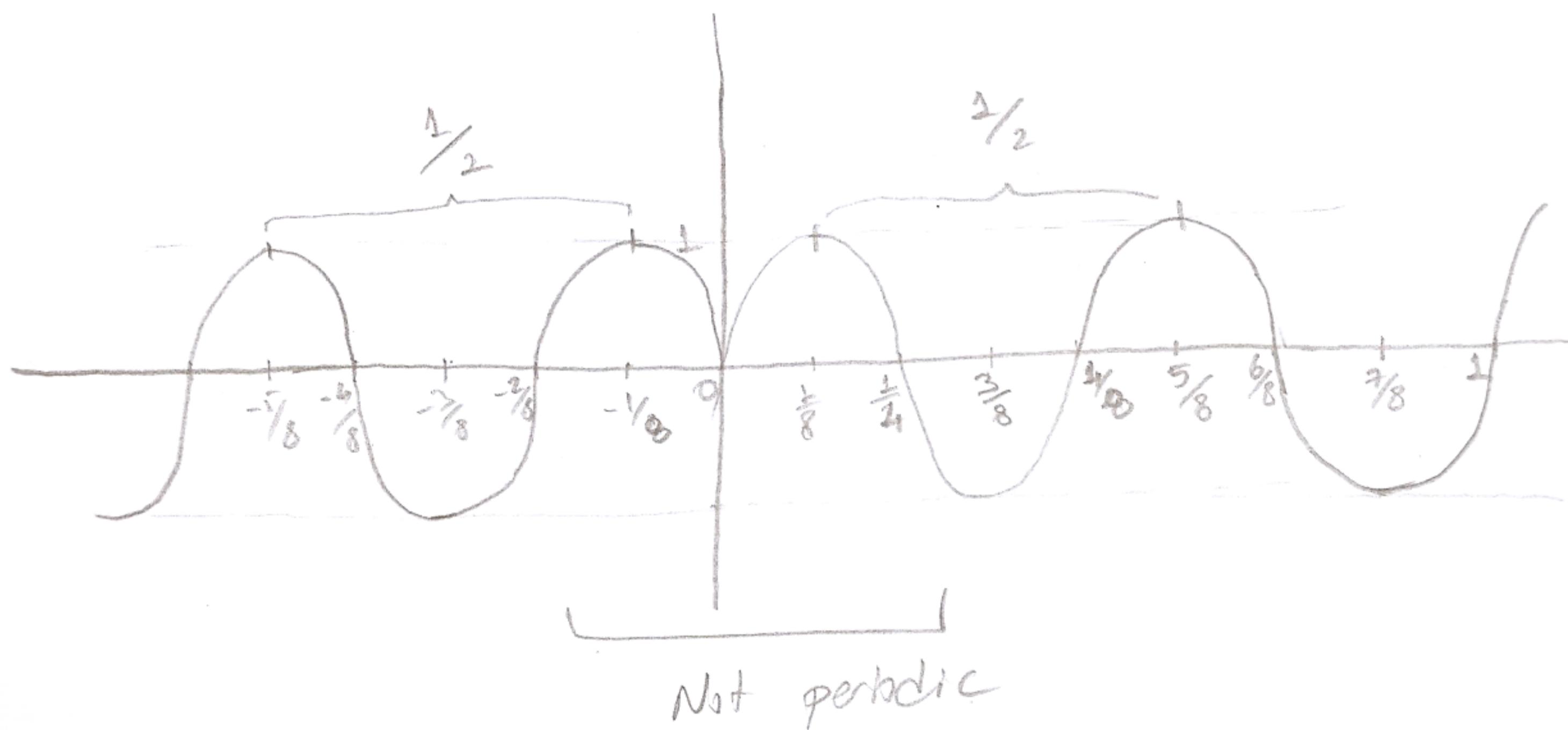
$$4\pi T = 2\pi m$$

$$T_0 = \frac{1}{2}$$

$$\begin{aligned}
 t < 0 \quad x(t) &= \frac{1}{2} \left\{ -\sin(4\pi t) \right\} \\
 \sin(4\pi t) &= \sin(4\pi(t+T))
 \end{aligned}$$

$$4\pi T = 2\pi m$$

$$T_0 = \frac{1}{2}$$



⑤

$$f) x(t) = \sum_{n=-\infty}^{\infty} e^{-(2t-n)} u(2t-n)$$

For any real t , $\lfloor t \rfloor$ represents floor of t (largest integer which is less than or equal to t).

Since $2t-n \geq 0$ for all integers $n \leq 2t$, and since $2t-n < 0$ for all integers $n \geq \lfloor 2t \rfloor + 1$,

$$u(2t-n) = \begin{cases} 1 & \text{for all } n \leq \lfloor 2t \rfloor \\ 0 & \text{for all } n \geq \lfloor 2t \rfloor + 1 \end{cases}$$

$x(t)$ given in the problem statement can be rewritten as follows:

$$\begin{aligned} x(t) &= \sum_{n=-\infty}^{\lfloor 2t \rfloor} e^{-(2t-n)} = e^{-2t} \sum_{m=-\lfloor 2t \rfloor}^{\infty} e^{-m} \\ &= e^{-2t} \frac{e^{\lfloor 2t \rfloor}}{1 - e^{-1}} = \frac{e^{-2(t-\lfloor t \rfloor)}}{1 - e^{-1}} \end{aligned}$$

where $m = -n$. Since $t - \lfloor t \rfloor = (t+1) - \lfloor t+1 \rfloor$ for all t ,

we have: $x(t+1) = x(t)$. Therefore, 1 is a period of x .

To see that it's fundamental period of x , if $0 < T < 1$

the equation $x(t) = x(t+T)$ will not hold for any integer t .

Because, if t is integer and T is a fraction between 0 & 1,
then $t - \lfloor t \rfloor = 0$ whereas $t + T - \lfloor t + T \rfloor = t + T - t = T \neq 0$.

Hence, 1 is the smallest period of x and is therefore the fundamental period.

(6)

Periodicity Ex 2: Determine whether or not each of the following discrete-time signals is periodic. If the signal is periodic, determine its fundamental period.

a) $x[n] = \sin\left(\frac{6\pi}{7}n + 1\right)$

$$x[n] \stackrel{?}{=} x[n+N]$$

$$\begin{aligned} \sin\left(\frac{6\pi}{7}n + 1\right) &= \sin\left(\frac{6\pi}{7}(n+N) + 1\right) \\ &= \sin\left(\frac{6\pi}{7}n + \frac{6\pi}{7}N + 1\right) \end{aligned}$$

$$\frac{6\pi N}{7} = 2\pi m, \quad m \in \mathbb{Z}$$

$$N = m\left(\frac{2\pi \cdot 7}{6\pi}\right) = m\left(\frac{7}{3}\right)$$

N must be integer in discrete-time
so $m=3$
 $N=7$

b) $x[n] = \cos\left(\frac{n}{8} - \pi\right)$

$$x[n] \stackrel{?}{=} x[n+N] \Rightarrow \cos\left(\frac{n+N}{8} - \pi\right) = \cos\left(\frac{n}{8} - \pi\right)$$

$$\frac{N}{8} = 2\pi m \Rightarrow N = 16\pi m \rightarrow \text{irrational}$$

so $x[n]$
aperiodic

(7)

$$d) \quad x[n] = \cos\left(\frac{\pi}{2}n\right) \cos\left(\frac{\pi}{4}n\right)$$

Using trigonometric product to sum formula

$$\cos(u) \cdot \cos(v) = \frac{1}{2} [\cos(u+v) + \cos(u-v)]$$

$$\cos\left(\frac{\pi}{2}n\right) \cdot \cos\left(\frac{\pi}{4}n\right) = \frac{1}{2} [\cos\left(\frac{3\pi}{4}n\right) + \cos\left(\frac{\pi}{4}n\right)]$$

Period of $\cos\left(\frac{3\pi}{4}n\right)$

$$\begin{aligned} \cos\left(\frac{3\pi}{4}n\right) &= \cos\left(\frac{3\pi}{4}(n+N)\right) \\ &= \cos\left(\frac{3\pi}{4}n + \frac{3\pi N}{4}\right) \end{aligned}$$

$$\frac{3\pi N}{4} = 2\pi m \Rightarrow N = \frac{8}{3}m \Rightarrow N \text{ must be integer}$$

