

## GowinSynthesis

## **User Guide**

SUG550-1.6E, 06/30/2023

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## **Revision History**

Date	Version	Description
08/02/2019	1.0E	Initial version published.
12/09/2019	1.1E	"Naming Rules of Objects before/after Synthesis" added. (Gowin Software V1.9.3 and above)
03/03/2020	1.2E	VHDL syntax design supported. (Gowin Software V1.9.5 and above)
05/29/2020	1.3E	<ul><li> "Synthesis Naming Rules" modified.</li><li> "syn_srlstyle" and "syn_noprune" attribute added.</li></ul>
09/14/2020	1.4E	"black_box_pad_pin" attribute added.
10/28/2021	1.5E	<ul><li> "parallel_case" and "syn_black_box" attributes added;</li><li> Chapter 6 Report Document modified.</li></ul>
06/30/2023	1.6E	Chapter 5 Synthesis Constraints Support updated.

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1 About This Guide 1.1 Purpose

# 1 About This Guide

## 1.1 Purpose

It mainly describes the function and operation of GowinSynthesis and aims to help you learn this software and improve design efficiency. The software screenshots in this manual are based on Gowin Software V1.9.8.01. As the software is subject to change without notice, some information may not remain relevant and may need to be adjusted according to the software that is in use.

### 1.2 Related Documents

The user guides are available on the GOWINSEMI Website. You can find the related documents at <a href="www.gowinsemi.com">www.gowinsemi.com</a>: <a href="SUG100">SUG100</a>, <a href="Gowinsemi.com">Gowin</a></a> <a href="Software User Guide">Software User Guide</a>

## 1.3 Terminology and Abbreviation

Table 1-1 shows the abbreviations and terminology that are used in this guide.

Table 1-1 Abbreviations and Terminology

Terminology and Abbreviation	Meaning
BSRAM	Block Static Random Access Memory
DSP	Digital Signal Processing
FPGA	Field Programmable Gate Array
FSM	Finite State Machine
GSC	Gowin Synthesis Constraint File
SSRAM	Shadow Static Random Access Memory

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## 1.4 Support and Feedback

Gowin Semiconductor provides customers with comprehensive technical support. If you have any questions, comments, or suggestions, please feel free to contact us directly by the following ways.

Website: <a href="www.gowinsemi.com">www.gowinsemi.com</a>
E-mail: <a href="mailto:support@gowinsemi.com">support@gowinsemi.com</a>

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# **2** Overview

This is the user guide of Gowin RTL design synthesis tool GowinSynthesis.

GowinSynthesis is designed in-house by GOWINSEMI, and it uses Gowin original EDA algorithm to realize RTL design extraction, arithmetic optimization, inference, resource sharing, parallel synthesis and mapping based on product hardware characteristics and hardware circuit resources, which can quickly optimize your RTL design, resource check, and timing analysis.

GowinSynthesis, targeting Gowin FPGA chips, provides the most efficient design implementation method for FPGA designers. GowinSynthesis can generate a post-synthesis netlist based on the Gowin device primitive library, which can be used as an input file for the PnR, achieving the optimal balance of area and speed, improving software compilation efficiency, and routing rate. The software has the following features:

- Supports Verilog/SystemVerilog, VHDL and mixed design input.
- Supports ultra-large-scale design, providing an excellent synthesis solution for complex programmable logic designs.
- Supports inferred mapping of look-up tables, registers, latches, and arithmetic logic units.
- Supports memory inferred mapping and logic resources balancing.
- Supports DSP inferred mapping and logic resources balancing.
- Supports synthesis optimization of FSM
- Supports synthesis attributes and instructions to meet the synthesis requirements in different applications.

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# 3 GowinSynthesis Usage

## 3.1 Input and Output of GowinSynthesis

GowinSynthesis reads user's RTL file in the format of project file (.prj), and the project file is automatically generated by Gowin Software. In addition to the specified user RTL file, GowinSynthesis project file also specifies the synthesis device, the user constraints file including attribute constraints, post-synthesis netlists file (.vg), and some synthesis options, such as top module, files including paths, etc.

## 3.2 Use GowinSynthesis for Synthesis

Right-click "Synthesize" in the "Process" view of Gowin Software, select "Configuration". In the Configuration page, you can specify top module, set include path, and select language versions, and configure options.

Double-click "Synthesize" in "Process" view of Gowin Software to perform synthesis, and the "Output" view will output the synthesis information. The synthesis report and gate level netlist file will be generated after synthesis. Double-click the "Synthesis Report" and "Netlist File" in the "Process" view to check the specific contents.

For the further detailed operation, see <u>SUG100, Gowin Software User</u> <u>Guide</u> > 4.4.3 Synthesize.

## 3.3 Naming Rules of Objects Pre/Post Synthesis

For user-friendly verification and debugging, GowinSynthesis synthesis tool will reserve the original user RTL design info. in maximum, such as the module info., primitive/module instance name, the user-defined wire/reg name in user designs, etc. For the objects that must be optimized/converted and to be regenerated, the name will be created according to the user-defined wiring name and some derivative rules. The rules are described in the sections below.

## 3.3.1 Naming of the Post-synthesis Netlist File

The post-synthesis netlist file name depends on the specified output netlist file name in project file (\*.prj).

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If the post-synthesis netlist file name is not specified in project file, the name is created as the same as the project file with a .vg suffix.

