人工智能学院《现代控制论》



自抗扰控制系统稳定性的基本理论结果

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ADRC as a powerful solution has been explored in many domain of control engineering:

- > 电机控制 (speed control of induction motor and permanent-magnet synchronous, noncircular turning process, etc)
- ➤ 飞行控制 (attitude control of space ship, high-speed vehicles, UAV, morphing aircraft, etc)
- ▶ 机器人 (force control, uncalibrated hand-eye coordination, AUV, etc)
- 热工过程 (unstable heat conduction systems, boiler-turbine-generator systems, ALSTOM gasifier, fractional-order system, etc)
- ▶ 电力电子装置 (DC-to DC power converters, rectifiers, inverters, HVDC SMC systems, etc)
- ▶ 行器
- ▶ 电力系统、发动机、汽车、内燃机。。。

ADRC in U.S.: Milestones

> 1997: made the 1st successful ADRC hardware test on a servo mechanism

>2008: \$1M venture capital, grew by \$5M in 2012.

▶2010: 1st factory implementation, 10 Parker extrusion lines (挤压机生产线)

at a Parker Hannifin Extrusion Plant in North America

(cpk: from 2.3 to >8; avg. 节能 57%)



>2011: implemented ADRC in several high energy particle accelerators (高能 粒子加速器) in the National Superconducting Cyclotron (超导回旋加速器) Lab in the U.S.

>2011: Texas Instrument adopts ADRC; 3 patents granted.

▶2013: Texas Instrument released the ADRC based motion control chips (德州仪器的运动控制芯片)

从学术到工业的跨越



本次课程目的:通过理论分析了解 自抗扰控制系统的主要特点

自抗扰的发展历程: 理论研究是远远落后于应用

致力于处理大范围的不确定性,导致理论分析较困难

不确定系统: 非线性, 时变, 多输入多输出

扰动信号: 不连续

状态反馈: 非线性结构

实现闭环系统的预期动态,缺少分析方法



Control input

20

25

自抗扰控制系统的特点: 在不确定性保证了闭环系统输出达到期望的瞬态

Simulation results of flight control: roll rate tracking under ±40% random derivations of parameters

闭环系统的角速率响应 控制输入 150 Reference signal Actual trajectory 10 100 Cases 100 50 100 Cases da (deg) (s/gab) d -50 -10 -100 -15 -20 -150 5 10 15 5 10 15 20 25 Time(s)

 $\dot{p} = b(t)u + F(p, u, t)$

P:角速率

b(t): 控制输入增益

F:总扰动

u:控制输入

线性ESO的估计误差分析



$$\begin{cases} \dot{X}_1 = X_2 \\ \dots \end{cases}$$

不确定系统:
$$\begin{cases} X_1 & X_2 \\ \dots \\ \dot{X}_{n-1} = X_n \\ \dot{X}_n = X_{n+1} + \overline{B}(t)U(t) \end{cases}$$

总扰动:
$$X_{n+1} = F(X,t) + d(t) + (B(X,t) - \overline{B}(t))U(t)$$

线性ESO: $\begin{cases} \dot{\hat{X}}_1 = \hat{X}_2 + l_1(\hat{X}_1 - Y) \\ \dots \\ \dot{\hat{X}}_n = \hat{X}_{n+1} + l_n(\hat{X}_1 - Y) + \overline{B}(t)U(t) \\ \dot{\hat{X}}_{n+1} = l_{n+1}(\hat{X}_1 - Y) \end{cases}$

ESO的带宽:
$$\omega_e$$
 $s^{n+1} + \sum_{i=1}^{n+1} l_i s^{n+1-i} = (s + \omega_e)^{n+1}$

经典的理论分析结果:

[Q Zheng, L Gao, Z Gao, 2007CDC] Assuming that $\dot{X}_{n+1}(.)$ is a bounded function, then there exists a finite $T_1 > 0$ such that

$$\left|X_i - \hat{X}_i\right| \le O\left(\frac{1}{\omega_e^k}\right) \qquad \forall t \ge T_1 > 0, \, \omega_e > 0, \ i = 1, 2, ..., n+1$$

