## Image Transforms: A brief introduction

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"Image and video processing in the compressed domain", CRC Press, 2011.

## $f(x,y) = \sum_{j} \sum_{i} \lambda_{ij} b_{ij}(x,y)$

### Image Transform

- Image in continuous form: f(x,y): A 2-D function, where (x,y) in R².
   Let B be a set of basis functions: can be extended in the analysis
- in the analysis.

$$B = \{b_i(x,y) \mid i = ..., -1, 0, 1, 2, 3, ....\}, b_i(x,y) \text{ in } R \text{ or } C.$$

Let f(x,y) be expanded using B as follows:

$$f(x,y) = \sum_{i} \lambda_{i} b_{i}(x,y)$$
 Coefficients of transform

The **transform** of f w.r.t. B is given by  $\{\lambda_i | i = \dots -1,0,1,2,3,\dots\}$ .

Indexing may be multidimensional say,  $\lambda_{ij}$ .

## Orthogonal Expansion $f(x) = \sum_{i} \lambda_i b_i(x)$ and 1-D Transforms

$$f(x) = \sum_{i} \lambda_{i} b_{i}(x)$$

- Inner product:  $\langle f, g \rangle = \int f(x)g^*(x)dx$
- Orthogonal expansion: If B satisfies:

$$\langle b_i.b_j \rangle = 0, \quad \textit{for } i \neq j,$$
 comma  $comma = c_i, \quad \textit{Otherwise (for } i = j), \ \textit{where } c_i > 0.$ 

- Transform coefficients in O.E.:  $\lambda_i = \frac{1}{c_i} < f.b_i > 1$  If  $c_i = 1$ , it becomes orthonormal expansion. Forward transform  $\lambda_i = < f.b_i > 1$
- Inverse transform:  $f(x) = \int_{i-\infty}^{\infty} \lambda_i . b_i(x) di$

## **Fourier** transform

Complete base 
$$B = \{e^{-j\omega x} | -\infty < \omega < \infty\}$$

$$B = \{e^{-j\omega x} | -\infty < \omega < \infty$$

Orthogonality:

$$\int_{-\infty}^{\infty} e^{j\omega x} dx = \begin{cases} 2\pi\delta(\omega), & \text{for } \omega = 0, \\ 0, & \text{Otherwise.} \end{cases}$$

Fourier Transform:

$$\mathbb{F}(f(x)) = \hat{f}(j\omega) = \int_{-\infty}^{\infty} f(x)e^{-j\omega x} dx$$

**Inverse Transform:** 

$$f(x) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \hat{f}(j\omega) \cdot e^{j\omega x} d\omega$$

Full reconstruction

$$e^{-j\omega x} = \cos(\omega x) - j\sin(\omega x)$$

$$\hat{f}(j\omega) = \int_{-\infty}^{\infty} f(x)(\cos(\omega x) - j\sin(\omega x)) dx$$

$$C = \{\cos(\omega x) | -\infty < \omega < \infty\}$$

$$S = \{\sin(\omega x) | -\infty < \omega < \infty\}$$
Orthogonal But not complete!

# Even and odd functions

$$\cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2}$$
$$\sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

- Even: f(-x)=f(x) for all x.
- Odd: f(-x) = -f(x) for all x.  $\rightarrow f(0) = 0$ .
- For even f(x):  $\int_{-\infty}^{\infty} f(x)(\sin(\omega x)) dx = 0$
- For odd f(x):  $\int_{-\infty}^{\infty} f(x)(\cos(\omega x)) dx = 0$

 Full reconstruction possible with cosines (sines) only if it is even (odd).

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### Discrete representation

Discrete representation of a function:

$$f(n) = \{f(nX_{{f 0}})|n\in{\Bbb Z}\}$$
 Set of integers Sampling interval

- Can be considered as a vector in an infinite dimensional vector space.
- In our context, it is of a finite dimensional space, e.g.  $\{f(n), n=0,1,..N-1\}$ , or
- $f = [f(0) f(1) \dots f(N-1)]^T$ .

### Discrete Transform

- For *n*-dimensional vector X any linear transform, e.g.  $Y_{mx1} = B_{mxn} X_{nx1}$
- Has inverse transform if B is square of size (nxn) and invertible.
- Rows of B are called basis vectors.
- $Y(i) = \langle \boldsymbol{b}_i^{*T}, X \rangle$  dot product or inner product.