Least possible value
of N is 8

Period of $\cos\left(\frac{\pi}{4}n\right)$

$$\begin{aligned} \cos\left(\frac{\pi}{4}n\right) &= \cos\left(\frac{\pi}{4}(n+N)\right) \\ &= \cos\left(\frac{\pi}{4}n + \frac{\pi}{4}N\right) \end{aligned}$$

$$\frac{\pi N}{4} = 2\pi m \Rightarrow N = 8m \Rightarrow N = 8$$

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$$e) \quad x[n] = 2\cos\left(\frac{\pi}{4}n\right) + \sin\left(\frac{\pi}{8}n\right) - 2\cos\left(\frac{\pi}{2}n + \frac{\pi}{6}\right)$$

Fundamental period of $\cos\left(\frac{\pi}{4}n\right)$

$$\begin{aligned} \cos\left(\frac{\pi}{4}n\right) &= \cos\left(\frac{\pi}{4}(n+N)\right) \\ &= \cos\left(\frac{\pi n}{4} + \frac{\pi N}{4}\right) \Rightarrow \frac{\pi N}{4} = 2\pi m \\ &\quad N=8 \end{aligned}$$

Fund. period of $\sin\left(\frac{\pi}{8}n\right)$

$$\begin{aligned} \cos\left(\frac{\pi}{8}n\right) &= \cos\left(\frac{\pi}{8}(n+N)\right) \\ &= \cos\left(\frac{\pi n}{8} + \frac{\pi N}{8}\right) \Rightarrow \frac{\pi N}{8} = 2\pi m \\ &\quad N=16 \end{aligned}$$

Fund. period of $\cos\left(\frac{\pi}{2}n + \frac{\pi}{6}\right)$

$$\begin{aligned} \cos\left(\frac{\pi}{2}n + \frac{\pi}{6}\right) &= \cos\left(\frac{\pi}{2}n + \frac{\pi N}{2} + \frac{\pi}{6}\right) \\ \frac{\pi N}{2} &= 2\pi m \Rightarrow N=4 \end{aligned}$$

Hence, $x[n]$ is periodic and its fundamental period is the least common multiple of 8, 16, and 4 which is $N=16$

⑨

General properties of systems Ex 3: A system may or may not be

- 1) Memoryless
- 2) Time invariant
- 3) Linear
- 4) Causal
- 5) Stable

Determine which of these properties hold and which do not hold for each of the following discrete-time systems. Justify your answers.

In each example, $y(t)$ denotes the system output and $x(t)$ is the system input.

a) $y(t) = x(t-2) + x(2-t)$



— At the first glance we see that the system has memory.

— Noncausal \rightarrow depend on future

— Stable \rightarrow for bounded input gives bounded output

Is it time-invariant?

$$x_1(t) = x(t-t_0)$$

$$y_1(t) = x_1(t-2) + x_1(2-t)$$

$$= x(t-2-t_0) + \underline{x(2-t-t_0)}$$

$$y(t) = x(t-2) + x(2-t)$$

$$y(t-t_0) = x(t-t_0-2) + \underline{x(2-t+t_0)}$$

This parts differ from each other

Is it linear

$$x_1(t) \Rightarrow y_1(t) = x_1(t-2) + x_1(2-t)$$

so the system is time-varying

$$x_2(t) \Rightarrow y_2(t) = x_2(t-2) + x_2(2-t)$$

$$x(t) = a_1 x_1(t) + a_2 x_2(t) \Rightarrow y(t) = a_1 y_1(t) + a_2 y_2(t)$$

⑩ $y(t) = a_1 y_1(t) + a_2 y_2(t) \Rightarrow \text{Linear}$

b) $y(t) = [\cos(3t)] x(t)$



- Memoryless \rightarrow not depend on past
- Causal \rightarrow not depend on future
- Stable \rightarrow for bounded input gives bounded output

Is it time-invariant?

$$x(t) = \dots$$

$$x_1(t) = x(t-t_0)$$

$$y(t) = [\cos(3t)] x(t)$$

$$y_1(t) = [\cos(3t)] x_1(t)$$

$$y(t-t_0) = \underbrace{[\cos(3(t-t_0))]}_{\text{Differs from } x(t)} x(t-t_0)$$

$$= \underbrace{[\cos(3t)]}_{\text{Differs from } y_1(t)} x(t-t_0)$$

Differs from each other. Hence,
the system is time-varying

Is it linear?

$$x_1(t) \Rightarrow y_1(t) = [\cos(3t)] x_1(t)$$

$$x_2(t) \Rightarrow y_2(t) = [\cos(3t)] x_2(t)$$

$$x(t) = a_1 x_1(t) + a_2 x_2(t) \Rightarrow y(t) = \cos(3t) [a_1 x_1(t) + a_2 x_2(t)]$$

$$= \underbrace{\cos(3t) \cdot a_1 x_1(t)}_{a_1 y_1(t)} + \underbrace{\cos(3t) a_2 x_2(t)}_{a_2 y_2(t)}$$

$$= a_1 y_1(t) + a_2 y_2(t)$$

Linear system

$$c) \quad y(t) = \int_{-\infty}^{2t} x(z) dz$$

- Has memory \rightarrow depends on past and future
- Noncausal
- Diverge

Is it time-invariant?

$$x(t) = \dots$$

$$y(t) = \int_{-\infty}^{2t} x(z) dz$$

$$y_1(t-t_0) = \int_{-\infty}^{2t-2t_0} x(z) dz \neq y_1(t) \rightarrow \text{time varying}$$

$$x_1(t) = x(t-t_0)$$

$$y_1(t) = \int_{-\infty}^{2t} x_1(z) dz$$

Is it linear?

$$x_1(t) \Rightarrow y_1(t) = \int_{-\infty}^{2t} x_1(z) dz$$

$$x_2(t) \Rightarrow y_2(t) = \int_{-\infty}^{2t} x_2(z) dz$$

$$x(t) = a_1 x_1(t) + a_2 x_2(t) \Rightarrow y(t) = \int_{-\infty}^{2t} (a_1 x_1(z) + a_2 x_2(z)) dz$$

$$y(t) = \underbrace{\int_{-\infty}^{2t} a_1 x_1(z) dz}_{a_1 y_1(t)} + \underbrace{\int_{-\infty}^{2t} a_2 x_2(z) dz}_{a_2 y_2(t)}$$

$$y(t) = a_1 y_1(t) + a_2 y_2(t)$$

So the system is linear

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CONVOLUTION

Time-Invariance

Cont-Time :

$$x(t) \rightarrow y(t)$$

then $x(t-t_0) \rightarrow y(t-t_0)$

Disc-Time:

$$x[n] \rightarrow y[n]$$

$$x[n-n_0] \rightarrow y[n-n_0]$$

Linearity :