### 3.3.2 Naming of the Post-synthesis Netlist Module

The post-synthesis netlist module name is consistent with RTL design name. Modules that are instantiated multiple times are distinguished by the suffix "\_idx", and the module's instantiation name is consistent with the RTL design.

### 3.3.3 Naming of the Post-synthesis Netlist Instance

If there is Instance in user RTL design and it is not optimized in the process of synthesis, the Instance name stays the same in the post-synthesis netlist.

For the Instance generated in the process of synthesis, the Instance name comes from the the external output signal name of the function design module in the user RTL. If the function design module has multiple output signal, the Instance name depends on the first output signal name.

For the Instance generated in the process of synthesis, the Instance name contains the names described above, and they are also followed by a suffix based on type. The "buf" types are suffixed with "\_ibuf", "\_obuf", "\_iobuf", and for others with "\_s", the number after "s" is the number of times the name is referenced by the node.

When flatten is specified to output, if the original submodule Instance name needs hierarchy, "/" can be used as hierarchical separator.

## 3.3.4 Naming of the Post-synthesis Netlist Wiring

For the user-defined wire/reg signal in RTL design, if the signal is not optiminzed in the process of synthesis, the related module name of the post-synthesis netlist stays the same.

In the process of synthesis using GowinSynthesis, some whole functional design modules in some RTL designs will be replaced or optimized. After synthesis, the output signal name of these functional design modules in netlist will be kept. For the internal signal of these modules, the name will be derived from the name of the output signal. The related digital suffix (\_idx) will be added on the basis of the original signal name.

When the multi-bit signal name (the bus format) is used as the derived signal name and other signal names or Instance name is derived, the bus bit in the singal name will be kept in the form of "\_idx".

When flatten is specified to output, "\_" can be used as hierarchical separator.

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# 4 HDL Code Support

## 4.1 Register HDL Code Support

## 4.1.1 An Introduction to Register Features

The registers contain flip-flops and latches.

#### Flip-flop

The flip-flops are all D flip-flops, and the initial value is assigned when defined. There are two types of reset/set: Synchronous and asynchronous. Synchronous reset/set means that only when the posedge or negedge of CLK arrives, and reset/set is at high level, can the reset/set be completed. Asynchronous reset/set means the level change from low to high of reset/set will lead to the output change immediately, but not controlled by CLK.

#### Latch

The latches trigger includes high-level trigger and low-level trigger, and the initial value is assigned when defined. In FPGA design, it is best to avoid latches. High level trigger is when the control signal is high; the latch allows the data signal to pass through. Low level trigger is when the control signal is low; latch allows the data signal to pass through.

## 4.1.2 Constraints Related with Register

You can constrain register by the preserve attribute. When this constraint exists, except the registers that are suspended will be optimized, all the other registers will be preserved in the synthesis results. Please see <a href="mailto:synthesis">synthesis</a> results.

## 4.1.3 Register Code Example

The initial value of Gowin synchronous reset clock flip-flop can only be set to 0. The initial value of synchronous set clock flip-flop can only be set to 1. When the initial value of the synchronous clock flip-flop in the RTL is different from the initial value of Gowin synchronous clock, GowinSynthesis will convert the type of synchronous clock flip-flop based on the initial value in the RTL. Asynchronous clock flip-flop does not be handled. Specific conversion strategies are:

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RTL is designed as synchronous reset clock flip-flop. When the initial value is set to 1, GowinSynthesis will replace it with synchronous set clock flip-flop. Add related logic to the original synchronous reset signal to realize synchronous set function.

RTL is designed as synchronous set clock flip-flop. When the initial value is set to 0, GowinSynthesis will replace it with normal flip-flop. Add related logic to the original synchronous set signal as the input of the flip-flop data end.

#### Not specify the Initial Value of Flip-flop

The only difference between CLK posedge flip-flop and CLK negedge flip-flop is that CLK triggers in different ways, so the following list is only the examples of CLK rising edge flip-flop.

```
Example 1 can be synthesized as synchronous reset clock flip-flop.
```

```
module top (q, d, clk, reset).

input d.

input clk.

input reset.

output q.

reg q_reg.

always @(posedge clk)begin

if(reset)

q_reg<=1'b0.

else

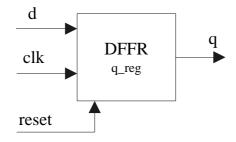
q_reg<=d.

end

assign q = q_reg.
```

Synchronous reset clock flip-flop is as shown in Figure 4-1.

Figure 4-1 Synchronous Reset Clock Flip-flop in Example 1



endmodule

Example 2 can be synthesized as synchronous set flip-flop with clock enable.

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```
module top (q, d, clk, ce, set).

input d.

input clk.

input set.

output q.

reg q_reg.

always @(posedge clk)begin

if(set)

q_reg<=1'b1.

else if(ce)

q_reg<=d.

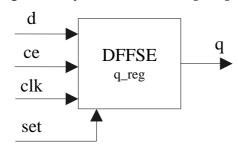
end

assign q = q_reg.

endmodule
```

Synchronous set flip-flop with clock enable is as shown in Figure 4-2.

Figure 4-2 Synchronous Set Flip-flop with Clock Enable in Example 2



Example 3 can be synthesized as asynchronous reset flip-flop with clock enable.

