

Introduction to RF Module



Introduction to the RF Module

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Contents

Introduction
The Use of the RF Module
The RF Module Interfaces
The Physics List by Space Dimension and Study Type
The Model Library
Tutorial Example: Impedance Matching of a Lossy Ferrite
3-port Circulator
Introduction
Impedance Matching
Model Definition
The Lossy Ferrite Material Model
References

Introduction

The RF Module is used by engineers and scientists to understand, predict, and design electromagnetic wave propagation and resonance effects in high-frequency applications. Simulations of this kind result in more powerful and efficient products and engineering methods. It allows its users to quickly and accurately predict electromagnetic field distributions, transmission, reflection, and power dissipation in a proposed design. Compared to traditional prototyping, it offers the benefits of lower cost and the ability to evaluate and predict entities that are not directly measurable in experiments. It also allows the exploration of operating conditions that would destroy a real prototype or be hazardous.

This module covers electromagnetic fields and waves in two-dimensional and three-dimensional spaces along with traditional circuit-based modeling of passive and active devices. All modeling formulations are based on Maxwell's equations or subsets and special cases of these together with material laws for propagation in various media. The modeling capabilities are accessed via predefined physics user interfaces, referred to as RF interfaces, which allow you to set up and solve electromagnetic models. The RF interfaces cover the modeling of electromagnetic fields and waves in frequency domain, time domain, eigenfrequency, and mode analysis.

Under the hood, the RF interfaces formulate and solve the differential form of Maxwell's equations together with the initial and boundary conditions. The equations are solved using the finite element method with numerically stable edge element discretization in combination with state-of-the-art algorithms for preconditioning and solution of the resulting sparse equation systems. The results are presented using predefined plots of electric and magnetic fields, S-parameters, power flow, and dissipation. You can also display your results as plots of expressions of the physical quantities that you define freely, or as tabulated derived values obtained from the simulation.

The work flow is straightforward and can be described by the following steps: define the geometry, select materials, select a suitable RF interface, define boundary and initial conditions, define the finite element mesh, select a solver, and visualize the results. All these steps are accessed from the COMSOL Desktop. The solver selection step is usually carried out automatically using default settings, which are tuned for each specific RF interface.

The RF Module's Model Library describes the interfaces and their different features through tutorial and benchmark examples for the different formulations. The library includes models addressing antennas, ferrite devices, microwave heating phenomena, passive devices, scattering and RCS analysis, transmission lines and waveguides in RF and microwave engineering, tutorial models for

education, and benchmark models for verification and validation of the RF interfaces

This introduction is intended to give you a jump start in your modeling work. It has examples of the typical use of the RF Module, a list of the interfaces with a short description, and a tutorial example that introduces the modeling workflow.

The Use of the RF Module

The RF interfaces are used to model electromagnetic fields and waves in high frequency applications. The latter means that it covers the modeling of devices that are above about 0.1 electromagnetic wavelength in size. Thus, it may be used to model microscale devices or human size devices operating at frequencies above 10 MHz.

RF simulations are frequently used to extract S-parameters characterizing the transmission and reflection of a device. Figure 1 shows the electric field distribution in a dielectric loaded H-bend waveguide component. A rectangular TE_{10} waveguide mode is launched into the inport at the near end of the device and absorbed by the outport at the far end. The bend region is filled by silica glass.

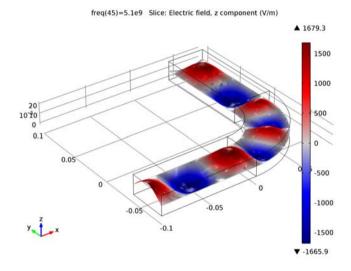


Figure 1: Electric field distribution in a dielectric loaded H-bend waveguide. From the RF Module Model Library model H-Bend Waveguide 3D.

The transmission and reflection of the device is obtained in the form of S-parameters, see Figure 2.

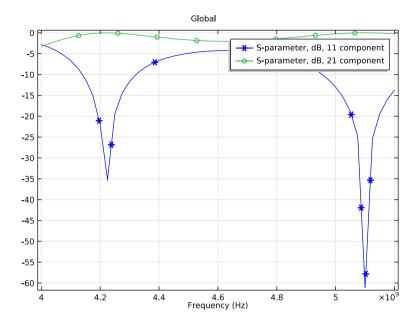


Figure 2: The S-parameters, on a dB scale, as a function of the frequency.

S-parameters can be exported in the Touchstone file format for further use in system simulations and are, by themselves, useful performance measures of a design.

In Figure 3 and Figure 4, a model from the RF Module Model Library shows how a human head absorbs a radiated wave from an antenna held next to the ear. The temperature is increased by the absorbed radiation.

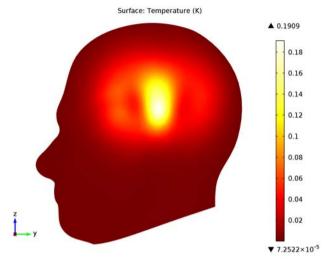


Figure 3: The local increase in temperature in a human head due to absorption of electromagnetic energy from an antenna held next to the ear. From the RF Module Model Library model Absorbed Radiation (SAR) in the Human Brain.

The SAR (specific absorption rate) value is of specific interest to designers of mobile telephones and is readily obtained from the simulation.

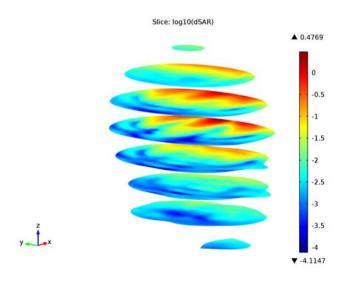


Figure 4: Log-scale slice plot of the local specific absorption rate (SAR) in a human head.

The RF Module also offers a comprehensive set of features for 2D modeling including both source driven wave propagation and mode analysis. Figure 5 shows mode analysis of a step-index profile optical fiber.

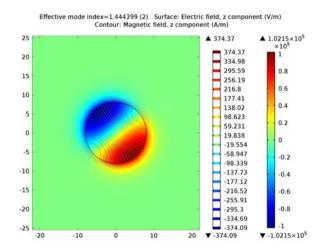


Figure 5: The surface plot visualizes the longitudinal component of the electric field in the fiber core. From the RF Module Model Library model Step Index Fiber.

Both in 2D and 3D, the analysis of periodic structures is popular. Figure 6 is an example of a plane wave incidence on a wire grating with a dielectric substrate.

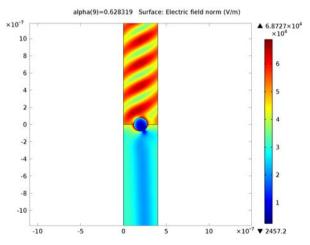


Figure 6: Electric field norm for TE incidence at π /5. From the RF Module Model Library model Plasmonic Wire Grating.

It is also possible to perform Body-of-Revolution (BOR) simulations in 2D axisymmetry. Figure 7 illustrates the modeling of a monoconical antenna with coaxial feed.

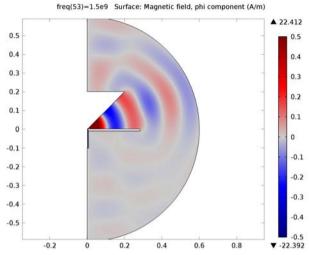


Figure 7: Axisymmetric model of a monoconical antenna with coaxial feed. The azimuthal component of the magnetic field is shown. From the RF Module Model Library model Conical Antenna.

