

5 Properties of Linear, Time-Invariant Systems

Solutions to Recommended Problems

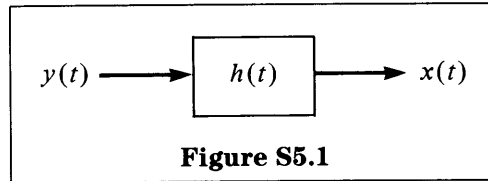
S5.1

The inverse system for a continuous-time accumulation (or integration) is a differentiator. This can be verified because

$$\frac{d}{dt} \left[\int_{-\infty}^t x(\tau) d\tau \right] = x(t)$$

Therefore, the input-output relation for the inverse system in Figure S5.1 is

$$x(t) = \frac{dy(t)}{dt}$$



S5.2

(a) We want to show that

$$h[n] - ah[n - 1] = \delta[n]$$

Substituting $h[n] = a^n u[n]$, we have

$$a^n u[n] - aa^{n-1} u[n - 1] = a^n (u[n] - u[n - 1])$$

But

$$u[n] - u[n - 1] = \delta[n] \quad \text{and} \quad a^n \delta[n] = a^0 \delta[n] = \delta[n]$$

- (b) (i) The system is not memoryless since $h[n] \neq k\delta[n]$.
(ii) The system is causal since $h[n] = 0$ for $n < 0$.
(iii) The system is stable for $|a| < 1$ since

$$\sum_{n=0}^{\infty} |a|^n = \frac{1}{1 - |a|}$$

is bounded.

(c) The system is not stable for $|a| > 1$ since $\sum_{n=0}^{\infty} |a|^n$ is not finite.

S5.3

(a) Consider $x(t) = \delta(t) \rightarrow y(t) = h(t)$. We want to verify that $h(t) = e^{-2t}u(t)$, so

$$\frac{dy(t)}{dt} = -2e^{-2t}u(t) + e^{-2t}\delta(t), \quad \text{or}$$

$$\frac{dy(t)}{dt} + 2y(t) = e^{-2t}\delta(t),$$

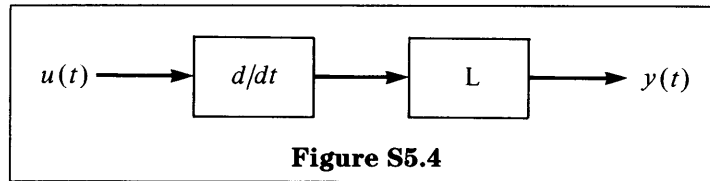
but $e^{-2t} \delta(t) = \delta(t)$ because both functions have the same effect on a test function within an integral. Therefore, the impulse response is verified to be correct.

- (b) (i) The system is not memoryless since $h(t) \neq k\delta(t)$.
(ii) The system is causal since $h(t) = 0$ for $t < 0$.
(iii) The system is stable since $h(t)$ is absolutely integrable.

$$\begin{aligned} \int_{-\infty}^{\infty} |h(t)| dt &= \int_0^{\infty} e^{-2t} dt = -\frac{1}{2}e^{-2t} \Big|_0^{\infty} \\ &= \frac{1}{2} \end{aligned}$$

S5.4

By using the commutative property of convolution we can exchange the two systems to yield the system in Figure S5.4.



Now we note that the input to system L is

$$\frac{du(t)}{dt} = \delta(t),$$

so $y(t)$ is the impulse response of system L. From the original diagram,

$$\frac{ds(t)}{dt} = y(t)$$

Therefore,

$$h(t) = \frac{ds(t)}{dt}$$

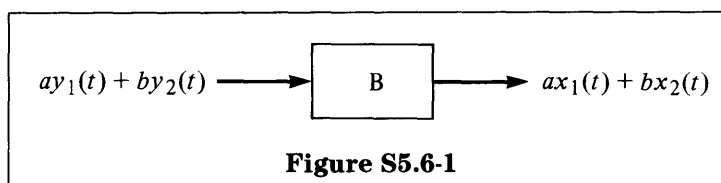
S5.5

- (a) By definition, an inverse system cascaded with the original system is the identity system, which has an impulse response $h(t) = \delta(t)$. Therefore, if the cascaded system has an input of $\delta(t)$, the output $w(t) = h(t) = \delta(t)$.
(b) Because the system is an identity system, an input of $x(t)$ produces an output $w(t) = x(t)$.

