Bidirectional Multi-Hop Communication via Two Relays Using Nested Voronoi Codes

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OUTLINE

- 1. Basic Principles
- 2. One Relay (Separated Two-Way Relay Channel)
- 3. Two Relays (Separated Two-Way Two-Relay Channel)
- 4. Conclusion



• Three nodes / devices

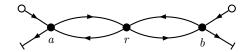


- Three nodes / devices
- ullet One message from a to b



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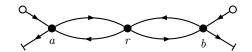
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- Three nodes / devices
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- No direct link

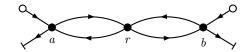
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Message Exchange via a Relay



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Start with toy problems:



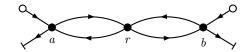
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Basic Principles •0000

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No noise at the nodes

Message Exchange via a Relay

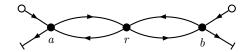


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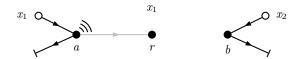
Start with toy problems:

- No noise at the nodes
- Binary input and output alphabets
- Half-duplex constraint (i.e. a node either listens or talks but not both at the same time)

ROUTING

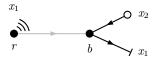


ROUTING

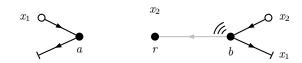


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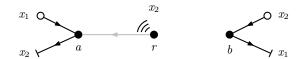




ROUTING



ROUTING



ROUTING



ROUTING

Exchanging 2 bits x_1 and x_2 with a routing approach:



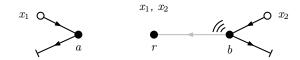
sum rate: 2 bits in 4 transmissions



 x_2



NETWORK CODING



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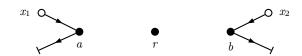
NETWORK CODING

The relay node r can reach both a and b. Broadcasting saves one transmission.



sum rate: 2 bits in 3 transmissions

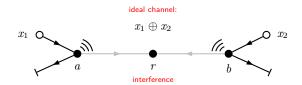
Physical-Layer Network Coding



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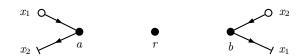
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The relay node doesn't need to know the individual bits.

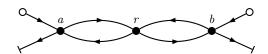


sum rate: 2 bits in 2 transmissions (for ideal channel)

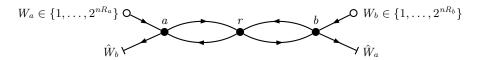
OBVIOUS QUESTION

How does it work for more realistic channel models?

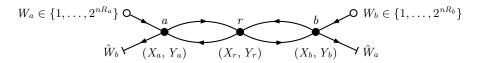
GAUSSIAN CHANNEL MODEL



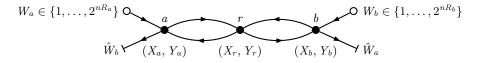
Gaussian Channel Model



GAUSSIAN CHANNEL MODEL



Gaussian Channel Model



Additive Gaussian noise:

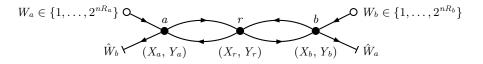
$$Y_a = X_r + Z_a$$

$$Y_r = X_a + X_b + Z_r$$

$$Y_b = X_r + Z_b$$

Successive channel uses:

$$\boldsymbol{X}_a = (X_{a,1}, \dots, X_{a,n})$$
, etc.



Additive Gaussian noise:

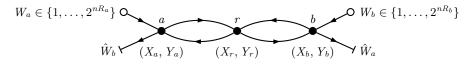
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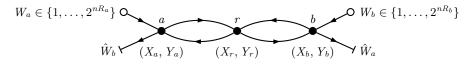
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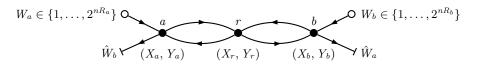
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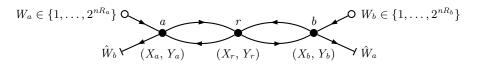
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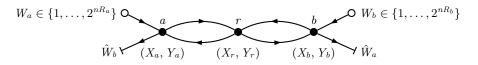
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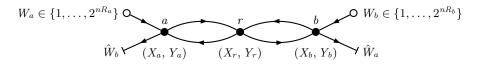
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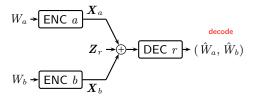
What (R_a, R_b) are achievable $(p_e \to 0 \text{ as } n \to \infty)$?