The RF Module has a vast range of tools to evaluate and export the results, for example, evaluation of far-field, feed impedance, and scattering matrices (S-parameters). S-parameters can be exported in the Touchstone file format.

The full-wave electromagnetic field modeling can also be combined with circuit-based modeling. This is an ideal basis for design, exploration, and optimization. More complex system models can be exploited using circuit-based modeling while maintaining links to full field models for key devices in the circuit allows for design innovation and optimization on both levels.

The RF Module Interfaces

The RF interfaces are based upon Maxwell's equations or subsets and special cases of these together with material laws. In the module, these laws of physics are translated by the RF interfaces to sets of partial differential equations with corresponding initial and boundary conditions.

The RF physics define a number of features. Each feature represents a term or condition in the underlying Maxwell-based formulation and may be defined in a geometric entity of the model, such as a domain, boundary, edge (for 3D models), or point.

Figure 8 uses the Coaxial Waveguide Coupling model from the RF Module Model Library to show the Model Builder window and the settings window for the selected Wave Equation, Electric 1 feature node. The Wave Equation, Electric 1 node adds the terms representing Electromagnetic Waves to the model equations in a selected geometrical domain in the model.

Furthermore, the Wave Equation, Electric 1 feature node may link to the Materials feature node to obtain physical properties such as relative permittivity in this case the relative permittivity of a user-defined dielectric. The properties, defined by the Dielectric material, can be functions of the modeled physical quantities, such as temperature. In the same fashion, the Perfect Electric

Conductor 1 boundary condition feature adds the reflecting boundary conditions that limit the Electromagnetic Waves domain.

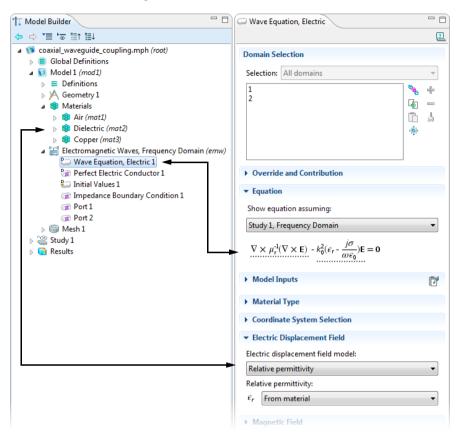


Figure 8: The Model Builder (left), and the Wave Equation, Electric settings window (right). The Equation section shows the model equations and the terms added by the Wave Equation, Electric 1 node to the model equations. The added terms are underlined with a dotted line. The text also explains the link between the Dielectric node and the values for the relative permittivity.

Figure 9 shows the RF interfaces as displayed in the Model Wizard for this module.

```
Radio Frequency
Electromagnetic Waves, Frequency Domain (emw)
Electromagnetic Waves, Time Explicit (ewte)
Electromagnetic Waves, Transient (temw)
Transmission Line (tl)
```

Figure 9: The RF Module physics as displayed in the Model Wizard.

This module includes RF interfaces () for frequency-domain modeling and time-domain modeling, respectively. It also has the Microwave Heating interface found under Heat Transfer. Also see "The Physics List by Space Dimension and Study Type" on page 11. A brief overview of the RF interfaces follows.

ELECTROMAGNETIC WAVES, FREQUENCY DOMAIN

The Electromagnetic Waves, Frequency Domain interface (solves a frequency-domain wave equation for the electric field. The sources can be in the form of point dipoles, line currents, or incident fields on boundaries or domains. It is used primarily to model electromagnetic wave propagation in different media and structures. Variants of the formulation solves an eigenvalue problem to find the eigenfrequencies of a structure or, at a prescribed frequency, solves an eigenvalue problem to find the propagating modes in waveguides and transmission lines. Some typical applications that are simulated with the interface are waveguides and transmission lines, filters and resonators, antennas, and RF connectors and couplers.

ELECTROMAGNETIC WAVES, TIME EXPLICIT

The Electromagnetic Waves, Time Explicit interface () solves a system of two first-order partial differential equations (Faraday's law and Maxwell-Ampère's law) for the electric and magnetic fields using the Time Explicit Discontinuous Galerkin method. The sources can be in the form of volumetric electric or magnetic currents or electric surface currents or fields on boundaries. It is used primarily to model electromagnetic wave propagation in linear media. Typical applications involve the transient propagation of electromagnetic pulses.

ELECTROMAGNETIC WAVES, TRANSIENT

The Electromagnetic Waves, Transient interface () solves a time-domain wave equation for the electric field. The sources can be in the form of point dipoles, line currents, or incident fields on boundaries or domains. It is used primarily to model electromagnetic wave propagation in different media and structures when a time-domain solution is required—for example, for non-sinusoidal waveforms or for nonlinear media. Typical applications involve the propagation of electromagnetic pulses and the generation of harmonics in nonlinear optical media.

Transmission Line

The Transmission Line interface (1947), solves the time-harmonic transmission line equation for the electric potential. The interface is used when solving for electromagnetic wave propagation along one-dimensional transmission lines and

is available in 1D, 2D and 3D. The Eigenfrequency and Frequency Domain study types are available. The frequency domain study is used for source-driven simulations for a single frequency or a sequence of frequencies. Typical applications involve the design of impedance matching elements and networks.

MICROWAVE HEATING

The Microwave Heating interface (combines the features of an Electromagnetic Waves, Frequency Domain interface with those of the Heat Transfer interface. The predefined interaction adds the electromagnetic losses from the electromagnetic waves as a heat source. This interface is based on the assumption that the electromagnetic cycle time is short compared to the thermal time scale (adiabatic assumption).

ELECTRICAL CIRCUITS

The Electrical Circuit interface (🔀) can be connected to an RF interface. The lumped voltage and current variables from the circuits are translated into boundary conditions applied to the distributed field model. Typical applications include the modeling of transmission lines and antenna feeding.

The Physics List by Space Dimension and Study Type

The table below list the interfaces available specifically with this module in addition to the COMSOL Multiphysics basic license.

PHYSICS	ICON	TAG	SPACE DIMENSION	PRESET STUDY TYPE	
₹ AC/DC					
Electrical Circuit	<u>\$0</u>	cir	Not space dependent	stationary; frequency domain; time dependent	
Heat Transfer					
Electromagnetic Heating					
Microwave Heating	(ME	mh	3D, 2D, 2D axisymmetric	stationary; frequency domain; time dependent; boundary mode analysis; frequency-stationary; frequency transient	
Radio Frequency					
Electromagnetic Waves, Frequency Domain	(p-1-1)	emw	3D, 2D, 2D axisymmetric	eigenfrequency; frequency domain; frequency-domain modal; boundary mode analysis	
Electromagnetic Waves, Time Explicit	ir-id M©	ewte	3D, 2D, 2D axisymmetric	time dependent	
Electromagnetic Waves, Transient	₩ #	temw	3D, 2D, 2D axisymmetric	eigenfrequency; time dependent; time dependent modal	
Transmission Line	A B	tl	3D, 2D, 1D	eigenfrequency; frequency domain	

The Model Library

To open any RF Module Model Library model, select View > Model Library III from the main menu in COMSOL Multiphysics. In the Model Library window that opens, expand the RF Module folder and browse or search the contents. Click Open Model and PDF to open the model in COMSOL Multiphysics and a PDF to read background theory about the model including the step-by-step instructions to build it.

