

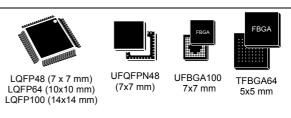
# STM32L073x8 STM32L073xB STM32L073xZ

Ultra-low-power 32-bit MCU Arm<sup>®</sup>-based Cortex<sup>®</sup>-M0+, up to 192KB Flash, 20KB SRAM, 6KB EEPROM, LCD, USB, ADC, DACs

Datasheet - production data

### **Features**

- Ultra-low-power platform
  - 1.65 V to 3.6 V power supply
  - -40 to 125 °C temperature range
  - 0.29 μA Standby mode (3 wakeup pins)
  - 0.43 µA Stop mode (16 wakeup lines)
  - 0.86 µA Stop mode + RTC + 20-Kbyte RAM retention
  - Down to 93 μA/MHz in Run mode
  - 5 µs wakeup time (from Flash memory)
  - 41 μA 12-bit ADC conversion at 10 ksps
- Core: Arm<sup>®</sup> 32-bit Cortex<sup>®</sup>-M0+ with MPU
- E COLLIN OF BUILDING WILL
  - From 32 kHz up to 32 MHz max.
  - 0.95 DMIPS/MHz
- Memories
  - Up to 192-Kbyte Flash memory with ECC (2 banks with read-while-write capability)
  - 20-Kbyte RAM
  - 6 Kbytes of data EEPROM with ECC
  - 20-byte backup register
  - Sector protection against R/W operation
- Up to 84 fast I/Os (78 I/Os 5V tolerant)
- Reset and supply management
  - Ultra-safe, low-power BOR (brownout reset) with 5 selectable thresholds
  - Ultra-low-power POR/PDR
  - Programmable voltage detector (PVD)
- Clock sources
  - 1 to 25 MHz crystal oscillator
  - 32 kHz oscillator for RTC with calibration
  - High speed internal 16 MHz factory-trimmed RC (+/- 1%)
  - Internal low-power 37 kHz RC
  - Internal multispeed low-power 65 kHz to 4.2 MHz RC
  - Internal self calibration of 48 MHz RC for USB
  - PLL for CPU clock
- Pre-programmed bootloader
  - USB, USART supported
- Development support
  - Serial wire debug supported
- LCD driver for up to 4x52 or 8x48 segments



- Support contrast adjustment
- Support blinking mode
- Step-up converted on board
- Rich Analog peripherals
  - 12-bit ADC 1.14 Msps up to 16 channels (down to 1.65 V)
  - 2 x 12-bit channel DACs with output buffers (down to 1.8 V)
  - 2x ultra-low-power comparators (window mode and wake up capability, down to 1.65 V)
- Up to 24 capacitive sensing channels supporting touchkey, linear and rotary touch sensors
- 7-channel DMA controller, supporting ADC, SPI, I2C, USART, DAC, Timers
- 11x peripheral communication interfaces
  - 1x USB 2.0 crystal-less, battery charging detection and LPM
  - 4x USART (2 with ISO 7816, IrDA), 1x UART (low power)
  - Up to 6x SPI 16 Mbits/s
  - 3x I2C (2 with SMBus/PMBus)
- 11x timers: 2x 16-bit with up to 4 channels, 2x 16-bit with up to 2 channels, 1x 16-bit ultra-low-power timer, 1x SysTick, 1x RTC, 2x 16-bit basic for DAC, and 2x watchdogs (independent/window)
- CRC calculation unit, 96-bit unique ID
- True RNG and firewall protection
- All packages are ECOPACK2

#### Table 1. Device summary

Reference	Part number
STM32L073x8	STM32L073V8
STM32L073xB	STM32L073VB, STM32L073RB, STM32L073CB
STM32L073xZ	STM32L073VZ, STM32L073RZ, STM32L073CZ

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STM32L073xx Introduction

### 1 Introduction

The ultra-low-power STM32L073xx are offered in 6 different package types from 48 to 100 pins. Depending on the device chosen, different sets of peripherals are included, the description below gives an overview of the complete range of peripherals proposed in this family.

These features make the ultra-low-power STM32L073xx microcontrollers suitable for a wide range of applications:

- Gas/water meters and industrial sensors
- · Healthcare and fitness equipment
- · Remote control and user interface
- PC peripherals, gaming, GPS equipment
- Alarm system, wired and wireless sensors, video intercom

This STM32L073xx datasheet should be read in conjunction with the STM32L0x3xx reference manual (RM0367).

For information on the Arm<sup>®(a)</sup> Cortex<sup>®</sup>-M0+ core please refer to the Cortex<sup>®</sup>-M0+ Technical Reference Manual, available from the www.arm.com website.

Figure 1 shows the general block diagram of the device family.

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Description STM32L073xx

### 2 Description

The ultra-low-power STM32L073xx microcontrollers incorporate the connectivity power of the universal serial bus (USB 2.0 crystal-less) with the high-performance Arm Cortex-M0+32-bit RISC core operating at a 32 MHz frequency, a memory protection unit (MPU), high-speed embedded memories (up to 192 Kbytes of Flash program memory, 6 Kbytes of data EEPROM and 20 Kbytes of RAM) plus an extensive range of enhanced I/Os and peripherals.

The STM32L073xx devices provide high power efficiency for a wide range of performance. It is achieved with a large choice of internal and external clock sources, an internal voltage adaptation and several low-power modes.

The STM32L073xx devices offer several analog features, one 12-bit ADC with hardware oversampling, two DACs, two ultra-low-power comparators, several timers, one low-power timer (LPTIM), four general-purpose 16-bit timers and two basic timer, one RTC and one SysTick which can be used as timebases. They also feature two watchdogs, one watchdog with independent clock and window capability and one window watchdog based on bus clock.

Moreover, the STM32L073xx devices embed standard and advanced communication interfaces: up to three I2Cs, two SPIs, one I2S, four USARTs, a low-power UART (LPUART), and a crystal-less USB. The devices offer up to 24 capacitive sensing channels to simply add touch sensing functionality to any application.

The STM32L073xx also include a real-time clock and a set of backup registers that remain powered in Standby mode.

Finally, their integrated LCD controller has a built-in LCD voltage generator that allows to drive up to 8 multiplexed LCDs with contrast independent of the supply voltage.

The ultra-low-power STM32L073xx devices operate from a 1.8 to 3.6 V power supply (down to 1.65 V at power down) with BOR and from a 1.65 to 3.6 V power supply without BOR option. They are available in the -40 to +125 °C temperature range. A comprehensive set of power-saving modes allows the design of low-power applications.



STM32L073xx Description

### 2.1 Device overview

Table 2. Ultra-low-power STM32L073xxx device features and peripheral counts

Peripheral		STM32L073 V8	STM32L073 CB	STM32L073 VB	STM32L073 RB	STM32L073 CZ	STM32L073 VZ	STM32L073 RZ		
Flash (Kby	tes)	64 Kbytes 128 Kbytes 192 Kbytes								
Data EEPR	OM (Kbytes)	3 Kbytes	3 Kbytes 6 Kbytes							
RAM (Kbyt	es)				20 Kbytes					
	General- purpose		4							
Timers	Basic		2							
	LPTIMER				1					
RTC/SYST	ICK/IWDG/WWDG				1/1/1/1					
	SPI/I2S				6(4) <sup>(1)</sup> /1					
Commu-	I <sup>2</sup> C				3					
nication	USART				4					
interraces	LPUART				1					
	USB/(VDD_USB)		1/(1)							
GPIOs		84	37	84	51 <sup>(2)</sup>	37	84	51 <sup>(2)</sup>		
Clocks: HSE/LSE/H	ISI/MSI/LSI	1/1/1/1/1								
12-bit sync Number of	hronized ADC channels	1 16	1 10	1 16 <sup>(2)</sup>		1 10	1 16 <sup>(2)</sup>			
12-bit DAC Number of		2 2								
LCD COM x SEC	3	1 4x52 or 8x48	1 4x18	1 4x52 or 8x48	1 4x32 or 8x28 <sup>(2)</sup>	1 4x18	1 4x52 or 8x48	1 4x32 or 8x28 <sup>(2)</sup>		
Comparato	ors	2								
Capacitive sensing channels		24	17	24	24 <sup>(2)</sup>	17	24	24 <sup>(2)</sup>		
Max. CPU f	requency	32 MHz								
Operating voltage		1.8 V to 3.6 V (down to 1.65 V at power-down) with BOR option 1.65 to 3.6 V without BOR option								
Operating temperatures					mperature: –40 mperature: –40					
Packages		LQFP100, UFBGA100	LQFP48, UFQFPN48	LQFP100, UFBGA100	LQFP64, TFBGA64	LQFP48, UFQFPN48	LQFP100, UFBGA100	LQFP64, TFBGA64		

<sup>1. 4</sup> SPI interfaces are USARTs operating in SPI master mode.

<sup>2.</sup> TFBGA64 has one GPIO, one ADC input, one capacitive sensing channel and one COMxSEG (4x31 or 8x27) less than LQFP64.

Description STM32L073xx

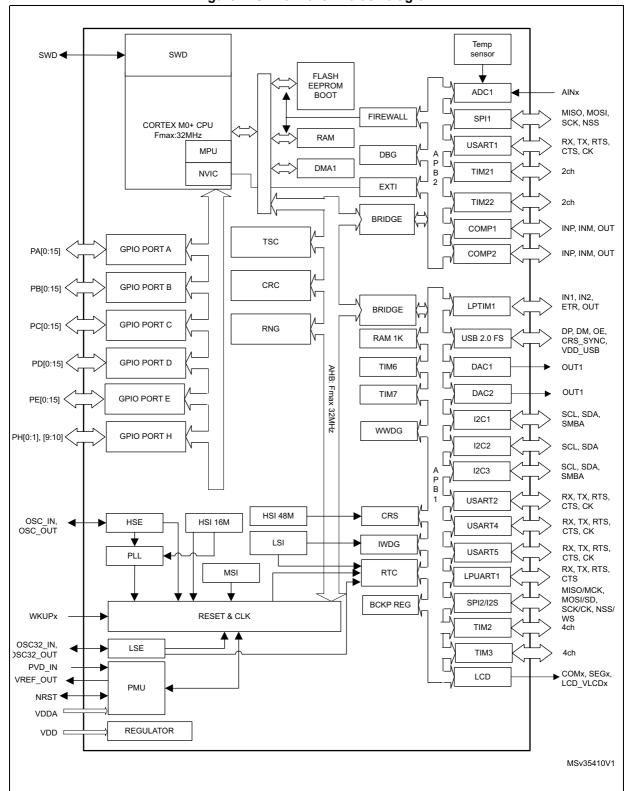


Figure 1. STM32L073xx block diagram



STM32L073xx Description

### 2.2 Ultra-low-power device continuum

The ultra-low-power family offers a large choice of core and features, from 8-bit proprietary core up to Arm<sup>®</sup> Cortex<sup>®</sup>-M4, including Arm<sup>®</sup> Cortex<sup>®</sup>-M3 and Arm<sup>®</sup> Cortex<sup>®</sup>-M0+. The STM32Lx series are the best choice to answer your needs in terms of ultra-low-power features. The STM32 ultra-low-power series are the best solution for applications such as gaz/water meter, keyboard/mouse or fitness and healthcare application. Several built-in features like LCD drivers, dual-bank memory, low-power run mode, operational amplifiers, 128-bit AES, DAC, crystal-less USB and many other definitely help you building a highly cost optimized application by reducing BOM cost. STMicroelectronics, as a reliable and long-term manufacturer, ensures as much as possible pin-to-pin compatibility between all STM8Lx and STM32Lx on one hand, and between all STM32Lx and STM32Fx on the other hand. Thanks to this unprecedented scalability, your legacy application can be upgraded to respond to the latest market feature and efficiency requirements.



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### 3 Functional overview

### 3.1 Low-power modes

The ultra-low-power STM32L073xx support dynamic voltage scaling to optimize its power consumption in Run mode. The voltage from the internal low-drop regulator that supplies the logic can be adjusted according to the system's maximum operating frequency and the external voltage supply.

There are three power consumption ranges:

- Range 1 (V<sub>DD</sub> range limited to 1.71-3.6 V), with the CPU running at up to 32 MHz
- Range 2 (full V<sub>DD</sub> range), with a maximum CPU frequency of 16 MHz
- Range 3 (full V<sub>DD</sub> range), with a maximum CPU frequency limited to 4.2 MHz

Seven low-power modes are provided to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

#### Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs. Sleep mode power consumption at 16 MHz is about 1 mA with all peripherals off.

#### Low-power run mode

This mode is achieved with the multispeed internal (MSI) RC oscillator set to the low-speed clock (max 131 kHz), execution from SRAM or Flash memory, and internal regulator in low-power mode to minimize the regulator's operating current. In Low-power run mode, the clock frequency and the number of enabled peripherals are both limited.

#### • Low-power sleep mode

This mode is achieved by entering Sleep mode with the internal voltage regulator in low-power mode to minimize the regulator's operating current. In Low-power sleep mode, both the clock frequency and the number of enabled peripherals are limited; a typical example would be to have a timer running at 32 kHz.

When wakeup is triggered by an event or an interrupt, the system reverts to the Run mode with the regulator on.

#### Stop mode with RTC

The Stop mode achieves the lowest power consumption while retaining the RAM and register contents and real time clock. All clocks in the  $V_{CORE}$  domain are stopped, the PLL, MSI RC, HSE crystal and HSI RC oscillators are disabled. The LSE or LSI is still running. The voltage regulator is in the low-power mode.

Some peripherals featuring wakeup capability can enable the HSI RC during Stop mode to detect their wakeup condition.

The device can be woken up from Stop mode by any of the EXTI line, in 3.5 µs, the processor can serve the interrupt or resume the code. The EXTI line source can be any GPIO. It can be the PVD output, the comparator 1 event or comparator 2 event (if internal reference voltage is on), it can be the RTC alarm/tamper/timestamp/wakeup events, the USB/USART/I2C/LPUART/LPTIMER wakeup events.



#### Stop mode without RTC

The Stop mode achieves the lowest power consumption while retaining the RAM and register contents. All clocks are stopped, the PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillators are disabled.

Some peripherals featuring wakeup capability can enable the HSI RC during Stop mode to detect their wakeup condition.

The voltage regulator is in the low-power mode. The device can be woken up from Stop mode by any of the EXTI line, in 3.5  $\mu$ s, the processor can serve the interrupt or resume the code. The EXTI line source can be any GPIO. It can be the PVD output, the comparator 1 event or comparator 2 event (if internal reference voltage is on). It can also be wakened by the USB/USART/I2C/LPUART/LPTIMER wakeup events.

#### Standby mode with RTC

The Standby mode is used to achieve the lowest power consumption and real time clock. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSE crystal and HSI RC oscillators are also switched off. The LSE or LSI is still running. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32 KHz oscillator, RCC\_CSR register).

The device exits Standby mode in 60 µs when an external reset (NRST pin), an IWDG reset, a rising edge on one of the three WKUP pins, RTC alarm (Alarm A or Alarm B), RTC tamper event, RTC timestamp event or RTC Wakeup event occurs.

#### Standby mode without RTC

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillators are also switched off. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32 KHz oscillator, RCC\_CSR register).

The device exits Standby mode in 60 µs when an external reset (NRST pin) or a rising edge on one of the three WKUP pin occurs.

Note:

The RTC, the IWDG, and the corresponding clock sources are not stopped automatically by entering Stop or Standby mode. The LCD is not stopped automatically by entering Stop mode.



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Table 3. Functionalities depending on the operating power supply range

Operating newer cumply	Functionalities depending on the operating power supply range					
Operating power supply range <sup>(1)</sup>	DAC and ADC operation	, , , , , ,				
V <sub>DD</sub> = 1.65 to 1.71 V	ADC only, conversion time up to 570 ksps	Range 2 or range 3	Not functional			
V <sub>DD</sub> = 1.71 to 1.8 V <sup>(2)</sup>	ADC only, conversion time up to 1.14 Msps	Range 1, range 2 or range 3	Functional <sup>(3)</sup>			
V <sub>DD</sub> = 1.8 to 2.0 V <sup>(2)</sup>	Conversion time up to 1.14 Msps	Range1, range 2 or range 3	Functional <sup>(3)</sup>			
V <sub>DD</sub> = 2.0 to 2.4 V	Conversion time up to 1.14 Msps	Range 1, range 2 or range 3	Functional <sup>(3)</sup>			
V <sub>DD</sub> = 2.4 to 3.6 V	Conversion time up to 1.14 Msps	Range 1, range 2 or range 3	Functional <sup>(3)</sup>			

GPIO speed depends on V<sub>DD</sub> voltage range. Refer to Table 62: I/O AC characteristics for more information about I/O speed.

Table 4. CPU frequency range depending on dynamic voltage scaling

CPU frequency range	Dynamic voltage scaling range
16 MHz to 32 MHz (1ws) 32 kHz to 16 MHz (0ws)	Range 1
8 MHz to 16 MHz (1ws) 32 kHz to 8 MHz (0ws)	Range 2
32 kHz to 4.2 MHz (0ws)	Range 3

CPU frequency changes from initial to final must respect "fcpu initial <4\*fcpu final". It must also respect 5
μs delay between two changes. For example to switch from 4.2 MHz to 32 MHz, you can switch from 4.2
MHz to 16 MHz, wait 5 μs, then switch from 16 MHz to 32 MHz.</li>

<sup>3.</sup> To be USB compliant from the I/O voltage standpoint, the minimum  $\rm V_{\rm DD\_USB}$  is 3.0 V.

Table 5. Functionalities depending on the working mode (from Run/active down to standby) <sup>(1)(2)</sup>

			Low-	Low-		Stop	Standby	
IPs	Run/Active	Sleep	power run	power sleep		Wakeup capability		Wakeup capability
CPU	Y		Y					
Flash memory	0	0	0	0			-	
RAM	Y	Υ	Υ	Υ	Υ		I	
Backup registers	Y	Y	Y	Y	Υ		Υ	
EEPROM	0	0	0	0			I	
Brown-out reset (BOR)	0	0	0	0	0	0	0	0
DMA	0	0	0	0				
Programmable Voltage Detector (PVD)	0	0	0	0	0	0	-	
Power-on/down reset (POR/PDR)	Y	Y	Y	Υ	Υ	Y	Υ	Y
High Speed Internal (HSI)	0	0			(3)			
High Speed External (HSE)	0	0	0	0				
Low Speed Internal (LSI)	0	0	0	0	0		0	
Low Speed External (LSE)	0	0	0	0	0		0	
Multi-Speed Internal (MSI)	0	0	Y	Y				
Inter-Connect Controller	Y	Y	Y	Y	Υ		ł	
RTC	0	0	0	0	0	0	0	
RTC Tamper	0	0	0	0	0	0	0	0
Auto WakeUp (AWU)	0	0	0	0	0	0	0	0
LCD	0	0	0	0	0			
USB	0	0				0		
USART	0	0	0	0	O <sup>(4)</sup>	0	-	
LPUART	0	0	0	0	O <sup>(4)</sup>	0	-	
SPI	0	0	0	0			-	
I2C	0	0			O <sup>(5)</sup>	0	-	
ADC	0	0					-	

Table 5. Functionalities depending on the working mode (from Run/active down to standby) (continued)(1)(2)

			Low-	Low-	Stop		Standby	
IPs	Run/Active	Sleep	power run	power sleep		Wakeup capability		Wakeup capability
DAC	0	0	0	0	0			
Temperature sensor	0	0	0	0	0			
Comparators	0	0	0	0	0	0		
16-bit timers	0	0	0	0				
LPTIMER	0	0	0	0	0	0		
IWDG	0	0	0	0	0	0	0	0
WWDG	0	0	0	0				
Touch sensing controller (TSC)	0	0						
SysTick Timer	0	0	0	0				
GPIOs	0	0	0	0	0	0		2 pins
Wakeup time to Run mode	0 μs	0.36 µs	3 µs	32 µs	3.5 µs		50 µs	
					0.4 μA (No RTC) V <sub>DD</sub> =1.8 V			28 μΑ (No ) V <sub>DD</sub> =1.8 V
Consumption	Down to 140 μΑ/ΜΗz	Down to 37 µA/MHz	Down to	Down to	0.8 μA (with RTC) V <sub>DD</sub> =1.8 V		0.65 μA (with RTC) V <sub>DD</sub> =1.8 V	
V <sub>DD</sub> =1.8 to 3.6 V (Typ)	(from Flash memory)	1 '	4.5 μA	0.4 μA (No RTC) V <sub>DD</sub> =3.0 V		0.29 μA (No RTC) V <sub>DD</sub> =3.0 V		
						(with RTC) <sub>DD</sub> =3.0 V		5 μA (with ) V <sub>DD</sub> =3.0 V

<sup>1.</sup> Legend:

- 2. The consumption values given in this table are preliminary data given for indication. They are subject to slight changes.
- Some peripherals with wakeup from Stop capability can request HSI to be enabled. In this case, HSI is woken up by the peripheral, and only feeds the peripheral which requested it. HSI is automatically put off when the peripheral does not need it anymore.
- 4. UART and LPUART reception is functional in Stop mode. It generates a wakeup interrupt on Start. To generate a wakeup on address match or received frame event, the LPUART can run on LSE clock while the UART has to wake up or keep
- 5. I2C address detection is functional in Stop mode. It generates a wakeup interrupt in case of address match. It will wake up the HSI during reception.

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<sup>&</sup>quot;Y" = Yes (enable).
"O" = Optional can be enabled/disabled by software)
"-" = Not available

### 3.2 Interconnect matrix

Several peripherals are directly interconnected. This allows autonomous communication between peripherals, thus saving CPU resources and power consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep, Low-power run, Low-power sleep and Stop modes.

Table 6. STM32L0xx peripherals interconnect matrix

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low- power run	Low- power sleep	Stop
COMPx	TIM2,TIM21, TIM22	Timer input channel, trigger from analog signals comparison	Y	Y	Υ	Y	-
COIVIFX	LPTIM	Timer input channel, trigger from analog signals comparison	Υ	Υ	Y	Y	Υ
TIMx	TIMx	Timer triggered by other timer	Y	Y	Υ	Y	-
RTC	TIM21	Timer triggered by Auto wake-up		Y	Υ	Y	-
RIC	LPTIM	Timer triggered by RTC event	Y	Y	Υ	Y	Υ
All clock source	TIMx	Clock source used as input channel for RC measurement and trimming	Y	Y	Υ	Y	-
USB	CRS/HSI48	the clock recovery system trims the HSI48 based on USB SOF	Υ	Y	-	-	-
	TIM3	USB_SOF is channel input for calibration	Y	Y	1	-	-
	TIMx	Timer input channel and trigger	Y	Y	Y	Y	-
GPIO	LPTIM	Timer input channel and trigger	Y	Y	Y	Y	Υ
	ADC,DAC	Conversion trigger	Υ	Υ	Y	Υ	-

### 3.3 Arm® Cortex®-M0+ core with MPU

The Cortex-M0+ processor is an entry-level 32-bit Arm Cortex processor designed for a broad range of embedded applications. It offers significant benefits to developers, including:

- a simple architecture that is easy to learn and program
- ultra-low power, energy-efficient operation
- excellent code density
- deterministic, high-performance interrupt handling
- upward compatibility with Cortex-M processor family
- platform security robustness, with integrated Memory Protection Unit (MPU).

The Cortex-M0+ processor is built on a highly area and power optimized 32-bit processor core, with a 2-stage pipeline Von Neumann architecture. The processor delivers exceptional energy efficiency through a small but powerful instruction set and extensively optimized design, providing high-end processing hardware including a single-cycle multiplier.

The Cortex-M0+ processor provides the exceptional performance expected of a modern 32-bit architecture, with a higher code density than other 8-bit and 16-bit microcontrollers.

Owing to its embedded Arm core, the STM32L073xx are compatible with all Arm tools and software.

### **Nested vectored interrupt controller (NVIC)**

The ultra-low-power STM32L073xx embed a nested vectored interrupt controller able to handle up to 32 maskable interrupt channels and 4 priority levels.

The Cortex-M0+ processor closely integrates a configurable Nested Vectored Interrupt Controller (NVIC), to deliver industry-leading interrupt performance. The NVIC:

- includes a Non-Maskable Interrupt (NMI)
- provides zero jitter interrupt option
- provides four interrupt priority levels

The tight integration of the processor core and NVIC provides fast execution of Interrupt Service Routines (ISRs), dramatically reducing the interrupt latency. This is achieved through the hardware stacking of registers, and the ability to abandon and restart load-multiple and store-multiple operations. Interrupt handlers do not require any assembler wrapper code, removing any code overhead from the ISRs. Tail-chaining optimization also significantly reduces the overhead when switching from one ISR to another.

To optimize low-power designs, the NVIC integrates with the sleep modes, that include a deep sleep function that enables the entire device to enter rapidly stop or standby mode.

This hardware block provides flexible interrupt management features with minimal interrupt latency.

### 3.4 Reset and supply management

### 3.4.1 Power supply schemes

 V<sub>DD</sub> = 1.65 to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through V<sub>DD</sub> pins.

- V<sub>SSA</sub>, V<sub>DDA</sub> = 1.65 to 3.6 V: external analog power supplies for ADC reset blocks, RCs and PLL. V<sub>DDA</sub> and V<sub>SSA</sub> must be connected to V<sub>DD</sub> and V<sub>SS</sub>, respectively.
- V<sub>DD\_USB</sub> = 1.65 to 3.6V: external power supply for USB transceiver, USB\_DM (PA11) and USB\_DP (PA12). To guarantee a correct voltage level for USB communication V<sub>DD\_USB</sub> must be above 3.0V. If USB is not used this pin must be tied to V<sub>DD</sub> or V<sub>SS</sub>. On packages without VDD\_USB pin, V<sub>DD\_USB</sub> voltage is internally connected to V<sub>DD</sub> voltage.

### 3.4.2 Power supply supervisor

The devices have an integrated ZEROPOWER power-on reset (POR)/power-down reset (PDR) that can be coupled with a brownout reset (BOR) circuitry.

Two versions are available:

- The version with BOR activated at power-on operates between 1.8 V and 3.6 V.
- The other version without BOR operates between 1.65 V and 3.6 V.

After the  $V_{DD}$  threshold is reached (1.65 V or 1.8 V depending on the BOR which is active or not at power-on), the option byte loading process starts, either to confirm or modify default thresholds, or to disable the BOR permanently: in this case, the VDD min value becomes 1.65 V (whatever the version, BOR active or not, at power-on).

When BOR is active at power-on, it ensures proper operation starting from 1.8 V whatever the power ramp-up phase before it reaches 1.8 V. When BOR is not active at power-up, the power ramp-up should guarantee that 1.65 V is reached on  $V_{DD}$  at least 1 ms after it exits the POR area.

Five BOR thresholds are available through option bytes, starting from 1.8 V to 3 V. To reduce the power consumption in Stop mode, it is possible to automatically switch off the internal reference voltage ( $V_{REFINT}$ ) in Stop mode. The device remains in reset mode when  $V_{DD}$  is below a specified threshold,  $V_{POR/PDR}$  or  $V_{BOR}$ , without the need for any external reset circuit.

Note: The start-up time at power-on is typically 3.3 ms when BOR is active at power-up, the start-up time at power-on can be decreased down to 1 ms typically for devices with BOR inactive at power-up.

The devices feature an embedded programmable voltage detector (PVD) that monitors the  $V_{DD/VDDA}$  power supply and compares it to the  $V_{PVD}$  threshold. This PVD offers 7 different levels between 1.85 V and 3.05 V, chosen by software, with a step around 200 mV. An interrupt can be generated when  $V_{DD/VDDA}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD/VDDA}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

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### 3.4.3 Voltage regulator

The regulator has three operation modes: main (MR), low power (LPR) and power down.

- MR is used in Run mode (nominal regulation)
- LPR is used in the Low-power run, Low-power sleep and Stop modes
- Power down is used in Standby mode. The regulator output is high impedance, the kernel circuitry is powered down, inducing zero consumption but the contents of the registers and RAM are lost except for the standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE crystal 32 KHz oscillator, RCC CSR).

### 3.5 Clock management

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

### Clock prescaler

To get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler.

#### Safe clock switching

Clock sources can be changed safely on the fly in Run mode through a configuration register.

### • Clock management

To reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.