Orthogonality condition:

$$< \boldsymbol{b}_{i}^{*T}. \boldsymbol{b}_{j} > = 0 \text{ if } i \neq j$$
  
=  $c_{i}$ , otherwise

 $B = \begin{bmatrix} \boldsymbol{b}_0^{*T} \\ \boldsymbol{b}_1^{*T} \\ \vdots \\ \boldsymbol{h}^{*T} \end{bmatrix}$ 

## Discrete Fourier Transform

(DFT)

$$b_k(n) = \frac{1}{\sqrt{N}} e^{j2\pi \frac{k}{N}n}$$
, for  $0 \le n \le N - 1$ , and  $0 \le k \le N - 1$ .

$$\hat{f}(k) = \sum_{n=0}^{N-1} f(n)e^{-j2\pi \frac{k}{N}n} \text{ for } 0 \le k \le N-1.$$
  $\hat{f}(N+k) = \hat{f}(k)$ 

$$f(n)=rac{1}{N}\sum_{k=0}^{N-1}\hat{f}(k)e^{j2\pi\frac{k}{N}n}$$
 for  $0\leq n\leq N-1$ . Hermitian Transpose

$$\mathbf{X} = \mathbf{F}\mathbf{X}$$
  $\mathbf{F} = \left[e^{-j2\pi\frac{k}{N}n}\right]_{0 \le (k,n) \le N-1}$ .  $\mathbf{F}^{-1} = \frac{1}{N}\mathbf{F}^H$ 

A single period

$$f(n+N)=f(n)$$

DFT: Fourier series of a periodic function frequency:  $1/(NX_0)$ 

**Fundamental** 

## Generalized

# Discrete Fourier

$$\mathbf{F}_{\alpha,\beta} = \begin{bmatrix} e^{-j2\pi \frac{k+\alpha}{N}(n+\beta)} \end{bmatrix}_{0 \le (k,n) \le N-1}.$$

$$\mathbf{F}_{\alpha,\beta}^{-1} = \frac{1}{N} \mathbf{F}_{\alpha,\alpha}^{H} = \frac{1}{N} \mathbf{F}_{\alpha,\alpha}^{*}.$$

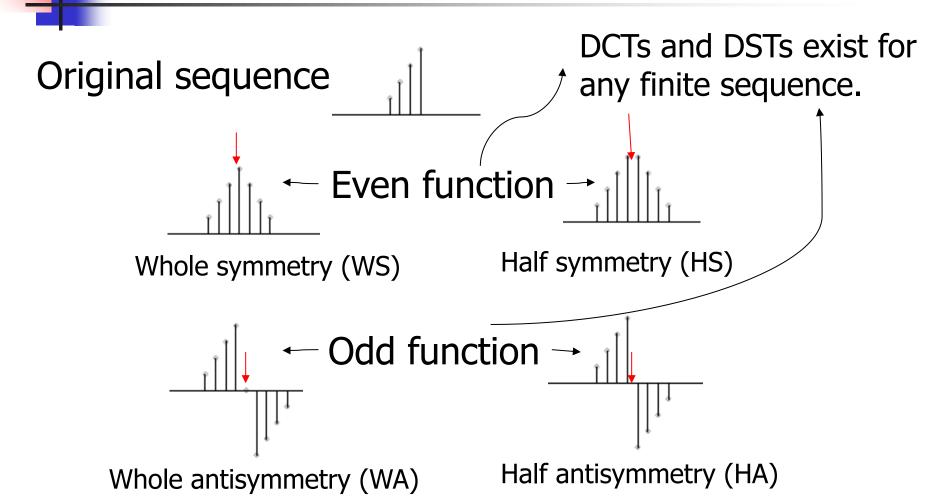
$$b_k^{(\alpha,\beta)}(n) = \frac{1}{\sqrt{N}} e^{j2\pi \frac{k+\alpha}{N}(n+\beta)}, \text{ for } 0 \le n \le N-1, \text{ and } 0 \le k \le N-1$$

$$\hat{f}_{\alpha,\beta}(k) = \sum_{n=0}^{N-1} f(n)e^{-j2\pi\frac{k+\alpha}{N}(n+\beta)}, \text{ for } 0 \le k \le N-1$$

$$f(n) = \frac{1}{N} \sum_{k=0}^{N-1} \hat{f}_{\alpha,\beta}(k) e^{j2\pi \frac{k+\alpha}{N}(n+\beta)}, \text{ for } 0 \le n \le N-1$$