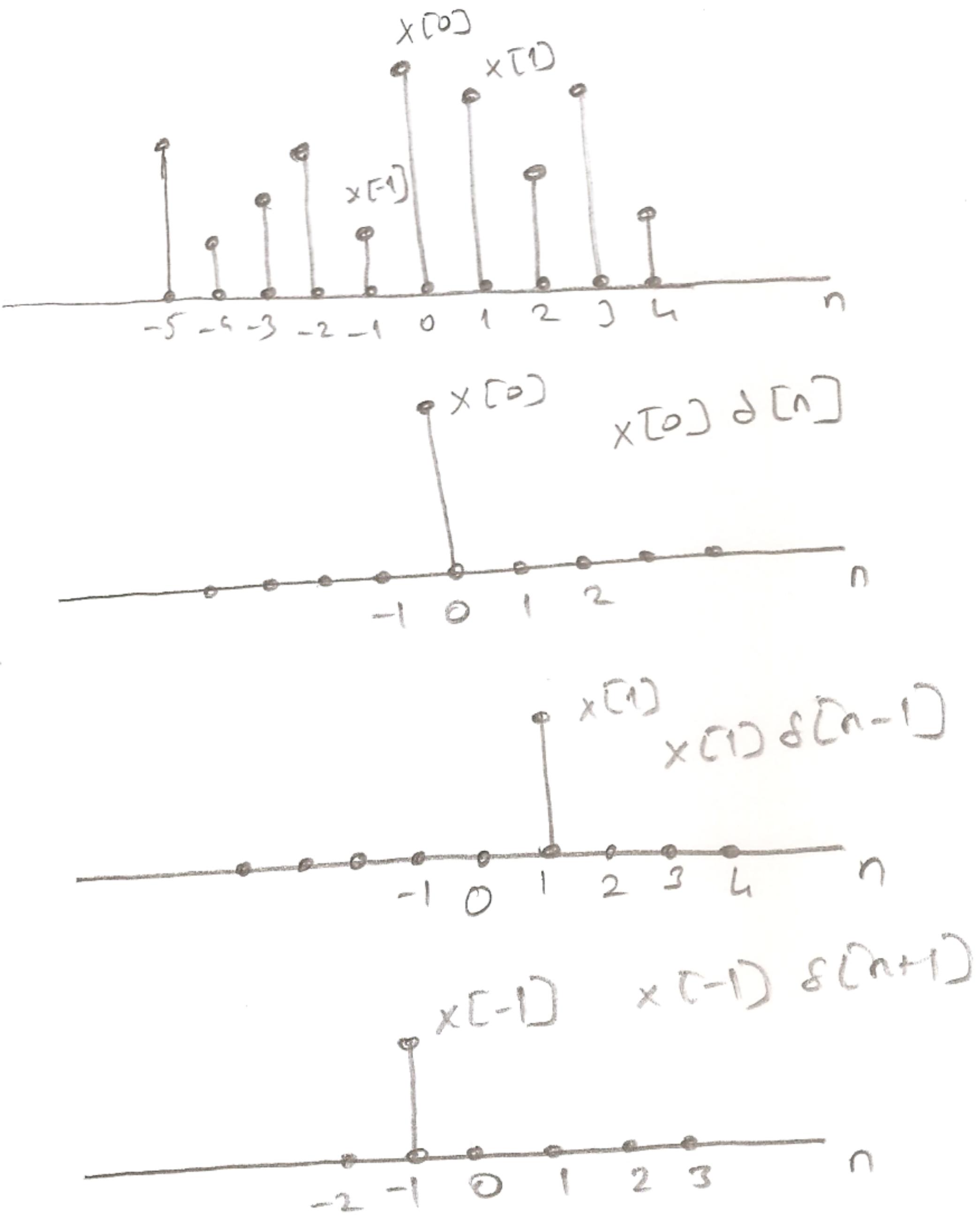
$$\phi_k \rightarrow \psi_k$$

Then

$$a_1\phi_1 + a_2\phi_2 + \dots \rightarrow a_1\psi_1 + a_2\psi_2 + \dots$$

Main Ideas of Convolution!

- Decompose input signal into a linear combination of basic signals
- Choose basic signals so that response easy to compute
- Decompose a signal a linear combination of delayed impulses. That leads to representation for LTI systems which is referred to convolution.



$$x[n] = x[0]\delta[n] + x[1]\delta[n-1]$$

$$+ x[-1]\delta[n+1] + \dots$$

$$= \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$

If we are talking about linear system, its response to that linear combination is linear combination of the responses.

If we denote the response to the delayed impulse as

$$\underbrace{\delta[n-k]}_{\text{impulse}} \rightarrow \boxed{\text{LTI system}} \rightarrow \underbrace{h_k[n]}_{\text{response of system}}$$

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Then response to general input

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n-k]$$

is

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h_k[n]$$

The system is time invariant, then the response to impulse at time k is as exactly the same as the response to an impulse at time 0, shifted over time k .

Hence,

$$h_k[n] = h_0[n-k]$$

And it is generally useful to, rather than carrying around $h_0[n]$, just simply define $h_0[n]$ as $h[n]$, which is the unit impulse response of the system.

So the consequence, then, is for a linear time-invariant system, the output can be expressed as this sum

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

Calculating and plotting convolution Ex 1:

$$x[n] = \delta[n] + 2\delta[n-1] - \delta[n-3] \quad \text{and}$$

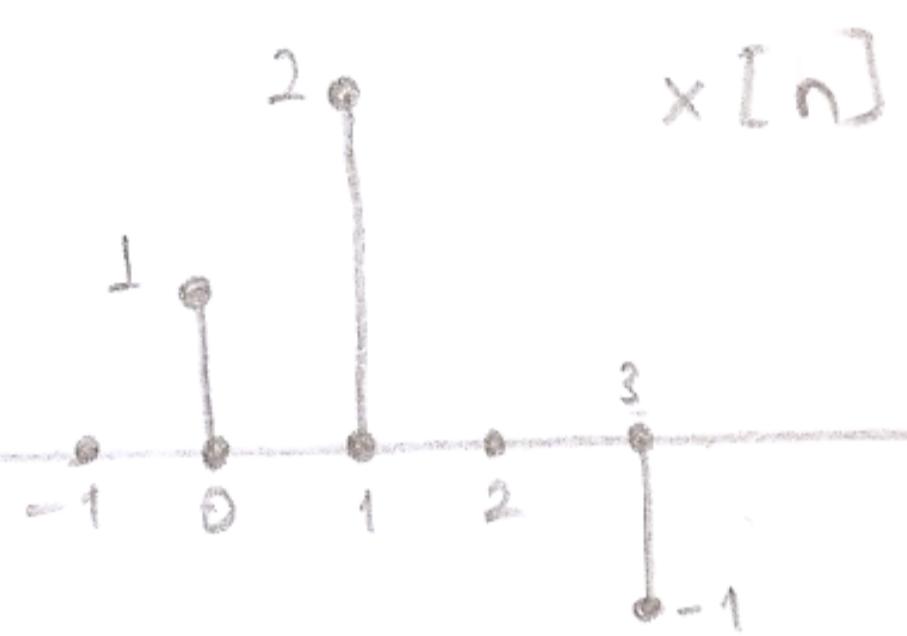
$$h[n] = 2\delta[n+1] + 2\delta[n-1]$$

Compute and plot each of the following convolutions:

a) $y_1[n] = x[n] * h[n]$

$$y_1[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

This means that shift impulse response signal from $-\infty$ to ∞ and multiply signal corresponding value of $x[n]$



While shifting $h[n]$, only

n for three value of h

$x[n]$ different from 0.

