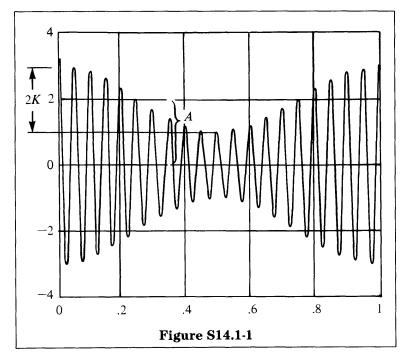
14 Demonstration of Amplitude Modulation

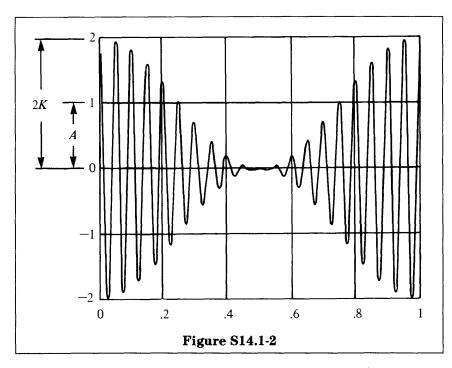
Solutions to Recommended Problems

S14.1

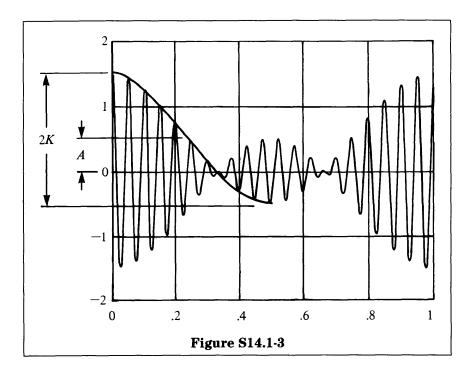
(a) We see in Figure S14.1-1 that the modulating cosine wave has a peak amplitude of 2K = 2, so that K = 1. At the point in time when the modulating cosine wave is zero, the total signal is A = 2, so K/A = 0.5. Therefore, the signal has 50% modulation. See Figure S14.1-1.



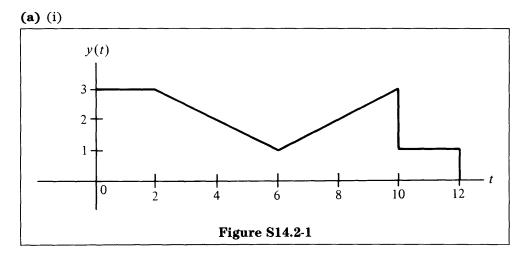
(b) 2K = 2, K = 1, A = 1, so K/A = 1, and the signal has 100% modulation. See Figure S14.1-2.

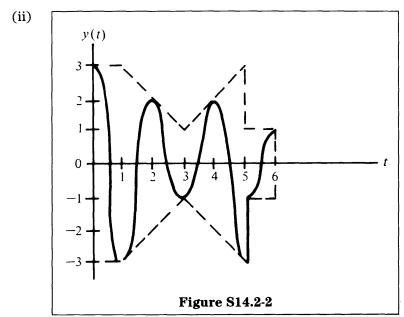


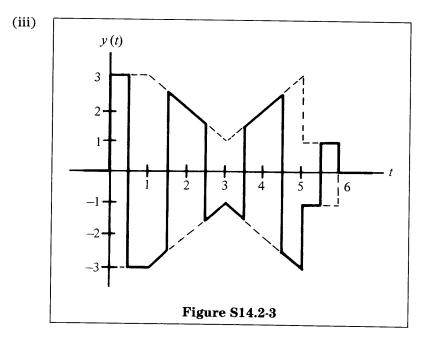
(c) 2K = 2, K = 1, A = 0.5, so K/A = 2, and the signal has 200% modulation.



