HW2

Chap 3

3.

Equations of motion in Laplace:

$$(s^{2} + 3s + 2)X_{1}(s) - (s + 2)X_{2}(s) - sX_{3}(s) = 0$$

$$-(s + 2)X_{1}(s) + (2s^{2} + 2s + 2)X_{2}(s) - sX_{3}(s) = F(s)$$

$$-sX_{1}(s) - sX_{2}(s) + (2s^{2} + 3s)X_{3}(s) = 0$$

Equations of motion in the time domain:

$$\frac{d^2x_1}{dt^2} + 3\frac{dx_1}{dt} + 2x_1 - \frac{dx_2}{dt} - 2x_2 - \frac{dx_3}{dt} = 0$$

$$-\frac{dx_1}{dt} - 2x_1 + 2\frac{d^2x_2}{dt^2} + 2\frac{dx_2}{dt} + 2x_2 - \frac{dx_3}{dt} = f(t)$$

$$-\frac{dx_1}{dt} - \frac{dx_2}{dt} + 2\frac{d^2x_3}{dt^2} + 3\frac{dx_3}{dt} = 0$$

Define state variables:

$$z_1 = x_1 \qquad \text{or} \qquad x_1 = z_1 \tag{1}$$

$$z_2 = \frac{dx_1}{dt} \quad \text{or} \quad \frac{dx_1}{dt} = z_2 \tag{2}$$

$$z_3 = x_2$$
 or $x_2 = z_3$ (3)

$$z_4 = \frac{dx_2}{dt} \quad \text{or} \quad \frac{dx_2}{dt} = z_4 \tag{4}$$

$$z_5 = x_3$$
 or $x_3 = z_5$ (5)

$$z_6 = \frac{dx_3}{dt} \quad \text{or} \quad \frac{dx_3}{dt} = z_6 \tag{6}$$

Substituting Eq. (1) in (2), (3) in (4), and (5) in (6), we obtain, respectively:

$$\frac{dz_1}{dt} = z_2 \tag{7}$$

$$\frac{dz_3}{dt} = z_4 \tag{8}$$

$$\frac{dz_5}{dt} = z_6 \tag{9}$$

Substituting Eqs. (1) through (6) into the equations of motion in the time domain and solving for the derivatives of the state variables and using Eqs. (7) through (9) yields the state equations:

$$\frac{dz_1}{dt} = z_2$$

$$\frac{dz_2}{dt} = -2z_1 - 3z_2 + 2z_3 + z_4 + z_6$$

$$\frac{dz_3}{dt} = z_4$$

$$\frac{dz_4}{dt} = z_1 + \frac{1}{2}z_2 - z_3 - z_4 + \frac{1}{2}z_6 + \frac{1}{2}f(t)$$

$$\frac{dz_5}{dt} = z_6$$

$$\frac{dz_6}{dt} = \frac{1}{2}z_2 + \frac{1}{2}z_4 - \frac{3}{2}z_6$$

The output is $x_1 = z_1$.

In vector-matrix form:

4T
$$\mathbf{Z} = \begin{bmatrix}
0 & 1 & 0 & 0 & 0 & 0 \\
-2 & -3 & 2 & 1 & 0 & 1 \\
0 & 0 & 0 & 1 & 0 & 0 \\
1 & 0.5 & -1 & -1 & 0 & 0.5 \\
0 & 0 & 0 & 0 & 0 & 1 \\
0 & 0.5 & 0 & 0.5 & 0 & -1.5
\end{bmatrix}
\mathbf{Z} + \begin{bmatrix}
0 \\
0 \\
0 \\
0.5 \\
0 \\
0
\end{bmatrix}
f(t)$$

$$\mathbf{y} = \begin{bmatrix}
1 & 0 & 0 & 0 & 0 & 0
\end{bmatrix} \mathbf{Z}$$

***請不要將
$$\dot{x} = [A]x + [B]u$$
寫成 $\dot{x} = [A]x + [Bu]$

5.

10T
$$\Theta_2$$
 Θ_3 50 D Kg-m² 100 N-m /rad 100 N-m sec/rad

Writing the equations of motion,

$$(1600s^2 + 100)\theta_2 - 100\theta_3 = 4T$$
$$-100\theta_2 + (50s^2 + 100s + 100)\theta_3 = 0$$

Taking the inverse Laplace transform and simplifying,

$$\theta_{2} + 0.0625\theta_{2} - 0.0625\theta_{3} = 0.0025T$$

$$-2\theta_{2} + \theta_{3} + 2\theta_{3} + 2\theta_{3} = 0$$

Defining the state variables as

$$x_1 = \theta_2, x_2 = \dot{\theta}_2, x_3 = \theta_3, x_4 = \dot{\theta}_3$$

Writing the state equations using the equations of motion and the definitions of the state variables