$\alpha$	$\beta$	Transform name	Notation
0	0	Discrete Fourier Transform $(DFT)$	$\hat{f}(k)$
0	$\frac{1}{2}$	Odd Time Discrete Fourier Transform $(OTDFT)$	$\hat{f}_{0,\frac{1}{2}}(k)$
$\frac{1}{2}$	0	Odd Frequency Discrete Fourier Transform $(OFDFT)$	$\hat{f}_{\frac{1}{2},0}(k)$
$\frac{1}{2}$	$\frac{1}{2}$	Odd Frequency Odd Time Discrete Fourier Transform $({\cal O}^2DFT)$	$\hat{f}_{\frac{1}{2},\frac{1}{2}}(k)$

# Symmetric / Antisymmetric extension of a finite sequence

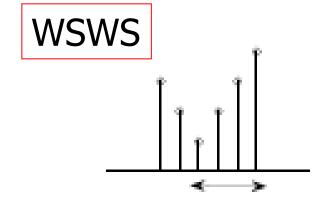


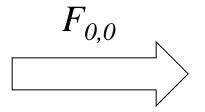


### Discrete Cosine / Sine Transforms

$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$

 Types of symmetric / antisymmetric extensions at the two ends of a sequence and a type of GDFT→ DCTs / DSTs



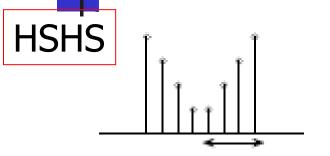


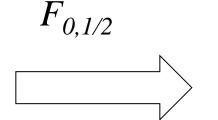
Type-I Even DCT

$$C_{1e}(x(n)) = X_{Ie}(k) = \sqrt{\frac{2}{N}} \alpha^2(k) \sum_{n=0}^{N} x(n) \cos\left(\frac{2\pi nk}{2N}\right), \ 0 \le k \le N,$$

# Discrete Cosine / Sine Transforms

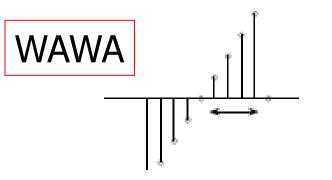
$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$

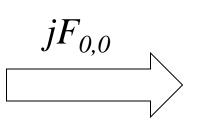




### Type-2 Even DCT

$$C_{2e}(x(n)) = X_{IIe}(k) = \sqrt{\frac{2}{N}}\alpha(k)\sum_{n=0}^{N-1}x(n)\cos\left(\frac{2\pi k(n+\frac{1}{2})}{2N}\right), \ 0 \le k \le N-1$$



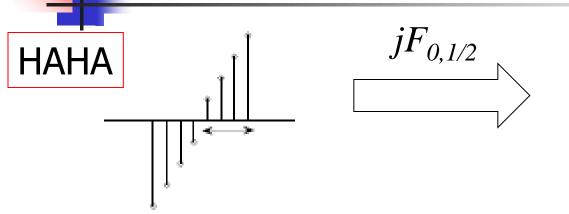


### Type-1 Even DST

$$S_{1e}(x(n)) = X_{sIe}(k) = \sqrt{\frac{2}{N}} \sum_{i=1}^{N-1} x(n) \sin\left(\frac{2\pi kn}{2N}\right), \ 1 \le k \le N-1$$

## Discrete Cosine / Sine Transforms

$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$



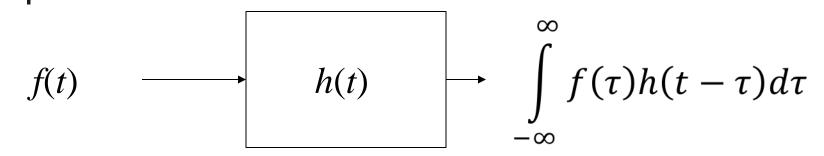
Type-2 Even DST

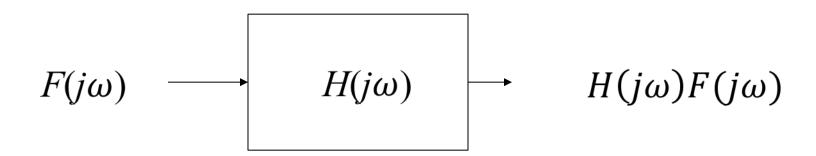
$$S_{2e}(x(n)) = X_{sIIe}(k) = \sqrt{\frac{2}{N}} \alpha(k) \sum_{n=0}^{N-1} x(n) \sin\left(\frac{2\pi k(n+\frac{1}{2})}{2N}\right), \ 1 \le k \le N-1$$

There exist 16 different types of DCTs and DSTs. Type-II Even DCT is used in signal, image, and video compression.