#### System clock source

Three different clock sources can be used to drive the master clock SYSCLK:

- 1-25 MHz high-speed external crystal (HSE), that can supply a PLL
- 16 MHz high-speed internal RC oscillator (HSI), trimmable by software, that can supply a PLLMultispeed internal RC oscillator (MSI), trimmable by software, able to generate 7 frequencies (65 kHz, 131 kHz, 262 kHz, 524 kHz, 1.05 MHz, 2.1 MHz, 4.2 MHz). When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be trimmed by software down to a ±0.5% accuracy.

#### Auxiliary clock source

Two ultra-low-power clock sources that can be used to drive the LCD controller and the real-time clock:

- 32.768 kHz low-speed external crystal (LSE)
- 37 kHz low-speed internal RC (LSI), also used to drive the independent watchdog.
   The LSI clock can be measured using the high-speed internal RC oscillator for greater precision.

#### RTC and LCD clock source

The LSI, LSE or HSE sources can be chosen to clock the RTC and the LCD, whatever the system clock.

#### USB clock source

A 48 MHz clock trimmed through the USB SOF or LSE supplies the USB interface.



#### Startup clock

After reset, the microcontroller restarts by default with an internal 2.1 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.

### Clock security system (CSS)

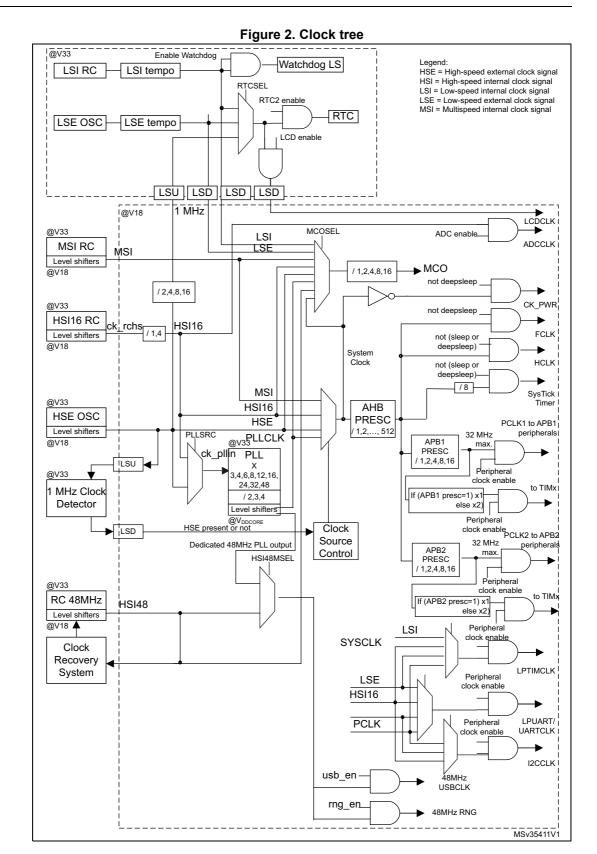
This feature can be enabled by software. If an HSE clock failure occurs, the master clock is automatically switched to HSI and a software interrupt is generated if enabled. Another clock security system can be enabled, in case of failure of the LSE it provides an interrupt or wakeup event which is generated if enabled.

Clock-out capability (MCO: microcontroller clock output)
 It outputs one of the internal clocks for external use by the application.

Several prescalers allow the configuration of the AHB frequency, each APB (APB1 and APB2) domains. The maximum frequency of the AHB and the APB domains is 32 MHz. See *Figure 2* for details on the clock tree.



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### 3.6 Low-power real-time clock and backup registers

The real time clock (RTC) and the 5 backup registers are supplied in all modes including standby mode. The backup registers are five 32-bit registers used to store 20 bytes of user application data. They are not reset by a system reset, or when the device wakes up from Standby mode.

The RTC is an independent BCD timer/counter. Its main features are the following:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format
- Automatically correction for 28, 29 (leap year), 30, and 31 day of the month
- Two programmable alarms with wake up from Stop and Standby mode capability
- Periodic wakeup from Stop and Standby with programmable resolution and period
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 1 ppm resolution, to compensate for quartz crystal inaccuracy
- 2 anti-tamper detection pins with programmable filter. The MCU can be woken up from Stop and Standby modes on tamper event detection.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event. The MCU can be woken up from Stop and Standby modes on timestamp event detection.

The RTC clock sources can be:

- A 32.768 kHz external crystal
- A resonator or oscillator
- The internal low-power RC oscillator (typical frequency of 37 kHz)
- The high-speed external clock

### 3.7 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions, and can be individually remapped using dedicated alternate function registers. All GPIOs are high current capable. Each GPIO output, speed can be slowed (40 MHz, 10 MHz, 2 MHz, 400 kHz). The alternate function configuration of I/Os can be locked if needed following a specific sequence in order to avoid spurious writing to the I/O registers. The I/O controller is connected to a dedicated IO bus with a toggling speed of up to 32 MHz.

#### **Extended interrupt/event controller (EXTI)**

The extended interrupt/event controller consists of 29 edge detector lines used to generate interrupt/event requests. Each line can be individually configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 84 GPIOs can be connected to the 16 configurable interrupt/event lines. The 13 other lines are connected to PVD, RTC, USB, USARTs, I2C, LPUART, LPTIMER or comparator events.



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### 3.8 Memories

The STM32L073xx devices have the following features:

• 20 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states. With the enhanced bus matrix, operating the RAM does not lead to any performance penalty during accesses to the system bus (AHB and APB buses).

- The non-volatile memory is divided into three arrays:
  - 64, 128 or 192 Kbytes of embedded Flash program memory
  - 6 Kbytes of data EEPROM
  - Information block containing 32 user and factory options bytes plus 8 Kbytes of system memory

Flash program and data EEPROM are divided into two banks. This allows writing in one bank while running code or reading data from the other bank.

The user options bytes are used to write-protect or read-out protect the memory (with 4 Kbyte granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no protection
- Level 1: memory readout protected.
  - The Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected
- Level 2: chip readout protected, debug features (Cortex-M0+ serial wire) and boot in RAM selection disabled (debugline fuse)

The firewall protects parts of code/data from access by the rest of the code that is executed outside of the protected area. The granularity of the protected code segment or the non-volatile data segment is 256 bytes (Flash memory or EEPROM) against 64 bytes for the volatile data segment (RAM).

The whole non-volatile memory embeds the error correction code (ECC) feature.

### 3.9 Boot modes

At startup, BOOT0 pin and nBOOT1 option bit are used to select one of three boot options:

- Boot from Flash memory
- Boot from System memory
- Boot from embedded RAM

The boot loader is located in System memory. It is used to reprogram the Flash memory by using USB (PA11, PA12), USART1(PA9, PA10) or USART2(PA2, PA3). See STM32<sup>™</sup> microcontroller system memory boot mode AN2606 for details.

### 3.10 Direct memory access (DMA)

The flexible 7-channel, general-purpose DMA is able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA controller supports circular buffer management, avoiding the generation of interrupts when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with software trigger support for each channel. Configuration is done by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI, I<sup>2</sup>C, USART, LPUART, general-purpose timers, DAC, and ADC.

### 3.11 Liquid crystal display (LCD)

The LCD drives up to 8 common terminals and 48 segment terminals to drive up to 384 pixels.

- Internal step-up converter to guarantee functionality and contrast control irrespective of V<sub>DD</sub>. This converter can be deactivated, in which case the V<sub>LCD</sub> pin is used to provide the voltage to the LCD
- Supports static, 1/2, 1/3, 1/4 and 1/8 duty
- Supports static, 1/2, 1/3 and 1/4 bias
- Phase inversion to reduce power consumption and EMI
- Up to 8 pixels can be programmed to blink
- Unneeded segments and common pins can be used as general I/O pins
- LCD RAM can be updated at any time owing to a double-buffer
- The LCD controller can operate in Stop mode
- V<sub>I CD</sub> rails decoupling capability

### 3.12 Analog-to-digital converter (ADC)

A native 12-bit, extended to 16-bit through hardware oversampling, analog-to-digital converter is embedded into STM32L073xx device. It has up to 16 external channels and 3 internal channels (temperature sensor, voltage reference,  $V_{LCD}$  voltage measurement). Three channels, PA0, PA4 and PA5, are fast channels, while the others are standard channels.

The ADC performs conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC frequency is independent from the CPU frequency, allowing maximum sampling rate of 1.14 MSPS even with a low CPU speed. The ADC consumption is low at all frequencies (~25  $\mu$ A at 10 kSPS, ~240  $\mu$ A at 1MSPS). An auto-shutdown function guarantees that the ADC is powered off except during the active conversion phase.

The ADC can be served by the DMA controller. It can operate from a supply voltage down to 1.65 V.

The ADC features a hardware oversampler up to 256 samples, this improves the resolution to 16 bits (see AN2668).



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An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions and timers.

### 3.13 Temperature sensor

The temperature sensor ( $T_{SENSE}$ ) generates a voltage  $V_{SENSE}$  that varies linearly with temperature.

The temperature sensor is internally connected to the ADC\_IN18 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

Calibration value name	Description	Memory address
TSENSE_CAL1	TS ADC raw data acquired at temperature of 30 °C, V <sub>DDA</sub> = 3 V	0x1FF8 007A - 0x1FF8 007B
TSENSE_CAL2	TS ADC raw data acquired at temperature of 130 °C V <sub>DDA</sub> = 3 V	0x1FF8 007E - 0x1FF8 007F

Table 7. Temperature sensor calibration values

### 3.13.1 Internal voltage reference (V<sub>REFINT</sub>)

VREFINT CAL

The internal voltage reference ( $V_{REFINT}$ ) provides a stable (bandgap) voltage output for the ADC and Comparators.  $V_{REFINT}$  is internally connected to the ADC\_IN17 input channel. It enables accurate monitoring of the  $V_{DD}$  value (when no external voltage,  $V_{REF+}$ , is available for ADC). The precise voltage of  $V_{REFINT}$  is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

14.010 01 11110	man romage romeromes means	
Calibration value name	Description	Memory address
	Raw data acquired at	

0x1FF8 0078 - 0x1FF8 0079

Table 8. Internal voltage reference measured values

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temperature of 25 °C

 $V_{DDA} = 3 V$ 

### 3.13.2 V<sub>LCD</sub> voltage monitoring

This embedded hardware feature allows the application to measure the  $V_{LCD}$  supply voltage using the internal ADC channel ADC\_IN16. As the  $V_{LCD}$  voltage may be higher than  $V_{DDA}$ , and thus outside the ADC input range, the ADC input is connected to LCD\_VLCD1 (which provides  $1/3V_{LCD}$  when the LCD is configured 1/3Bias and  $1/4V_{LCD}$  when the LCD is configured 1/4Bias or 1/2Bias).

### 3.14 Digital-to-analog converter (DAC)

Two 12-bit buffered DACs can be used to convert digital signal into analog voltage signal output. An optional amplifier can be used to reduce the output signal impedance.

This digital Interface supports the following features:

- One data holding register (for each channel)
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channels with independent or simultaneous conversions
- DMA capability (including the underrun interrupt)
- External triggers for conversion
- Input reference voltage V<sub>REF+</sub>

Six DAC trigger inputs are used in the STM32L073xx. The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

### 3.15 Ultra-low-power comparators and reference voltage

The STM32L073xx embed two comparators sharing the same current bias and reference voltage. The reference voltage can be internal or external (coming from an I/O).

- One comparator with ultra low consumption
- One comparator with rail-to-rail inputs, fast or slow mode.
- The threshold can be one of the following:
  - DAC output
  - External I/O pins
  - Internal reference voltage (V<sub>RFFINT</sub>)
  - submultiple of Internal reference voltage(1/4, 1/2, 3/4) for the rail to rail comparator.

Both comparators can wake up the devices from Stop mode, and be combined into a window comparator.

The internal reference voltage is available externally via a low-power / low-current output buffer (driving current capability of 1  $\mu$ A typical).

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### 3.16 Touch sensing controller (TSC)

The STM32L073xx provide a simple solution for adding capacitive sensing functionality to any application. These devices offer up to 24 capacitive sensing channels distributed over 8 analog I/O groups.

Capacitive sensing technology is able to detect the presence of a finger near a sensor which is protected from direct touch by a dielectric (such as glass, plastic). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists of charging the sensor capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. To limit the CPU bandwidth usage, this acquisition is directly managed by the hardware touch sensing controller and only requires few external components to operate.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library, which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

Table 9. Capacitive sensing GPIOs available on STM32L073xx devices

Group	Capacitive sensing signal name	Pin name	Group	Capacitive sensing signal name	Pin name
	TSC_G1_IO1	PA0		TSC_G5_IO1	PB3
1	TSC_G1_IO2	PA1	5	TSC_G5_IO2	PB4
'	TSC_G1_IO3	PA2	3	TSC_G5_IO3	PB6
	TSC_G1_IO4	PA3		TSC_G5_IO4	PB7
	TSC_G2_IO1	PA4 <sup>(1)</sup>		TSC_G6_IO1	PB11
2	TSC_G2_IO2	PA5	6	TSC_G6_IO2	PB12
2	TSC_G2_IO3	PA6	0	TSC_G6_IO3	PB13
	TSC_G2_IO4	PA7		TSC_G6_IO4	PB14
	TSC_G3_IO1	PC5		TSC_G7_IO1	PC0
3	TSC_G3_IO2	PB0	7	TSC_G7_IO2	PC1
3	TSC_G3_IO3	PB1	,	TSC_G7_IO3	PC2
	TSC_G3_IO4	PB2		TSC_G7_IO4	PC3
	TSC_G4_IO1	PA9		TSC_G8_IO1	PC6
4	TSC_G4_IO2	PA10	8	TSC_G8_IO2	PC7
	TSC_G4_IO3	PA11	0	TSC_G8_IO3	PC8
	TSC_G4_IO4	PA12		TSC_G8_IO4	PC9

This GPIO offers a reduced touch sensing sensitivity. It is thus recommended to use it as sampling capacitor I/O.

### 3.17 Timers and watchdogs

The ultra-low-power STM32L073xx devices include three general-purpose timers, one low-power timer (LPTIM), one basic timer, two watchdog timers and the SysTick timer.

Table 10 compares the features of the general-purpose and basic timers.

Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
TIM2, TIM3	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
TIM21, TIM22	16-bit	Up, down, up/down	Any integer between 1 and 65536	No	2	No
TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

Table 10. Timer feature comparison

### 3.17.1 General-purpose timers (TIM2, TIM3, TIM21 and TIM22)

There are four synchronizable general-purpose timers embedded in the STM32L073xx device (see *Table 10* for differences).

### TIM2, TIM3

TIM2 and TIM3 are based on 16-bit auto-reload up/down counter. It includes a 16-bit prescaler. It features four independent channels each for input capture/output compare, PWM or one-pulse mode output.

The TIM2/TIM3 general-purpose timers can work together or with the TIM21 and TIM22 general-purpose timers via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs.

TIM2/TIM3 have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

#### TIM21 and TIM22

TIM21 and TIM22 are based on a 16-bit auto-reload up/down counter. They include a 16-bit prescaler. They have two independent channels for input capture/output compare, PWM or one-pulse mode output. They can work together and be synchronized with the TIM2/TIM3, full-featured general-purpose timers.

They can also be used as simple time bases and be clocked by the LSE clock source (32.768 kHz) to provide time bases independent from the main CPU clock.

### 3.17.2 Low-power Timer (LPTIM)

The low-power timer has an independent clock and is running also in Stop mode if it is clocked by LSE, LSI or an external clock. It is able to wakeup the devices from Stop mode.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous / one shot mode
- Selectable software / hardware input trigger
- Selectable clock source
  - Internal clock source: LSE, LSI, HSI or APB clock
  - External clock source over LPTIM input (working even with no internal clock source running, used by the Pulse Counter Application)
- Programmable digital glitch filter
- Encoder mode

### 3.17.3 Basic timer (TIM6, TIM7)

These timers can be used as a generic 16-bit timebase.

### 3.17.4 SysTick timer

This timer is dedicated to the OS, but could also be used as a standard downcounter. It is based on a 24-bit downcounter with autoreload capability and a programmable clock source. It features a maskable system interrupt generation when the counter reaches '0'.

### 3.17.5 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 37 kHz internal RC and, as it operates independently of the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes. The counter can be frozen in debug mode.

### 3.17.6 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.



#### **Communication interfaces** 3.18

#### I<sup>2</sup>C bus 3.18.1

Up to three I<sup>2</sup>C interfaces (I2C1 and I2C3) can operate in multimaster or slave modes.

Each I<sup>2</sup>C interface can support Standard mode (Sm, up to 100 kbit/s), Fast mode (Fm, up to 400 kbit/s) and Fast Mode Plus (Fm+, up to 1 Mbit/s) with 20 mA output drive on some I/Os.

7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (2 addresses, 1 with configurable mask) are also supported as well as programmable analog and digital noise filters.

rabio : companicon or izo analog and algital intere				
	Analog filter	Digital filter		
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2C peripheral clocks		
Benefits	Available in Stop mode	Extra filtering capability vs. standard requirements.     Stable length		
Drawbacks	Variations depending on temperature, voltage, process	Wakeup from Stop on address match is not available when digital filter is enabled		

Table 11. Comparison of I2C analog and digital filters

In addition, I2C1 and I2C3 provide hardware support for SMBus 2.0 and PMBus 1.1: ARP capability, Host notify protocol, hardware CRC (PEC) generation/verification, timeouts verifications and ALERT protocol management. I2C1/I2C3 also have a clock domain independent from the CPU clock, allowing the I2C1/I2C3 to wake up the MCU from Stop mode on address match.

Table 12. STM32L073xx I<sup>2</sup>C implementation

Each I2C interface can be served by the DMA controller.

Refer to Table 12 for an overview of I2C interface features.

I2C features<sup>(1)</sup> **I2C1 I2C2** Χ 7-bit addressing mode Х Χ Χ 10-bit addressing mode

**I2C3** Х Χ Standard mode (up to 100 kbit/s) Χ Χ Χ Fast mode (up to 400 kbit/s) Χ Χ Χ Fast Mode Plus with 20 mA output drive I/Os (up to 1  $X^{(2)}$ Χ Χ Mbit/s) Independent clock Χ Х **SMBus** Χ Χ Wakeup from STOP Χ Х

X = supported.

<sup>2.</sup> See Table 16: STM32L073xx pin definition on page 42 for the list of I/Os that feature Fast Mode Plus capability



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### 3.18.2 Universal synchronous/asynchronous receiver transmitter (USART)

The four USART interfaces (USART1, USART2, USART4 and USART5) are able to communicate at speeds of up to 4 Mbit/s.

They provide hardware management of the CTS, RTS and RS485 driver enable (DE) signals, multiprocessor communication mode, master synchronous communication and single-wire half-duplex communication mode. USART1 and USART2 also support SmartCard communication (ISO 7816), IrDA SIR ENDEC, LIN Master/Slave capability, auto baud rate feature and has a clock domain independent from the CPU clock, allowing to wake up the MCU from Stop mode using baudrates up to 42 Kbaud.

All USART interfaces can be served by the DMA controller.

*Table 13* for the supported modes and features of USART interfaces.

USART modes/features <sup>(1)</sup>	USART1 and USART2	USART4 and USART5
Hardware flow control for modem	X	X
Continuous communication using DMA	Х	Х
Multiprocessor communication	Х	X
Synchronous mode <sup>(2)</sup>	Х	X
Smartcard mode	X	-
Single-wire half-duplex communication	X	Х
IrDA SIR ENDEC block	X	-
LIN mode	X	-
Dual clock domain and wakeup from Stop mode	X	-
Receiver timeout interrupt	X	-
Modbus communication	X	-
Auto baud rate detection (4 modes)	X	-
Driver Enable	X	X

**Table 13. USART implementation** 

### 3.18.3 Low-power universal asynchronous receiver transmitter (LPUART)

The devices embed one Low-power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock. It can wake up the system from Stop mode using baudrates up to 46 Kbaud. The Wakeup events from Stop mode are programmable and can be:

- Start bit detection
- Or any received data frame
- Or a specific programmed data frame

<sup>1.</sup> X = supported.

<sup>2.</sup> This mode allows using the USART as an SPI master.

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

LPUART interface can be served by the DMA controller.

### 3.18.4 Serial peripheral interface (SPI)/Inter-integrated sound (I2S)

Up to two SPIs are able to communicate at up to 16 Mbits/s in slave and master modes in full-duplex and half-duplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes.

The USARTs with synchronous capability can also be used as SPI master.

One standard I2S interfaces (multiplexed with SPI2) is available. It can operate in master or slave mode, and can be configured to operate with a 16-/32-bit resolution as input or output channels. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When the I2S interfaces is configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

The SPIs can be served by the DMA controller.

Refer to *Table 14* for the differences between SPI1 and SPI2.

 SPI features<sup>(1)</sup>
 SPI1
 SPI2

 Hardware CRC calculation
 X
 X

 I2S mode
 X

 TI mode
 X
 X

Table 14. SPI/I2S implementation

### 3.18.5 Universal serial bus (USB)

The STM32L073xx embed a full-speed USB device peripheral compliant with the USB specification version 2.0. The internal USB PHY supports USB FS signaling, embedded DP pull-up and also battery charging detection according to Battery Charging Specification Revision 1.2. The USB interface implements a full-speed (12 Mbit/s) function interface with added support for USB 2.0 Link Power Management. It has software-configurable endpoint setting with packet memory up to 1 Kbyte and suspend/resume support. It requires a precise 48 MHz clock which can be generated from the internal main PLL (the clock source must use a HSE crystal oscillator) or by the internal 48 MHz oscillator in automatic trimming mode. The synchronization for this oscillator can be taken from the USB data stream itself (SOF signalization) which allows crystal-less operation.

<sup>1.</sup> X = supported.

### 3.19 Clock recovery system (CRS)

The STM32L073xx embed a special block which allows automatic trimming of the internal 48 MHz oscillator to guarantee its optimal accuracy over the whole device operational range. This automatic trimming is based on the external synchronization signal, which could be either derived from USB SOF signalization, from LSE oscillator, from an external signal on CRS\_SYNC pin or generated by user software. For faster lock-in during startup it is also possible to combine automatic trimming with manual trimming action.

### 3.20 Cyclic redundancy check (CRC) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

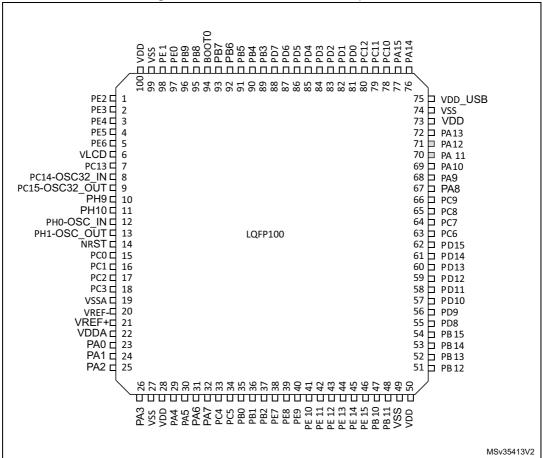
Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at linktime and stored at a given memory location.

### 3.21 Serial wire debug port (SW-DP)

An Arm SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.

### 4 Pin descriptions

Figure 3. STM32L073xx LQFP100 pinout



- 1. The above figure shows the package top view.
- 2. The I/O pins supplied by VDD\_USB are shown in grey.

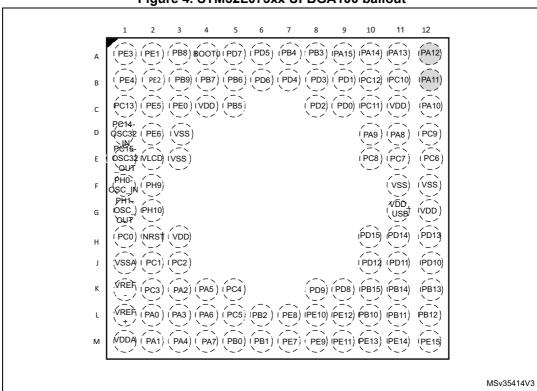


Figure 4. STM32L073xx UFBGA100 ballout

- 1. The above figure shows the package top view.
- 2. The I/O pins supplied by VDD\_USBare shown in grey.

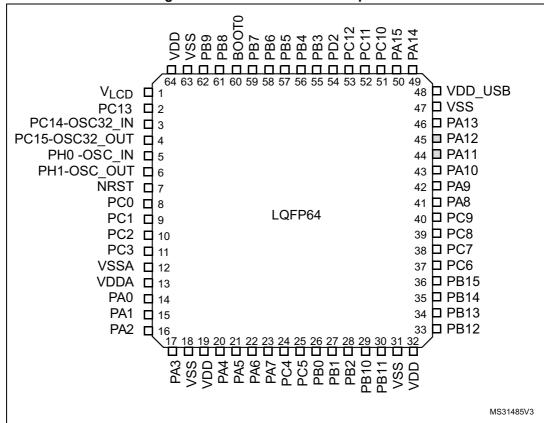


Figure 5. STM32L073xx LQFP64 pinout

- 1. The above figure shows the package top view.
- 2. The I/O pins supplied by VDD\_USB are shown in grey.

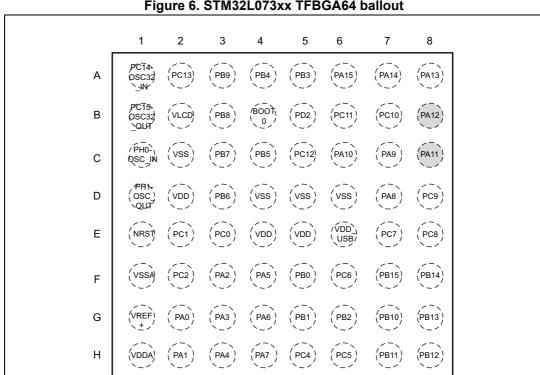


Figure 6. STM32L073xx TFBGA64 ballout

MSv37829V2

<sup>1.</sup> The above figure shows the package top view.

<sup>2.</sup> the I/O pins supplied by VDD\_USB are shown in grey.

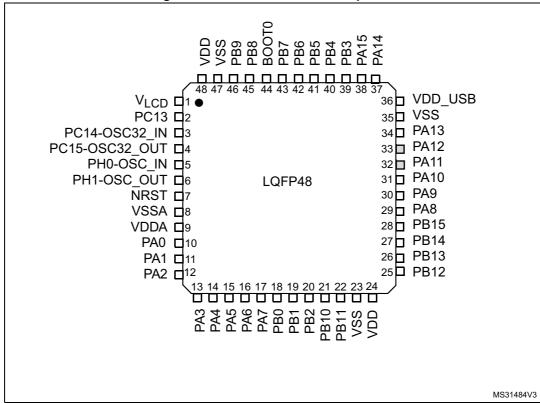


Figure 7. STM32L073xx LQFP48 pinout

- 1. The above figure shows the package top view.
- 2. The I/O pins supplied by VDD\_USB are shown in grey.

VSS
PB9
BOOT0
PB7
PB6
PB5
PB7
PB7
PB7
PB7
PB7
PB7
PB7
PB7
PB7 47 46 45 47 47 47 40 40 40 33 33 33 33 VLCD [ 36 VDD\_USB PC13 🗌 35 □ vss 34 | PA13 PC14-OSC32 IN PC15-OSC32\_OUT 33 🔲 PA12 32 PA11 PH0-OSC IN [ PH1-OSC\_OUT 31 ☐ PA10 UFQFPN48 NRST [ 30 | PA9 VSSA [ 29 🗖 PA8 28 PB15 VDDA 🗆 PA0 🔲 10 27 🗖 PB14 26 PB13 PA1 🔲 11 PA2 🗌 12 25 PB12 PA3
PA4
PA5
PA6
PA7
PB0
PB1
PB1
PB1
PB1
VSS
VDD MSv62439V1

Figure 8. STM32L073xx UFQFPN48

- The above figure shows the package top view.
- 2. The I/O pins supplied by VDD\_USB are shown in grey.