The MPH-files in the COMSOL model libraries can have two formats—Full MPH-files or Compact MPH-files.

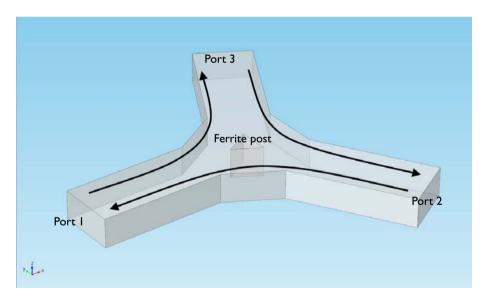
- Full MPH-files, including all meshes and solutions. In the Model Library these models appear with the icon. If the MPH-file's size exceeds 25MB, a tip with the text "Large file" and the file size appears when you position the cursor at the model's node in the Model Library tree.
- Compact MPH-files with all settings for the model but without built meshes and solution data to save space on the DVD (a few MPH-files have no solutions for other reasons). You can open these models to study the settings and to mesh and re-solve the models. It is also possible to download the full versions—with meshes and solutions—of most of these models through Model Library Update. In the Model Library these models appear with the 🛓 icon. If you position the cursor at a compact model in the Model Library window, a No solutions stored message appears. If a full MPH-file is available for download, the corresponding node's context menu includes a Model Library Update item.

A model from the Model Library is used as a tutorial in this guide. See "Tutorial Example: Impedance Matching of a Lossy Ferrite 3-port Circulator", which starts on the next page.

Tutorial Example: Impedance Matching of a Lossy Ferrite 3-port Circulator

Introduction

A microwave circulator is a nonreciprocal multiport device. It has the property that a wave incident on port 1 is routed into port 3 but a wave incident on port 3 is not routed back into port 1 but is instead routed into port 2, and so on. This property of circulators is used to isolate microwave components from each other. A typical example is when connecting a transmitter and a receiver to a common antenna. By connecting the transmitter, receiver, and antenna to different ports of a circulator, the transmitted power is routed to the antenna whereas any power received by the antenna goes into the receiver. Circulators typically rely on the use of ferrites, a special type of highly permeable and low-loss magnetic material that is anisotropic for a small RF signal when biased by a much larger static magnetic field. In the example, a three-port circulator is constructed from three rectangular waveguide sections joining at 120° and with a ferrite post inserted at the center of the joint.



The post is magnetized by a static H_0 bias field along its axis. The bias field is supplied by external permanent magnets which are not explicitly modeled in this tutorial.

Impedance Matching

An important step in the design of any microwave device is to match its input impedance for a given operating frequency. Impedance matching is equivalent to minimizing the reflections at the inport. The parameters that need to be determined are the size of the ferrite post and the width of the wider waveguide section surrounding the ferrite. In this tutorial, these are varied in order to minimize the reflectance. The scattering parameters (S-parameters) used as measures of the reflectance and transmittance of the circulator are automatically computed.

The nominal frequency for the design of the device is selected to 3 GHz. The circulator can be expected to perform reasonably well in a narrow frequency band around 3 GHz so a frequency range of 2.6 – 3.4 GHz is studied. It is desired that the device operates in single mode. Thus a rectangular waveguide cross section of 6.67 cm by 3.33 cm is selected. This sets the cut-off frequency for the fundamental TE_{10} mode to 2.25 GHz. The cut-off frequencies for the two nearest higher modes, the TE₂₀ and TE₀₁ modes, are both at 4.5 GHz, leaving a reasonable safety margin.

Model Definition

One of the rectangular ports is excited by the fundamental TE₁₀ mode. At the ports, the boundaries are transparent to the TE_{10} mode. The following equation applies to the electric field vector **E** inside the circulator:

$$\nabla \times (\boldsymbol{\mu}_r^{-1} \nabla \times \boldsymbol{E}) \! - \! k_0^2 \! \left(\boldsymbol{\epsilon}_r - \frac{j\boldsymbol{\sigma}}{\omega \epsilon_0} \! \right) \! \boldsymbol{E} \; = \; 0$$

where μ_r denotes the relative permeability tensor, ω the angular frequency, σ the conductivity tensor, ε_0 the permittivity of vacuum, ε_r the relative permittivity tensor, and k_0 is the free space wave number. The conductivity is zero everywhere. Losses in the ferrite are introduced as complex-valued permittivity and permeability tensors. The magnetic permeability is of key importance as it is the anisotropy of this parameter that is responsible for the nonreciprocal behavior of the circulator. For simplicity, the rather complicated material expressions are predefined in a text file that is imported into the model. As a reference, the expressions are also included in the next section.

The Lossy Ferrite Material Model

Complete treatises on the theory of magnetic properties of ferrites can be found in Ref. 1 and Ref. 2. The model assumes that the static magnetic bias field, H_0 , is much stronger than the alternating magnetic field of the microwaves, so the quoted expressions are a linearization for a small-signal analysis around this

operating point. Under these assumptions, and including losses, the anisotropic permeability of a ferrite magnetized in the positive *z* direction is given by:

$$[\boldsymbol{\mu}] = \begin{bmatrix} \mu & j\kappa & 0 \\ -j\kappa & \mu & 0 \\ 0 & 0 & \mu_0 \end{bmatrix}$$

where

$$\kappa = -j\mu_0 \chi_{xy}$$

$$\mu = \mu_0(1 + \chi_{xx})$$

and the unique elements of the magnetic susceptibility tensor χ are given by:

$$\chi_{xx} = \frac{\omega_0 \omega_{\rm m} (\omega_0^2 - \omega^2) + \omega_0 \omega_{\rm m} \omega^2 \alpha^2}{(\omega_0^2 - \omega^2 (1 + \alpha^2))^2 + 4\omega_0^2 \omega^2 \alpha^2} - j \frac{\alpha \omega \omega_{\rm m} (\omega_0^2 + \omega^2 (1 + \alpha^2))}{(\omega_0^2 - \omega^2 (1 + \alpha^2))^2 + 4\omega_0^2 \omega^2 \alpha^2}$$

$$\chi_{xy} = \frac{2\omega_{0}\omega_{m}\omega^{2}\alpha}{\left(\omega_{0}^{2} - \omega^{2}(1 + \alpha^{2})\right)^{2} + 4\omega_{0}^{2}\omega^{2}\alpha^{2}} + j\frac{\omega\omega_{m}(\omega_{0}^{2} - \omega^{2}(1 + \alpha^{2}))}{\left(\omega_{0}^{2} - \omega^{2}(1 + \alpha^{2})\right)^{2} + 4\omega_{0}^{2}\omega^{2}\alpha^{2}}$$

where

$$\omega_0 = \mu_0 \gamma H_0$$

$$\omega_{\rm m} = \mu_0 \gamma M_s$$

$$\alpha = \frac{\mu_0 \gamma \Delta H}{2\omega}$$

Here μ_0 denotes the permeability of free space; ω is the angular frequency of the microwave field; ω_0 is the precession resonance frequency (Larmor frequency) of a spinning electron in the applied magnetic bias field, H_0 ; ω_m is the electron Larmor frequency at the saturation magnetization of the ferrite, M_s ; and γ is the gyromagnetic ratio of the electron. For a lossless ferrite ($\alpha = 0$), the permeability becomes infinite at $\omega = \omega_0$. In the lossy ferrite ($\alpha \neq 0$), this resonance becomes finite and is broadened. The loss factor, α , is related to the line width, ΔH , of the susceptibility curve near the resonance as given by the last expression above. The material data,

$$M_s = 5.41 \cdot 10^4 \text{ A/m}, \ \epsilon_r = 14.5$$

with an effective loss tangent of $2 \cdot 10^{-4}$ and $\Delta H = 3.18 \cdot 10^{3}$ A/m, are taken for aluminum garnet from Ref. 2. The applied bias field is set to $H_0 = 7.96 \cdot 10^3$ A/m. The electron gyromagnetic ratio taken from Ref. 2 is 1.759·10¹¹ C/kg.

