Relaying Strategies for the Two-Way Gaussian Relay Channel

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- 1 Basic Principles
- 2 Relaying Strategies
- 3 Sum Rate Comparison
- 4 Conclusion

The Network Model

The Network Model



■ Three nodes / devices

Basic Principles

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Sum Rate Comparison



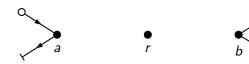
■ Three nodes / devices

Basic Principles

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One message from a to b

The Network Model



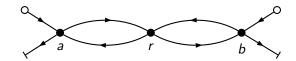
■ Three nodes / devices

Basic Principles

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- One message from a to b
- One message from b to a

The Network Model



- Three nodes / devices
- One message from a to b
- One message from b to a
- No direct link

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Sum Rate Comparison

Start with Toy Problems

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Sum Rate Comparison

No noise at the nodes

Start with Toy Problems

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The resulting model can illustrate the basic ideas behind network coding (NC) and physical layer network coding (PLNC).

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Sum Rate Comparison

Exchanging 2 bits x_1 and x_2 with a routing approach:



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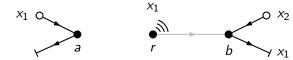
Exchanging 2 bits x_1 and x_2 with a routing approach:





Sum Rate Comparison

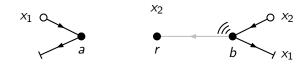
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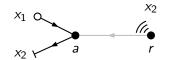
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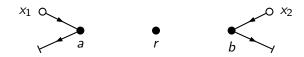


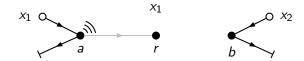
Routing

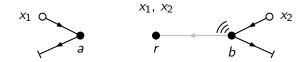
Exchanging 2 bits x_1 and x_2 with a routing approach:



sum rate: 2 bits in 4 transmissions







Network Coding







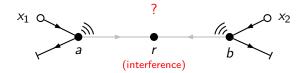
Network Coding

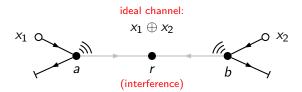
The relay node r can reach both a and b. Broadcasting saves one transmission.



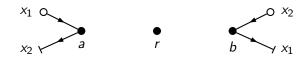
sum rate: 2 bits in 3 transmissions











The relay node doesn't need to know the individual bits.



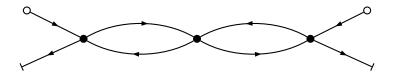
sum rate: 2 bits in 2 transmissions (for ideal channel)

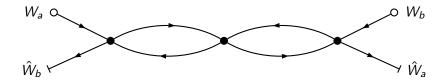
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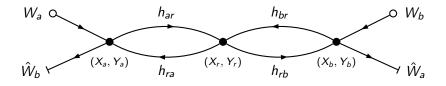
Obvious Question

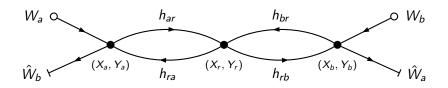
How does it work for more realistic channel models?

Two-Way Gaussian Relay Channel Model





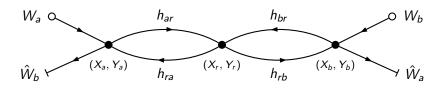




$$Y_r = h_{ar}X_a + h_{br}X_b + Z_r$$

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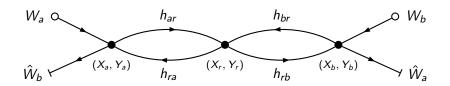


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Channel coefficients h_{ii} are time-invariant.

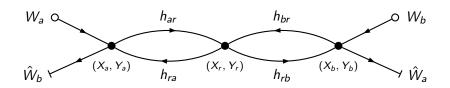


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Channel coefficients h_{ij} are time-invariant. The inputs X_a, X_b and X_r are average power constrained to P.