DECODE-AND-FORWARD (DF)

See [Rankov-Wittneben '05], [Knopp '06].

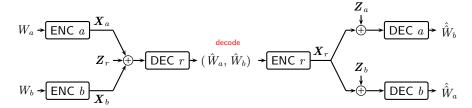
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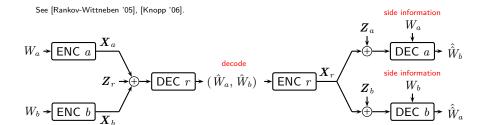


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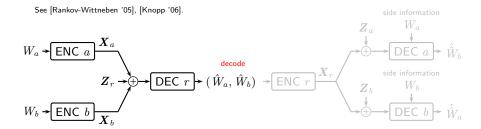
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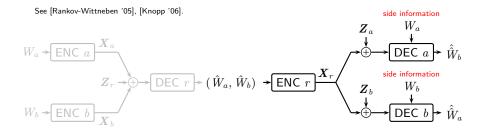


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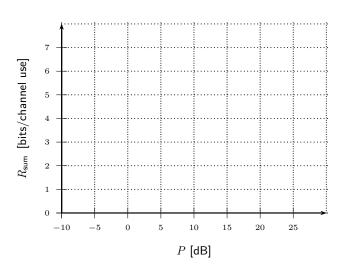


Multi-acces channel

Decode-And-Forward (DF)

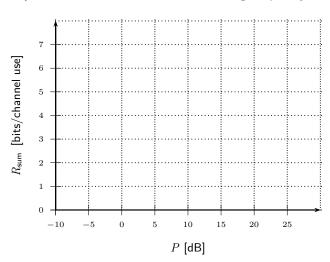


- Multi-acces channel
- Broadcasting with user side information

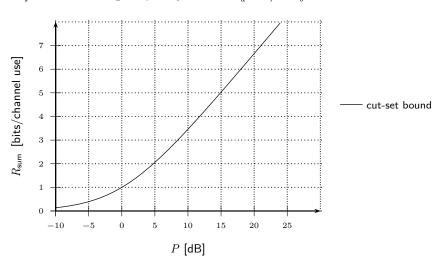


Sum Rate
$$R_{\text{sum}} = R_a + R_b$$

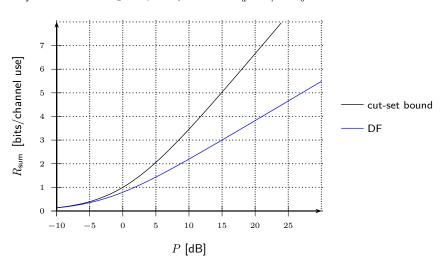
Symmetric case:
$$P_a = P_r = P_b = P$$
 and $\sigma_a^2 = \sigma_r^2 = \sigma_b^2 = 1$



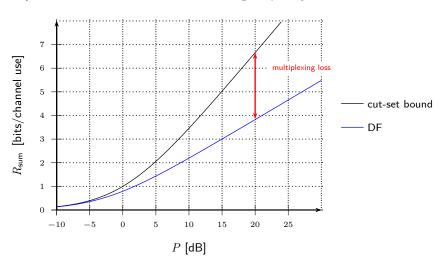
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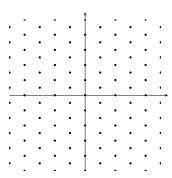


 Multiplexing loss: the relay tries to understand something that it doesn't really need to know, i.e., both messages individually.

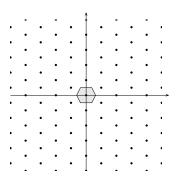
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- Use structured codes: Voronoi codes

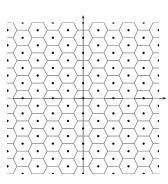
• (coding) Lattice Λ_c



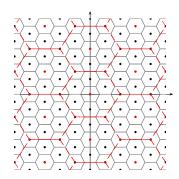
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- Voronoi regions

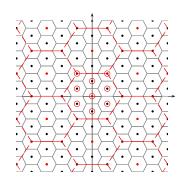


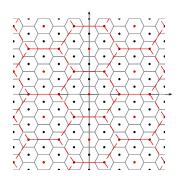
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- Nested (shaping) lattice $\Lambda \subset \Lambda_c$



