

Complex Combinational Logic: Implementation and Design Tradeoffs

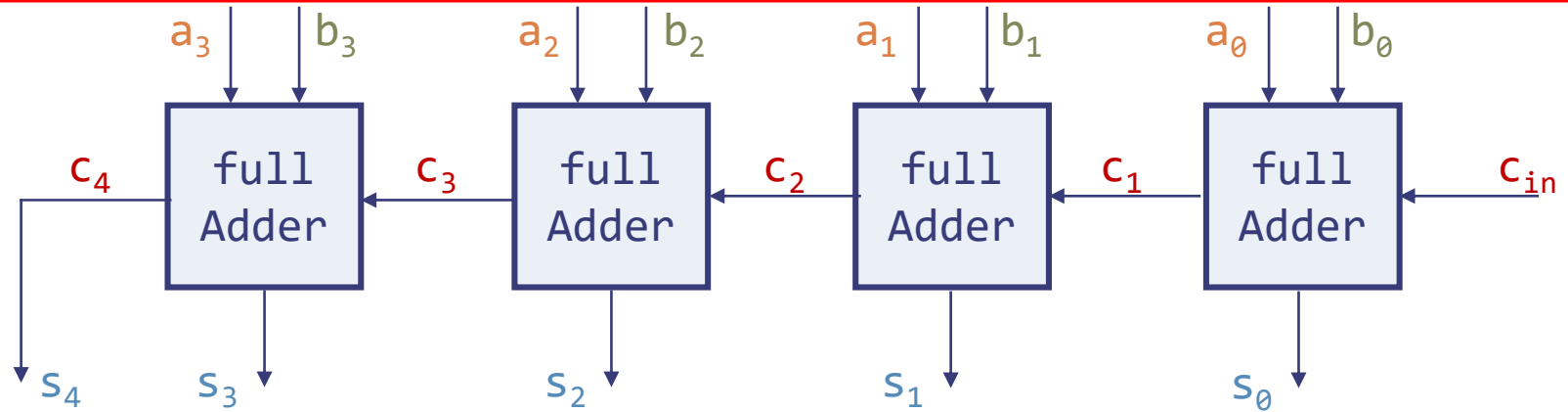
Lecture Goals

- Learn some advanced Minispec features that enable implementing large circuits succinctly
 - Parametric functions
 - Type inference and user-defined types
 - Loops and control-flow statements

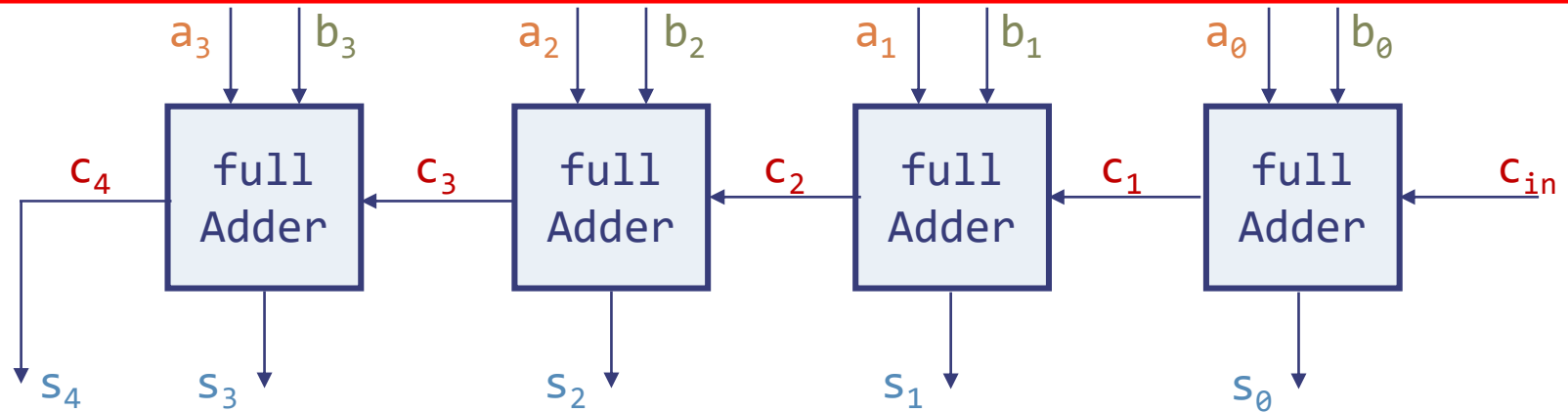
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 - Type inference and user-defined types
 - Loops and control-flow statements
- Study design tradeoffs in combinational logic by analyzing different adder implementations

Reminder: 4-bit Ripple-Carry Adder

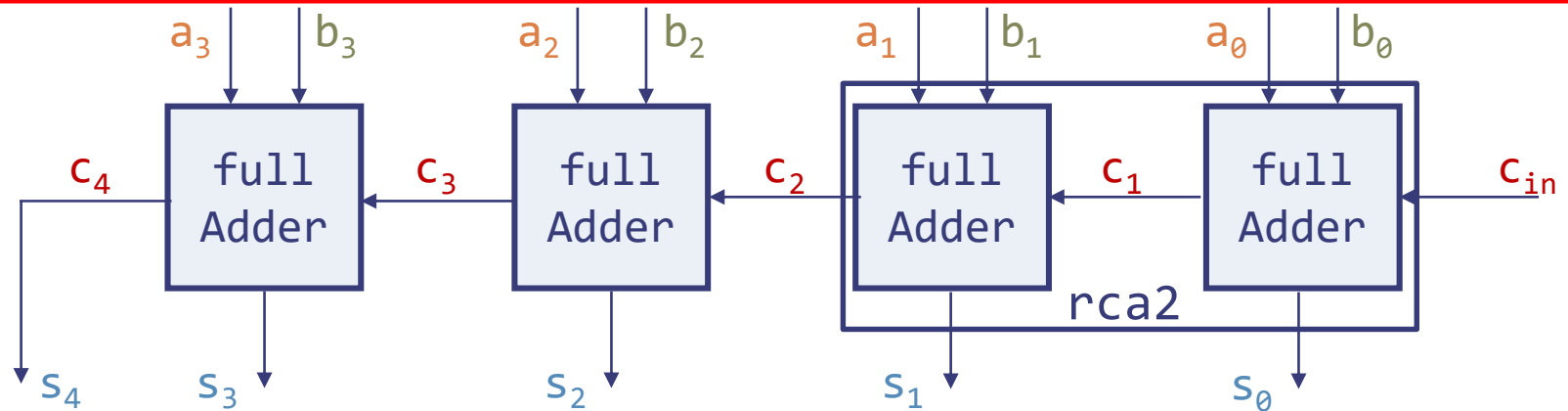


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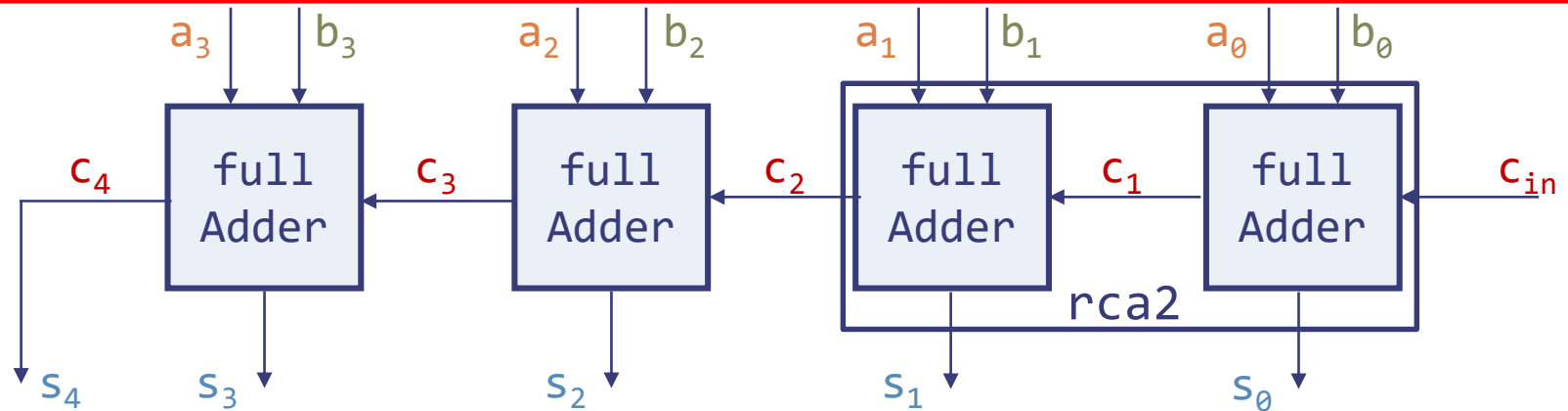
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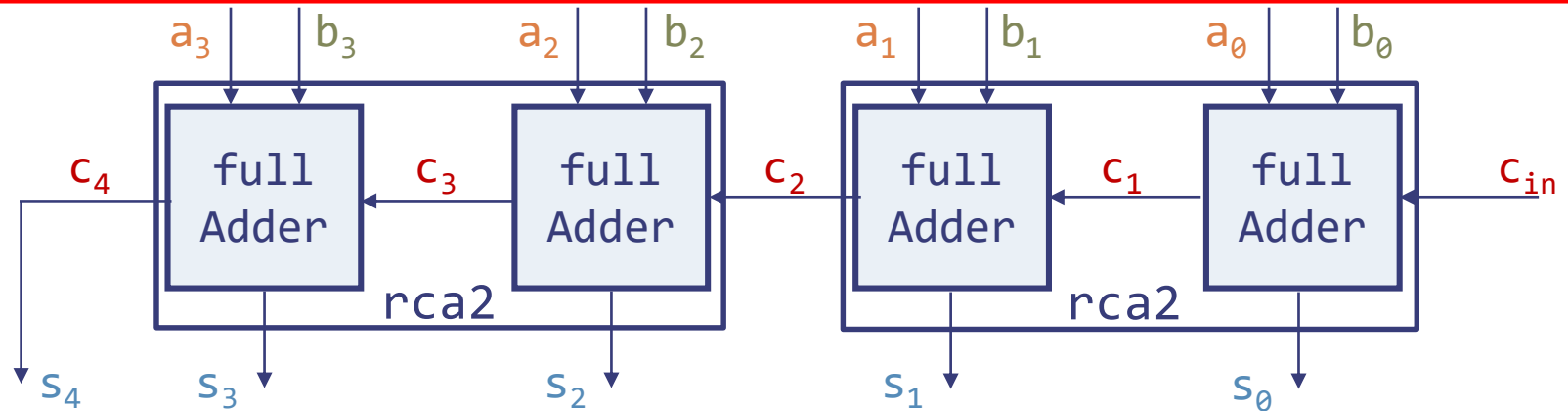
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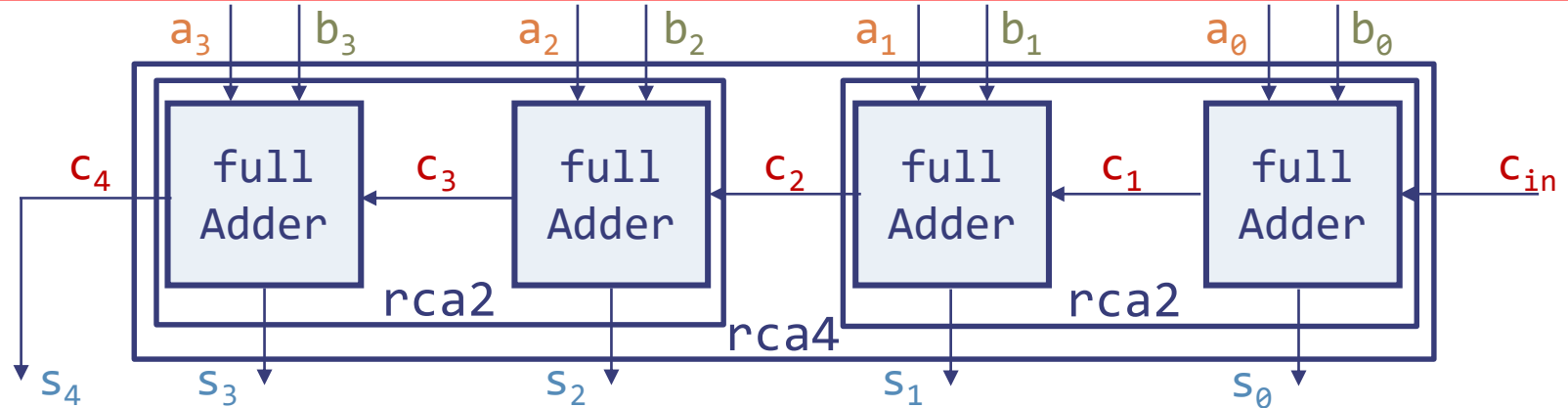
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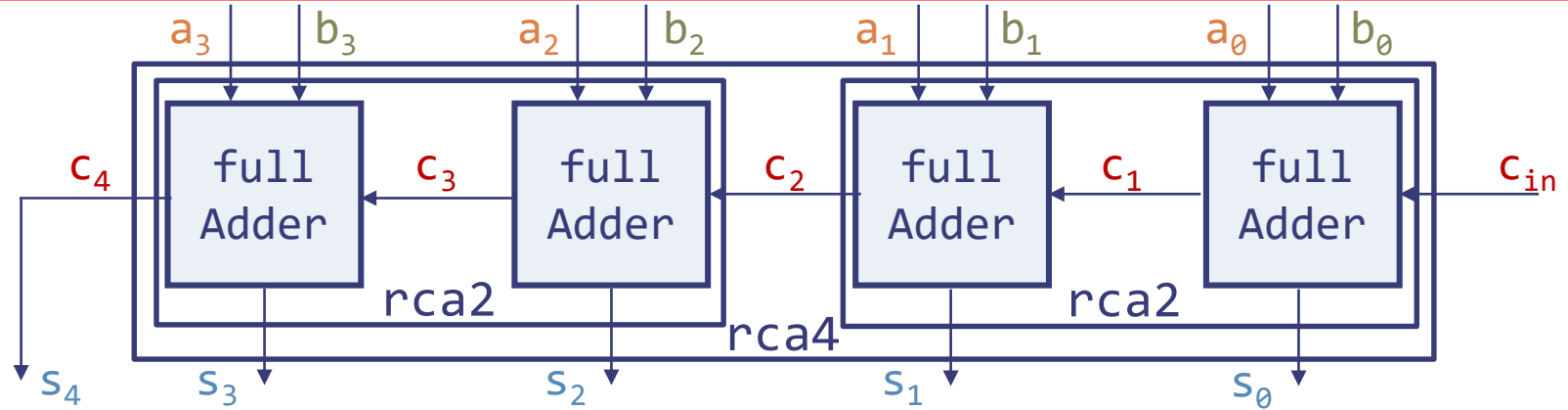
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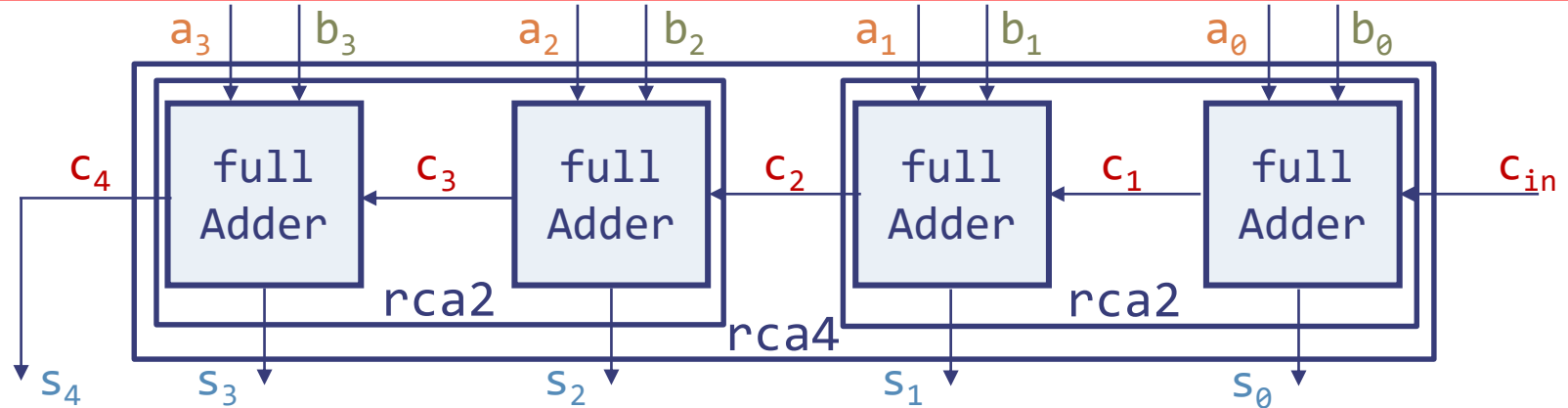


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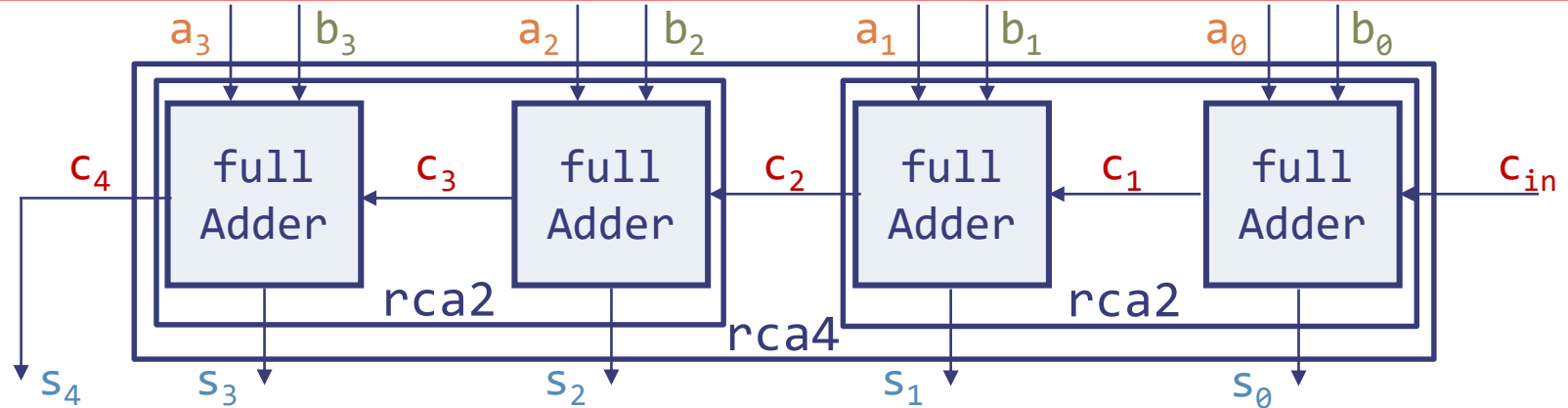
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- Problem 2: If we build large functions from smaller ones, have to write many functions!