线性ESO的估计误差分析: 定理1



定理1:假设总扰动的导数 \dot{X}_{n+1} 有界,则

$$\lim_{t\to\infty} \left| X_i(t) - \hat{X}_i(t) \right| \le O\left(\frac{1}{\omega_e^k}\right) \qquad k = n+2-i, \qquad \omega_e > 0, \qquad i = 1, 2, ..., n+1.$$

$$k=n+2-i,$$

$$\omega_e > 0$$
,

$$i = 1, 2, ..., n + 1.$$

证明: 首先定义ESO的估计误差
$$\hat{E} \triangleq \begin{bmatrix} \hat{E}_1 \\ \hat{E}_2 \\ \vdots \\ \hat{E}_{n+1} \end{bmatrix} = \begin{bmatrix} X_1 - \hat{X}_1 \\ X_2 - \hat{X}_2 \\ \vdots \\ X_{n+1} - \hat{X}_{n+1} \end{bmatrix}$$

$$\begin{cases} \hat{E}_{1} = \hat{E}_{2} - l_{1}\hat{E}_{1} \\ \dots \\ \dot{\hat{E}}_{n} = \hat{E}_{n+1} - l_{n}\hat{E}_{1} \\ \dot{\hat{E}}_{n+1} = \dot{X}_{n+1} - l_{n+1}\hat{E}_{1} \end{cases}$$

进而得到ESO估计误差的动态方程
$$\begin{cases} \dot{\hat{E}}_1 = \hat{E}_2 - l_1 \hat{E}_1 \\ \dots \\ \dot{\hat{E}}_n = \hat{E}_{n+1} - l_n \hat{E}_1 \\ \dot{\hat{E}}_{n+1} = \dot{X}_{n+1} - l_{n+1} \hat{E}_1 \end{cases}$$
其中:
$$\begin{cases} s^{n+1} + \sum_{i=1}^{n+1} l_i s^{n+1-i} = (s + \omega_e)^{n+1} \\ or \\ l_i = C_{n+1}^i \omega_e^i \end{cases}$$



再引入新状态及等价变换:
$$\xi \triangleq \begin{bmatrix} \xi_1 \\ \xi_2 \\ \vdots \\ \xi_{n+1} \end{bmatrix} = \begin{bmatrix} \omega_e^n & 0 & \cdots & 0 \\ 0 & \omega_e^{n-1} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & 1 \end{bmatrix} \begin{bmatrix} \hat{E}_1 \\ \hat{E}_2 \\ \vdots \\ \hat{E}_{n+1} \end{bmatrix}$$

即有: $\xi_i = \omega_e^{n+1-i} \hat{E}_i, \dot{\xi} = \omega_e^{n+1-i} \dot{\hat{E}},$

进而得到新状态的动态方程:
$$\dot{\xi} = \begin{bmatrix} \omega_e^n & 0 & \cdots & 0 \\ 0 & \omega_e^{n-1} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & 1 \end{bmatrix} \dot{E}$$

$$= \begin{bmatrix} \omega_e^n & 0 & \cdots & 0 \\ 0 & \omega_e^{n-1} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & 1 \end{bmatrix} \begin{pmatrix} l_1 & 1 & \cdots & 0 \\ l_2 & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ l_{n+1} & 0 & \cdots & 0 \end{pmatrix} \dot{E} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix}$$

$$= \omega_e \begin{bmatrix} -C_{n+1}^1 & 1 & \cdots & 0 \\ -C_{n+1}^2 & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ -C_{n+1}^{n+1} & 0 & \cdots & 0 \end{bmatrix} \xi + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix}$$