```
module top (q, d, clk, ce, clear).

input d.

input clk.

input ce.

input clear.

output q.

reg q_reg.

always @(posedge clk or posedge clear)begin

if(clear)

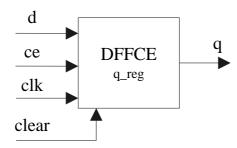
q_reg<=1'b0.
```

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```
else if(ce)
q_reg<=d.
end
assign q = q_reg.
endmodule
```

Asynchronous reset flip-flop with clock enable is as shown in Figure 4-3.

Figure 4-3 Asynchronous Reset Flip-flop with Clock Enable in Example 3



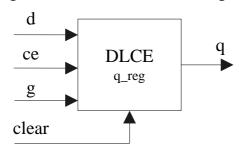
Example 4 can be synthesized as latch with reset and high level enable.

```
module top(d,g,clear,q,ce);
input d,g,clear,ce;
output q;
reg q_reg;
always @(g or d or clear or ce) begin
  if(clear)
        q_reg <= 0;
    else if(g && ce)
        q_reg <= d;
end
assign q = q_reg;
endmodule</pre>
```

The latch with reset and high level enable is shown as in Figure 4-4.

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Figure 4-4 Latch with Reset and High Level Enable in Example 4



#### Specify the Initial Value of Flip-flop

Example 5 is synchronous reset clock flip-flop with an initial value of 0. Set an initial value of 1 in RTL, which will be synthesized as synchronous set clock flip-flop with an initial value of 1 and a logic circuit for synchronous reset.

```
module top (q, d, clk, reset).

input d.

input clk.

input reset.

output q.

reg q_reg = 1'b1.

always @(posedge clk)begin

if(reset)

q_reg<=1'b0.

else

q_reg<=d.

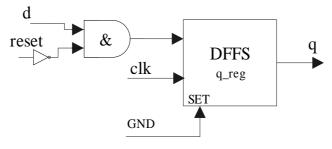
end

assign q = q_reg.

endmodule
```

Synchronous reset clock flip-flop above is as shown in Figure 4-5.

Figure 4-5 Synchronous Reset Clock Flip-flop and Logic Circuit in Example 5  $\,$ 



Example 6 is synchronous set clock flip-flop with an initial value of 1, but its initial value is 0 in RTL; and this synchronous set clock flip-flop will

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be synthesized as common clock flip-flop with an initial value of 0 and a logic circuit for synchronous set.

```
module top (q, d, clk, set).

input d.

input clk.

input set.

output q.

reg q_reg = 1'b0.

always @(posedge clk)begin

if(set)

q_reg<=1'b1.

else

q_reg<=d.

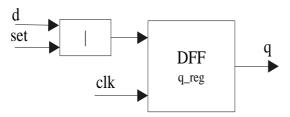
end

assign q = q_reg.

endmodule
```

The common clock flip-flop with the initial value of 0 and logic circuit are shown in Figure 4-6.

Figure 4-6 Common Clock Flip-flop with an Initial Value of 0 and Logic Circuit in Example 6



Example 7 is an asynchronous set flip-flop with an initial value of 1. module top (q, d, clk, ce, preset).

input d.

input clk.

input ce.

input preset.

output q.

 $reg q_reg = 1'b1.$ 

always @(posedge clk or posedge preset)begin

if(preset)

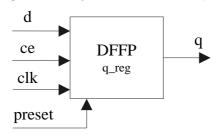
q\_reg<=1'b1.

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```
else if(ce)
q_reg<=d.
end
assign q = q_reg.
endmodule
```

Asynchronous set flip-flop above is as shown in Figure 4-7.

Figure 4-7 Asynchronous Set Flip-flop in Example 7



## 4.2 RAM HDL Code Support

#### 4.2.1 An Introduction to RAM Inference Function

RAM inference is one step in RTL synthesis to infer block memory primitives (BSRAM and SSRAM) in FPGA to implement memory functions in user design so that you can write device-independent RTL or use embedded block ram functionality in FPGA. For RTL memory blocks, GowinSynthesis will infer the RTL that meets the corresponding conditions to RAM module according to RTL description.

If the design needs to implement by BSRAM, the following principles need to be met:

- 1. All output registers have the same control signal.
- 2. RAM must be synchronous memory and can not connect to asynchronous control signal. GowinSynthesis does not support asynchronous RAM.
- 3. It needs to connect registers at read address or output port.

#### 4.2.2 An Introduction to RAM Features

#### **BSRAM**

There are four configuration modes for BSRAM: single-port, dual-port, semi-dual-port and read-only. Read mode includes pipeline and bypass. Write mode includes normal, write-through and read-before-write.

#### **SSRAM**

There are three configuration modes for SSRAM: Single-port, semi-dual-port and read-only. SSRAM does not support dual-port mode.

#### 4.2.3 Constraints Related with RAM Inference

Syn\_ramstyle specifies how memory is inferenced, and syn\_romstyle specifies how read-only memory is implemented.

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If the design needs to generate SSRAM or BSRAM, please use ram\_style, rom\_style or syn\_srlstyle constraint statement.

For constraint syntax use, please see syn\_ramstyle\_and syn\_romstyle.

## 4.2.4 RAM Inference Code Example

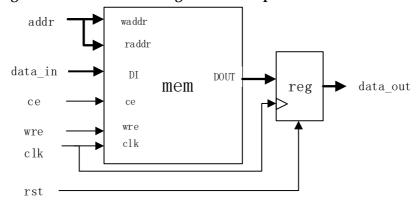
According to the different features of RAM, examples are as follows:

Example 1 is a memory with one write port, one read port and the same read and write address, which can be synthesized to a single port BSRAM in normal mode.