Parametric Types

- $\text{Bit\#}(n)$, an n -bit value, is a **parametric type**
 - n is the **parameter** (an Integer value)
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- Parameters can be Integers or types
 - Example: $\text{Vector}\#(n, T)$ is an n -element vector of T 's (e.g., $\text{Vector}\#(4, \text{Bit}\#(8))$ = 4-elem vector of 8-bit values)

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- A parametric function must be invoked with fixed parameters, which instantiates a *concrete* function
 - Example: Calling $\text{rca\#}(32)$ instantiates a 32-bit adder

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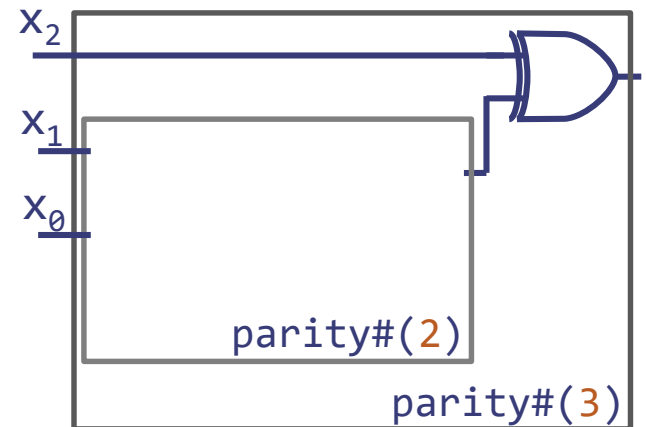
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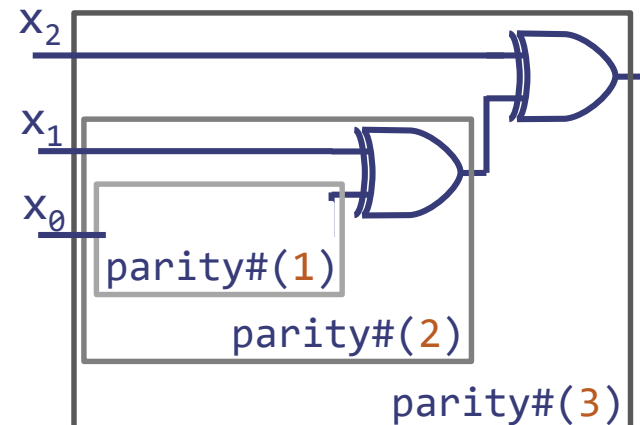


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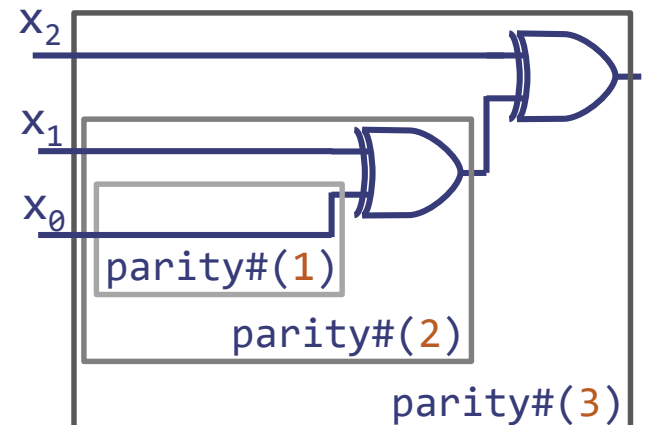


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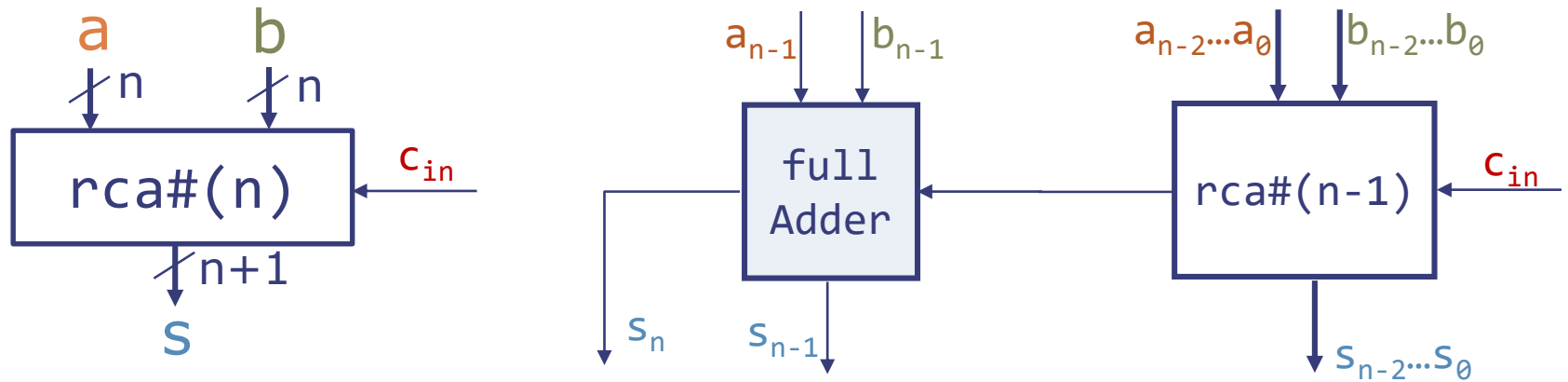
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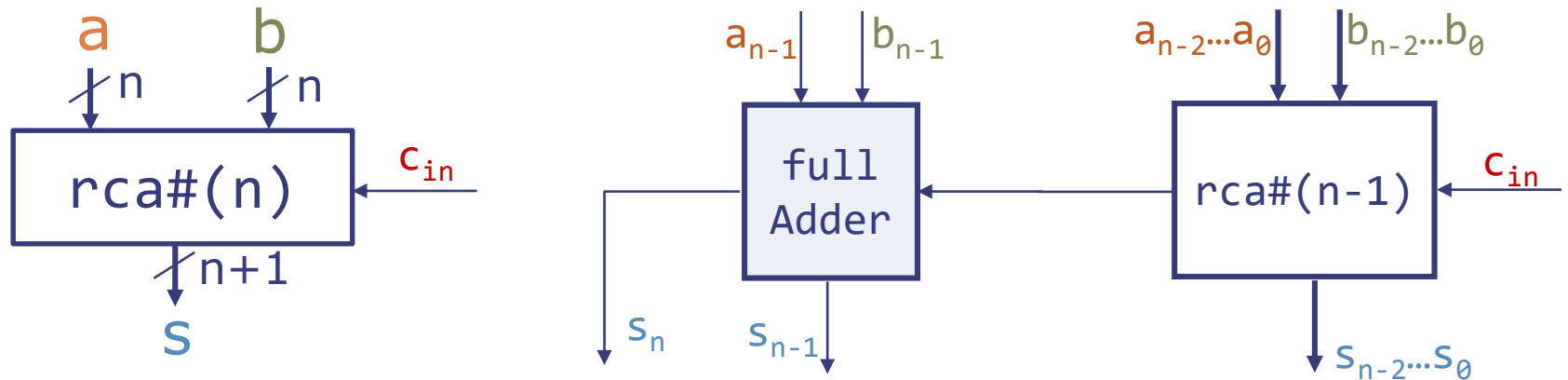
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- Integer supports the same operations as Bit#(n), (arithmetic, logical, comparisons, etc.)
 - But evaluated by compiler → operations on Integers never produce any hardware

N-bit Ripple-Carry Adder



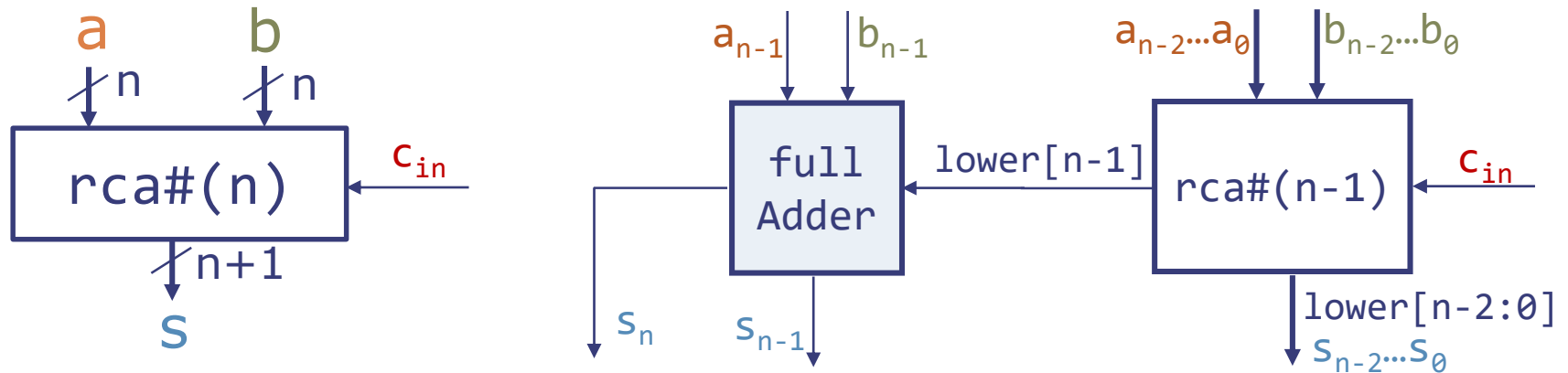
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```

```
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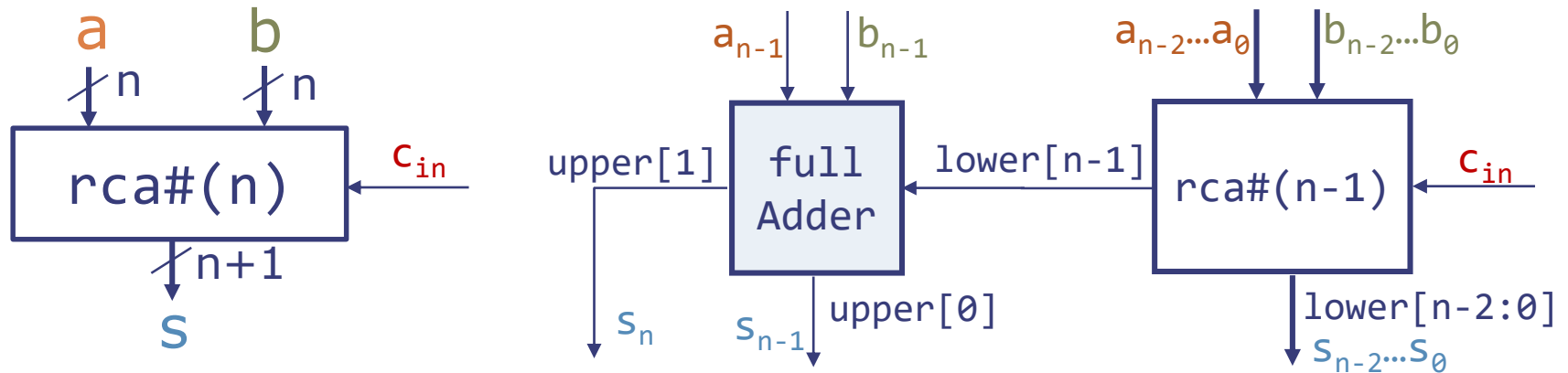
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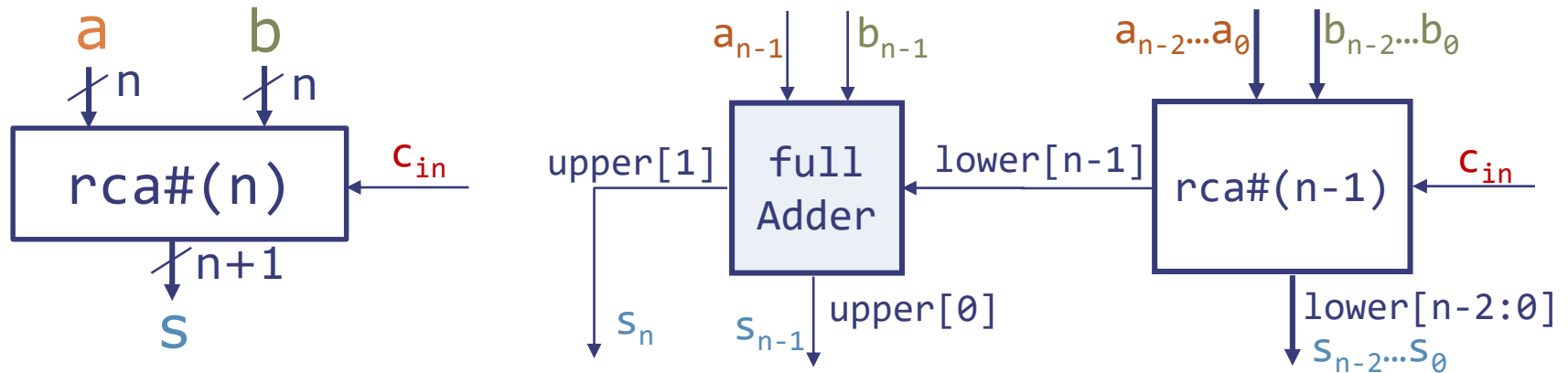
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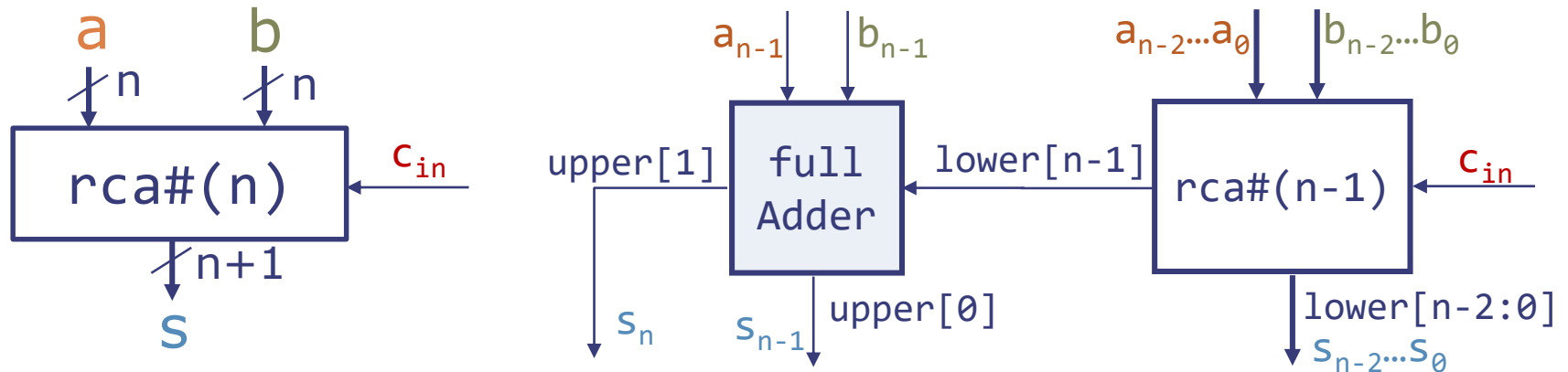
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endfunction
```

// Base case

```
function Bit#(2) rca#(1)(Bit#(1) a, Bit#(1) b, Bit#(1) cin);
    return fullAdder(a, b, cin);
endfunction
```

Type Inference

- You can omit the type of a variable by declaring it with the `let` keyword
- The compiler infers the variable's type from the type of the expression assigned to the variable

```
Bit#(4) x = 4'b0011;  
let y = x;           // y has type Bit#(4)  
let z = {x, x};       // z has type Bit#(8)  
let w = 2'b11;        // w has type Bit#(2)  
let n = 42;           // n has type Integer
```


User-Defined Types

- **Type synonyms** allow giving a different name to a type

```
typedef Bit#(8) Byte;
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    Byte green;  
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} Pixel;
```

```
Pixel p;  
p.red = 255;
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- Structs and enums are much clearer than using raw bits!
 - e.g., `Bit#(24) pixel; pixel[15:8]` versus `Pixel pixel; pixel.green`

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For Loops



- For loop statements allow compactly expressing a sequence of similar statements

```
Bit#(6) w = 0;  
for (Integer i = 0; i < 6; i = i + 1)  
    w[i] = z[i / 2];
```

For Loops



- For loop statements allow compactly expressing a sequence of similar statements

```
Bit#(6) w = 0;  
for (Integer i = 0; i < 6; i = i + 1)  
    w[i] = z[i / 2];
```

- For loops are not like loops in software programming languages!
 - **Fixed number of iterations**
(Integer induction variable!)
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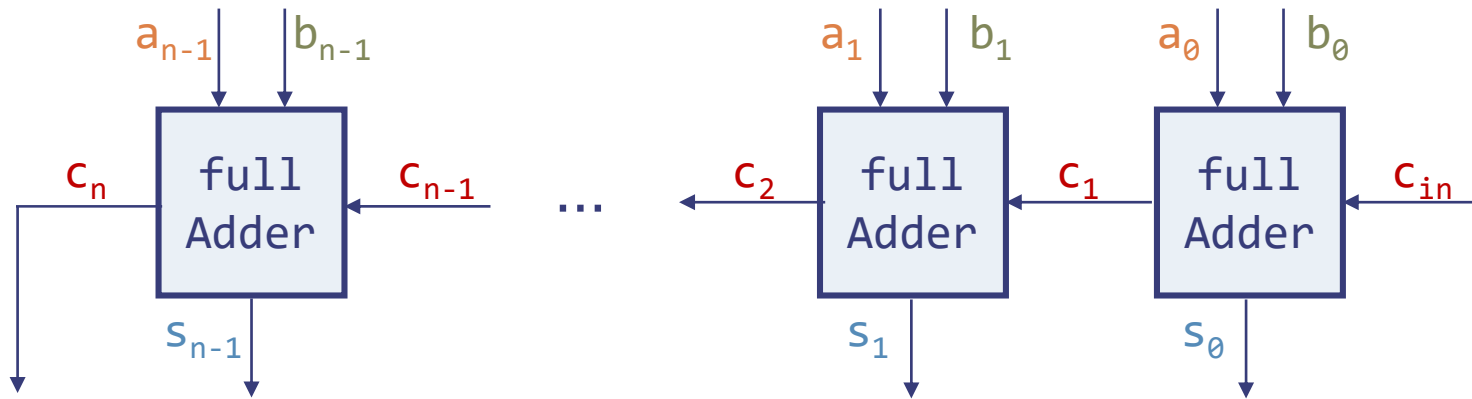
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- Example: The loop above is translated into this sequence:

```
w[0] = z[0];  
w[1] = z[0];  
w[2] = z[1];  
w[3] = z[1];  
w[4] = z[2];  
w[5] = z[2];
```

N-bit Ripple-Carry Adder with Loop



```
function Bit#(n+1) rca#(Integer n)(Bit#(n) a, Bit#(n) b, Bit#(1) cin);
    Bit#(n) s = 0;
    Bit#(n+1) c = {0, cin};
    for (Integer i = 0; i < n; i = i + 1) begin
        let x = fullAdder(a[i], b[i], c[i]);
        s[i] = x[0];
        c[i+1] = x[1];
    end
    return {c[n], s};
endfunction
```


Conditional Statements



- If statements have a syntax similar to software:

```
function Bit#(4) max(Bit#(4) a,  
                    Bit#(4) b);  
    Bit#(4) result = b;  
    if (a > b) result = a;  
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- Minispec also has case statements (see tutorial)

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Never forget that you're designing hardware

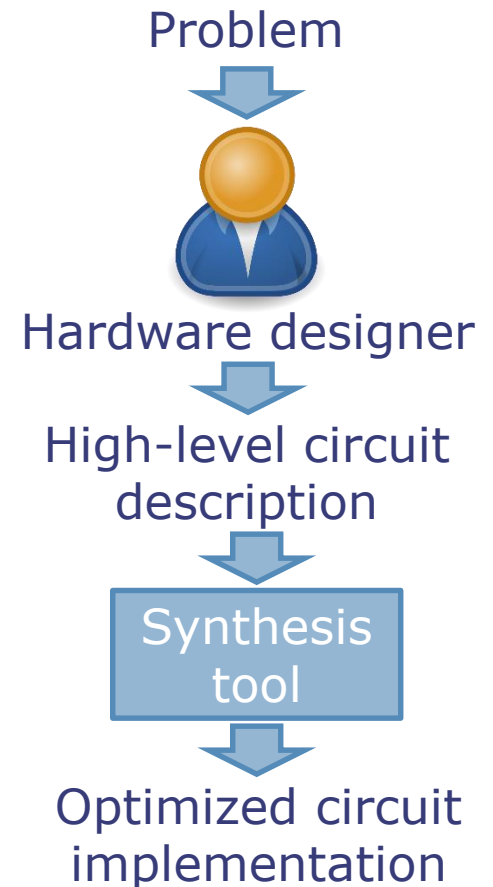
Design Tradeoffs in Combinational Circuits

Algorithmic Tradeoffs in Hardware Design

- Each function often allows many implementations with widely different delay, area, and power

