Complex-Domain Analysis of Discrete-Time Signals using the *z* Transform

(Chapter 4, Schaum's Outline of Signals and Systems)

Dr. Jason S. Skinner

November 12, 2014

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [2 of 65]

The z Transform: Bilateral Definition

For a general discrete-time signal x[n], the **bilateral** (or **two-sided**) z transform X(z) is defined as

$$X(z) = \sum_{n = -\infty}^{\infty} x[n]z^{-n}.$$
 (2)

The variable z is complex, in general, and is typically expressed in polar/exponential form as

$$z = re^{j\Omega}$$

where r is the magnitude of z and  $\Omega$  is the angle of z.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [1 of 65]

#### The z Transform: Introduction

Consider a discrete-time LTI system with an input  $x[n] = z^n$ , where z is a complex variable. Therefore,

$$x[n] = z^n \longrightarrow \mathsf{LTI} \; \mathsf{System} \longrightarrow y[n],$$

where the output is given by

$$y[n] = \sum_{k=-\infty}^{\infty} h[k]z^{n-k} = z^n \sum_{k=-\infty}^{\infty} h[k]z^{-k} = H(z)z^n.$$

where the function H(z) is referred to as the z transform of h[n] and is given by

$$H(z) = \sum_{k=-\infty}^{\infty} h[k]z^{-k}.$$
 (1)

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [3 of 65]

#### The z Transform: Region of Convergence

The range of values of the complex variables z for which the z transform converges is called the **region of convergence (ROC)**.

The **ROC** is illustrated in the following examples.

Complex-Domain Analysis of Discrete-Time Signals [4 of 65]

The z Transform: ROC Example 1

Determine the z transform and region of convergence for the signal  $% \left( z\right) =\left( z\right) +\left( z\right)$ 

$$x[n] = a^n u[n]$$
 for  $a$  real.

The z Transform: ROC Example 1

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [6 of 65]

The z Transform: ROC Example 2

Determine the z transform and region of convergence for the signal

$$x[n] = -a^n u[-n-1]$$
 for  $a$  real.

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [7 of 65]

Complex-Domain Analysis of Discrete-Time Signals [5 of 65]

The z Transform: ROC Example 2

Complex-Domain Analysis of Discrete-Time Signals [9 of 65]

The z Transform: Poles and Zeros of X(z)

The function X(z) will typically be a rational function in z. That is,

$$X(z) = \frac{b_0 z^m + b_1 z^{m-1} + \dots + b_m}{a_0 z^n + a_1 z^{n-1} + \dots + a_n},$$

where m and n are positive integers and the coefficients  $a_k$  and  $b_k$  are real constants.

X(z) is a called a **proper** rational function if n>m and an **improper** rational function if  $n\leq m$ .

The z Transform: Poles and Zeros of X(z)

A rational function X(z) can also be written as

$$X(z) = \frac{b_0}{a_0} \left[ \frac{(z-z_1)\cdots(z-z_m)}{(z-p_1)\cdots(z-p_n)} \right].$$

The roots,  $z_k$ , of the numerator polynomial are called the **zeros** of X(z) because X(z)=0 for those values of z.

The roots,  $p_k$ , of the denominator polynomial are called the **poles** of X(z) because  $X(z) \to \infty$  for those values of z.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [10 of 65]

The z Transform: Poles and Zeros of X(z)

The poles of X(z) lie outside the ROC since X(z) does not converge at the poles (by definition). Traditionally, an " $\times$ " is used to indicate each pole location in the complex z-plane.

The zeros may lie inside or outside the ROC. Traditionally, an " $\circ$ " is used to indicate each zero location in the complex z-plane.

Except for the scale factor  $\left(\frac{b_0}{a_0}\right)$ , X(z) is completely specified by its poles and zeros.

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [11 of 65]

The z Transform: Poles and Zeros of X(z)

A compact representation of X(z) in the complex z-plane is to show the location of poles and zeros in addition to the ROC.

This is illustrated in the following example.

Complex-Domain Analysis of Discrete-Time Signals [12 of 65]

Complex-Domain Analysis of Discrete-Time Signals [13 of 65]

#### The z Transform: Poles and Zeros - Example

Determine the z transform X(z) and region of convergence for the signal

$$x[n] = \left\lceil \left(\frac{1}{2}\right)^n + \left(\frac{1}{3}\right)^n \right\rceil u[n].$$

Plot the poles, zeros, and ROC for X(z).