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Table 15. Legend/abbreviations used in the pinout table

Nar	ne	Abbreviation	Definition		
Pin na	ame		d in brackets below the pin name, the pin function during and reset is the same as the actual pin name		
		S	Supply pin		
Pin t	ype	I	Input only pin		
		I/O	Input / output pin		
		FT	5 V tolerant I/O		
		FTf	5 V tolerant I/O, FM+ capable		
I/O stru	ucture	TC Standard 3.3V I/O			
		В	Dedicated BOOT0 pin		
		RST	Bidirectional reset pin with embedded weak pull-up resistor		
Not	es	Unless otherwise specifie	d by a note, all I/Os are set as floating inputs during and after reset.		
Pin functions	Alternate functions	Functio	ons selected through GPIOx_AFR registers		
THITUIICUOTIS	Additional functions	Functions directly selected/enabled through peripheral registers			

Table 16. STM32L073xx pin definition

	Piı	n num	ber								
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions	
-	-	-	1	B2	PE2	I/O	FT	-	LCD_SEG38, TIM3_ETR	-	
-	-	-	2	A1	PE3	I/O	FT	-	TIM22_CH1, LCD_SEG39, TIM3_CH1	-	
	-	-	3	B1	PE4	I/O	FT	-	TIM22_CH2, TIM3_CH2	-	
-	-	-	4	C2	PE5	I/O	FT	-	TIM21_CH1, TIM3_CH3	-	
-	-	-	5	D2	PE6	I/O	FT	-	TIM21_CH2, TIM3_CH4	RTC_TAMP3/WKUP3	
1	1	B2	6	E2	VLCD	S		_	-		
2	2	A2	7	C1	PC13	I/O	FT	-	-	RTC_TAMP1/RTC_TS/ RTC_OUT/WKUP2	

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber				•		(continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
3	3	A1	8	D1	PC14- OSC32_IN (PC14)	I/O	FT	ı	-	OSC32_IN
4	4	B1	9	E1	PC15- OSC32_OUT (PC15)	I/O	тс	ı	-	OSC32_OUT
-	-	-	10	F2	PH9	I/O	FT	-	-	-
-	-	-	11	G2	PH10	I/O	FT	-	-	-
5	5	C1	12	F1	PH0-OSC_IN (PH0)	I/O	TC	-	USB_CRS_SYNC	OSC_IN
6	6	D1	13	G1	PH1- OSC_OUT (PH1)	I/O	тс	-	-	OSC_OUT
7	7	E1	14	H2	NRST	I/O	-	-	-	-
-	8	E3	15	Н1	PC0	I/O	FTf	1	LPTIM1_IN1, LCD_SEG18, EVENTOUT, TSC_G7_IO1, LPUART1_RX, I2C3_SCL	ADC_IN10
-	9	E2	16	J2	PC1	I/O	FTf	1	LPTIM1_OUT, LCD_SEG19, EVENTOUT, TSC_G7_IO2, LPUART1_TX, I2C3_SDA	ADC_IN11
-	10	F2	17	J3	PC2	I/O	FTf	1	LPTIM1_IN2, LCD_SEG20, SPI2_MISO/I2S2_MCK, TSC_G7_IO3	ADC_IN12
-	11	-	18	K2	PC3	I/O	FT	ı	LPTIM1_ETR, LCD_SEG21, SPI2_MOSI/I2S2_SD, TSC_G7_IO4	ADC_IN13

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Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber						(continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
8	12	F1	19	J1	VSSA	S	-	-	-	-
-	-	-	20	K1	VREF-	S	-	-	-	-
-	-	G1	21	L1	VREF+	S	-	-	-	-
9	13	H1	22	M1	VDDA	S	-	-	-	-
10	14	G2	23	L2	PA0	I/O	тс	-	TIM2_CH1, TSC_G1_IO1, USART2_CTS, TIM2_ETR, USART4_TX, COMP1_OUT	COMP1_INM, ADC_IN0, RTC_TAMP2/WKUP1
11	15	H2	24	M2	PA1	I/O	FT	-	EVENTOUT, LCD_SEG0, TIM2_CH2, TSC_G1_IO2, USART2_RTS/ USART2_DE, TIM21_ETR, USART4_RX	COMP1_INP, ADC_IN1
12	16	F3	25	K3	PA2	I/O	FT	1	TIM21_CH1, LCD_SEG1, TIM2_CH3, TSC_G1_IO3, USART2_TX, LPUART1_TX, COMP2_OUT	COMP2_INM, ADC_IN2
13	17	G3	26	L3	PA3	I/O	FT	1	TIM21_CH2, LCD_SEG2, TIM2_CH4, TSC_G1_IO4, USART2_RX, LPUART1_RX	COMP2_INP, ADC_IN3
-	18	C2	27	E3	VSS	S	-	-	-	-
-	19	D2	28	Н3	VDD	S	-	-	-	-
14	20	НЗ	29	M3	PA4	I/O	тс	(1)	SPI1_NSS, TSC_G2_IO1, USART2_CK, TIM22_ETR	COMP1_INM, COMP2_INM, ADC_IN4, DAC_OUT1

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber				•		eminion (continueu)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
15	21	F4	30	K4	PA5	I/O	тс	-	SPI1_SCK, TIM2_ETR, TSC_G2_IO2, TIM2_CH1	COMP1_INM, COMP2_INM, ADC_IN5, DAC_OUT2
16	22	G4	31	L4	PA6	I/O	FT	-	SPI1_MISO, LCD_SEG3, TIM3_CH1, TSC_G2_IO3, LPUART1_CTS, TIM22_CH1, EVENTOUT, COMP1_OUT	ADC_IN6
17	23	H4	32	M4	PA7	I/O	FT	-	SPI1_MOSI, LCD_SEG4, TIM3_CH2, TSC_G2_IO4, TIM22_CH2, EVENTOUT, COMP2_OUT	ADC_IN7
-	24	H5	33	K5	PC4	I/O	FT	-	EVENTOUT, LCD_SEG22, LPUART1_TX	ADC_IN14
-	25	H6	34	L5	PC5	I/O	FT	-	LCD_SEG23, LPUART1_RX, TSC_G3_IO1	ADC_IN15
18	26	F5	35	M5	PB0	I/O	FT	ı	EVENTOUT, LCD_SEG5, TIM3_CH3, TSC_G3_IO2	LCD_VLCD3, ADC_IN8, VREF_OUT
19	27	G5	36	M6	PB1	I/O	FT	-	LCD_SEG6, TIM3_CH4, TSC_G3_IO3, LPUART1_RTS/ LPUART1_DE	ADC_IN9, VREF_OUT
20	28	G6	37	L6	PB2	I/O	FT	-	LPTIM1_OUT, TSC_G3_IO4, I2C3_SMBA	LCD_VLCD2
-	-	-	38	M7	PE7	I/O	FT	-	LCD_SEG45, USART5_CK/ USART5_RTS/ USART5_DE	-

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber				•		fillition (continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	1	-	39	L7	PE8	I/O	FT	-	LCD_SEG46, USART4_TX	-
-	1	-	40	M8	PE9	I/O	FT	-	TIM2_CH1, LCD_SEG47, TIM2_ETR, USART4_RX	-
-	ı	1	41	L8	PE10	I/O	FT	ı	TIM2_CH2, LCD_SEG40, USART5_TX	-
-	1	ı	42	М9	PE11	I/O	FT	ı	TIM2_CH3, USART5_RX	LCD_VLCD1
-	ı	ı	43	L9	PE12	I/O	FT	1	TIM2_CH4, SPI1_NSS	LCD_VLCD3
-	1	-	44	M10	PE13	I/O	FT	-	LCD_SEG41, SPI1_SCK	-
-	ı	ı	45	M11	PE14	I/O	FT	ı	LCD_SEG42, SPI1_MISO	-
-	ı	ı	46	M12	PE15	I/O	FT	ı	LCD_SEG43, SPI1_MOSI	ı
21	29	G7	47	L10	PB10	I/O	FT	-	LCD_SEG10, TIM2_CH3, TSC_SYNC, LPUART1_TX, SPI2_SCK, I2C2_SCL, LPUART1_RX	-
22	30	Н7	48	L11	PB11	I/O	FT	-	EVENTOUT, LCD_SEG11, TIM2_CH4, TSC_G6_IO1, LPUART1_RX, I2C2_SDA, LPUART1_TX	-
23	31	D6	49	F12	VSS	S		-	-	-
24	32	E5	50	G12	VDD	S		-	-	-

Table 16. STM32L073xx pin definition (continued)

	Piı	n num	ber						ennition (continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
25	33	Н8	51	L12	PB12	I/O	FT	-	SPI2_NSS/I2S2_WS, LCD_SEG12, LPUART1_RTS/ LPUART1_DE, TSC_G6_IO2, I2C2_SMBA, EVENTOUT	LCD_VLCD1
26	34	G8	52	K12	PB13	I/O	FTf	-	SPI2_SCK/I2S2_CK, LCD_SEG13, MCO, TSC_G6_IO3, LPUART1_CTS, I2C2_SCL, TIM21_CH1	-
27	35	F8	53	K11	PB14	I/O	FTf	ı	SPI2_MISO/I2S2_MCK, LCD_SEG14, RTC_OUT, TSC_G6_IO4, LPUART1_RTS/ LPUART1_DE, I2C2_SDA, TIM21_CH2	-
28	36	F7	54	K10	PB15	I/O	FT	1	SPI2_MOSI/I2S2_SD, LCD_SEG15, RTC_REFIN	-
-	1	-	55	K9	PD8	I/O	FT	-	LPUART1_TX, LCD_SEG28	-
-	-	-	56	K8	PD9	I/O	FT	-	LPUART1_RX, LCD_SEG29	-
-	-	-	57	J12	PD10	I/O	FT	-	LCD_SEG30	-
-	1	-	58	J11	PD11	I/O	FT	-	LPUART1_CTS, LCD_SEG31	-
-	1	ı	59	J10	PD12	I/O	FT	ı	LPUART1_RTS/ LPUART1_DE, LCD_SEG32	-
-	-	-	60	H12	PD13	I/O	FT	-	LCD_SEG33	-
-	-	-	61	H11	PD14	I/O	FT	-	LCD_SEG34	-

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber				_			
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	1	-	62	H10	PD15	I/O	FT	ı	USB_CRS_SYNC, LCD_SEG35	-
1	37	F6	63	E12	PC6	I/O	FT	1	TIM22_CH1, LCD_SEG24, TIM3_CH1, TSC_G8_IO1	-
-	38	E7	64	E11	PC7	I/O	FT	-	TIM22_CH2, LCD_SEG25, TIM3_CH2, TSC_G8_IO2	-
-	39	E8	65	E10	PC8	I/O	FT	ī	TIM22_ETR, LCD_SEG26, TIM3_CH3, TSC_G8_IO3	-
-	40	D8	66	D12	PC9	I/O	FTf	-	TIM21_ETR, LCD_SEG27, USB_NOE/TIM3_CH4, TSC_G8_IO4, I2C3_SDA	-
29	41	D7	67	D11	PA8	I/O	FTf	-	MCO, LCD_COM0, USB_CRS_SYNC, EVENTOUT, USART1_CK, I2C3_SCL	-
30	42	C7	68	D10	PA9	I/O	FTf	1	MCO, LCD_COM1, TSC_G4_IO1, USART1_TX,I2C1_SCL, I2C3_SMBA	-
31	43	C6	69	C12	PA10	I/O	FTf	ı	LCD_COM2, TSC_G4_IO2, USART1_RX, I2C1_SDA	-
32	44	C8	70	B12	PA11	I/O	FT	(2)	SPI1_MISO, EVENTOUT, TSC_G4_IO3, USART1_CTS, COMP1_OUT	USB_DM

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber						(continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
33	45	В8	71	A12	PA12	I/O	FT	(2)	SPI1_MOSI, EVENTOUT, TSC_G4_IO4, USART1_RTS/ USART1_DE, COMP2_OUT	USB_DP
34	46	A8	72	A11	PA13	I/O	FT	-	SWDIO, USB_NOE, LPUART1_RX	-
-	1	-	73	C11	VDD	S		ı	-	-
35	47	D5	74	F11	VSS	S		-	-	-
36	48	E6	75	G11	VDD_USB	S		-	-	-
37	49	A7	76	A10	PA14	I/O	FT	-	SWCLK, USART2_TX, LPUART1_TX	-
38	50	A6	77	A9	PA15	I/O	FT	-	SPI1_NSS, LCD_SEG17, TIM2_ETR, EVENTOUT, USART2_RX, TIM2_CH1, USART4_RTS/ USART4_DE	-
-	51	В7	78	B11	PC10	I/O	FT	-	LPUART1_TX, LCD_COM4/LCD_SEG2 8/LCD_SEG48, USART4_TX	-
-	52	В6	79	C10	PC11	I/O	FT	-	LPUART1_RX, LCD_COM5/LCD_SEG2 9/LCD_SEG49, USART4_RX	-
-	53	C5	80	B10	PC12	I/O	FT	-	LCD_COM6/LCD_SEG3 0/LCD_SEG50, USART5_TX, USART4_CK	-
-	-	-	81	C9	PD0	I/O	FT	-	TIM21_CH1, SPI2_NSS/I2S2_WS	-

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber						fillition (continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
-	-	-	82	В9	PD1	I/O	FT	-	SPI2_SCK/I2S2_CK	-
-	54	B5	83	C8	PD2	I/O	FT	-	LPUART1_RTS/ LPUART1_DE, LCD_COM7/LCD_SEG3 1/LCD_SEG51, TIM3_ETR, USART5_RX	-
-	-	-	84	В8	PD3	I/O	FT	-	USART2_CTS, LCD_SEG44, SPI2_MISO/I2S2_MCK	-
-	-	-	85	В7	PD4	I/O	FT	-	USART2_RTS/ USART2_DE, SPI2_MOSI/I2S2_SD	-
-	-	-	86	A6	PD5	I/O	FT	-	USART2_TX	-
-	-	-	87	В6	PD6	I/O	FT	-	USART2_RX	-
-	-	-	88	A5	PD7	I/O	FT	-	USART2_CK, TIM21_CH2	-
39	55	A5	89	A8	PB3	I/O	FT	1	SPI1_SCK, LCD_SEG7, TIM2_CH2, TSC_G5_IO1, EVENTOUT, USART1_RTS/ USART1_DE, USART5_TX	COMP2_INM
40	56	A4	90	A7	PB4	I/O	FTf	-	SPI1_MISO, LCD_SEG8, TIM3_CH1, TSC_G5_IO2, TIM22_CH1, USART1_CTS, USART5_RX, I2C3_SDA	COMP2_INP

Table 16. STM32L073xx pin definition (continued)

	Pi	n num	ber				•		continued)	
LQFP48/UFQFPN48	LQFP64	TFBGA64	LQFP100	UFBGA100	Pin name (function after reset)	Pin type	I/O structure	Note	Alternate functions	Additional functions
41	57	C4	91	C5	PB5	I/O	FT	-	SPI1_MOSI, LCD_SEG9, LPTIM1_IN1, I2C1_SMBA, TIM3_CH2/TIM22_CH2, USART1_CK, USART5_CK, USART5_RTS/ USART5_DE	COMP2_INP
42	58	D3	92	B5	PB6	I/O	FTf	-	USART1_TX, I2C1_SCL, LPTIM1_ETR, TSC_G5_IO3	COMP2_INP
43	59	СЗ	93	B4	PB7	I/O	FTf	-	USART1_RX, I2C1_SDA, LPTIM1_IN2, TSC_G5_IO4, USART4_CTS	COMP2_INP, PVD_IN
44	60	B4	94	A4	воото	I		-	-	-
45	61	В3	95	A3	PB8	I/O	FTf	-	LCD_SEG16, TSC_SYNC, I2C1_SCL	-
46	62	A3	96	В3	PB9	I/O	FTf	1	LCD_COM3, EVENTOUT, I2C1_SDA, SPI2_NSS/I2S2_WS	-
-	-	-	97	C3	PE0	I/O	FT	-	LCD_SEG36, EVENTOUT	-
-	-	-	98	A2	PE1	I/O	FT	ı	LCD_SEG37, EVENTOUT	-
47	63	D4	99	D3	VSS	S	-	-	-	-
48	64	E4	100	C4	VDD	S	-	-	-	-

<sup>1.</sup> PA4 offers a reduced touch sensing sensitivity. It is thus recommended to use it as sampling capacitor I/O.

<sup>2.</sup> These pins are powered by VDD\_USB. For all characteristics that refer to  $V_{DD}$ ,  $V_{DD\_USB}$  must be used instead.

	Table 17. Alternate functions port A											
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7			
F	Port	SPI1/SPI2/I2S2/ USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	SPI1/SPI2/I2S2/ I2C1/LCD/ TIM2/21	SPI1/SPI2/I2S2/ LPUART1/ USART5/USB/L PTIM1/TIM2/3/E VENTOUT/ SYS_AF	I2C1/TSC/ EVENTOUT	I2C1/USART1/2 /LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2/I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21 / EVENTOUT	I2C3/LPUART1/ COMP1/2/ TIM3			
	PA0	-	-	TIM2_CH1	TSC_G1_IO1	USART2_CTS	TIM2_ETR	USART4_TX	COMP1_OUT			
	PA1	EVENTOUT	LCD_SEG0	TIM2_CH2	TSC_G1_IO2	USART2_RTS/ USART2_DE	TIM21_ETR	USART4_RX	-			
	PA2	TIM21_CH1	LCD_SEG1	TIM2_CH3	TSC_G1_IO3	USART2_TX	-	LPUART1_TX	COMP2_OUT			
	PA3	TIM21_CH2	LCD_SEG2	TIM2_CH4	TSC_G1_IO4	USART2_RX	-	LPUART1_RX	-			
	PA4	SPI1_NSS	-	-	TSC_G2_IO1	USART2_CK	TIM22_ETR	-	-			
	PA5	SPI1_SCK	-	TIM2_ETR	TSC_G2_IO2		TIM2_CH1	-	-			
	PA6	SPI1_MISO	LCD_SEG3	TIM3_CH1	TSC_G2_IO3	LPUART1_CTS	TIM22_CH1	EVENTOUT	COMP1_OUT			
_	PA7	SPI1_MOSI	LCD_SEG4	TIM3_CH2	TSC_G2_IO4	-	TIM22_CH2	EVENTOUT	COMP2_OUT			
Port A	PA8	MCO	LCD_COM0	USB_CRS_ SYNC	EVENTOUT	USART1_CK	-	-	I2C3_SCL			
	PA9	MCO	LCD_COM1	-	TSC_G4_IO1	USART1_TX	-	I2C1_SCL	I2C3_SMBA			
	PA10	-	LCD_COM2	-	TSC_G4_IO2	USART1_RX	-	I2C1_SDA	-			
	PA11	SPI1_MISO	-	EVENTOUT	TSC_G4_IO3	USART1_CTS	-	-	COMP1_OUT			
	PA12	SPI1_MOSI	-	EVENTOUT	TSC_G4_IO4	USART1_RTS/ USART1_DE	-	-	COMP2_OUT			
	PA13	SWDIO	-	USB_NOE	-	-	-	LPUART1_RX	-			
	PA14	SWCLK	-	-	-	USART2_TX	-	LPUART1_TX	-			
	PA15	SPI1_NSS	LCD_SEG17	TIM2_ETR	EVENTOUT	USART2_RX	TIM2_CH1	USART4_RTS/ USART4_DE	-			



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Table 18. Alternate functions port B

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	Port	SPI1/SPI2/I2S2/ USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	SPI1/SPI2/I2S2/I 2C1/LCD/ TIM2/21	SPI1/SPI2/I2S2/ LPUART1/ USART5/USB/L PTIM1/TIM2/3/E VENTOUT/ SYS_AF	I2C1/TSC/ EVENTOUT	I2C1/USART1/2/ LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2/I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21/ EVENTOUT	I2C3/LPUART1/ COMP1/2/ TIM3
	PB0	EVENTOUT	LCD_SEG5	TIM3_CH3	TSC_G3_IO2	-	-	-	-
	PB1	-	LCD_SEG6	TIM3_CH4	TSC_G3_IO3	LPUART1_RTS/ LPUART1_DE	-	-	-
	PB2	-	-	LPTIM1_OUT	TSC_G3_IO4	-	-	-	I2C3_SMBA
	PB3	SPI1_SCK	LCD_SEG7	TIM2_CH2	TSC_G5_IO1	EVENTOUT	USART1_RTS/ USART1_DE	USART5_TX	-
	PB4	SPI1_MISO	LCD_SEG8	TIM3_CH1	TSC_G5_IO2	TIM22_CH1	USART1_CTS	USART5_RX	I2C3_SDA
	PB5	SPI1_MOSI	LCD_SEG9	LPTIM1_IN1	I2C1_SMBA	TIM3_CH2/ TIM22_CH2	USART1_CK	USART5_CK, USART5_RTS/ USART5_DE	-
	PB6	USART1_TX	I2C1_SCL	LPTIM1_ETR	TSC_G5_IO3	-	-	-	-
t B	PB7	USART1_RX	I2C1_SDA	LPTIM1_IN2	TSC_G5_IO4	-	-	USART4_CTS	-
Port	PB8	-	LCD_SEG16	-	TSC_SYNC	I2C1_SCL	-	-	-
	PB9	-	LCD_COM3	EVENTOUT	-	I2C1_SDA	SPI2_NSS/ I2S2_WS	-	-
	PB10	-	LCD_SEG10	TIM2_CH3	TSC_SYNC	LPUART1_TX	SPI2_SCK	I2C2_SCL	LPUART1_RX
	PB11	EVENTOUT	LCD_SEG11	TIM2_CH4	TSC_G6_IO1	LPUART1_RX	-	I2C2_SDA	LPUART1_TX
	PB12	SPI2_NSS/I2S2_WS	LCD_SEG12	LPUART1_RTS/ LPUART1_DE	TSC_G6_IO2		I2C2_SMBA	EVENTOUT	-
	PB13	SPI2_SCK/I2S2_CK	LCD_SEG13	MCO	TSC_G6_IO3	LPUART1_CTS	I2C2_SCL	TIM21_CH1	-
	PB14	SPI2_MISO/ I2S2_MCK	LCD_SEG14	RTC_OUT	TSC_G6_IO4	LPUART1_RTS/ LPUART1_DE	I2C2_SDA	TIM21_CH2	-
	PB15	SPI2_MOSI/ I2S2_SD	LCD_SEG15	RTC_REFIN	-	-	-	-	-

				Table 19. Alter	nate function	ns port C			
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	Port	SPI1/SPI2/I2S2/ USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	SPI1/SPI2/I2S2/I2C1/ LCD/ TIM2/21	SPI1/SPI2/I2S2/ LPUART1/ USART5/USB/ LPTIM1/TIM2/3 /EVENTOUT/SYS_AF	I2C1/TSC/ EVENTOUT	I2C1/USART1/2/ LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2 /I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21/E VENTOUT	I2C3/LPUART1/ COMP1/2/ TIM3
	PC0	LPTIM1_IN1	LCD_SEG18	EVENTOUT	TSC_G7_IO1			LPUART1_RX	I2C3_SCL
	PC1	LPTIM1_OUT	LCD_SEG19	EVENTOUT	TSC_G7_IO2			LPUART1_TX	I2C3_SDA
	PC2	LPTIM1_IN2	LCD_SEG20	SPI2_MISO/ I2S2_MCK	TSC_G7_IO3				
	PC3	LPTIM1_ETR	LCD_SEG21	SPI2_MOSI/ I2S2_SD	TSC_G7_IO4				
	PC4	EVENTOUT	LCD_SEG22	LPUART1_TX					
	PC5		LCD_SEG23	LPUART1_RX	TSC_G3_IO1				
	PC6	TIM22_CH1	LCD_SEG24	TIM3_CH1	TSC_G8_IO1				
O	PC7	TIM22_CH2	LCD_SEG25	TIM3_CH2	TSC_G8_IO2				
Port	PC8	TIM22_ETR	LCD_SEG26	TIM3_CH3	TSC_G8_IO3				
	PC9	TIM21_ETR	LCD_SEG27	USB_NOE/TIM3_CH4	TSC_G8_IO4				I2C3_SDA
	PC10	LPUART1_TX	LCD_COM4/LCD_SEG 28/LCD_SEG48					USART4_TX	
	PC11	LPUART1_RX	LCD_COM5/LCD_SEG 29/LCD_SEG49					USART4_RX	
	PC12		LCD_COM6/LCD_SEG 30/LCD_SEG50	USART5_TX				USART4_CK	
	PC13								
	PC14								
	PC15								



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Table 20. Alternate functions port D

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	Port	SPI1/SPI2/I2S2/ USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	USART1/2/ PUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/		I2C1/TSC/ EVENTOUT	I2C1/USART1/2/ LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2 /I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21/E VENTOUT	I2C3/LPUART1/ COMP1/2/TIM3
	PD0	TIM21_CH1	SPI2_NSS/I2S2_WS	-	-	-	-	-	-
	PD1	-	SPI2_SCK/I2S2_CK	-	-	-	-	-	-
	PD2	LPUART1_RTS/ LPUART1_DE	LCD_COM7/ LCD_SEG31/ LCD_SEG51	TIM3_ETR	-	-	-	USART5_RX	-
	PD3	USART2_CTS	LCD_SEG44	SPI2_MISO/ I2S2_MCK	-	-	-	-	-
	PD4	USART2_RTS/ USART2_DE	SPI2_MOSI/I2S2_SD	-	-	-	-	-	-
	PD5	USART2_TX	-	-	-	-	-	-	-
ţ.	PD6	USART2_RX	-	-	-	-	-	-	-
Port D	PD7	USART2_CK	TIM21_CH2	-	-	-	-	-	-
	PD8	LPUART1_TX	LCD_SEG28	-	-	-	-	-	-
	PD9	LPUART1_RX	LCD_SEG29	-	-	-	-	-	-
	PD10	-	LCD_SEG30	-	-	-	-	-	-
	PD11	LPUART1_CTS	LCD_SEG31	-	-	-	-	-	-
	PD12	LPUART1_RTS/ LPUART1_DE	LCD_SEG32	-	-	-	-	-	-
	PD13	-	LCD_SEG33	-	-	-	-	-	-
	PD14		LCD_SEG34	-	-	-	-	-	-
	PD15	USB_CRS_SYNC	LCD_SEG35	-	-	-	-	-	-

				Table 21.	Alternate fur	ctions port E			
		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
F	Port	SPI1/SPI2/I2S2/ USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	SPI1/SPI2/I2S2/I2C1 /LCD/TIM2/21	SPI1/SPI2/I2S2/ LPUART1/ USART5/USB/ LPTIM1/TIM2/3 /EVENTOUT/ SYS_AF	I2C1/TSC/ EVENTOUT	I2C1/USART1/2/ LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2 /I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21/ EVENTOUT	I2C3/LPUART1/ COMP1/2/TIM3
	PE0	-	LCD_SEG36	EVENTOUT	-	-	-	-	-
	PE1	-	LCD_SEG37	EVENTOUT	-	-	-	-	-
	PE2	-	LCD_SEG38	TIM3_ETR	-	-	-	-	-
	PE3	TIM22_CH1	LCD_SEG39	TIM3_CH1	-	-	-	-	-
	PE4	TIM22_CH2	-	TIM3_CH2	-	-	-	-	-
	PE5	TIM21_CH1	-	TIM3_CH3	-	-	-	-	-
	PE6	TIM21_CH2	-	TIM3_CH4	-	-	-	-	-
Port E	PE7	-	LCD_SEG45	-	-	-	-	USART5_CK, USART5_RTS/ USART5_DE	-
	PE8	-	LCD_SEG46	-	-	-	-	USART4_TX	-
	PE9	TIM2_CH1	LCD_SEG47	TIM2_ETR	-	-	-	USART4_RX	-
	PE10	TIM2_CH2	LCD_SEG40	-	-	-	-	USART5_TX	-
	PE11	TIM2_CH3	-	-	-	-	-	USART5_RX	
	PE12	TIM2_CH4	-	SPI1_NSS	-	-	-	-	-
	PE13	-	LCD_SEG41	SPI1_SCK	-	-	-	-	-
	PE14	-	LCD_SEG42	SPI1_MISO	-	-	-	-	-
	PE15	-	LCD_SEG43	SPI1_MOSI	-	-	-	-	-





### Table 22. Alternate functions port H

		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	Port	SPI1/SPI2/ I2S2/USART1/2/ LPUART1/USB/ LPTIM1/TSC/ TIM2/21/22/ EVENTOUT/ SYS_AF	SPI1/SPI2/I2S2 /I2C1/LCD/TIM2/21	SPI1/SPI2/I2S2/ LPUART1/ USART5/USB/ LPTIM1/TIM2/3/ EVENTOUT/ SYS_AF	I2C1/TSC/ EVENTOUT	I2C1/USART1/2/ LPUART1/ TIM3/22/ EVENTOUT	SPI2/I2S2/I2C2/ USART1/ TIM2/21/22	I2C1/2/ LPUART1/ USART4/ UASRT5/TIM21/ EVENTOUT	I2C3/ LPUART1/ COMP1/2/ TIM3
Ŧ	PH0	USB_CRS_SYNC	-	-	-	-	-	-	-
Port	PH1	-	-	-	-	-	-	-	-

Memory mapping STM32L073xx

# 5 Memory mapping

Refer to the product line reference manual for details on the memory mapping as well as the boundary addresses for all peripherals.