These are 0, 1, and 3

Hence,

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= x[0] h[n-0] +$$

$$x[1] h[n-1] +$$

$$x[3] h[n-3]$$

$$x[0] = 1, x[1] = 2, x[3] = -1$$

$$h[n] = 2\delta[n+1] + 2\delta[n-1]$$

$$h[n-1] = 2\delta[n] + 2\delta[n-2]$$

$$h[n-3] = 2\delta[n-2] + 2\delta[n-4]$$

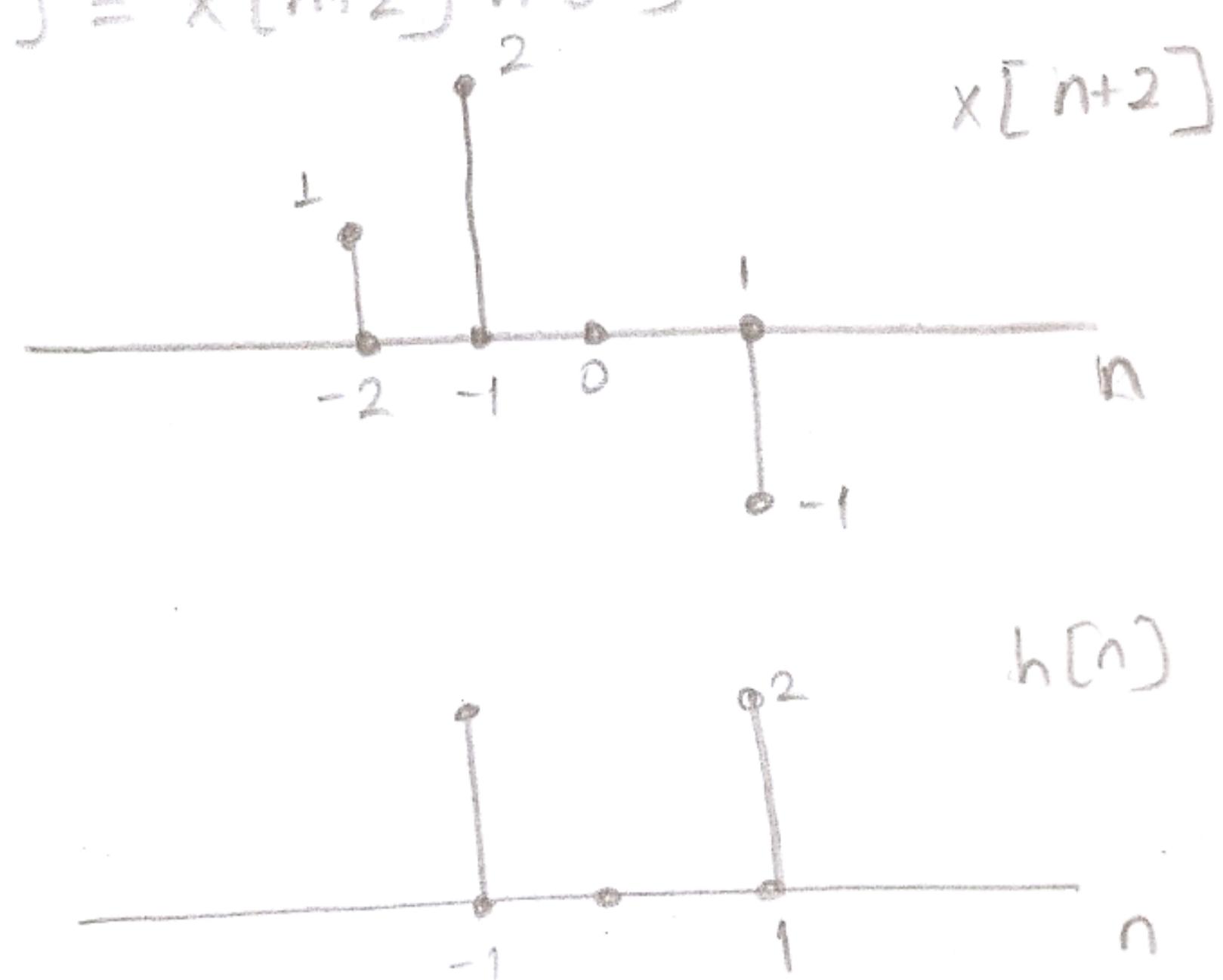
\Rightarrow
put all
this values



$$y[n] = 2\delta[n+1] + 2\delta[n-1] + 2[2\delta[n] + 2\delta[n-2]] - 2\delta[n-2] - 2\delta[n-4]$$

$$y[n] = 2\delta[n+1] + 4\delta[n] + 2\delta[n-1] + 2\delta[n-2] - 2\delta[n-4]$$

b) $y_2[n] = x[n+2] h[n]$



while shifting $h[n], x[n+2]$
has values different from
'0' for only $n=-2, -1, \text{ and } 1$.

$$\begin{aligned} y_2[n] &= \sum_{k=-\infty}^{k=\infty} x[k+2] h[n-k] = x[0] h[n+2] + x[1] h[n+1] + \\ &\quad x[3] h[n-1] \\ &= 2\delta[n+3] + 2\delta[n+1] + 2[2\delta[n+2] + 2\delta[n]] \\ &\quad - 2\delta[n] - 2\delta[n-2] \\ y_2[n] &= 2\delta[n+3] + 4\delta[n+2] + 2\delta[n+1] + 2\delta[n] - 2\delta[n-2] \end{aligned}$$

$y_2[n] = y_1[n+2]$ \rightarrow Time invariant system

c) $y_3[n] = x[n] * h[n+2]$ we have showed that the system
is time-invariant so

$$y_3[n] = y_1[n+2] = y_2[n]$$

⑦

Cts-time compute and sketch convolution Ex 2: Determine and sketch the convolution of the following two signals:

$$x(t) = \begin{cases} t+1, & 0 \leq t \leq 1 \\ 2-t, & 1 < t \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

$$h(t) = \delta(t+2) + 2\delta(t+1)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau = \int_{-\infty}^{\infty} x(\tau) [2\delta(\tau+2) + 2\delta(\tau+1)] d\tau \quad (\text{commutativity})$$

$$= \int_{-\infty}^{\infty} x(\tau) \underbrace{[2\delta(\tau+2) + 2\delta(\tau+1)]}_{\text{differ from '0' for two points } -2, -1} d\tau$$

$$= x(t+2) + 2x(t+1)$$

$$x(t+2) = \begin{cases} t+3, & -2 \leq t \leq -1 \\ -t, & -1 < t \leq 0 \\ 0, & \text{elsewhere} \end{cases}$$

$$x(t+1) = \begin{cases} t+2, & -1 \leq t \leq 0 \\ 1-t, & 0 < t \leq 1 \\ 0, & \text{elsewhere} \end{cases}$$

$$2x(t+1) = \begin{cases} 2t+4, & -1 \leq t \leq 0 \\ 2-2t, & 0 < t \leq 1 \\ 0, & \text{elsewhere} \end{cases}$$

$$y(t) = x(t+2) + 2x(t+1) = \begin{cases} t+3, & -2 \leq t < -1 \\ 3t+7, & t = -1 \\ t+1, & -1 < t \leq 0 \\ 2-2t, & 0 < t \leq 1 \\ 0, & \text{elsewhere} \end{cases}$$

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Discrete-time compute convolution Ex 3 (2.3)

Consider an input $x[n]$ and a unit impulse response $h[n]$ given by

$$x[n] = \left(\frac{1}{2}\right)^{n-2} u[n-2]$$

$$h[n] = u[n+2]$$

Determine and plot the output $y[n] = x[n] * h[n]$

Soln:

$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= \sum_{k=-\infty}^{\infty} \left(\frac{1}{2}\right)^{k-2} u[k-2] u[n-k+2]$$

$$k-2 \geq 0 \Rightarrow k \geq 2$$

$$n-k+2 \geq 0 \Rightarrow k \leq n+2$$

$$= \sum_{k=2}^{n+2} \left(\frac{1}{2}\right)^{k-2} = \underbrace{\sum_{m=0}^n \left(\frac{1}{2}\right)^k}_{k=m+2} \stackrel{\text{geometric series}}{=} \frac{1 - \left(\frac{1}{2}\right)^{n+1}}{1 - \frac{1}{2}} = 2 \left[1 - \left(\frac{1}{2}\right)^{n+1}\right]$$