The z Transform: Poles and Zeros - Example

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [14 of 65]

### The z Transform: Properties of the ROC

The ROC of X(z) depends on the nature of x[n]. If we assume that X(z) is a rational function of z, then there are five interesting properties of the ROC.

**Property 1:** The ROC does not contain any poles.

**Property 2:** If x[n] is a **finite-duration** sequence (x[n] = 0 except for a finite interval  $N_1 \le n \le N_2$ ) and X(z) converges for some value of z, then the ROC is the entire z-plane (except possibly z = 0).

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [15 of 65]

**Property 3:** If x[n] is a **right-sided** sequence  $(x[n] = 0 \text{ for } n < N_1 < \infty)$  and X(z) converges for some value of z, then the ROC is of the form

$$|z| > r_{\text{max}}$$

where  $r_{\text{max}}$  equals the largest magnitude of any of the poles of X(z).

Therefore, the ROC is the exterior of the circle

$$|z| = r_{\text{max}}$$

in the z-plane.

Complex-Domain Analysis of Discrete-Time Signals [17 of 65]

**Property 4:** If x[n] is a **left-sided** sequence  $(x[n] = 0 \text{ for } n > N_2 > -\infty)$  and X(z) converges for some value of z, then the ROC is of the form

$$|z| < r_{\min}$$
 or  $0 < |z| < r_{\min}$ 

where  $r_{\min}$  equals the smallest magnitude of any of the poles of X(z).

Therefore, the ROC is the interior of the circle

$$|z| = r_{\min}$$

in the z-plane with the possible exception of z = 0.

**Property 5:** If x[n] is a **two-sided** sequence (x[n] is an infinite-duration sequence that is neither right-sided nor left-sided) and X(z) converges for some value of z, then the ROC is of the form

$$r_1 < |z| < r_2$$

where  $r_1$  and  $r_2$  are the magnitudes of the two poles of X(z).

Therefore, the ROC is an annular ring in the z-plane between the circles

$$|z|=r_1$$
 and  $|z|=r_2$ 

not containing any poles.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [18 of 65]

The z Transform: Unilateral Definition

The unilateral (or one-sided) z transform is defined as

$$X(z) = \sum_{n=0}^{\infty} x[n]z^{-n}.$$
 (3)

Note that the bilateral and unilateral z transforms are equivalent if x[n]=0 for n<0, and the unilateral z transform ignores x[n] for n<0.

In other words, the unilateral z transform is sufficient for **right-sided** signals (typical real-world signals).

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [19 of 65]

#### The *z* Transform: Unilateral Definition

The unilateral z transform is useful for calculating the complete response of an LTI system to a causal input (typical real-world signal). It can take into account both the zero-state response of the **causal** system due to a **causal** input, as well as the zero-input response due to **nonzero** initial conditions.

From this point forward, we will ONLY use the **unilateral** z transform. All references henceforth to the z transform refer only to the **unilateral** z transform.

#### The z Transform: Representation

Equation 3 is sometimes considered an operator that transforms a signal x[n] into a function X(z) symbolically represented by

$$X(z) = \mathcal{Z}\left\{x[n]\right\},\,$$

and the signal x[n] and its z transform X(z) are said to form a z-transform pair denoted as

$$x[n] \longleftrightarrow X(z).$$

See Table 4-1 (page 153 of Schaum's Outline of Signals and Systems) for a listing of z transform pairs.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [22 of 65]

#### z-Transform Pairs: Unit Step Function u[n]

The z transform of a unit step function is given by

$$\begin{split} \mathcal{Z}\left\{u[n]\right\} &= \sum_{n=0}^{\infty} u[n] z^{-n} = \sum_{n=0}^{\infty} z^{-n} \\ &= \frac{1}{1-z^{-1}} = \frac{z}{z-1} \text{ for } |z| > 1. \end{split}$$

Therefore, the z-transform pair for a unit step function is

$$u[n] \longleftrightarrow \frac{z}{z-1}$$
 with ROC =  $|z| > 1$ .