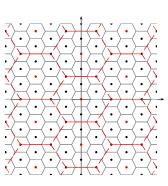
- (coding) Lattice Λ_c
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- Voronoi code:

$$\mathcal{C}(\Lambda_c/\Lambda) = (\Lambda_c + \mathbf{t}) \cap \mathcal{R}_V(\Lambda)$$

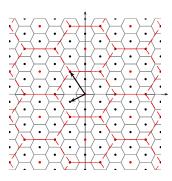




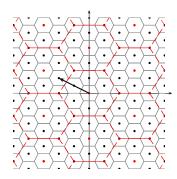
• Users a and b use the same Voronoi code $\mathcal{C}(\Lambda_c/\Lambda)$.



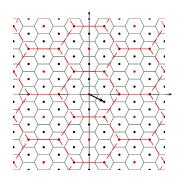
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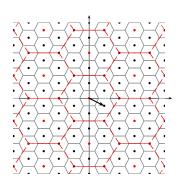
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- Side information allows each user to recover the lattice point of the other user based on $oldsymbol{V}$



RESULT FOR ARBITRARY CHANNEL CONDITIONS

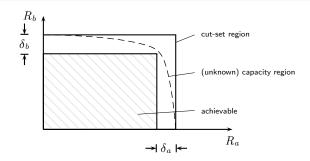
RESULT FOR ARBITRARY CHANNEL CONDITIONS

Theorem

The capacity region is achievable to within 1/2 bits per dimension.

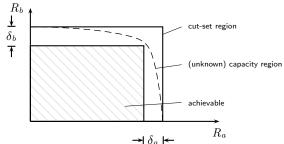
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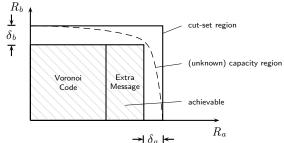
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Achievability strategy:

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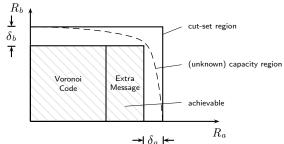


Achievability strategy:

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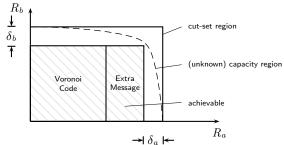


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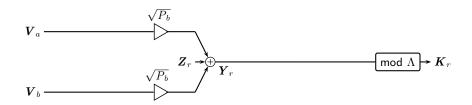
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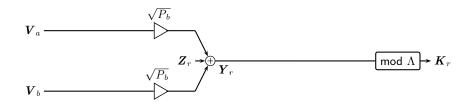


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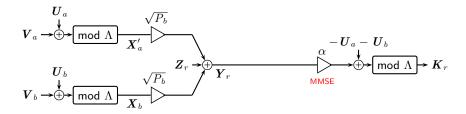
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(see also [Nam-Chung-Lee '10])

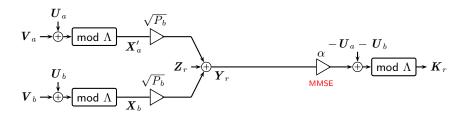




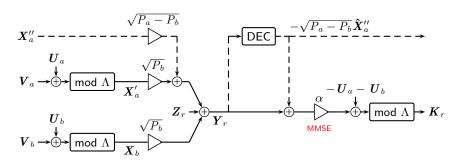
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- Dithering and linear MMSE estimation for effective noise reduction

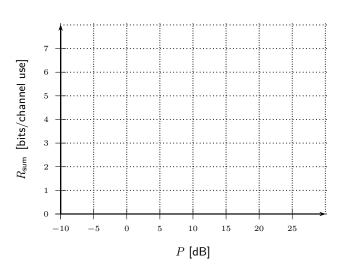


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- $K_r = (\alpha Y_r U_a U_b) \mod \Lambda = (V_a + V_b + \tilde{Z}_r) \mod \Lambda$