S14.2







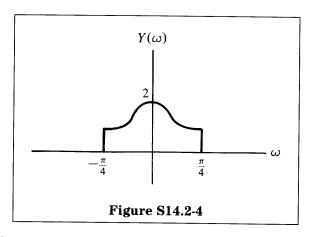
(b) (i)
$$Y(\omega) = \int_{-\infty}^{\infty} y(t)e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} x\left(\frac{t}{2}\right)e^{-j\omega t} dt, \qquad t' = \frac{t}{2}, \qquad dt' = \frac{1}{2} dt$$

$$= \int_{-\infty}^{\infty} x(t')e^{-j\omega 2t'} 2 dt'$$

$$= 2X(2\omega)$$

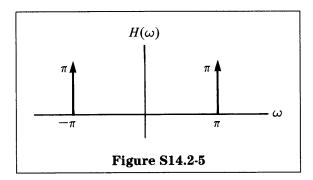
Therefore, $Y(\omega)$ is a compressed version of $X(\omega)$. See Figure S14.2-4.

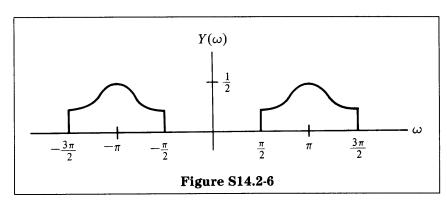


(ii) From the convolution theorem,

$$Y(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\Omega) H(\omega - \Omega) \ d\Omega,$$

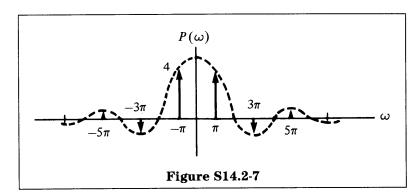
where $\cos \pi t \stackrel{\mathcal{F}}{\longleftrightarrow} H(\omega)$, and $H(\omega)$ is as shown in Figure S14.2-5. Therefore, $Y(\omega)$ is as given in Figure S14.2-6.



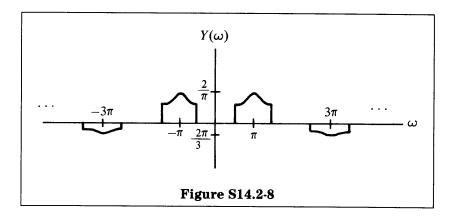


(iii)
$$Y(\omega) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(\Omega) P(\omega - \Omega) d\Omega$$

 $P(\omega)$ is an impulsive spectrum, as shown in Figure S14.2-7, because the corresponding p(t) is periodic. (Note that only odd harmonics are present.)



Therefore $Y(\omega)$ is as shown in Figure S14.2-8.



S14.3

- (a) ii
- **(b)** i
- (c) iii
- (d) vi
- **(e)** v
- **(f)** iv
- **(g)** vii
- (h) x
- (i) ix
- (j) viii

S14.4

(a) We are considering

$$X(\Omega) = \sum_{n=0}^{N-1} x[n]e^{-j\Omega n},$$

which is effectively the Fourier transform of a signal of infinite duration multiplied by a window of length N:

$$X(\Omega) = \sum_{n=-\infty}^{\infty} \cos \omega_0 n T(u[n] - u[n-N]) e^{-j\Omega n}$$

From the convolution theorem we can compute the Fourier transform of the product of these two sequences:

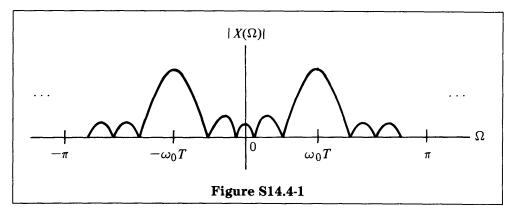
$$\cos \omega_0 nT \xrightarrow{\mathcal{F}} \pi[\delta(\Omega - \omega_0 T) + \delta(\Omega + \omega_0 T)], \quad -\pi < \Omega < \pi$$

$$u[n] - u[n - N] \xrightarrow{\mathcal{F}} \frac{1 - e^{-j\Omega N}}{1 - e^{-j\Omega}} = e^{-j\Omega(N-1)/2} \frac{\sin N\Omega/2}{\sin \Omega/2}$$

Therefore,

$$X(\Omega) = \frac{1}{2} e^{-j(\Omega - \omega_0 T)(N-1)/2} \frac{\sin[N(\Omega - \omega_0 T)/2]}{\sin[(\Omega - \omega_0 T)/2]} + \frac{1}{2} e^{-j(\Omega + \omega_0 T)(N-1)/2} \frac{\sin[N(\Omega + \omega_0 T)/2]}{\sin[(\Omega + \omega_0 T)/2]},$$

as shown in Figure S14.4-1. (Note that the spectrum is periodic with period 2π .)