发性的的估计误差分析:定理 【续)
$$\dot{\xi} = \omega_{e} \begin{bmatrix}
-C_{n+1}^{1} & 1 & 0 & \cdots & 0 \\
-C_{n+2}^{2} & 0 & 1 & \cdots & 0 \\
\vdots & \vdots & \vdots & \ddots & \vdots \\
-C_{n+1}^{n} & 0 & 0 & \cdots & 1 \\
-C_{n+1}^{n+1} & 0 & 0 & \cdots & 0
\end{bmatrix} \xi + \begin{bmatrix}
0 \\
0 \\
\vdots \\
0 \\
\dot{X}_{n+1}
\end{bmatrix}$$
因为矩阵
$$\begin{bmatrix}
-C_{n+1}^{1} & 1 & 0 & \cdots & 0 \\
-C_{n+2}^{2} & 0 & 1 & \cdots & 0 \\
\vdots & \vdots & \vdots & \ddots & \vdots \\
-C_{n+1}^{n} & 0 & 0 & \cdots & 1 \\
-C_{n+1}^{n+1} & 0 & 0 & \cdots & 0
\end{bmatrix}$$
为赫尔维兹矩阵,则存在一个正定矩阵 P 使得:
$$\begin{bmatrix}
-C_{n+1}^{1} & 1 & 0 & \cdots & 0
\end{bmatrix}^{T} \qquad \begin{bmatrix}
-C_{n+1}^{1} & 1 & 0 & \cdots & 0
\end{bmatrix}$$

$$\begin{bmatrix} -C_{n+1}^2 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -C_{n+1}^n & 0 & 0 & \cdots & 1 \\ -C_{n+1}^{n+1} & 0 & 0 & \cdots & 0 \end{bmatrix}$$

$$\begin{bmatrix} -C_{n+1}^{1} & 1 & 0 & \cdots & 0 \\ -C_{n+2}^{2} & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -C_{n+1}^{n} & 0 & 0 & \cdots & 1 \\ -C_{n+1}^{n+1} & 0 & 0 & \cdots & 0 \end{bmatrix}^{T} P + P \begin{bmatrix} -C_{n+1}^{1} & 1 & 0 & \cdots & 0 \\ -C_{n+2}^{2} & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -C_{n+1}^{n} & 0 & 0 & \cdots & 1 \\ -C_{n+1}^{n+1} & 0 & 0 & \cdots & 0 \end{bmatrix} = -I$$

定义矩阵
$$V(\xi) \triangleq \xi^T P \xi$$
 , 则有

$$\frac{dV(\xi)}{dt} = -\omega_e \xi^T \xi + 2\xi^T P \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix}$$



$$\frac{dV(\xi)}{dt} = -\omega_e \xi^T \xi + 2\xi^T P \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix}$$



$$\frac{dV(\xi)}{dt} \leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \left\| \xi \right\| c_{2} \left\| \dot{\boldsymbol{X}}_{n+1} \right\| \leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \frac{\sqrt{V(\xi)}}{\sqrt{c_{1}}} c_{2} \left\| \dot{\boldsymbol{X}}_{n+1} \right\|$$



$$\frac{d\sqrt{V(\xi)}}{dt} \leq -\omega_e \frac{\sqrt{V(\xi)}}{2c_2} + \frac{c_2}{\sqrt{c_1}} \left\| \dot{\boldsymbol{X}}_{n+1} \right\|$$



$$\frac{d\sqrt{V(\xi)}}{dt} \leq -\omega_e \frac{\sqrt{V(\xi)}}{2c_2} + \frac{c_2}{\sqrt{c_1}} \left\| \dot{X}_{n+1} \right\|$$

又由于 \dot{X}_{n+1} 有界,则存在一个常数 M,使得

$$\left\|\dot{X}_{n+1}\right\| \leq M, \quad \forall t \geq t_0.$$

有

$$\frac{d\sqrt{V(\xi)}}{dt} \le -\omega_e \frac{\sqrt{V(\xi)}}{2c_2} + \frac{c_2}{\sqrt{c_1}} \mathbf{M}, \qquad \forall t \ge t_0$$