ELEC 309: Signals and Systems

### z-Transform Pairs: Unit Impulse Function $\delta[n]$

The z transform of a unit impulse function is given by

$$\mathcal{Z}\left\{\delta[n]\right\} = \sum_{n=0}^{\infty} \delta[n] z^{-n} = z^{-0} = 1 \text{ for all } z.$$

Therefore, the z-transform pair for a unit impulse function is

$$\delta[n] \longleftrightarrow 1$$
 with ROC = all  $z$ .

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [23 of 65]

#### z-Transform Pairs: Other Common Signals

Table 1. Some z-Transform Pairs

x[n]	X(z)	ROC
$\delta[n]$	1	All $z$
$\delta[n-m]$	$z^{-m}$	All $z$ except $0$ if $m>0$ or $\infty$ if $m<0$
u[n]	$\frac{1}{1-z^{-1}}$ or $\frac{z}{z-1}$	z  > 1
$a^nu[n]$	$\frac{1}{1-az^{-1}}$ or $\frac{z}{z-a}$	z  >  a
$na^nu[n]$	$\frac{az^{-1}}{\left(1-az^{-1}\right)^2} \text{ or } \frac{az}{(z-a)^2}$	z  >  a

Complex-Domain Analysis of Discrete-Time Signals [24 of 65]

#### z-Transform Pairs: Other Common Signals

Table 2: Some z-Transform Pairs

x[n]	X(z)	ROC
$\cos\left(\Omega_0 n\right) u[n]$	$\frac{z^2 - \cos(\Omega_0)z}{z^2 - 2\cos(\Omega_0)z + 1}$	z  > 1
$\sin\left(\Omega_0 n\right) u[n]$	$\frac{\sin(\Omega_0)z}{z^2 - 2\cos(\Omega_0)z + 1}$	z  > 1
$r^n \cos(\Omega_0 n) u[n]$	$\frac{z^2 - r\cos(\Omega_0)z}{z^2 - 2r\cos(\Omega_0)z + r^2}$	z  > r
$r^n \sin\left(\Omega_0 n\right) u[n]$	$\frac{r\sin(\Omega_0)z}{z^2 - 2r\cos(\Omega_0)z + r^2}$	z  > r
$\begin{cases} a^n & 0 \le n \le N - 1 \\ 0 & \text{otherwise} \end{cases}$	$\frac{1-a^Nz^{-N}}{1-az^{-1}}$	z  > 0

#### z Transform Pairs: Example

Determine the z transform of

$$x[n] = \left(\frac{1}{2}\right)^n \cos(2n) u[n].$$

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [26 of 65]

#### z Transform Pairs: Example

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [27 of 65]

Complex-Domain Analysis of Discrete-Time Signals [25 of 65]

#### **Properties of the** *z* **Transform: Linearity**

lf

$$x_1[n] \longleftrightarrow X_1(z)$$
 with ROC  $= R_1$  and  $x_2[n] \longleftrightarrow X_2(z)$  with ROC  $= R_2$ ,

then

$$a_1x_1[n] + a_2x_2[n] \longleftrightarrow a_1X_1(z) + a_2X_2(z)$$

with ROC=  $R' \supset R_1 \cap R_2$ .

(The ROC of the resultant z transform is at least as large as the region in common between  $R_1$  and  $R_2$ . Usually, the ROC will simply be  $R' = R_1 \cap R_2$ .)

Complex-Domain Analysis of Discrete-Time Signals [28 of 65]

#### ELEC 309: Signals and Systems

#### **Properties of the** *z* **Transform: Time Shifting**

If 
$$x[n] \longleftrightarrow X(z)$$
 with ROC=  $R$ , then for  $m \ge 0$ ,

$$x[n-m] \longleftrightarrow z^{-m}X(z) + z^{-m+1}x[-1] + z^{-m+2}x[-2] + \dots + x[-m]$$
  
 $\longleftrightarrow z^{-m}X(z) + z^{-m}\sum_{i=1}^{m} z^{i}x[-i]$ 

and

$$x[n+m] \longleftrightarrow z^m X(z) - z^m x[0] - z^{m-1} x[1] - \dots - zx[m-1]$$

$$\longleftrightarrow z^m X(z) - z^m \sum_{i=0}^{m-1} z^{-i} x[i]$$

with ROC=  $R' = R \cap \{0 < |z| < \infty\}.$ 

(The ROC of the resultant z transform is the original ROC minus z=0 and/or infinite values of z.)