#### 6 Electrical characteristics

#### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean±3 $\sigma$ ).

#### 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A$  = 25 °C,  $V_{DD}$  = 3.6 V (for the 1.65 V  $\leq$ V $_{DD}$   $\leq$ 3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\sigma$ ).

#### 6.1.3 Typical curves

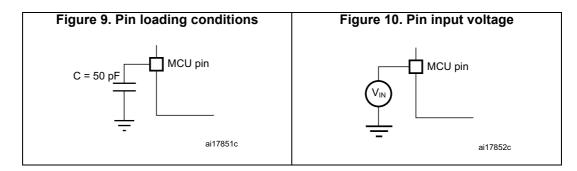
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 9.

#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 10*.



### 6.1.6 Power supply scheme

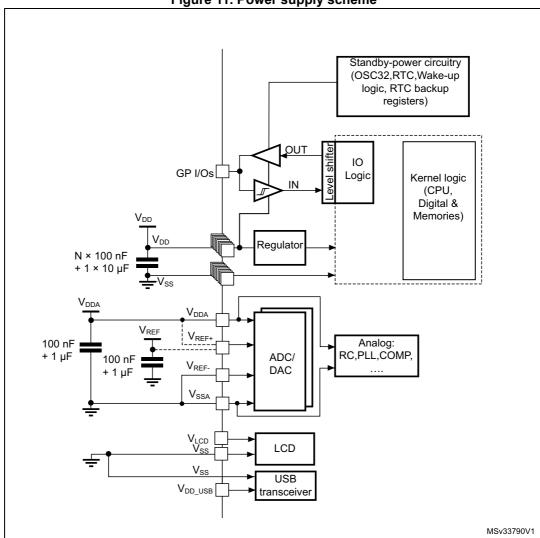


Figure 11. Power supply scheme



### 6.1.7 Optional LCD power supply scheme

Option 1
Option 2
C<sub>EXT</sub>

V<sub>DD</sub>
V<sub>DD</sub>
V<sub>DD</sub>
V<sub>DD</sub>
V<sub>DD</sub>
V<sub>DD</sub>
Step-up
Converter

LCD

MSv33791V1

Figure 12. Optional LCD power supply scheme

- 1. Option 1: LCD power supply is provided by a dedicated VLCD supply source, VSEL switch is open.
- 2. Option 2: LCD power supply is provided by the internal step-up converter, VSEL switch is closed, an external capacitance is needed for correct behavior of this converter.

### 6.1.8 Current consumption measurement

N × 100 nF + 1 × 10 μF NxVSS

Figure 13. Current consumption measurement scheme

### 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 23: Voltage characteristics*, *Table 24: Current characteristics*, and *Table 25: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 Qualification Standard. Extended mission profiles are available on demand.

Symbol	Definition	Min	Max	Unit	
V <sub>DD</sub> -V <sub>SS</sub>	External main supply voltage (including V <sub>DDA</sub> , V <sub>DD_USB</sub> , V <sub>DD</sub> ) <sup>(1)</sup>	-0.3	4.0		
V <sub>IN</sub> <sup>(2)</sup>	Input voltage on FT and FTf pins	V <sub>SS</sub> -0.3	V <sub>DD</sub> +4.0	]	
	Input voltage on TC pins	V <sub>SS</sub> -0.3	4.0	V	
	Input voltage on BOOT0	V <sub>SS</sub>	V <sub>DD</sub> +4.0		
	Input voltage on any other pin	V <sub>SS</sub> -0.3	4.0		
ΔV <sub>DD</sub>	Variations between different V <sub>DDx</sub> power pins	-	50		
V <sub>DDA</sub> -V <sub>DDx</sub>	Variations between any $V_{DDx}$ and $V_{DDA}$ power pins <sup>(3)</sup>	-	300	mV	
ΔV <sub>SS</sub>	Variations between all different ground pins including V <sub>REF-</sub> pin	-	50		
V <sub>REF+</sub> –V <sub>DDA</sub>	Allowed voltage difference for V <sub>REF+</sub> > V <sub>DDA</sub>	-	0.4	V	
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	see Secti	ion 6.3.11		

Table 23. Voltage characteristics

All main power (V<sub>DD</sub>,V<sub>DD USB</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

<sup>2.</sup> V<sub>IN</sub> maximum must always be respected. Refer to *Table 24* for maximum allowed injected current values.

<sup>3.</sup> It is recommended to power  $V_{DD}$  and  $V_{DDA}$  from the same source. A maximum difference of 300 mV between  $V_{DD}$  and  $V_{DDA}$  can be tolerated during power-up and device operation.  $V_{DD\_USB}$  is independent from  $V_{DD}$  and  $V_{DDA}$ : its value does not need to respect this rule.

**Table 24. Current characteristics** 

Symbol	Ratings	Max.	Unit			
$\Sigma I_{VDD}^{(2)}$	Total current into sum of all V <sub>DD</sub> power lines (source) <sup>(1)</sup>	105				
ΣI <sub>VSS</sub> <sup>(2)</sup>	Total current out of sum of all V <sub>SS</sub> ground lines (sink) <sup>(1)</sup>	105				
ΣI <sub>VDD_USB</sub>	Total current into V <sub>DD_USB</sub> power lines (source)	25				
I <sub>VDD(PIN)</sub>						
I <sub>VSS(PIN)</sub>	I <sub>VDD(PIN)</sub> Maximum current into each V <sub>DD</sub> power pin (source) <sup>(1)</sup> I <sub>VSS(PIN)</sub> Maximum current out of each V <sub>SS</sub> ground pin (sink) <sup>(1)</sup> Output current sunk by any I/O and control pin except FTf pins  Output current sunk by FTf pins					
		16				
I <sub>IO</sub>	Output current sunk by FTf pins	22				
	Output current sourced by any I/O and control pin	-16	mA			
	Total output current sunk by sum of all IOs and control pins except PA11 and PA12 <sup>(2)</sup>	90				
ΣΙ <sub>ΙΟ(PIN)</sub>	Total output current sunk by PA11 and PA12	25				
	Total output current sourced by sum of all IOs and control pins <sup>(2)</sup>	-90				
	Injected current on FT, FTf, RST and B pins	-5/+0 <sup>(3)</sup>				
I <sub>INJ(PIN)</sub>	Injected current on TC pin	± 5 <sup>(4)</sup>				
ΣΙ <sub>ΙΝJ(PIN)</sub>	Total injected current (sum of all I/O and control pins) <sup>(5)</sup>	± 25				

- All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.
- This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.
- Positive current injection is not possible on these I/Os. A negative injection is induced by V<sub>IN</sub><V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 23* for maximum allowed input voltage values.
- A positive injection is induced by V<sub>IN</sub> > V<sub>DD</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 23: Voltage characteristics* for the maximum allowed input voltage values.
- 5. When several inputs are submitted to a current injection, the maximum  $\Sigma I_{INJ(PIN)}$  is the absolute sum of the positive and negative injected currents (instantaneous values).

**Table 25. Thermal characteristics** 

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C



## 6.3 Operating conditions

### 6.3.1 General operating conditions

Table 26. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	32	
f <sub>PCLK1</sub>	Internal APB1 clock frequency	-	0	32	MHz
f <sub>PCLK2</sub>	Internal APB2 clock frequency	-	0	32	
		BOR detector disabled	1.65	3.6	
$V_{DD}$	Standard operating voltage	BOR detector enabled, at power-on	1.8	3.6	٧
		BOR detector disabled, after power-on	0 32 0 32 0 32 1.65 3.6		
$V_{DDA}$	Analog operating voltage (DAC not used)  Must be the same voltage as V <sub>DD</sub> <sup>(1)</sup>		1.65	3.6	٧
V <sub>DDA</sub>	Analog operating voltage (all features)  Must be the same voltage as V <sub>DD</sub> <sup>(1)</sup>		1.8	3.6	V
V <sub>DD</sub> _	Standard operating voltage, USB	USB peripheral used	3.0	3.6	V
USB	domain <sup>(2)</sup>	quency quency - BOR detector disabled BOR detector enabled, at power-on BOR detector disabled, after power-on BOR detector disabled, after power-on Ge (DAC not Must be the same voltage as V <sub>DD</sub> (1)  Ge (all Must be the same voltage as V <sub>DD</sub> (1)  USB peripheral used USB peripheral not used USB peripheral not used  1 and RST 2.0 V ≤V <sub>DD</sub> ≤3.6 V 1.65 V ≤V <sub>DD</sub> ≤2.0 V  1 and 1 and 1 and 2 and 2 and 2 and 3 and	0	3.6	V
VIN I	Input voltage on FT, FTf and RST	2.0 V ≤V <sub>DD</sub> ≤3.6 V	-0.3	5.5	
	pins <sup>(3)</sup>	1.65 V ≤V <sub>DD</sub> ≤2.0 V	-0.3	5.2	V
	Input voltage on BOOT0 pin	-	0	5.5	V
	Input voltage on TC pin	-	0 32 1.65 3.6 -on 1.8 3.6 Ner-on 1.65 3.6  D(1) 1.65 3.6  D(1) 1.8 3.6  3.0 3.6  0 3.6  -0.3 5.5  -0.3 5.5  -0.3 V <sub>DD</sub> +0.3  - 351  - 488  - 313  - 435  - 370  - 714  - 88  - 122  - 78  - 109  - 93		
		UFBGA100	-	351	
		LQFP100	-	488	
	Power dissipation at T <sub>A</sub> = 85 °C	TFBGA64	-	313	
	(range 6) or $T_A = 105  ^{\circ}\text{C}$ (range 7) (4)	LQFP64	-	435	
		LQFP48	0   32   No.		
Б		UFQFPN48	-	714	m\\/
$P_D$		UFBGA100	-	88	mW
		LQFP100	-	122	
	Power dissipation at T <sub>A</sub> = 125 °C (range 3) <sup>(4)</sup>	TFBGA64	-	78	
	(range 3) <sup>(4)</sup>	LQFP64	-	109	
		LQFP48	-	93	]
		UFQFPN48	-	179	]

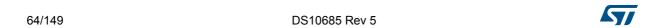


Table 26. General operating conditions (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
Та	Temperature range	Maximum power dissipation (range 6)	-40	85	
		Maximum power dissipation (range 7)	-40	105	
		Maximum power dissipation (range 3)	-40	125	°C
TJ	Junction temperature range (range 6)	-40 °C ≤T <sub>A</sub> ≤85 °	-40	105	
	Junction temperature range (range 7)	-40 °C ≤T <sub>A</sub> ≤105 °C	<del>-4</del> 0	125	
	Junction temperature range (range 3)	-40 °C ≤T <sub>A</sub> ≤125 °C	<del>-4</del> 0	130	

- 1. It is recommended to power  $V_{DD}$  and  $V_{DDA}$  from the same source. A maximum difference of 300 mV between  $V_{DD}$  and  $V_{DDA}$  can be tolerated during power-up and normal operation.
- $V_{DD\_USB} \text{ must respect the following conditions:} \\ \text{When V}_{DD} \text{ is powered-on (V}_{DD} < \text{V}_{DD\_min}), \text{V}_{DD\_USB} \text{ should be always lower than V}_{DD} \\ \text{When V}_{DD} \text{ is powered-down (V}_{DD} < \text{V}_{DD\_min}), \text{V}_{DD\_USB} \text{ should be always lower than V}_{DD} \\ \text{In operating mode, V}_{DD\_USB} \text{ could be lower or higher V}_{DD}. \\ \text{If the USB is not used, V}_{DD\_USB} \text{ must range from V}_{DD\_min} \text{ to V}_{DD\_max} \text{ to be able to use PA11 and PA12 as standard I/Os.} \\ \text{If the USB is not used and PA11/PA12 are not used as standard I/Os, VDD\_USB must be connected to a V}_{SS} \text{ or V}_{DD} \text{ power supply voltage (VDD\_USB must not be left floating).} \\$
- 3. To sustain a voltage higher than  $V_{DD}$ +0.3V, the internal pull-up/pull-down resistors must be disabled.
- If T<sub>A</sub> is lower, higher P<sub>D</sub> values are allowed as long as T<sub>J</sub> does not exceed T<sub>J</sub> max (see Table 90: Thermal characteristics on page 142).



### 6.3.2 Embedded reset and power control block characteristics

The parameters given in the following table are derived from the tests performed under the ambient temperature condition summarized in *Table 26*.

Table 27. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
	V via a time vata	BOR detector enabled	0	-	∞		
t <sub>VDD</sub> <sup>(1)</sup>	V <sub>DD</sub> rise time rate	BOR detector disabled	0	-	1000	] ,,,,,,	
ا VDD`	V fall time rate	BOR detector enabled	20	-	∞	µs/V	
	V <sub>DD</sub> fall time rate	BOR detector disabled	0	-	1000		
т (1)	Reset temporization	V <sub>DD</sub> rising, BOR enabled	-	2	3.3		
T <sub>RSTTEMPO</sub> <sup>(1)</sup>	Reset temporization	V <sub>DD</sub> rising, BOR disabled <sup>(2)</sup>	0.4	0.7	1.6	ms	
V	Power-on/power down reset	Falling edge	1	1.5	1.65		
V <sub>POR/PDR</sub>	threshold	Rising edge	1.3	1.5	1.65		
M	Drawn aut roast throabald 0	Falling edge	1.67	1.7	1.74		
$V_{BOR0}$	Brown-out reset threshold 0	Rising edge	1.69	1.76	1.8		
\/	Drawn aut roast throad ald 4	Falling edge	1.87	1.93	1.97		
V <sub>BOR1</sub>	Brown-out reset threshold 1	Rising edge	1.96	2.03	2.07		
M	Brown-out reset threshold 2	Falling edge	2.22	2.30	2.35		
$V_{BOR2}$		Rising edge	2.31	2.41	2.44		
M	Brown-out reset threshold 3	Falling edge	2.45	2.55	2.6		
$V_{BOR3}$		Rising edge	2.54	2.66	2.7	- V	
V	Brown-out reset threshold 4	Falling edge	2.68	2.8	2.85		
$V_{BOR4}$		Rising edge	2.78	2.9	2.95		
1/	Programmable voltage detector	Falling edge	1.8	1.85	1.88		
$V_{PVD0}$	threshold 0	Rising edge	1.88	1.94	1.99		
	DVD three-bald 4	Falling edge	1.98	2.04	2.09		
$V_{PVD1}$	PVD threshold 1	Rising edge	2.08	2.14	2.18	1	
.,	DVD the sector LLO	Falling edge	2.20	2.24	2.28		
$V_{PVD2}$	PVD threshold 2	Rising edge	2.28	2.34	2.38		
	DVD the sector LLO	Falling edge	2.39	2.44	2.48		
$V_{PVD3}$	PVD threshold 3	Rising edge	2.47	2.54	2.58		
	DVD three shorts 4	Falling edge	2.57	2.64	2.69		
$V_{PVD4}$	PVD threshold 4	Rising edge	2.68	2.74	2.79		
.,	DVD the select of	Falling edge	2.77	2.83	2.88		
$V_{PVD5}$	PVD threshold 5	Rising edge	2.87	2.94	2.99		

Symbol	Parameter	Conditions		Тур	Max	Unit	
V <sub>PVD6</sub>	PVD threshold 6	Falling edge	2.97	3.05	3.09	\ /	
		Rising edge	3.08	3.15	3.20	V	
$V_{hyst}$	Hysteresis voltage	BOR0 threshold	-	40	-		
		All BOR and PVD thresholds excepting BOR0	-	100	-	mV	

Table 27. Embedded reset and power control block characteristics (continued)

### 6.3.3 Embedded internal reference voltage

The parameters given in *Table 29* are based on characterization results, unless otherwise specified.

Table 28. Embedded internal reference voltage calibration values

Calibration value name	Description	Memory address
VREFINT_CAL	Raw data acquired at temperature of 25 °C V <sub>DDA</sub> = 3 V	0x1FF8 0078 - 0x1FF8 0079

Table 29. Embedded internal reference voltage<sup>(1)</sup>

Symbol	Parameter	Conditions		Тур	Max	Unit
V <sub>REFINT out</sub> <sup>(2)</sup>	Internal reference voltage	– 40 °C < T <sub>J</sub> < +125 °C	1.202	1.224	1.242	V
T <sub>VREFINT</sub>	Internal reference startup time	-	-	2	3	ms
V <sub>VREF_MEAS</sub>	V <sub>DDA</sub> and V <sub>REF+</sub> voltage during V <sub>REFINT</sub> factory measure	-	2.99	3	3.01	V
A <sub>VREF_MEAS</sub>	Accuracy of factory-measured V <sub>REFINT</sub> value <sup>(3)</sup>			-	±5	mV
T <sub>Coeff</sub> <sup>(4)</sup>	Temperature coefficient	-40 °C < T <sub>J</sub> < +125 °C	-	25	100	ppm/°C
A <sub>Coeff</sub> <sup>(4)</sup>	Long-term stability	1000 hours, T= 25 °C	-	-	1000	ppm
V <sub>DDCoeff</sub> <sup>(4)</sup>	Voltage coefficient	3.0 V < V <sub>DDA</sub> < 3.6 V	-	-	2000	ppm/V
T <sub>S_vrefint</sub> (4)(5)	ADC sampling time when reading the internal reference voltage	-	5	10	-	μs
T <sub>ADC_BUF</sub> <sup>(4)</sup>	Startup time of reference voltage buffer for ADC	-	-	-	10	μs
I <sub>BUF_ADC</sub> <sup>(4)</sup>	Consumption of reference voltage buffer for ADC	-	-	13.5	25	μА
I <sub>VREF_OUT</sub> <sup>(4)</sup>	VREF_OUT output current <sup>(6)</sup>	-	-	-	1	μΑ
C <sub>VREF_OUT</sub> <sup>(4)</sup>	VREF_OUT output load	-	-	-	50	pF



<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Valid for device version without BOR at power up. Please see option "D" in Ordering information scheme for more details.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>LPBUF</sub> <sup>(4)</sup>	Consumption of reference voltage buffer for VREF_OUT and COMP	-	-	730	1200	nA
V <sub>REFINT_DIV1</sub> <sup>(4)</sup>	1/4 reference voltage	-	24	25	26	
V <sub>REFINT_DIV2</sub> <sup>(4)</sup>	1/2 reference voltage	-	49	50	51	% V <sub>REFINT</sub>
V <sub>REFINT_DIV3</sub> <sup>(4)</sup>	3/4 reference voltage	-	74	75	76	INEI IIVI

Table 29. Embedded internal reference voltage<sup>(1)</sup> (continued)

- Refer to Table 41: Peripheral current consumption in Stop and Standby mode for the value of the internal reference current consumption (I<sub>REFINT</sub>).
- 2. Guaranteed by test in production.
- 3. The internal  $V_{REF}$  value is individually measured in production and stored in dedicated EEPROM bytes.
- 4. Guaranteed by design.
- 5. Shortest sampling time can be determined in the application by multiple iterations.
- 6. To guarantee less than 1% VREF\_OUT deviation.

#### 6.3.4 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code. The current consumption is measured as described in *Figure 13: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to Dhrystone 2.1 code if not specified otherwise.

The current consumption values are derived from the tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26: General operating conditions* unless otherwise specified.

The MCU is placed under the following conditions:

- All I/O pins are configured in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time and prefetch is adjusted depending on fHCLK frequency and voltage range to provide the best CPU performance unless otherwise specified.
- When the peripherals are enabled f<sub>APB1</sub> = f<sub>APB2</sub> = f<sub>APB</sub>
- When PLL is ON, the PLL inputs are equal to HSI = 16 MHz (if internal clock is used) or HSE = 16 MHz (if HSE bypass mode is used)
- The HSE user clock applied to OSCI\_IN input follows the characteristic specified in Table 43: High-speed external user clock characteristics
- For maximum current consumption V<sub>DD</sub> = V<sub>DDA</sub> = 3.6 V is applied to all supply pins
- For typical current consumption V<sub>DD</sub> = V<sub>DDA</sub> = 3.0 V is applied to all supply pins if not specified otherwise

The parameters given in *Table 51*, *Table 26* and *Table 27* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.



Table 30. Current consumption in Run mode, code with data processing running from Flash memory

Symbol	Parameter	Conditio	f <sub>HCLK</sub> (MHz)	Тур	Max <sup>(1)</sup>	Unit	
			Range3,	1	190	250	
			Vcore=1.2 V	2	345	380	μΑ
			VOS[1:0]=11	4	650	670	
		f <sub>HSE</sub> = f <sub>HCLK</sub> up to	Range2,	4	0,8	0,86	
		16MHz included, f <sub>HSE</sub> = f <sub>HCLK</sub> /2 above	Vcore=1.5 V	8	1,55	1,7	mA
I <sub>DD</sub> (Run		16 MHz (PLL ON) <sup>(2)</sup>	VOS[1:0]=10	16	2,95	3,1	
			Range1,	8	1,9	2,1	IIIA
	Supply current in Run		Vcore=1.8 V	16	3,55	3,8	μΑ
from Flash memory)	mode code executed from Flash memory		VOS[1:0]=01 32	32	6,65	7,2	
memory)	nomination y		Range3, Vcore=1.2 V	0,065	39	130	
		MSI clock source		0,524	115	210	
			VOS[1:0]=11	4,2	700	770	
		HSI clock source	Range2, Vcore=1.5 V VOS[1:0]=10	16	2,9	3,2	mA
		(16MHz)	Range1, Vcore=1.8 V VOS[1:0]=01	32	7,15	7,4	IIIA

<sup>1.</sup> Guaranteed by characterization results at 125  $^{\circ}\text{C},$  unless otherwise specified.

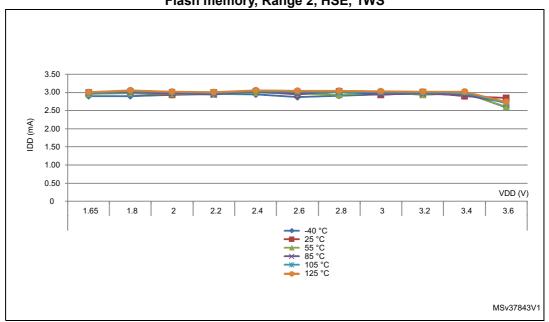
<sup>2.</sup> Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 31. Current consumption in Run mode vs code type, code with data processing running from Flash memory

Symbol	Parameter		Conditions				Unit
				Dhrystone		650	
				CoreMark		655	
			Range 3, $V_{CORE}$ =1.2 V, $VOS[1:0]$ =11	Fibonacci	4 MHz	485	μA
	Supply			while(1)		385	,
	current in Run mode,	f <sub>HSE</sub> = f <sub>HCLK</sub> up to 16 MHz included, f <sub>HSE</sub> = f <sub>HCLK</sub> /2 above 16 MHz (PLL ON) <sup>(1)</sup>		* **		375	
from Flash	code executed			Dhrystone		6,65	
memory)	from Flash memory	WII IZ (FLL ON)		CoreMark		6,9	
mem	memory		Range 1, V <sub>CORE</sub> =1.8 V,	Fibonacci	32 MHz	6,75	mA
			VOS[1:0]=01	while(1)		5,8	
				while(1), prefetch OFF		5,5	

<sup>1.</sup> Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Figure 14.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25/55/85/105 °C, Run mode, code running from Flash memory, Range 2, HSE, 1WS



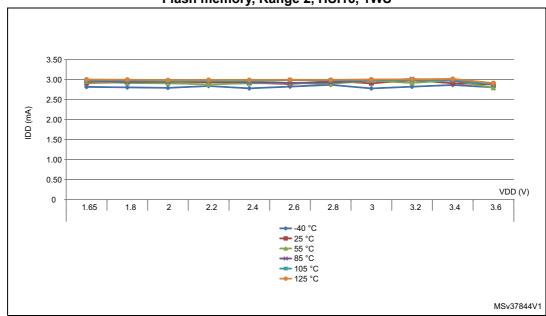


Figure 15.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25/55/85/105 °C, Run mode, code running from Flash memory, Range 2, HSI16, 1WS

Table 32. Current consumption in Run mode, code with data processing running from RAM

Symbol	Parameter	Conditio	f <sub>HCLK</sub> (MHz)	Тур	Max <sup>(1)</sup>	Unit	
			Range3,	1	175	230	
I. (Dur			Vcore=1.2 V	2	315	360	μΑ
			VOS[1:0]=11	4	570	630	
		$f_{HSE} = f_{HCLK}$ up to	Range2,	4	0,71	0,78	
		16 MHz included, $f_{HSE} = f_{HCLK}/2$ above	Vcore=1.5 V	8	1,35	1,6	
		16 MHz (PLL ON) <sup>(2)</sup>	PLL ON) <sup>(2)</sup> VOS[1:0]=10	16	2,7	3	
			Range1,	8	1,7	1,9	IIIA
	Supply current in Run mode code executed from RAM, Flash memory switched off		Vcore=1.8 V VOS[1:0]=01	16	3,2	3,7	
I <sub>DD</sub> (Run from RAM)				32	6,65	7,1	
			Range3,	0,065	38	98	
		MSI clock	Vcore=1.2 V	0,524	105	160	μΑ
			VOS[1:0]=11	4,2	615	710	
		HSI clock source	Range2, Vcore=1.5 V VOS[1:0]=10	16	2,85	3	
		(16 MHz)	Range1, Vcore=1.8 V VOS[1:0]=01	32	6,85	7,3	MA

<sup>1.</sup> Guaranteed by characterization results at 125 °C, unless otherwise specified.



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2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 33. Current consumption in Run mode vs code type, code with data processing running from RAM<sup>(1)</sup>

Symbol	Parameter		Conditions			Тур	Unit
I <sub>DD</sub> (Run from RAM)  Supply current in Run mode, code executed from RAM, Flash memory switched off			Range 3, V <sub>CORE</sub> =1.2 V, VOS[1:0]=11    CoreMark   Fibonacci   While(1)	Dhrystone		570	
				CoreMark	4 8411-	670	
		f -f unto			4 IVIDZ	410	μΑ
	·	thse = thclk up to 16 MHz included, fhse = thclk/2 above 16 MHz (PLL ON) <sup>(2)</sup>		while(1)		375	
	,			Dhrystone		6,65	
	,		Range 1,	CoreMark	32 MHz	6,95	mΛ
			V <sub>CORE</sub> =1.8 V, VOS[1:0]=01	Fibonacci	32 IVITZ	5,9	mA
				while(1)	1	5,2	

<sup>1.</sup> Guaranteed by characterization results, unless otherwise specified.