```
module normal(data_out, data_in, addr, clk,ce, wre,rst).
output [7:0]data_out.
input [7:0]data in.
input [7:0]addr.
input clk,wre,ce,rst.
reg [7:0] mem [255:0].
reg [7:0] data_out.
always@(posedge clk or posedge rst)
if(rst)
   data out <= 0.
else
     if(ce & !wre)
         data out <= mem[addr].
 always @(posedge clk)
    if (ce & wre)
         mem[addr] <= data in.
endmodule
```

The above single-port BSRAM circuit diagram is shown in Figure 4-8.

Figure 4-8 RAM Circuit Diagram in Example 1



Example 2 is a memory with one write port, one read port and the

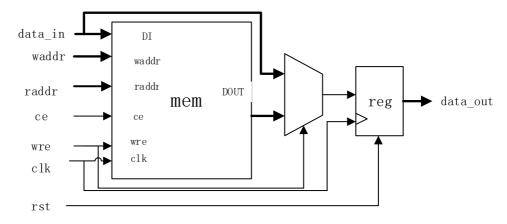
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same read and write address. When wre is 1, input data can be transferred directly to output, which can be synthesized to single-port BSRAM in normal write mode.

```
module wt11(data out, data in, addr, clk, wre,rst);
output [31:0]data_out;
input [31:0]data_in;
input [6:0]addr;
input clk,wre,rst;
reg [31:0] mem [127:0];
reg [31:0] data_out;
always@(posedge clk or posedge rst)
if(rst)
    data out <= 0;
else if(wre)
     data out <= data in;
else
     data_out <= mem[addr];</pre>
always @(posedge clk)
if (wre)
     mem[addr] <= data in;
endmodule
```

The above single-port BSRAM circuit diagram is shown in Figure 4-9.

Figure 4-9 RAM Circuit Diagram in Example 2



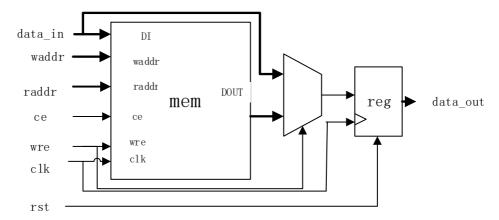
Example 3 is a memory with one write port, one read port and the same read and write address. When wre is 1, input data is written to memory, which can be synthesized to single-port BSRAM in read-before-write mode.

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```
module read_first_01(data_out, data_in, addr, clk, wre);
output [31:0]data_out;
input [31:0]data_in;
input [6:0]addr;
input clk,wre;
reg [31:0] mem [127:0];
reg [31:0] data_out;
always @(posedge clk)
begin
    if (wre)
        mem[addr] <= data_in;
        data_out <= mem[addr];
end
endmodule
```

The above single-port BSRAM circuit diagram is shown in Figure 4-10.

Figure 4-10 RAM Circuit Diagram in Example 3



Example 4 is a memory with two write ports and one read port. One of the two write ports has a wre signal and the other does not. The read port absorbs asynchronous reset register. This example can be synthesized to asynchronous reset dual-port BSRAM with A port in normal write mode and B port in read-before-write mode or in register output read mode.

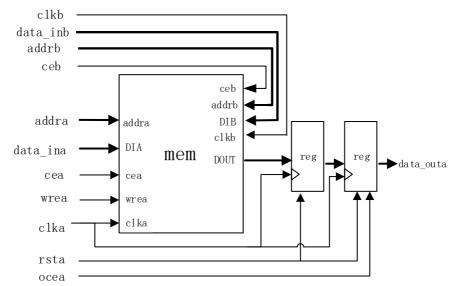
```
module read_first_02_1(data_outa, data_ina, addra, clka, rsta,cea, wrea,ocea, data_inb, addrb, clkb, ceb);
output [17:0]data_outa;
input [17:0]data_ina,data_inb;
input [6:0]addra,addrb;
input clka, rsta,cea, wrea,ocea;
input clkb, ceb;
```

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```
reg [17:0] mem [127:0];
reg [17:0] data_outa;
reg [17:0] data_out_rega,data_out_regb;
always @(posedge clkb)
if (ceb)
   mem[addrb] <= data_inb;
always@(posedge clka or posedge rsta)
if(rsta)
    data_out_rega <= 0;
else begin
    data_out_rega <= mem[addra];
end
always@(posedge clka or posedge rsta)
if(rsta)
    data_outa <= 0;
else if (ocea)
    data_outa <= data_out_rega;
always @(posedge clka)
if (cea & wrea)
   mem[addra] <= data_ina;
endmodule
```

The above dual-port BSRAM circuit diagram is shown in Figure 4-11.

Figure 4-11 RAM Circuit Diagram in Example 4



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Example 5 is a memory with one read port and one write port and different read and write addresses, which can be synthesized to semi-dual-port BSRAM in normal write mode or in bypass read mode.