由Gronwall-Bellman不等式有

$$\sqrt{V(\xi)} \le e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))} + \frac{c_2}{\sqrt{c_1}} M \int_{t_0}^t e^{-\frac{\omega_e}{2c_2}(t-s)} ds, \qquad \forall t \ge t_0$$



由矩阵 $V(\xi) \triangleq \xi^T P \xi$ 定义,

则有

$$\|\xi\| \leq \sqrt{\frac{V(\xi)}{c_1}} \leq e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{\frac{V(\xi(t_0))}{c_1}} + \frac{c_2}{c_1} M \int_{t_0}^t e^{-\frac{\omega_e}{2c_2}(t-s)} ds, \qquad \forall t \geq t_0$$

又由于
$$\xi_i = \omega_e^{n+1-i} \hat{E}_i, \hat{E}_i = \frac{1}{\omega_e^{n+1-i}} \xi_i,$$
 则有

$$\begin{aligned} \left| \hat{E}_{i}(t) \right| &= \frac{1}{\omega_{e}^{n+1-i}} \left| \xi_{i}(t) \right| \\ &\leq \frac{1}{\omega_{e}^{n+1-i}} \left\| \xi(t) \right\| \\ &\leq \frac{1}{\omega_{e}^{n+1-i}} \left[e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \sqrt{\frac{V(\xi(t_{0}))}{c_{1}}} + \frac{c_{2}}{c_{1}} M \int_{t_{0}}^{t} e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} ds \right]. \end{aligned}$$



又由于

$$\begin{split} &\lim_{t \to \infty} \left| \hat{E}_i(t) \right| \leq \lim_{t \to \infty} \frac{1}{\omega_e^{n+1-i}} \left[e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{\frac{V(\xi(t_0))}{c_1}} + \frac{c_2}{c_1} M \int_{t_0}^t e^{-\frac{\omega_e}{2c_2}(t-s)} ds \right] \\ &\leq \lim_{t \to \infty} \frac{1}{\omega_e^{n+1-i}} \left[e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{\frac{V(\xi(t_0))}{c_1}} + \frac{c_2}{c_1} M \int_{t_0}^t e^{-\frac{\omega_e}{2c_2}(t-s)} ds \right] \\ &\leq \lim_{t \to \infty} \frac{1}{\omega_e^{n+1-i}} \left[e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{\frac{V(\xi(t_0))}{c_1}} + \frac{c_2}{c_1} M \frac{2c_2}{\omega_e} \left(1 - e^{-\frac{\omega_e}{2c_2}(t-t_0)} \right) \right] \\ &\leq \frac{1}{\omega_e^{n+1-i}} \frac{c_2}{c_1} M \frac{2c_2}{\omega_e}. \end{split}$$

所以,
$$\left|X_{i}-\hat{X}_{i}\right| \leq O\left(\frac{1}{\omega_{e}^{k}}\right)$$
 $k=n+2-i, \forall t \geq T_{1} > t_{0}, \omega_{e} > 0, i=1,2,...,n+1$ 证毕。

线性ESO的估计误差分析: 定理2



定理2: 假设总扰动的导数, 即 $\dot{X}_{n+1}(.)$, 满足 $\lim_{t\to\infty}\dot{X}_{n+1}(.)=0$, 则:

$$\lim_{t \to \infty} |X_i - \hat{X}_i| = 0 \qquad i = 1, 2, ..., n+1$$

证明:同上定义矩阵 $V(\xi) \triangleq \xi^T P \xi$,有

$$\frac{dV(\xi)}{dt} = -\omega_e \xi^T \xi + 2\xi^T P \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix}$$

$$\frac{dV(\xi)}{dt} \leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \left\| \xi \right\| c_{2} \left\| \dot{\boldsymbol{X}}_{n+1} \right\| \leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \frac{\sqrt{V(\xi)}}{\sqrt{c_{1}}} c_{2} \left\| \dot{\boldsymbol{X}}_{n+1} \right\|$$

$$\frac{d\sqrt{V(\xi)}}{dt} \leq -\omega_e \frac{\sqrt{V(\xi)}}{2c_2} + \frac{c_2}{\sqrt{c_1}} \left\| \dot{\boldsymbol{X}}_{n+1} \right\|$$