(b)
$$X(\Omega_k) = \sum_{n=0}^{N-1} x[n]e^{-j\Omega_k n}$$

$$X\left(\frac{2\pi k}{N}\right) = \sum_{n=0}^{N-1} \cos \omega_0 n T e^{-j(2\pi k/N)n}$$

$$= \sum_{n=0}^{N-1} \frac{1}{2} e^{j\omega_0 n T} e^{-j(2\pi k/N)n} + \sum_{n=0}^{N-1} \frac{1}{2} e^{-j\omega_0 n T} e^{-j(2\pi k/N)n}$$

$$= \frac{1}{2} \left(\frac{1 - e^{j(\omega_0 T - 2\pi k/N)N}}{1 - e^{j(\omega_0 T - 2\pi k/N)}}\right) + \frac{1}{2} \frac{1 - e^{j(-\omega_0 T - 2\pi k/N)N}}{1 - e^{j(-\omega_0 T - 2\pi k/N)}}$$

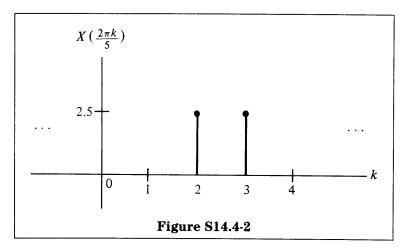
(i) For $\omega_0 T = 2\pi(\frac{2}{5})$ and N = 5, the first term is zero for

$$k = \ldots -3, 2, 7, \ldots$$

However, when k = 2 we have the ratio of

$$\frac{1}{2} \left(\frac{1 - e^{j2\pi(2/5 - k/5)5}}{1 - e^{j2\pi(2/5 - k/5)}} \right) = \frac{0}{0}$$

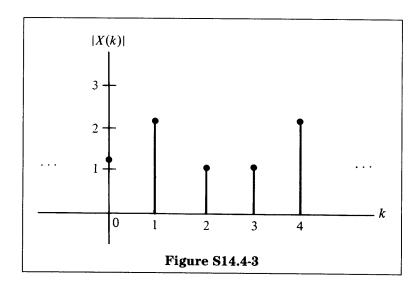
and we treat the limit as $k \to 0$. Using L'Hôpital's rule, we have $\frac{1}{2}(5) = 2.5$. Similarly, the second term is zero except when $k = \ldots -2, 3, 8, \ldots$. Taking the limit yields 2.5. So $X(2\pi k/5)$ is as shown in Figure S14.4-2.



Note that $X(2\pi k/5)$ is periodic in k with period 5 since $X(\Omega)$ is periodic in Ω with period 2π .

$$(ii) \quad X \bigg(\frac{2\pi k}{N} \bigg) = \frac{1}{2} \left(\frac{1 - e^{j(\omega_0 T - 2\pi k/N)N}}{1 - e^{j(\omega_0 T - 2\pi k/N)}} \right) + \frac{1}{2} \left(\frac{1 - e^{j(-\omega_0 T - 2\pi k/N)N}}{1 - e^{j(-\omega_0 T - 2\pi k/N)}} \right)$$

Now $\omega_0 T = 2\pi \frac{3}{10}$, and the numerator and denominator are nonzero for all k. Evaluating the preceding expression yields X(k) as shown in Figure S14.4-3.



MIT OpenCourseWare http://ocw.mit.edu

Resource: Signals and Systems Professor Alan V. Oppenheim

The following may not correspond to a particular course on MIT OpenCourseWare, but has been provided by the author as an individual learning resource.

For information about citing these materials or our Terms of Use, visit: http://ocw.mit.edu/terms.