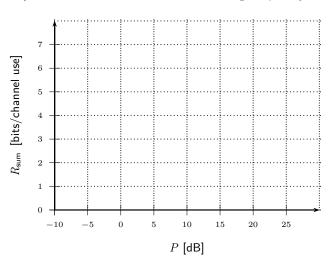
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- $K_r = (\alpha Y_r U_a U_b) \mod \Lambda = (V_a + V_b + \tilde{Z}_r) \mod \Lambda$
- Superposition coding with successive cancellation

Sum Rate $R_{\text{sum}} = R_a + R_b$



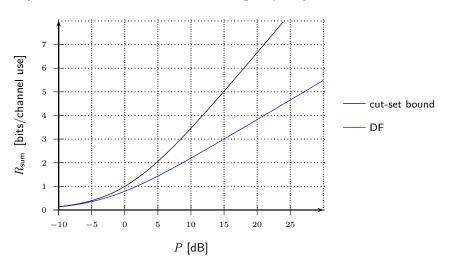
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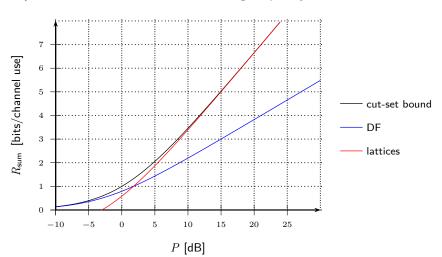
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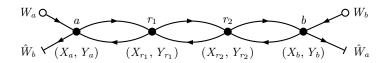


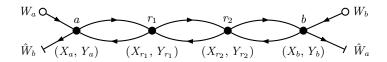
Sum Rate
$$R_{\text{sum}} = R_a + R_b$$

Symmetric case:
$$P_a = P_r = P_b = P$$
 and $\sigma_a^2 = \sigma_r^2 = \sigma_b^2 = 1$



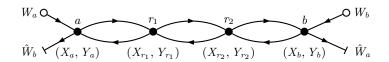
Two Relays •000000





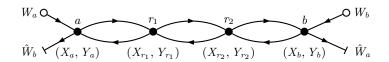
Multi-hop extension of the separated two-way relay channel

Two Relays

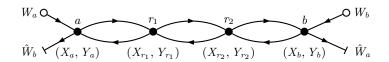


- Multi-hop extension of the separated two-way relay channel
- Both relays are necessary to enable the message exchange.

Two Relays



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- No sinks or sources at the relays

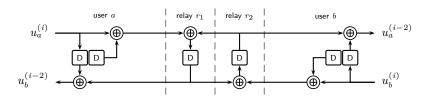


- Multi-hop extension of the separated two-way relay channel
- Both relays are necessary to enable the message exchange.
- No sinks or sources at the relays
- Transmission strategy: physical-layer network coding with nested Voronoi codes

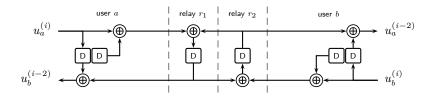
 No noise, binary inputs and outputs, finite field physical layer, full-duplex nodes

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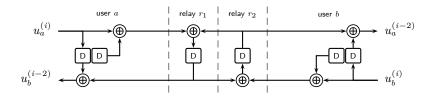


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- Each user "sees" the packet stream of the other user delayed by two blocks.
- User rates approach 1 bit per channel use for many blocks.

Channel model:

$$Y_a = X_{r_1} + Z_a$$

$$Y_{r_1} = X_a + X_{r_2} + Z_{r_1}$$

$$Y_{r_2} = X_{r_1} + X_b + Z_{r_2}$$

$$Y_b = X_{r_2} + Z_b$$

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- Same power constraint P for all nodes
- Gaussian noise with zero mean, variances $\sigma_a^2, \, \sigma_{r_1}^2, \, \sigma_{r_2}^2, \, \sigma_b^2$

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- Successive channel uses, transmission block i: $X_a^{(i)}$, etc.

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Theorem

One can achieve all rate pairs (R_a, R_b) satisfying

$$\begin{split} R_{a} &< \min \left(\tilde{\mathbf{C}} \left(P/\sigma_{r_{1}}^{2} \right), \tilde{\mathbf{C}} \left(P/\sigma_{r_{2}}^{2} \right), \mathbf{C} \left(P/\sigma_{b}^{2} \right) \right) \\ R_{b} &< \min \left(\tilde{\mathbf{C}} \left(P/\sigma_{r_{2}}^{2} \right), \tilde{\mathbf{C}} \left(P/\sigma_{r_{1}}^{2} \right), \mathbf{C} \left(P/\sigma_{a}^{2} \right) \right) \end{split}$$

where $C(x) \triangleq \log_2(1/2 + x)/2$ and $C(x) \triangleq \log_2(1 + x)/2$.