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [30 of 65]

Properties of the z Transform: Multiplication by  $z_0^n$ 

lf

$$x[n] \longleftrightarrow X(z)$$
 with ROC =  $R$ ,

then

$$z_0^n x[n] \longleftrightarrow X\left(\frac{z}{z_0}\right)$$

with ROC=  $R' = |z_0|R$ .

(The ROC of the resultant z transform expands or contracts by the factor  $|z_0|$ . A pole at  $z=p_k$  or a zero at  $z=z_k$  moves to  $z=z_0p_k$  or  $z=z_0z_k$ , respectively.)

# Properties of the z Transform: Time Shifting Special Cases

$$x[n-1]u[n]\longleftrightarrow z^{-1}X(z)+x[-1] \text{ with ROC}=R'=R\cap\{0<|z|\}$$
 
$$x[n+1]u[n]\longleftrightarrow zX(z)-zx[0] \text{ with ROC}=R'=R\cap\{|z|<\infty\}$$

For z transform:

- $\bullet$   $z^{-1}$  often called **unit-delay operator**
- z often called unit-advance operator

For Laplace transform:

- $\bullet$   $s^{-1}$  corresponds to time-domain integration
- s corresponds to time-domain differentiation

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [31 of 65]

## Properties of the z Transform: Multiplication by $z_0^n$ Special Case

$$e^{j\Omega_0 n}x[n]\longleftrightarrow X\left(e^{-j\Omega_0}z\right)$$

with ROC= R' = R.

(All poles and zeros are simply rotated by the angle  $\Omega$ , and the ROC of the resultant z transform is unchanged.)

#### Properties of the z Transform: Time Reversal

lf

$$x[n] \longleftrightarrow X(z)$$
 with ROC =  $R$ ,

then

$$x[-n] \longleftrightarrow X\left(\frac{1}{z}\right)$$

with ROC=  $R' = \frac{1}{R}$ .

(The ROC of the resultant z transform is the inversion of the original ROC, indicating the fact that a right-sided sequence becomes left-sided if time-reversed, and vice versa. A pole at  $z=p_k$  or a zero at  $z=z_k$  moves to  $z=1/p_k$  or  $z=1/z_k$ , respectively, after time reversal.)

ELEC 309: Signals and Systems

### Properties of the z Transform: Multiplication by n or Differentiation in the z-Domain

lf

$$x[n] \longleftrightarrow X(z)$$
 with ROC =  $R$ ,

then

$$nx[n] \longleftrightarrow -z \frac{dX(z)}{dz}$$

with ROC= R' = R.

(The ROC of the resultant z transform is the original ROC.)

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [34 of 65]

#### **Properties of the** *z* **Transform: Accumulation**

If  $x[n] \longleftrightarrow X(z)$  with ROC = R, then

$$\sum_{k=0}^{n} x[k] \longleftrightarrow \frac{1}{1-z^{-1}} X(z) = \frac{z}{z-1} X(z) \text{ and}$$

$$\sum_{k=-\infty}^{n} x[k] \longleftrightarrow \frac{1}{1-z^{-1}} X(z) = \frac{z}{z-1} X(z) + \frac{z}{z-1} \sum_{k=-\infty}^{0} x[k]$$

with ROC=  $R' \supset R \cap \{|z| > 1\}$ .

(Note that  $\sum\limits_{k=-\infty}^n x[k]$  is the discrete-time counterpart to integration in the discrete-time domain and is called accumulation.)

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [35 of 65]

#### **Properties of the** *z* **Transform: Convolution**

lf

$$x_1[n] \longleftrightarrow X_1(z)$$
 with ROC  $= R_1$  and  $x_2[n] \longleftrightarrow X_2(z)$  with ROC  $= R_2$ ,

then

$$x_1[n] * x_2[n] \longleftrightarrow X_1(z)X_2(z)$$

with ROC=  $R' \supset R_1 \cap R_2$ .

(The ROC of the resultant z transform is at least as large as the region in common between  $R_1$  and  $R_2$ . Usually, the ROC will simply be  $R'=R_1\cap R_2$ .)