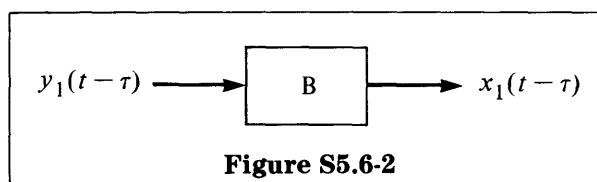
Solutions to Optional Problems

S5.6

- (a) If $y(t) = ay_1(t) + by_2(t)$, we know that since system A is linear, $x(t) = ax_1(t) + bx_2(t)$. Since the cascaded system is an identity system, the output $w(t) = ax_1(t) + bx_2(t)$.



- (b) If $y(t) = y_1(t - \tau)$, then since system A is time-invariant, $x(t) = x_1(t - \tau)$ and also $w(t) = x_1(t - \tau)$.



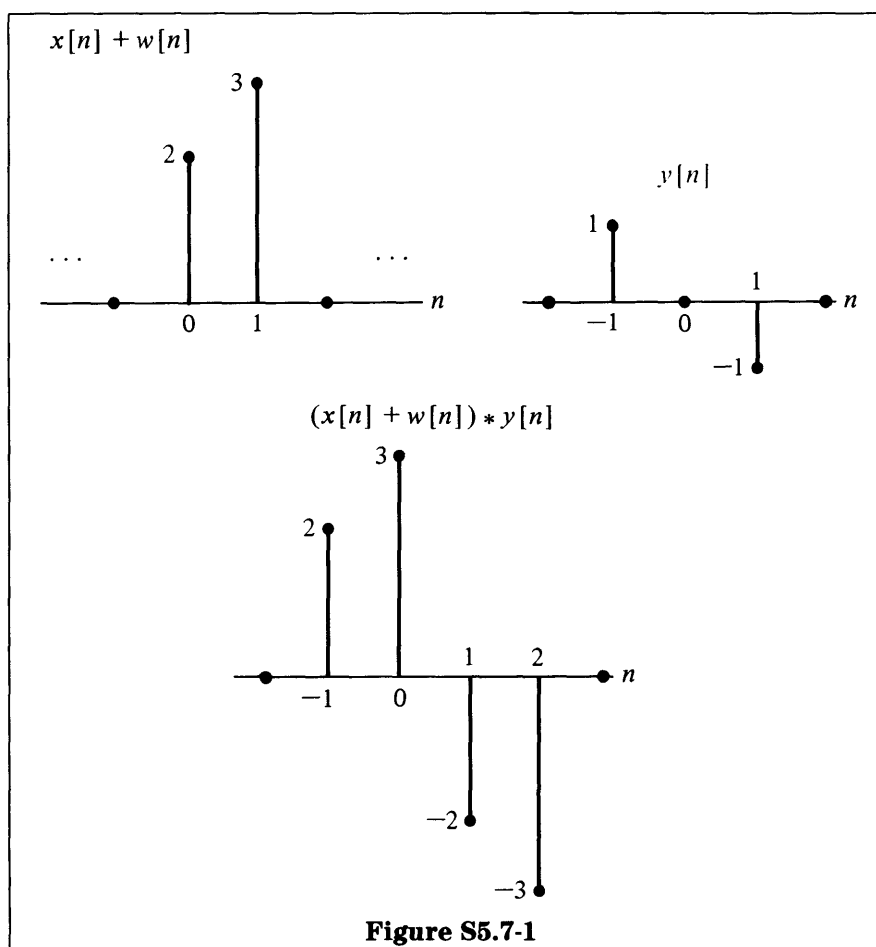
- (c) From the solutions to parts (a) and (b), we see that system B is linear and time-invariant.

S5.7

- (a) The following signals are obtained by addition and graphical convolution:

$$(x[n] + w[n]) * y[n] \quad (\text{see Figure S5.7-1})$$

$$x[n] * y[n] + w[n] * y[n] \quad (\text{see Figure S5.7-2})$$



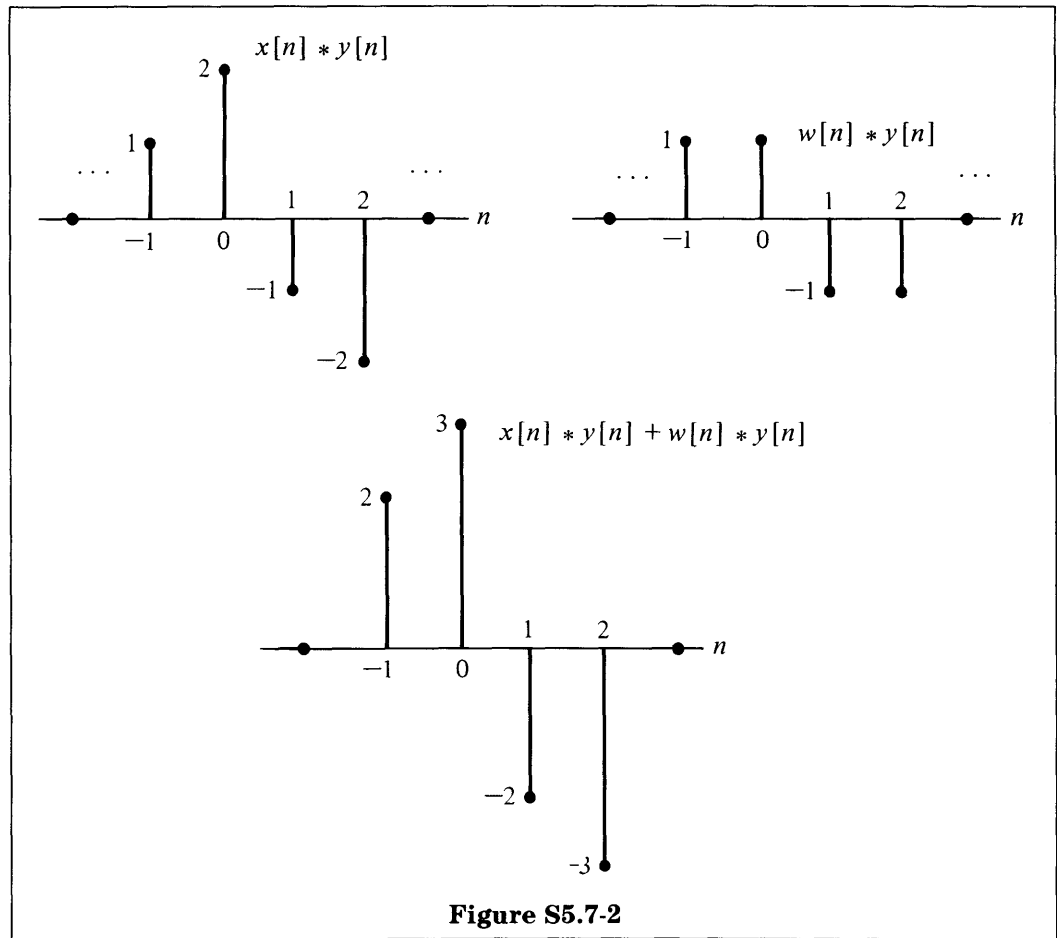


Figure S5.7-2

Therefore, the distributive property $(x + w) * y = x * y + w * y$ is verified.

(b) Figure S5.7-3 shows the required convolutions and multiplications.

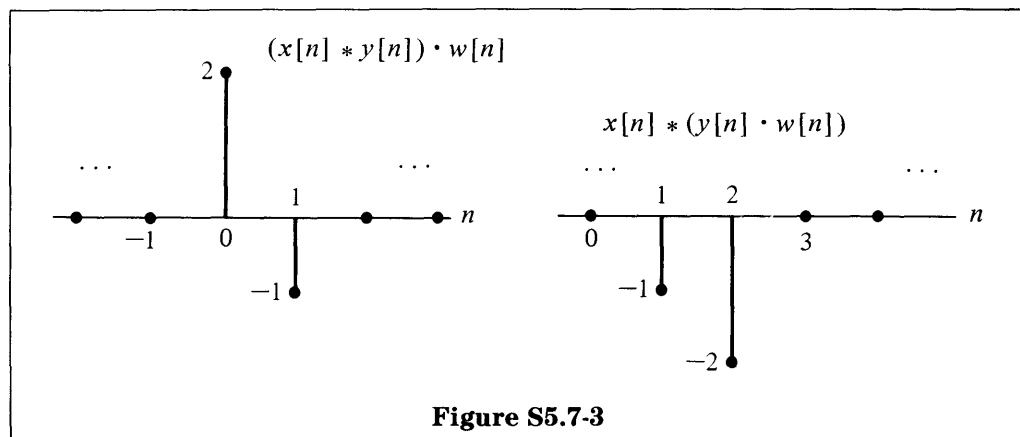


Figure S5.7-3

Note, therefore, that $(x[n] * y[n]) * w[n] \neq x[n] * (y[n] * w[n])$.