In vector-matrix form,

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -0.0625 & 0 & 0.0625 & 0 \\ 0 & 0 & 0 & 1 \\ 2 & 0 & -2 & -2 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 0.0025 \\ 0 \\ 0 \end{bmatrix} T$$

$$y = \begin{bmatrix} 4 & 0 & 0 & 0 \end{bmatrix} \mathbf{x}$$

13.(a)

a.
$$G(s)=C(sI-A)^{-1}B$$

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -1 & -3 & -2 \end{bmatrix}; B = \begin{bmatrix} 0 \\ 0 \\ 23 \end{bmatrix}; C = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$
$$(sI - A)^{-1} = \frac{1}{s^3 + 2s^2 + 3s + 1} \begin{bmatrix} s^2 + 2s + 3 & s + 2 & 1 \\ -1 & s(s + 2) & s \\ -s & -(3s + 1) & s^2 \end{bmatrix}$$

Therefore, $G(s) = \frac{23}{s^3 + 2s^2 + 3s + 1}$. Note that in this case the result could have also been obtained by inspection.

Since $T_m = J_{eq} \frac{d\omega_m}{dt} + D_{eq}\omega_m$, and $T_m = K_t \, i_a$,

$$J_{eq} \frac{d\omega_m}{dt} + D_{eq} \omega_m = K_t \; i_a \eqno(1)$$

Or,

$$\frac{d\omega_m}{dt} \quad = -\frac{D_{eq}}{J_{eq}} \quad \omega_m + \frac{K_t}{J_{eq}} \quad i_a \label{eq:delta_m}$$

$$But,\,\omega_m\ = \frac{N_2}{N_1} \quad \omega_L.$$

Substituting in (1) and simplifying yields the first state equation,

$$\frac{d\omega_L}{dt} \quad = \quad -\frac{D_{eq}}{J_{eq}} \quad \omega_L + \frac{K_t}{J_{eq}} \quad \frac{N_1}{N_2} \quad i_a \label{eq:delta_eps}$$

The second state equation is:

$$\frac{d\theta_L}{dt} \quad = \omega_L$$

Since

$$e_a = R_a i_a + L_a \frac{di_a}{dt} + K_b \omega_m = R_a i_a + L_a \frac{di_a}{dt} + K_b \frac{N_2}{N_1} + \omega_L,$$

the third state equation is found by solving for $\frac{di_a}{dt}$. Hence,

$$\frac{di_a}{dt} \quad = -\frac{K_b}{L_a} \; \frac{N_2}{N_1} \quad \omega_L - \frac{R_a}{L_a} \quad i_a + \frac{1}{L_a} \quad e_a \label{eq:diameter}$$

Thus the state variables are: $x_1 = \omega_L, \, x_2 = \theta_L$, and $x_3 = i_a$.

Finally, the output is $y = \theta_m = \frac{N_2}{N_1}$ θ_L .

In vector-matrix form,

$$\dot{\mathbf{x}} = \begin{bmatrix} -\frac{D_{eq}}{J_{eq}} & 0 & \frac{K_t}{J_{eq}} \frac{N_1}{N_2} \\ 1 & 0 & 0 \\ -\frac{K_b}{L_a} \frac{N_2}{N_1} & 0 & -\frac{R_a}{L_a} \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{L_a} \end{bmatrix} \mathbf{e}_a \; ; \; \mathbf{y} = \begin{bmatrix} 0 & \frac{N_2}{N_1} & 0 \end{bmatrix} \mathbf{x}$$

$$\mathbf{x} = \begin{bmatrix} \mathbf{\omega}_{L} \\ \mathbf{\theta}_{L} \\ \mathbf{i}_{a} \end{bmatrix}$$

Chap 4

13. (a)

$$\begin{split} &\omega_n{}^2 = 16 \text{ r/s}, \, 2\zeta\omega_n = 3. \text{ Therefore } \zeta = 0.375, \, \omega_n = 4. \, T_s = \frac{4}{\zeta\omega_n} \\ &= 2.667 \text{ s; } T_P = \frac{\pi}{\omega_n\sqrt{1-\zeta^2}} \\ &= 0.8472 \text{ s; } \% \text{OS} = \frac{1}{2} \\ &= -\zeta\pi \\ &= -$$

16.

$$\mathbf{a.}\ \zeta = \frac{-\ln{(\frac{\%\text{OS}}{100})}}{\sqrt{\pi^2 + \ln^2{(\frac{\%\text{OS}}{100})}}} = 0.517, \ \omega_n = \frac{4}{\zeta T_s} = 15.474. \ \text{Therefore, poles} = -\zeta \omega_n \ \pm j \omega_n \sqrt{1-\zeta^2}$$
$$= -8 \pm j 13.246.$$