$$y[n] = 2 \left[1 - \left(\frac{1}{2}\right)^{n+1}\right] u(n)$$

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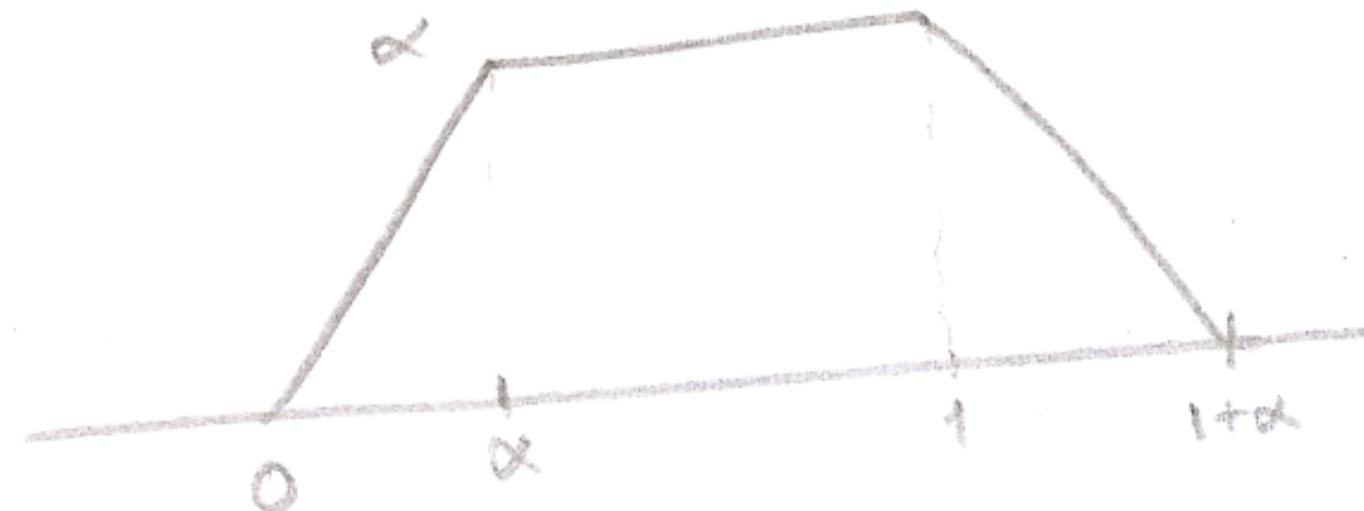
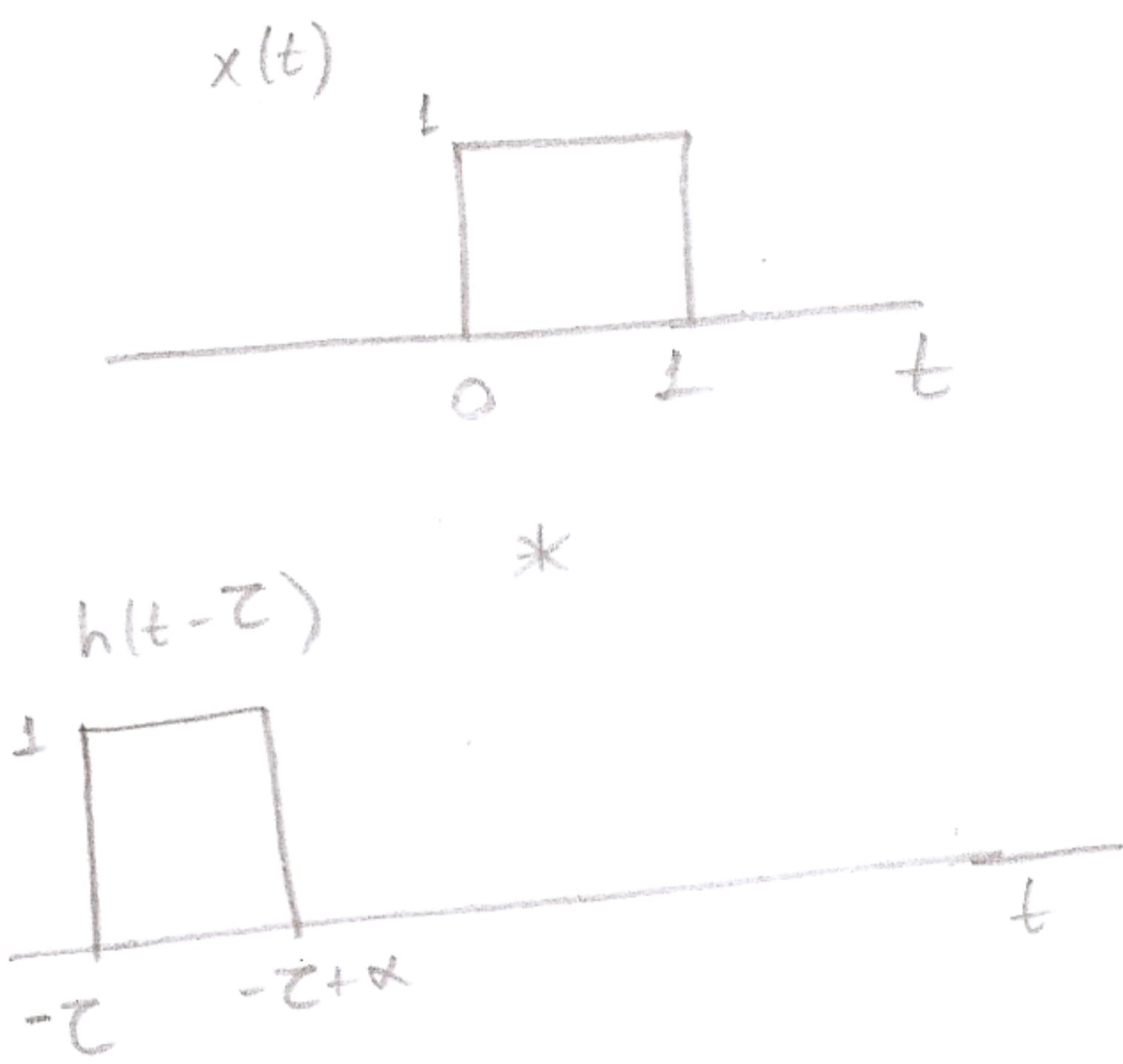
Determine and sketch convolution Ex 4 (2.10)

Suppose that

$$x(t) = \begin{cases} 1, & 0 \leq t \leq L \\ 0, & \text{elsewhere} \end{cases}$$

and $h(t) = x(t/\alpha)$, where $0 < \alpha \leq 1$

Determine and sketch $y(t) = x(t) * h(t)$



$$y(t) = \begin{cases} t, & 0 \leq t \leq \alpha \\ 1, & \alpha \leq t \leq L \\ 1-t, & L \leq t \leq (\alpha+L) \\ 0, & \text{elsewhere} \end{cases}$$

(20)

Compute convolution Ex 5: (2.11)

$$x(t) = u(t-3) - u(t-5) \quad \text{and} \quad h(t) = e^{-3t}u(t)$$

Compute $y(t) = x(t) * h(t)$

Sol'n: $y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$

$$= \int_{-\infty}^{\infty} [u(z-3) - u(z-5)] e^{-3(t-z)} dz$$

$\begin{cases} > 0 \\ < 0 \end{cases}$ $\begin{cases} > 0 \\ < 0 \end{cases}$

$\begin{cases} t > 3 \\ t < 5 \end{cases} \Rightarrow$ $\begin{cases} t > 3 \\ t < 5 \end{cases}$

$t-5 < 0 \Rightarrow t < 5$

combine this constraints

case 1:

$$3 < t < 5 \Rightarrow y(t) = \int_3^t e^{-3t+3z} dz = e^{-3t} \int_3^t e^{3z} dz$$

$$= e^{-3t} \left[\frac{e^{3z}}{3} \Big|_3^t \right]$$

$$= e^{-3t} \left[\frac{e^{3t}}{3} - \frac{e^9}{3} \right] = \frac{1 - e^{-3(t-3)}}{3}$$

case 2: $t > 5$

$$y(t) = \int_3^5 e^{-3t+2z} dz = \int_3^5 e^{-3t} \cdot e^{2z} dz = e^{-3t} \int_3^5 e^{2z} dz$$

$$= e^{-3t} \left[\frac{e^{2z}}{2} \Big|_3^5 \right] = e^{-3t} \left[\frac{e^{10} - e^6}{2} \right]$$

(21)

$$y(t) = \begin{cases} \frac{1 - e^{-3(t-2)}}{3}, & 3 \leq t < 5 \\ e^{-3t} \left[\frac{e^{15} - e^9}{3} \right], & t > 5 \\ 0, & \text{elsewhere} \end{cases}$$

(22)

Compute convolution Ex 6 (2.21) :

Compute the convolution $y[n] = x[n] * h[n]$ of the following pairs of signals

$$\left. \begin{array}{l} x[n] = \alpha^n u[n] \\ h[n] = \beta^n u[n] \end{array} \right\} \alpha \neq \beta$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k] = \sum_{k=-\infty}^{\infty} \alpha^k u[k] \beta^{n-k} u[n-k]$$

$$u[k] = 1 \text{ for } k \geq 0 \quad \left. \right\} n \geq k \geq 0$$

$$u[n-k] = 1 \text{ for } n \geq k$$

$$y[n] = \sum_{k=0}^n \alpha^k \beta^{n-k} = \beta^n \sum_{k=0}^n \alpha^k \beta^{-k} = \beta^n \sum_{k=0}^n \left(\frac{\alpha}{\beta}\right)^k$$

$$= \beta^n \left(\frac{1 - \left(\frac{\alpha}{\beta}\right)^{n+1}}{1 - \frac{\alpha}{\beta}} \right) = \beta^n \left(\frac{\beta^{n+1} - \alpha^{n+1}}{\beta^{n+1} - \alpha} \cdot \frac{\cancel{\beta}}{\cancel{\beta - \alpha}} \right)$$

from geometric series formula

$$= \frac{\beta^{n+1} - \alpha^{n+1}}{\beta - \alpha}$$

If $n < 0$, $y[n] = 0$. Hence,

$$y[n] = \left[\frac{\beta^{n+1} - \alpha^{n+1}}{\beta - \alpha} \right] u[n]$$

$$b) \quad x[n] = h[n] = \alpha^n u[n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} \alpha^k u[k] \alpha^{n-k} u[n-k]$$

$\left. \begin{array}{l} u[k] = 1 \text{ for } k \geq 0, \text{ so } k \geq 0 \\ u[n-k] = 1 \text{ for } n-k \geq 0, \text{ so } n-k \geq 0 \Rightarrow n \geq k \end{array} \right\} n \geq k \geq 0$

$$y[n] = \sum_{k=0}^n \alpha^k \alpha^{n-k} = \alpha^n \sum_{k=0}^n \alpha^0 = \alpha^n (n+1)$$

$$y[n] = 0 \text{ for } n < 0$$

$$\underline{y[n] = \alpha^n (n+1) u[n]}$$

$$c) \quad x[n] = \left(-\frac{1}{2}\right)^n u[n-4]$$

$$h[n] = 4^n u[2-n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k] = \sum_{k=-\infty}^{\infty} \left(-\frac{1}{2}\right)^k u[k-4] 4^{n-k} u[2-n+k]$$

$$u[k-4] = 1 \text{ for } k-4 \geq 0, \text{ so } k \geq 4$$

$$u[2-n+k] = 1 \text{ for } 2-n+k \geq 0 \text{ so } k \geq n-2$$

$$\underline{\text{case 1: } n-2 \geq 4 \Rightarrow n \geq 6}$$

$$y[n] = \sum_{k=n-2}^{\infty} \left(-\frac{1}{2}\right)^k 4^{n-k} = \sum_{k=n-2}^{\infty} \left(-\frac{1}{2}\right)^k \left(\frac{1}{4}\right)^k 4^n$$

$$= \sum_{k=n-2}^{\infty} \left(-\frac{1}{8}\right)^k \cdot 4^n = 4^n \sum_{k=n-2}^{\infty} \left(-\frac{1}{8}\right)^k = 4^n \left(-\frac{1}{8}\right)^{n-2} \left(\frac{1}{1+\frac{1}{8}}\right)$$