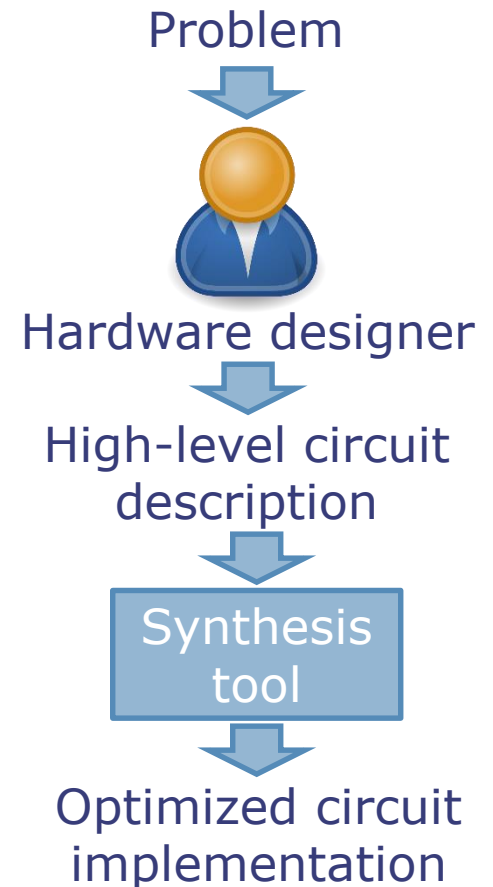
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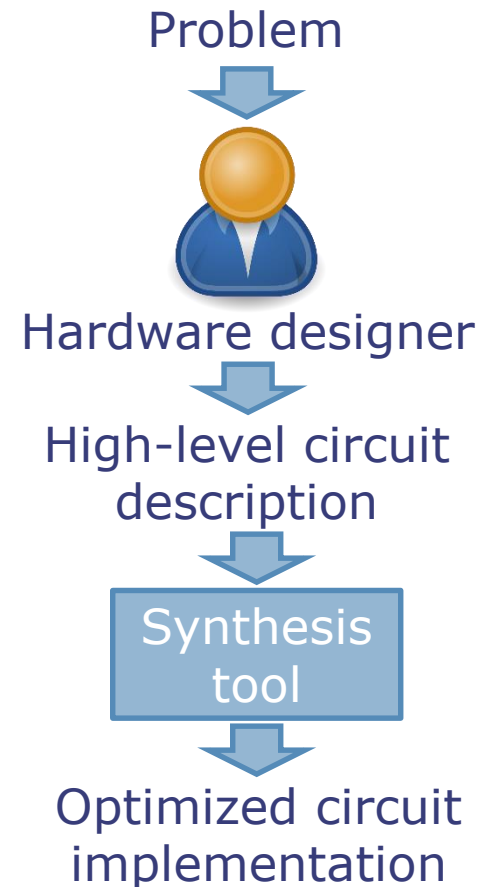
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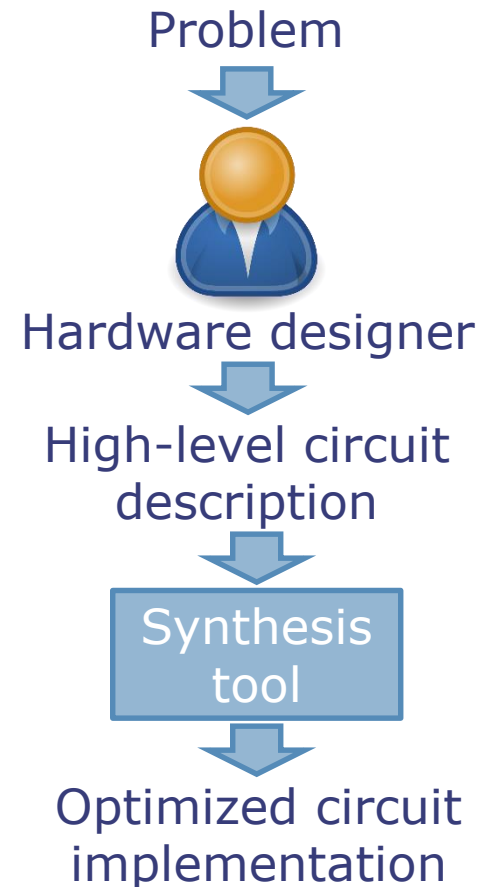
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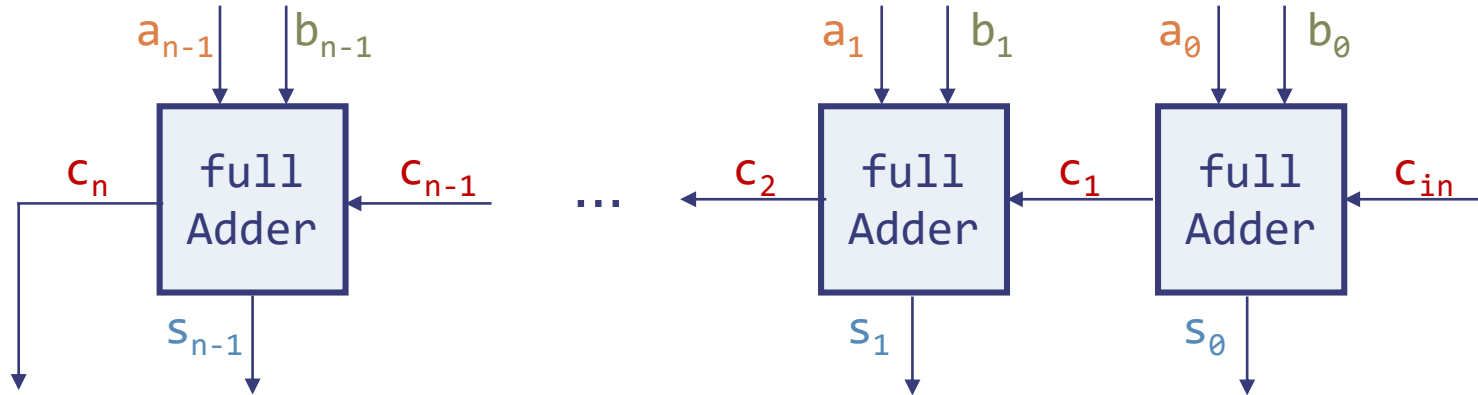


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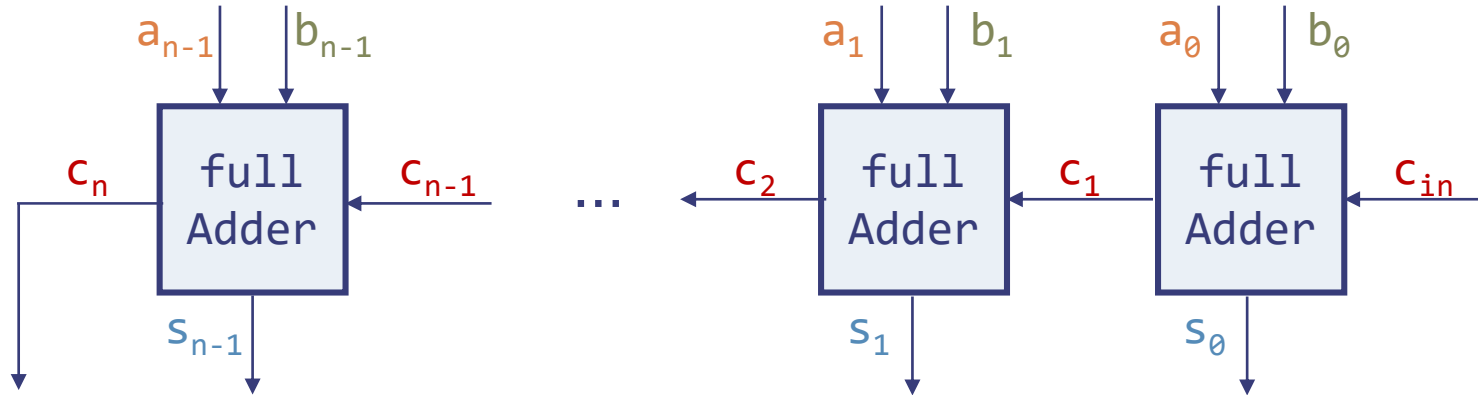
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- Case study: Building a better adder



Ripple-Carry Adder: Simple but Slow

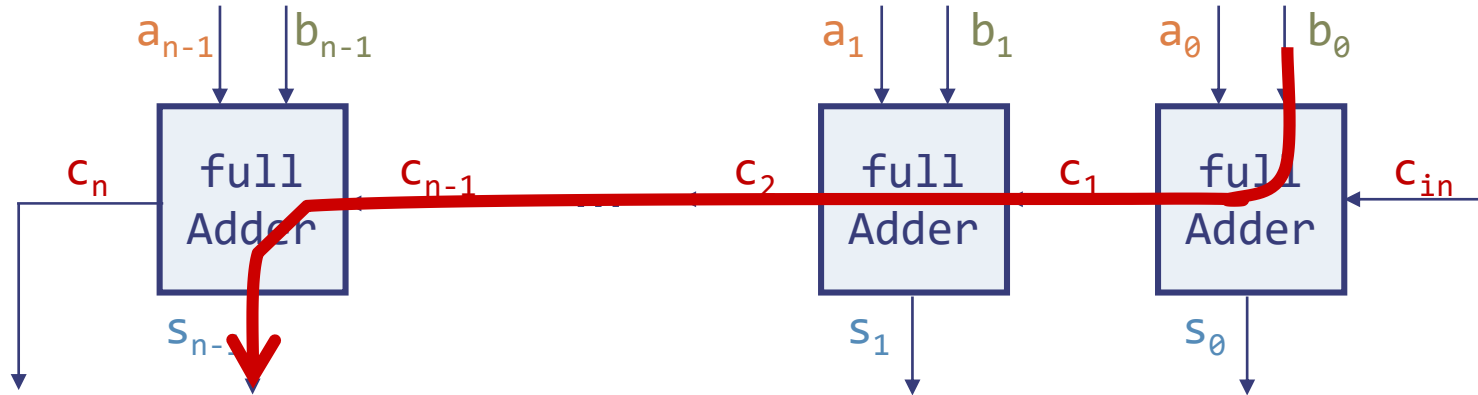


Ripple-Carry Adder: Simple but Slow



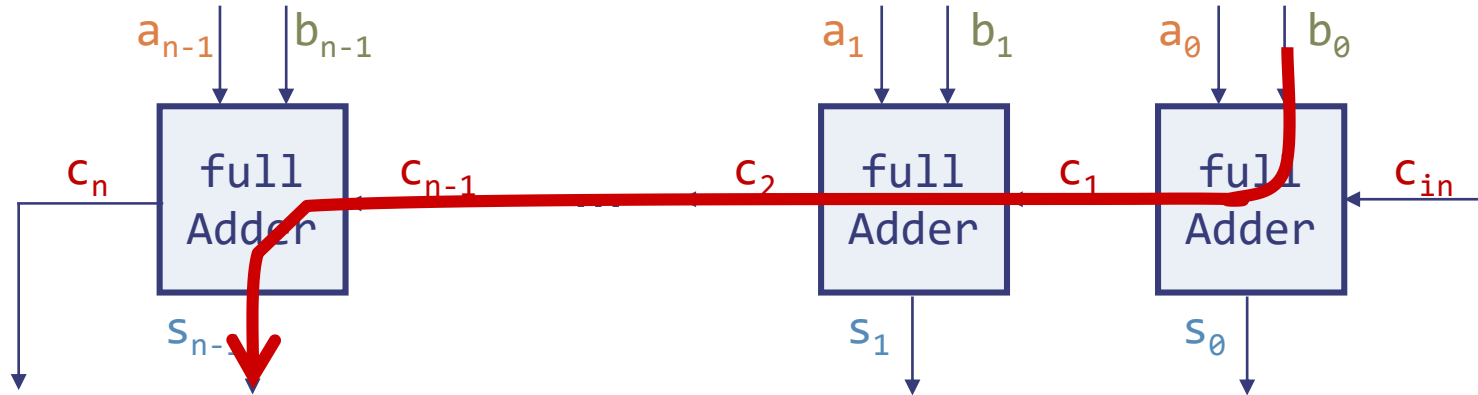
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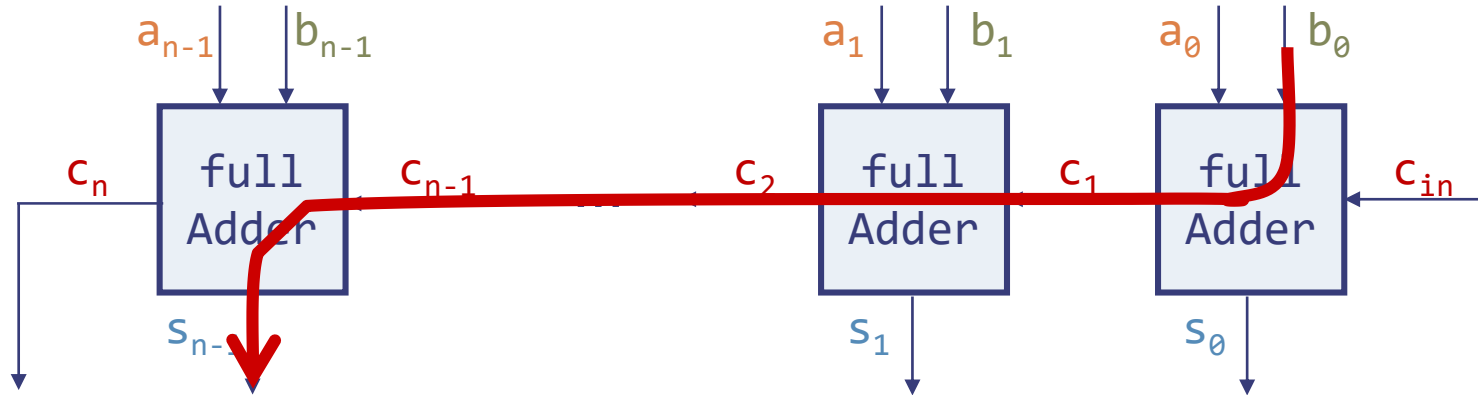
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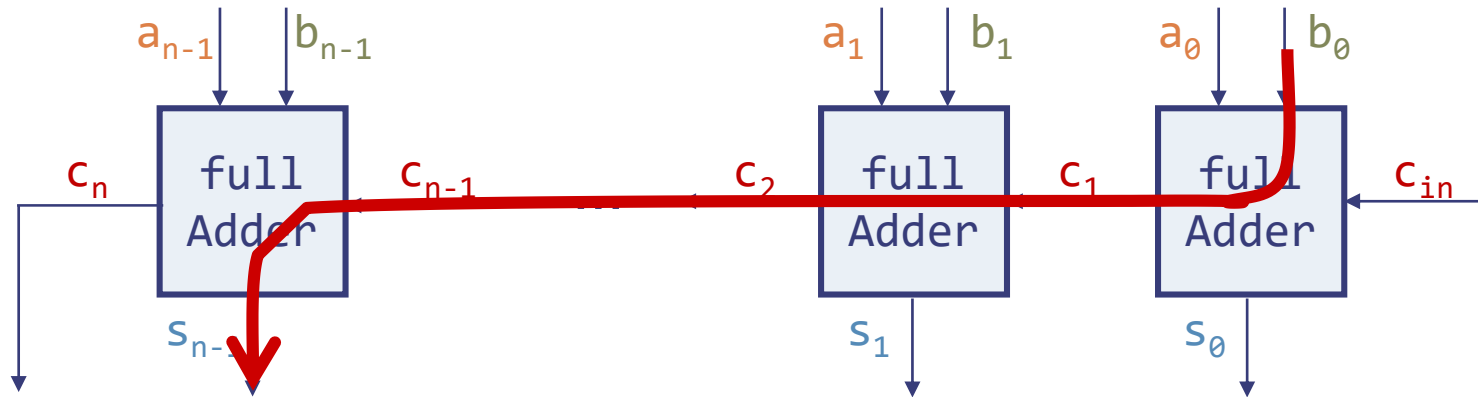
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- $\Theta(n)$ is read "order n" and tells us that the latency of our adder grows **linearly** with the number of bits of the operands

Asymptotic Analysis

- Formally, $g(n) = \Theta(f(n))$ iff there exist $C_2 \geq C_1 > 0$ such that for all but *finitely many* integers $n \geq 0$,

$$C_2 \cdot f(n) \geq g(n) \geq C_1 \cdot f(n)$$

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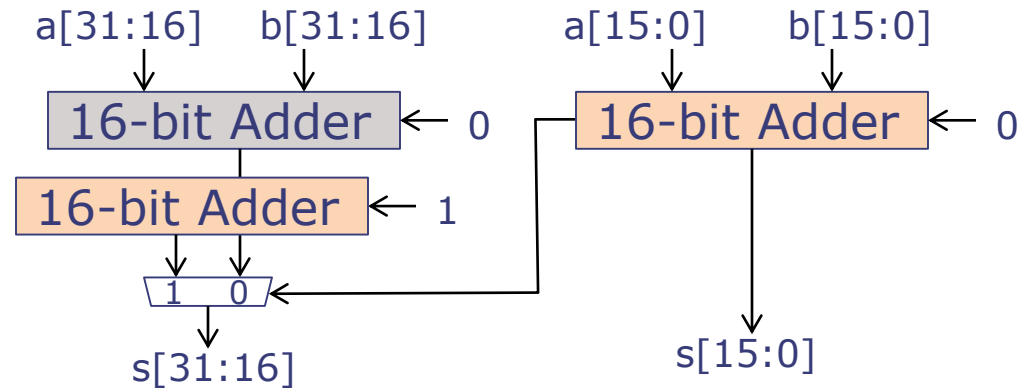
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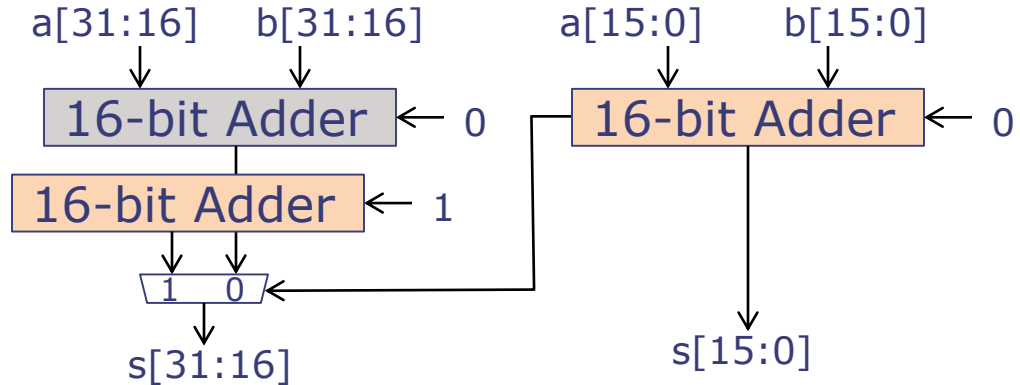
- Example: $n^2 + 2n + 3 = \Theta(n^2)$ (read “is of order n^2 ”) since $2n^2 > n^2 + 2n + 3 > n^2$ except for a few small integers

Carry-Select Adder Trades Area for Speed



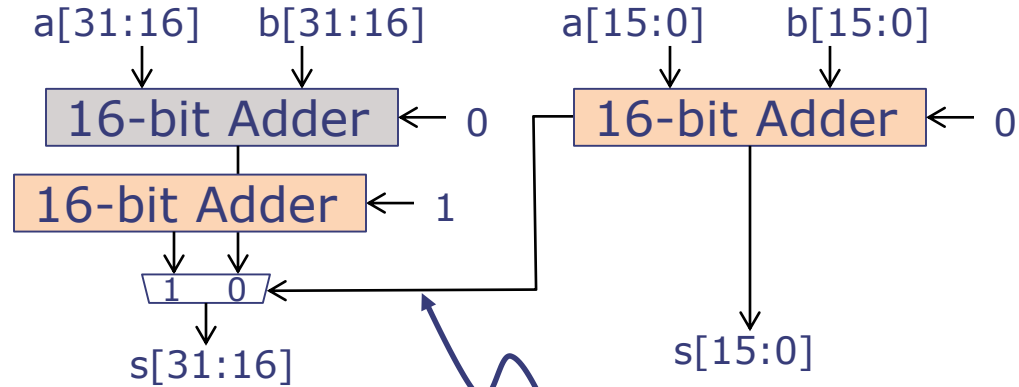
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Two copies of the high half of the adder: one assumes carry-in of "0", the other carry-in of "1"



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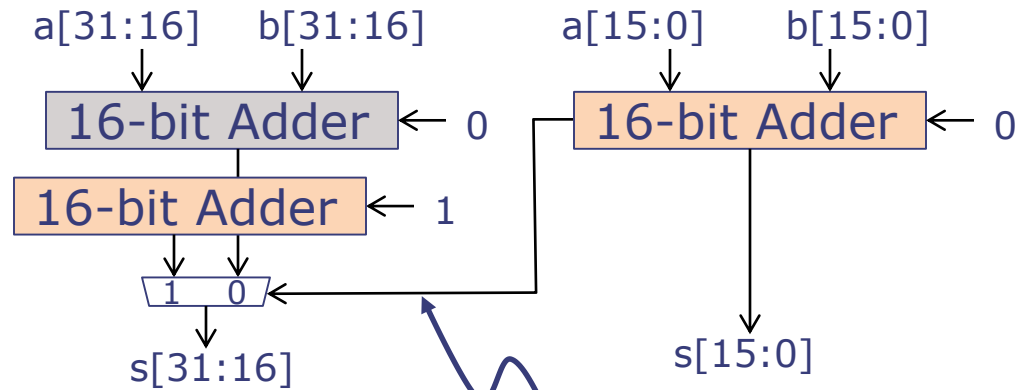
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The carry-out of the low half selects the correct version of the high-half addition.