```
module read_first_wp_pre_1(data_out, data_in, waddr, raddr,clk, rst,ce);
output [10:0]data_out;
input [10:0]data_in;
```

```
input [6:0]raddr,waddr;
input clk, rst,ce;
```

reg [10:0] mem [127:0];

reg [10:0] data out;

always@(posedge clk or posedge rst)

if(rst)

```
data_out <= 0;
```

else if(ce)

data out <= mem[raddr];</pre>

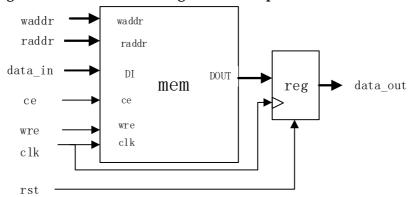
always @(posedge clk)

if (ce) mem[waddr] <= data\_in;</pre>

endmodule

The above semi-dual-port BSRAM circuit diagram is shown in Figure 4-12.

Figure 4-12 RAM Circuit Diagram in Example 5



Example 6 is a memory with one read port and an initial value, which can be synthesized to asynchronous set read-only memory in bypass read mode.

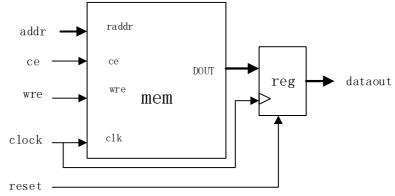
```
module test_invce (clock,ce,oce,reset,addr,dataout);
input clock,ce,oce,reset;
input [5:0] addr;
output [7:0] dataout;
```

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```
reg [7:0] dataout;
always @(posedge clock or posedge reset)
if(reset) begin
   dataout <= 0;
end else begin
if (ce & oce) begin
case (addr)
6'b000000: dataout <= 32'h87654321;
6'b000001: dataout <= 32'h18765432;
6'b000010: dataout <= 32'h21876543;
6'b111110: dataout <= 32'hdef89aba;
6'b111111: dataout <= 32'hef89abce:
default: dataout <= 32'hf89abcde:
endcase
end
end
endmodule
```

The above read-only memory circuit diagram is shown in Figure 4-13.

Figure 4-13 pROM Circuit Diagram in Example 6



Example 7 is a memory with shift-register mode, which can be synthesized to simple-dual-port BSRAM in normal mode.

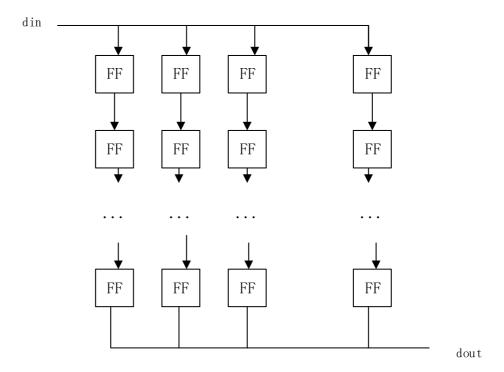
```
module seqshift_bsram (clk, din, dout);
parameter SRL_WIDTH = 65;
parameter SRL_DEPTH = 16;
input clk;
input [SRL_WIDTH-1:0] din;
```

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```
output [SRL_WIDTH-1:0] dout;
reg [SRL_WIDTH-1:0] regBank[SRL_DEPTH-1:0];
integer i;
always @(posedge clk) begin
    for (i=SRL_DEPTH-1; i>0; i=i-1) begin
        regBank[i] <= regBank[i-1];
    end
    regBank[0] <= din;
end
assign dout = regBank[SRL_DEPTH-1];
endmodule</pre>
```

The above semi-dual-port BSRAM circuit diagram is shown in Figure 4-14.

Figure 4-14 RAM Circuit Diagram in Example 7



#### Note!

For more examples, please see <u>GowinSynthesis Inference Coding Template</u> at Gowinsemi official website.

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## 4.3 DSP HDL Code Support

#### 4.3.1 Basic Introduction to DSP Inference

DSP inference is an algorithm that infers and permutes multiplication and partial addition in user design to DSP in RTL synthesis. When designing RTL, you can either instantiate DSP or write DSP description in device-independent RTL. For the multiplication and addition module of RTL, GowinSynthesis will permute RTL description meeting corresponding conditions with corresponding DSP module.

DSP module has functions of multiplication, addition and register. GowinSynthesis uses logic circuits to realize multiplier functions when the current device does not support DSP modules.

#### 4.3.2 Introduction to DSP Features

Gowin DSP includes multiplier, multiply add accumulator and preadder. The following functions are supported:

- 1. Supports multiplication permutation of different sign bits input
- 2. Supports synchronous or asynchronous mode
- 3. Supports multiplication chain addition
- 4. Supports multiplication accumulation
- 5. Supports pre-add function
- 6. Supports register absorption, including input register, output register, bypass register

#### 4.3.3 Constraints Related with DSP

Syn\_dspstyle is used to control the multipliers or specific objects using DSP or logic circuits.

Syn\_perserve is used to reserve registers. When register around the DSP has this property, the DSP cannot absorb this register.

For the constraint statements, please see syn\_dspstyle and syn\_preserve.

## 4.3.4 DSP Inference Code Example

Example 1 can be synthesized to synchronous set multiplier with sign bit. The input registers are ina and inb. The output register is out\_reg, and the bypass register is pp\_reg.

