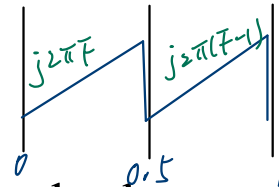


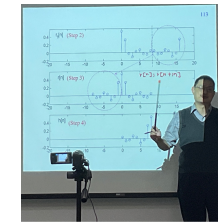
R11945072 3/6/23

## Homework 2 (Due: 4/12)



$$H(F) = H(F+1)$$

(page 111-114)



- (1) Write a Matlab or Python code that uses the frequency sampling method to design a (2k+1)-point discrete differentiation filter  $H(F) = j2\pi F$  when  $-0.5 < F < 0.5$  ( $k$  is an input parameter and can be any integer). (25 scores)

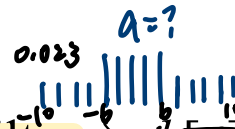
The transition band is assigned to reduce the error (unnecessary to optimize). (i) The impulse response and (ii) the imaginary part of the frequency response (DTFT of  $r[n]$ , see pages 113 and 114) of the designed filter should be shown. The code should be handed out by [NTU Cool](#).

沖激響應

虛部誤差

- (2) Can the techniques of the weight function and the transition band be applied in the FIR filter designed by (a) the MSE method and (b) the frequency sampling method? Why? (10 scores)

拾遺補

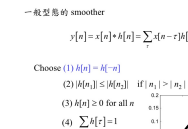
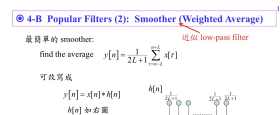


- (3) Suppose that the smooth filter is  $h[n] = a$  for  $|n| \leq 5$ ,  $h[n] = 0.023$  for  $6 \leq |n| \leq 10$ , and  $h[n] = 0$  otherwise. (a) What is the value of  $a$ ? (b) What is the efficient way to implement the convolution  $y[n] = x[n] * h[n]$ ? (10 scores)

①  $11a + 10(0.023) = 1$

②

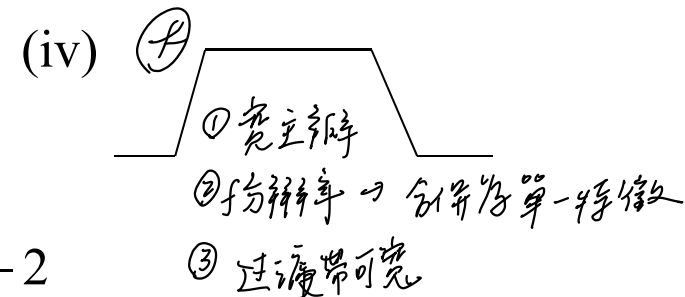
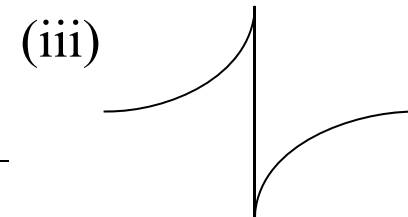
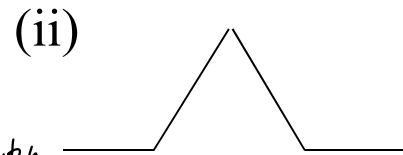
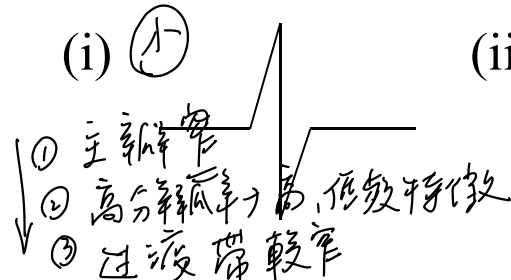
分成 sum-a  
sum-b



/sliding windows

可變變公式拆開單獨計算，避免重複捲積造成

- (4) The following figures are the impulse responses of some filters. Which one is a suitable smoother when we want to extract (a) small scaled features? (b) large scaled features? Also illustrate the reasons. (10 scores)



- (5) If the z-transform of  $h[n]$  is  $H(z) = \frac{2z^4 - 2z^3 + 3z^2 + z - 2}{z^2 + 0.2z - 0.24}$

(a) Determine the cepstrum of  $h[n]$ .