<sup>2.</sup> Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 34. Current consumption in Sleep mode

Symbol	Parameter	Condition	-	f <sub>HCLK</sub> (MHz)	Тур	Max <sup>(1)</sup>	Unit
			Range3,	1	43,5	110	
			Vcore=1.2 V	2	72	140	
			VOS[1:0]=11	4	130	200	
		$f_{HSE} = f_{HCLK}$ up to 16 MHz included, $f_{HSE} = f_{HCLK}/2$ above	Range2,	4	160	110 140	
			Vcore=1.5 V	8	305	380	
		16 MHz (PLL ON) <sup>(2)</sup>	VOS[1:0]=10	16	590	690	
			Range1,	8	370	460	
	Supply current in Sleep mode, Flash		Vcore=1.8 V	16	715	840	0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0
	memory switched		VOS[1:0]=01	32	1650	2000	
	OFF		Range3,	0,065	18	93	
		MSI clock	Vcore=1.2 V	0,524	31,5	110	
			VOS[1:0]=11	4,2	140	230	
		HSI clock source	Range2, Vcore=1.5 V VOS[1:0]=10	16	665	850	
I <sub>DD</sub>		(16 MHz)	Range1, Vcore=1.8 V VOS[1:0]=01	32	1750	2100	
(Sleep)			Range3,	1	57,5	130	μΑ
			Vcore=1.2 V VOS[1:0]=11	2	84	160	-
				4	150	220	
		f <sub>HSE</sub> = f <sub>HCLK</sub> up to	Range2,	4	170	240	
		16MHz included, $f_{HSE} = f_{HCLK}/2$ above	Vcore=1.5 V	8	315	400	
		16 MHz (PLL ON) <sup>(2)</sup>	VOS[1:0]=10	16	605	710	
			Range1,	8	380	470	
	Supply current in		Vcore=1.8 V	16	730	860	
	Sleep mode, Flash memory switched		VOS[1:0]=01	32	1650	2000	
	ON		Range3,	0,065	29,5	110	
		MSI clock	Vcore=1.2 V	0,524	44,5	120	
			VOS[1:0]=11	4,2	150	240	
		HSI clock source	Range2, Vcore=1.5 V VOS[1:0]=10	16	680	930	
		(16MHz)	Range1, Vcore=1.8 V VOS[1:0]=01	32	1750	2200	

<sup>1.</sup> Guaranteed by characterization results at 125  $^{\circ}\text{C},$  unless otherwise specified.



2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 35. Current consumption in Low-power run mode

Symbol	Parameter		Condition	-	f <sub>HCLK</sub> (MHz)	Тур	Max <sup>(1)</sup>	Unit
				$T_A = -40 \text{ to } 25^{\circ}\text{C}$		9,45	12	
			MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	0,032	14	58	
			f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 105°C	0,032	21	64	
				T <sub>A</sub> = 125°C		36,5	160	
		All peripherals		$T_A = -40 \text{ to } 25^{\circ}\text{C}$		14,5	18	
		OFF, code executed from	MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	0,065	19,5	60	
		RAM, Flash	f <sub>HCLK</sub> = 65kHz	T <sub>A</sub> = 105°C	0,005	26	65	
		memory switched OFF, V <sub>DD</sub> from		T <sub>A</sub> = 125°C		42	160	
		1.65 to 3.6 V		$T_A = -40 \text{ to } 25^{\circ}\text{C}$		26,5	30	
			MSI clock=131 kHz, f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 55°C	0,131	27,5	60	
				T <sub>A</sub> = 85°C		31	66	
	Cummbu			T <sub>A</sub> = 105°C		37,5	77	
I <sub>DD</sub>	Supply current in			T <sub>A</sub> = 125°C		53,5	170	
(LP Run)	Low-power run mode			$T_A = -40 \text{ to } 25^{\circ}\text{C}$		24,5	34	μΑ
	run mode		MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	0,032	30	82	
			f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 105°C	0,032	38,5	90	
				T <sub>A</sub> = 125°C		58	120	
		All peripherals		$T_A = -40 \text{ to } 25^{\circ}\text{C}$		30,5	40	
		OFF, code	MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	0,065	36,5	88	
		executed from Flash memory,	f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 105°C	0,005	45	96	
		VDD from 1.65 V		T <sub>A</sub> = 125°C		64,5	120	
		to 3.6 V		$T_A = -40 \text{ to } 25^{\circ}\text{C}$		45	56	
			MSI clock =	T <sub>A</sub> = 55°C		48	96	
			131 kHz,	T <sub>A</sub> = 85°C	0,131	51	110	
			f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 105°C		59,5	120	
				T <sub>A</sub> = 125°C		79,5	150	

<sup>1.</sup> Guaranteed by characterization results at 125 °C, unless otherwise specified.



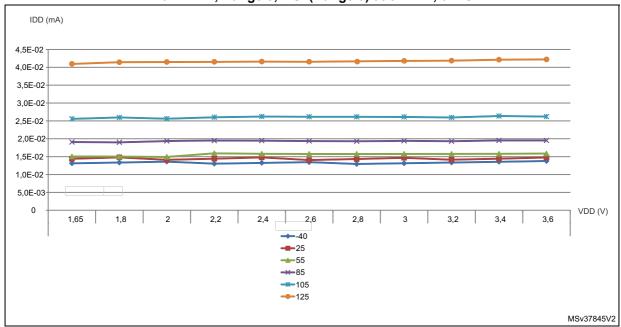


Figure 16.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25 °C, Low-power run mode, code running from RAM, Range 3, MSI (Range 0) at 64 KHz, 0 WS

Table 36. Current consumption in Low-power sleep mode

Symbol	Parameter		Condition				
			MSI clock = 65 kHz, f <sub>HCLK</sub> = 32 kHz, Flash memory OFF	$T_A = -40 \text{ to } 25^{\circ}\text{C}$	4,7	-	
				$T_A = -40 \text{ to } 25^{\circ}\text{C}$	17	24	
			MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	19,5	30	
			f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 105°C	23	47	
		All marinhanala		T <sub>A</sub> = 125°C	32,5	70	
	Supply current in	All peripherals OFF, code		$T_A = -40 \text{ to } 25^{\circ}\text{C}$	17	24	
(LP Sleep)	Low-power sleep mode	executed from Flash memory, V <sub>DD</sub>	MSI clock = 65 kHz,	T <sub>A</sub> = 85°C	20	31	μA
	mode	from 1.65 to 3.6 V	f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 105°C	23,5	47	
				T <sub>A</sub> = 125°C	32,5	70	
				$T_A = -40 \text{ to } 25^{\circ}\text{C}$	19,5	27	
				T <sub>A</sub> = 55°C	20,5	28	
			MSI clock = 131kHz, f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85°C	22,5	33	
			HOLK	T <sub>A</sub> = 105°C	26	50	
				T <sub>A</sub> = 125°C	35	73	

<sup>1.</sup> Guaranteed by characterization results at 125 °C, unless otherwise specified.

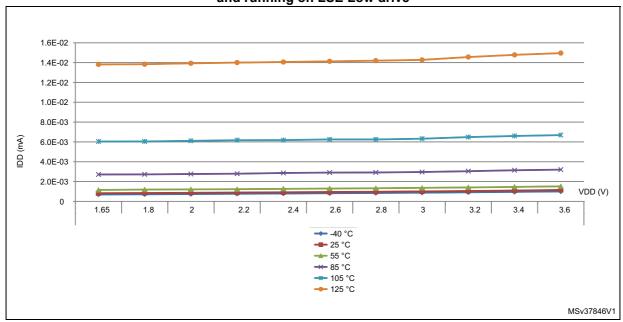


Table 37. Typical and maximum current consumptions in Stop mode

Symbol	Parameter	Conditions	Тур	Max <sup>(1)</sup>	Unit
		$T_A = -40 \text{ to } 25^{\circ}\text{C}$	0,43	1,00	
		T <sub>A</sub> = 55°C	0,735	2,50	
I <sub>DD</sub> (Stop)	Supply current in Stop mode	T <sub>A</sub> = 85°C	2,25	4,90	μA
		T <sub>A</sub> = 105°C	5,3	13,00	
		T <sub>A</sub> = 125°C	12,5	28,00	

<sup>1.</sup> Guaranteed by characterization results at 125 °C, unless otherwise specified.

Figure 17.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25/55/ 85/105/125 °C, Stop mode with RTC enabled and running on LSE Low drive



1.4E-02 1.2E-02 1.0E-02 8.0E-03 IDD (mA) 6.0E-03 4.0E-03 2.0E-03 VDD (V) 2.6 3.2 3.4 3.6 -40 °C **---** 25 °C <u>►</u> 55 °C - 105 °C - 125 °C MSv37847V1

Figure 18.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25/55/85/105/125 °C, Stop mode with RTC disabled, all clocks OFF

Table 38. Typical and maximum current consumptions in Standby mode

Symbol	Parameter	Conditi	Тур	Max <sup>(1)</sup>	Unit	
			$T_A = -40 \text{ to } 25^{\circ}\text{C}$	0,855	1,70	
			T <sub>A</sub> = 55 °C	-	2,90	
		Independent watchdog and LSI enabled	T <sub>A</sub> = 85 °C	-	3,30	
			T <sub>A</sub> = 105 °C	-	4,10	
I <sub>DD</sub>	Supply current in Standby		T <sub>A</sub> = 125 °C	-	8,50	
(Standby)	mode		$T_A = -40 \text{ to } 25^{\circ}\text{C}$	0,29	0,60	μA
			T <sub>A</sub> = 55 °C	0,32	1,20	
		Independent watchdog and LSI OFF	T <sub>A</sub> = 85 °C	0,5	2,30	
			T <sub>A</sub> = 105 °C	0,94	3,00	
			T <sub>A</sub> = 125 °C	2,6	7,00	

<sup>1.</sup> Guaranteed by characterization results at 125 °C, unless otherwise specified

Table 39. Average current consumption during Wakeup

Symbol	parameter	System frequency	Current consumption during wakeup	Unit
		HSI	1	
I <sub>DD</sub> (Wakeup from Stop)		HSI/4	0,7	
	Supply current during Wakeup from Stop mode	MSI clock = 4,2 MHz	0,7	
13347	Stop Medic	MSI clock = 1,05 MHz	0,4	
		MSI clock = 65 KHz	0,1	mA
I <sub>DD</sub> (Reset)	Reset pin pulled down	-	0,21	
I <sub>DD</sub> (Power-up)	BOR ON	-	0,23	
I <sub>DD</sub> (Wakeup from	With Fast wakeup set	MSI clock = 2,1 MHz	0,5	
StandBy)	With Fast wakeup disabled	MSI clock = 2,1 MHz	0,12	

## On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in the following tables. The MCU is placed under the following conditions:

- $\bullet$   $\,$  all I/O pins are in input mode with a static value at  $V_{DD}$  or  $V_{SS}$  (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
  - with all peripherals clocked OFF
  - with only one peripheral clocked on

Table 40. Peripheral current consumption in Run or Sleep mode<sup>(1)</sup>

		Typical	consumption, V	/ <sub>DD</sub> = 3.0 V, T <sub>A</sub> =	25 °C	
Peripheral		Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	Range 3, V <sub>CORE</sub> =1.2 V VOS[1:0] = 11	Low-power sleep and run	Unit
	CRS	2.5	2	2	2	
	DAC1/2	4	3.5	3	2.5	
	I2C1	11	9.5	7.5	9	
	I2C3	11	9	7	9	
	LCD1	4	3.5	3	2.5	
	LPTIM1	10	8.5	6.5	8	
	LPUART1	8	6.5	5.5	6	
	SPI2	9	4.5	3.5	4	
APB1	USART2	14.5	12	9.5	11	μΑ/MHz (f <sub>HCLK</sub> )
	USART4	5	4	3	5	(TICER)
	USART5	5	4	3	5	
	USB	8.5	4.5	4	4.5	
	TIM2	10.5	8.5	7	9	
	TIM3	12	10	8	11	
	TIM6	3.5	3	2.5	2	
	TIM7	3.5	3	2.5	2	
	WWDG	3	2	2	2	



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Table 40. Peripheral current consumption in Run or Sleep mode<sup>(1)</sup> (continued)

	-	Typical	Typical consumption, V <sub>DD</sub> = 3.0 V, T <sub>A</sub> = 25 °C				
Peri	ipheral	Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	Range 3, V <sub>CORE</sub> =1.2 V VOS[1:0] = 11	Low-power sleep and run	Unit	
	ADC1 <sup>(2)</sup>	5.5	5	3.5	4		
	SPI1	4	3	3	2.5		
	USART1	14.5	11.5	9.5	12		
APB2	TIM21	7.5	6	5	5.5	μΑ/MHz	
AFBZ	TIM22	7	6	5	6	(f <sub>HCLK</sub> )	
	FIREWALL	1.5	1	1	0.5		
	DBGMCU	1.5	1	1	0.5		
	SYSCFG	2.5	2	2	1.5		
	GPIOA	3.5	3	2.5	2.5		
	GPIOB	3.5	2.5	2	2.5		
Cortex- M0+ core	GPIOC	8.5	6.5	5.5	7	μΑ/MHz	
I/O port	GPIOD	1	0.5	0.5	0.5	(f <sub>HCLK</sub> )	
	GPIOE	8	6	5	6		
	GPIOH	1.5	1	1	0.5		
	CRC	1.5	1	1	1		
	FLASH	0(3)	0(3)	0(3)	0(3)		
AHB	DMA1	10	8	6.5	8.5	μΑ/MHz (f <sub>HCLK</sub> )	
	RNG	5.5	1	0.5	0.5	('HCLK)	
	TSC	3	2.5	2	3		
All e	enabled	204	162	130	202	μΑ/ΜΗz (f <sub>HCLK</sub> )	
PWR		2.5	2	2	1	μΑ/ΜΗz (f <sub>HCLK</sub> )	

<sup>1.</sup> Data based on differential I<sub>DD</sub> measurement between all peripherals OFF an one peripheral with clock enabled, in the following conditions: f<sub>HCLK</sub> = 32 MHz (range 1), f<sub>HCLK</sub> = 16 MHz (range 2), f<sub>HCLK</sub> = 4 MHz (range 3), f<sub>HCLK</sub> = 64kHz (Low-power run/sleep), f<sub>APB1</sub> = f<sub>HCLK</sub>, f<sub>APB2</sub> = f<sub>HCLK</sub>, default prescaler value for each peripheral. The CPU is in Sleep mode in both cases. No I/O pins toggling. Not tested in production.

<sup>2.</sup> HSI oscillator is OFF for this measure.

<sup>3.</sup> Current consumption is negligible and close to 0  $\mu$ A.

Comphal	Devimberal	Typical consur	nption, T <sub>A</sub> = 25 °C	1114
Symbol	Peripheral –	V <sub>DD</sub> =1.8 V	V <sub>DD</sub> =3.0 V	- Unit
I <sub>DD(PVD / BOR)</sub>	-	0.7	1.2	
I <sub>REFINT</sub>	-	-	1.7	
-	LSE Low drive <sup>(2)</sup>	0.11	0,13	
-	LSI	0.27	0.31	
-	IWDG	0.2	0.3	
-	LPTIM1, Input 100 Hz	0.01	0,01	μΑ
-	LPTIM1, Input 1 MHz	11	12	
-	LPUART1	-	0,5	
-	RTC	0.16	0,3	
-	LCD1 (static duty)	0.15	0.15	
-	LCD1 (1/8 duty)	1.6	2.6	μA

Table 41. Peripheral current consumption in Stop and Standby mode<sup>(1)</sup>

## 6.3.5 Wakeup time from low-power mode

The wakeup times given in the following table are measured with the MSI or HSI16 RC oscillator. The clock source used to wake up the device depends on the current operating mode:

- Sleep mode: the clock source is the clock that was set before entering Sleep mode
- Stop mode: the clock source is either the MSI oscillator in the range configured before entering Stop mode, the HSI16 or HSI16/4.
- Standby mode: the clock source is the MSI oscillator running at 2.1 MHz

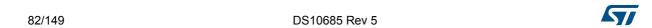
All timings are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.

<sup>1.</sup> LCD, LPTIM, LPUART peripherals can operate in Stop mode but not in Standby mode.

LSE Low drive consumption is the difference between an external clock on OSC32\_IN and a quartz between OSC32\_IN and OSC32\_OUT.-

Table 42. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>WUSLEEP</sub>	Wakeup from Sleep mode	f <sub>HCLK</sub> = 32 MHz	7	8	
t <sub>WUSLEEP_</sub>	Wakeup from Low-power sleep mode,	f <sub>HCLK</sub> = 262 kHz Flash memory enabled	7	8	Number of clock
LP	f <sub>HCLK</sub> = 262 kHz	f <sub>HCLK</sub> = 262 kHz Flash memory switched OFF	9	10	cycles
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz	5.0	8	
	Wakeup from Stop mode, regulator in Run mode	f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	4.9	7	
		f <sub>HCLK</sub> = f <sub>HSI</sub> /4 = 4 MHz	8.0	11	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 1	5.0	8	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 2	5.0	8	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 3	5.0	8	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2.1 MHz	7.3	13	
t <sub>WUSTOP</sub>	Wakeup from Stop mode, regulator in low-power mode	f <sub>HCLK</sub> = f <sub>MSI</sub> = 1.05 MHz	13	23	
	power mode	f <sub>HCLK</sub> = f <sub>MSI</sub> = 524 kHz	28	38	μs
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 262 kHz	51	65	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 131 kHz	100	120	
		f <sub>HCLK</sub> = MSI = 65 kHz	190	260	
		f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	4.9	7	
		$f_{HCLK} = f_{HSI}/4 = 4 \text{ MHz}$	8.0	11	
		f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	4.9	7	
	Wakeup from Stop mode, regulator in low- power mode, code running from RAM	$f_{HCLK} = f_{HSI}/4 = 4 \text{ MHz}$	7.9	10	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz	4.7	8	
t	Wakeup from Standby mode FWU bit = 1	f <sub>HCLK</sub> = MSI = 2.1 MHz	65	130	
twustdby	Wakeup from Standby mode FWU bit = 0	f <sub>HCLK</sub> = MSI = 2.1 MHz	2.2	3	ms



#### 6.3.6 **External clock source characteristics**

## High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO. The external clock signal has to respect the I/O characteristics in Section 6.3.12. However, the recommended clock input waveform is shown in Figure 19.

Table 43. High-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	User external clock source	CSS is ON or PLL is used	1	8	32	MHz
f <sub>HSE_ext</sub> V <sub>HSEH</sub> V <sub>HSEL</sub>	frequency	CSS is OFF, PLL not used	0	8	32	MHz
V <sub>HSEH</sub>	OSC_IN input pin high level voltage		0.7V <sub>DD</sub>	-	V <sub>DD</sub>	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage		V <sub>SS</sub>	ı	0.3V <sub>DD</sub>	v
$\begin{array}{c} t_{\text{w(HSE)}} \\ t_{\text{w(HSE)}} \end{array}$	OSC_IN high or low time		12	ı	-	ns
$t_{r(HSE)} \ t_{f(HSE)}$	OSC_IN rise or fall time	-	-	-	20	113
C <sub>in(HSE)</sub>	OSC_IN input capacitance		-	2.6	-	pF
DuCy <sub>(HSE)</sub>	Duty cycle		45	ı	55	%
IL	OSC_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	μΑ

<sup>1.</sup> Guaranteed by design.

 $V_{\mathsf{HSEH}}$ 90% 10%  $V_{HSEL}$ -T<sub>HSE</sub>  $f_{\mathsf{HSE\_ext}}$ EXTERNAL CLOCK SOURCE OSC\_IN STM32Lxx ai18232c

Figure 19. High-speed external clock source AC timing diagram

## Low-speed external user clock generated from an external source

The characteristics given in the following table result from tests performed using a lowspeed external clock source, and under ambient temperature and supply voltage conditions summarized in Table 26.

Table 44. Low-speed external user clock characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User external clock source frequency		1	32.768	1000	kHz
V <sub>LSEH</sub>	OSC32_IN input pin high level voltage		0.7V <sub>DD</sub>	-	V <sub>DD</sub>	V
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage	-	V <sub>SS</sub>	-	0.3V <sub>DD</sub>	V
t <sub>w(LSE)</sub>	OSC32_IN high or low time		465	-	-	ns
$\begin{matrix} t_{r(LSE)} \\ t_{f(LSE)} \end{matrix}$	OSC32_IN rise or fall time		-	-	10	113
C <sub>IN(LSE)</sub>	OSC32_IN input capacitance	-	-	0.6	-	pF
DuCy <sub>(LSE)</sub>	Duty cycle	-	45	-	55	%
IL	OSC32_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	μΑ

<sup>1.</sup> Guaranteed by design, not tested in production

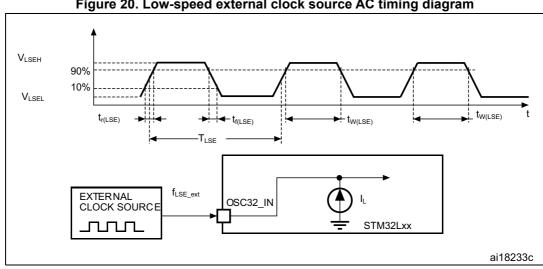


Figure 20. Low-speed external clock source AC timing diagram

### High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 1 to 25 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 45*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>OSC_IN</sub>	Oscillator frequency	-	1		25	MHz
R <sub>F</sub>	Feedback resistor	-	-	200	-	kΩ
G <sub>m</sub>	Maximum critical crystal transconductance	Startup	-	-	700	μA /V
t <sub>SU(HSE)</sub>	Startup time	V <sub>DD</sub> is stabilized	i	2	1	ms

Table 45. HSE oscillator characteristics<sup>(1)</sup>

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 21*).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ . Refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website *www.st.com*.

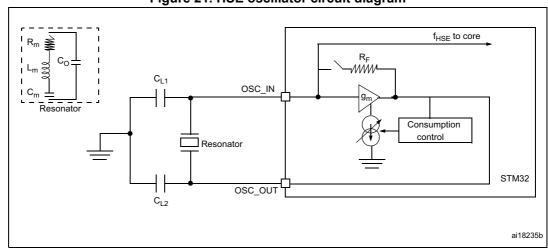


Figure 21. HSE oscillator circuit diagram

<sup>1.</sup> Guaranteed by design.

Guaranteed by characterization results. t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

### Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 46*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions <sup>(2)</sup>	Min <sup>(2)</sup>	Тур	Max	Unit
f <sub>LSE</sub>	LSE oscillator frequency		-	32.768	-	kHz
	Maximum critical crystal	LSEDRV[1:0]=00 lower driving capability	-	-	0.5	
		LSEDRV[1:0]= 01 medium low driving capability	-	-	0.75	uA/V
G <sub>m</sub>	transconductance	LSEDRV[1:0] = 10 medium high driving capability	-	-	1.7	μΑνν
		LSEDRV[1:0]=11 higher driving capability	-	-	2.7	
t <sub>SU(LSE)</sub> <sup>(3)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	S

Table 46. LSE oscillator characteristics<sup>(1)</sup>

- Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
- Guaranteed by characterization results. t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer. To increase speed, address a lower-drive quartz with a high- driver mode.

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

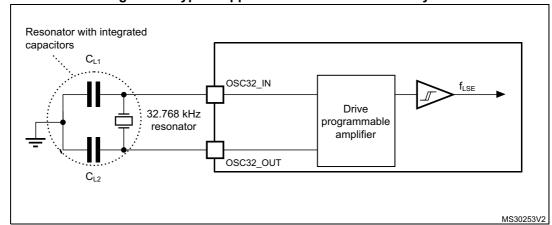


Figure 22. Typical application with a 32.768 kHz crystal

Note: An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.

<sup>1.</sup> Guaranteed by design.

### 6.3.7 Internal clock source characteristics

The parameters given in *Table 47* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.

## High-speed internal 16 MHz (HSI16) RC oscillator

Table 47. 16 MHz HSI16 oscillator characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI16</sub>	Frequency	V <sub>DD</sub> = 3.0 V	-	16	-	MHz
TRIM <sup>(1)(2)</sup>	HSI16 user-	Trimming code is not a multiple of 16	-	±0.4	0.7	%
TRIM` ^` /	trimmed resolution	Trimming code is a multiple of 16	-	-	±1.5	%
	$V_{DDA}$ = 3.0 V, $T_A$ = 25 °C	-1 <sup>(3)</sup>	-	1 <sup>(3)</sup>	%	
	Accuracy of the	V <sub>DDA</sub> = 3.0 V, T <sub>A</sub> = 0 to 55 °C	-1.5	-	1.5	%
۸۵۵		$V_{DDA}$ = 3.0 V, $T_A$ = -10 to 70 °C	-2	-	2	%
ACC <sub>HSI16</sub>	factory-calibrated HSI16 oscillator	$V_{DDA}$ = 3.0 V, $T_{A}$ = -10 to 85 °C	-2.5	-	2	%
	Tiorro oscillator	$V_{DDA}$ = 3.0 V, $T_{A}$ = -10 to 105 °C	-4	-	2	%
		V <sub>DDA</sub> = 1.65 V to 3.6 V T <sub>A</sub> = -40 to 125 °C	-5.45	-	3.25	%
t <sub>SU(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator startup time	-	-	3.7	6	μs
I <sub>DD(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator power consumption	-	-	100	140	μΑ

<sup>1.</sup> The trimming step differs depending on the trimming code. It is usually negative on the codes which are multiples of 16 (0x00, 0x10, 0x20, 0x30...0xE0).

- 2. Guaranteed by characterization results.
- 3. Guaranteed by test in production.

Figure 23. HSI16 minimum and maximum value versus temperature 4.009 3.00% 2.00% 1.65V min 0,00 3V typ 60 20 40 120 3.6V max 1.65V max ■ 3.6V min 4 00 -5.00% -6.00% MSv34791V1

## High-speed internal 48 MHz (HSI48) RC oscillator

Table 48. HSI48 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI48</sub>	Frequency		-	48	-	MHz
TRIM	HSI48 user-trimming step		0.09 <sup>(2)</sup>	0.14	0.2 <sup>(2)</sup>	%
DuCy <sub>(HSI48)</sub>	Duty cycle		45 <sup>(2)</sup>	-	55 <sup>(2)</sup>	%
ACC <sub>HSI48</sub>	Accuracy of the HSI48 oscillator (factory calibrated before CRS calibration)	T <sub>A</sub> = 25 °C	-4 <sup>(3)</sup>	-	4 <sup>(3)</sup>	%
t <sub>su(HSI48)</sub>	HSI48 oscillator startup time		-	-	6 <sup>(2)</sup>	μs
I <sub>DDA(HSI48)</sub>	HSI48 oscillator power consumption		-	330	380 <sup>(2)</sup>	μΑ

- 1.  $V_{DDA}$  = 3.3 V,  $T_A$  = -40 to 125 °C unless otherwise specified.
- 2. Guaranteed by design.
- 3. Guaranteed by characterization results.

### Low-speed internal (LSI) RC oscillator

Table 49. LSI oscillator characteristics

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>LSI</sub> <sup>(1)</sup>	LSI frequency	26	38	56	kHz
D <sub>LSI</sub> <sup>(2)</sup>	LSI oscillator frequency drift 0°C ≤T <sub>A</sub> ≤85°C	-10	-	4	%
t <sub>su(LSI)</sub> (3)	LSI oscillator startup time	-	-	200	μs
I <sub>DD(LSI)</sub> <sup>(3)</sup>	LSI oscillator power consumption	-	400	510	nA

- 1. Guaranteed by test in production.
- 2. This is a deviation for an individual part, once the initial frequency has been measured.
- 3. Guaranteed by design.