Complex-Domain Analysis of Discrete-Time Signals [36 of 65]

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [37 of 65]

#### Properties of the z Transform: Example 1

Find the z transform of

$$x[n] = u[n-5].$$

Properties of the z Transform: Example 1

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [38 of 65]

### Properties of the z Transform: Example 2

Find the z transform of

$$x[n] = \left(\frac{1}{2}\right)^n (u[n] - u[n-5]).$$

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [39 of 65]

Properties of the z Transform: Example 2

#### The Inverse z Transform: Definition

The **inverse** z **transform** is the inversion of the ztransform to determine the signal x[n] from its z transform X(z).

It is symbolically denoted as

$$x[n] = \mathcal{Z}^{-1} \left\{ X(z) \right\}.$$

The Inverse z Transform: Contour-Integral Formula

The **inverse** z **transform** can be written formally as as the evaluation of an integral in the complex z-plane of the form

$$x[n] = \frac{1}{2\pi j} \oint_C X(z) z^{n-1} dz.$$

where C is a counterclockwise contour of integration enclosing the origin.

The evaluation of the inverse z transform integral requires an understanding of complex variable theory.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [42 of 65]

#### The Inverse z Transform: Use of Tables

A simpler method to determine the **inverse** *z* **transform** is to express X(z) as a sum

$$X(z) = X_1(z) + \cdots + X_m(z),$$

where  $X_1(z), \ldots, X_m(z)$  are functions with known inverse z transforms  $x_1[n], \ldots, x_m[n]$  (given in tables of z transforms).

From the linearity property, it follows that

$$x[n] = x_1[n] + \dots + x_m[n].$$

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [43 of 65]

### The Inverse z Transform: Use of Tables Example 1

Find the inverse z transform of

$$X(z) = \frac{z}{z - \frac{1}{2}} \text{ with ROC} = |z| > \frac{1}{2}.$$

Complex-Domain Analysis of Discrete-Time Signals [44 of 65]

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [45 of 65]

# The Inverse *z* Transform: Use of Tables Example 2

Find the inverse z transform of

$$X(z) = z^{-2}$$
 with ROC =  $|z| > 0$ .

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [46 of 65]

The Inverse z Transform: Use of Tables Example 2

The Inverse z Transform: Use of Tables

Example 1

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [47 of 65]

# The Inverse z Transform: Use of Tables Example 3

Find the inverse z transform of

$$X(z) = rac{1 - rac{1}{(5z)^5}}{1 - 0.2z^{-1}}$$
 with ROC  $= |z| > 0$ .

Complex-Domain Analysis of Discrete-Time Signals [48 of 65]

Complex-Domain Analysis of Discrete-Time Signals [49 of 65]

### The Inverse *z* Transform: Use of Tables Example 3

#### The Inverse *z* Transform: Partial-Fraction Expansion

If X(z) is of the form

ELEC 309: Signals and Systems

$$X(z) = \frac{N(z)}{D(z)} = k \frac{(z - z_1) \cdots (z - z_m)}{(z - p_1) \cdots (z - p_n)},$$

then X(z) is a rational function and a simple technique involving partial-fraction expansion can be used for the inversion of X(z).

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [50 of 65]

### The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case for $m \le n$

If  $n \geq m$  and all the poles  $(p_1, \ldots, p_n)$  of X(z) are distinct (all roots of D(z) are different), then

$$\frac{X(z)}{z} = \frac{c_0}{z} + \frac{c_1}{z - p_1} + \frac{c_2}{z - p_2} + \dots + \frac{c_n}{z - p_n}$$
$$= \frac{c_0}{z} + \sum_{k=1}^{n} \frac{c_k}{z - p_k}$$

where the coefficients are given by

$$c_0 = X(z)igg|_{z=0}$$
 and  $c_k = (z-p_k)rac{X(z)}{z}igg|_{z=p_k}$ .

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [51 of 65]

### The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case for $m \le n$

Therefore, we have

$$X(z) = c_0 + c_1 \frac{z}{z - p_1} + c_2 \frac{z}{z - p_2} + \dots + c_n \frac{z}{z - p_n}$$
 (4)

$$= c_0 + \sum_{k=1}^n c_k \frac{z}{z - p_k}$$
 (5)

We can then use our tables of z transforms and infer the ROC for each term from the overall ROC of X(z) to determine the overall inverse z-transform.

