人工智能学院 现代控制论



卡尔曼滤波简介

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概要



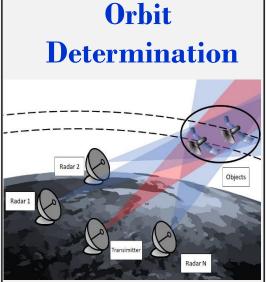
- □ 研究背景和问题描述
- 口 卡尔曼滤波算法及其主要特性
- 口 扩展卡尔曼滤波算法

研究背景



滤波: 最优的状态估计

Yolatility Estimation 实际系统





Observation	Disturbed stock	Radar	Measurements of
data	price	measurements	sensors in network
System	Volatility	Position and	Position and
state		velocity	velocity



准备知识: 随机变量

随机变量:随机试验各种结果的实值单值函数

X: 将随机试验各种结果映射到实数

(扔硬币出现正面记为1,扔硬币出现反面记为0)

ightharpoonup 随机变量X最基本的特性:概率分布函数 (probability distribution function, PDF) $F_X(x) = P(X \le x)$

➤ PDF满足的特性:

$$F_X(x) \in [0,1]$$
 $F_X(-\infty) = 0$
 $F_X(\infty) = 1$
 $F_X(a) \le F_X(b) \text{ if } a \le b$
 $P(a < X \le b) = F_X(b) - F_X(a)$

准备知识: 概率分布函数(PDF)和概率密度函数(pdf)

> 概率密度函数(Probability density function, pdf): PDF的导数

$$f_X(x) = \frac{dF_X(x)}{dx}$$

➤ pdf满足的特性:

$$F_X(x) = \int_{-\infty}^x f_X(z) dz$$

$$f_X(x) \ge 0$$

$$\int_{-\infty}^{\infty} f_X(x) dx = 1$$

$$P(a < x \le b) = \int_a^b f_X(x) dx$$

准备知识:条件概率分布函数和条件概率密度函数



> 给定事件A条件下的概率分布和概率密度

$$F_X(x|A) = P(X \le x|A)$$

$$= \frac{P(X \le x, A)}{P(A)}$$

$$f_X(x|A) = \frac{dF_X(x|A)}{dx}$$

$$f_{X_1|X_2}(x_1|x_2) = P[(X_1 \le x_1)|(X_2 = x_2)]$$
$$= \frac{f_{X_1,X_2}(x_1,x_2)}{f_{X_2}(x_2)}$$

准备知识: 期望和方差



> g(X)为随机变量的一个函数 (将随机变量映射为随机变量)

则
$$g(X)$$
的期望为 $E[g(X)] = \int_{-\infty}^{\infty} g(x) f_X(x) dx$

> X的期望为:
$$\overline{x} \triangleq E(X) = \int_{-\infty}^{\infty} x f_X(x) dx$$

> X的方差:
$$\sigma_X^2 = E[(X - \bar{x})^2]$$
$$= \int_{-\infty}^{\infty} (x - \bar{x})^2 f_X(x) dx$$

$$\triangleright$$
 常用标记: $X \sim (\bar{x}, \sigma^2)$

$$ightharpoonup$$
 条件期望: $E(X|A) = \int_{-\infty}^{\infty} x f_X(x|A) dx$

准备知识:高维随机变量



$$F_X(x) = P(X \le x)$$

 $F_Y(y) = P(Y \le y)$

ightharpoonup 联合分布函数: $F_{XY}(x,y) = P(X \le x, Y \le y)$

(扔两个硬币,出现正面+1,出现反面+0)

$$\begin{array}{rcl} F(x,y) & \in & [0,1] \\ F(x,-\infty) = F(-\infty,y) & = & 0 \\ F(\infty,\infty) & = & 1 \\ F(a,c) & \leq & F(b,d) & \text{if } a \leq b \text{ and } c \leq d \\ P(a < x \leq b, c < y \leq d) & = & F(b,d) + F(a,c) - F(a,d) - F(b,c) \\ F(x,\infty) & = & F(x) \\ F(\infty,y) & = & F(y) \end{array}$$