18.

a. The impedance equations are:

$$(2s2 + s)\theta_1 - s\theta_2 = T$$
$$-s\theta_1 + (s+1)\theta_2 = 0$$

Solving for θ_2

$$\theta_2 = \frac{\begin{vmatrix} 2s^2 + s & T \\ -s & 0 \end{vmatrix}}{\begin{vmatrix} 2s^2 + s & -s \\ -s & s + 1 \end{vmatrix}} = \frac{T}{2s^2 + 2s + 1}$$

So

$$\frac{\theta_2(s)}{T(s)} = \frac{\frac{1}{2}}{s^2 + s + \frac{1}{2}}$$

b. $\omega_n = \frac{1}{\sqrt{2}}$, $2\xi\omega_n = 1$ or $\xi = \frac{1}{\sqrt{2}}$. $T_s = \frac{4}{\xi\omega_n} = 8$ sec. $T_p = \frac{\pi}{\omega_n\sqrt{1-\xi^2}} = 6.28 \text{sec}$ and $\%\text{OS} = \frac{1}{2}$

$$100e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}} = 4.32 \%.$$

a.
$$\frac{s+5}{s(s+1)(s^2+3s+10)} = \frac{0.5}{s} - \frac{0.5}{s+1} - \frac{1}{\sqrt{31}} \frac{\frac{\sqrt{31}}{2}}{(s+1.5)^2 + 7.75}$$

The amplitude of residue of the pole at -1 is larger than the amplitude of the sinusoid, so a pole-zero cancellation cannot be assumed.

b.
$$\frac{s+5}{s(s+2)(s^2+4s+15)} = \frac{0.167}{s} - \frac{0.136}{s+2} - 0.0303 \frac{(s+2)+3.32\sqrt{11}}{(s+2)^2+11}$$

Since the amplitudes of the sinusoids are of the same order of magnitude as the residue of the pole at -2, pole-zero cancellation cannot be assumed.

c.

$$C(s) = \frac{(s+5)}{s(s+4.5)(s^2+2s+20)} = \frac{0.055}{s} - \frac{0.0036}{s+4.5} - \frac{0.052(s+1)+0.3\sqrt{19}}{(s+1)^2+19}$$

Since the amplitudes of the sinusoids much larger than the residue of the pole at -4.5, a pole-zero

cancellation can be assumed. Since $2\zeta\omega_n$ = 2, and $\omega_n = \sqrt{20} = 4.47211$, $\zeta = 0.224$,

%OS =
$$e^{-\zeta\pi/\sqrt{1-\zeta^2}}$$
 x100 = 48.64%, $T_S = \frac{4}{\zeta\omega_n}$ = 4 sec, $T_p = \frac{\pi}{\omega_n\sqrt{1-\zeta^2}}$ = 0.72 sec; $\omega_n T_r = 1.23$,

therefore, $T_r = 0.275$ sec.

d.

$$C(s) = \frac{(s+5)}{s(s+4.9)(s^2+5s+20)} = \frac{0.051}{s} - \frac{0.0010}{s+4.9} - \frac{0.05(s+2.5)+0.702\sqrt{13.75}}{(s+2.5)^2+13.75}$$

Since the amplitude of the sinusoids are several orders of magnitude larger than the residue of the pole at -

4.9, a pole-zero cancellation can be assumed. Since $2\zeta\omega_n$ = 5, and ω_n = $\sqrt{20}$ = 4.47211, ζ = 00.56,

$$\% \text{OS} = e^{-\zeta \pi / \sqrt{1 - \zeta^2}} \times 100 = 12.03\% \text{ , } \mathsf{T_S} = \frac{4}{\zeta \omega_\mathsf{n}} \\ = 0.39 \text{ sec, } \mathsf{T_p} = \frac{\pi}{\omega_\mathsf{n} \sqrt{1 - \zeta^2}} \\ = 0.847 \text{ sec, } \mathsf{T_r} = 0.39 \text{ sec.}$$

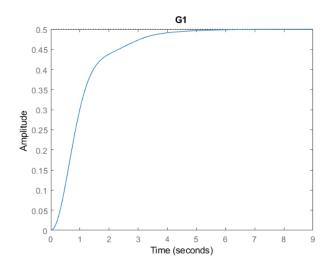
26.