由Gronwall-Bellman不等式有

$$\sqrt{V(\xi)} \le e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))} + \frac{c_2}{\sqrt{c_1}} \int_{t_0}^t e^{-\frac{\omega_e}{c_2}(t-s)} \left\| \dot{X}_{n+1}(s) \right\| ds$$



又由于 $\lim_{n\to\infty} \dot{X}_{n+1} = 0$, 则有

$$\lim_{t\to\infty} \varepsilon(t) = 0$$

$$\varepsilon(t) \triangleq \sup_{s \in \left[\frac{t}{2}, \infty\right)} \left\| \dot{X}_{n+1}(s) \right\|$$

因此有

$$\begin{split} & \sqrt{V(\xi)} \leq e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \sqrt{V(\xi(t_{0}))} + \frac{c_{2}}{\sqrt{c_{1}}} \left(\int_{t_{0}}^{t/2} + \int_{t/2}^{t} \right) e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} \left\| \dot{X}_{n+1}(s) \right\| ds \\ & \leq e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \sqrt{V(\xi(t_{0}))} + \frac{c_{2}}{\sqrt{c_{1}}} \int_{t_{0}}^{t/2} e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} \left\| \dot{X}_{n+1}(s) \right\| ds + \varepsilon(t) \frac{c_{2}}{\sqrt{c_{1}}} \int_{t/2}^{t} e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} ds \\ & \leq e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \sqrt{V(\xi(t_{0}))} + \frac{c_{2}}{\sqrt{c_{1}}} M \int_{t_{0}}^{t/2} e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} ds + \varepsilon(t) \frac{c_{2}}{\sqrt{c_{1}}} \int_{t/2}^{t} e^{-\frac{\omega_{e}}{2c_{2}}(t-s)} ds \\ & \leq e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \sqrt{V(\xi(t_{0}))} + \frac{c_{2}}{\sqrt{c_{1}}} M \frac{2c_{2}}{\omega_{e}} \left(e^{-\frac{\omega_{e}}{2c_{2}}\frac{t}{2}} - e^{-\frac{\omega_{e}}{2c_{2}}(t-t_{0})} \right) + \varepsilon(t) \frac{c_{2}}{\sqrt{c_{1}}} \frac{2c_{2}}{\omega_{e}} \left(1 - e^{-\frac{\omega_{e}}{2c_{2}}\frac{t}{2}} \right). \end{split}$$



进一步可以得到:

$$\lim_{t \to \infty} \sqrt{V(\xi)} \leq \lim_{t \to \infty} e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))}$$

$$+ \lim_{t \to \infty} \frac{c_2}{\sqrt{c_1}} M \frac{2c_2}{\omega_e} \left(e^{-\frac{\omega_e}{2c_2} \frac{t}{2}} - e^{-\frac{\omega_e}{2c_2}(t-t_0)} \right)$$

$$+ \lim_{t \to \infty} \varepsilon(t) \frac{c_2}{\sqrt{c_1}} \frac{2c_2}{\omega_e} \lim_{t \to \infty} \left(1 - e^{-\frac{\omega_e}{2c_2} \frac{t}{2}} \right)$$

$$= 0$$

线性ESO的估计误差分析: 定理3



定理3: 假设总扰动的导数有界,则

$$\left|X_{i}(t) - \hat{X}_{i}(t)\right| \leq O\left(\frac{1}{\omega_{e}^{k}}\right) \qquad \forall t \geq T \triangleq t_{0} + 2(n+1)c_{2} \max\left\{\frac{\ln \omega_{e}}{\omega_{e}}, 0\right\}, \ \omega_{e} > 0, \ i = 1, 2, ..., n+1$$