References

- 1. R.E. Collin, Foundations for Microwave Engineering, 2nd ed., IEEE Press/ Wiley-Interscience, 2000.
- 2. D.M. Pozar, Microwave Engineering, 3rd ed., John Wiley & Sons Inc, 2004.

Model Wizard

These step-by-step instructions guide you through the design and modeling of the lossy three-port circulator in 3D. The first part involves the geometric design and impedance matching at a nominal frequency of 3 GHz. After that, a frequency sweep is performed to see how well it performs in a frequency band of 400 MHz centered at 3 GHz. Finally, computation and a Touchstone file export of the entire S-parameter matrix is performed.

- Open COMSOL Multiphysics. In the Model Wizard, the default Space Dimension is 3D. Click Next .
- 2 In the Add Physics tree under Radio Frequency, double-click Electromagnetic Waves, Frequency Domain 🔀 to add it to the Selected physics list. Click Next ⇒.
- 3 Under Studies>Preset Studies click Frequency Domain m.
- 4 Click Finish ...

Global Definitions - Parameters and Variables

The geometry is set up using a parameterized approach. This allows you to match the input impedance to that of the connecting waveguide sections by variation of two geometric design parameters. These are dimensionless numbers used to scale selected geometric building blocks.

Note: In this section, two parameters are entered and a set of variables imported from a file to prepare for drawing the circulator geometry, which is described in the section "Geometry Sequence" on page 40. Alternatively, a predefined Model Library file containing the geometry, parameters, and variables can be imported, as described in "Geometry" on page 18. If you import the geometry, you only need to review this section for information.

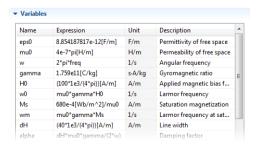
- In the Model Builder, right-click Global Definitions and select Parameters P. .
- 2 In the Parameters settings window under Parameters, enter these settings in the Parameters table.



The lossy ferrite material model is set up by referring to global variables. For convenience the definitions are stored in an external text file that is imported into the model. The external text file also contains comments.

Note: The location of the text files vary based on the installation. For example, if the installation is on your hard drive, the file path might be similar to C:\Program Files\COMSOL43a\models\.

- In the Model Builder, right-click Global Definitions and choose Variables
- ² Go to the Variables settings window. Under Variables click Load from File *▶*.
- 3 Browse to the Model Library folder \RF_Module\Ferrimagnetic_Devices and double-click the file lossy_circulator_3d_parameters.txt. The variables are imported into the table.



Geometry

In the Global Definitions section, you entered parameters and imported variables in preparation for drawing the geometry. To learn how to draw the circulator, go to "Geometry Sequence" on page 40.

To save time, a predefined model containing the parameters, variables, and geometry can instead be opened from the Model Library.

- From the View menu, select Model Library
- 2 In the Model Library, under RF Module>Ferrimagnetic Devices double-click lossy_circulator_3d_geom to open it.

Once the geometry is either drawn or imported, you can then experiment with different dimensions and update the values of sc_chamfer and sc_ferrite and re-run the geometry sequence.

Materials

The next step is to add material settings to the model. The air that fills most of the volume is available as a built-in material. The lossy ferrite has material assigned to it later, and illustrates how external material data can also be entered directly into the electromagnetic waves model. The walls of the waveguide sections are modeled as perfect conductors and do not require a material.

- From the main menu, select View>Material Browser
- 2 Go to the Material Browser window. In the Materials tree under Built-In, right-click Air and choose Add Material to Model 4.

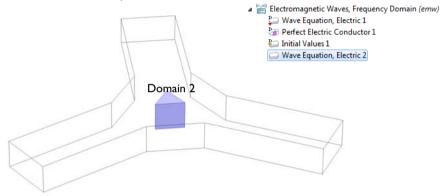


Electromagnetic Waves, Frequency Domain

The ferrite enters in the physics interface as a separate, user-defined equation model, referring to the global variables defined on page 16.

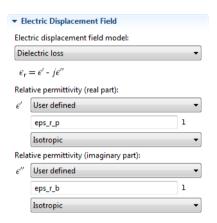
Wave Equation, Electric 2

- In the Model Builder right-click Electromagnetic Waves, Frequency Domain (emw) and and at the domain level choose Wave Equation, Electric . A Wave Equation, Electric 2 node is added to the Model Builder. The nodes with a 'D' in the upper left corner indicate a default node.
- 2 Select Domain 2 only.

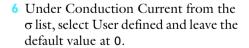


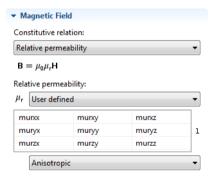
There are many ways to select geometric entities. When you know the domain to add, such as in this exercise, you can click the Paste Selection button and enter the information in the Selection field. In this example enter 2 in the Paste Selection window. For more information about selecting geometric entities in the Graphics window, see the *COMSOL Multiphysics Reference Manual*

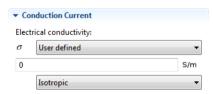
- 3 Go to the Wave Equation, Electric settings window. Under Electric Displacement Field:
 - From the Electric displacement field model list, select Dielectric loss.
 - From the ε' list, select User defined. In the associated field, enter eps_r_p.
 - From the ε'' list, select User defined. In the associated field, enter eps r b.



- 4 Under Magnetic Field from the μ_r list, select User defined and Anisotropic.
- 5 In the μ_r table, enter the settings as in the figure to the right.



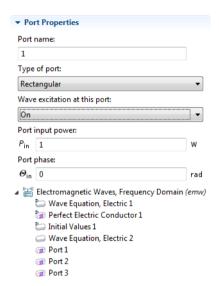




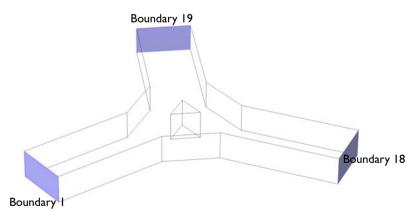
Now add ports for excitation and transmission.

Port 1, Port 2, and Port 3

- In the Model Builder, right-click Electromagnetic Waves, Frequency Domain (emw) 🕍 and at the boundary level choose Port 🝙.
- 2 Select Boundary 1 only.
- 3 Go to the Port settings window. Under Port Properties from the Type of port list, select Rectangular.
- 4 From the Wave excitation at this port list, select On.
- 5 Right-click Electromagnetic Waves, Frequency Domain (emw) and add another Port anode. For Port 2:
 - Select Boundary 18
 - Select Rectangular as the Type of port list
- 6 Add another Port ⋒ node. For Port 3:
 - Select Boundary 19
 - Select Rectangular as the Type of port list



The node sequence in the Model Builder should match the above figure.