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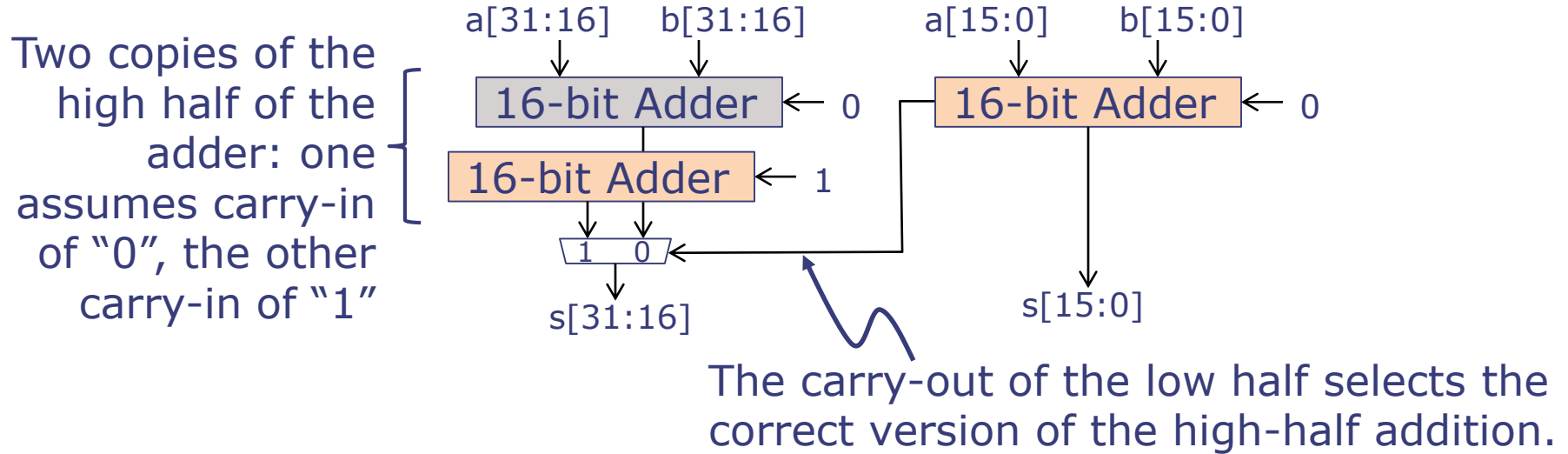
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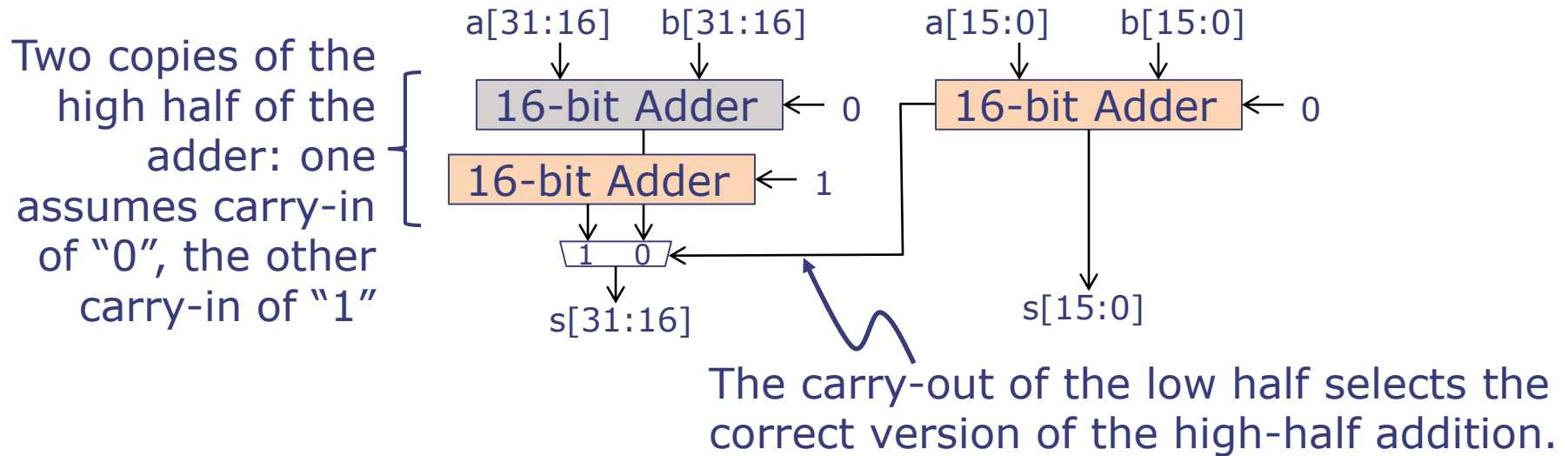
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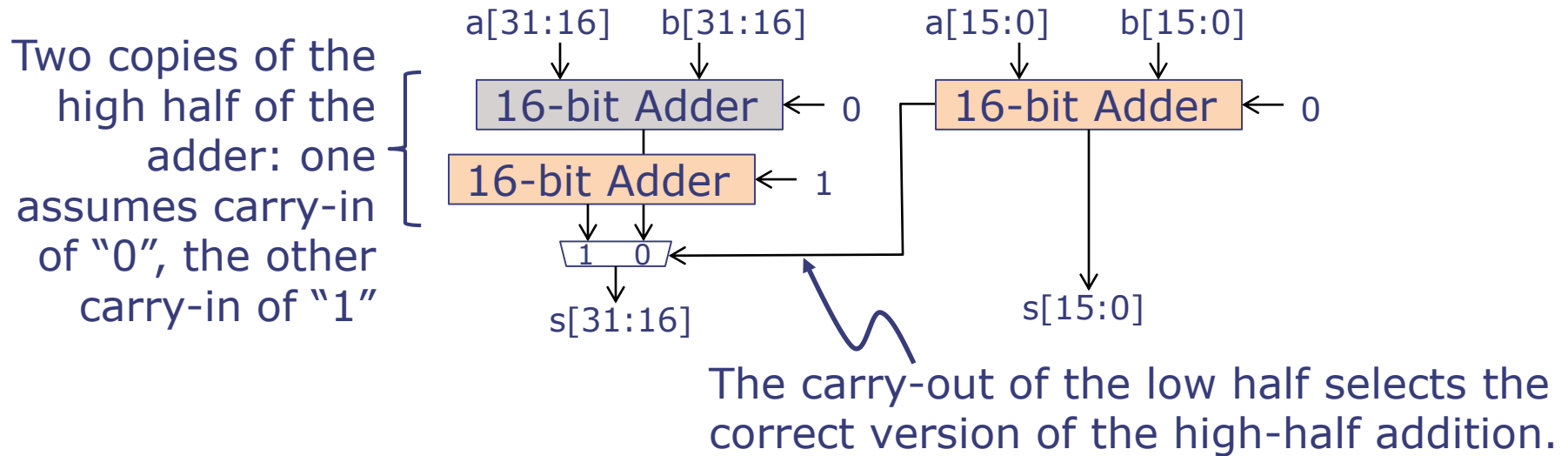
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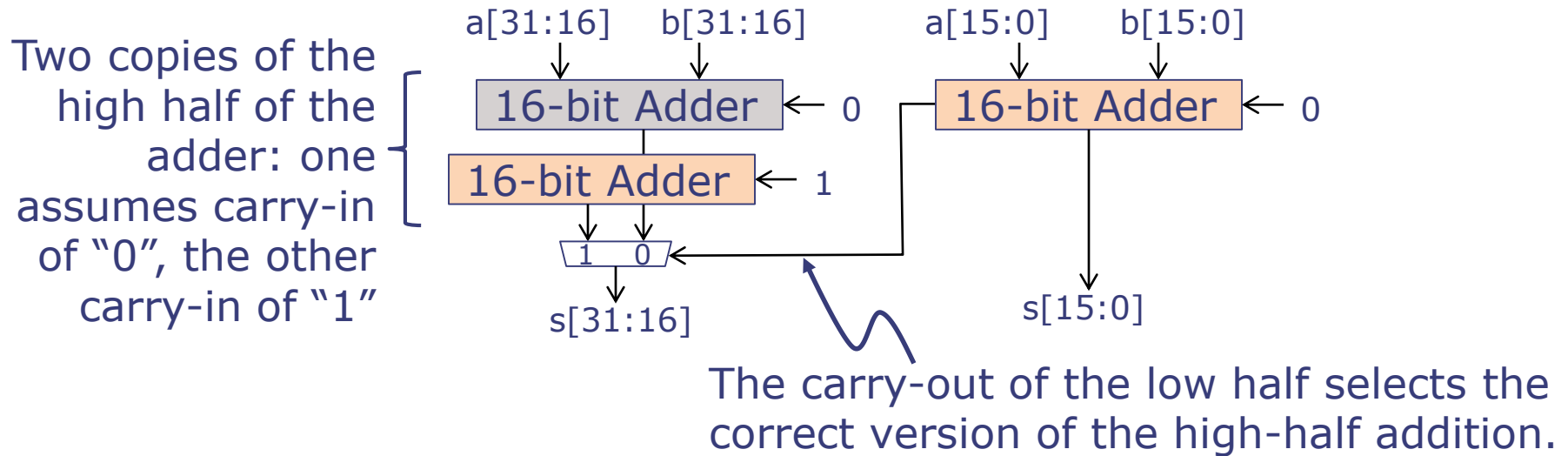
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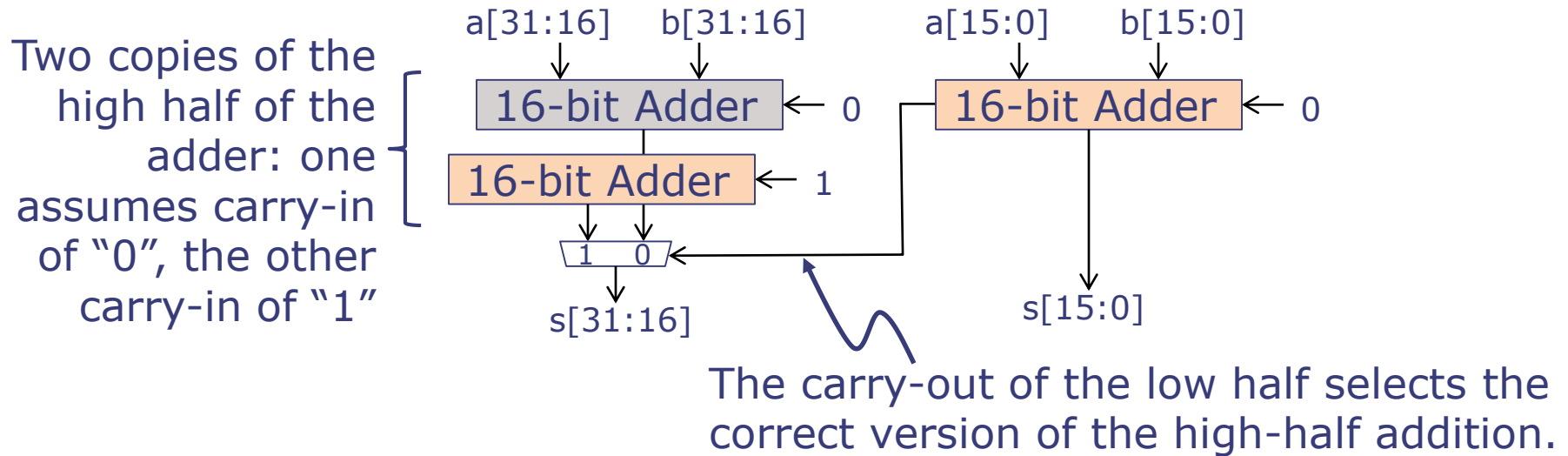
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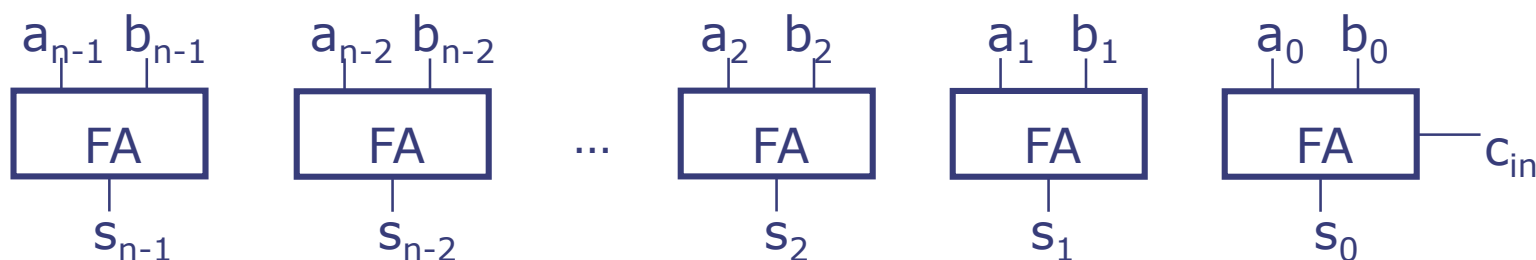
*Drawbacks? Consumes much more area than ripple-carry adder
Wide mux adds significant delay (lab 4)*

Carry-Lookahead Adders (CLAs)

- CLAs compute all carry bits in $\Theta(\log n)$ delay

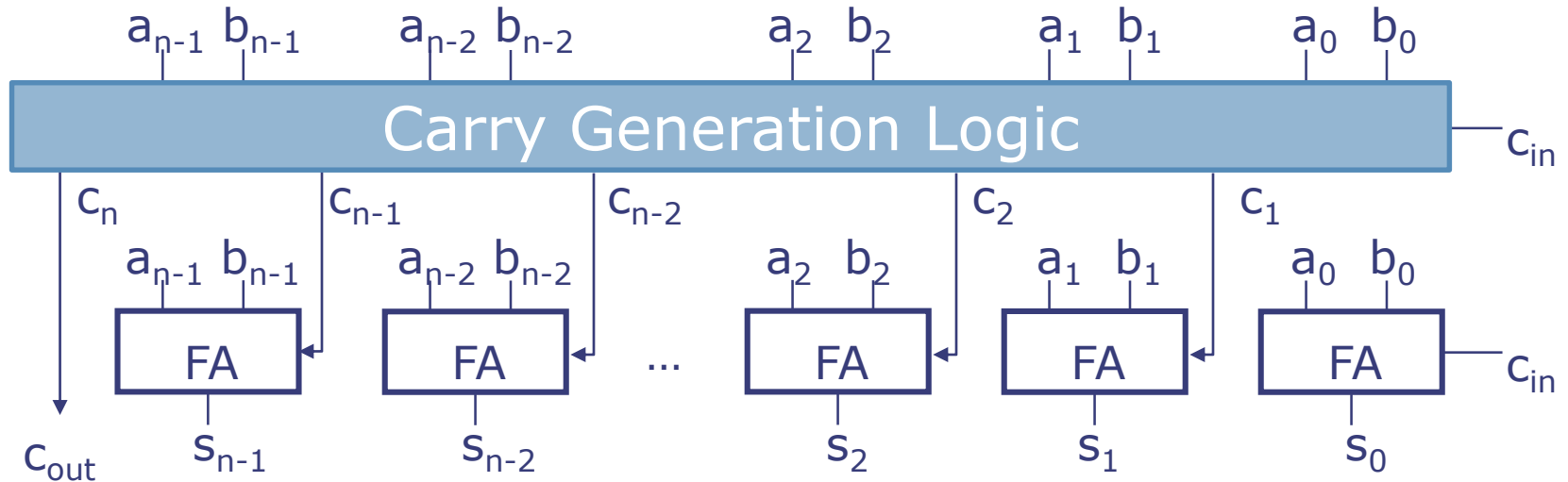
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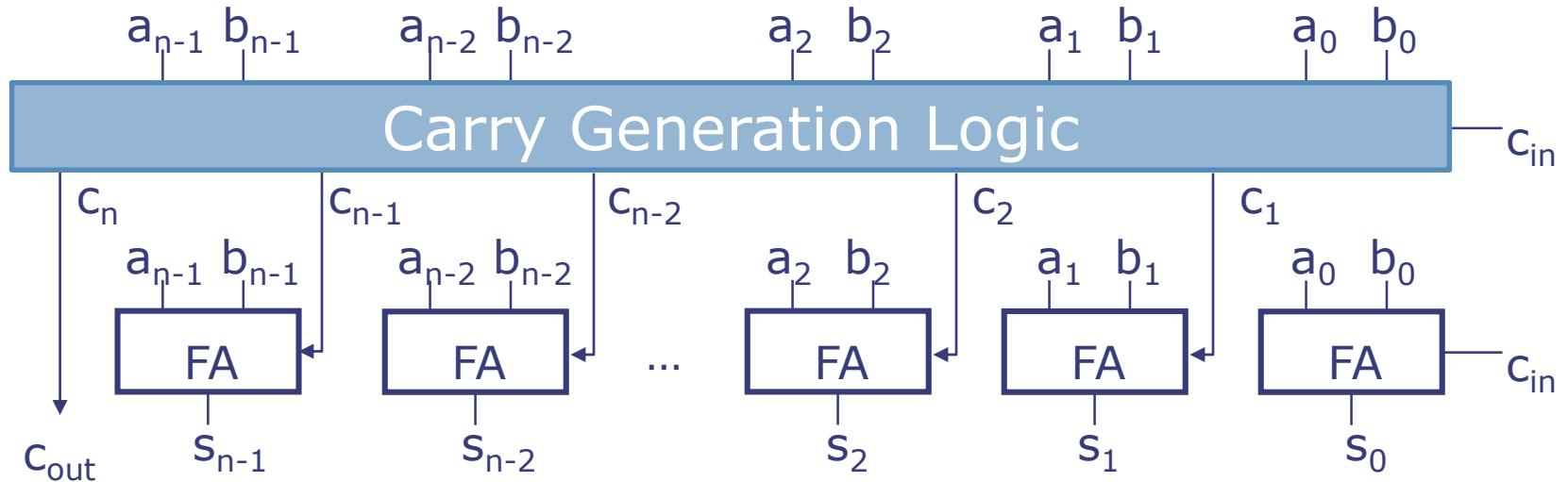
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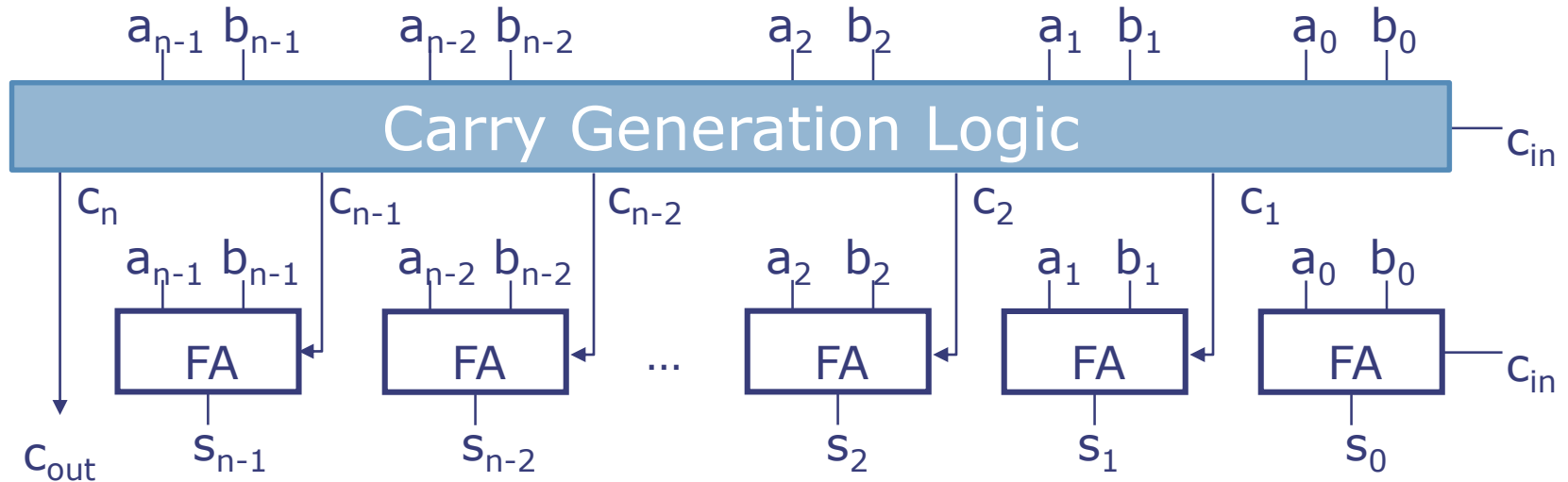
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- Key idea: Transform chain of carry computations into a tree

Carry-Lookahead Adders (CLAs)

- CLAs compute all carry bits in $\Theta(\log n)$ delay



- Key idea: Transform chain of carry computations into a tree
 - Transforming a chain of associative operations (e.g., AND, OR, XOR) into a tree is easy
 - But how to do this with carries?