证明:同上,定义矩阵 $V(\xi) \triangleq \xi^T P \xi$,有

$$\begin{split} \frac{dV(\xi)}{dt} &= -\omega_{e} \xi^{T} \xi + 2 \xi^{T} P \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \dot{X}_{n+1} \end{bmatrix} \\ \frac{dV(\xi)}{dt} &\leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \|\xi\| c_{2} \|\dot{X}_{n+1}\| \leq -\omega_{e} \frac{V(\xi)}{2c_{2}} + 2 \frac{\sqrt{V(\xi)}}{\sqrt{c_{1}}} c_{2} \|\dot{X}_{n+1}\| \\ \frac{d\sqrt{V(\xi)}}{dt} &\leq -\omega_{e} \frac{\sqrt{V(\xi)}}{2c_{2}} + \frac{c_{2}}{\sqrt{c_{1}}} \|\dot{X}_{n+1}\| \end{split}$$



又由于 \dot{X}_{n+1} 有界,则存在一个常数 M, 使得

$$\|\dot{X}_{n+1}\| \leq M, \quad \forall t \geq t_0.$$

因此有

$$\frac{d\sqrt{V(\xi)}}{dt} \le -\omega_e \frac{\sqrt{V(\xi)}}{2c_2} + \frac{c_2}{\sqrt{c_1}} \underline{M}$$

由Gronwall-Bellman不等式有

$$\sqrt{V(\xi)} \le e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))} + \frac{c_2}{\sqrt{c_1}} M \int_{t_0}^t e^{-\frac{\omega_e}{c_2}(t-s)} ds$$

设初始估计误差满足:

$$\left| \hat{E}(t_0) \right| < \rho_0$$

则有

$$\|\xi(t_0)\| \le \overline{\mu}_e(\omega_e)\rho_0$$

$$\overline{\mu}_e(\omega_e) \triangleq \max \left\{ 1, \frac{1}{\omega_e^{n-1}} \right\}$$



设初始估计误差满足:

$$\left| \hat{E}(t_0) \right| < \rho_0$$

则有

$$\left\| \xi(t_0) \right\| \leq \overline{\mu}_e(\omega_e) \rho_0$$

$$\overline{\mu}_e(\omega_e) \triangleq \max\left\{1, \omega_e^n\right\}$$

则对
$$\sqrt{V(\xi)} \le e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))} + \frac{c_2}{\sqrt{c_1}} M \int_{t_0}^t e^{-\frac{\omega_e}{c_2}(t-s)} ds$$
 的右端第1项得如下分析结果

$$e^{-\frac{\omega_e}{2c_2}(t-t_0)} \sqrt{V(\xi(t_0))} \leq \frac{1}{\omega_e^{n+1}} \sqrt{c_2} \rho_0 \overline{\mu}_e(\omega_e) \leq \max\left\{\frac{1}{\omega_e^{n+1}}, \frac{1}{\omega_e}\right\} \sqrt{c_2} \rho_0.$$



又由于
$$\xi_i = \omega_e^{n+1-i} \hat{E}_i, \hat{E}_i = \frac{1}{\omega_e^{n+1-i}} \xi_i,$$
 则

$$\exists \forall t \geq T \triangleq t_0 + 2(n+1)c_2 \max \left\{ \frac{\ln \omega_e}{\omega_e}, 0 \right\}, \ \omega_e > 0, \ i = 1, 2, ..., n+1$$

$$\left| \hat{E}_i(t) \right| = \frac{1}{\omega_e^{n+1-i}} \left\| \xi_i(t) \right\|$$

$$\leq \frac{1}{\omega_e^{n+1-i}} \left[e^{-\frac{\omega_e}{2c_2}t} \sqrt{\frac{V(\xi(t_0))}{c_1}} + \frac{c_2}{c_1} M \int_{t_0}^t e^{-\frac{\omega_e}{2c_2}(t-s)} ds \right]$$

$$\leq \frac{1}{\omega_e^{n+1-i}} \left[\max \left\{ \frac{1}{\omega_e^{n+1}}, \frac{1}{\omega_e} \right\} \sqrt{c_2} \rho_0 + \frac{c_2}{c_1} M \frac{2c_2}{\omega_e} \right]$$