```
module top(a,b,c,clock,reset,ce).

parameter a_width = 18.

parameter b_width = 18.

parameter c_width = 36.

input signed [a_width-1:0] a.

input signed [b_width-1:0] b.
```

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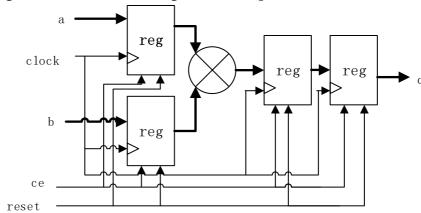
```
input clock.
input reset.
input ce.
output signed [c_width-1:0] c.
reg signed [a width-1:0] ina.
reg signed [b_width-1:0] inb.
reg signed [c_width-1:0] pp_reg.
reg signed [c_width-1:0] out_reg.
wire signed [c_width-1:0] mult_out.
always @(posedge clock) begin
   if(reset)begin
       ina<=0.
       inb<=0.
   end else begin
       if(ce)begin
           ina<=a.
          inb<=b.
       end
   end
end
assign mult_out=ina*inb.
always @(posedge clock) begin
   if(reset)begin
       pp_reg<=0.
   end else begin
       if(ce)begin
          pp_reg<=mult_out.
       end
   end
end
always @(posedge clock) begin
   if(reset)begin
       out_reg<=0.
   end else begin
       if(ce)begin
```

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```
out_reg<=pp_reg.
end
end
end
assign c=out_reg.
endmodule
```

The above multiplier circuit diagram is shown in Figure 4-15.

Figure 4-15 DSP Circuit Diagram in Example 1



Example 2 can be synthesized to a multiplier accumulator in asynchronous mode, which has input registers a0\_reg, a1\_reg, b0\_reg and b1\_reg, output register s\_reg and bypass registers p0\_reg and p1\_reg.

```
module top(a0, a1, b0, b1, s, reset, clock, ce).
parameter a0 width=18.
parameter a1_width=18.
parameter b0_width=18.
parameter b1_width=18.
parameter s width=37.
input unsigned [a0 width-1:0] a0.
input unsigned [a1_width-1:0] a1.
input unsigned [b0_width-1:0] b0.
input unsigned [b1_width-1:0] b1.
input reset, clock, ce.
output unsigned [s_width-1:0] s.
wire unsigned [s width-1:0] p0, p1, p.
reg unsigned [a0_width-1:0] a0_reg.
reg unsigned [a1_width-1:0] a1_reg.
reg unsigned [b0_width-1:0] b0_reg.
```

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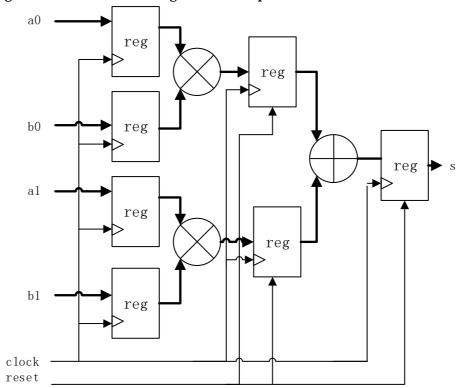
```
reg unsigned [b1_width-1:0] b1_reg.
reg unsigned [s_width-1:0] p0_reg, p1_reg, s_reg.
always @(posedge clock or posedge reset)
begin
   if(reset)begin
       a0_reg <= 0.
       a1\_reg \le 0.
       b0 reg \le 0.
       b1_reg <= 0.
   end else begin
       if(ce)begin
           a0\_reg \le a0.
           a1 reg \le a1.
           b0\_reg \le b0.
           b1 reg \le b1.
       end
   end
end
assign p0 = a0\_reg*b0\_reg.
assign p1 = a1\_reg*b1\_reg.
always @(posedge clock or posedge reset)
begin
   if(reset)begin
       p0_reg <= 0.
       p1 reg \le 0.
   end else begin
       if(ce)begin
           p0\_reg \le p0.
           p1\_reg \le p1.
       end
   end
end
assign p = p0\_reg - p1\_reg.
always @(posedge clock or posedge reset)
begin
```

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```
if(reset)begin
    s_reg <= 0.
end else begin
    if(ce) begin
    s_reg <= p.
end
end
end
end
assign s = s_reg.
endmodule</pre>
```

The above multiplier accumulator circuit diagram is shown in Figure 4-16.

Figure 4-16 DSP Circuit Diagram in Example 2



Example 3 can be synthesized to two unsigned bit multipliers, which are in chain addition relation.

```
module top(a0, a1, a2, b0, b1, b2, a3, b3, s).

parameter a_width=18.

parameter b_width=18.

parameter s_width=36.

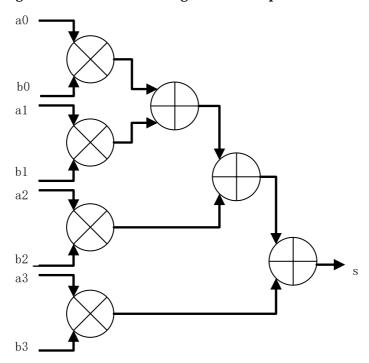
input unsigned [a_width-1:0] a0, a1, a2, b0, b1, b2, a3, b3.
```

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```
output unsigned [s_width-1:0] s.
assign s=a0*b0+a1*b1+a2*b2+a3*b3.
endmodule
```

The above multiplier accumulator circuit diagram is shown in Figure 4-17.