Carry-Lookahead Adder Details

NOTE: Remaining slides are optional material that will not be on a quiz but will be helpful for Lab 4 and the Design Project

Building a Carry-Lookahead Adder

- Step 1: Generate the output carry in $\Theta(\log n)$ delay
- Step 2: Extend step 1 to generate all carries in $\Theta(\log n)$ delay

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 - Step 1 leverages that **function composition is associative**
 - Step 2 uses the **parallel scan** (a.k.a. parallel prefix) algorithm

Function composition is associative

Basic math reminder ☺

Function compo

Function composition is associative

Basic math reminder ☺

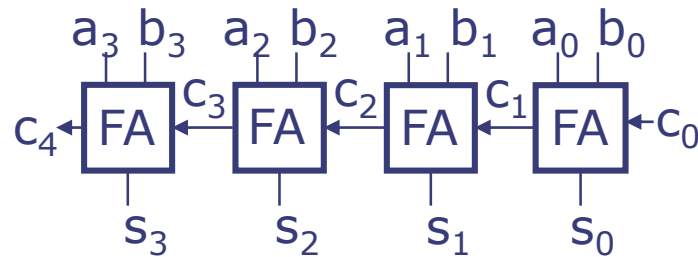
$$\begin{array}{c} f_1 \circ f_2 \circ f_3 = (f_1 \circ f_2) \circ f_3 = f_1 \circ (f_2 \circ f_3) \\ \text{Fu} \\) \\ 3 \ 3 \ 3 \ 2 \ 2 \ 2 \ 1 \ 1 \ 3 \ 3 \ 3 \ 2 \ 2 \ 1 \ 1 \ 1 \ 3 \ 3 \ 2 \ 2 \ 2 \ 1 \ 1 \ 1 \end{array}$$

- Function composition is always associative:

$$f_1 \circ f_2 \circ f_3 = (f_1 \circ f_2) \circ f_3 = f_1 \circ (f_2 \circ f_3)$$

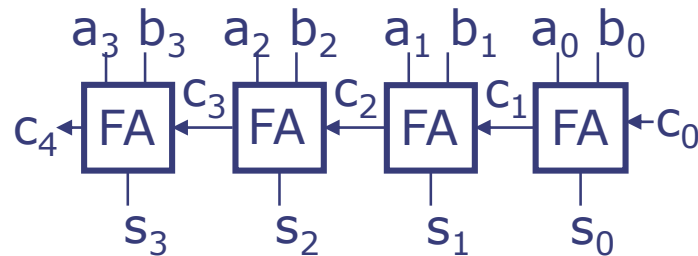
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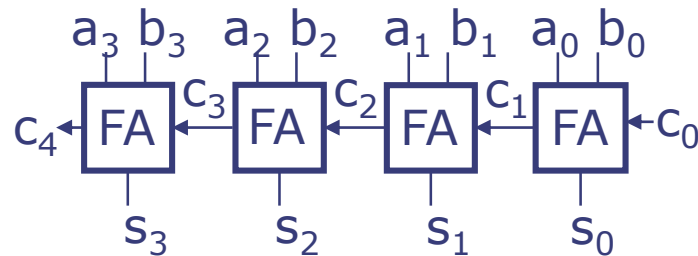


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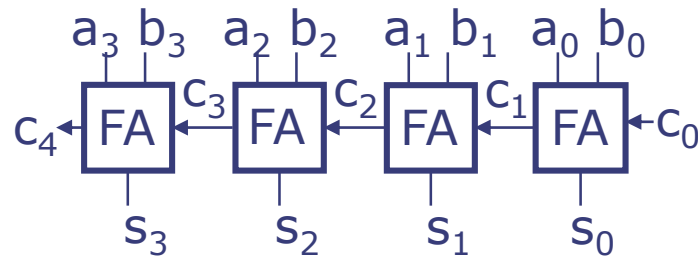
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- A chain of functions f_i , each with 1-bit input and output
- Each f_i is determined by the values of a_i and b_i

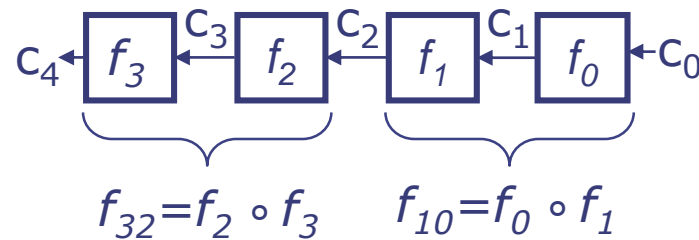
Turning function chains into trees

- Because function composition is associative, we can turn a chain of functions into a tree by first composing the functions...



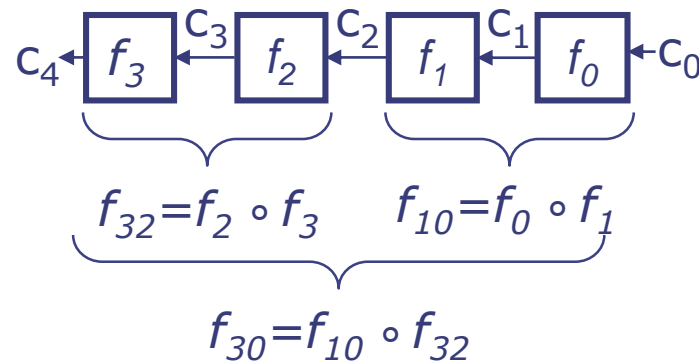
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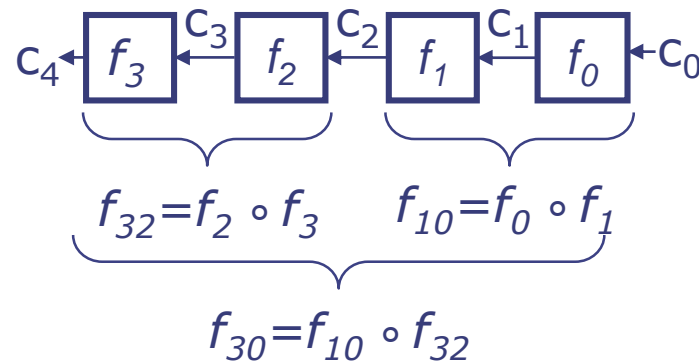
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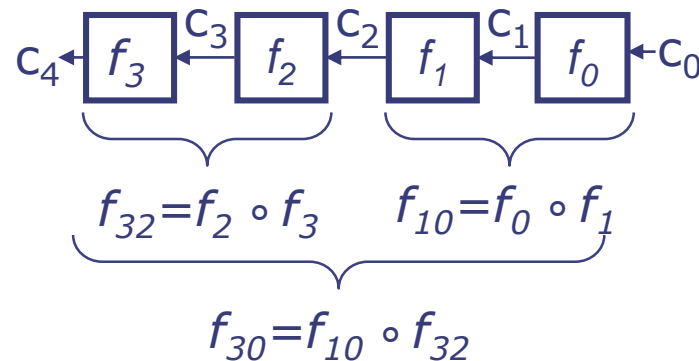
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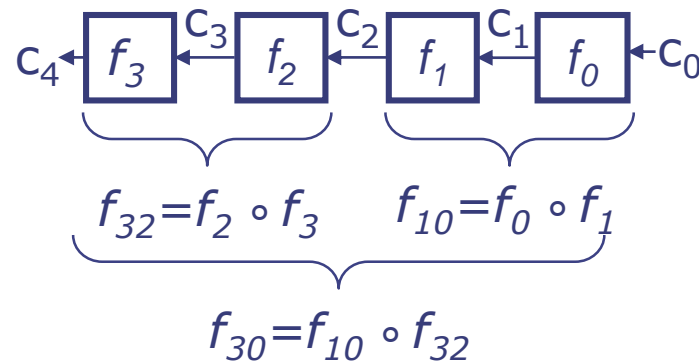


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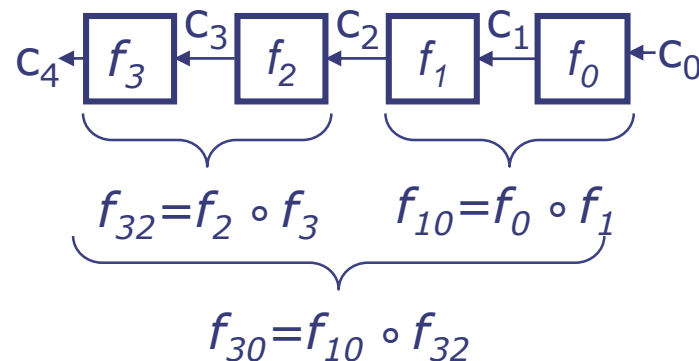


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How does delay grow with chain length n ? $\Theta(\log n)$

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How does delay grow with chain length n ? $\Theta(\log n)$

- Very general trick: Can turn any chain of functions into a tree, *if you can compose them efficiently*

1-bit input, 1-bit output functions

in	out
0	0
1	0

in	out
0	0
1	1

in	out
0	1
1	1

in	out
0	1
1	0

1-bit input, 1-bit output functions

kill

in	out
0	0
1	0

in	out
0	0
1	1

in	out
0	1
1	1

in	out
0	1
1	0

1-bit input, 1-bit output functions

kill

in	out
0	0
1	0

in	out
0	0
1	1

generate

in	out
0	1
1	1

in	out
0	1
1	0

1-bit input, 1-bit output functions

kill

in	out
0	0
1	0

propagate

in	out
0	0
1	1

generate

in	out
0	1
1	1

in	out
0	1
1	0

1-bit input, 1-bit output functions

kill

in	out
0	0
1	0

propagate

in	out
0	0
1	1

generate

in	out
0	1
1	1

invert

in	out
0	1
1	0

1-bit input, 1-bit output functions

kill	
in	out
0	0
1	0

propagate	
in	out
0	0
1	1

generate	
in	out
0	1
1	1

invert	
in	out
0	1
1	0

- *How many bits do we need to enumerate these functions?*

1-bit input, 1-bit output functions

kill	
in	out
0	0
1	0

propagate	
in	out
0	0
1	1

generate	
in	out
0	1
1	1

invert	
in	out
0	1
1	0

- How many bits do we need to enumerate these functions? **2 bits (only 4 choices!)**

Composing 1-bit functions

f	g	$f \circ g$
kill	kill	
kill	generate	
kill	propagate	
kill	invert	
generate	kill	
generate	generate	
generate	propagate	
generate	invert	
propagate	kill	
propagate	generate	
propagate	propagate	
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invert	kill	
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propagate	kill	kill
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propagate	invert	invert
invert	kill	kill
invert	generate	generate
invert	propagate	invert
invert	invert	propagate

Composing 1-bit functions

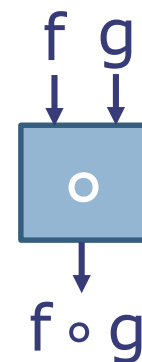
f	g	$f \circ g$
kill	kill	kill
kill	generate	generate
kill	propagate	kill
kill	invert	generate
generate	kill	kill
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generate	propagate	generate
generate	invert	kill
propagate	kill	kill
propagate	generate	generate
propagate	propagate	propagate
propagate	invert	invert
invert	kill	kill
invert	generate	generate
invert	propagate	invert
invert	invert	propagate

This is just a
combinational
function with
two 2-bit inputs
and one 2-bit
output!

Composing 1-bit functions

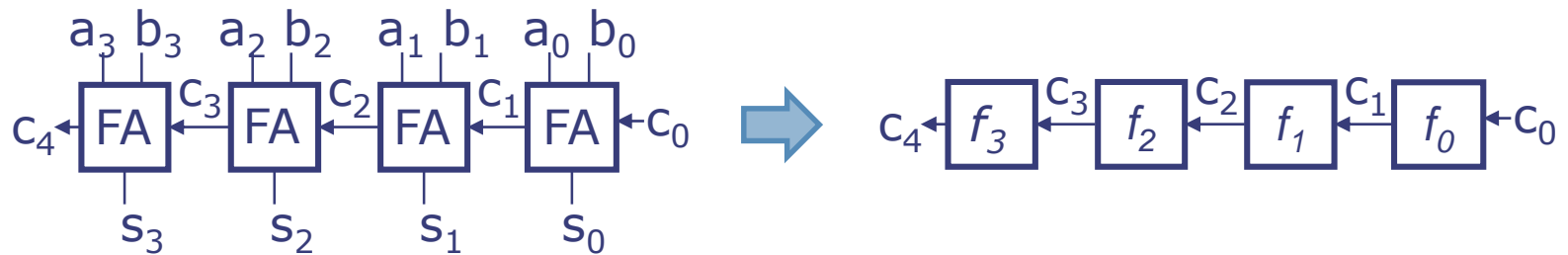
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propagate	generate	generate
propagate	propagate	propagate
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This is just a combinational function with two 2-bit inputs and one 2-bit output!



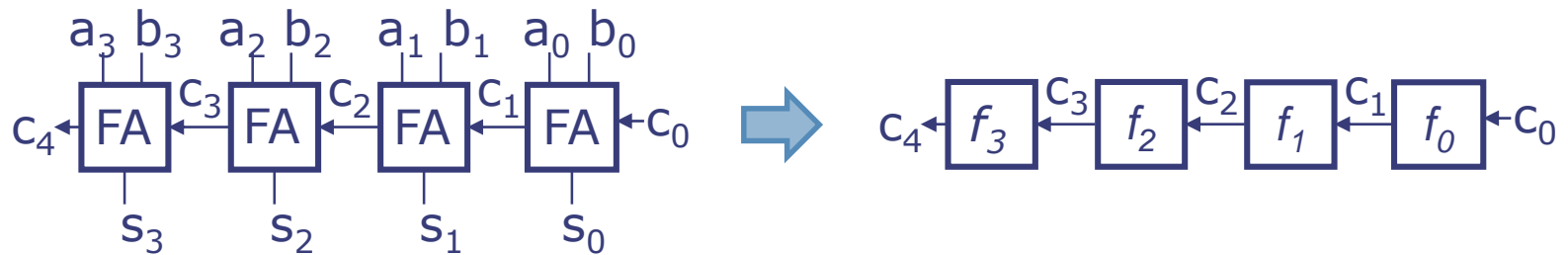
Deriving the initial functions

- Remember, to derive the carry-out, a ripple-carry adder can be seen as a chain of functions f_i , each determined by the values of a_i and b_i



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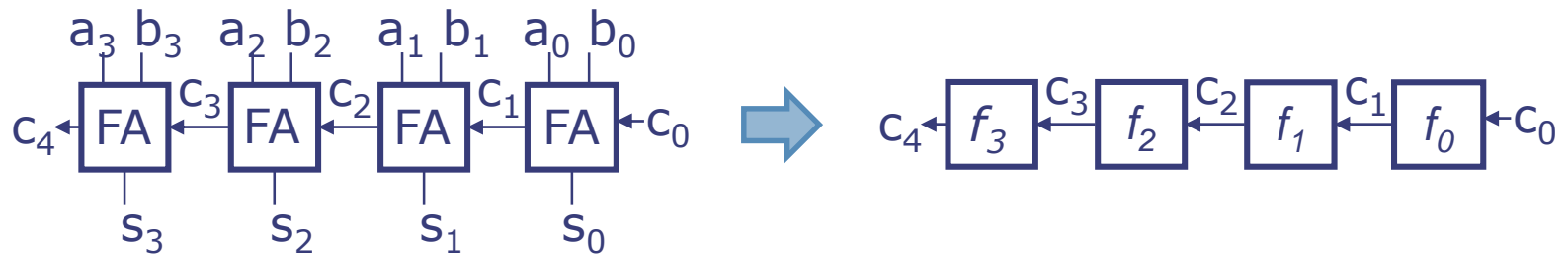
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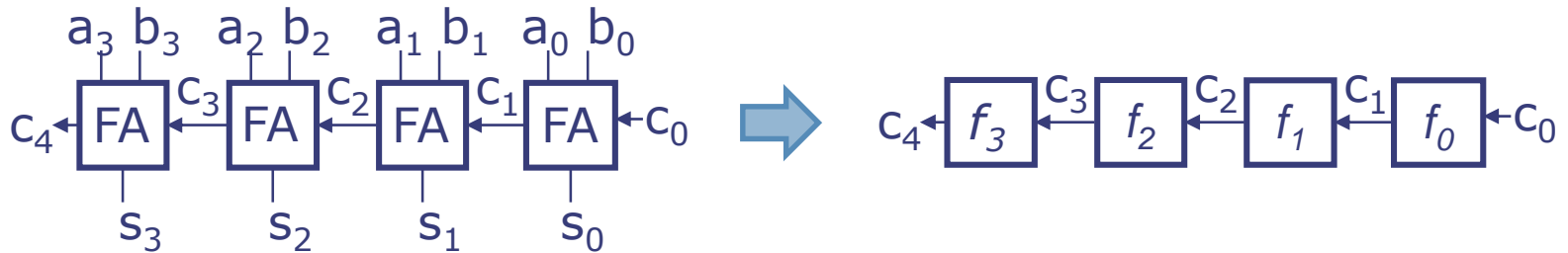


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a	b	c_{in}	c_{out}
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	1

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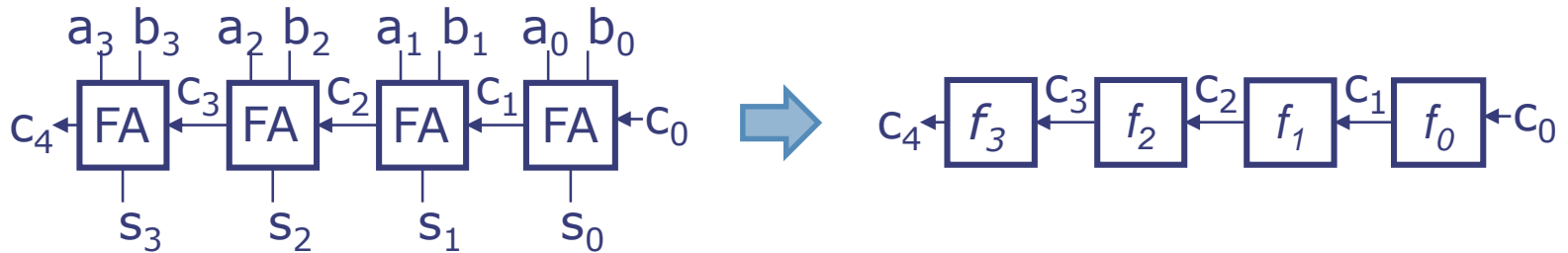


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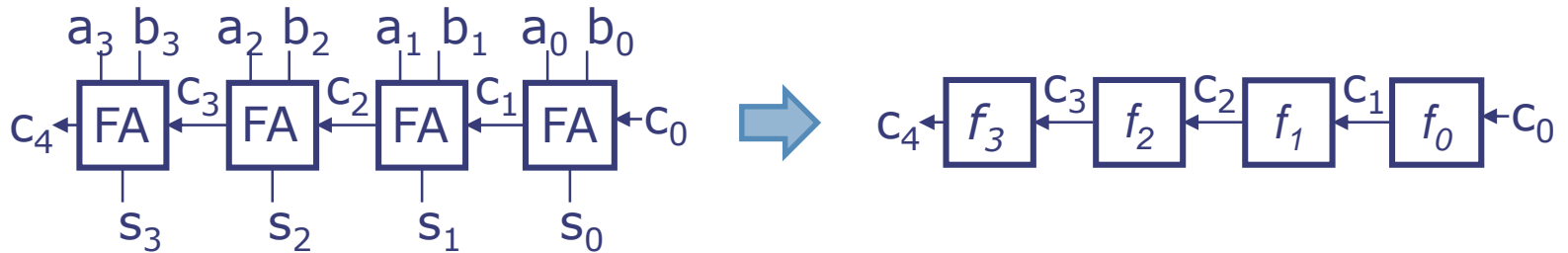