准备知识: 高维密度函数



ightharpoonup 联合密度函数: $f_{XY}(x,y) = \frac{\partial^2 F_{XY}(x,y)}{\partial x \partial y}$

$$F(x,y) = \int_{-\infty}^{x} \int_{-\infty}^{y} f(z_1, z_1) dz_1 dz_2$$

$$f(x,y) \geq 0$$

$$\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x,y) dx dy = 1$$

$$P(a < x \leq b, c < y \leq d) = \int_{c}^{d} \int_{a}^{b} f(x,y) dx dy$$

$$f(x) = \int_{-\infty}^{\infty} f(x,y) dy$$

$$f(y) = \int_{-\infty}^{\infty} f(x,y) dx$$

准备知识:独立性



ightharpoonup 随机变量X和Y独立: $P(X \le x, Y \le y) = P(X \le x)P(Y \le y)$ for all x, y

ightharpoonup 随机变量X和Y独立: $\begin{array}{rcl} F_{XY}(x,y) & = & F_X(x)F_Y(y) \\ f_{XY}(x,y) & = & f_X(x)f_Y(y) \end{array}$

准备知识: 高维随机变量的相关函数



$$ightharpoonup$$
 互相关函数矩阵: $R_{XY} = E(XY^T)$

$$= \begin{bmatrix} E(X_1Y_1) & \cdots & E(X_1Y_m) \\ \vdots & & \vdots \\ E(X_nY_1) & \cdots & E(X_nY_m) \end{bmatrix}$$

$$ightharpoonup$$
 互协方差矩阵: $C_{XY} = E[(X - \bar{X})(Y - \bar{Y})^T]$
= $E(XY^T) - \bar{X}\bar{Y}^T$

准备知识: 高维随机变量的协方差阵



 \triangleright 自相关函数矩阵 $R_X = E[XX^T]$

$$= \begin{bmatrix} E[X_1^2] & \cdots & E[X_1X_n] \\ \vdots & & \vdots \\ E[X_nX_1] & \cdots & E[X_n^2] \end{bmatrix}$$

ightharpoonup 自协方差矩阵 $C_X = E[(X - \bar{X})(X - \bar{X})^T]$

$$E = E[(X - \bar{X})(X - \bar{X})^{T}]$$

$$= \begin{bmatrix} E[(X_{1} - \bar{X}_{1})^{2}] & \cdots & E[(X_{1} - \bar{X}_{1})(X_{n} - \bar{X}_{n})] \\ \vdots & & \vdots \\ E[(X_{n} - \bar{X}_{n})(X_{1} - \bar{X}_{1})] & \cdots & E[(X_{n} - \bar{X}_{n})^{2}] \end{bmatrix}$$

$$= \begin{bmatrix} \sigma_{1}^{2} & \cdots & \sigma_{1n} \\ \vdots & & \vdots \\ \sigma_{n1} & \cdots & \sigma_{n}^{2} \end{bmatrix}$$

自协方差矩阵为半正定矩阵 $z^T C_X z = z^T E[(X - \bar{X})(X - \bar{X})^T]z$

Note that
$$\sigma_{ij} = \sigma_{ji}$$
 so $C_X = C_X^T$.

$$z^{T}C_{X}z = z^{T}E[(X - \bar{X})(X - \bar{X})^{T}]z$$

$$= E[z^{T}(X - \bar{X})(X - \bar{X})^{T}z]$$

$$= E[(z^{T}(X - \bar{X}))^{2}]$$
> 0

准备知识: 随机过程



随机过程:一列随机变量X(t),t为时间

$$F_X(x,t) = P(X(t) \le x)$$

$$F_X(x,t) = P[X_1(t) \le x_1 \text{ and } \cdots X_n(t) \le x_n(t)]$$

均为时变的函数:

$$f_X(x,t) = \frac{d^n F_X(x,t)}{dx_1 \cdots dx_n}$$

$$\bar{x}(t) = \int_{-\infty}^{\infty} x f(x,t) dx$$

$$C_X(t) = E\left\{ \left[X(t) - \bar{x}(t) \right] \left[X(t) - \bar{x}(t) \right]^T \right\}$$
$$= \int_{-\infty}^{\infty} \left[x - \bar{x}(t) \right] \left[x - \bar{x}(t) \right]^T f(x, t) dx$$

$$R_X(t_1, t_2) = E[X(t_1)X^T(t_2)]$$

$$C_X(t_1, t_2) = E\left\{ \left[X(t_1) - \bar{X}(t_1) \right] \left[X(t_2) - \bar{X}(t_2) \right]^T \right\}$$