### Multi-speed internal (MSI) RC oscillator

Table 50. MSI oscillator characteristics

Symbol	Parameter	Condition	Тур	Max	Unit
		MSI range 0	65.5	-	
		MSI range 1	131	-	kHz
		MSI range 2	262	-	NI IZ
f <sub>MSI</sub>	Frequency after factory calibration, done at V <sub>DD</sub> = 3.3 V and T <sub>A</sub> = 25 °C	MSI range 3	524	-	
	TOD ON TANK TA	MSI range 4	1.05	-	
		MSI range 5	2.1	-	MHz
		MSI range 6	4.2	-	



Table 50. MSI oscillator characteristics (continued)

Symbol	Parameter	Condition	Тур	Max	Unit
ACC <sub>MSI</sub>	Frequency error after factory calibration	-	±0.5	-	%
	MSI oscillator frequency drift 0 °C ≤T <sub>A</sub> ≤85 °C	-	±3	-	
		MSI range 0	- 8.9	+7.0	
		MSI range 1	- 7.1	+5.0	
D <sub>TEMP(MSI)</sub> <sup>(1)</sup>		MSI range 2	- 6.4	+4.0	%
TEMP (M31)	MSI oscillator frequency drift $V_{DD} = 3.3 \text{ V}, -40 \text{ °C} \leq T_A \leq 110 \text{ °C}$	MSI range 3	- 6.2	+3.0	
	TDD	MSI range 4	- 5.2	+3.0	
		MSI range 5	- 4.8	+2.0	
		MSI range 6	- 4.7	+2.0	
D <sub>VOLT(MSI)</sub> <sup>(1)</sup>	MSI oscillator frequency drift 1.65 V ≤V <sub>DD</sub> ≤3.6 V, T <sub>A</sub> = 25 °C	-	-	2.5	%/V
		MSI range 0	0.75	-	
	MSI oscillator power consumption	MSI range 1	1	-	
		MSI range 2	1.5	-	
I <sub>DD(MSI)</sub> <sup>(2)</sup>		MSI range 3	2.5	-	μA
		MSI range 4	4.5	-	
		MSI range 5	8	-	
		MSI range 6	15	-	
		MSI range 0	30	-	
		MSI range 1	20	-	
		MSI range 2	15	-	
		MSI range 3	10	-	
t	MSI oscillator startup time	MSI range 4	6	-	116
<sup>t</sup> su(MSI)	INIST OSCINATOR STATUP TIME	MSI range 5	5	-	μs
		MSI range 6, Voltage range 1 and 2	3.5	-	
		MSI range 6, Voltage range 3	5	-	

Symbol	Parameter	Condition	Тур	Max	Unit
Symbol t <sub>STAB(MSI)</sub> (2)		MSI range 0	-	40	
		MSI range 1	-	20	
		MSI range 2	-	10	
	MSI oscillator stabilization time	MSI range 3	-	4	
(2)		MSI range 4	-	2.5	0
<sup>L</sup> STAB(MSI)` ′		MSI range 5	-	2	μs
		MSI range 6, Voltage range 1 and 2	-	2	
		MSI range 3, Voltage range 3	-	3	
f <sub>OVER(MSI)</sub>	MSI oscillator frequency overshoot	Any range to range 5	-	4	MHz
	INOI OSCIIIAIOI ITEQUETICY OVEISTIOOI	Any range to range 6	-	6	IVII IZ

Table 50. MSI oscillator characteristics (continued)

### 6.3.8 PLL characteristics

The parameters given in *Table 51* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.

Value **Symbol Parameter** Unit Max<sup>(1)</sup> Min Typ PLL input clock<sup>(2)</sup> MHz 2 24 f<sub>PLL IN</sub> % PLL input clock duty cycle 45 55 PLL output clock 2 32 MHz f<sub>PLL\_OUT</sub> PLL input = 16 MHz t<sub>LOCK</sub> 115 160 μs PLL VCO = 96 MHz Jitter ±600 Cycle-to-cycle jitter ps I<sub>DDA</sub>(PLL) Current consumption on V<sub>DDA</sub> 220 450 μΑ I<sub>DD</sub>(PLL) Current consumption on V<sub>DD</sub> 120 150

Table 51. PLL characteristics

<sup>1.</sup> This is a deviation for an individual part, once the initial frequency has been measured.

<sup>2.</sup> Guaranteed by characterization results.

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by  $f_{\text{PLL\_OUT}}$ .

## 6.3.9 Memory characteristics

### **RAM** memory

Table 52. RAM and hardware registers

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
VRM	Data retention mode <sup>(1)</sup>	STOP mode (or RESET)	1.65	-	-	V

Minimum supply voltage without losing data stored in RAM (in Stop mode or under Reset) or in hardware registers (only in Stop mode).

### Flash memory and data EEPROM

Table 53. Flash memory and data EEPROM characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
$V_{DD}$	Operating voltage Read / Write / Erase	-	1.65	-	3.6	V
+	Programming time for	Erasing	-	3.28	3.94	ms
<sup>t</sup> prog	word or half-page	Programming	-	3.28	3.94	1115
	Average current during the whole programming / erase operation		-	500	700	μA
I <sub>DD</sub>	Maximum current (peak) during the whole programming / erase operation	T <sub>A</sub> = 25 °C, V <sub>DD</sub> = 3.6 V	-	1.5	2.5	mA

<sup>1.</sup> Guaranteed by design.

Table 54. Flash memory and data EEPROM endurance and retention

Symbol	Parameter	Conditions	Value	Unit
Symbol	raiailletei	Conditions	Value Min <sup>(1)</sup> 10 100 0.2	Oille
	Cycling (erase / write) Program memory	T <sub>A</sub> = -40°C to 105 °C	10	
N <sub>CYC</sub> <sup>(2)</sup>	Cycling (erase / write) EEPROM data memory		100	kcycles
INCYC.	Cycling (erase / write) Program memory		0.2	RCYCIES
	Cycling (erase / write) EEPROM data memory	T <sub>A</sub> = -40°C to 125 °C	2	

Symbol	Parameter	Conditions	Value Min <sup>(1)</sup>	Unit
	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 85 °C	T - 105 °C	30	
	Data retention (EEPROM data memory) after 100 kcycles at T <sub>A</sub> = 85 °C	-T <sub>RET</sub> = +85 °C	30	
t <sub>RET</sub> <sup>(2)</sup>	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 105 °C	T <sub>RFT</sub> = +105 °C		Vooro
'RET` ′	Data retention (EEPROM data memory) after 100 kcycles at T <sub>A</sub> = 105 °C	1 RET - +103 C	10	years
	Data retention (program memory) after 200 cycles at T <sub>A</sub> = 125 °C	T - 1125 °C	10	
	Data retention (EEPROM data memory) after 2 kcycles at T <sub>A</sub> = 125 °C	T <sub>RET</sub> = +125 °C		

Table 54. Flash memory and data EEPROM endurance and retention (continued)

#### 6.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 55*. They are based on the EMS levels and classes defined in application note AN1709.

Symbol	Parameter	Conditions	Level/ Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP100, $T_A$ = +25 °C, $f_{HCLK}$ = 32 MHz conforms to IEC 61000-4-2	3B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on V <sub>DD</sub> and V <sub>SS</sub> pins to induce a functional disturbance	$V_{DD} = 3.3 \text{ V, LQFP100, T}_{A} = +25$ °C, $f_{HCLK} = 32 \text{ MHz}$ conforms to IEC 61000-4-4	4A

**Table 55. EMS characteristics** 

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> Characterization is done according to JEDEC JESD22-A117.

### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical data corruption (control registers...)

#### Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

#### **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 56. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. frequency range at 32 MHz	Unit
		V 3 6 V	0.1 to 30 MHz	-7	
	Peak level	$V_{DD} = 3.6 \text{ V},$ $T_A = 25 ^{\circ}\text{C},$	30 to 130 MHz	14	dΒμV
S <sub>EMI</sub>	reak level	LQFP100 package   130 MHz to 1 GH	130 MHz to 1 GHz	9	
		compliant with IEC 61967-2	EMI Level	2	-



## 6.3.11 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 57. ESD absolute maximum ratings

Symbol	Ratings	Ratings Conditions		Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	T <sub>A</sub> = +25 °C, conforming to ANSI/JEDEC JS-001	2	2000	V
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C, conforming to ANSI/ESD STM5.3.1.	C4	500	V

<sup>1.</sup> Guaranteed by characterization results.

### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 58. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +125 °C conforming to JESD78A	II level A



### 6.3.12 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DD}$  (for standard pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of  $-5 \mu A/+0 \mu A$  range), or other functional failure (for example reset occurrence oscillator frequency deviation).

The test results are given in the Table 59.

Table 59. I/O current injection susceptibility

		Functional s	Functional susceptibility		
Symbol	Description	Negative injection	Positive injection	Unit	
	Injected current on BOOT0	-0	NA <sup>(1)</sup>		
I <sub>INJ</sub>	Injected current on PA0, PA4, PA5, PC15, PH0 and PH1	-5	0	mA	
1145	Injected current on any other FT, FTf pins	-5 <sup>(2)</sup>	NA <sup>(1)</sup>		
	Injected current on any other pins	-5 <sup>(2)</sup>	+5		

<sup>1.</sup> Current injection is not possible.

<sup>2.</sup> It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

## 6.3.13 I/O port characteristics

## General input/output characteristics

Unless otherwise specified, the parameters given in *Table 60* are derived from tests performed under the conditions summarized in *Table 26*. All I/Os are CMOS and TTL compliant.

Table 60. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL</sub>	Input low level voltage	TC, FT, FTf, RST I/Os	-	-	0.3V <sub>DD</sub>	
"-	-	BOOT0 pin	-	-	0.14V <sub>DD</sub> <sup>(1)</sup>	
V <sub>IH</sub>	Input high level voltage	All I/Os	0.7 V <sub>DD</sub>	-	-	V
	I/O Schmitt trigger voltage hysteresis	Standard I/Os	-	10% V <sub>DD</sub> <sup>(3)</sup>	-	
V <sub>hys</sub>	(2)	BOOT0 pin	-	0.01	-	
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> All I/Os except for PA11, PA12, BOOT0 and FTf I/Os	-	-	±50	
		V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> , PA11 and PA12 I/Os	-	-	-50/+250	nA
		PA11 and PA12 I/Os  V <sub>SS</sub> ≤V <sub>IN</sub> ≤V <sub>DD</sub> FTf I/Os	-	±100		
l <sub>lkg</sub>	Input leakage current (4)	V <sub>DD</sub> ≤V <sub>IN</sub> ≤5 V All I/Os except for PA11, PA12, BOOT0 and FTf I/Os	-	-	200	nA
		V <sub>DD</sub> ≤V <sub>IN</sub> ≤5 V FTf I/Os	-	-	500	
		V <sub>DD</sub> ≤V <sub>IN</sub> ≤5 V PA11, PA12 and BOOT0	-	-	10	μА
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(5)</sup>	$V_{IN} = V_{SS}$	25	45	65	kΩ
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(5)</sup>	$V_{IN} = V_{DD}$	25	45	65	kΩ
C <sub>IO</sub>	I/O pin capacitance	-	-	5	-	pF

<sup>1.</sup> Guaranteed by characterization.

<sup>2.</sup> Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results.

<sup>3.</sup> With a minimum of 200 mV. Guaranteed by characterization results.

<sup>4.</sup> The max. value may be exceeded if negative current is injected on adjacent pins.

Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This MOS/NMOS contribution to the series resistance is minimum (~10% order).

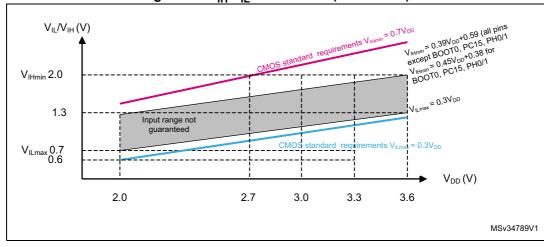
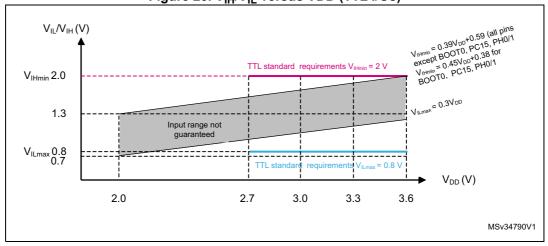


Figure 24. V<sub>IH</sub>/V<sub>IL</sub> versus VDD (CMOS I/Os)

Figure 25. V<sub>IH</sub>/V<sub>IL</sub> versus VDD (TTL I/Os)



### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 15$  mA with the non-standard  $V_{OL}/V_{OH}$  specifications given in *Table 61*.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 6.2*:

- The sum of the currents sourced by all the I/Os on  $V_{DD}$ , plus the maximum Run consumption of the MCU sourced on  $V_{DD}$ , cannot exceed the absolute maximum rating  $I_{VDD(\Sigma)}$  (see *Table 24*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub> plus the maximum Run consumption of the MCU sunk on V<sub>SS</sub> cannot exceed the absolute maximum rating I<sub>VSS(Σ)</sub> (see *Table 24*).

### **Output voltage levels**

Unless otherwise specified, the parameters given in *Table 61* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*. All I/Os are CMOS and TTL compliant.

Table 61. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup> ,	-	0.4	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	$10^{-40} \text{ IIA}$ $2.7 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	V <sub>DD</sub> -0.4	-	
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	Output low level voltage for an I/O  TTL port <sup>(2)</sup> , I <sub>IO</sub> =+ 8 mA		0.4	
V <sub>OH</sub> (3)(4)	Output high level voltage for an I/O pin	TTL port <sup>(2)</sup> , $I_{IO} = -6 \text{ mA}$ $2.7 \text{ V} \le V_{DD} \le 3.6 \text{ V}$	2.4	-	
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	$I_{IO}$ = +15 mA 2.7 V $\leq$ V <sub>DD</sub> $\leq$ 3.6 V	-	1.3	V
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	$I_{IO}$ = -15 mA 2.7 V $\leq$ V <sub>DD</sub> $\leq$ 3.6 V	V <sub>DD</sub> -1.3	-	
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub> = +4 mA 1.65 V ≤V <sub>DD</sub> < 3.6 V	-	0.45	
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	$I_{IO} = -4 \text{ mA}$ 1.65 V $\leq$ V <sub>DD</sub> $\leq$ 3.6 V	V <sub>DD</sub> -0.45	-	
V <sub>OLFM+</sub> (1)(4)	Output low level voltage for an FTf	$I_{IO} = 20 \text{ mA}$ 2.7 V $\leq$ V <sub>DD</sub> $\leq$ 3.6 V	-	0.4	
VOLFM+`^^	I/O pin in Fm+ mode	I <sub>IO</sub> = 10 mA 1.65 V ≤V <sub>DD</sub> ≤ 3.6 V	-	0.4	

The I<sub>IO</sub> current sunk by the device must always respect the absolute maximum rating specified in *Table 24*.
The sum of the currents sunk by all the I/Os (I/O ports and control pins) must always be respected and must not exceed ΣI<sub>IO(PIN)</sub>.

<sup>2.</sup> TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

<sup>3.</sup> The  $I_{IO}$  current sourced by the device must always respect the absolute maximum rating specified in *Table 24*. The sum of the currents sourced by all the I/Os (I/O ports and control pins) must always be respected and must not exceed  $\Sigma I_{IO(PIN)}$ .

<sup>4.</sup> Guaranteed by characterization results.

### Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 26* and *Table 62*, respectively.

Unless otherwise specified, the parameters given in *Table 62* are derived from tests performed under ambient temperature and V<sub>DD</sub> supply voltage conditions summarized in *Table 26*.

Table 62. I/O AC characteristics<sup>(1)</sup>

OSPEEDRx[1:0]	Symbol	Parameter	Conditions	Min	Max <sup>(2)</sup>	Unit	
bit value <sup>(1)</sup>	Cymbol	raramotor	Conditions		Max	Oiiit	
	f	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	400	kHz	
00	f <sub>max(IO)out</sub>	Maximum frequency	$C_L = 50 \text{ pF}, V_{DD} = 1.65 \text{ V to } 2.7 \text{ V}$	-	100	KI IZ	
	t <sub>f(IO)out</sub>	Output rise and fall time	$C_L = 50 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	125	ns	
	t <sub>r(IO)out</sub>	Output rise and fail time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	320	115	
	f	Maximum frequency <sup>(3)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	2	MHz	
04	f <sub>max(IO)out</sub>	maximum frequency.	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	0.6	IVITZ	
01	t <sub>f(IO)out</sub>	Outrant size and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	30		
	t <sub>r(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	65	ns	
	_	(3)	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	10		
10	F <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	- 2 N	MHz	
10	t <sub>f(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	13		
	t <sub>r(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	28	ns	
	F	Marian (3)	C <sub>L</sub> = 30 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	35	NAL 1-	
44	F <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	10	MHz	
11	t <sub>f(IO)out</sub>	Outrot since and fall times	C <sub>L</sub> = 30 pF, V <sub>DD</sub> = 2.7 V to 3.6 V	-	6		
	t <sub>r(IO)out</sub>	Output rise and fall time	C <sub>L</sub> = 50 pF, V <sub>DD</sub> = 1.65 V to 2.7 V	-	17	ns	
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	1	MHz	
	t <sub>f(IO)out</sub>	Output fall time	$C_L = 50 \text{ pF}, V_{DD} = 2.5 \text{ V to } 3.6 \text{ V}$	-	10		
Fm+	t <sub>r(IO)out</sub>	Output rise time		-	30	ns	
configuration <sup>(4)</sup>	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	350	KHz	
	t <sub>f(IO)out</sub>	Output fall time	$C_L = 50 \text{ pF}, V_{DD} = 1.65 \text{ V to } 3.6 \text{ V}$	-	15		
	t <sub>r(IO)out</sub>	Output rise time		-	60	ns	
-	t <sub>EXTIpw</sub>	Pulse width of external signals detected by the EXTI controller	-	8	-	ns	

<sup>1.</sup> The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the line reference manual for a description of GPIO Port configuration register.

<sup>4.</sup> When Fm+ configuration is set, the I/O speed control is bypassed. Refer to the line reference manual for a detailed description of Fm+ I/O configuration.



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<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> The maximum frequency is defined in Figure 26.

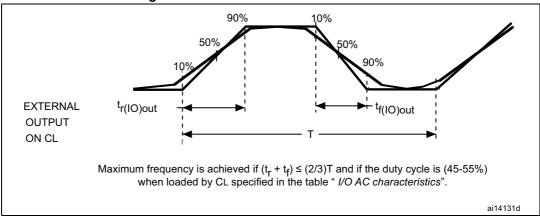


Figure 26. I/O AC characteristics definition

## 6.3.14 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R<sub>PU</sub>, except when it is internally driven low (see *Table 63*).

Unless otherwise specified, the parameters given in *Table 63* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL(NRST)</sub> <sup>(1)</sup>	NRST input low level voltage	-	$V_{SS}$	-	0.8	
V <sub>IH(NRST)</sub> <sup>(1)</sup>	NRST input high level voltage	-	1.4	-	$V_{DD}$	
V <sub>OL(NRST)</sub> <sup>(1)</sup>	NRST output low level	I <sub>OL</sub> = 2 mA 2.7 V < V <sub>DD</sub> < 3.6 V	-	-	0.4	٧
VOL(NRST)	voltage	I <sub>OL</sub> = 1.5 mA 1.65 V < V <sub>DD</sub> < 2.7 V	-	-	0.4	
V <sub>hys(NRST)</sub> <sup>(1)</sup>	NRST Schmitt trigger voltage hysteresis	-	-	10%V <sub>DD</sub> <sup>(2)</sup>	-	mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(3)</sup>	$V_{IN} = V_{SS}$	25	45	65	kΩ
V <sub>F(NRST)</sub> <sup>(1)</sup>	NRST input filtered pulse	-	-	-	50	ns
V <sub>NF(NRST)</sub> <sup>(1)</sup>	NRST input not filtered pulse	-	350	-	-	ns

Table 63. NRST pin characteristics

<sup>1.</sup> Guaranteed by design.

<sup>2. 200</sup> mV minimum value

<sup>3.</sup> The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is around 10%.

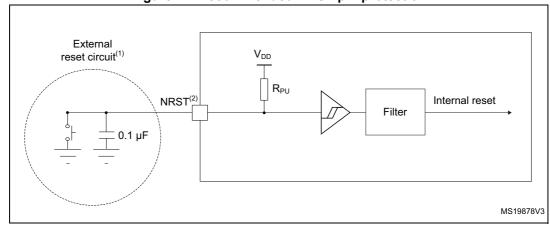


Figure 27. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- 2. The external capacitor must be placed as close as possible to the device.
- The user must ensure that the level on the NRST pin can go below the V<sub>IL(NRST)</sub> max level specified in Table 63. Otherwise the reset will not be taken into account by the device.

### 6.3.15 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 64* are derived from tests performed under ambient temperature, f<sub>PCLK</sub> frequency and V<sub>DDA</sub> supply voltage conditions summarized in *Table 26: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

Table 64. ADC characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	Analog supply voltage for	Fast channel	1.65	-	3.6	V
$V_{DDA}$	ADC ON	Standard channel	1.75 <sup>(1)</sup>	-	3.6	V
V <sub>REF+</sub>	Positive reference voltage	-	1.65		$V_{DDA}$	V
V <sub>REF-</sub>	Negative reference voltage	-	-	0	-	
	Current consumption of the	1.14 Msps	-	200	-	
,	ADC on V <sub>DDA</sub> and V <sub>REF+</sub>	10 ksps	-	40	-	
IDDA (ADC)	Current consumption of the	1.14 Msps	-	70	-	μA
	ADC on V <sub>DD</sub> <sup>(2)</sup>	10 ksps	-	1	-	
		Voltage scaling Range 1	0.14	-	16	
f <sub>ADC</sub>	ADC clock frequency	Voltage scaling Range 2	0.14	-	8	MHz
		Voltage scaling Range 3	0.14	-	4	
f <sub>S</sub> <sup>(3)</sup>	Sampling rate	12-bit resolution	0.01	-	1.14	MHz
f <sub>TRIG</sub> (3)	External trigger frequency	f <sub>ADC</sub> = 16 MHz, 12-bit resolution	-	-	941	kHz
		-	-	-	17	1/f <sub>ADC</sub>
V <sub>AIN</sub>	Conversion voltage range	-	0	-	V <sub>REF+</sub>	V

Table 64. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R <sub>AIN</sub> <sup>(3)</sup>	External input impedance	See Equation 1 and Table 65 for details	-	-	50	kΩ
R <sub>ADC</sub> (3)(4)	Sampling switch resistance	-	1		kΩ	
C <sub>ADC</sub> <sup>(3)</sup>	Internal sample and hold capacitor	-	-	-	8	pF
t <sub>CAL</sub> (3)(5)	Calibration time	f <sub>ADC</sub> = 16 MHz		5.2		μs
CAL	Cambration time	-		83		1/f <sub>ADC</sub>
		ADC clock = HSI16	1.5 ADC cycles + 2 f <sub>PCLK</sub> cycles	-	1.5 ADC cycles + 3 f <sub>PCLK</sub> cycles	-
W <sub>LATENCY</sub> <sup>(6)</sup>	ADC_DR register write latency	ADC clock = PCLK/2	-	4.5	-	f <sub>PCLK</sub> cycle
		ADC clock = PCLK/4	-	8.5	-	f <sub>PCLK</sub> cycle
		$f_{ADC} = f_{PCLK}/2 = 16 \text{ MHz}$	0.266			μs
		$f_{ADC} = f_{PCLK}/2$	8.5			1/f <sub>PCLK</sub>
t <sub>latr</sub> (3)	Trigger conversion latency	$f_{ADC} = f_{PCLK}/4 = 8 \text{ MHz}$	0.516			μs
		$f_{ADC} = f_{PCLK}/4$	16.5			1/f <sub>PCLK</sub>
		$f_{ADC} = f_{HSI16} = 16 \text{ MHz}$	0.252	-	0.260	μs
Jitter <sub>ADC</sub>	ADC jitter on trigger conversion	f <sub>ADC</sub> = f <sub>HSI16</sub>	-	1	-	1/f <sub>HSI16</sub>
t <sub>S</sub> <sup>(3)</sup>	Sampling time	f <sub>ADC</sub> = 16 MHz	0.093	ı	10.03	μs
ıs, ,	Sampling time	-	1.5	-	160.5	1/f <sub>ADC</sub>
t <sub>UP_LDO</sub> (3)(5)	Internal LDO power-up time	-	10		10	μs
t <sub>STAB</sub> (3)(5)	ADC stabilization time	-	14		1/f <sub>ADC</sub>	
<b>4</b> (3)	Total conversion time	f <sub>ADC</sub> = 16 MHz, 12-bit resolution	0.875	-	10.81	μs
t <sub>ConV</sub> <sup>(3)</sup>	(including sampling time)	12-bit resolution	14 to 173 (t <sub>S</sub> for successive			1/f <sub>ADC</sub>

<sup>1.</sup>  $V_{DDA}$  minimum value can be decreased in specific temperature conditions. Refer to Table 65:  $R_{AIN}$  max for  $f_{ADC}$  = 16 MHz.

<sup>2.</sup> A current consumption proportional to the APB clock frequency has to be added (see *Table 40: Peripheral current consumption in Run or Sleep mode*).

<sup>3.</sup> Guaranteed by design.

Standard channels have an extra protection resistance which depends on supply voltage. Refer to Table 65: R<sub>AIN</sub> max for f<sub>ADC</sub> = 16 MHz.

<sup>5.</sup> This parameter only includes the ADC timing. It does not take into account register access latency.

<sup>6.</sup> This parameter specifies the latency to transfer the conversion result into the ADC\_DR register. EOC bit is set to indicate the conversion is complete and has the same latency.

## Equation 1: R<sub>AIN</sub> max formula

$$R_{AIN} < \frac{T_S}{f_{ADC} \times C_{ADC} \times ln(2^{N+2})} - R_{ADC}$$

The simplified formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).

Table 65.  $R_{AIN}$  max for  $f_{ADC} = 16 \text{ MHz}^{(1)}$ 

		B may for	$R_{AIN}$ max for standard channels (k $\Omega$ )						
T <sub>s</sub> (cycles)	t <sub>S</sub> (µs)	R <sub>AIN</sub> max for fast channels (kΩ)	V <sub>DD</sub> > 2.7 V	V <sub>DD</sub> > 2.4 V	V <sub>DD</sub> > 2.0 V	V <sub>DD</sub> > 1.8 V	V <sub>DD</sub> > 1.75 V	V <sub>DD</sub> > 1.65 V and T <sub>A</sub> > -10 °C	V <sub>DD</sub> > 1.65 V and T <sub>A</sub> > 25 °C
1.5	0.09	0.5	< 0.1	NA	NA	NA	NA	NA	NA
3.5	0.22	1	0.2	< 0.1	NA	NA	NA	NA	NA
7.5	0.47	2.5	1.7	1.5	< 0.1	NA	NA	NA	NA
12.5	0.78	4	3.2	3	1	NA	NA	NA	NA
19.5	1.22	6.5	5.7	5.5	3.5	NA	NA	NA	< 0.1
39.5	2.47	13	12.2	12	10	NA	NA	NA	5
79.5	4.97	27	26.2	26	24	< 0.1	NA	NA	19
160.5	10.03	50	49.2	49	47	32	< 0.1	< 0.1	42

<sup>1.</sup> Guaranteed by design.

Table 66. ADC accuracy<sup>(1)(2)(3)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ET	Total unadjusted error		-	2	4	
EO	Offset error		-	1	2.5	LSB
EG	Gain error	1.65 V < V <sub>DDA</sub> = V <sub>REF+</sub> < 3.6 V, range 1/2/3	-	1	2	
EL	Integral linearity error		-	1.5	2.5	
ED	Differential linearity error		-	1	1.5	
	Effective number of bits		10.2	11		bits
ENOB	Effective number of bits (16-bit mode oversampling with ratio =256) <sup>(4)</sup>		11.3	12.1	-	
SINAD	Signal-to-noise distortion		63	69	-	
	Signal-to-noise ratio		63	69	-	dB
SNR	Signal-to-noise ratio (16-bit mode oversampling with ratio =256) <sup>(4)</sup>		70	76	-	
THD	Total harmonic distortion		-	-85	-73	



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ET	Total unadjusted error	1.65 V < V <sub>REF+</sub> <v<sub>DDA &lt; 3.6 V, range 1/2/3</v<sub>	-	2	5	LSB
EO	Offset error		-	1	2.5	
EG	Gain error		-	1	2	
EL	Integral linearity error		-	1.5	3	
ED	Differential linearity error		-	1	2	
ENOB	Effective number of bits		10.0	11.0	-	bits
SINAD	Signal-to-noise distortion		62	69	-	
SNR	Signal-to-noise ratio		61	69	-	dB
THD	Total harmonic distortion		-	-85	-65	

Table 66. ADC accuracy<sup>(1)(2)(3)</sup> (continued)

- 1. ADC DC accuracy values are measured after internal calibration.
- ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current.
   Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.12 does not affect the ADC accuracy.
- Better performance may be achieved in restricted V<sub>DDA</sub>, frequency and temperature ranges.
- 4. This number is obtained by the test board without additional noise, resulting in non-optimized value for oversampling mode.