Mesh

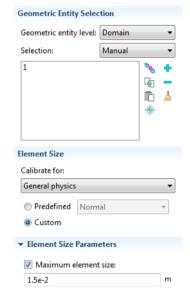
The mesh respects the geometry but in addition some physics related considerations are needed. The mesh needs to resolve the local wavelength and, for lossy domains, the skin depth. The skin depth in the ferrite is large compared to the size of the domain so the main concern is to resolve the local wavelength. This is done by providing maximum mesh sizes per domain. The rule of thumb is to use a maximum element size that is one fifth of the local wavelength (at the maximum frequency) or less.

Free Tetrahedral I

- In the Model Builder, right-click Mesh 1 ⊚ and choose Free Tetrahedral △.
- 2 Right-click Free Tetrahedral 1 △ and choose Size △.
- 3 Go to the Size settings window. Under Geometric Entity Selection from the Geometric entity level list, select Domain.
- 4 Select Domain 1 only. This is the domain filled with air inside the waveguide.
- 5 Under Element Size click the Custom button.
- 6 Under Element Size Parameters select the Maximum element size check box. Enter 1.5e-2 in the field.

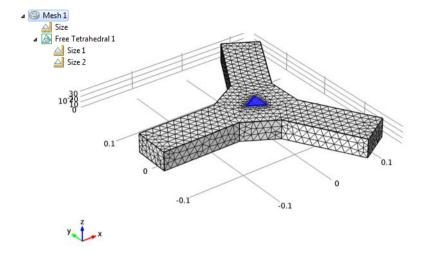
Size 2

- In the Model Builder, right-click Free Tetrahedral 1 ♠ and choose Size ຝ. A second Size node is added to the sequence.
- 2 Go to the Size settings window. Under Geometric Entity Selection from the Geometric entity level list, select Domain.



- 3 Select Domain 2 only. This is the ferrite post domain.
- 4 Under Element Size click the Custom button.
- 5 Under Element Size Parameters select the Maximum element size check box. Enter 4.5e-3 in the field.
- 6 In the Size settings window, click Build All

The node sequence in the Model Builder and the mesh should match the figure below.



Study I

The final step is to solve for the nominal frequency and inspect the results for possible modeling errors.

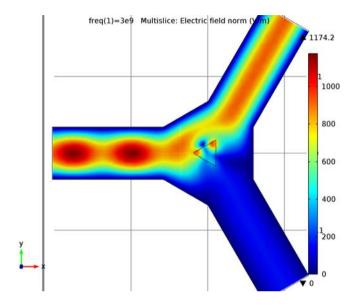
- In the Model Builder, expand the Study 1 node, then click Step 1: Frequency Domain w.
- 2 Go to the Frequency Domain settings window. Under Study Settings, in the Frequencies field, enter 3e9.
- 3 In the Model Builder, right-click Study 1 and choose Compute =.

Results

Electric Field

The default multislice plot shows the electric field norm. It is best viewed from above, so click the Go to XY View button on the Graphics toolbar.

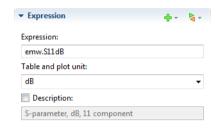
The electric field norm gives a good indication of where the main power is flowing and where there are standing waves due to reflections from the impedance mismatch at the center.



Definitions - Probes

The remaining work is to vary the two design parameters to minimize reflections at the nominal frequency. To do this, perform parametric sweeps over the design parameters (scale factors). To avoid accumulating a lot of data while solving, throw away the solutions and log only the S-parameter representing reflection in a table. For this purpose, add a Global Variable Probe to the model.

- In the Model Builder under Model 1, right-click Definitions ≡ and choose Probes>Global Variable Probe ▶.
- 2 Go to the Global Variable Probe settings window. In the upper-right corner of the Expression section, click Replace Expression § .
- 3 From the menu, choose Electromagnetic Waves, Frequency Domain>Ports> S-parameter, dB, 11 component (emw.S11dB)(or when you know the variable name, enter emw.S11dB in the Expression field).



Study I

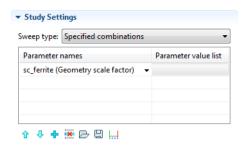
Modify the study to vary the scale factor determining the size of the ferrite post. The study type is still Frequency Domain.

Parametric Sweep

The parametric sweep over the scale factor is added as an extension to the frequency domain study.

- In the Model Builder, right-click Study 1 and choose Parametric Sweep #.
- 2 Go to the Parametric Sweep settings window. Under Study Settings click Add + under the Parameter names table.

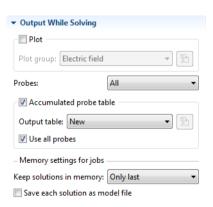
3 In the Parameter names list, select sc_ferrite (geometry scale factor).



- 4 Under Study Settings click the Range | button under the Parameter names table.
- 5 In the Range dialog box:
 - In the Start field, enter 0.5.
 - In the Step field, enter 0.003.
 - In the Stop field, enter 0.53.

Note: Or enter range (0.5,0.003,0.53) in the Parameter value list.

- 6 Click Replace.
- 7 Under Output While Solving, select the Accumulated probe table check box under Probes.
- 8 Choose Only last from the Keep solutions in memory list.
- **9** In the Model Builder, right-click Study 1 and choose Compute ■ (or press F8).



Results

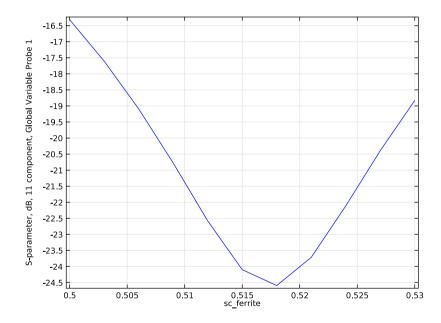
Probe I D Plot Group 2

The probe with the reflection coefficient versus the scale parameter is automatically logged to a table. A dedicated 1D plot group is also created but it plots the S-parameter versus frequency. To plot versus the geometry parameter, proceed as follows.

- In the Model Builder under Results, expand the Tables node.
- 2 Select Accumulated Probe Table 1 ■.
- 3 In the Table window, right-click the freq column and select Delete Column ▼.

Note: The Table window table is by default located in the lower-right side of the COMSOL Desktop. Or select View>Table ## from the main menu to open the window.

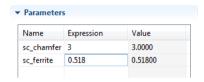
4 In the Table window, click the Table Graph [1].



The plot of the S-parameter indicates a minimum for a scale factor of 0.518. Freeze the parameter at this value and add a new study to vary the next scale factor.

Global Definitions - Parameters

- In the Model Builder under Global Definitions, click Parameters P₁.
- 2 In the Parameters settings window under Parameters, in the Expression column enter 0.518 in the sc_ferrite row.



Study I

Parametric Sweep

- In the Model Builder under Study 1, click Parametric Sweep #.
- **2** Go to the Parametric Sweep settings window. Under Study Settings in the Parameter names list, select sc_chamfer (Geometry scale factor).
- 3 Under Study Settings click the Range | button under the Parameter names table.
- 4 In the Range dialog box:
 - In the Start field, enter 2.8.
 - In the Step field, enter 0.04.
 - In the Stop field, enter 3.2.

Note: Or enter range (2.8,0.04,3.2) in the Parameter value list.