is is

离散时变线性系统的滤波问题

$$\begin{cases} X_{k+1} = A_k X_k + B_k \mathbf{u}_k + w_k \\ Y_{k+1} = C_{k+1} X_{k+1} + n_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

$$\{n_k\}$$
和 $\{w_k\}$ 为随机过程

$$E(n_k) = 0, \forall k > 0$$

$$E(w_k) = 0, \forall k > 0$$

$$C_{n}(k, j) = \begin{cases} 0, \forall k \neq j \\ R_{k}, \forall k = j \end{cases}$$

$$C_{w}(k, j) = \begin{cases} 0, \forall k \neq j \\ O, \forall k \neq j \end{cases}$$

$$O_{m}(k, j) = \begin{cases} 0, \forall k \neq j \\ O_{m}(k, j) = j \end{cases}$$



离散时变线性系统的滤波问题

$$\begin{cases} X_{k+1} = A_k X_k + B_k \mathbf{u}_k + w_k \\ Y_{k+1} = C_{k+1} X_{k+1} + n_{k+1} \end{cases}$$

$$k = 0, 1, 2, \dots$$

$$\{n_k\}$$
和 $\{w_k\}$ 为随机过程

$$E(n_k) = 0, \forall k > 0$$
$$E(w_k) = 0, \forall k > 0$$

$$C_{n}(k, j) = \begin{cases} 0, \forall k \neq j \\ R_{k}, \forall k = j \end{cases}$$

$$C_{w}(k, j) = \begin{cases} 0, \forall k \neq j \\ O_{k}, \forall k = j \end{cases}$$

目标:利用量测获得 X, 的线性最小方差估计

$$\hat{X}_k = \underset{Z_k \in \{\text{linear function of } Y_i, 1 \leq i \leq k\}}{\arg\min} E\left\{ \left(Z_k - X_k\right) \left(Z_k - X_k\right)^T \right\}$$

概要



- □ 研究背景和问题描述
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卡尔曼滤波算法及参数设计



喜散线性系统:
$$\begin{cases} X_{k+1} = A_k X_k + B_k \mathbf{u}_k + w_k \\ Y_{k+1} = C_{k+1} X_{k+1} + n_{k+1} \end{cases}$$
 $k = 0, 1, 2, ...$

卡尔曼滤波:

$$\overline{X}_{k+1} = A_k \hat{X}_k + B_k \mathbf{u}_k$$
: 预测过程

$$\bar{P}_{k+1} = A_k P_k A_k^T + Q_k$$
: 预测误差的协方差阵

$$\hat{X}_{k+1} = \overline{X}_{k+1} + K_{k+1}(Y_{k+1} - C_{k+1}\overline{X}_{k+1})$$
: 更新过程

$$K_{k+1} = (\overline{P}_{k+1}C_{k+1}^T)(C_{k+1}\overline{P}_{k+1}C_{k+1}^T + R_{k+1})^{-1}$$
: 更新增益

$$P_{k+1} = \bar{P}_{k+1} - K_{k+1} \bar{P}_{k+1}$$
: 估计(滤波)误差的协方差阵

1960s, Kalman

考虑线性定常系统:

$$\begin{cases} X_{k+1} = AX_k + Bu_k + w_k \\ Y_{k+1} = CX_{k+1} + n_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

$$\begin{cases} \hat{X}_{k+1} = \overline{X}_{k+1} + K_{k+1} (Y_{k+1} - C\overline{X}_{k+1}) \\ \overline{X}_{k+1} = A\hat{X}_k + Bu_k \\ K_{k+1} = (\overline{P}_{k+1}C^T)(C \ \overline{P}_{k+1}C^T + R)^{-1} \\ P_{k+1} = \overline{P}_{k+1} - K_{k+1}C \ \overline{P}_{k+1} \\ \overline{P}_{k+1} = AP_k A^T + Q \end{cases}$$

定理: 卡尔曼滤波满足如下特性:

1. 线性最小方差估计

2.
$$P_k = \mathbb{E}\left\{ (X_k - \hat{X}_k)(X_k - \hat{X}_k)^T \right\}$$

3.当(A,C)可观时,滤波误差均方有界



引理1. 设x和y分别是具有前二阶矩的随机向量,则利用y对x的线性无偏最小方差估计为

$$\widehat{x}_y = \mathbb{E}\{x\} + R_{xy}R_y^{-1}(y - \mathbb{E}\{y\}),$$

其中

$$R_{xy} \triangleq \mathbb{E}\{(x - \mathbb{E}\{x\})(y - \mathbb{E}\{y\})^T\},\$$

$$R_{y} \triangleq \mathbb{E}\{(y - \mathbb{E}\{y\})(y - \mathbb{E}\{y\})^T\}.$$



引理1的证明:由估计的线性无偏性可知 \hat{x}_{ν} 具有如下形式:

$$\hat{x}_y = \mathbb{E}\{x\} + K(y - \mathbb{E}\{y\}).$$

由此易得

$$\mathbb{E}\left\{(\hat{x}_{y}-x)(\hat{x}_{y}-x)^{T}\right\}$$

$$=\mathbb{E}\{(x-\mathbb{E}\{x\}-K(y-\mathbb{E}\{y\}))(x-\mathbb{E}\{x\}-K(y-\mathbb{E}\{y\}))^{T}\}$$

$$=R_{x}-KR_{yx}-R_{xy}K^{T}+KR_{y}K^{T}$$

$$=(K-R_{xy}R_{y}^{-1})R_{y}(K-R_{xy}R_{y}^{-1})^{T}+R_{x}-R_{xy}R_{y}^{-1}R_{yx}$$
其中最后一个等式的第1项为非负定项,故欲令估计误差的协

方差阵最小,当且仅当 $K = R_{xy}R_y^{-1}$.



引理2. 设u为已知向量,x和y是具有前二阶矩的随机向量,记y对x的线性无偏最小方差估计为 \hat{x}_y ,则利用y对 $z ext{$\infty$} Ax + Bu + Cy$ 的线性无偏最小方差估计为 $\hat{z}_y = A\hat{x}_y + Bu + Cy$.

证明过程



引理3. 设x, y和y是具有前二阶矩的随机向量, 记 $w \triangleq [y^T \ y^T]^T$, 若利用y对x和y的估计分别为 \hat{x}_y 和 \hat{y}_y ,则利用w对x的估计满足

$$\hat{x}_w = \hat{x}_\mathcal{Y} + K(y - \hat{y}_\mathcal{Y}),$$

其中

$$K = R_{xy|y}R_{y|y}^{-1},$$

$$R_{xy|y} \triangleq \mathbb{E}\{(x - \hat{x}_y)(y - \hat{y}_y)^T\},$$

$$R_{y|y} \triangleq \mathbb{E}\{(y - \hat{y}_y)(y - \hat{y}_y)^T\}.$$



引理3的证明: 记 $e = x - \hat{x}_{y} - K(y - \hat{y}_{y})$, 则由引理2

知 $\hat{e}_w = \hat{x}_w - \hat{x}_y - K(y - y_y)$,从而只需验证 $\hat{e}_w = 0$. 根据引理1易得

$$\mathbb{E}\left\{\left(x-\hat{x}_{\mathcal{Y}}\right)(\mathcal{Y}-\mathbb{E}\{\mathcal{Y}\})^{T}\right\}=\mathbf{0},$$

$$\mathbb{E}\left\{\left(y-\hat{y}_{\mathcal{Y}}\right)(\mathcal{Y}-\mathbb{E}\{\mathcal{Y}\})^{T}\right\}=\mathbf{0}.$$

从而 $\mathbb{E}\{e(y - \mathbb{E}\{y\})^T\} = \mathbf{0}$. 进而有

$$\mathbb{E}\{e(y - \mathbb{E}\{y\})^T\} = \mathbb{E}\left\{e(y - \hat{y}_y)^T\right\} = \mathbf{0}.$$

注意到 $\mathbb{E}\{e\} = \mathbf{0}$,从而 $R_{ew} = \mathbb{E}\{e(w - \mathbb{E}\{w\})^T\} = \mathbf{0}$.则根据引理1知 $\hat{e}_w = 0$.