S5.8

Consider

$$\begin{aligned} y(t) &= x(t) * h(t) = \int_{-\infty}^{\infty} x(t - \tau)h(\tau) d\tau \\ &= \int_{-\infty}^{\infty} x(\tau)h(t - \tau) d\tau \end{aligned}$$

$$\begin{aligned} \text{(a)} \quad y'(t) &= \int_{-\infty}^{\infty} x'(t - \tau)h(\tau) d\tau = x'(t) * h(t) \\ &= \int_{-\infty}^{\infty} x(\tau)h'(t - \tau) d\tau = x(t) * h'(t), \end{aligned}$$

where the primes denote d/dt .

$$\begin{aligned} \text{(b)} \quad y(t) &= x(t) * h(t), \\ y(t) &= x(t) * u_{-1}(t) * u_1(t) * h(t), \\ y(t) &= \int_{-\infty}^t x(\tau) d\tau * h'(t) \end{aligned}$$

$$\begin{aligned} \text{(c)} \quad y(t) &= x(t) * h(t), \\ y(t) &= x(t) * u_1(t) * h(t) * u_{-1}(t), \\ y(t) &= \int_{-\infty}^t x'(\tau) * h(\tau) d\tau \end{aligned}$$

$$\begin{aligned} \text{(d)} \quad y(t) &= x(t) * h(t) \\ &= x(t) * u_1(t) * h(t) * u_{-1}(t), \\ y(t) &= x'(t) * \int_{-\infty}^t h(\tau) d\tau \end{aligned}$$

S5.9

(a) True.

$$\int_{-\infty}^{\infty} |h(t)| dt = \sum_{k=-\infty}^{\infty} \int_0^T |h(t)| dt = \infty$$

(b) False. If $h(t) = \delta(t - t_0)$ for $t_0 > 0$, then the inverse system impulse response is $\delta(t + t_0)$, which is noncausal.

(c) False. Suppose $h[n] = u[n]$. Then

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} u[n] = \infty$$

(d) True, assuming $h[n]$ is finite-amplitude.

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-K}^L |h[n]| = M \quad (\text{a number})$$

(e) False. $h(t) = u(t)$ implies causality, but $\int_{-\infty}^{\infty} u(t) dt = \infty$ implies that the system is not stable.

(f) False.

$$\begin{aligned} h_1(t) &= \delta(t - t_1), & t_1 > 0 & \quad \text{Causal} \\ h_2(t) &= \delta(t + t_2), & t_2 > 0 & \quad \text{Noncausal} \\ h(t) &= h_1(t) * h_2(t) = \delta(t + t_2 - t_1), & t_2 \leq t_1 & \quad \text{Causal} \end{aligned}$$

(g) False. Suppose $h(t) = e^{-t}u(t)$. Then

$$\int_{-\infty}^{\infty} e^{-t}u(t) dt = -e^{-t} \Big|_0^{\infty} = 1 \quad \text{Stable}$$

The step response is

$$\begin{aligned} \int_{-\infty}^{\infty} u(t - \tau)e^{-\tau}u(\tau) d\tau &= \int_0^t e^{-\tau} d\tau \\ &= (1 - e^{-t})u(t), \\ \int_0^{\infty} (1 - e^{-t}) dt &= t + e^{-t} \Big|_0^{\infty} = \infty \end{aligned}$$

(h) True. We know that $u[n] = \sum_{k=0}^{\infty} \delta[n - k]$ and, from superposition, $s[n] = \sum_{k=0}^{\infty} h[n - k]$. If $s[n] \neq 0$ for some $n < 0$, there exists some value of $h[k] \neq 0$ for some $k < 0$. If $s[n] = 0$ for all $n < 0$, $h[k] = 0$ for all $k < 0$.

S5.10

$$\begin{aligned} \text{(a)} \quad \int_{-\infty}^{\infty} g(\tau)u_1(\tau) d\tau &= -g'(0), \\ g(\tau) &= x(t - \tau), \quad t \text{ fixed}, \\ \int_{-\infty}^{\infty} x(t - \tau)u_1(\tau) d\tau &= -\frac{dg(\tau)}{d\tau} \Big|_{\tau=0} = -\frac{dx(t - \tau)}{d\tau} \Big|_{\tau=0} \\ &= \frac{dx(t - \tau)}{dt} \Big|_{\tau=0} = \frac{dx(t)}{dt} \end{aligned}$$

$$\begin{aligned} \text{(b)} \quad \int_{-\infty}^{\infty} g(t)f(t)u_1(t) dt &= -\frac{d}{dt}[g(t)f(t)] \Big|_{t=0} \\ &= -[g'(t)f(t) + g(t)f'(t)] \Big|_{t=0} \\ &= -[g'(0)f(0) + g(0)f'(0)], \end{aligned}$$

$$\int g(t)[f(0)u_1(t) - f'(0)\delta(t)] dt = -f(0)g'(0) - f'(0)g(0)$$

So when we use a test function $g(t)$, $f(t)u_1(t)$ and $f(0)u_1(t) - f'(0)\delta(t)$ both produce the same operational effect.