Figure 4-17 DSP Circuit Diagram in Example 3



Example 4 can be synthesized to a multiplier with sign-bit 0 and a preadder with sign-bit 0. An input port of this multiplier is connected to the output port b of the preadder.

```
module top(a, bX, bY, p).

parameter a_width=36.

parameter b_width=18.

parameter p_width=54.

input [a_width-1:0] a.

input [b_width-1:0] bX, bY.

output [p_width-1:0] p.

wire [b_width-1:0] b.

assign b = bX + bY.

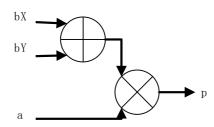
assign p = a*b.

endmodule
```

The above multiplier accumulator circuit diagram is shown in Figure 4-18.

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Figure 4-18 DSP Circuit Diagram in Example 4



Example 5 can be synthesized to a multiplier accumulator with sign-bit 0, and the output register is s.

```
module acc(a, b, s, accload, reset, ce, clock).
parameter a_width=36. //18 36
parameter b width=18. //18 36
parameter s_width=54. //54
input unsigned [a_width-1:0] a.
input unsigned [b_width-1:0] b.
input accload, reset, ce, clock.
output unsigned [s width-1:0] s.
wire unsigned [s_width-1:0] s_sel.
wire unsigned [s_width-1:0] p.
reg [s_width-1:0] s.
assign p = a*b.
assign s_sel = (accload == 1'b1) ? s : 54'h0000000.
always @(posedge clock)
begin
   if(reset)begin
       s <= 0.
   end else begin
       if(ce)begin
           s \le s_sel + p.
       end
   end
end
endmodule
```

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The above multiplier accumulator circuit diagram is shown in Figure

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gnd accload clock reset ce

Figure 4-19 DSP Circuit Diagram in Example 5

#### Note!

For more examples, please see <u>GowinSynthesis\_Inference\_Coding\_Template</u> at Gowinsemi official website.

# 4.4 Synthesis Implementation Rules for Finite State Machine

### 4.4.1 Synthesis Rules for Finite State Machine

GowinSynthesis supports the synthesis of Finate State Machine (FSM), and the encoding mode supports one-hot code, gray code, binary code, etc. The synthesis results of finite state machine are related to its encoding mode, code number, code bit width and code constraints. If not setting the code constraints, GowinSynthesis automatically selects the one-hot code or gray code to realize state machine. If setting code constraints, the code method specified by the constraints should be implemented first. For the code constraints of state machine, please see syn encoding.

#### Note!

It should be noted that if the output of the finite state machine drives the output port directly, GowinSynthesis will not synthesize it as a state machine, and the code constraints of the state machine will be ignored.

## 4.4.2 Finite State Machine Code Example

The synthesis rules for finite state machine are described below.

#### **One-hot Code State Machine**

If the state machine adopts one-hot code for coding in RTL design, GowinSynthesis will select one-hot code by default to realize the functions of the state machine with no code constraints. In the case of code constraints, the functions of the state machine are realized according to the encoding mode specified by the constraints. One-hot encoding mode example is as follows:

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```
reg [3:0] state,next_state;
parameter state0=4'b0001;
parameter state1=4'b0010;
parameter state2=4'b0100;
parameter state3=4'b1000;
```

In the above example, RTL uses one-hot code and GowinSynthesis uses one-hot code to implement.

#### **Gray Code State Machine**

If the state machine adopts gray code for coding in the RTL design, GowinSynthesis will select gray code by default to realize the functions of the state machine with no code constraints. In the case of code constraints, the functions of the state machine are realized according to the encoding mode specified by the constraints. Gray code example is as follows:

```
reg [3:0] state,next_state;
parameter state0=2'b00;
parameter state1=2'b01;
parameter state2=2'b11;
parameter state3=2'b10;
```

In the above example, RTL uses gray code and GowinSynthesis uses gray code to implement.

#### **Binary Code or other Codes State Machines**

If the state machine adopts binary code in RTL design, which is neither one-hot code nor gray code, GowinSynthesis will select the corresponding code according to the number of codes and bit width for implementation with no constraints. The selection principle is as follows: If the number of code is greater than the effective bit width of code, gray code will be used for implementation. If the number of code is less than or equal to the effective bit width of code, one-hot code will be used for implementation. In the case of code constraints, the functions of state machine are realized according to the encoding mode specified by the constraints.

```
Example 1

reg [5:0] state,next_state;

parameter state0= 6'b000001;

parameter state1= 6'b000001;

parameter state2= 6'b000000;

parameter state3= 6'b010101;
```

In the above example, the number of code is 4, the bit width of code is 6, and the effective bit width is 5, so the number of code is less than the effective bit width of code, and the implementation is carried out by one-hot code.

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```
Example 2

reg [2:0] state,next_state;

parameter state0=3'b001;

parameter state1=3'b010;

parameter state2=3'b011;

parameter state3=3'b100;
```

In the above example, the number of code is 4, and the effective bit width of code is 3. The number of code is larger than the effective bit width of code, and the implementation is carried out by gray code.

```
Example 3

reg [5:0] state,next_state.

parameter state0= 1,

parameter state1= 3,

parameter state2= 6,

parameter state3= 15.
```

In the above example, the number of code is 4 in decimal, and the effective bit width converted into binary is 4 bits. The number of code is equal to the effective bit width of code, and the implementation is carried out by one-hot code.