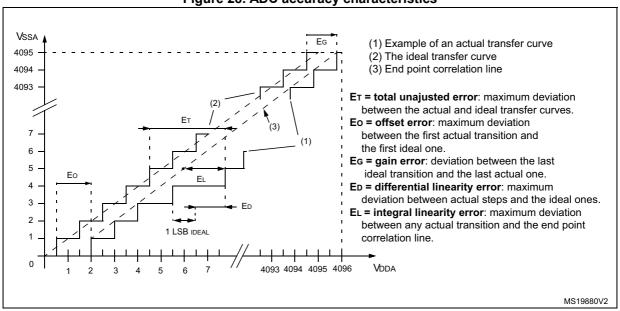


Figure 28. ADC accuracy characteristics

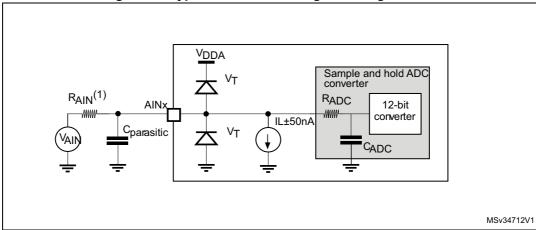


Figure 29. Typical connection diagram using the ADC

- 1. Refer to *Table 64: ADC characteristics* for the values of R<sub>AIN</sub>, R<sub>ADC</sub> and C<sub>ADC</sub>.
- C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.

### General PCB design guidelines

Power supply decoupling should be performed as shown in *Figure 30* or *Figure 31*, depending on whether  $V_{REF+}$  is connected to  $V_{DDA}$  or not. The 10 nF capacitors should be ceramic (good quality). They should be placed as close as possible to the chip.

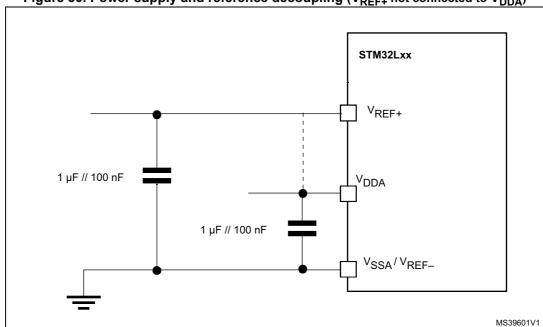


Figure 30. Power supply and reference decoupling (V<sub>REF+</sub> not connected to V<sub>DDA</sub>)

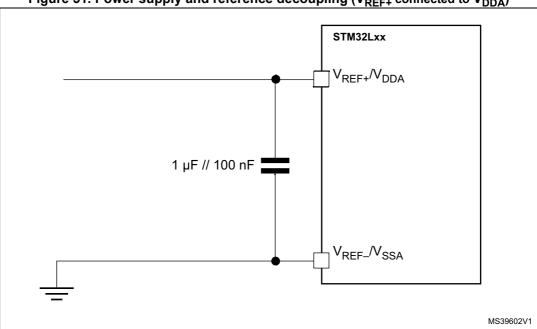


Figure 31. Power supply and reference decoupling ( $V_{REF+}$  connected to  $V_{DDA}$ )

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# 6.3.16 DAC electrical characteristics

Data guaranteed by design, not tested in production, unless otherwise specified.

**Table 67. DAC characteristics** 

Symbol	Parameter	Cond	litions	Min	Тур	Max	Unit	
$V_{DDA}$	Analog supply voltage	-		1.8	-	3.6	V	
V <sub>REF+</sub>	Reference supply voltage	V <sub>REF+</sub> must always be below V <sub>DDA</sub>		1.8	-	3.6	٧	
V <sub>REF-</sub>	Lower reference voltage	-			V <sub>SSA</sub>		V	
(1)	Current consumption on V <sub>REF+</sub> supply V <sub>REF+</sub> = 3.3 V	No load, mid (0x800)	ldle code	-	130	220		
I <sub>DDVREF+</sub> <sup>(1)</sup>		No load, worst code (0x000)		-	220	350	μA	
(2)	Current consumption on V <sub>DDA</sub> supply, V <sub>DDA</sub> = 3.3 V	No load, mid (0x800)	Idle code	-	210	320		
I <sub>DDA</sub> <sup>(2)</sup>		No load, worst code (0xF1C)		-	320	520	μΑ	
R <sub>1</sub> <sup>(3)</sup>	Resistive load	DAC output to to ON R	R <sub>L</sub> connected to V <sub>SSA</sub>	5	-	-	1.0	
KL <sup>(*)</sup>			R <sub>L</sub> connected to V <sub>DDA</sub>	25	-	-	kΩ	
C <sub>L</sub> <sup>(3)</sup>	Capacitive load	DAC output buffer ON		-	-	50	pF	
R <sub>O</sub>	Output impedance	DAC output buffer OFF		12	16	20	kΩ	
V <sub>DAC_OUT</sub>	Voltage on DAC_OUT output	DAC output	buffer ON	0.2	-	V <sub>DDA</sub> – 0.2	V	
		DAC output	buffer OFF	0.5	-	V <sub>REF+</sub> – 1LSB	mV	

Table 67. DAC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
DNL <sup>(2)</sup>	Differential non linearity <sup>(4)</sup>	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON	-	1.5	3		
		No R <sub>LOAD</sub> , C <sub>L</sub> ≤ 50 pF DAC output buffer OFF	-	1.5	3		
INL <sup>(2)</sup>	Integral non linearity <sup>(5)</sup>	$C_L \le 50 \text{ pF}, R_L \ge 5 \text{ k}\Omega$ DAC output buffer ON	-	2	4		
		No $R_{LOAD}$ , $C_{L} \le 50 pF$ DAC output buffer OFF	-	2	4	LSB	
Offset <sup>(2)</sup>	Offset error at code 0x800 <sup>(6)</sup>	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$ DAC output buffer ON	-	±10	±25		
Offset(=)		No R <sub>LOAD</sub> , C <sub>L</sub> ≤ 50 pF DAC output buffer OFF	-	±5	±8		
Offset1 <sup>(2)</sup>	Offset error at code 0x001 <sup>(7)</sup>	No $R_{LOAD}$ , $C_{L} \le 50 pF$ DAC output buffer OFF	-	±1.5	±5		
dOffset/dT <sup>(2)</sup>	Offset error temperature coefficient (code 0x800)	$V_{DDA} = 3.3V$ $V_{REF+} = 3.0 V$ $T_A = 0 \text{ to } 50 ^{\circ}\text{C}$ DAC output buffer OFF	-20	-10	0	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	
		$V_{DDA} = 3.3V$ $V_{REF+} = 3.0 V$ $T_A = 0 \text{ to } 50 ^{\circ}\text{C}$ DAC output buffer ON	0	20	50	μV/°C	
Gain <sup>(2)</sup>	Gain error <sup>(8)</sup>	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON	-	+0.1 / -0.2%	+0.2 / -0.5%	0/	
		No $R_{LOAD}$ , $C_{L} \le 50 pF$ DAC output buffer OFF	-	+0 / -0.2%	+0 / -0.4%	%	
dGain/dT <sup>(2)</sup>	Gain error temperature coefficient	$V_{DDA} = 3.3V$ $V_{REF+} = 3.0 V$ $T_A = 0 \text{ to } 50 ^{\circ}\text{C}$ DAC output buffer OFF	-10	-2	0	μV/°C	
		$V_{DDA} = 3.3V$ $V_{REF+} = 3.0 V$ $T_{A} = 0 \text{ to } 50 ^{\circ}\text{C}$ DAC output buffer ON	-40	-8	0	μ.ν. σ	
TUE <sup>(2)</sup>	Total unadjusted error	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$ DAC output buffer ON	-	12	30	- LSB	
		No $R_{LOAD}$ , $C_{L} \le 50 pF$ DAC output buffer OFF	-	8	12		

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>SETTLING</sub>	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes till DAC_OUT reaches final value ±1LSB	$C_L \le 50$ pF, $R_L \ge 5$ k $\Omega$	-	7	12	μs
Update rate	Max frequency for a correct DAC_OUT change (95% of final value) with 1 LSB variation in the input code	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$	-	-	1	Msps
t <sub>WAKEUP</sub>	Wakeup time from off state (setting the ENx bit in the DAC Control register) <sup>(9)</sup>	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$	-	9	15	μs
PSRR+	V <sub>DDA</sub> supply rejection ratio (static DC measurement)	$C_L \le 50 \text{ pF, } R_L \ge 5 \text{ k}\Omega$	-	-60	-35	dB

Table 67. DAC characteristics (continued)

- 1. Guaranteed by characterization results.
- 2. Guaranteed by design, not tested in production.
- 3. Connected between DAC\_OUT and V<sub>SSA</sub>.
- 4. Difference between two consecutive codes 1 LSB.
- 5. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
- 6. Difference between the value measured at Code (0x800) and the ideal value =  $V_{REF+}/2$ .
- 7. Difference between the value measured at Code (0x001) and the ideal value.
- 8. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFF when buffer is OFF, and from code giving 0.2 V and (V<sub>DDA</sub> 0.2) V when buffer is ON.
- 9. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).

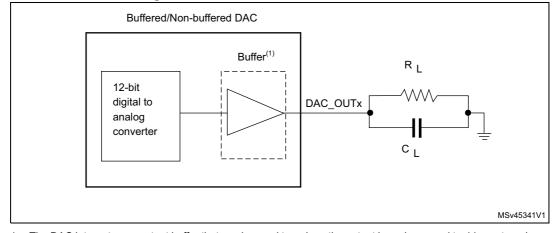


Figure 32. 12-bit buffered/non-buffered DAC

 The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC\_CR register. Electrical characteristics STM32L073xx

# 6.3.17 Temperature sensor characteristics

Table 68. Temperature sensor calibration values

Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, V <sub>DDA</sub> = 3 V	0x1FF8 007A - 0x1FF8 007B
TS_CAL2	TS ADC raw data acquired at temperature of 130 °C, V <sub>DDA</sub> = 3 V	0x1FF8 007E - 0x1FF8 007F

Table 69. Temperature sensor characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with temperature	-	±1	<u>+2</u>	°C
Avg_Slope <sup>(1)</sup>	Average slope	1.48	1.61	1.75	mV/°C
V <sub>130</sub>	Voltage at 130°C ±5°C <sup>(2)</sup>	640	670	700	mV
I <sub>DDA(TEMP)</sub> (3)	Current consumption	-	3.4	6	μA
t <sub>START</sub> (3)	Startup time	-	-	10	ше
T <sub>S_temp</sub> <sup>(4)(3)</sup>	ADC sampling time when reading the temperature	10	-	-	μs

- 1. Guaranteed by characterization results.
- 2. Measured at  $V_{DD}$  = 3 V ±10 mV. V130 ADC conversion result is stored in the TS\_CAL2 byte.
- 3. Guaranteed by design.
- 4. Shortest sampling time can be determined in the application by multiple iterations.

### 6.3.18 Comparators

Table 70. Comparator 1 characteristics

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit
$V_{DDA}$	Analog supply voltage	-	1.65		3.6	V
R <sub>400K</sub>	R <sub>400K</sub> value	-	-	400	-	kΩ
R <sub>10K</sub>	R <sub>10K</sub> value	-	-	10	-	K22
V <sub>IN</sub>	Comparator 1 input voltage range	-	0.6	-	$V_{DDA}$	V
t <sub>START</sub>	Comparator startup time	-	-	7	10	II.C
td	Propagation delay <sup>(2)</sup>	-	-	3	10	μs
Voffset	Comparator offset	-	-	±3	±10	mV
d <sub>Voffset</sub> /dt	Comparator offset variation in worst voltage stress conditions	$V_{DDA} = 3.6 \text{ V}, V_{IN+} = 0 \text{ V}, V_{IN-} = V_{REFINT}, T_A = 25 ^{\circ}\text{C}$	0	1.5	10	mV/1000 h
I <sub>COMP1</sub>	Current consumption <sup>(3)</sup>	-	-	160	260	nA

<sup>1.</sup> Guaranteed by characterization.

<sup>2.</sup> The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

<sup>3.</sup> Comparator consumption only. Internal reference voltage not included.

Table 71. Comparator 2 characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
$V_{DDA}$	Analog supply voltage	-	1.65	-	3.6	V
V <sub>IN</sub>	Comparator 2 input voltage range	-	0	-	$V_{DDA}$	V
+.	Comparator startup time	Fast mode	-	15	20	
t <sub>START</sub>	Comparator startup time	Slow mode	-	20	25	
+	Propagation delay <sup>(2)</sup> in slow mode	1.65 V ≤V <sub>DDA</sub> ≤2.7 V	-	1.8	3.5	116
t <sub>d slow</sub>	Propagation delay Vill slow mode	2.7 V ≤V <sub>DDA</sub> ≤3.6 V	-	2.5	6	μs
4	Propagation delay <sup>(2)</sup> in fast mode	1.65 V ≤V <sub>DDA</sub> ≤2.7 V	-	0.8	2	
t <sub>d fast</sub>	Propagation delay \(\tag{\text{in last mode}}\)	2.7 V ≤V <sub>DDA</sub> ≤3.6 V	-	1.2	4	
V <sub>offset</sub>	Comparator offset error		-	<u>±4</u>	±20	mV
dThreshold/ dt	Threshold voltage temperature coefficient	$V_{DDA} = 3.3V, T_A = 0 \text{ to } 50 \text{ °C},$ $V = V_{REFINT},$ $3/4 V_{REFINT},$ $1/2 V_{REFINT},$ $1/4 V_{REFINT}.$	-	15	30	ppm /°C
1	Current consumption(3)	Fast mode	-	3.5	5	
I <sub>COMP2</sub>	Current consumption <sup>(3)</sup>	Slow mode	-	0.5	2	μA

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

<sup>3.</sup> Comparator consumption only. Internal reference voltage (required for comparator operation) is not included.

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#### 6.3.19 Timer characteristics

#### **TIM timer characteristics**

The parameters given in the *Table 72* are guaranteed by design.

Refer to Section 6.3.13: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Symbol	Parameter Conditions Mi		Min	Max	Unit
t	Timer resolution time		1	-	t <sub>TIMxCLK</sub>
<sup>t</sup> res(TIM)	Time resolution time	f <sub>TIMxCLK</sub> = 32 MHz	31.25	-	ns
f	Timer external clock frequency on CH1		0	f <sub>TIMxCLK</sub> /2	MHz
$f_{EXT}$	to CH4	f <sub>TIMxCLK</sub> = 32 MHz	0	16	MHz
Res <sub>TIM</sub>	Timer resolution	-		16	bit
	16-bit counter clock period when	-	1	65536	t <sub>TIMxCLK</sub>
t <sub>COUNTER</sub>	internal clock is selected (timer's prescaler disabled)	f <sub>TIMxCLK</sub> = 32 MHz	0.0312	2048	μs
t <sub>MAX_COUNT</sub>	Maximum possible count	-	-	65536 × 65536	t <sub>TIMxCLK</sub>
	Iviaximum possible count	f <sub>TIMxCLK</sub> = 32 MHz	ı	134.2	s

Table 72. TIMx characteristics<sup>(1)</sup>

### 6.3.20 Communications interfaces

### I<sup>2</sup>C interface characteristics

The I<sup>2</sup>C interface meets the timings requirements of the I<sup>2</sup>C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The  $I^2C$  timing requirements are guaranteed by design when the  $I^2C$  peripheral is properly configured (refer to the reference manual for details). The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and VDDIOx is disabled, but is still present. Only FTf I/O pins support Fm+ low level output current maximum requirement (refer to Section 6.3.13: I/O port characteristics for the I2C I/Os characteristics).

All I<sup>2</sup>C SDA and SCL I/Os embed an analog filter (see *Table 73* for the analog filter characteristics).

<sup>1.</sup> TIMx is used as a general term to refer to the TIM2, TIM6, TIM21, and TIM22 timers.

The analog spike filter is compliant with I<sup>2</sup>C timings requirements only for the following voltage ranges:

- Fast mode Plus: 2.7 V ≤V<sub>DD</sub> ≤3.6 V and voltage scaling Range 1
- Fast mode:
  - 2 V ≤V<sub>DD</sub> ≤3.6 V and voltage scaling Range 1 or Range 2.
  - V<sub>DD</sub> < 2 V, voltage scaling Range 1 or Range 2, C<sub>load</sub> < 200 pF.

In other ranges, the analog filter should be disabled. The digital filter can be used instead.

Note: In Standard mode, no spike filter is required.

Table 73. I2C analog filter characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
		Range 1		100 <sup>(3)</sup>	
t <sub>AF</sub>	Maximum pulse width of spikes that are suppressed by the analog filter	Range 2	50 <sup>(2)</sup>	-	ns
	3	Range 3		-	

- 1. Guaranteed by characterization results.
- 2. Spikes with widths below  $t_{AF(min)}$  are filtered.
- 3. Spikes with widths above  $t_{AF(max)}$  are not filtered

#### **SPI** characteristics

Unless otherwise specified, the parameters given in the following tables are derived from tests performed under ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 26*.

Refer to Section 6.3.12: I/O current injection characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 74. SPI characteristics in voltage Range 1 (1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode			16	
		Slave mode receiver	-	-	16	
f <sub>SCK</sub> 1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode Transmitter 1.71 <v<sub>DD&lt;3.6V</v<sub>	-	-	12 <sup>(2)</sup>	MHz
		Slave mode Transmitter 2.7 <v<sub>DD&lt;3.6V</v<sub>	-	-	16 <sup>(2)</sup>	
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%



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Table 74. SPI characteristics in voltage Range 1 <sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4*Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2*Tpclk	-	-	
$t_{w(SCKH)} \ t_{w(SCKL)}$	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+	
t <sub>su(MI)</sub>	Data input setup time	Master mode	0	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	3	-	-	
t <sub>h(MI)</sub>	- Data input hold time	Master mode	7	-	-	
t <sub>h(SI)</sub>	Data input noid time	Slave mode	3.5	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	15	-	36	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	10	-	30	
<b>+</b>		Slave mode 1.65 V <v<sub>DD&lt;3.6 V</v<sub>	-	18	41	
t <sub>v(SO)</sub>	Data output valid time	Slave mode 2.7 V <v<sub>DD&lt;3.6 V</v<sub>	-	18	25	
t <sub>v(MO)</sub>		Master mode	-	4	7	
t <sub>h(SO)</sub>	Data output hold time	Slave mode	10	-	-	
t <sub>h(MO)</sub>	Data output Hold tillle	Master mode	0	-	-	

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> The maximum SPI clock frequency in slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while  $\text{Duty}_{(SCK)} = 50\%$ .

Table 75. SPI characteristics in voltage Range 2 (1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode			8	
f <sub>SCK</sub> 1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode Transmitter 1.65 <v<sub>DD&lt;3.6V</v<sub>	_	-	8	MHz
" C(SCK)		Slave mode Transmitter 2.7 <v<sub>DD&lt;3.6V</v<sub>			8 <sup>(2)</sup>	
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4*Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2*Tpclk	-	-	
t <sub>w(SCKH)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2	
t <sub>su(MI)</sub>	Data input setup time	Master mode	0	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	3	-	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	11	-	-	
t <sub>h(SI)</sub>	Data input noid time	Slave mode	4.5	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	18	-	52	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	12	-	42	
t <sub>v(SO)</sub>	Data output valid time	Slave mode	-	20	56.5	
t <sub>v(MO)</sub>	2010 001put 10110 11110	Master mode	-	5	9	
t <sub>h(SO)</sub>	Data output hold time	Slave mode	13	-	-	
t <sub>h(MO)</sub>	Data output noid time	Master mode	3	-	-	

<sup>1.</sup> Guaranteed by characterization results.

<sup>2.</sup> The maximum SPI clock frequency in slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty<sub>(SCK)</sub> = 50%.

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Table 76. SPI characteristics in voltage Range 3 (1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>SCK</sub>	CDI plant fraguancy	Master mode			2	MHz
1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode	-	-	2 <sup>(2)</sup>	IVITZ
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4*Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2*Tpclk	-	-	
t <sub>w(SCKH)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2	
t <sub>su(MI)</sub>	Data input setup time	Master mode	1.5	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	6	-	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	13.5	-	-	
t <sub>h(SI)</sub>	Data input hold time	Slave mode	16	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	30	-	70	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	40	-	80	
t <sub>v(SO)</sub>	Data output valid time	Slave mode	-	30	70	
t <sub>v(MO)</sub>		Master mode	-	7	9	
t <sub>h(SO)</sub>	Data output hold time	Slave mode	25	-	-	
t <sub>h(MO)</sub>	Data output noid time	Master mode	8	-	-	

<sup>1.</sup> Guaranteed by characterization results.

The maximum SPI clock frequency in slave transmitter mode is determined by the sum of t<sub>v(SO)</sub> and t<sub>su(MI)</sub> which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having t<sub>su(MI)</sub> = 0 while Duty<sub>(SCK)</sub> = 50%.

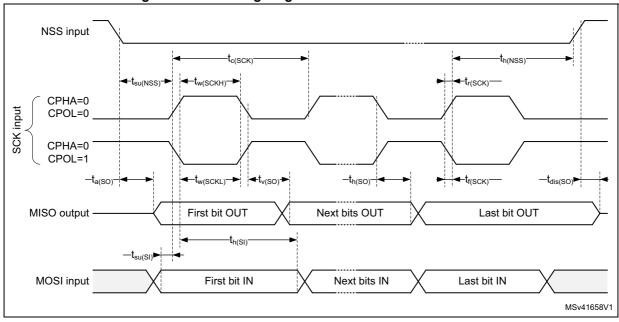
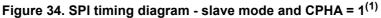
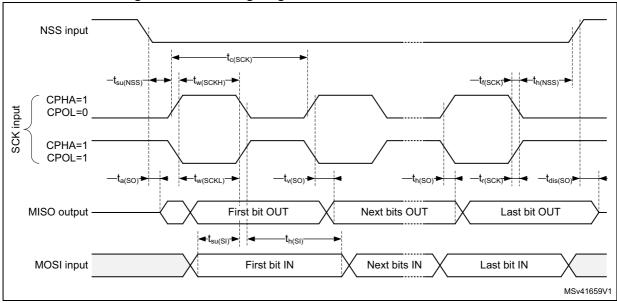


Figure 33. SPI timing diagram - slave mode and CPHA = 0





1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

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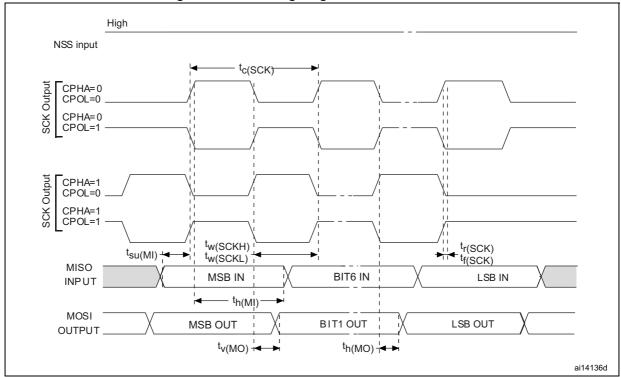


Figure 35. SPI timing diagram - master mode<sup>(1)</sup>

1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .



### **I2S** characteristics

Table 77. I2S characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>MCK</sub>	I2S Main clock output	-	256 x 8K	256xFs <sup>(2)</sup>	MHz
f	129 alook froguenay	Master data: 32 bits	-	64xFs	MHz
f <sub>CK</sub>	I2S clock frequency Slave data: 32 bits	Slave data: 32 bits	-	64xFs	IVITIZ
D <sub>CK</sub>	I2S clock frequency duty cycle	Slave receiver	30	70	%
t <sub>v(WS)</sub>	WS valid time	Master mode	-	15	
t <sub>h(WS)</sub>	WS hold time	Master mode	11	-	
t <sub>su(WS)</sub>	WS setup time	Slave mode	6	-	
t <sub>h(WS)</sub>	WS hold time	Slave mode	2	-	
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	0	-	
t <sub>su(SD_SR)</sub>	Data input setup time	Slave receiver	6.5	-	ns
t <sub>h(SD_MR)</sub>	Data input hold time	Master receiver	18	-	115
t <sub>h(SD_SR)</sub>	Data input noid time	Slave receiver	15.5	-	
t <sub>v(SD_ST)</sub>	Data output valid time	Slave transmitter (after enable edge)	-	77	
t <sub>v(SD_MT)</sub>	Data output valid time	Master transmitter (after enable edge)	-	8	
t <sub>h(SD_ST)</sub>	Data output hold time	Slave transmitter (after enable edge)	18	-	
t <sub>h(SD_MT)</sub>	Data output noid time	Master transmitter (after enable edge)	1.5	-	

<sup>1.</sup> Guaranteed by characterization results.

### Note:

Refer to the I2S section of the product reference manual for more details about the sampling frequency (Fs),  $f_{MCK}$ ,  $f_{CK}$  and  $D_{CK}$  values. These values reflect only the digital peripheral behavior, source clock precision might slightly change them. DCK depends mainly on the ODD bit value, digital contribution leads to a min of (I2SDIV/(2\*I2SDIV+ODD) and a max of (I2SDIV+ODD)/(2\*I2SDIV+ODD). Fs max is supported for each mode/condition.

<sup>2. 256</sup>xFs maximum value is equal to the maximum clock frequency.

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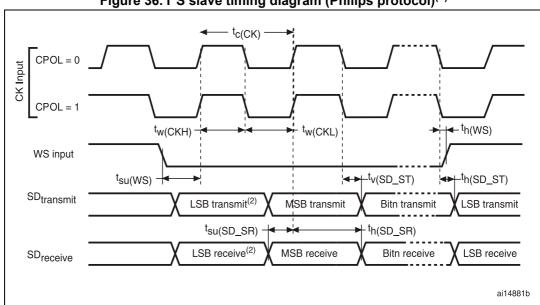


Figure 36. I<sup>2</sup>S slave timing diagram (Philips protocol)<sup>(1)</sup>

- Measurement points are done at CMOS levels:  $0.3 \times V_{DD}$  and  $0.7 \times V_{DD}$ .
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first

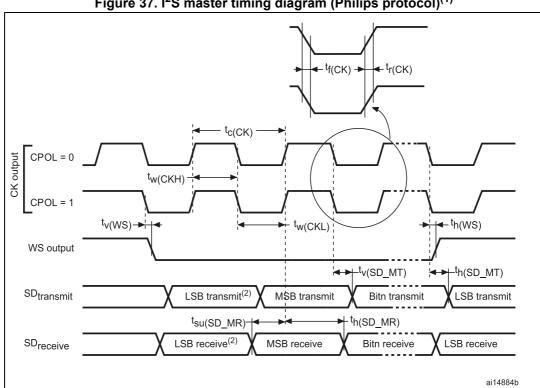


Figure 37. I<sup>2</sup>S master timing diagram (Philips protocol)<sup>(1)</sup>

- 1. Guaranteed by characterization results.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first

### **USB** characteristics

The USB interface is USB-IF certified (full speed).

Table 78. USB startup time

Symbol	Parameter	Max	Unit
t <sub>STARTUP</sub> <sup>(1)</sup>	USB transceiver startup time	1	μs

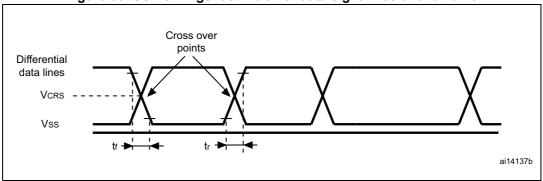
<sup>1.</sup> Guaranteed by design.

Table 79. USB DC electrical characteristics

Symbol	Parameter	Conditions	Min. <sup>(1)</sup>	Max. <sup>(1)</sup>	Unit					
Input levels										
V <sub>DD</sub>	USB operating voltage	-	3.0	3.6	V					
V <sub>DI</sub> <sup>(2)</sup>	Differential input sensitivity	I(USB_DP, USB_DM)	0.2	-						
V <sub>CM</sub> <sup>(2)</sup>	Differential common mode range	Includes V <sub>DI</sub> range	0.8	2.5	V					
V <sub>SE</sub> <sup>(2)</sup>	Single ended receiver threshold	-	1.3	2.0						
Output lev	Output levels									
V <sub>OL</sub> <sup>(3)</sup>	Static output level low	$R_L$ of 1.5 k $\Omega$ to 3.6 $V^{(4)}$	-	0.3	V					
V <sub>OH</sub> <sup>(3)</sup>	Static output level high	$R_L$ of 15 k $\Omega$ to $V_{SS}^{(4)}$	2.8	3.6	] <b>'</b>					

- 1. All the voltages are measured from the local ground potential.
- 2. Guaranteed by characterization results.
- 3. Guaranteed by test in production.
- 4.  $R_L$  is the load connected on the USB drivers.