定理的证明

 $il Y_k \triangleq [Y_0^T \ Y_1^T \cdots \ Y_k^T]^T 对 X_k, X_{k+1} 和 Y_{k+1}$ 的线性无偏最小方差估计分别为 $\hat{X}_{k}, \bar{X}_{k+1}$ 和 \bar{Y}_{k+1} ,则根据引理2可得 $\bar{X}_{k+1} = A\hat{X}_{k} + Bu_{k}, \bar{Y}_{k+1} = C\bar{X}_{k+1}$. 同时记 \hat{X}_k 和 \bar{X}_{k+1} 的估计误差协方差阵为 P_k 和 \bar{P}_{k+1} ,则 $\bar{P}_{k+1} = AP_kA^T + Q$. 进一步地 ,将引理3中的x,y和y分别替换成 X_{k+1} , y_k 和 Y_{k+1} ,可得 $\hat{X}_{k+1} = \bar{X}_{k+1} +$ $K_{k+1}(Y_{k+1}-CX_{k+1}), \perp$ $K_{k+1} = \mathbb{E}\{(X_{k+1} - \bar{X}_{k+1})(Y_{k+1} - \bar{Y}_{k+1})^T\}(\mathbb{E}\{(Y_{k+1} - \bar{Y}_{k+1})(Y_{k+1} - \bar{Y}_{k+1})^T\})^{-1}$ $= \bar{P}_{\nu+1} C^T (C\bar{P}_{\nu+1}C^T + R)^{-1}$ $P_{k+1} = (\mathbf{I} - K_{k+1}C)\bar{P}_{k+1}(\mathbf{I} - K_{k+1}C)^T + K_{k+1}RK_{k+1} = \bar{P}_{k+1} - K_{k+1}C\bar{P}_{k+1}.$

定理的证明 (续)



记 $Y_{k(n)} \triangleq [Y_{k-n+1}^T \ Y_{k-n}^T \ \cdots \ Y_k^T]^T, M \triangleq$ $[C^T \ (CA)^T \ \cdots \ (CA^{n-1})^T]^T, 由于(A, C) 能观,亦即M是列满 秩的,从而可以构造如下估计$

$$\tilde{X}_k = \begin{cases} A^k \mathbb{E}\{x_0\}, & k < n \\ A^{n-1} (M^T M)^{-1} M^T Y_{k(n)}, & k \ge n \end{cases}$$

易得 \tilde{X}_k 的估计误差协方差阵 \tilde{P}_k 是有界的。由于 \tilde{P}_k 是最小方差的线性估计,从而 $P_k \leq \tilde{P}_k$ 也是有界的。

概要



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非线性系统的滤波问题:

$$\begin{cases} X_{k+1} = f(X_k) + w_k \\ Y_{k+1} = g(X_{k+1}) + v_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

$$X_k$$
: state, Y_k : measured output,

$$(f(\bullet), g(\bullet))$$
: known C^1 functions

 $(\{n_k\}, \{w_k\})$: uncorrelated zero-mean stochastic noise

Objective: linear minimum variance estimation, i.e.,

$$\hat{X}_k = \underset{Z_k \in \{\text{linear function of } Y_i, 1 \leq i \leq k\}}{\arg\min} E\left\{ \left(Z_k - X_k\right) \left(Z_k - X_k\right)^T \right\}$$

对付非线性的主要方法:线性化

<u>线性化</u> <u>linearization</u>



Kalman Filter, Kalman.

H-infinity Filter, Banavar, etc.

Set-valued Filter, Zhu, etc.

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线性系统的卡尔曼滤波



$$\begin{cases} X_{k+1} = AX_k + BF(X_k, k) + w_k \\ Y_{k+1} = CX_{k+1} + n_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

 X_k : state, Y_k : measured output

(A,B,C): known matrices

 $F(X_k,k)$: nonlinear dynamics

字文章
$$\begin{cases} \hat{X}_{k+1} = \overline{X}_{k+1} + K_{k+1} (Y_{k+1} - C\overline{X}_{k+1}) & \text{Parameters design:} \\ \overline{X}_{k+1} = A\hat{X}_k + Bu_k \\ K_{k+1} = (\overline{P}_{k+1}C^T)(C\ \overline{P}_{k+1}C^T + R\)^{-1} \\ P_{k+1} = \overline{P}_{k+1} - K_{k+1}C\ \overline{P}_{k+1} \\ \overline{P}_{k+1} = AP_kA^T + Q \end{cases}$$
 Parameters design:
$$\begin{cases} Q_k = E\left(w_k w_k^T\right) \\ R_{k+1} = E\left(n_{k+1}n_{k+1}^T\right) > 0 \\ P_0 = E\left(X_0 - EX_0\right)(X_0) \\ \overline{X}_0 = EX_0 \end{cases}$$