Channel model:

$$Y_a = X_{r_1} + Z_a$$

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- Gaussian noise with zero mean, variances $\sigma_a^2, \, \sigma_{r_1}^2, \, \sigma_{r_2}^2, \, \sigma_b^2$
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where $C(x) \triangleq \log_2(1/2 + x)/2$ and $C(x) \triangleq \log_2(1 + x)/2$.

Capacity region to within 1/2 bit per dimension

Channel model:

$$Y_a = X_{r_1} + Z_a$$

$$Y_{r_1} = X_a + X_{r_2} + Z_{r_1}$$

$$Y_{r_2} = X_{r_1} + X_b + Z_{r_2}$$

$$Y_b = X_{r_2} + Z_b$$

- Same power constraint P for all nodes
- Gaussian noise with zero mean, variances $\sigma_a^2,\,\sigma_{r_1}^2,\,\sigma_{r_2}^2,\,\sigma_b^2$
- Successive channel uses, transmission block i: $X_a^{(i)}$, etc.

Theorem

One can achieve all rate pairs (R_a, R_b) satisfying

$$R_{a} < \min \left(\tilde{C} \left(P/\sigma_{r_{1}}^{2} \right), \tilde{C} \left(P/\sigma_{r_{2}}^{2} \right), C \left(P/\sigma_{b}^{2} \right) \right)$$

$$R_{b} < \min \left(\tilde{C} \left(P/\sigma_{r_{2}}^{2} \right), \tilde{C} \left(P/\sigma_{r_{1}}^{2} \right), C \left(P/\sigma_{a}^{2} \right) \right)$$

where $\tilde{C}(x) \triangleq \log_2(1/2 + x)/2$ and $C(x) \triangleq \log_2(1 + x)/2$.

- Capacity region to within 1/2 bit per dimension
- Essentially optimal at high SNR

MLAN CONVERSION

e [Erez-Zamir '08].	
	J

MLAN Conversion

See [Erez-Zamir '08].

Blockwise modulo-lattice additive noise channel:

$$\begin{split} \tilde{\boldsymbol{Y}}_a^{(i)} &= (\tilde{\boldsymbol{X}}_{r_1}^{(i)} + \tilde{\boldsymbol{Z}}_a^{(i)}) \mod \Lambda \\ \tilde{\boldsymbol{Y}}_{r_1}^{(i)} &= (\tilde{\boldsymbol{X}}_a^{(i)} + \tilde{\boldsymbol{X}}_{r_2}^{(i)} + \tilde{\boldsymbol{Z}}_{r_1}^{(i)}) \mod \Lambda \\ \tilde{\boldsymbol{Y}}_{r_2}^{(i)} &= (\tilde{\boldsymbol{X}}_{r_1}^{(i)} + \tilde{\boldsymbol{X}}_b^{(i)} + \tilde{\boldsymbol{Z}}_{r_2}^{(i)}) \mod \Lambda \\ \tilde{\boldsymbol{Y}}_b^{(i)} &= (\tilde{\boldsymbol{X}}_{r_2}^{(i)} + \tilde{\boldsymbol{Z}}_b^{(i)}) \mod \Lambda \end{split}$$

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MLAN Conversion

See [Erez-Zamir '08].

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where

•
$$\tilde{\pmb{X}}_a^{(i)}, \tilde{\pmb{X}}_{r_1}^{(i)}, \tilde{\pmb{X}}_{r_2}^{(i)}, \tilde{\pmb{X}}_b^{(i)} \in \mathcal{R}_V(\Lambda)$$
 are the input signals in block i .

MLAN Conversion

Two Relays

See [Erez-Zamir '08].