(Hint:  $z = 2^{-0.5}$  is one of the zeros of  $H(z)$ ) 表  $2z^{-1}$  为因式

(b) Convert the IIR filter into the minimum phase filter.

(20 scores)

- (6) Suppose that the cepstrum of a signal  $x[n]$  is

$$\hat{x}[2] = 0.7, \quad \hat{x}[n] = 0 \quad \text{otherwise}$$

Determine  $x[n]$  using the Z transform and  $\exp(\cdot)$ .

(10 scores)

Hilbert transform?

(2.)



R11945072 張栢彰

(1)

見檔案, m

(2)

也可以

① MSE: 最小化所需  $f$  響應和濾波器實際  $f$  之間的均方誤差設計 FIR, 此方法用所需  $f$  與標準計算濾波器係數

a. 加權函數: 可使訊號分佈不同寬帶比重, 不太關鍵的寬帶可進行必要弱化

b. transition band: 可使過渡有更陡峭/平滑的過渡

Sampling: 由 DFT 計算其係數所需之頻率響應 (through 均勻間隔)

a. 加權函數: 可給予不同权重, 使其更逼近所需響應、做出逼近符合之 signal.

b. transition band: 也可以、可使其於過渡帶的特徵做更大取樣而影響其特徵

(3)

①

$$11a + 10(0.023) = 1$$

$$a = 0.07 \text{ \#}$$


②

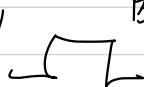
分成  $\text{sum}_a, \text{sum}_b$ , 主要通過使用兩個滑動窗口分別處理  $h[n]$  的不同值, 最終獲得卷積  $y[n]$ .

→ 如此可有多次計算卷積,

∴ 不需對所有  $h[n]$  值做單獨的卷積運算

(4)

(a) 

(b) 

因其具有較窄的主頻響應，高低頻分辨率窄，過渡帶寬 (可分為單一 feature)

(5)

過渡帶較窄

If the z-transform of  $h[n]$  is  $H(z) = \frac{2z^4 - 2z^3 + 3z^2 + z - 2}{z^2 + 0.2z - 0.24}$

(a) Determine the cepstrum of  $h[n]$ .

(Hint:  $z = 2^{-0.5}$  is one of the zeros of  $H(z)$ ) 表  $2z^{-1}$  為因式

(b) Convert the IIR filter into the minimum phase filter.

$$z^2 + \frac{1}{5}z - \frac{6}{5}z$$

(a)

$$H(z) = 2 \frac{(z + z^{-0.5})(z - z^{-0.5})(z - \frac{1+\sqrt{3}i}{2})(z - \frac{1-\sqrt{3}i}{2})}{(z + 0.6)(z - 0.4)} \times \left( \left( \frac{1+\sqrt{3}i}{2} \right) \times \left( \frac{1-\sqrt{3}i}{2} \right) \right) \times \left( \frac{1+1}{4} = 2 \right)$$

$$= 4 \frac{(z + z^{-0.5})(z - z^{-0.5})(z - \frac{1+\sqrt{3}i}{2})(z - \frac{1-\sqrt{3}i}{2})}{(z + 0.6)(z - 0.4)} \quad (1)$$

$$= 4 \frac{(z^2 - \frac{1}{2})(z^2 - \frac{1-\sqrt{3}i}{2}z - \frac{1+\sqrt{3}i}{2} + \frac{8}{4})}{z^2(1 + 0.6z^{-1})(1 - 0.4z^{-1})} = \frac{4(1 - z^{-2})(z^2 + 2)}{(1 + 0.6z^{-1})(1 - 0.4z^{-1})}$$

$$\hat{H}(z) = \log(H(z))$$

$$= \log(4) + \log(1 - z^{-0.5}z^{-1}) + \log(1 + z^{-0.5}z^{-1}) + \log(1 - \frac{z}{1+\sqrt{3}i}) + \log(1 - \frac{z}{1-\sqrt{3}i}) - \log(1 + 0.6z^{-1}) - \log(1 - 0.4z^{-1})$$

$$h[n] = \begin{cases} \log 4, & n=0 \\ \frac{-z^{-0.5n}}{n} + \frac{(z^{-0.5})^n}{n} + \frac{(0.6)^n}{n} + \frac{(0.4)^n}{n}, & n>0 \\ \frac{(z^{-0.5})^{-n}}{n} + \frac{(z^{-0.5})^n}{n}, & n<0 \end{cases}$$