- We can derive these functions from the full adder:

a	b	c_{in}	c_{out}	
0	0	0	0	kill
0	0	1	0	
0	1	0	0	
0	1	1	1	
1	0	0	0	
1	0	1	1	
1	1	0	1	
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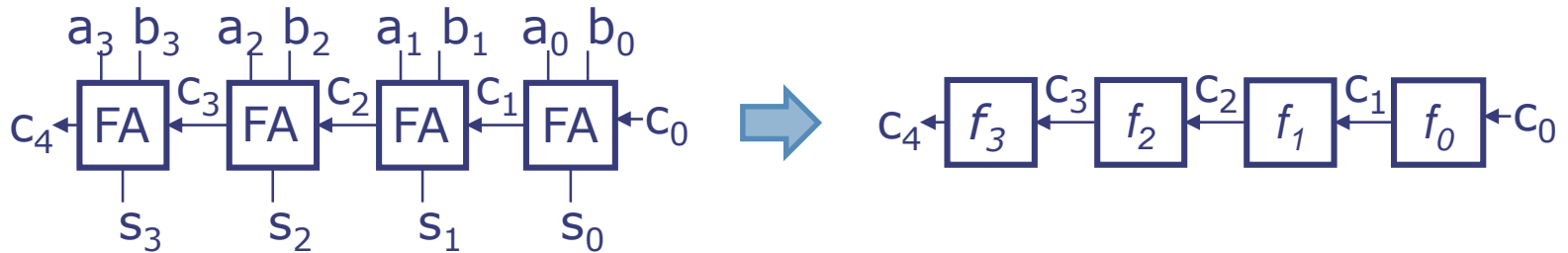


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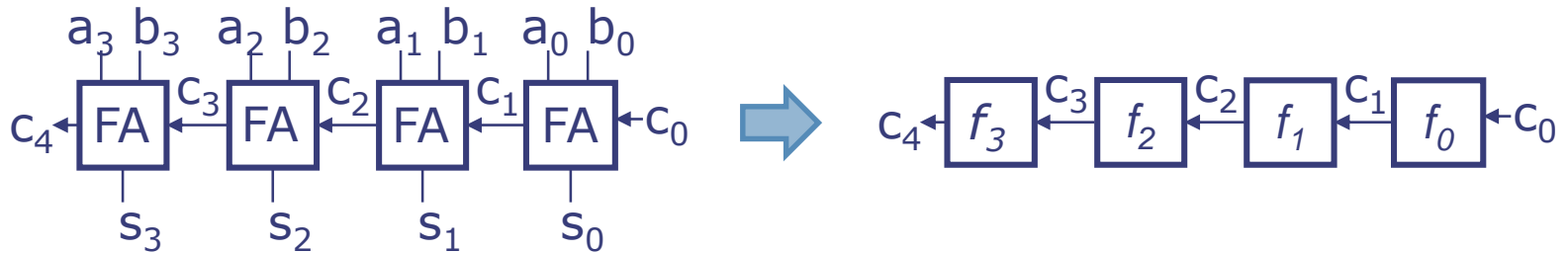


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a	b	c_{in}	c_{out}	
0	0	0	0	kill
0	0	1	0	
0	1	0	0	propagate
0	1	1	1	
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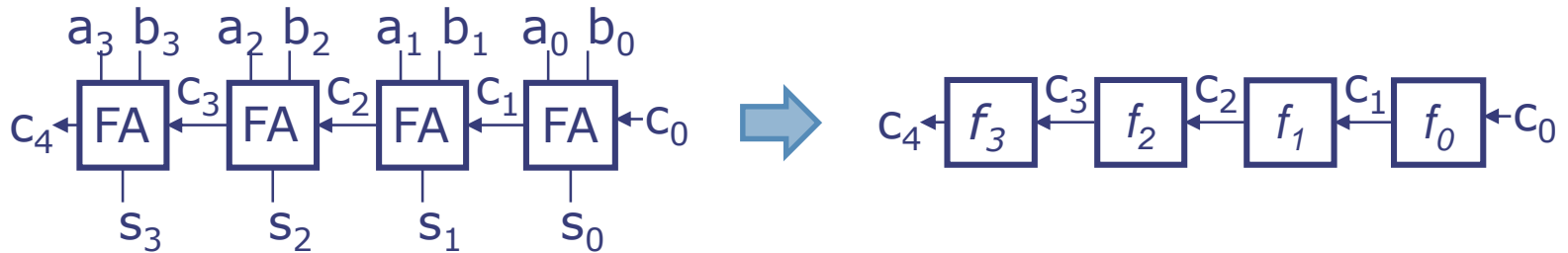


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a	b	c_{in}	c_{out}	
0	0	0	0	kill propagate
0	0	1	0	
0	1	0	0	
0	1	1	1	
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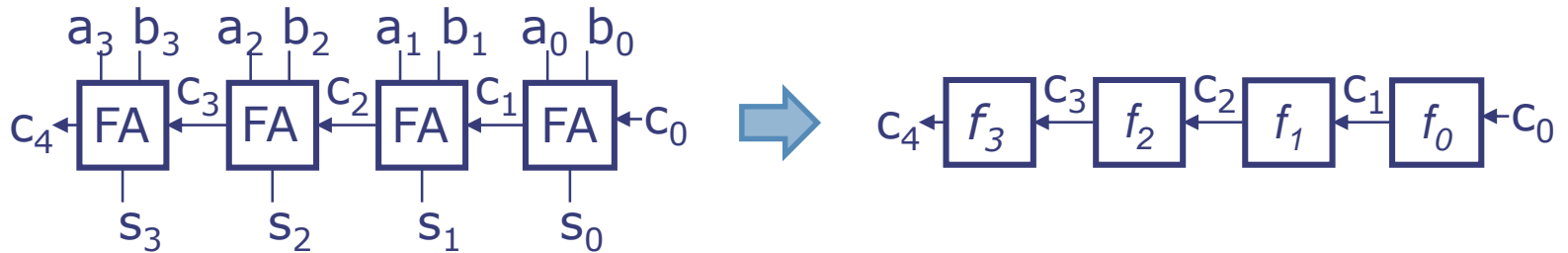


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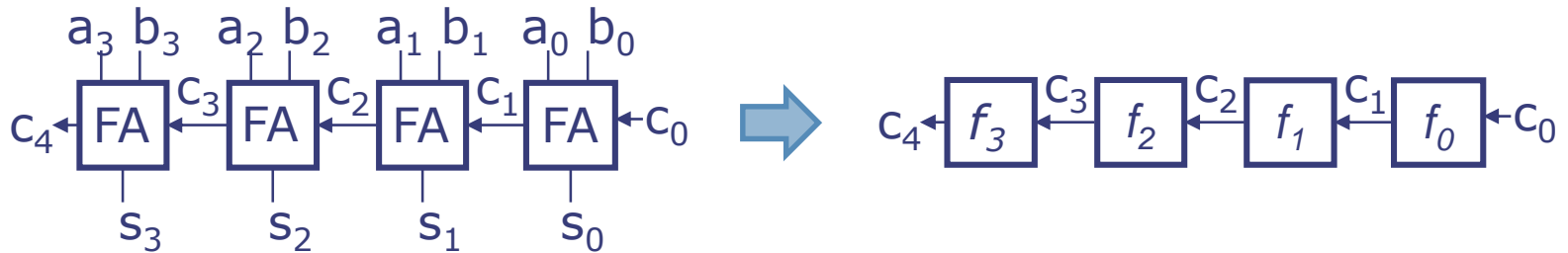


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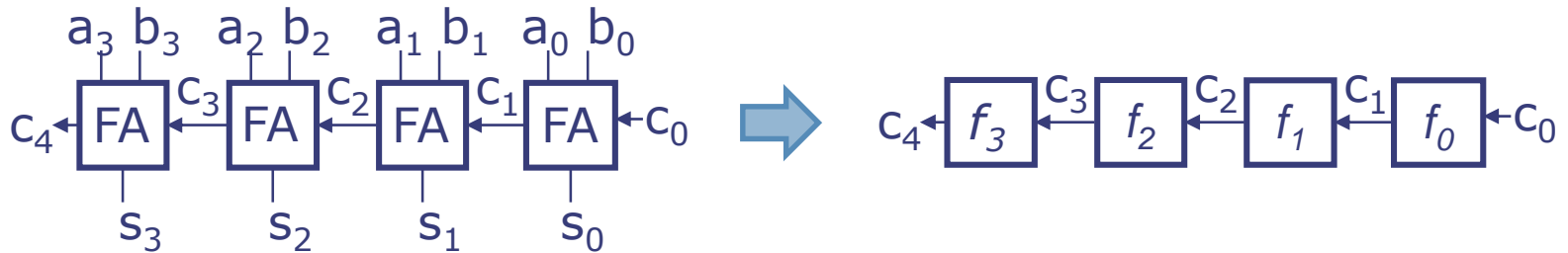


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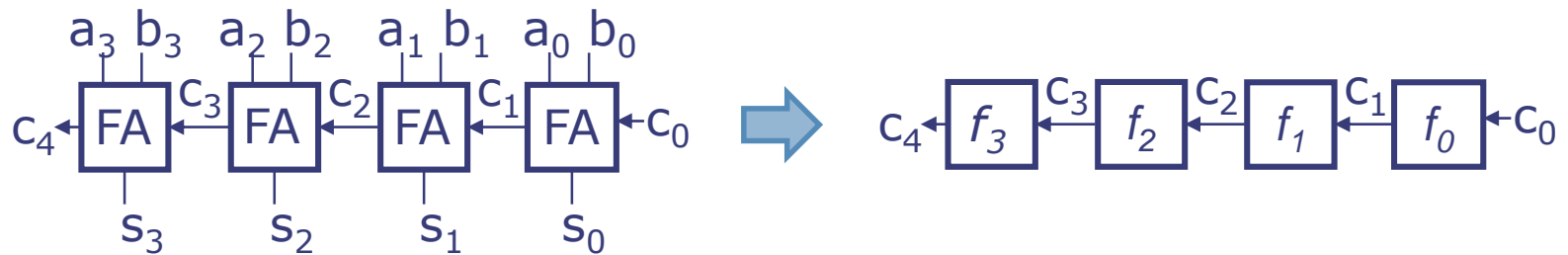
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Deriving the initial functions

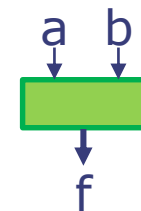
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kill	1	0
generate	0	1
generate	1	1
propagate	0	0
propagate	1	1
invert	0	1
invert	1	0

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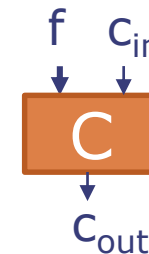
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invert not used in CLAs

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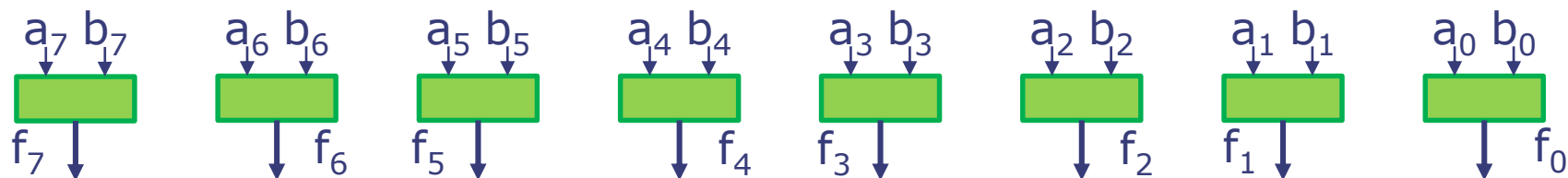
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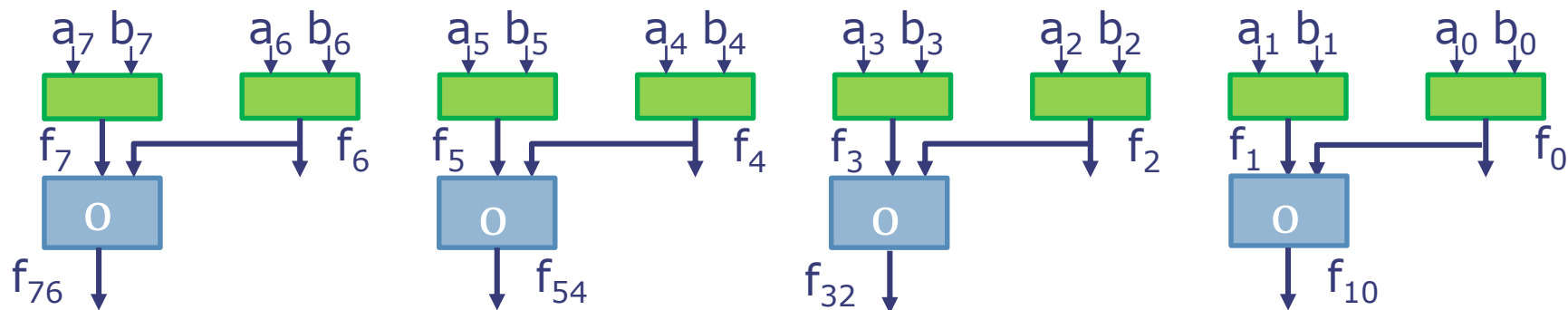
Step 1: Generating the output carry