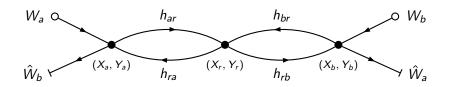


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Two-phase protocol:

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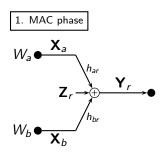
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What are achievable rate pairs (R_a, R_b) (in bits per channel use) and achievable sum rates $R = R_a + R_b$?

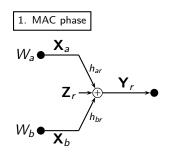
See [Rankov-Wittneben '05].

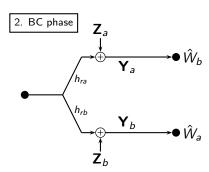
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Amplify-and-Forward (AF)

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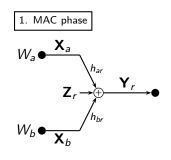


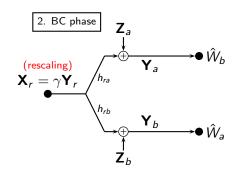


Sum Rate Comparison

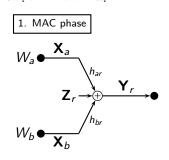
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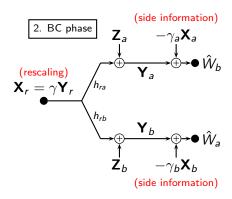
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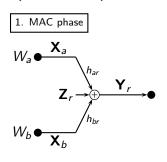
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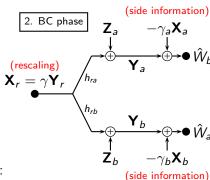




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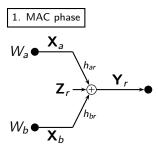


$$\begin{split} R_{a} &< \frac{1}{4} \log_{2} \left(1 + \frac{h_{ra}^{2} h_{br}^{2} P}{h_{ra}^{2} P + h_{ar}^{2} P + h_{br}^{2} P + 1} \right) \\ R_{b} &< \frac{1}{4} \log_{2} \left(1 + \frac{h_{rb}^{2} h_{ar}^{2} P}{h_{rb}^{2} P + h_{br}^{2} P + 1} \right) \end{split}$$

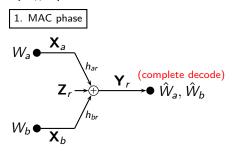
See [Knopp '06].

Sum Rate Comparison

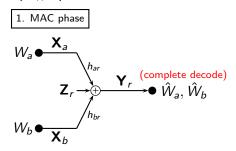
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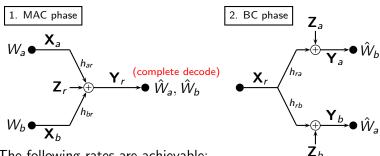


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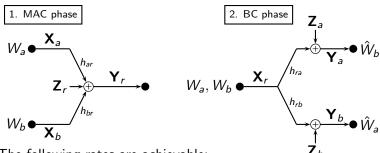
$$R_a < rac{1}{4} \log_2(1 + h_{ar}^2 P)$$
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See [Knopp '06].



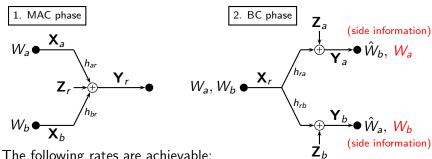
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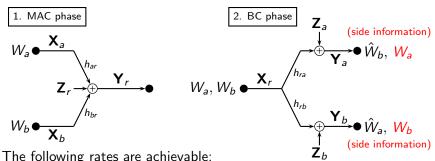
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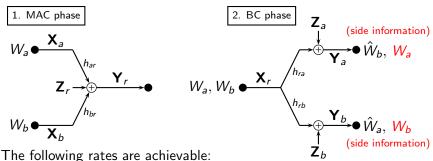
See [Knopp '06].



$$R_a < \min\left(rac{1}{4}\log_2(1+h_{ar}^2P), rac{1}{4}\log_2(1+h_{rb}^2P)
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Sum Rate Comparison

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PLNC with Structured Codes

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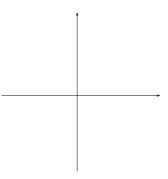
Multiplexing loss: the relay tries to understand something that it doesn't really need to know (i.e. both messages individually).