Complex-Domain Analysis of Discrete-Time Signals [52 of 65]

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [53 of 65]

# The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case Example 1

Find the inverse z transform of

$$X(z) = \frac{z}{2z^2 - 3z + 1} \text{ with ROC} = |z| > 1$$

using the partial-fraction expansion technique.

The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case Example 1

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [54 of 65]

## The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case Example 2

Find the inverse z transform of

$$X(z) = \frac{3}{z-2}$$
 with ROC =  $|z| > 2$ 

using the partial-fraction expansion technique.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [55 of 65]

The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case Example 2

### The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case for m>n

If m>n and all the poles  $(p_1,\ldots,p_n)$  of X(z) are distinct (all roots of D(z) are different), then a polynomial of z of order (m-n) must be added to the right-hand side of Equation 4.

Therefore, for m>n, the complete partial-fraction expansion would have the form

$$X(z) = \sum_{q=0}^{m-n} b_q z^q + \sum_{k=1}^n c_k \frac{z}{z - p_k}$$

The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case for m>n Example 1

Find the inverse z transform of

$$X(z) = \frac{2z^3 - 5z^2 + z + 3}{(z - 1)(z - 2)}$$
 with ROC =  $|z| > 2$ 

using the partial-fraction expansion technique.

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [58 of 65]

The Inverse z Transform: Partial-Fraction Expansion Simple Pole Case for m > n Example 1

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [59 of 65]

## The Inverse z Transform: Partial-Fraction Expansion Multiple Pole Case

If D(z) has multiple factors of the form  $(z - p_i)^r$ , we say that  $p_i$  is a **multiple pole** of X(z) with **multiplicity** r. Then, the partial-fraction expansion of X(z) will consist of terms of the form

$$\frac{\lambda_1}{z-p_i} + \frac{\lambda_2}{(z-p_i)^2} + \dots + \frac{\lambda_r}{(z-p_i)^r}$$

where the coefficients  $\lambda_k$  are determined from the formula

$$\lambda_{r-k} = \frac{1}{k!} \cdot \frac{d^k}{dz^k} \left[ (z - p_i)^r \frac{X(z)}{z} \right] \Big|_{z=p_i}.$$

# The Inverse z Transform: Partial-Fraction Expansion Multiple Pole Case Example

Find the inverse z transform of

$$X(z) = \frac{z}{(z-1)(z-2)^2} \text{ with ROC} = |z| > 2$$

using the partial-fraction expansion technique.

The Inverse z Transform: Partial-Fraction Expansion Multiple Pole Case Example

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [62 of 65]

#### The Inverse z Transform: Power Series Expansion

The defining expression for the z transform is a power series where the sequence values x[n] are the coefficients of  $z^{-n}$ . Therefore, if X(z) is given as a power series of the form

$$X(z) = \sum_{n = -\infty}^{\infty} x[n]z^{-n}$$
  
= \cdots + x[-2]z^2 + x[-1]z + x[0] + x[1]z^{-1} + \cdots ,

we can determine any particular value of the sequence by finding the coefficient of the appropriate power of  $z^{-1}$ .

ELEC 309: Signals and Systems

ELEC 309: Signals and Systems

Complex-Domain Analysis of Discrete-Time Signals [63 of 65]

The Inverse z Transform: Power Series Expansion

This approach may not provide a closed-form solution.

It is very useful for a finite-length sequence where X(z) may have not simpler form than a polynomial in  $z^{-1}$ .

For rational z-transforms, a power series expansion can be obtained by long division.

Complex-Domain Analysis of Discrete-Time Signals [64 of 65]

Complex-Domain Analysis of Discrete-Time Signals [65 of 65]

# The Inverse z Transform: Partial-Fraction Expansion Power Series Expansion Example

Find the inverse z transform of

$$X(z) = \frac{z}{2z^2 - 3z + 1} \text{ with ROC} = |z| > 1$$

using the power series expansion technique.

# The Inverse z Transform: Partial-Fraction Expansion Power Series Expansion Example

ELEC 309: Signals and Systems