(24) 24 $= (-2)^{n+6} \cdot \underline{(9/8)}$

Case 2: $n-2 < 4$

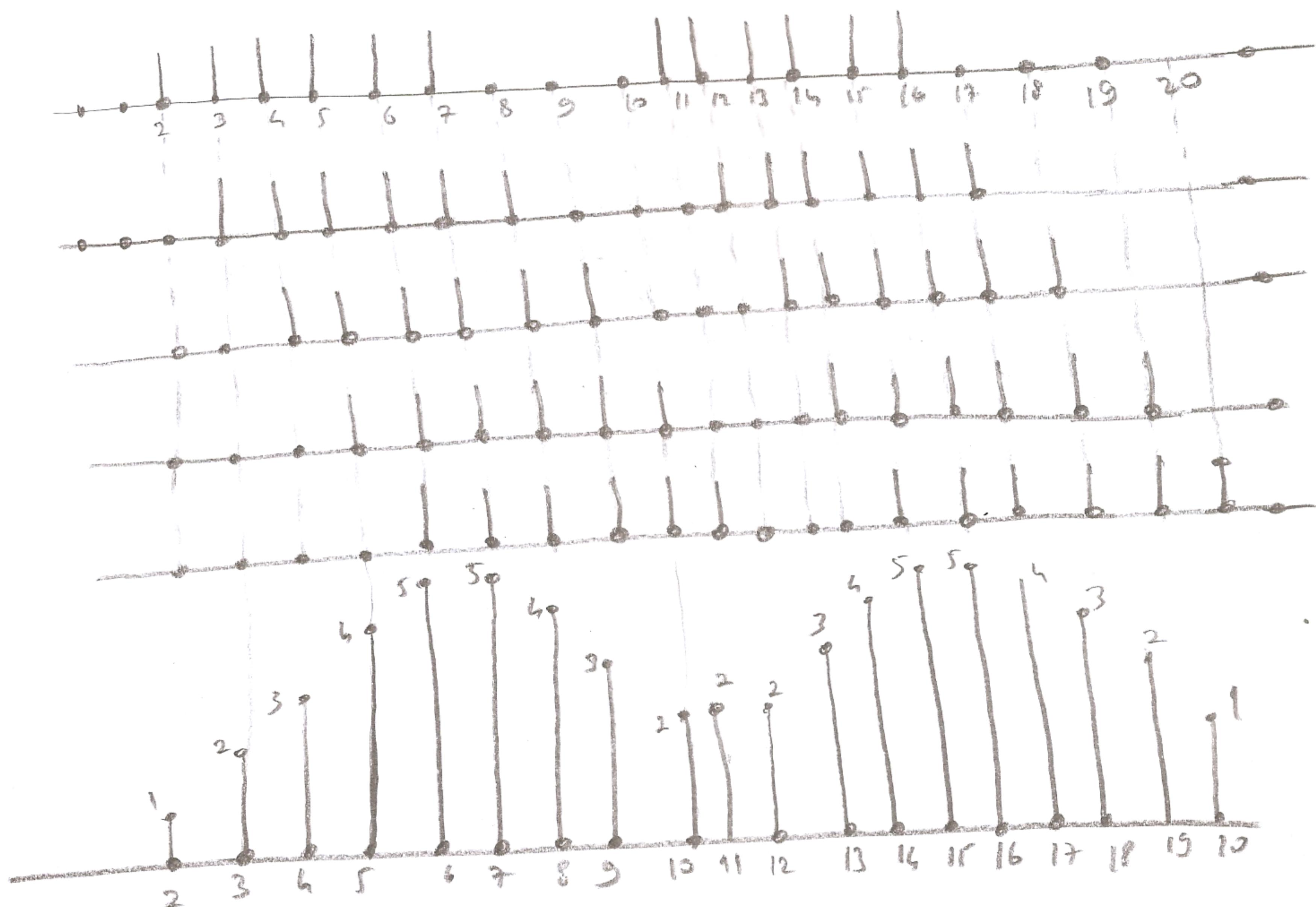
$$\begin{aligned}y[n] &= \sum_{k=4}^{\infty} \left(-\frac{1}{2}\right)^k 4^{n-k} = \sum_{k=4}^{\infty} \left(-\frac{1}{2}\right)^k \cdot 4^n \cdot 4^{-k} \\&= 4^n \sum_{k=4}^{\infty} \left(-\frac{1}{2}\right) \left(\frac{1}{4}\right)^k = 4^n \sum_{k=4}^{\infty} \left(-\frac{1}{8}\right)^k \\&= 4^n \left(-\frac{1}{8}\right)^4 \left(\frac{1}{1+\frac{1}{8}}\right) = 4^n \left(-\frac{1}{8}\right)^4 \left(\frac{8}{9}\right)\end{aligned}$$

$$y[n] = \begin{cases} (-2)^{-n+6} \cdot (8/9), & n \geq 6 \\ 4^n \left(-\frac{1}{8}\right)^4 \left(\frac{8}{9}\right), & n < 6 \\ 0, & \text{otherwise} \end{cases}$$

d)



$$y[n] = x[0] h[n] + x[1] h[n-1] + x[2] h[n-2] + \\ x[3] h[n-3] + x[4] h[n-4]$$



Cts-time convolution Ex 7 (22)

For each of the following pairs of waveforms, use the convolution integral to find the response $y(t)$ of the LTI system with impulse response $h(t)$ to the input $x(t)$.

$$a) x(t) = e^{-\alpha t} u(t)$$

$$h(t) = e^{-\beta t} u(t)$$

$$\alpha \neq \beta$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau = \int_{-\infty}^{\infty} e^{-\alpha \tau} u(\tau) e^{-\beta(t-\tau)} u(t-\tau) d\tau$$

$$u(\tau) = 1 \text{ for } \tau > 0$$

$$u(t-\tau) = 1 \text{ for } t-\tau > 0 \Rightarrow t > \tau \quad \left. \right\} 0 < \tau < t$$

$$\int_0^t e^{-\alpha \tau} e^{-\beta(t-\tau)} d\tau = \int_0^t e^{-\alpha \tau} e^{-\beta t} e^{\beta \tau} d\tau$$

$$= e^{-\beta t} \int_0^t e^{\tau(\beta-\alpha)} d\tau = e^{-\beta t} \left[\frac{e^{\tau(\beta-\alpha)}}{\beta-\alpha} \Big|_0^t \right]$$

$$= e^{-\beta t} \left[\frac{e^{t(\beta-\alpha)}}{\beta-\alpha} - \frac{1}{\beta-\alpha} \right] = e^{-\beta t} \left[\frac{e^{t(\beta-\alpha)} - 1}{\beta-\alpha} \right]$$

$$= \frac{e^{-\alpha t} - e^{-\beta t}}{\beta-\alpha} \quad y(t) \text{ is zero for } t < 0$$

$$y(t) = \left[\frac{e^{-\alpha t} - e^{-\beta t}}{\beta-\alpha} \right] u(t) \quad \alpha \neq \beta$$

$$b) \quad x(t) = u(t) - 2u(t-2) + u(t-5)$$

$$h(t) = e^{2t} u(1-t)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$= \int_{-\infty}^{\infty} [u(\tau) - 2u(\tau-2) + u(\tau-5)] e^{2(t-\tau)} u(1-t+\tau) d\tau$$

differ from zero

if $0 < \tau < 5$

① Also for $0 < \tau < 2$ equal to 1
 $2 < \tau < 5$ equal to -1

$$= \int_0^2 h(t-\tau) d\tau - \int_2^5 h(t-\tau) d\tau$$

$$= \int_0^2 e^{2(t-\tau)} u(1-t+\tau) d\tau - \int_2^5 e^{2(t-\tau)} u(1-t+\tau) d\tau$$

② $1-t+\tau > 0$ combine ① and ② constraints
 $\tau > t-1$ and split range of t according to this
constraints.