Blockwise modulo-lattice additive noise channel:

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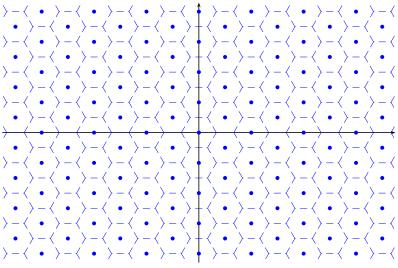
where

- $\tilde{X}_a^{(i)}, \tilde{X}_{r_1}^{(i)}, \tilde{X}_{r_2}^{(i)}, \tilde{X}_b^{(i)} \in \mathcal{R}_V(\Lambda)$ are the input signals in block i.
- Effective noise $\tilde{Z}_a^{(i)}$, $\tilde{Z}_{r_1}^{(i)}$, $\tilde{Z}_{r_2}^{(i)}$, $\tilde{Z}_b^{(i)}$ is statistically independent of the inputs and has reduced effective noise power.

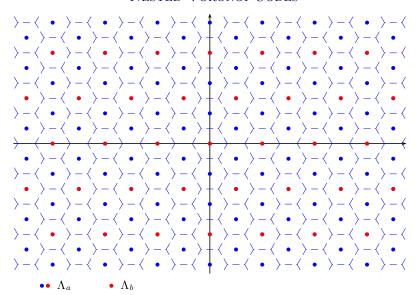
NESTED VORONOI CODES



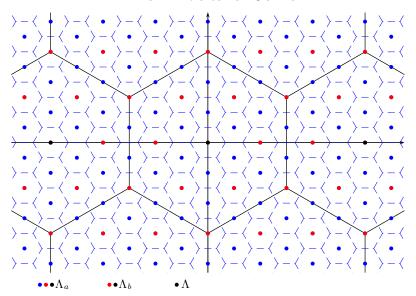
Nested Voronoi Codes



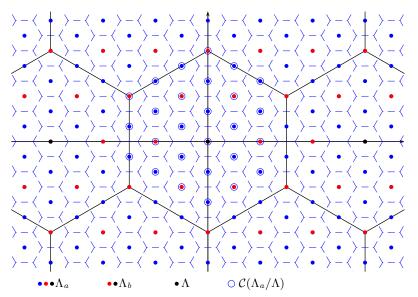
NESTED VORONOI CODES



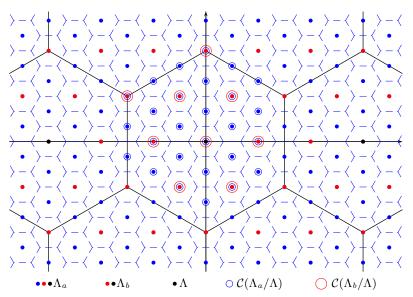
NESTED VORONOI CODES

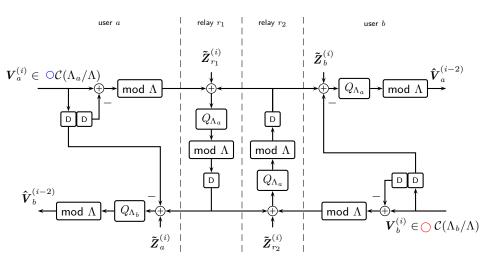


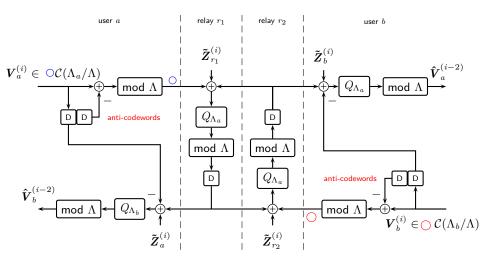
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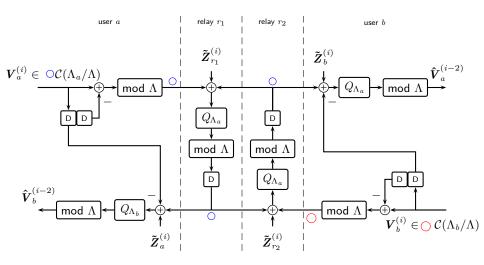