(b) 若零點幅度大於單位圓 1，則換成  $\frac{1}{\text{conj}(z)}$

∴ 原式又有  $(z - \frac{1+\sqrt{3}i}{2})(z - \frac{1-\sqrt{3}i}{2})$  之零點過單位圓 (若過單位圓=1者，便其換成  $\frac{1}{\text{conj}(z)}$ )

$$4 \frac{(z + z^{-0.5})(z - z^{-0.5})(z - \frac{1+\sqrt{3}i}{2})(z - \frac{1-\sqrt{3}i}{2})}{(z + 0.6)(z - 0.4)} \times \frac{1}{\frac{1+\sqrt{3}i}{2}} \times \frac{1}{\frac{1-\sqrt{3}i}{2}} = \frac{\frac{1}{2}z^4 - \frac{1}{2}z^3 + \frac{3}{4}z^2 + \frac{1}{4}z - \frac{1}{2}}{z^2 + 0.2z + 0.24}$$

(6) 1. 計算 cepstrum:

$$c[n] = x[n] = \{0, 0, 0.7, 0, 0, \dots\}$$

2. Magnitude and phase spectra:

$$|X[k]| = \exp(\angle X[k]) = \{1, 1, \exp(0.7), 1, 1, \dots\}$$

$$\text{Phase spectrum: } \arg X[k] = 0$$

3. Z-transform:

$$X(z) = \sum_{k=-\infty}^{\infty} x[k] z^{-k} \exp(j \times \arg X[k]) \times z^{-k}$$

'i magnitude of spectrum is 1 all value ( $k=2$  除外),  $\angle$  phase spectrum = 0

$$\therefore X(z) = 1 + \exp(0.7) \times z^{-2}$$

4. inverse Z-transform.

$$\Rightarrow X[n] = Z^{-1}\{X(z)\} = Z^{-1}\{1 + \exp(0.7) \times z^{-2}\}$$

$$\Rightarrow \underline{X[n] = \delta[n] + \exp(0.7) \times \delta[n-2]} \quad \#$$

'i signal  $x[n]$  有值 1 於  $n=0$  時, 值  $\exp(0.7) \approx 2.01$  於  $n=2$ , 其他值 = 0

(7)

(a) 不是全通滤波器: poles & zeros 皆位於圓的內部, 有助於平滑振幅並應用於實際應用

② Casality (因果關係): 最小相位滤波器因果關係, 其 output 只取決於當下、過去的值, 可允許更快 impulse response 並大幅減少延遲

(b)

解析信號生成: 信號為複值, 實部為原始, 虛部為 Hilbert Transform

$\Rightarrow$  可用於通信中幅度、相位解調, 包絡檢測等特別有用。

瞬時頻率估計: 透過計算並解析信號相位之時間導數達成, 可分析具有呼吸頻率特性之語音、音樂信號

(c)

1. 模型簡單：將 time domain 的卷積運算轉換為加法運算來簡化問題。

→ 識別、分離所需信號、分集更容易

2. 噪聲的穩定性：比均衡技術處理噪聲更穩定。可有效分析所需信號，多位分量 → 提高信號質量，降低誤碼率。

(8)

Extra: ② Kalman, Particle filter 在信號處理用來做？

Ans: 二者皆是用於估計系統狀態的遞推算法，常用於導航、目標追蹤等領域

(1) Kalman: 線性濾波：主要用於估計線性高斯系統。結合了系統模型的先驗知識和觀測數據來估計系統狀態

特性：計算效率高(線性)，但對非線性或非高斯系統，有性能限制。

(2) Particle: 蒙特卡洛方法之遞推濾波 → 非線性，非高斯系統狀態。

使用一組粒子(樣本)來近似系統後驗分布，以觀測數據做粒子权重更新、取樣

特性：可calculate 非線性，非高斯 → but 計算效率低 (高維更是！)