If $t \leq 1$, range of t depends on constraint ①

$$y(t) = \int_0^2 e^{2(t-\tau)} d\tau - \int_2^5 e^{2(t-\tau)} d\tau$$

$$= \int_0^2 e^{2t} \cdot e^{-2\tau} d\tau - \int_2^5 e^{2t} \cdot e^{-2\tau} d\tau$$

(28)

$$\begin{aligned}
&= e^{2t} \left[\int_0^2 e^{-2z} dz - \int_2^5 e^{-2z} dz \right] \\
&= e^{2t} \left[-\frac{e^{-2z}}{2} \Big|_0^2 + \frac{e^{-2z}}{2} \Big|_2^5 \right] \\
&= e^{2t} \left[-\frac{e^{-4}}{2} + \frac{1}{2} + \frac{e^{-10}}{2} - \frac{e^{-4}}{2} \right] \\
&= \frac{e^{2t}}{2} \left[+e^{-10} + 1 - 2e^{-4} \right]
\end{aligned}$$

If $1 < t < 3$, smaller part of constraint ① constrained $t-1$

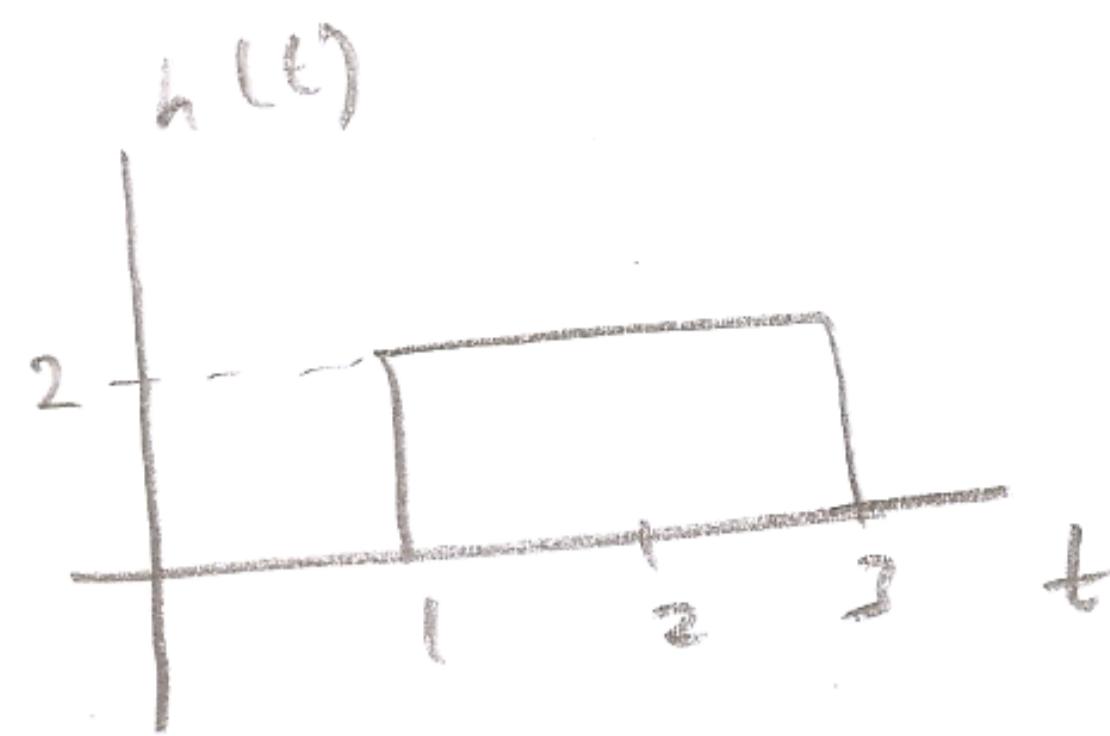
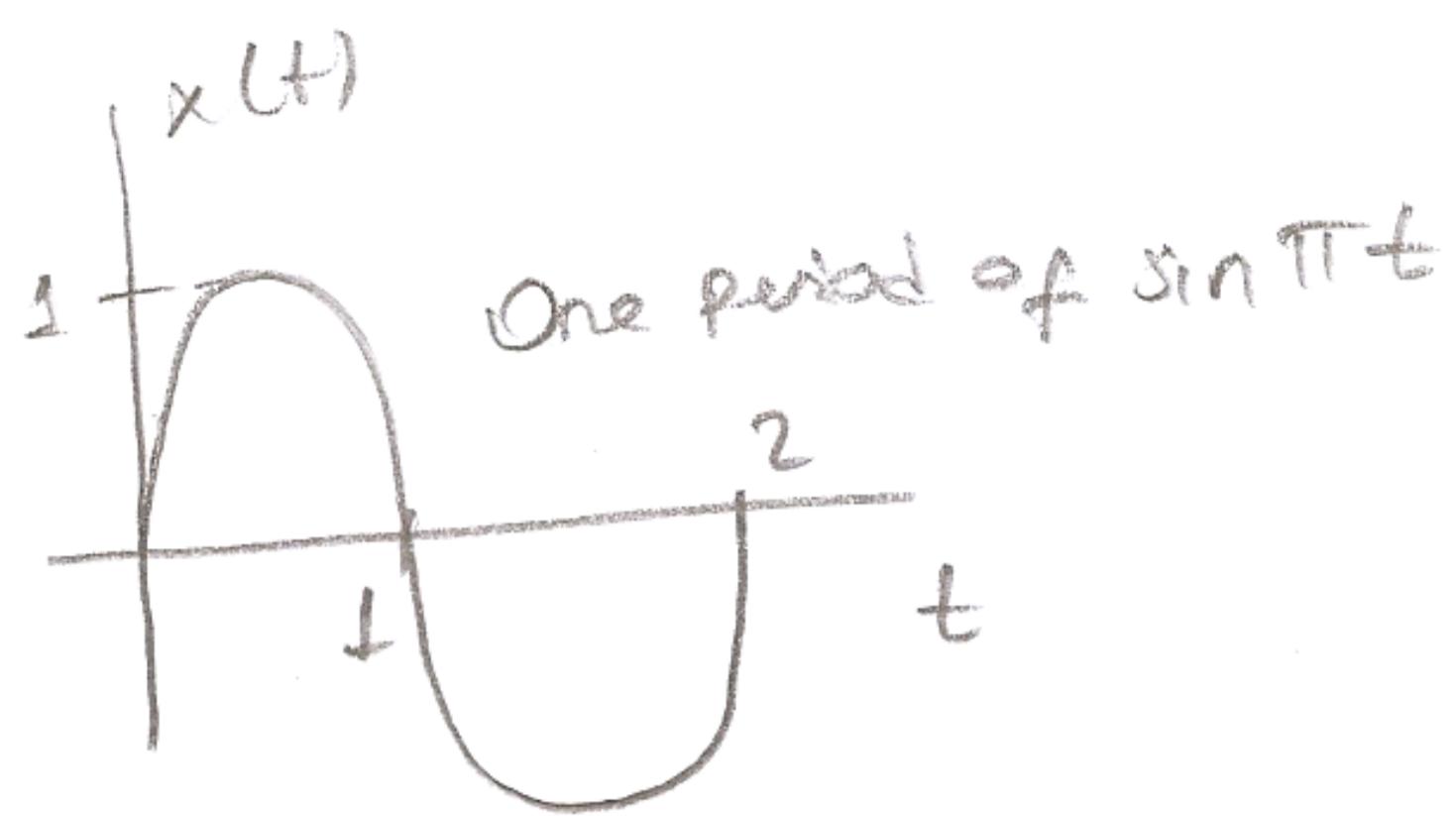
$$\begin{aligned}
y(t) &= \int_{t-1}^2 e^{2(t-z)} dz - \int_2^5 e^{2(t-z)} dz \\
&= \frac{e^{2t}}{2} \left[-e^{-4} + e^{-2t+2} + e^{-10} - e^{-4} \right] = \frac{e^{2t}}{2} \left[e^{-10} - 2e^{-4} + e^{-2t+2} \right]
\end{aligned}$$