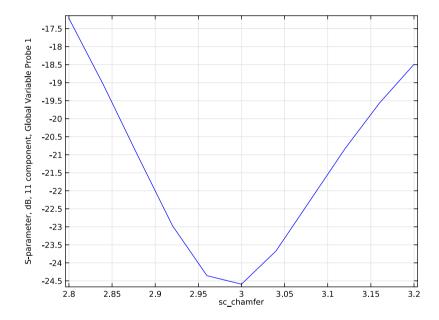
- 5 Click Replace.
- 6 In the Model Builder, right-click Study 1 and choose Compute ■.

Results

Probe I D Plot Group 2

Again, the probe with the reflection coefficient versus the frequency is automatically logged to a table. To get the desired plot versus the geometry parameter, proceed as follows.

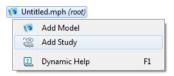
- In the Model Builder, go to Results and select Accumulated Probe Table 1
- 2 In the Table window, right-click the freq column and select Delete Column **※**. In the Table window, click the Table Graph **■**.



The plot of the S-parameter indicates a minimum for a scale factor of about 3.0. Leave the parameter at this value and add a study for the frequency response.

Model Wizard

- I Open the Model Wizard by right-clicking the root node and selecting Add Study **☼**.
- 2 Under Preset Studies select Frequency Domain m.
- 3 Click Finish .



Electromagnetic Waves (emw)

At this stage, it is convenient to partially invoke some higher-level control of what port is excited. This is done by letting the value of a parameter decide. The parameter can in turn be controlled by the solver when computing the full S-parameter matrix by exciting one port at a time. Another advantage is that when this control is activated, a plot group for the S-parameters is automatically created.

- In the Model Builder click the Electromagnetic Waves, Frequency Domain (emw) anode.
- 2 Go to the Electromagnetic Waves, Frequency Domain settings window. Under Port Sweep Settings select the Activate port sweep check box.



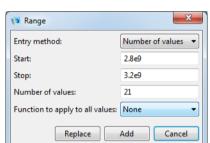
Note: The default variable name PortName is automatically added to the Port parameter name field but must be declared as a global parameter to be available for the parametric sweep.

Global Definitions - Parameters

- In the Model Builder under Global Definitions, click Parameters P. .
- 2 In the Parameters settings window under Parameters:
 - In the Name column, enter PortName
 - In the Expression column, enter 1

Study 2

- In the Model Builder under Study 2, click Step 1: Frequency Domain w.
- **2** Go to the Frequency Domain settings window. Under Study Settings click the Range _ button.
- 3 From the Entry method list, select Number of values.
 - In the Start field, enter 2.8e9.
 - In the Stop field, enter 3.2e9.
 - In the Number of values field, enter 21.
- 4 Click Replace.
- 5 In the Model Builder, right-click Study 2 and choose Compute (or press F8).



Study 1

Study 2

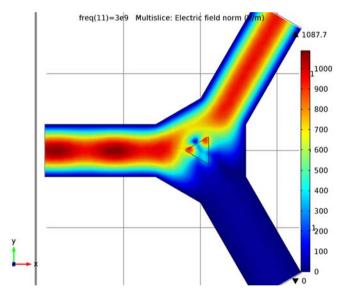
XX Step 1: Frequency Domain

Results

The probe with the reflection coefficient versus the frequency is automatically logged to a table and plotted while solving. At the last frequency, there are pronounced standing waves. Look at the center frequency instead.

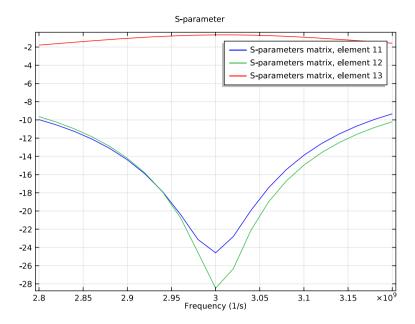
- In the Model Builder under Results, click Electric Field 1 🝙.
- 2 Go to the 3D Plot Group settings window. Under Data from the Parameter value (freq) list, select 3e9.
- 3 Click the Plot ✓ button. Click the Go to XY View button w.





At the center frequency most of the standing waves are gone.

Finally look at all the S-parameters plotted versus the frequency. As the port sweep was activated a plot group has been created automatically.



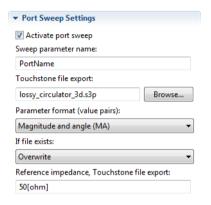
This is the frequency response of the final design.

Electromagnetic Waves (emw)

So far, only the first port has been excited so the full S-parameter matrix remains to be calculated by exciting one port at a time. This is also needed in order to confirm that the circulator behaves as desired. In the RF Module, this procedure is referred to as performing a port sweep. During this stage, the S-parameters can optionally be exported to a Touchstone file for documentation purposes and for use in external system simulation tools. Add the name of the Touchstone file.

- In the Model Builder click the Electromagnetic Waves, Frequency Domain (emw) 🛗 node.
- 2 Go to the Electromagnetic Waves, Frequency Domain settings window. Under Port Sweep Settings in the Touchstone file export field, enter: lossy circulator 3d.s3p.

Note: The Touchstone file is saved in the start directory of COMSOL Multiphysics unless a different file path is specified.



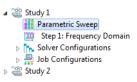
Study I

Reuse the first study for the port sweep. The study is solved for a single frequency to keep simulation time to a minimum although it is possible to solve for a range of frequencies. All solutions are needed in order to display the S-parameter matrix in a table so this setting has to be changed.

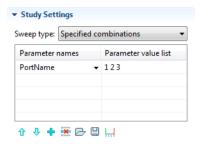
Parametric Sweep

The parametric sweep is used to control which port is excited. It overrides the settings on individual port features and drives one port at a time using 1 W of input power.

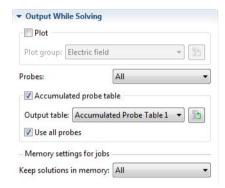
- In the Model Builder under Study 1, click Parametric Sweep #.
- 2 Go to the Parametric Sweep settings window. Under Study Settings in the Parameter names list, select PortName.



3 In the Parameter value list, enter (space separated) 1 2 3.



- 4 Under Output While Solving, choose All from the Keep solutions in memory list.
- 5 In the Model Builder, right-click Study $1 \cong \text{ and choose Compute } = .$

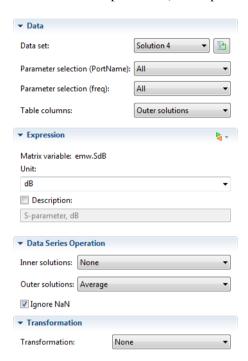


Results

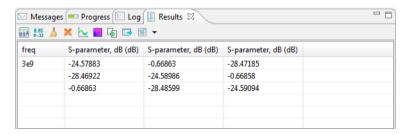
After the computation finishes, the Touchstone file can be inspected in a text editor. The S-parameter matrix can also be displayed in a table.

- In the Model Builder under Results, right-click Derived Values 1855 and choose Global Matrix Evaluation
- 2 Go to the Global Matrix Evaluation settings window. Under Data from the Data set list, choose Solution 4.
- 3 In the upper-right corner of the Expression section, click Replace Expression þ

4 From the menu, choose Electromagnetic Waves, Frequency Domain>Ports>S-parameter, dB>S-parameter, dB (emw.SdB).



5 Click the Evaluate = button and the S-parameter matrix displays in the Table window located in the lower-right side of the COMSOL Desktop. Or select View>Table ## from the main menu to open the Table window.