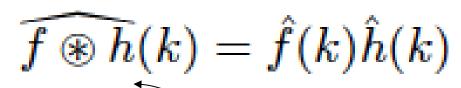


# Convolution Multiplication Property (CMP)





**CMP for Fourier Transform** 



CMP for DFT holds for

### CMP for DFT

Linear convolution

circular convolution.

$$f(n)$$
  $\longrightarrow$   $h(n)$   $\longrightarrow \sum_{m=-\infty}^{\infty} f(m)h(n-m)$ 

- Periodic convolution: Convolution between two finite sequences with periodic extension.
- It is defined if both have the same period, providing a periodic sequence with the same period.

#### **Circular Convolution**

$$f \circledast h(n) = \sum_{m=0}^{N-1} f(m)h(n-m),$$
  
=  $\sum_{m=0}^{n} f(m)h(n-m) + \sum_{m=n+1}^{N-1} f(m)h(n-m+N).$ 

### Antiperiodic extension and skew-circular convolution

- Antiperiodic function with an antiperiod N, if f(x+N)=-f(x).
- An antiperiodic function of antiperiod  $N \rightarrow$ a periodic function of period 2N.
- Skew-circular convolution: convolution between two antiperiodic extended sequences of the same antiperiod.

$$f(\hat{\mathbf{S}})h(n) = \sum_{m=0}^{N-1} f(m)h(n-m),$$
  
= 
$$\sum_{m=0}^{n} f(m)h(n-m) - \sum_{m=n+1}^{N-1} f(m)h(n-m+N)$$



# CMPs for DCTs and DSTs

$$u(n) = x(n) \circledast y(n)$$

$$w(n) = x(n) \otimes y(n)$$

$$C_{1e}(u(n)) = \sqrt{2N}C_{1e}(x(l))C_{1e}(y(m))$$

$$C_{2e}(u(n)) = \sqrt{2N}C_{2e}(x(l))C_{1e}(y(m))$$
Input Filter response
$$S_{2e}(u(n)) = \sqrt{2N}C_{2e}(x(l))S_{1e}(y(m))$$

$$S_{2e}(u(n)) = \sqrt{2N}S_{2e}(x(l))C_{1e}(y(m))$$

$$C_{3e}(w(n)) = \sqrt{2N}C_{3e}(x(l))C_{3e}(y(m))$$

# Properties of Type-II DCT

$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$
 
$$X = C_N.\mathbf{x}, \quad C_N^{-1} = C_N^T$$

- Matrix form:  $C_N = \left[ \sqrt{\frac{2}{N}}.\alpha(k)cos(\frac{\pi k(2n+1)}{2N}) \right]_{0 \leq (k,n) \leq N-1}.$  N-point DCT
- Each row is either symmetric (even row) or antisymmetric (odd row).

$$C_N(k, N-1-n) = \begin{cases} C_N(k, n) & \text{for } k \text{ even} \\ -C_N(k, n) & \text{for } k \text{ odd} \end{cases}$$

Sub-band relationship:

$$x_L(n) = \frac{1}{2} \{x(2n) + x(2n+1)\},\$$
  
 $x_H(n) = \frac{1}{2} \{x(2n) - x(2n+1)\}, \quad n = 0, 1, \dots, \frac{N}{2} - 1.$ 

## Type-II DCT:

## Sub-band relation $C_N = \left[\sqrt{\frac{2}{N}}.\alpha(k)cos(\frac{\pi k(2n+1)}{2N})\right]_{0 \le (k,n) \le N-1}$

$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$

$$C_N = \left[ \sqrt{\frac{2}{N}} \cdot \alpha(k) \cos(\frac{\pi k (2n+1)}{2N}) \right]_{0 \le (k,n)}$$