PLNC with Structured Codes

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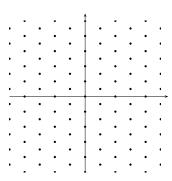
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- PLNC idea: the relay should only try to understand a linear combination of the transmitted messages.
- Translation of this linear combination of messages to the signals (physical layer) can be done with nested lattice codes.

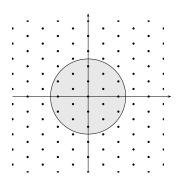


Lattices, Lattice Codes and Nested Lattice Codes

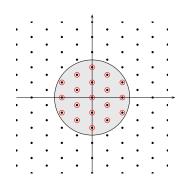
■ Lattice Λ_f



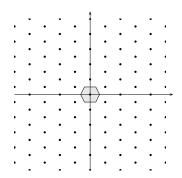
- Lattice Λ_f
- lacksquare Region ${\cal R}$



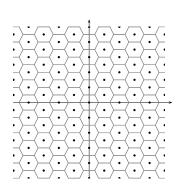
- Lattice Λ_f
- \blacksquare Region \mathcal{R}
- Lattice code $(\Lambda_f + \mathbf{t}) \cap \mathcal{R}$



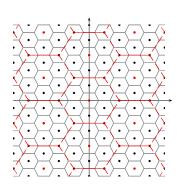
- Lattice Λ_f
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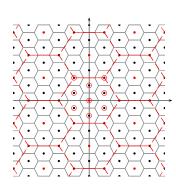
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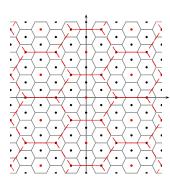


- Lattice Λ_f
- \blacksquare Region \mathcal{R}
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- Fundamental region $V(\Lambda_f)$
- Nested lattice $\Lambda \subseteq \Lambda_f$

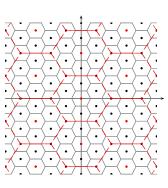


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- Nested lattice $\Lambda \subset \Lambda_f$
- Nested lattice code $\mathcal{C} = (\Lambda_f + \mathbf{t}) \cap \mathcal{V}(\Lambda)$

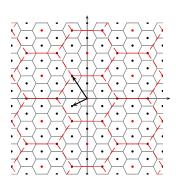




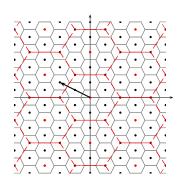
All nodes now use the same nested lattice code.



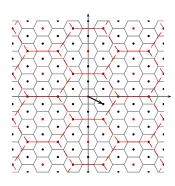
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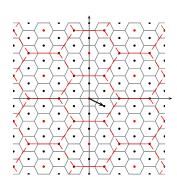
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- Relay receives $\mathbf{V}_a + \mathbf{V}_b$.



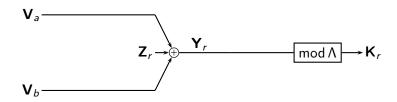
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- Side information allows each user. to recover the lattice point of the other user.

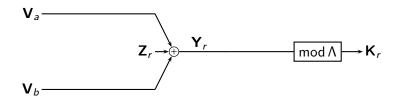


Sum Rate Comparison

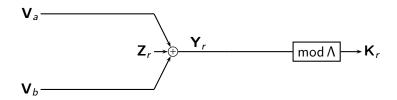


Sum Rate Comparison

See [Baik-Chung '08].

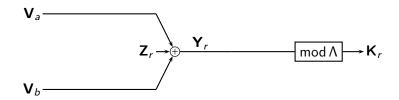


For easier illustration the channel gains are all assumed to be 1.