Note: The matrix is nonsymmetric, which is typical for a device based on a gyrotropic material. There are three groups of matrix elements that differ in the fourth digit. These should in theory be equal within each group so this gives some indication of the discretization errors.

Definitions

As a final step, create a plot and also use it as a model thumbnail. From this plot, it should be possible to identify the model at first glance so it has to display the geometry and some characteristic simulation results. First change to the default 3D view and switch off the grid.

- I Click the Go to Default 3D View ↓ button on the Graphics toolbar.
- 2 In the Model Builder right-click Definitions and choose View ...
- **3** Go to the View settings window.
- 4 Under View click to clear the Show grid check box.

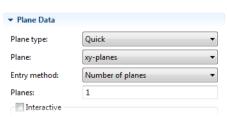


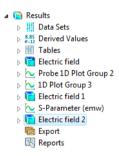
Results

- In the Model Builder under Results, click Electric Field 2 .
- **2** Go to the 3D Plot Group settings window. Under Plot Settings from the View list, choose View 3.
- 3 Click the Plot ✓ button.

Next, delete the multislice and create a single slice.

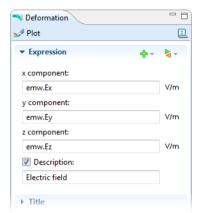
- In the Model Builder under Electric Field 2, right-click Multislice 1 ★ and choose Delete ★ (or press Delete on the keyboard).
- 2 Click Yes to confirm.
- 3 Right-click Electric Field 2 🛅 and choose Slice 🐚.
- **4** In the Slice settings window, under Plane data from the Plane list, choose xy-planes.
- 5 In the Planes field, enter 1.





Add a deformation proportional to the electric field to the slice.

- I Right-click Slice 1 **⋒** and choose Deformation **¬**.
- 2 Go to the Deformation settings window. In the upper-right corner of the Expression section, click Replace Expression ▶ .
- **3** From the menu, choose Electromagnetic Waves, Frequency Domain>Electric>Electric field (emw.Ex,emw.Ey,emw.Ez).
- **4** Under Expression select the Description check box.

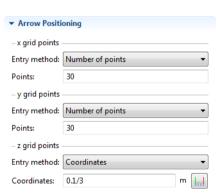


Display the magnetic field as arrows. Use logarithmic length scaling to make sure that the arrows are clearly visible everywhere. Place the arrows well above the slice.

- In the Model Builder, right-click Electric Field 2 and choose Arrow Volume delta.
- 2 Go to the Arrow Volume settings window. In the upper-right corner of the Expression section, click Replace Expression ▶ .

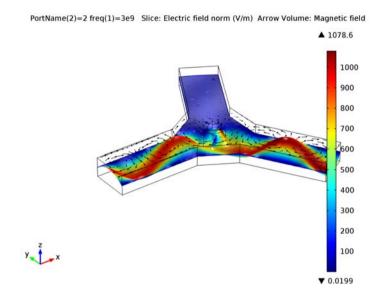


- **3** From the menu, choose Electromagnetic Waves, Frequency Domain>Magnetic>Magnetic field (emw.Hx,emw.Hy,emw.Hz).
- **4** Under Expression select the Description check box.
- 5 Under Arrow Positioning:
 - In the Points field for x grid points, enter 30.
 - In the Points field for y grid points, enter 30
 - For z grid points from the Entry method list, select Coordinates.
 - For z grid points in the Coordinates field, enter 0.1/3.
- 6 Under Coloring and Style from the Arrow length list, choose Logarithmic. From the Color list, choose Black.



The port excitation can now be selected for the plot group. For the model thumbnail, select the second port.

- In the Model Builder, click Electric Field 2 .
- 2 In the 3D Plot Group settings window under Data, choose 2 from the Parameter value (PortName) list.
- 3 Click the Plot ✓ button.



Select this plot to use as a model thumbnail.

- In the Model Builder under Results click Electric Field 2 .
- **2** From the File menu, choose Save Model Thumbnail.

To view the thumbnail image, click the Root node and look under the Model Thumbnail section.

This concludes the modeling session unless you want to practice drawing a "Geometry Sequence" on page 40.

Geometry Sequence

In "Global Definitions - Parameters and Variables" on page 16 parameters were entered to prepare for drawing the circulator geometry. Once the geometry is created, you can then experiment with different dimensions and update the values of sc_chamfer and sc_ferrite and rerun the geometry sequence. These step-by-step instructions build the same geometry that is contained in the Model Library file lossy_circulator_3d_geom (imported in the section "Geometry" on page 18).

The geometry is built by first defining a 2D cross section of the 3D geometry in a work plane. The 2D geometry is then extruded into 3D.

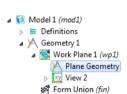
Note: You need to complete the first two sections "Model Wizard" on page 16 and "Global Definitions - Parameters and Variables" on page 16 before defining the geometry.

Start by defining one arm of the circulator, then copy and rotate it twice to build all three arms.

Rectangle I

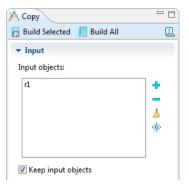
- In the Model Builder, right-click Geometry 1 / and choose Work Plane \(\bigs_{\text{.}} \).
- 2 Under Work Plane 1 (wp1), click Plane Geometry

 ∧.
- 4 Go to the Rectangle settings window. Under Size in the:
 - Width field, enter 0.2-0.1/(3*sqrt(3)).
 - Height field, enter 0.2/3.
- 5 Under Position in the:
 - xw field, enter -0.2.
 - yw field, enter -0.1/3.
- 6 Click the Build Selected | button.

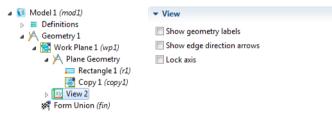


Copy I

- In the Model Builder under Work Plane 1 (wp1), right-click Plane Geometry ▲ and choose Transforms>Copy ▶.
- 2 Select the object r1 by left-clicking in the graphics window (the rectangle is highlighted red) and then right-click the rectangle (it is highlighted blue). The object r1 is added to the Input objects list on the Copy settings window. Click the Build Selected button.



Note: To turn on the geometry labels in the Graphics window, in the Model Builder under Geometry 1>WorkPlane 1>Plane Geometry, click the View 2 node. Go to the View settings window and select the Show geometry labels check box

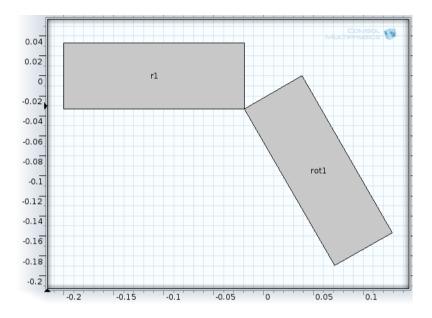


Rotate I

- In the Model Builder under Work Plane 1 (wp1), right-click Plane Geometry

 ∧ and choose Transforms>Rotate ⊚.
- 2 Select only the object copy1 (left-click twice before right-clicking or you get r1).
- 3 Go to the Rotate settings window. Under Rotation Angle in the Rotation field, enter 120.

4 Click the Build Selected button and then click the Zoom Extents button on the Graphics toolbar. The geometry should match the figure so far.