### Sub-band relationship:

$$x_L(n) = \frac{1}{2} \{x(2n) + x(2n+1)\},\$$
  
 $x_H(n) = \frac{1}{2} \{x(2n) - x(2n+1)\}, \quad n = 0, 1, \dots, \frac{N}{2} - 1.$   
 $X_L(k) = C_{2e} (x_L(n)) \quad S_H(k) = S_{2e} (x_H(n))$ 

$$X(k) = \sqrt{2}\cos(\frac{\pi k}{2N})\overline{X_L}(k) + \sqrt{2}\sin(\frac{\pi k}{2N})\overline{S_H}(k), \ 0 \le k \le N - 1,$$

$$\overline{X_L}(k) = \begin{cases} X_L(k), & 0 \le k \le \frac{N}{2} - 1, \\ 0, & k = \frac{N}{2}, \\ -X_L(N-k), & \frac{N}{2} + 1 \le k \le N - 1, \end{cases} \quad \overline{S_H}(k) = \begin{cases} S_H(k), & 0 \le k \le \frac{N}{2} - 1, \\ \sqrt{2} \sum_{n=0}^{\frac{N}{2} - 1} (-1)^n x_H(n), & k = \frac{N}{2}, \\ S_H(N-k), & \frac{N}{2} + 1 \le k \le N - 1, \end{cases}$$

# Type-II DCT: Sub $\overline{X_L}(k) = \begin{cases} X_L(k), & 0 \le k \le \frac{N}{2} - 1, \\ 0, & k = \frac{N}{2}, \\ -X_L(N-k), & \frac{N}{2} + 1 \le k \le N - 1, \end{cases}$ band Approximation

Sub-band approximation:

$$X(k) = \begin{cases} \sqrt{2}\cos(\frac{\pi k}{2N})\overline{X_L}(k), & k \in \{0, 1, \dots, \frac{N}{2} - 1\}, \\ 0, & \text{otherwise.} \end{cases}$$

Low-pass truncated approximation:

$$X(k) = \begin{cases} \sqrt{2} \cdot \overline{X_L}(k), & k \in \{0, 1, \dots, \frac{N}{2} - 1\}, \\ 0, & \text{otherwise.} \end{cases}$$

Provide conversion from N-point DCT to N/2-point DCT and vice versa.



# Block composition and decomposition

### $X^{(N)} = C_N \mathbf{x}$

 To convert M adjacent N-point DCT blocks to a single MxN-point DCT block.

$$X^{(MN)} = A_{(M,N)} [X_0^{(N)} X_1^{(N)} \dots X_{M-1}^{(N)}]^T,$$
 Inversion decomposes a large block. 
$$\begin{bmatrix} C_N^{-1} & 0_N & 0_N & \cdots & 0_N & 0_N \\ 0_N & C_N^{-1} & 0_N & \cdots & 0_N & 0_N \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ 0_N & 0_N & 0_N & \cdots & C_N^{-1} & 0_N \\ 0_N & 0_N & 0_N & \cdots & 0_N & C_N^{-1} \end{bmatrix},$$
 NxN zero matrix

### A typical example

 Convert two 4-point DCT block to a single 8-point block

$$X^{(8)} = A_{(2,4)} \begin{bmatrix} X_1^{(4)} \\ X_2^{(4)} \end{bmatrix}$$

$$A_{(2,4)}^{-1} \begin{bmatrix} X_1^{(4)} \\ X_2^{(4)} \end{bmatrix} \rightarrow X^{(8)}$$

$$f(x,y) = \sum_{i} \sum_{i} \lambda_{ij} b_{ij}(x,y)$$

### 2-D Transforms

■ Easily extendable if basis functions are separable, i.e.  $B=\{b_{ij}(x,y)=g_i(x).g_j(y)\}.$ 

They could be from two different sets, say b(x,y)=g(x).h(y).