因此
$$|X_i - \hat{X}_i| \le O\left(\frac{1}{\omega_e^k}\right)$$
 $k = n + 2 - i, \forall t \ge T_1 > t_0, \omega_e > 0, i = 1, 2, ..., n + 1.$ 证毕。

线性自抗扰控制的跟踪误差分析: 定理4



Uncertain
$$\begin{cases} \dot{X}_1 = X_2 \\ \cdots \\ \dot{X}_{n-1} = X_n \\ \dot{X}_n = X_{n+1} + \overline{B}(t)U(t) \end{cases}$$

总扰动: $X_{n+1} = F(X,t) + d(t) + (B(X,t) - \overline{B}(t))U(t)$

Linear
$$\begin{cases} \dot{\hat{X}}_{1} = \hat{X}_{2} - l_{1}(\hat{X}_{1} - Y) \\ \dots \\ \dot{\hat{X}}_{n} = -l_{n}(\hat{X}_{1} - Y) - \hat{X}_{n+1} + \overline{B}(t)U(t) \\ \dot{\hat{X}}_{n+1} = -l_{n+1}(\hat{X}_{1} - Y) \end{cases}$$

控制器设计:
$$U(t) = \begin{cases} 0, t < T_1 \\ \overline{B}^{-1}(-k_1\hat{X}_1 - k_2\hat{X}_2 \cdots - k_n\hat{X}_n - \hat{X}_{n+1}), t \ge T \end{cases}$$

定理4: 考虑上述闭环系统,存在 ω_e^* , 使得 $\forall \omega_e \geq \omega_e^*$:

$$||y(t)-y^*(t)|| \le \gamma_1^* \frac{\max\{\ln(\rho_2\omega_e),\ln\omega_e,1\}}{\omega_e}, \quad \forall t \ge t_0 \longrightarrow \mathbf{m}$$
 ணč性能: **接近预期轨迹**

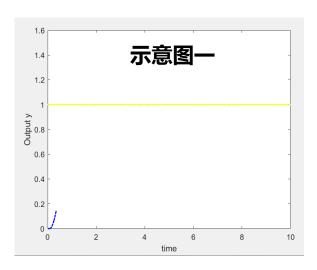
其中 $(\gamma_1^*, \gamma_3^*, \gamma_2^*)$ 为正常数。



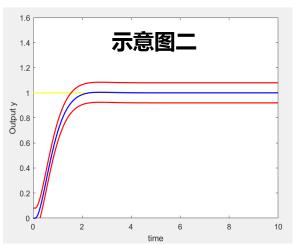
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 瞬态性能:接近预期轨迹

其中 $(\gamma_1^*,\gamma_3^*,\gamma_2^*)$ 为正常数。



y → r (t → ∞), 但超调、震荡...



输出曲线在理想轨线(蓝线)的小范围内(红线之间)波动

$$y^* = x_1^*$$

$$\begin{bmatrix} \dot{x}_1^* \\ \dot{x}_2^* \\ \vdots \\ \dot{x}_n^* \end{bmatrix} = A_k \begin{bmatrix} x_1^* \\ x_2^* \\ \vdots \\ x_n^* \end{bmatrix} - \begin{bmatrix} r \\ r^{(1)} \\ \vdots \\ r^{(n-1)} \end{bmatrix} + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ r^{(n)} \end{bmatrix}$$

$$A_k = \begin{bmatrix} 0 & 1 & \cdots & 0 \\ \vdots & \ddots & \ddots & \vdots \\ 0 & \cdots & 0 & 1 \\ k_{1,c} & k_{2,c} & \cdots & k_{n,c} \end{bmatrix}$$