If $3 \leq t < 6$, smaller part of constraint ② constrained $t-1$

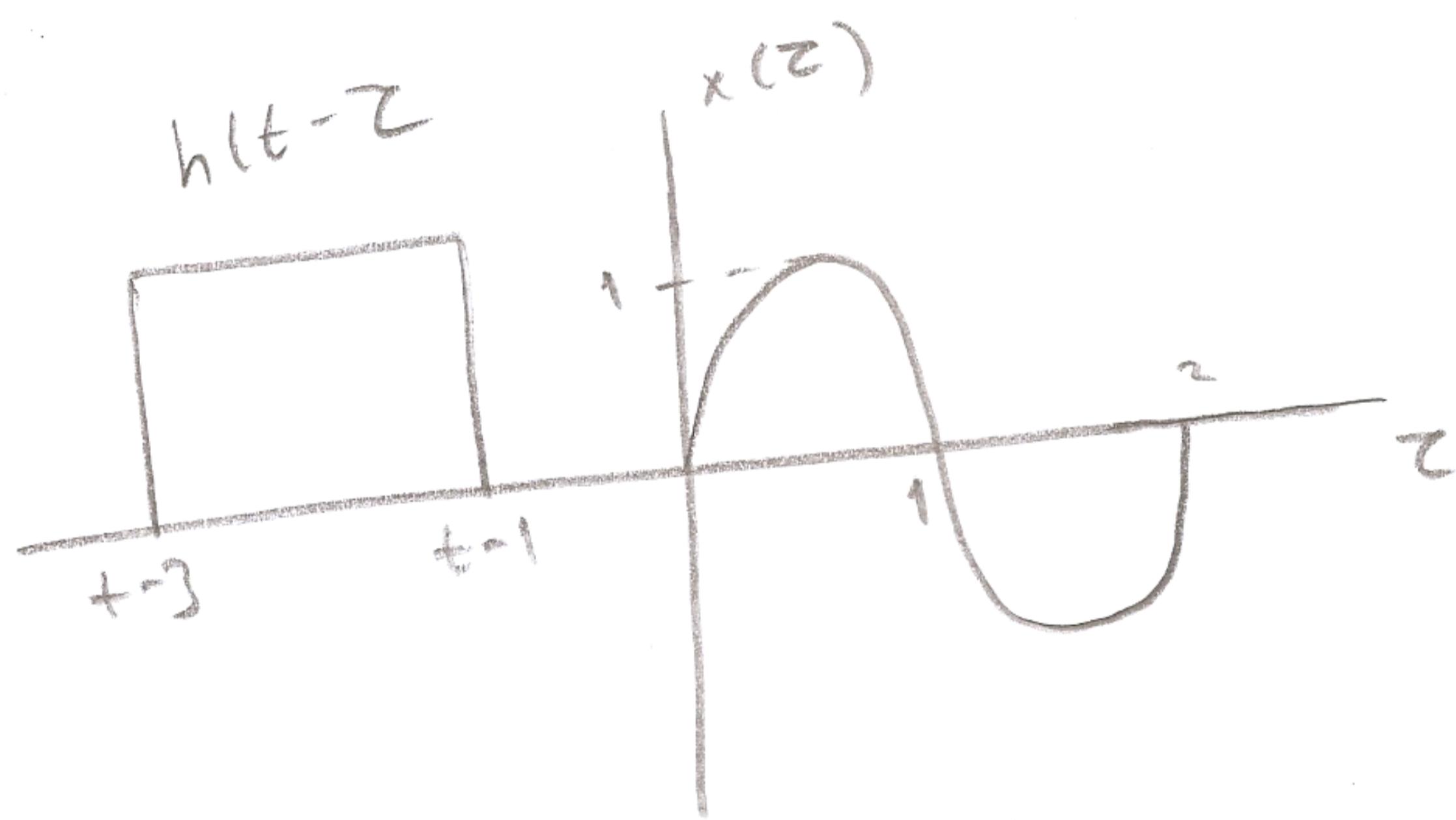
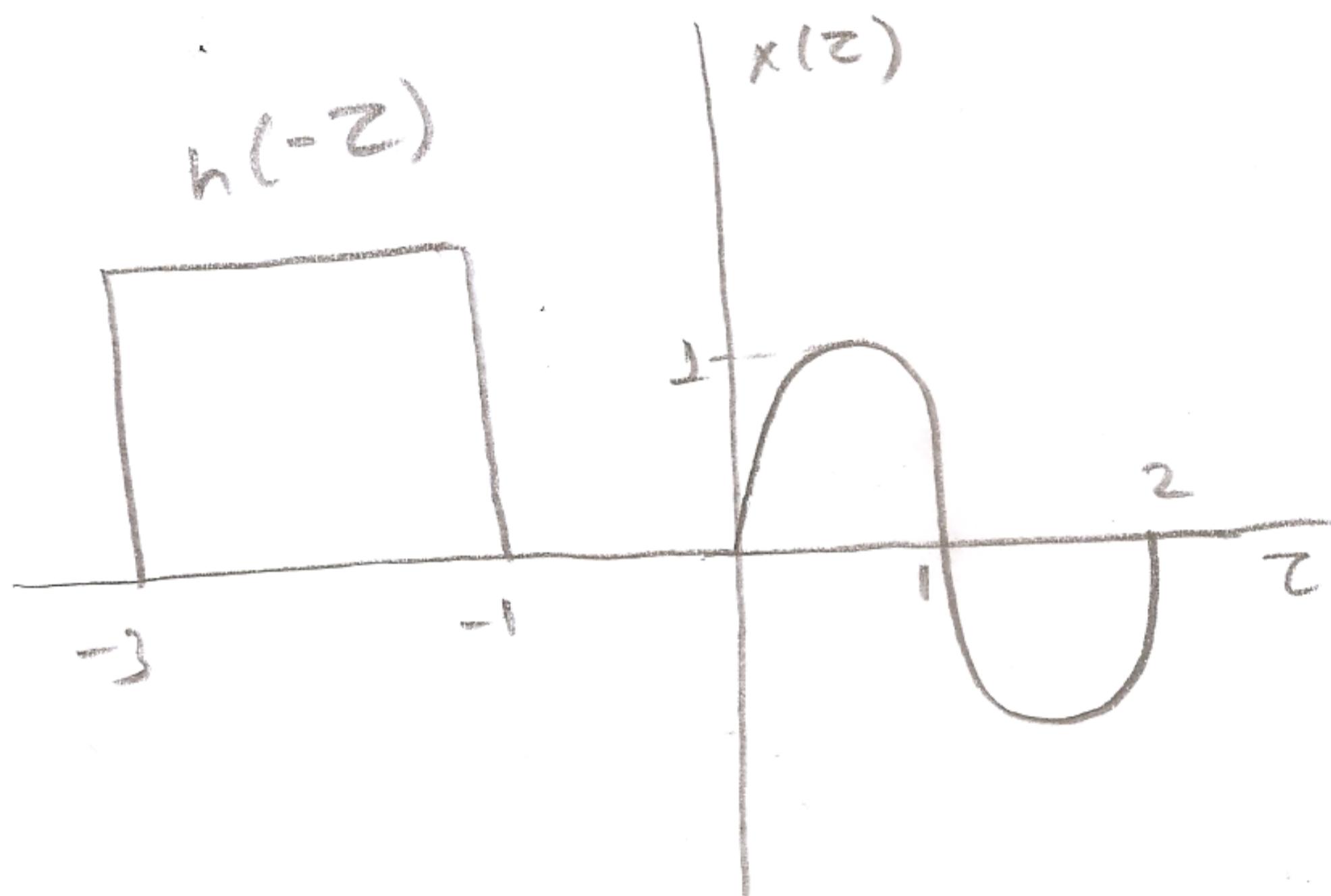
$$y(t) = - \int_{t-1}^5 e^{2(t-z)} dz = \frac{e^{2t}}{2} \left[e^{-10} - e^{-2t+2} \right]$$

$$y(t) = \begin{cases} \frac{e^{2t}}{2} [e^{-10} - 2e^{-4} + 1], & t < 1 \\ \frac{e^{2t}}{2} [e^{-10} - 2e^{-4} + e^{-2t+2}], & 1 < t < 3 \\ \frac{e^{2t}}{2} [e^{-10} - e^{-2t+2}], & 3 \leq t < 6 \\ 0, & t \geq 6 \end{cases}$$

c)



$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$



$$y(t) = \begin{cases} 0 & t < 1 \\ \int_0^{t-1} 2\pi z \sin z dz, & 1 \leq t < 3 \\ \int_0^2 2\pi z \sin z dz, & 3 \leq t < 5 \\ 0 & t \geq 5 \end{cases}$$

(30)

(7)

$$y(t) = \begin{cases} 0 & , t < 1 \\ \frac{2}{\pi} [1 - \cos(\pi(t-1))] & , 1 \leq t < 3 \\ \frac{2}{\pi} [\cos((t-3)\pi) - 1] & , 3 \leq t < 5 \\ 0 & , t \geq 5 \end{cases}$$

(31)