Nested Voronoi Codes

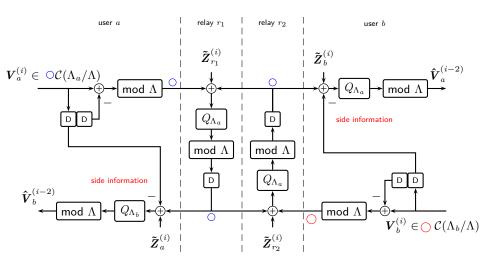


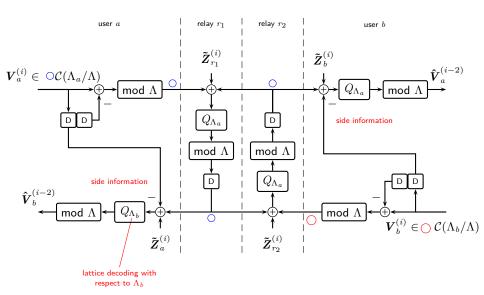


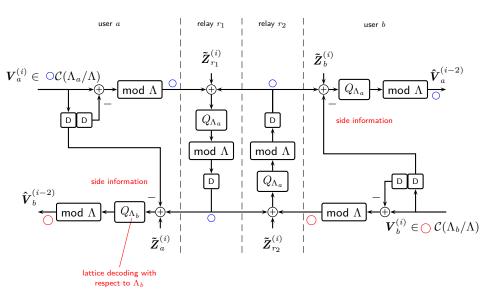




relays "protect" linear combinations of codewords







FINAL REMARKS

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• Superposition coding for unequal power constraints does not lead to a gap with respect to the upper bound in general.

Final Remarks

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- Different shaping lattices might provide a solution.

FINAL REMARKS

- Superposition coding for unequal power constraints does not lead to a gap with respect to the upper bound in general.
- Different shaping lattices might provide a solution.
- Generalization to L relays (assuming full separation) is possible.

CONCLUSION

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 Structured codes appear to be a powerful tool to show achievability of rates in communication networks.

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- Structured codes appear to be a powerful tool to show achievability of rates in communication networks.
- Involved principles: physical-layer network coding, broadcasting, using side information
- In certain networks the interference of users can be harnessed.
- However, a strategy to show full achievability of the upper bound is still not available.

References



B. Rankov and A. Wittneben.

Spectral efficient signaling for half-duplex relay channels.

2005, Proc. of Asilomar Conference on Signals, Systems and Computers



R. Knopp.

Two-way radio networks with a star topology.

2006 International Zurich Seminar on Communications



U. Erez and R. Zamir.

A Modulo-Lattice Transformation for Multiple-Access Channels 2008, IEEE Convention of Electrical and Electronics Engineers in Israel



W. Nam and S.-Y. Chung and Y. Lee.

Capacity of the Gaussian Two-Way Relay Channel to within 1/2 bit 2010. IEEE Transactions on Information Theory

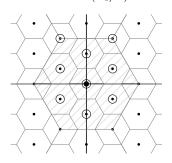
DOWNLINK

	1	 $2^{n(R_a - R_b)}$
1	$X_r(1,1)$	 $\boldsymbol{X}_r(1,M)$
2	$egin{aligned} oldsymbol{X}_r(1,1) \ oldsymbol{X}_r(2,1) \end{aligned}$	 $\boldsymbol{X}_r(2,M)$
:	:	:
2^{nR_b}	$X_r(N,1)$	 $\boldsymbol{X}_r(N,M)$

Downlink

DOWNLINK

$$oldsymbol{V} \in \mathcal{C}(\Lambda_c/\Lambda)$$

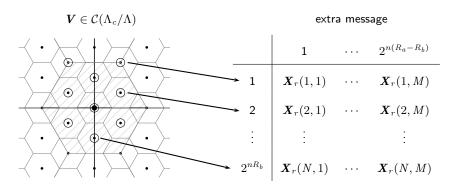


	I		
	1	• • •	$2^{n(R_a - R_b)}$
1	$X_r(1,1)$		$\boldsymbol{X}_r(1,M)$
2	$X_r(2,1)$		$\boldsymbol{X}_r(2,M)$
:	:		÷
2^{nR_b}	$X_r(N,1)$		$\boldsymbol{X}_r(N,M)$

Conclusion 00000

Downlink

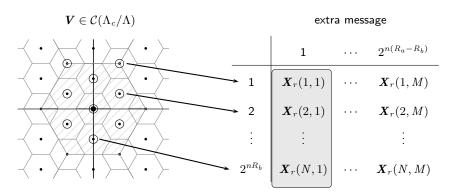
DOWNLINK



Downlink

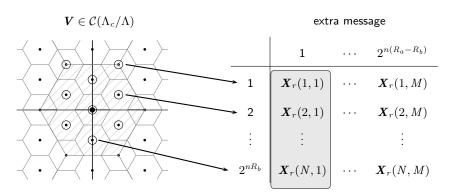
- Random codebook: $N=2^{nR_b}$ rows, $M=2^{n(R_a-R_b)}$ columns
- User b decodes with respect to the whole codebook.