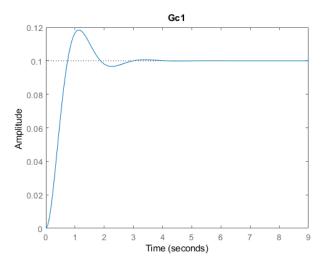
a.
$$|sI - A| = \begin{bmatrix} s & 0 & 0 \\ 0 & s & 0 \\ 0 & 0 & s \end{bmatrix} - \begin{bmatrix} 3 & 2 & 1 \\ 1 & 1 & 0 \\ 1 & 5 & 0 \end{bmatrix} = \begin{bmatrix} s - 3 & -2 & -1 \\ -1 & s - 1 & 0 \\ -1 & -5 & s \end{bmatrix} = s^3 - 4s^2 - 4$$

b. Solving $s^3 - 4s^2 - 4 = 0$ gives poles at 4.2242 and -0.1121±j0.967

第7題(4-22),滿多同學 c 小題寫錯,因此跑一次 matlab 給大家看結果,下次記得確實判斷再寫答案。 G 開頭為原始轉移函數, Gc 為極零點對消後的轉移函數,大家可以觀察一下前後差別。

7.a.





info_G1 = struct with fields:

RiseTime: 2.0048
SettlingTime: 3.8604
SettlingMin: 0.4509

SettlingMax: 0.5000

Overshoot: 0
Undershoot: 0

Peak: 0.5000

PeakTime: 10.5458

info Gc1 = struct with fields:

RiseTime: 0.5029

SettlingTime: 2.6100

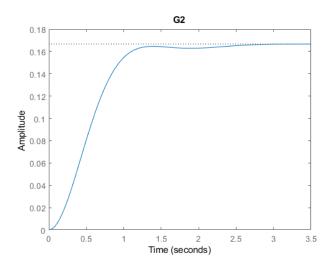
SettlingMin: 0.0924

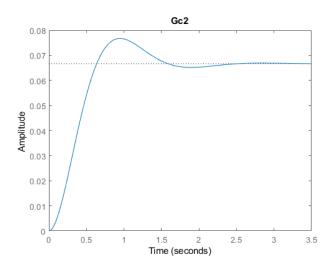
SettlingMax: 0.1184

Overshoot: 18.3964

Undershoot: 0

Peak: 0.1184
PeakTime: 1.1359





info_G2 = struct with fields:

RiseTime: 0.7459

SettlingTime: 2.1035

SettlingMin: 0.1505

SettlingMax: 0.1666

Overshoot: 0

Undershoot: 0

Peak: 0.1666

PeakTime: 3.2927

info_Gc2 = struct with fields:

RiseTime: 0.4316

SettlingTime: 2.0314

SettlingMin: 0.0617

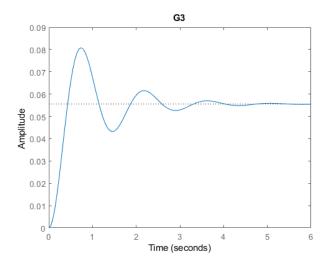
SettlingMax: 0.0767

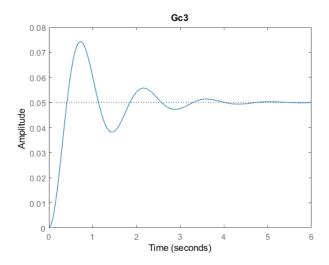
Overshoot: 15.0389

Undershoot: 0

Peak: 0.0767

PeakTime: 0.9441





info_G3 = struct with fields:

RiseTime: 0.2887

SettlingTime: 3.7770

SettlingMin: 0.0433

SettlingMax: 0.0808

Overshoot: 45.3994

Undershoot: 0

Peak: 0.0808

PeakTime: 0.7368

info_Gc3 = struct with fields:

RiseTime: 0.2767

SettlingTime: 3.7804

SettlingMin: 0.0382

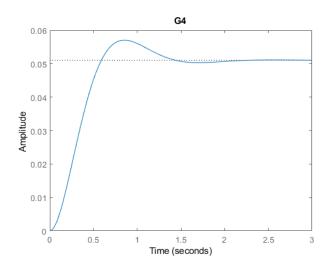
SettlingMax: 0.0743

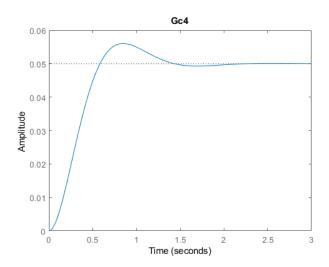
Overshoot: 48.5150

Undershoot: 0

Peak: 0.0743

PeakTime: 0.7368





info_G4 = struct with fields:

RiseTime: 0.3974

SettlingTime: 1.3117

SettlingMin: 0.0464

SettlingMax: 0.0570

Overshoot: 11.7478

Undershoot: 0

Peak: 0.0570

PeakTime: 0.8474

info_Gc4 = struct with fields:

RiseTime: 0.3936

SettlingTime: 1.3088

SettlingMin: 0.0459

SettlingMax: 0.0560

Overshoot: 12.0265

Undershoot: 0

Peak: 0.0560

PeakTime: 0.8474