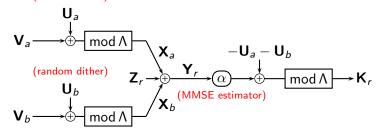
The number of lattice points that can be reliably distinguished at the relay is bounded by $M < \frac{\text{Vol}(\mathcal{V}(\Lambda))}{\text{Vol}_{\text{noise}}} \approx P^{n/4}$ for very large n and ideal shaping gain of Λ . Then the rate is limited by

$$R < \frac{1}{n}\log_2 M = \frac{1}{4}\log_2(P).$$



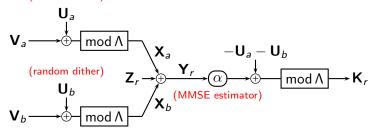
We can do better!

(random dither)



See [Baik-Chung '08].

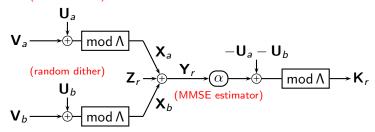
(random dither)



Dither variables are uniformly distributed over $\mathcal{V}(\Lambda)$ but known to all nodes.

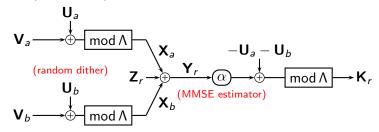
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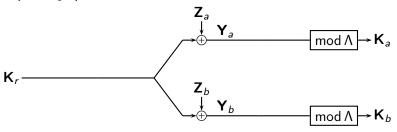
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- Introduction of the linear MMSE estimator α reduces the effective noise.

(random dither)

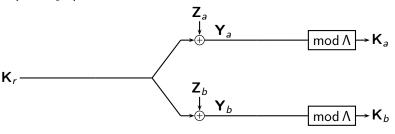


- Dither variables are uniformly distributed over $\mathcal{V}(\Lambda)$ but known to all nodes.
- Introduction of the linear MMSE estimator α reduces the effective noise.
- $\mathbf{K}_r = (\alpha \mathbf{Y}_r \mathbf{U}_a \mathbf{U}_b) \mod \Lambda = (\mathbf{V}_a + \mathbf{V}_b + \mathbf{Z}_{eff}) \mod \Lambda$

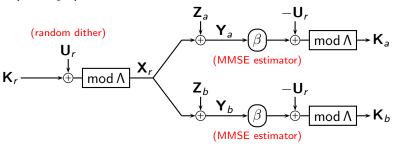
Sum Rate Comparison



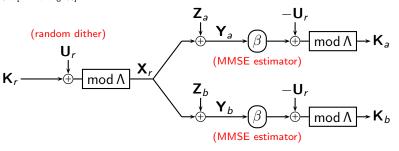
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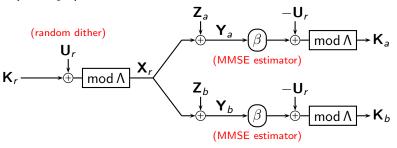
■ Relay broadcasts the noisy lattice point.



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- MMSE estimator and dither for effective noise reduction.



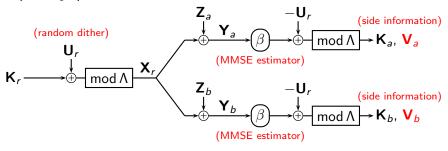
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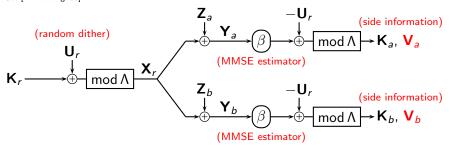


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See [Baik-Chung '08].



Remark: For general channel conditions superposition coding can be employed to improve the achievable rates.

Sum Rate Comparison

Comparison of achievable sum rates $R = R_a + R_b$ when all channel gains are assumed to be 1.