$$\begin{aligned} \text{(c)} \quad \int_{-\infty}^{\infty} x(\tau)u_2(\tau) d\tau &= x(\tau)u_1(\tau) \Big|_{-\infty}^{\infty} - \int_{-\infty}^{\infty} \frac{dx}{d\tau} u_1(\tau) d\tau \\ &= -\int_{-\infty}^{\infty} \frac{dx}{d\tau} u_1(\tau) d\tau = -\frac{dx}{d\tau} u_1(\tau) \Big|_{-\infty}^{\infty} + \int_{-\infty}^{\infty} \frac{d^2x}{d\tau^2} u_1(\tau) d\tau \\ &= \frac{d^2x}{d\tau^2} \Big|_{\tau=0} \end{aligned}$$

$$\text{(d)} \quad \int g(\tau)f(\tau)u_2(\tau) d\tau = g''(\tau)f(\tau) + 2g'(\tau)f'(\tau) + g(\tau)f''(\tau) \Big|_{\tau=0}$$

Noting that $2g'(\tau)f'(\tau)|_{\tau=0} = -2f'(0)\int g(\tau)u_1(\tau) d\tau$, we have an equivalent operational definition:

$$f(\tau)u_2(\tau) = f(0)u_2(\tau) - 2f'(0)u_1(\tau) + f''(0)\delta(\tau)$$

S5.11

- (a) $h(t) * g(t) = \int_{-\infty}^{\infty} h(t - \tau)g(\tau) d\tau = \int_0^t h(t - \tau)g(\tau) d\tau$ since $h(t) = 0$ for $t < 0$ and $g(t) = 0$ for $t < 0$. But if $t < 0$, this integral is obviously zero. Therefore, the cascaded system is causal.
- (b) By the definition of stability we know that for any bounded input to H, the output of H is also bounded. This output is also the input to system G. Since the input to G is bounded and G is stable, the output of G is bounded. Therefore, a bounded input to the cascaded system produces a bounded output. Hence, this system is stable.

S5.12

We have a total system response of

$$\begin{aligned} h &= \{[(h_1 * h_2) + (h_2 * h_1) - (h_2 * h_1)] * h_1 + h_1^{-1}\} * h_2^{-1} \\ h &= (h_2 * h_1) + (h_1^{-1} * h_2^{-1}) \end{aligned}$$

S5.13

We are given that $y[n] = x[n] * h[n]$.

$$\begin{aligned} y[n] &= \sum_{k=-\infty}^{\infty} x[n-k]h[k] \\ |y[n]| &= \left| \sum_{k=-\infty}^{\infty} x[n-k]h[k] \right| \\ \max\{|y[n]|\} &= \max \left\{ \left| \sum_{k=-\infty}^{\infty} x[n-k]h[k] \right| \right\} \\ &\leq \max \sum_{k=-\infty}^{\infty} |x[n-k]| |h[k]| \\ &\leq \sum_{k=-\infty}^{\infty} \max\{|x[n-k]|\} |h[k]| \\ &= \max\{|x[n]|\} \sum_{k=-\infty}^{\infty} |h[k]| \end{aligned}$$

We can see from the inequality

$$\max\{|y[n]|\} \leq \max\{|x[n]|\} \sum_{k=-\infty}^{\infty} |h[k]|$$

that $\sum_{k=-\infty}^{\infty} |h[k]| \leq 1 \Rightarrow \max\{|y[n]|\} \leq \max\{|x[n]|\}$. This means that $\sum_{k=-\infty}^{\infty} |h[k]| \leq 1$ is a sufficient condition. It is necessary because some $x[n]$ always exists that yields $y[n] = \sum_{k=-\infty}^{\infty} |h[k]|$. ($x[n]$ consists of a sequence of +1's and -1's.) Therefore, since $\max\{x[n]\} = 1$, it is necessary that $\sum_{k=-\infty}^{\infty} |h[k]| \leq 1$ to ensure that $y[n] \leq \max\{|x[n]|\} = 1$.

MIT OpenCourseWare
<http://ocw.mit.edu>

Resource: Signals and Systems
Professor Alan V. Oppenheim

The following may not correspond to a particular course on MIT OpenCourseWare, but has been provided by the author as an individual learning resource.

For information about citing these materials or our Terms of Use, visit: <http://ocw.mit.edu/terms>.