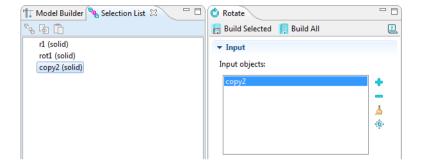
Сору 2

- I Right-click Plane Geometry ★ and choose Transforms>Copy 凝.
- 2 Select the object r1 only and add it to the Input objects list in the Copy settings window.
- 3 Click the Build Selected 🚪 button.

Rotate 2

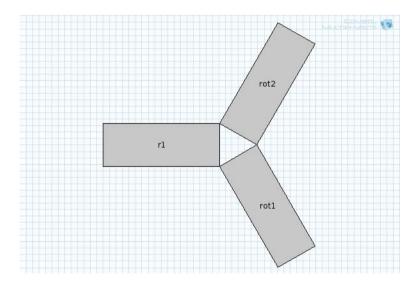
- I Right-click Plane Geometry ★ and choose Transforms>Rotate ⊚.
- 2 Select the object copy2 only.

Note: If you cannot locate copy2 in the Graphics window, select View>Selection List and click to highlight copy2 in the list. Then right-click copy2 in the list to add it to the Input objects list.



- 3 Go to the Rotate settings window. Under Rotation Angle in the Rotation field, enter - 120.
- 4 Click the Build Selected 🚪 button and then click the Zoom Extents 👰 button on the Graphics toolbar.

The geometry should match this figure.



Next, unite the three arms to one object.

Union 1

- I Under Work Plane 1 (wp1), right-click Plane Geometry ⋈ and choose Boolean Operations>Union ■.
- 2 Select the objects r1, rot1, and rot2 only and add these to the Input objects list in the Union settings window.
- 3 Click the Build Selected 🚪 button. There is one object created called unil.

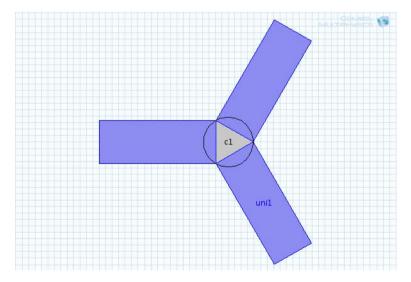
Now, build the central connecting region and add the ferrite domain. During these stages, the geometric design parameters are used. Start by creating a triangle connecting the arms and then by subtracting a copy of what has already been drawn from a circle of proper radius.

Circle 1

- I Under Work Plane 1 (wp1), right-click Plane Geometry ∧ and choose Circle ○.
- 2 In the Circle settings window under Size and Shape, enter 0.2/(3*sqrt(3)) in the Radius field.
- 3 Click the Build Selected # button.

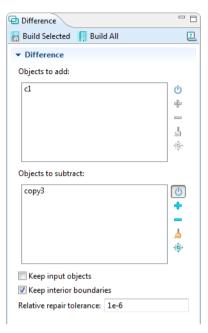
Сору 3

- I Under Work Plane 1 (wp1), right-click Plane Geometry ∧ and choose Transforms>Copy ₄. A Copy 3 node is added to the sequence.
- 2 Select the object unil only.

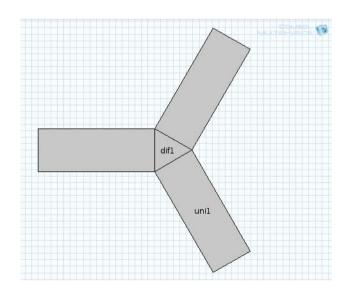


Difference I

- I Right-click Plane Geometry ★ and choose Boolean Operations>Difference ...
- 2 Select the object c1 only to add it to the Objects to add list in the Difference settings window. Go to the Difference settings window. To the right of the Objects to subtract section, click the Activate Selection button ...
- 3 Select View>Selection List and click to highlight copy3 in the list. Then right-click copy3 in the list to add it to the Objects to subtract list.
- 4 Click the Build Selected 🚪 button.



The geometry should match this figure so far.



Now, rotate the newly created triangle 180 degrees and use one scaled copy of it to create linear fillets for impedance matching. Use another scaled copy to define the ferrite.

Rotate 3

- In the Model Builder under Work Plane 1 (wp1), right-click Plane Geometry A and choose Transforms>Rotate .
- 2 Select the object difl only.
- 3 Go to the Rotate settings window. Under Rotation Angle in the Rotation field, enter 180.
- 4 Click the Build Selected | button.

Copy 4

- 2 Select the object rot3 only.

It is now time to apply the first scaling for the impedance matching.

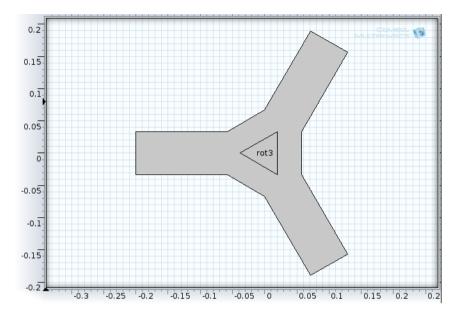
Scale 1

- Right-click Plane Geometry A and choose Transforms-Scale .
- 2 Go to the Scale settings window. Under Scale Factor in the Factor field, enter sc chamfer (one of the parameters entered in the step "Global Definitions -Parameters and Variables" on page 16).
- 3 Select the object copy4 only.
- 4 Click the Build Selected 🚪 button.

Union 2

- I Right-click Plane Geometry ⋈ and choose Boolean Operations>Union ⊕.
- 2 Select the objects unil and scal only. Use the Selection List to select the objects if required.
- 3 Go to the Union settings window. Under Union click to clear the Keep interior boundaries check box.
- 4 Click the Build Selected 🚪 button.

The geometry should match this figure.



Next, apply the scaling for the ferrite region.

Scale 2

- In the Model Builder right-click Plane Geometry ∧ and choose Transforms>Scale ⋈.
- 2 Select the object rot3 only.
- 3 Go to the Scale settings window. Under Scale Factor in the Factor field, enter sc ferrite.
- 4 Click the Build Selected # button.

Extruding the 2D cross-section into a 3D solid geometry finalizes the geometry.

Extrude 1

- I Right-click Work Plane 1 (wp1)

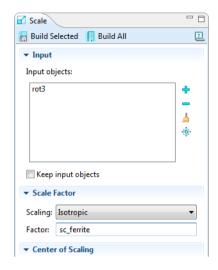
 sample and choose Extrude

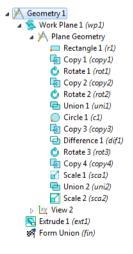
 sample.
- 2 Go to the Extrude settings window. Under Distances from Plane in the associated table, enter 0.1/3 in the Distances (m) column.
- 3 Click the Build Selected ☐ button and then click the Zoom Extents ⊕ button on the Graphics toolbar.

Form Union

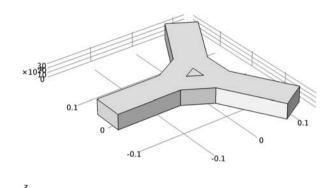
- 2 In the Form Union settings window, click Build All

The final sequence of Geometry nodes in the Model Builder should match the figure to the right.





The last step finalizes the geometry and turns it into a form suitable for the simulation by removing duplicate faces, for example. It is performed automatically when material is added or when physics are defined, but it is good practice to perform it manually as any error messages from this step may be confusing when appearing at a later stage. The geometry should match this figure.



Note: If you skipped to this section to learn how to create the geometry, you can now return to the next tutorial step: "Materials" on page 18.