Sum Rate Comparison

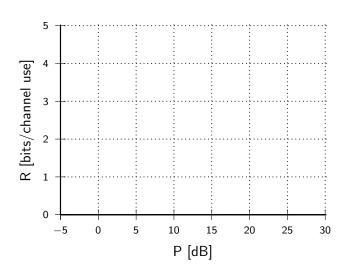
Comparison of achievable sum rates $R = R_a + R_b$ when all channel gains are assumed to be 1.

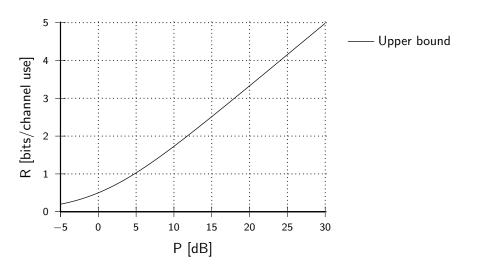
Upper bound:
$$R < \frac{1}{2}\log_2\left(1+P\right)$$

$$AF: \qquad R < \frac{1}{2}\log_2\left(1+P\frac{P}{3P+1}\right)$$

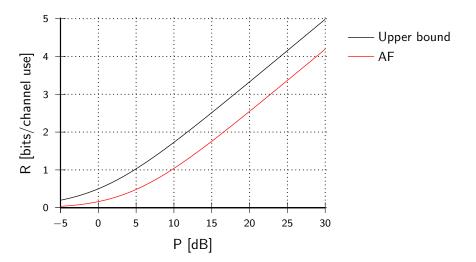
$$DF: \qquad R < \frac{1}{4}\log_2\left(1+2P\right)$$

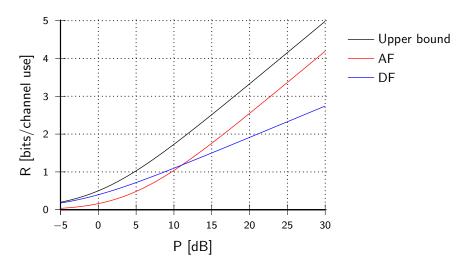
$$MF: \qquad R < \frac{1}{2}\log_2\left(\frac{1}{2}+\frac{P}{2}\right)$$

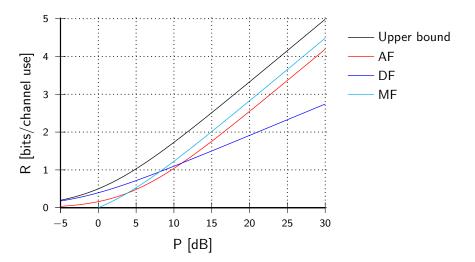




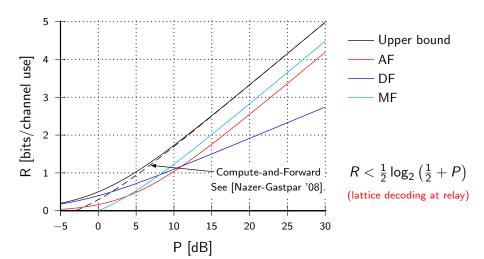












Conclusion

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- Involved principles: physical layer network coding, broadcasting, using side information
- In certain networks the interference of users can be harnessed.
- However a strategy to show achievability of the upper bound for this network for all channel conditions is still not available.

- I.-J. Baik and S.-Y. Chung.
 - Network coding for two-way relay channels using lattices.
 - 2008, IEEE International Conference on Communications
- R. Knopp.
 - Two-way radio networks with a star topology.
 - 2006. International Zurich Seminar on Communications
- B. Nazer and M. Gastpar. Compute-and-forward: harnessing interference with structured codes.
 - 2008, International Symposium on Information Theory
- B. Rankov and A. Wittneben. Spectral efficient signaling for half-duplex relay channels. 2005, Proc. of Asilomar Conference on Signals, Systems and Computers