Figure 38. USB timings: definition of data signal rise and fall time



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	Driver characteristics <sup>(1)</sup>										
Symbol Parameter Conditions Min Max Ur											
t <sub>r</sub>	Rise time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	20	ns						
t <sub>f</sub>	Fall Time <sup>(2)</sup>	C <sub>L</sub> = 50 pF	4	20	ns						
t <sub>rfm</sub>	Rise/ fall time matching	t <sub>r</sub> /t <sub>f</sub>	90	110	%						
V <sub>CRS</sub>	Output signal crossover voltage		1.3	2.0	V						

Table 80. USB: full speed electrical characteristics

### 6.3.21 LCD controller

The devices embed a built-in step-up converter to provide a constant LCD reference voltage independently from the  $V_{DD}$  voltage. An external capacitor  $C_{ext}$  must be connected to the  $V_{LCD}$  pin to decouple this converter.

**Table 81. LCD controller characteristics** 

Symbol	Parameter	Min	Тур	Max	Unit
$V_{LCD}$	LCD external voltage	-	-	3.6	
V <sub>LCD0</sub>	LCD internal reference voltage 0	-	2.6	-	1
V <sub>LCD1</sub>	LCD internal reference voltage 1	-	2.73	-	
V <sub>LCD2</sub>	LCD internal reference voltage 2	-	2.86	-	1
V <sub>LCD3</sub>	LCD internal reference voltage 3	-	2.98	-	٧
V <sub>LCD4</sub>	LCD internal reference voltage 4	-	3.12	-	
V <sub>LCD5</sub>	LCD internal reference voltage 5	-	3.26	-	1
V <sub>LCD6</sub>	LCD internal reference voltage 6	-	3.4	-	1
V <sub>LCD7</sub>	LCD internal reference voltage 7	-	3.55	-	
C <sub>ext</sub>	V <sub>LCD</sub> external capacitance	0.1	-	2	μF
I <sub>LCD</sub> <sup>(1)</sup>	Supply current at V <sub>DD</sub> = 2.2 V	-	3.3	-	
LCD, ,	Supply current at V <sub>DD</sub> = 3.0 V	-	3.1	-	μA
R <sub>Htot</sub> <sup>(2)</sup>	Low drive resistive network overall value	5.28	6.6	7.92	ΜΩ
R <sub>L</sub> <sup>(2)</sup>	High drive resistive network total value	192	240	288	kΩ
V <sub>44</sub>	Segment/Common highest level voltage	-	-	$V_{LCD}$	V



<sup>1.</sup> Guaranteed by design.

<sup>2.</sup> Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).

Table 81. LCD controller characteristics (continued)

Symbol	Parameter	Min	Тур	Max	Unit
V <sub>34</sub>	Segment/Common 3/4 level voltage	-	3/4 V <sub>LCD</sub>	-	
V <sub>23</sub>	Segment/Common 2/3 level voltage	-	2/3 V <sub>LCD</sub>	-	
V <sub>12</sub>	Segment/Common 1/2 level voltage	-	1/2 V <sub>LCD</sub>	-	V
V <sub>13</sub>	Segment/Common 1/3 level voltage	-	1/3 V <sub>LCD</sub>	-	V
V <sub>14</sub>	Segment/Common 1/4 level voltage	-	1/4 V <sub>LCD</sub>	-	
V <sub>0</sub>	Segment/Common lowest level voltage	0	-	-	,
ΔVxx <sup>(3)</sup>	Segment/Common level voltage error T <sub>A</sub> = -40 to 85 ° C	-	-	± 50	mV

LCD enabled with 3 V internal step-up active, 1/8 duty, 1/4 bias, division ratio= 64, all pixels active, no LCD connected.

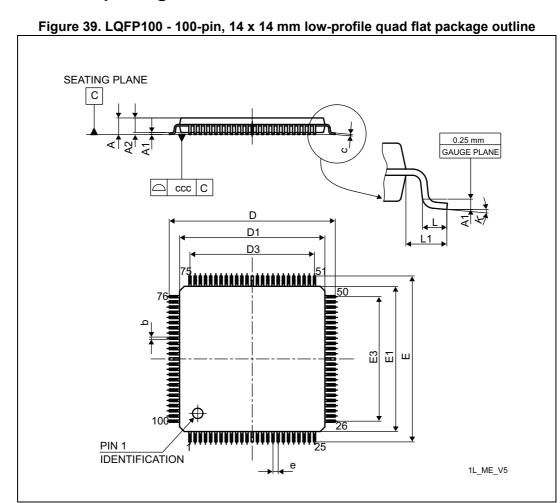
<sup>2.</sup> Guaranteed by design.

<sup>3.</sup> Guaranteed by characterization results.

# 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status *are available at www.st.com.* ECOPACK is an ST trademark.

# 7.1 LQFP100 package information



1. Drawing is not to scale. Dimensions are in millimeters.

Table 82. LQPF100 - 100-pin, 14 x 14 mm low-profile quad flat package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059



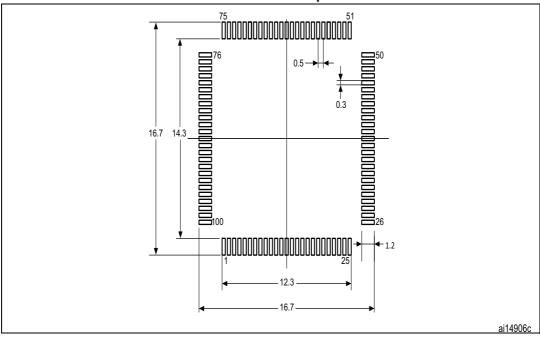
STM32L073xx Package information

Table 82. LQPF100 - 100-pin, 14 x 14 mm low-profile quad flat package mechanical data (continued)

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	-	12.000	-	-	0.4724	-
Е	15.800	16.000	16.200	0.6220	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591
E3	-	12.000	-	-	0.4724	-
е	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°
ccc	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 40. LQFP100 - 100-pin, 14 x 14 mm low-profile quad flat recommended footprint



1. Dimensions are expressed in millimeters.

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### **Device marking for LQFP100**

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

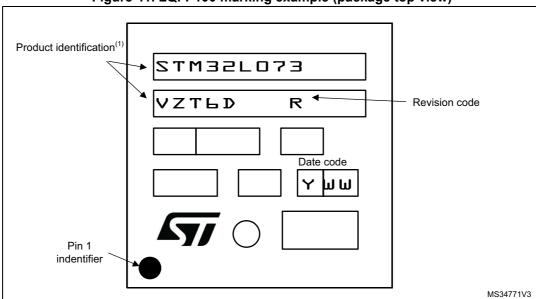


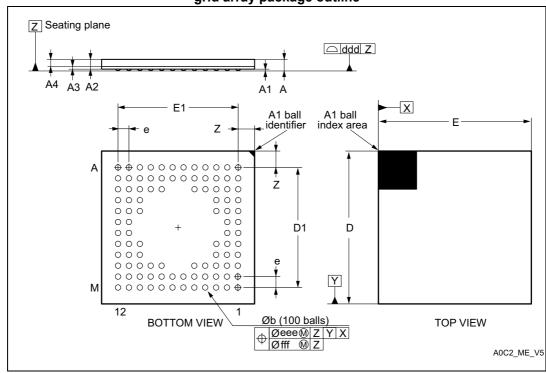
Figure 41. LQFP100 marking example (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



# 7.2 UFBGA100 package information

Figure 42. UFBGA100 - 100-pin, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package outline



1. Drawing is not to scale.

Table 83. UFBGA100 - 100-pin, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
Α	-	-	0.600	-	-	0.0236
A1	-	-	0.110	-	-	0.0043
A2	-	0.450	-	-	0.0177	-
A3	-	0.130	-	-	0.0051	0.0094
A4	-	0.320	-	-	0.0126	-
b	0.240	0.290	0.340	0.0094	0.0114	0.0134
D	6.850	7.000	7.150	0.2697	0.2756	0.2815
D1	-	5.500	-	-	0.2165	-
Е	6.850	7.000	7.150	0.2697	0.2756	0.2815
E1	-	5.500	-	-	0.2165	-
е	-	0.500	-	-	0.0197	-
Z	-	0.750	-	-	0.0295	-

Table 83. UFBGA100 - 100-pin, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data (continued)

Symbol	millimeters			inches <sup>(1)</sup>		
	Min.	Тур.	Max.	Min.	Тур.	Max.
ddd	-	-	0.080	-	-	0.0031
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 43. UFBGA100 - 100-pin, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package recommended footprint

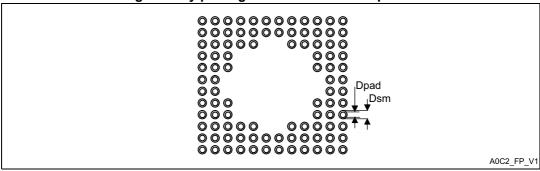


Table 84. UFBGA100 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5
Dpad	0.280 mm
Dsm	0.370 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.280 mm
Stencil thickness	Between 0.100 mm and 0.125 mm

### **Device marking for UFBGA100**

The following figure gives an example of topside marking versus ball A 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

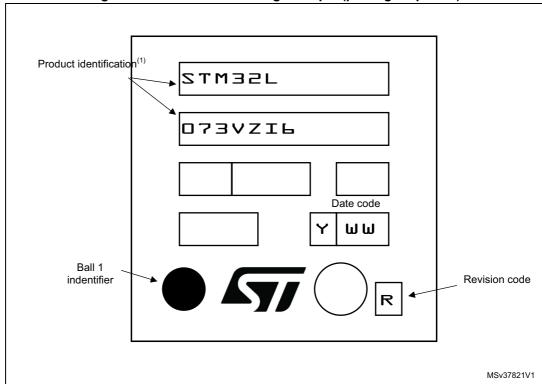


Figure 44. UFBGA100 marking example (package top view)

<sup>1.</sup> Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

# 7.3 LQFP64 package information

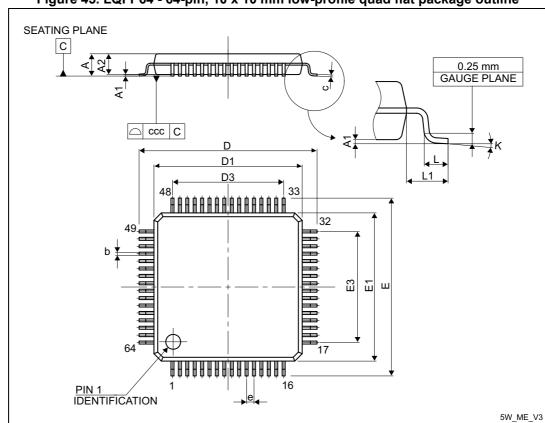


Figure 45. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 85. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data

Symbol		millimeters			inches <sup>(1)</sup>	ches <sup>(1)</sup>	
Зупівої	Min	Тур	Max	Min	Тур	Max	
Α	-	-	1.600	-	-	0.0630	
A1	0.050	-	0.150	0.0020	-	0.0059	
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571	
b	0.170	0.220	0.270	0.0067	0.0087	0.0106	
С	0.090	-	0.200	0.0035	-	0.0079	
D	-	12.000	-	-	0.4724	-	
D1	-	10.000	-	-	0.3937	-	
D3	-	7.500	-	-	0.2953	-	
E	-	12.000	-	-	0.4724	-	
E1	-	10.000	-	-	0.3937	-	

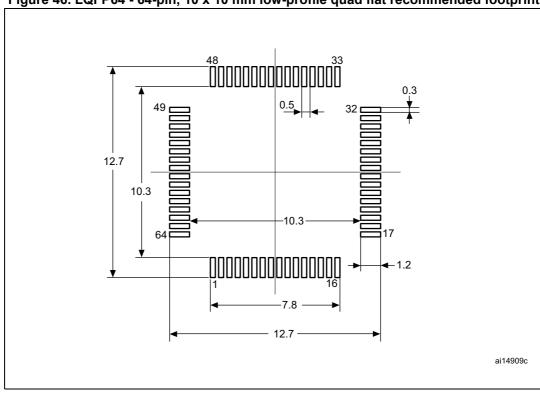


Table 85. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data (continued)

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
E3	-	7.500	-	-	0.2953	-
е	-	0.500	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 46. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat recommended footprint



1. Dimensions are expressed in millimeters.

### **Device marking for LQFP64**

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

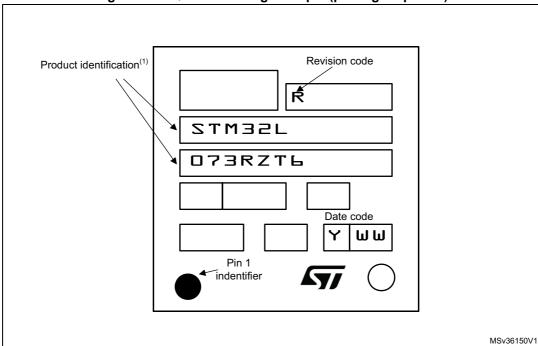


Figure 47. LQFP64 marking example (package top view)

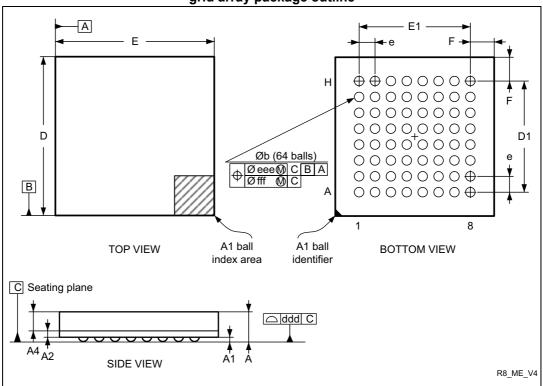


<sup>1.</sup> Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

STM32L073xx Package information

# 7.4 TFBGA64 package information

Figure 48. TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline



1. Drawing is not to scale.

Table 86. TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline

Symbol	millimeters			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.200	-	-	0.0472
A1	0.150	-	-	0.0059	-	-
A2	-	0.200	-	-	0.0079	-
A4	-	-	0.600	-	-	0.0236
b	0.250	0.300	0.350	0.0098	0.0118	0.0138
D	4.850	5.000	5.150	0.1909	0.1969	0.2028
D1	-	3.500	-	-	0.1378	-
E	4.850	5.000	5.150	0.1909	0.1969	0.2028
E1	-	3.500	-	-	0.1378	-
е	-	0.500	-	-	0.0197	-
F	-	0.750	-	-	0.0295	-

Table 86. TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline (continued)

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
ddd	-	-	0.080	-	-	0.0031
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 49. TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch, thin profile fine pitch ball ,grid array recommended footprint

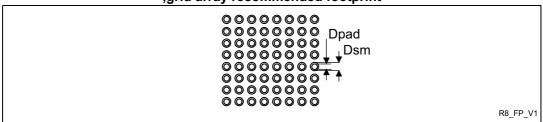


Table 87. TFBGA64 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5
Dpad	0.280 mm
Dsm	0.370 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.280 mm
Stencil thickness	Between 0.100 mm and 1.125 mm
Pad trace width	0.100 mm

### **Device marking for TFBGA64**

The following figure gives an example of topside marking versus ball A 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Product identification<sup>(1)</sup>

LO73RBHL

Date code = Year + week
Y WW

Revision code

MSv37820V1

Figure 50. TFBGA64 marking example (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

# 7.5 LQFP48 package information

Figure 51. LQFP48 - 48-pin, 7 x 7 mm low-profile quad flat package outline

1. Drawing is not to scale.

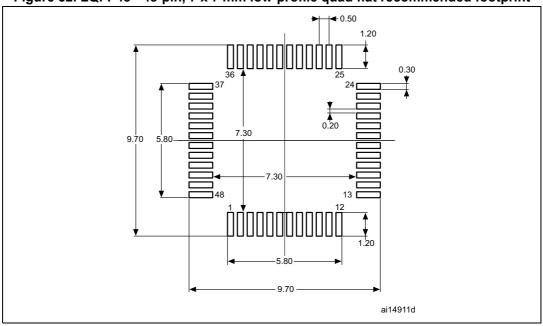


Table 88. LQFP48 - 48-pin, 7 x 7 mm low-profile quad flat package mechanical data

Symbol		millimeters	s inches <sup>(1</sup>			
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.500	-	-	0.2165	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.500	-	-	0.2165	-
е	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080			0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 52. LQFP48 - 48-pin, 7 x 7 mm low-profile quad flat recommended footprint



1. Dimensions are expressed in millimeters.

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### **Device marking for LQFP48**

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Product identification<sup>(1)</sup>

STM32L

Date code

Y WW

Revision code

MSv37819V1

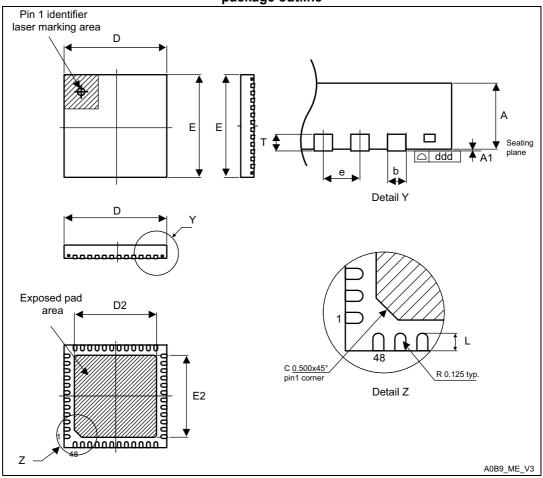
Figure 53. LQFP48 marking example (package top view)

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



# 7.6 UFQFPN48 package information

Figure 54. UFQFPN48 - 48 leads, 7x7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package outline



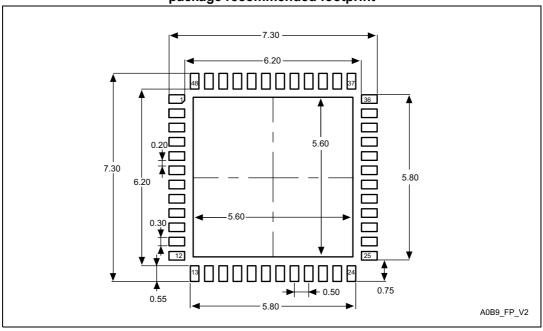
- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- 3. There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this back-side pad to PCB ground.

Table 89. UFQFPN48 - 48 leads, 7x7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package mechanical data

		millimeters				
Symbol	Min	Тур	Max	Min	Тур	Max
A	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
E	6.900	7.000	7.100	0.2717	0.2756	0.2795
D2	5.500	5.600	5.700	0.2165	0.2205	0.2244
E2	5.500	5.600	5.700	0.2165	0.2205	0.2244
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
Т	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 55. UFQFPN48 - 48 leads, 7x7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package recommended footprint



<sup>1.</sup> Dimensions are expressed in millimeters.

### **Device marking for UFQFPN48**

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Product identification<sup>(1)</sup>

STMBPLO73

CZULD

Date code

Y WW

Revision code

MSv62438V1

Figure 56. UFQFPN48 marking example (package top view)

<sup>1.</sup> Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

## 7.7 Thermal characteristics

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

 $T_J \max = T_A \max + (P_D \max \times \Theta_{JA})$ 

#### Where:

- T<sub>A</sub> max is the maximum ambient temperature in °C,
- Θ<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D$  max is the sum of  $P_{INT}$  max and  $P_{I/O}$  max ( $P_D$  max =  $P_{INT}$  max +  $P_{I/O}$ max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

P<sub>I/O</sub> max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \Sigma (V_{OL} \times I_{OL}) + \Sigma ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP48 - 7 x 7 mm / 0.5 mm pitch	54	
	Thermal resistance junction-ambient UFQFPN48 - 7 x 7 mm / 0.5 mm pitch	28	
$\Theta_{\sf JA}$	Thermal resistance junction-ambient LQFP64 - 10 x 10 mm / 0.5 mm pitch	46	oc/w
JA	Thermal resistance junction-ambient TFBGA64 - 5 x 5 mm / 0.5 mm pitch	64	0,,,,
1	Thermal resistance junction-ambient LQFP100 - 14 x 14 mm / 0.5 mm pitch	41	
	Thermal resistance junction-ambient UFBGA100 - 7 x 7 mm / 0.5 mm pitch	57	

Table 90. Thermal characteristics

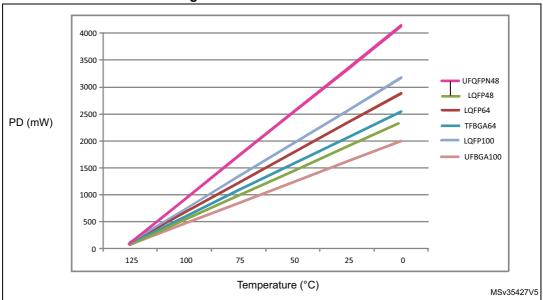


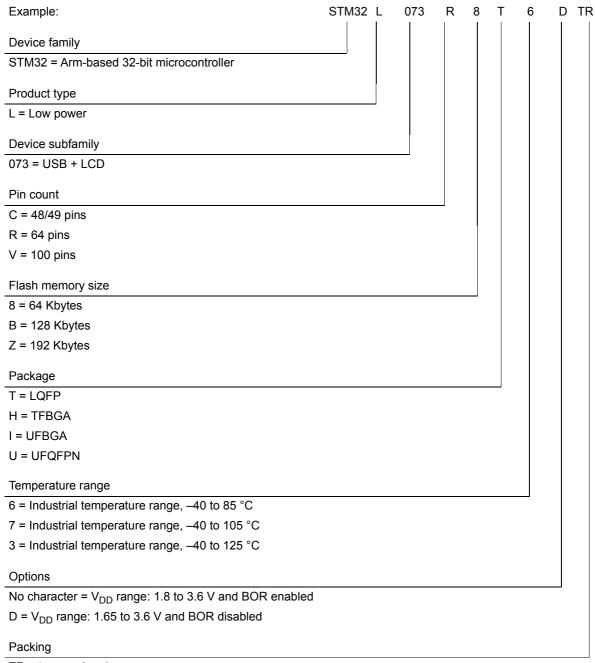
Figure 57. Thermal resistance

### 7.7.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.

Ordering information STM32L073xx

# 8 Ordering information



TR = tape and reel

No character = tray or tube

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

STM32L073xx Revision history

# 9 Revision history

**Table 91. Document revision history** 

Date	Revision	Changes
03-Aug-2015	1	Initial release
26-Oct-2015	2	Changed confidentiality level to public. Updated datasheet status to "production data". Modified ultra-low-power platform features on cover page. Changed number of GPIOs for LQFP48 for 37 in Table 2: Ultra-low-power STM32L073xxx device features and peripheral counts. Changed LCD_VLCD1 into LCD_VLCD2 in Section 3.13.2: $V_{LCD}$ voltage monitoring. In Section 6: Electrical characteristics, updated notes related to values guaranteed by characteristicn. Updated $ \Delta V_{SS} $ definition to include $V_{REF-}$ in Table 23: Voltage characteristics. Added $\Sigma V_{DD\_USB}$ and updated $\Sigma I_{IO(PIN)}$ in Figure 24: Current characteristics. Updated Table 56: EMI characteristics. Updated $T_{RIG}$ and $T_{R$

Revision history STM32L073xx

Table 91. Document revision history (continued)

Date	Revision	Changes
22-Mar-2016	3	Updated number of SPIs on cover page and in <i>Table 2: Ultra-low-power STM32L073xxx device features and peripheral counts</i> .  Changed minimum comparator supply voltage to 1.65 V on cover page. Added minimum DAC supply voltage on cover page.  Added number of fast and standard channels in <i>Section 3.12: Analog-to-digital converter (ADC)</i> .  Updated <i>Section 3.18.2: Universal synchronous/asynchronous receiver transmitter (USART)</i> and <i>Section 3.18.4: Serial peripheral interface (SPI)/Inter-integrated sound (I2S)</i> to mention the fact that USARTs with synchronous mode feature can be used as SPI master interfaces.  Added baudrate allowing to wake up the MCU from Stop mode in <i>Section 3.18.2: Universal synchronous/asynchronous receiver transmitter (USART)</i> and <i>Section 3.18.3: Low-power universal asynchronous receiver transmitter (USART)</i> .  Section 6.3.15: 12-bit ADC characteristics:  — <i>Table 64: ADC characteristics:</i> Distinction made between V <sub>DDA</sub> for fast and standard channels; added note 1.  Added note 4. related to R <sub>ADC</sub> .  Updated f <sub>TRIG</sub> Updated f <sub>TRIG</sub> Updated <i>Table 65: R<sub>AIN</sub> max for f<sub>ADC</sub> = 16 MHz</i> for f <sub>ADC</sub> = 16 MHz and distinction made between fast and standard channels.  Updated R <sub>O</sub> and added Note 2. in <i>Table 67: DAC characteristics</i> .  Added <i>Table 71: USART/LPUART characteristics</i> .



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STM32L073xx Revision history

Table 91. Document revision history (continued)

Date	Revision	Changes
		Memories and I/Os moved after Core in <i>Features</i> .  Removed column "I/O operation" from <i>Table 3: Functionalities depending on the operating power supply range</i> and added note related to GPIO speed.  Updated V <sub>DD_USB</sub> in <i>Section 3.4.1: Power supply schemes</i> .  In <i>Section 4: Pin descriptions</i> , renamed USB_OE into USB_NOE.  In <i>Section 5: Memory mapping</i> , replaced memory mapping schematic by reference to the reference manual.  Added mission profile compliance with JEDEC JESD47 in <i>Section 6.2: Absolute maximum ratings</i> .
		Updated minimum and maximum values of I/O weak pull-up equivalent resistor (R <sub>PU</sub> ) and weak pull-down equivalent resistor
		(R <sub>PD</sub> ) in <i>Table 60: I/O static characteristics</i> .
12-Sep-2017	4	Updated minimum and maximum values of NRST weak pull-up equivalent resistor (R <sub>PU</sub> ) in <i>Table 63: NRST pin characteristics</i> . Added note 2. related to the position of the external capacitor below <i>Figure 27: Recommended NRST pin protection</i> .
		Updated R <sub>AIN</sub> in <i>Table 64: ADC characteristics</i> .
		Updated Figure 32: 12-bit buffered/non-buffered DAC and added note below figure.
		Updated t <sub>AF</sub> maximum value for range 1 in <i>Table 73: I2C analog filter characteristics</i> .
		Removed Table 90: USART/LPUART characteristics.
		NSS timing waveforms updated in Figure 33: SPI timing diagram - slave mode and CPHA = 0 and Figure 34: SPI timing diagram - slave mode and CPHA = $1^{(1)}$ .
		Updated Figure 48: TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline.
		Added reference to optional marking or inset/upset marks in all package device marking sections. Updated note below marking schematics.

Revision history STM32L073xx

Table 91. Document revision history (continued)

Date Ro	evision	Changes
08-Oct-2019	evision 5	Added UFQFPN48 package.  Updated Arm logo and added Arm word mark notice in Section 1: Introduction.  Removed Cortex logo.  Updated Table 5: Functionalities depending on the working mode (from Run/active down to standby) to change I2C functionality to disabled in Low-power Run and Low-power Sleep modes.  Updated V <sub>DD_USB</sub> description in Section 3.4.1: Power supply schemes.  Replaced LCD_VLCD2 by LCD_VLCD1 in Section 3.13.2: V <sub>LCD</sub> voltage monitoring.  Changed USARTx_RTS, USARTx_RTS_DE into USARTx_RTS_DE into LPUART1_RTS_LPUART1_DE in Section 4: Pin descriptions and in all alternate function tables. In Table 16: STM32L073xx pin definition changed PB2 and PB12/PE11 additional functions to LCD_VLCD2 and LCD_VLCD1, respectively. Updated V <sub>DD_USB</sub> and note 2. in Table 26: General operating conditions.  Updated t <sub>AF</sub> maximum value for range 1 in Table 73: I2C analog filter characteristics.  Updated paragraph introducing all package marking schematics to add the new sentence "The printed markings may differ depending on the supply chain."  Updated Figure 48: TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline, Figure 49: TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch thin profile fine pitch ball grid array package outline, Figure 49: TFBGA64 – 64-ball, 5 x 5 mm, 0.5 mm pitch, thin profile fine pitch ball grid array recommended footprint and Figure 87: TFBGA64

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