DOWNLINK



- Random codebook: $N=2^{nR_b}$ rows, $M=2^{n(R_a-R_b)}$ columns
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- User a decodes with respect to a column of the codebook.

Downlink



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- User b decodes with respect to the whole codebook.
- User a decodes with respect to a column of the codebook.
- Optimal downlink strategy.

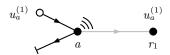
Another Toy Problem

- Noiseless, finite field physical layer and half-duplex nodes
- Users exchange bits (packets) $u_a^{(1)}, u_a^{(2)}, \ldots$ and $u_b^{(1)}, u_b^{(2)}, \ldots$



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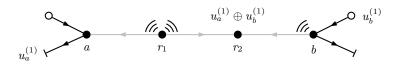






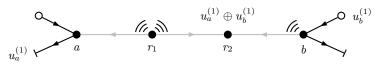
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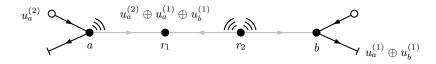
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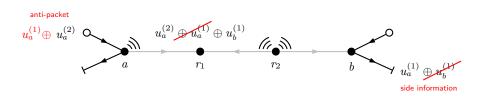
can be discarded

Another Toy Problem

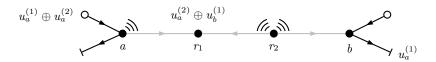
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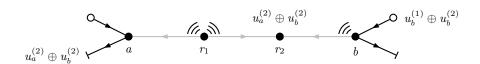
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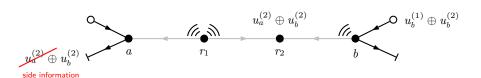
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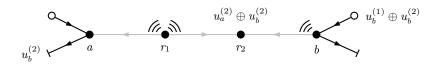
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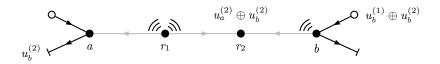
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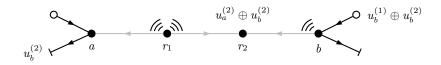


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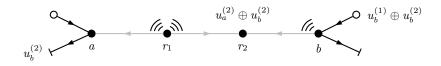
• Steady state

- Noiseless, finite field physical layer and half-duplex nodes
- Users exchange bits (packets) $u_a^{(1)}, u_a^{(2)}, \ldots$ and $u_b^{(1)}, u_b^{(2)}, \ldots$



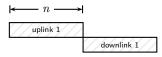
- Steady state
- Both users send a new packet or receive a packet in each block

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- Steady state
- Both users send a new packet or receive a packet in each block
- Sum rate approaches 2 bits per 2 transmissions

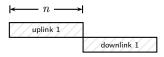
FULL-DUPLEX BLOCK TRANSMISSION



Two phases:

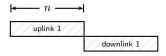
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FULL-DUPLEX BLOCK TRANSMISSION



Two phases:

1. Uplink: users a and b transmit to the relay r.



Two phases:

- 1. Uplink: users a and b transmit to the relay r.
- 2. Downlink: relay r broadcasts to the users a and b.



uplink 1	uplink 2	uplink 3
	downlink 1	downlink 2

uplink M	
downlink $M-1$	downlink M

Two phases:

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- 2. Downlink: relay r broadcasts to the users a and b.



uplink 1	uplink 2	uplink 3
	downlink 1	downlink 2

uplink Mdownlink M-1 downlink M

Two phases:

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Choose number of blocks M (uplink-downlink pairs) large, such that

$$\frac{MnR_a}{(M+1)n} \approx R_a, \qquad \frac{MnR_b}{(M+1)n} \approx R_b.$$

 $\longleftarrow n \longrightarrow$

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What rate pairs (R_a,R_b) (in bits per channel use) are achievable with $p_e \to 0$ as $n \to \infty$?