Putting it all together



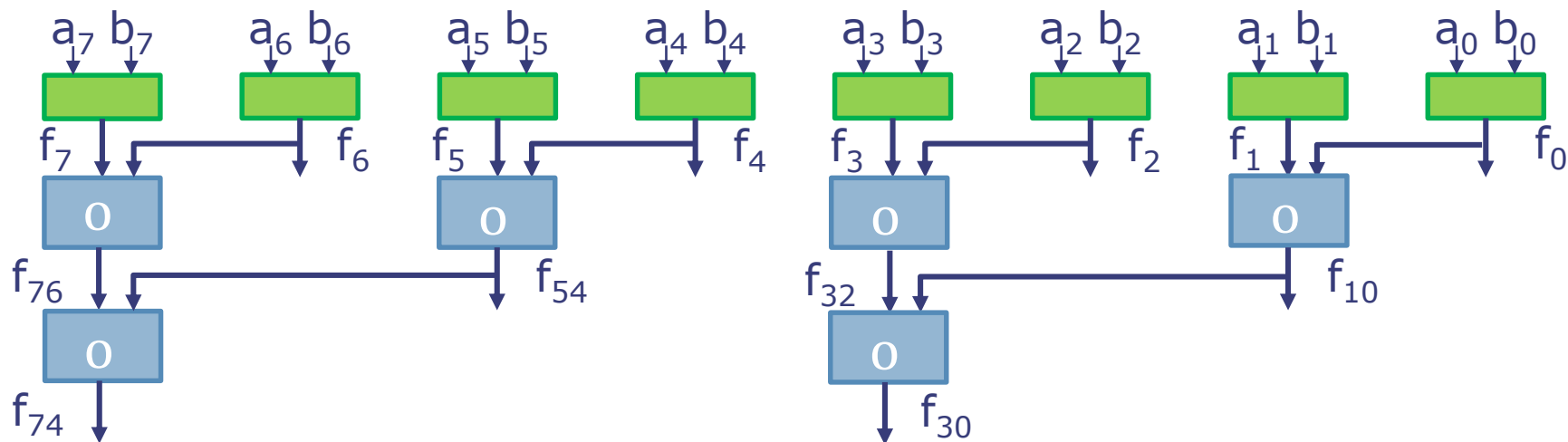
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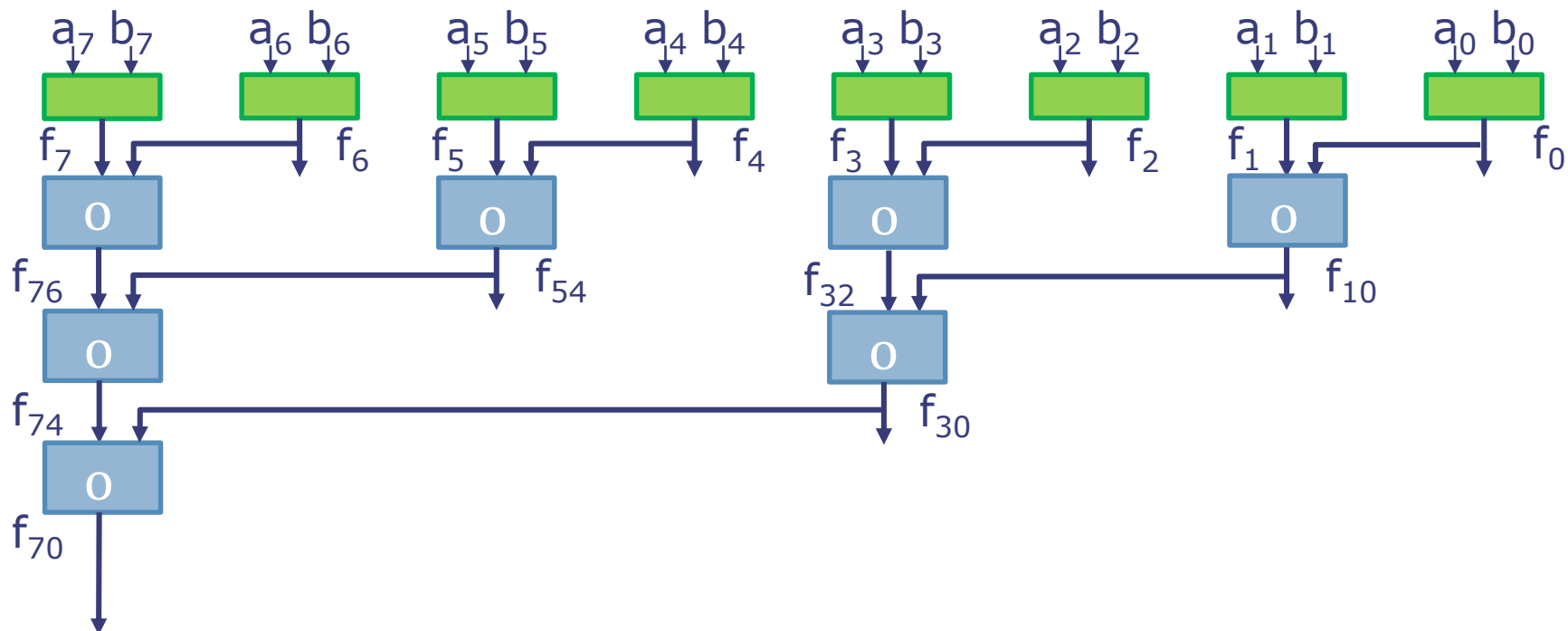
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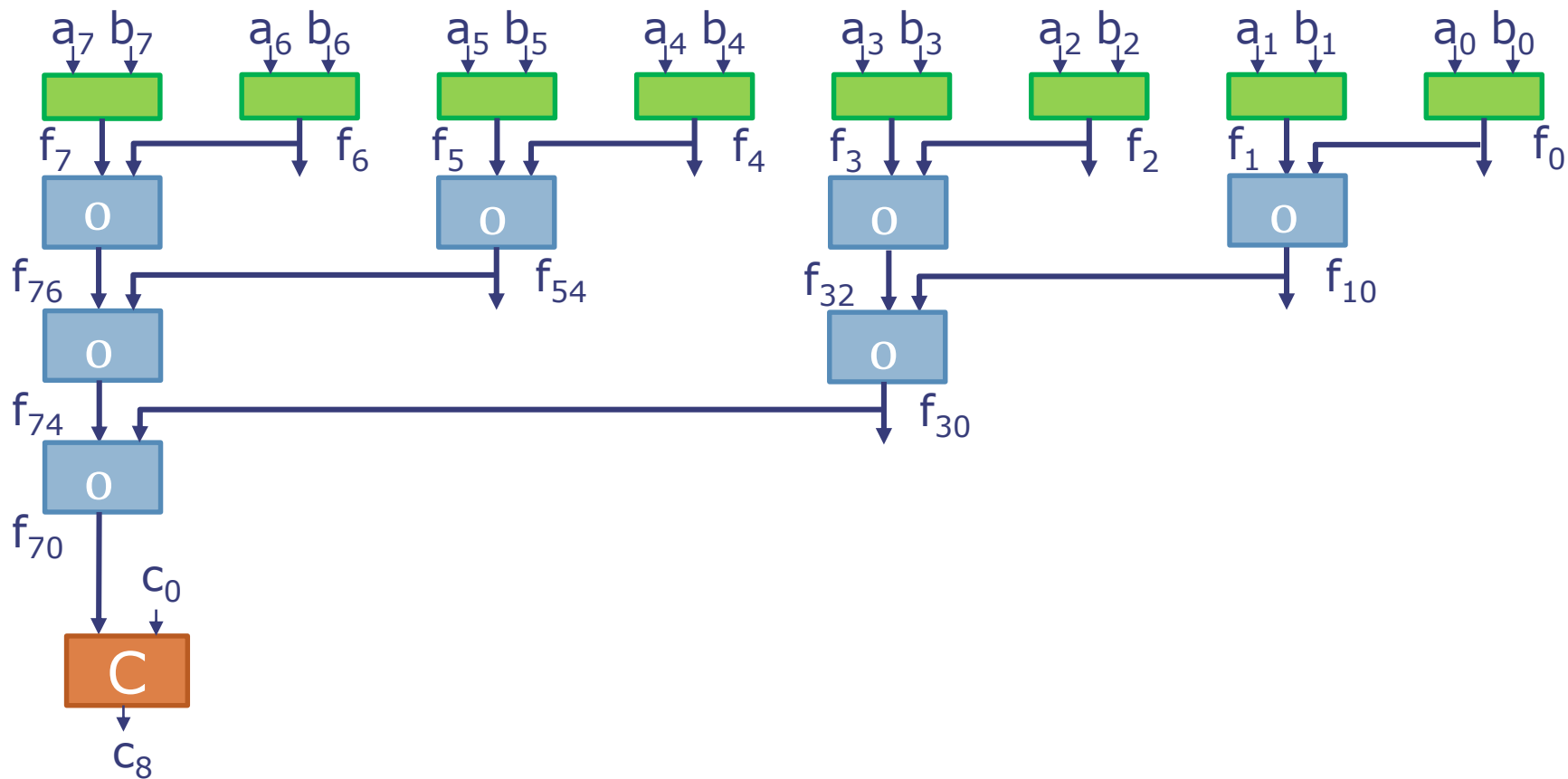
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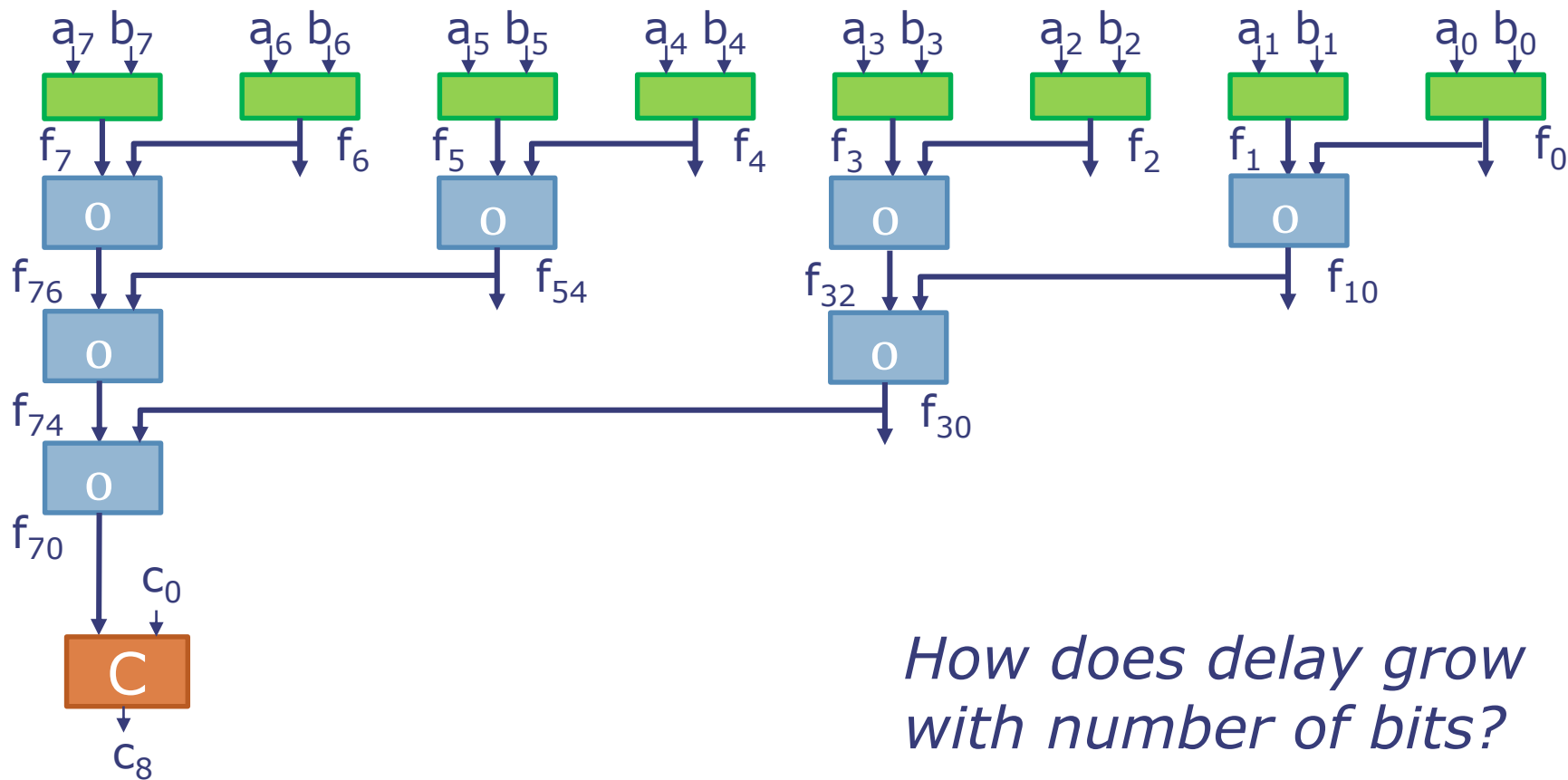
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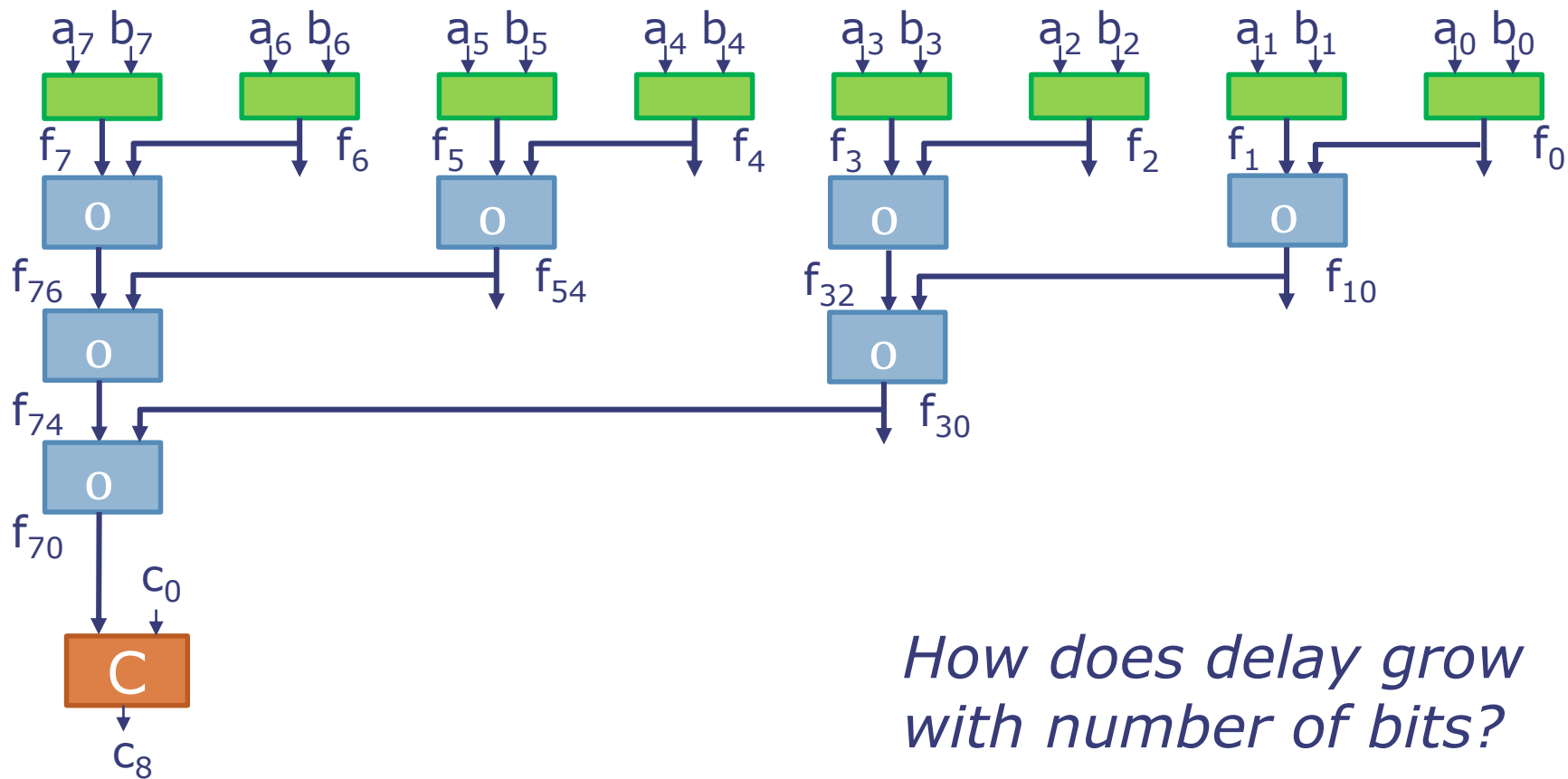
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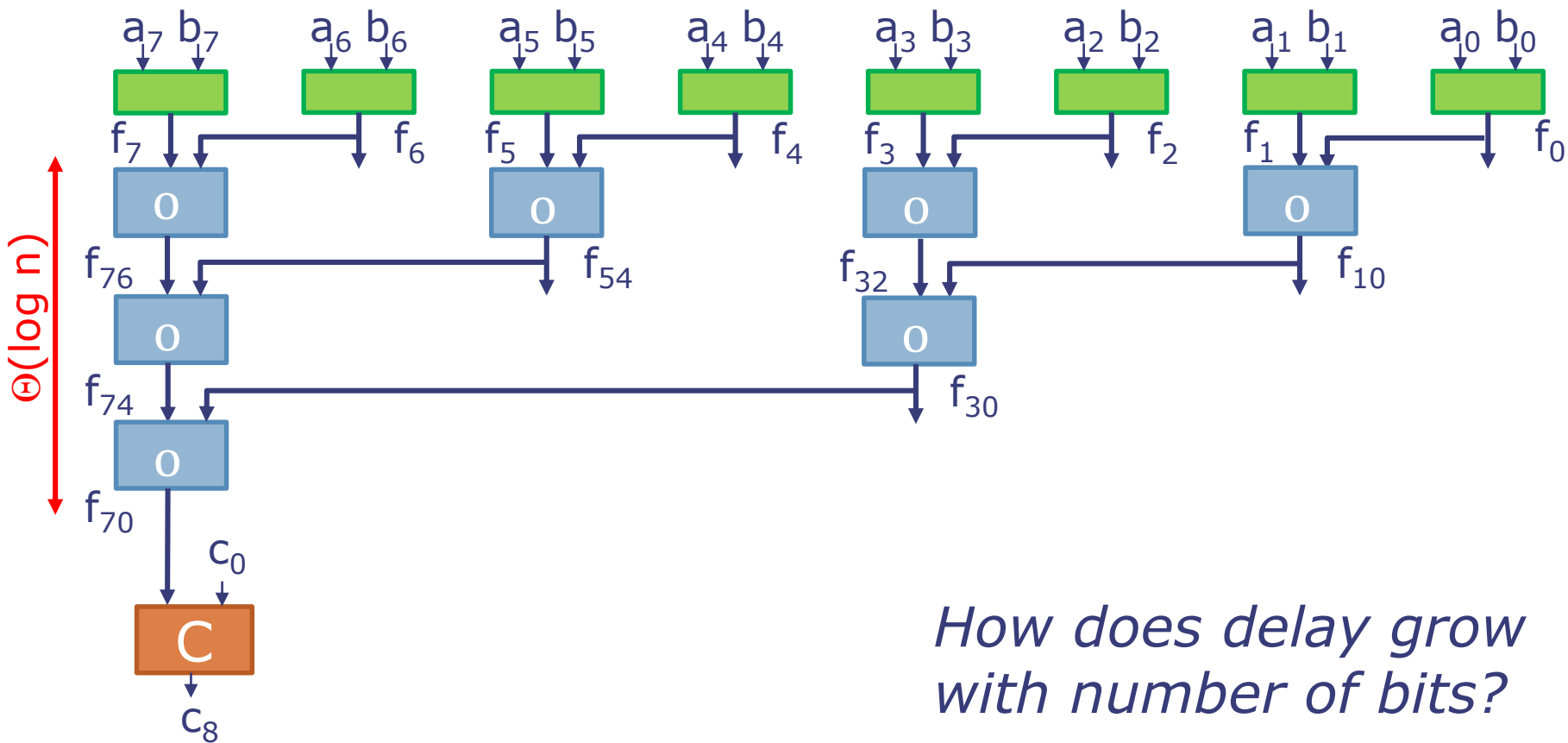


How does delay grow with number of bits?

$\Theta(\log n)$

Step 1: Generating the output carry

Putting it all together



$\Theta(\log n)$

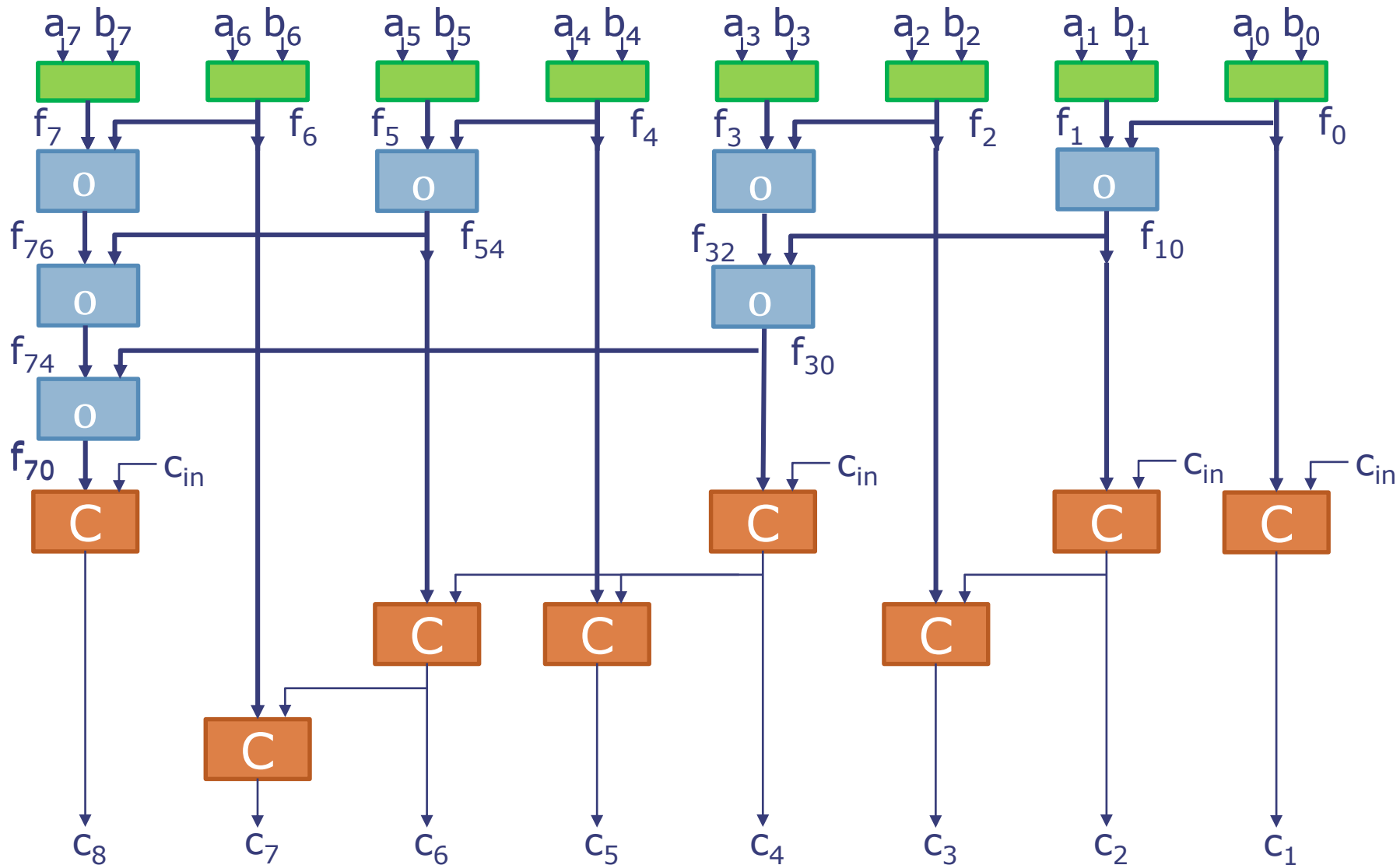
Step 2: Generating all carries

- So far we have seen how to generate a single output carry, but we need all intermediate ones too
- This is a specific application of the parallel scan (a.k.a. parallel prefix) algorithm

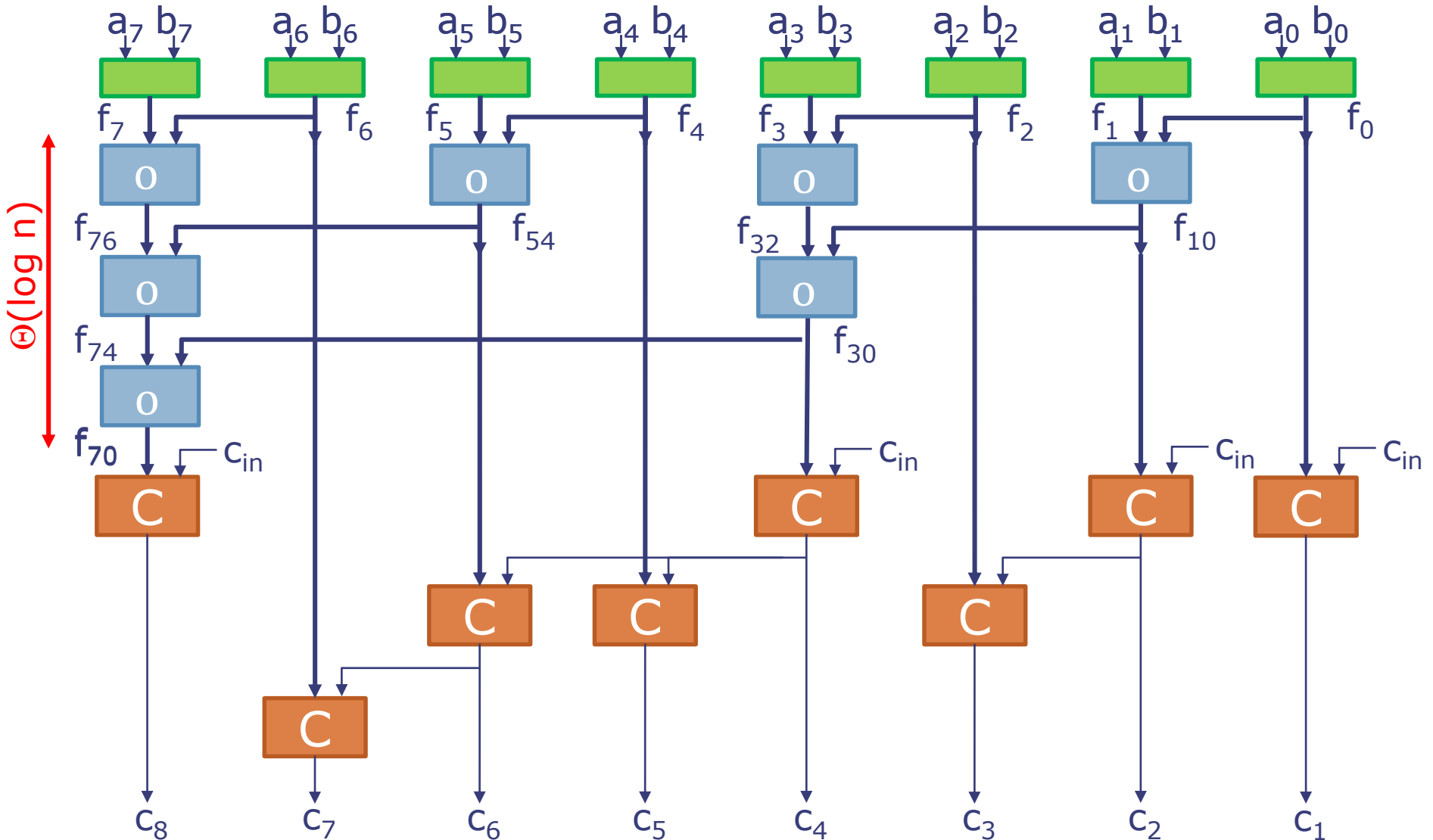
Step 2: Generating all carries

- So far we have seen how to generate a single output carry, but we need all intermediate ones too
- This is a specific application of the parallel scan (a.k.a. parallel prefix) algorithm
- Two main options:
 - Brent-Kung CLA: Low area but some extra delay
 - Kogge-Stone CLA: High area but lower delay