Parameters design:

linear model

$$\begin{cases} Q_k = E\left(w_k w_k^T\right) \\ R_{k+1} = E\left(n_{k+1} n_{k+1}^T\right) > 0 \\ P_0 = E\left(X_0 - EX_0\right) \left(X_0 - EX_0\right)^T \\ \overline{X}_0 = EX_0 \end{cases}$$

Require the exact information of : noise's model initial value's model





$$\begin{cases} X_{k+1} = f(X_k) + w_k \\ Y_{k+1} = g(X_{k+1}) + v_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

$$A_k = \frac{\partial f}{\partial X} \bigg|_{\hat{X}_k},$$

$$C_{k+1} = \frac{\partial g}{\partial X} \bigg|_{\bar{X}_{k+1}}$$

 X_k : state, Y_k : measured output, $(f(\bullet), g(\bullet))$: known C^1 functions $(\{v_k\}, \{w_k\})$: uncorrelated zero-mean

第k+1步: 在滤波值点的线性部分;

第k+1步: 在预测值点的线性部分

AMSS

扩展卡尔曼滤波

$$\begin{cases} X_{k+1} = f(X_k) + w_k \\ Y_{k+1} = g(X_{k+1}) + v_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

$$X_k$$
: state, Y_k : measured output, $(f(\bullet), g(\bullet))$: known C^1 functions $(\{v_k\}, \{w_k\})$: uncorrelated zero-mean

$$\begin{cases} \hat{X}_{k+1} = \overline{X}_{k+1} + K_{k+1} \left(Y_{k+1} - g \left(\overline{X}_{k+1} \right) \right) \\ \overline{X}_{k+1} = F \left(\hat{X}_{k}, k \right) \\ K_{k+1} = (\overline{P}_{k+1} \overline{C}_{k+1}^{T}) (\overline{C}_{k+1} \overline{P}_{k+1} \overline{C}_{k+1}^{T} + R_{k+1})^{-1} \\ P_{k+1} = \overline{P}_{k+1} - K_{k+1} \overline{C}_{k+1} \overline{P}_{k+1} \\ \overline{P}_{k+1} = \overline{A}_{k} P_{k} \overline{A}_{k}^{T} + Q_{k} \\ A_{k} = \frac{\partial f}{\partial X} \Big|_{\hat{X}} , C_{k+1} = \frac{\partial g}{\partial X} \Big|_{\overline{X}} \end{cases}$$

Parameters design:

$$\begin{cases} Q_k = E\left(w_k w_k^T\right) \\ R_{k+1} = E\left(n_{k+1} n_{k+1}^T\right) > 0 \\ P_0 = E\left(X_0 - EX_0\right) \left(X_0 - EX_0\right)^T \\ \overline{X}_0 = EX_0 \end{cases}$$

Utilizing the linear part at the point of estimation value

$$\begin{cases} X_{k+1} = f(X_k) + w_k \\ Y_{k+1} = g(X_{k+1}) + v_{k+1} \\ k = 0, 1, 2, \dots \end{cases}$$

Reif K., et al., Stochastic stability of the discrete-time extended Kalman filter, IEEE TAC, 1999

$$\begin{cases} \hat{X}_{k+1} = \overline{X}_{k+1} + K_{k+1} \left(Y_{k+1} - g \left(\overline{X}_{k+1} \right) \right) \\ \overline{X}_{k+1} = F \left(\hat{X}_{k}, k \right) \\ K_{k+1} = (\overline{P}_{k+1} \overline{C}_{k+1}^{T}) (\overline{C}_{k+1} \overline{P}_{k+1} \overline{C}_{k+1}^{T} + R_{k+1})^{-1} \\ P_{k+1} = \overline{P}_{k+1} - K_{k+1} \overline{C}_{k+1} \overline{P}_{k+1} \\ \overline{P}_{k+1} = \overline{A}_{k} P_{k} \overline{A}_{k}^{T} + Q_{k}, \qquad A_{k} = \frac{\partial f}{\partial X} \Big|_{\hat{X}_{k}}, C_{k+1} = \frac{\partial g}{\partial X} \Big|_{\overline{X}_{k+1}} \end{cases}$$

EKF算法的主要理论结果

- Exact information of nonlinear model
- Conditions: > Sufficiently small noise and initial error
 - > Uniformly observable condition

稳定性:
$$\sup_{k} \left\| E\left((\hat{X}_{k} - X_{k}) (\hat{X}_{k} - X_{k})^{T} \right) \right\| < \infty$$
 最优性? — 致性?





谢谢!