1-D basis function

- *B*: Orthogonal if  $G=\{g_i(x), i=1,2,...\}$  is orthogonal.
- B: Orthogonal and complete if G is so.
- Reuse of 1-D transform computation.

$$\lambda_{ij} = \sum_{i} g_j^*(y) \left( \sum_{i} f(x, y) g_i^*(x) \right)$$

## 2D Discrete Transform

$$Y_{mxn} = B_{mxm} X_{mxn} B_{nxn}^{T}$$

- Use of separability:
  - Transform columns.
  - Transform rows.
- Input:  $X_{m \times n}$  1-D Transform Matrix: B
- Transform columns:  $[Y_1]_{m \times n} = B_{m \times m} X_{m \times n}$
- Transform rows:  $Y_{mxn} = [B_{nxn}Y_1^T]^T$   $= Y_1B_{nxn}^T$  $= B_{mxm}X_{mxn}B_{nxn}^T$

$$\alpha(p) = \begin{cases} \sqrt{\frac{1}{2}}, & \text{for } p = 0 \text{ or } N, \\ 1, & \text{otherwise.} \end{cases}$$

#### Type-I:

$$X_{I}(k,l) = \frac{2}{N} \cdot \alpha^{2}(k) \cdot \alpha^{2}(l) \cdot \sum_{m=0}^{M} \sum_{n=0}^{N} (x(m,n) \cos(\frac{m\pi k}{M}) \cos(\frac{n\pi l}{N})),$$
  
  $0 \le k \le M, 0 \le l \le N.$ 

#### Type-II

$$\begin{array}{lcl} X_{II}(k,l) & = & \frac{2}{N}.\alpha(k).\alpha(l).\sum_{m=0}^{M-1}\sum_{n=0}^{N-1}(x(m,n)\cos(\frac{(2m+1)\pi k}{2M})\cos(\frac{(2n+1)\pi l}{2N})), \\ & & 0 \leq k \leq M-1, 0 \leq l \leq N-1. \end{array}$$

### Matrix Representation:

$$X = DCT(x) = C_M.x.C_N^T$$

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## 2D DCT: Sub-band relation

$$\begin{array}{rcl} x_{LL}(m,n) & = & \frac{1}{4}\{x(2m,2n) + x(2m+1,2n) \\ & + x(2m,2n+1) + x(2m+1,2n+1)\}, \ 0 \leq m,n \leq \frac{N}{2} - 1. \end{array}$$

#### Sub-band approximation:

2D DCT of 
$$x_{LL}(m,n)$$
 
$$X(k,l) = \left\{ \begin{array}{ll} 2cos(\frac{\pi k}{2N})\cos(\frac{\pi l}{2N})\overline{X_{LL}}(k,l), & k,l=0,1,....,\frac{N}{2}-1\\ 0, & \text{otherwise.} \end{array} \right.$$

#### Low-pass truncated approximation:

$$X(k,l) = \begin{cases} 2\overline{X_{LL}}(k,l), & k,l = 0,1,\dots,\frac{N}{2} - 1\\ 0, & \text{otherwise.} \end{cases}$$

# 2D DCT: Block composition and decomposition

$$X^{(LN\times MN)} = A_{(L,N)} \begin{bmatrix} X_{0,0}^{(N\times N)} & X_{0,1}^{(N\times N)} & \cdots & X_{0,M-1}^{(N\times N)} \\ X_{1,0}^{(N\times N)} & X_{1,1}^{(N\times N)} & \cdots & X_{1,M-1}^{(N\times N)} \\ \vdots & \vdots & \ddots & \vdots \\ X_{L-1,0}^{(N\times N)} & X_{L-1,1}^{(N\times N)} & \cdots & X_{L-1,M-1}^{(N\times N)} \end{bmatrix} A_{(M,N)}^{T}$$

$$\left[ \begin{array}{ccccc} X_{0,0}^{(N\times N)} & X_{0,1}^{(N\times N)} & \cdots & X_{0,M-1}^{(N\times N)} \\ X_{1,0}^{(N\times N)} & X_{1,1}^{(N\times N)} & \cdots & X_{1,M-1}^{(N\times N)} \\ \vdots & \vdots & \ddots & \vdots \\ X_{L-1,0}^{(N\times N)} & X_{L-1,1}^{(N\times N)} & \cdots & X_{L-1,M-1}^{(N\times N)} \end{array} \right] = A_{(L,N)}^{-1} X^{(LN\times MN)} A_{(M,N)}^{-1^T}$$

## 2D DCT: CMP

Circular convolution with respective symmetric extensions.

$$C_{2e}\{x(m,n) \not\vdash h(m,n)\} = C_{2e}\{x(m,n)\}C_{1e}\{h(m,n)\}$$
  
 $C_{1e}\{x(m,n) \not\vdash h(m,n)\} = C_{2e}\{x(m,n)\}C_{2e}\{h(m,n)\}$ 