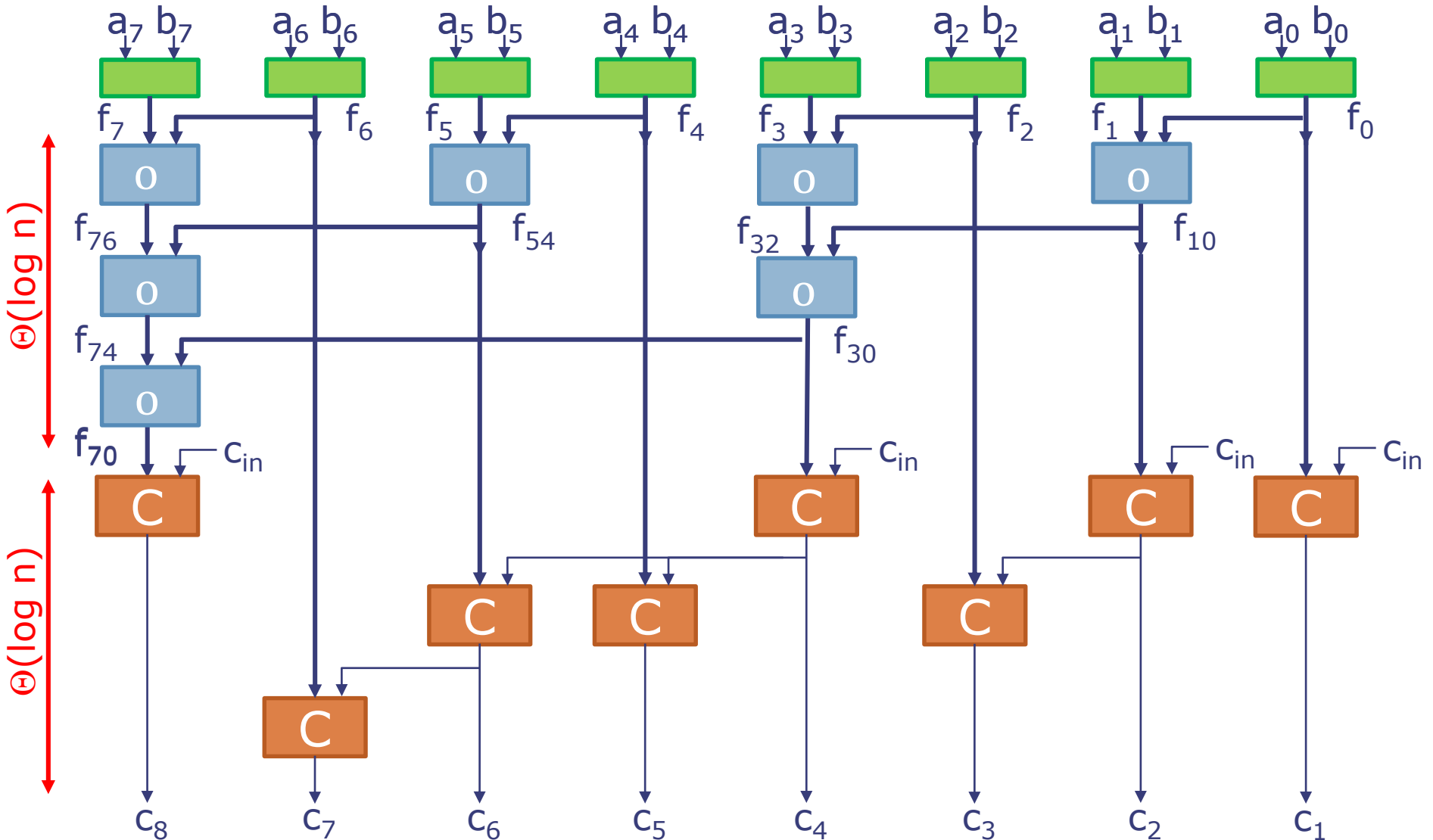
Option 1: Brent-Kung CLA



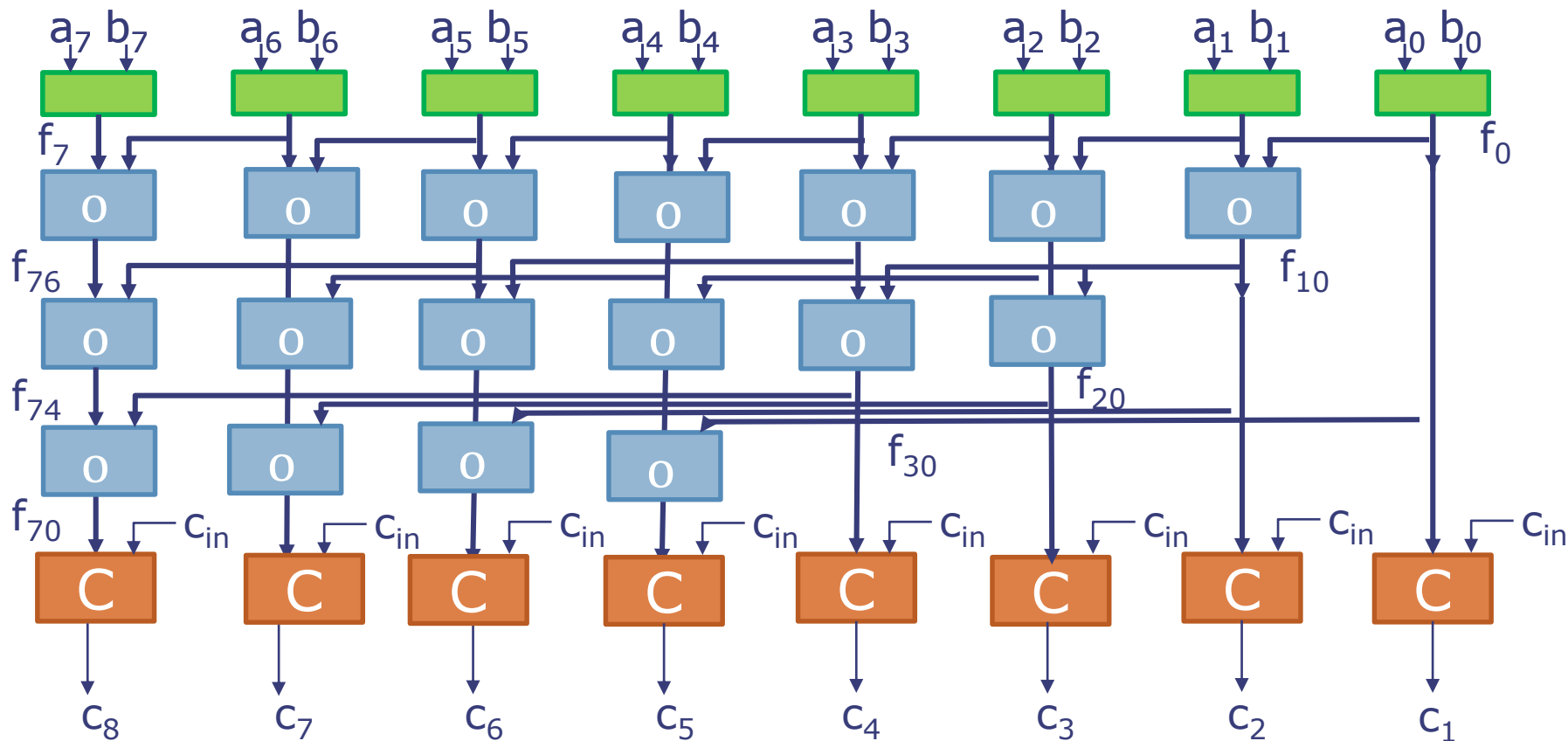
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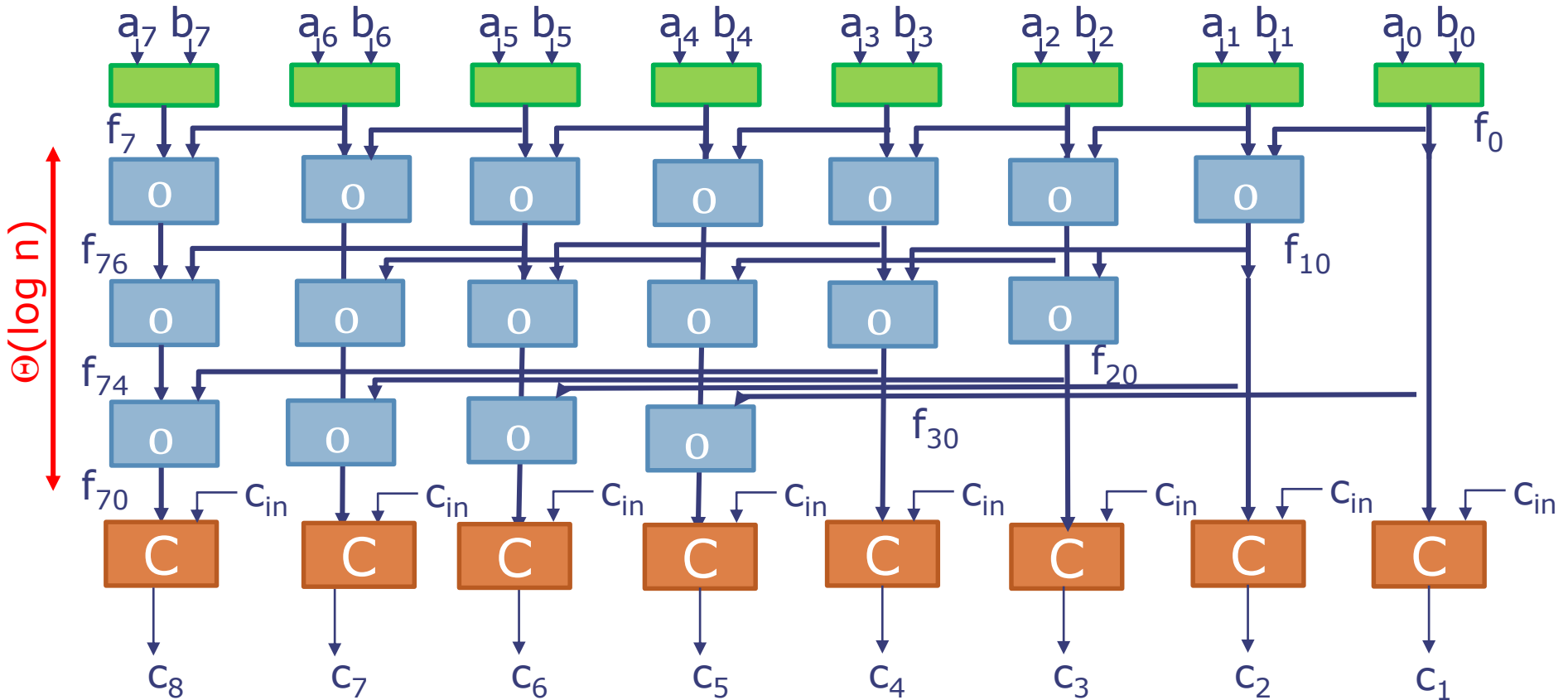
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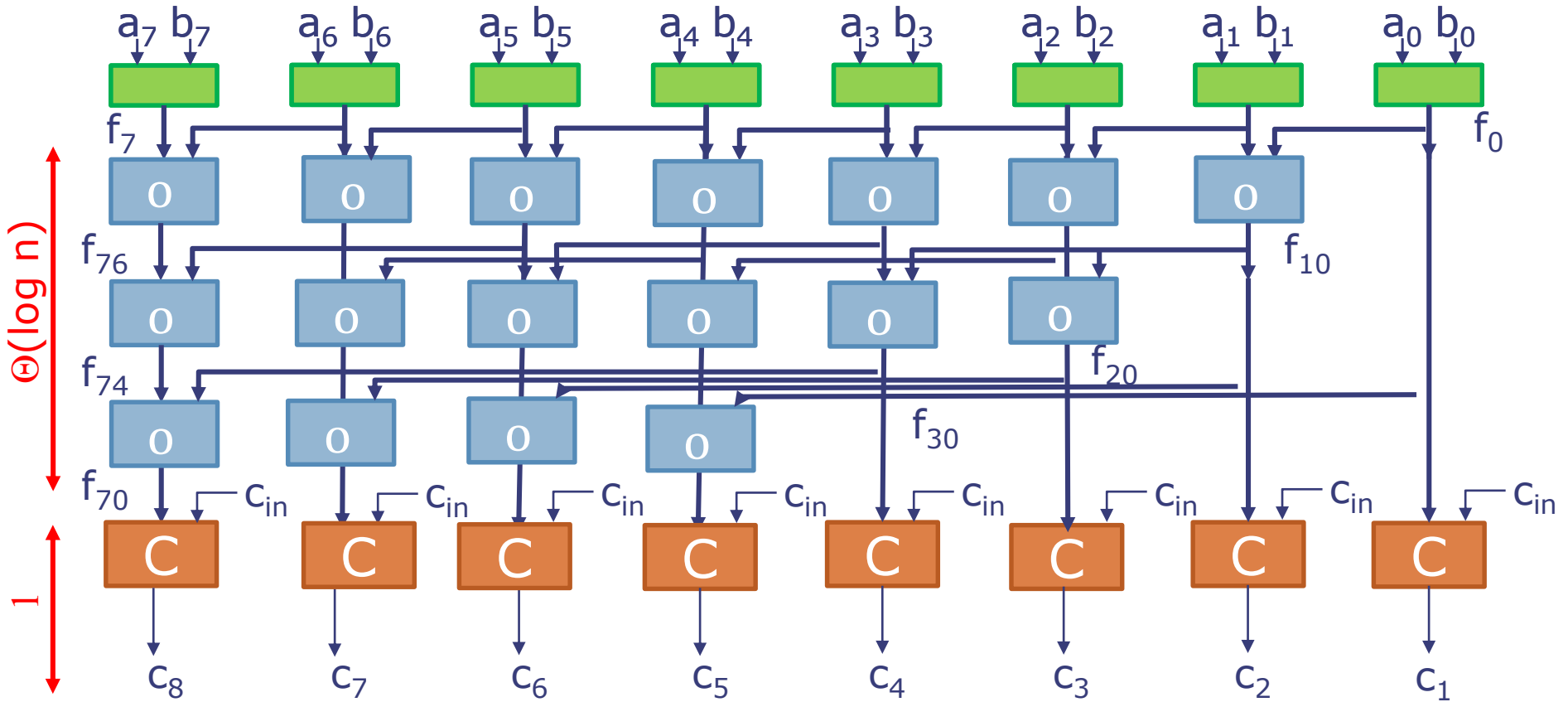
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CLA Nitty-Gritty: Choosing a good encoding for functions

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 - $g = ab$ (generate bit)
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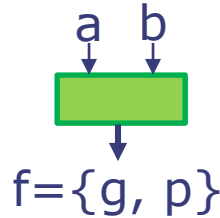
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- A common encoding is $f = \{g, p\}$, where:
 - $g = ab$ (generate bit)
 - $p = a+b$ (propagate bit)
- With this encoding,
 - $g = 0, p = 0 \rightarrow \text{kill}$
 - $g = 0, p = 1 \rightarrow \text{propagate}$
 - $g = 1, p = X \rightarrow \text{generate}$

CLA Building Blocks

with $f = \{g, p\}$ encoding

- Produce initial f signals



$$g = ab$$
$$p = a+b$$

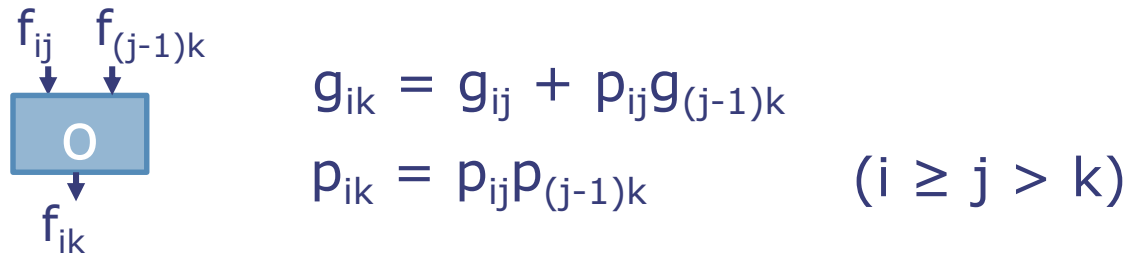
CLA Building Blocks

with $f = \{g, p\}$ encoding

- Produce initial f signals



- Compose f signals



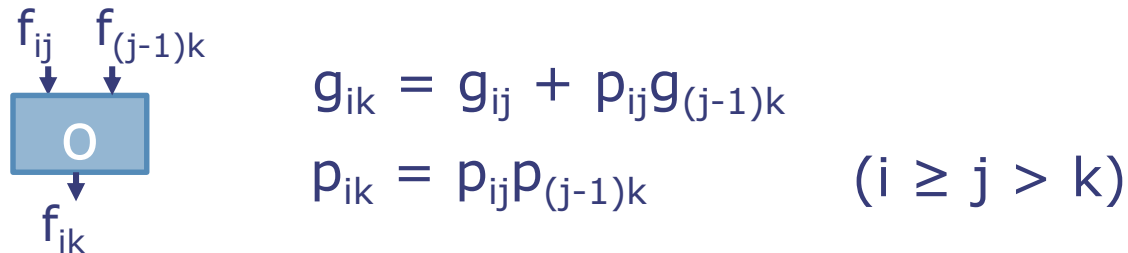
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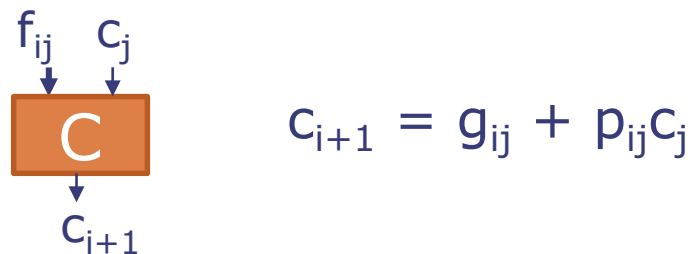
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- Compose f signals



- Produce individual carries



Carry-Lookahead Adder Takeaways

- There are many CLA designs
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- There are many CLA designs
 - We've seen Brent-Kung and Kogge-Stone CLAs
 - Some other types
 - Different variants for each type, e.g., using higher-radix trees to reduce depth
- This technique is useful beyond adders: computes any one-dimensional recurrence in $\Theta(\log n)$ delay
 - e.g., comparators, priority encoders, etc.

Summary

- Parametric functions let us write a generic description of a function that is then instantiated on demand
- Use for loops and if-else statements with care: their similarity to software can be confusing and they can lead to poor circuits

Summary

- Parametric functions let us write a generic description of a function that is then instantiated on demand
- Use for loops and if-else statements with care: their similarity to software can be confusing and they can lead to poor circuits
- Choosing the right algorithms is crucial to design good digital circuits—tools can only do so much!
- Carry-select and carry-lookahead adders achieve $\Theta(\log n)$ delay, but at the cost of extra area

Thank you!

Next lecture: Sequential Circuits