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# 5 Synthesis Constraints Support

Attribute constraints is used to set various attributes of optimization selection, function implement, output netlist format in synthesis so that the synthesis results can better meet the design function and usage. Attributes can be written in constraint files or embedded in source code.

This chapter describes the syntax for constraints in RTL files and GowinSynthesis Constraint (GSC) files. Verilog files are case-sensitive, so instructions and attributes must be typed exactly as described in the syntax. An attribute constraint must be written in the same line in a constraint statement and can not separate by breaks, and a semicolon must be added at the end of the statement.

#### **Constraints in RTL File**

Constraints in the RTL file must be added in the definition statement of the constraint object before the semicolon. If the constraint setting\_value in the statement is a string, then double quotes should be added before and after the setting\_value. If the setting\_value is a number, then no double quotes are required.

## **GSC**

GSC constraints include Instance constraints, Net constraints, Port constraints and global objects constraints. In order to distinguish the types, there are different syntaxes in writing. The constraint object must be enclosed by double quotation marks. Attribute name and the setting\_value need not be identified by double quotation marks or other sligns, and there can be spaces before and after the equal sign. GSC constraints support notes with "//". The examples are as follows:

INS "object" attributeName=setting value;

NET "object" attributeName=setting value;

PORT "object" attributeName=setting value;

GLOBAL attributeName=setting value;

The constraint statement begins with INS, and the object must be the name of instance. Instance includes module/entity instance and primitive instance. The name of instance does not contain parenthesis. In other

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words, don't write temp[15:0] when bus, just write temp instead.

The constraint statement begins with NET, and the constraint object must be the NET name.

The constraint statement begins with PORT, and the constraint object must be the PORT name.

The constraint statement begins with GLOBAL, indicating that the attribute constraint is global.

The name of the object in the constraint must match the one in the netlist. There can be no spaces in the name. Wildcards are supported in the name of the object. Use "/" to distinguish the hierarchy of names. Add w before object to distinguish when using wildcards, such as w "object".

The setting\_value may be the value specified directly by the user, inherited from the super-structure, or the default. The priority of values is direct value in GSC > direct value in RTL > inherited value in GSC > inherited value in RTL > default value. When there are multiple inherited values, take the value closest to the specified name (lowest level). For example, check the MULT\_STYLE attribute of A/D/C/mult1 ("/" indicates the hierarchy between module names), which has direct value of DSP. Check MULT\_STYLE attribute of "A/D/C", which has no direct value, and the MULT\_STYLE attribute can be inherited, so find and inherit the attribute value logic of "A/D". Check MULT\_STYLE of "A/D/C/mult1", which has both the inherited value and the direct value. The direct value DSP is finally taken because of the priority.

# 5.1 black\_box\_pad\_pin

#### Description

Specify that the IO pads of the black box are visible to the external environment. This attribute only works for the black box IO pads.

This attribute can only be specified in the RTL file.

#### **Syntax**

```
Verilog Constraint Syntax

Verilog object /* synthesis black_box_pad_pin=portList */;

VHDL Constraint Syntax

attribute black_box_pad_pin : string;

attribute black box pad pin of object: objectType is portList;
```

#### Note!

- object: It can be module or component defined by a black box.
- The PortList enclosed in double quotes is a space-less, comma-separated list of the ports in the black box.

### **Examples**

```
Verilog Constraint Example module top(clk, in1, in2, out1, out2,D,E);
```

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```
input clk;
    input [1:0]in1;
    input [1:0]in2;
    output [1:0]out1;
    output [1:0]out2;
    output D,E;
    black_box_add U2 (in1, in2, out2,D,E);
    endmodule
    module black box add(A, B, C, D,E)/* synthesis syn black box
black box pad pin="D,E" */;
    input [1:0]A;
    input [1:0]B;
    output [1:0]C;
    output D,E;
    endmodule
    VHDL Constraint Example
    library ieee;
    use ieee.std_logic_1164.all;
    entity top is
    generic (width : integer := 4);
    port (in1,in2: in std logic vector(width downto 0);
       clk: in std_logic;
       q : out std_logic_vector (width downto 0)
    );
    end top;
    architecture top1_arch of top is
    component test is
    generic (width1 : integer := 2);
    port (in1,in2 : in std_logic_vector(width1 downto 0);
       clk: in std logic;
       q : out std_logic_vector (width1 downto 0)
    );
    end component;
```

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```
attribute black_box_pad_pin : string;
attribute black_box_pad_pin of test : component is "q[4:0]";
begin
test123 : test generic map (width) port map (in1,in2,clk,q);
end top1_arch;
```

# 5.2 full case

# Description

full\_case is only used in Verilog design. When this attribute is added after a case, casex or casez statement, it means that all possible values are given and no redundant hardware is required to preserve the signal values.

This attribute can only be specified in RTL files and only support Verilog.

# **Syntax**

```
RTL Constraint Syntax verilog case /* synthesis full_case*/
```

# Example

RTL Constraint Example

Example 1 specifies that part of the circuit no longer requires redundant hardware to preserve signal values

```
module top(...);
.....
always @(select or a or b or c or d)
begin
casez(select) /* synthesis full_case*/
4'b???1: out=a;
.....
4'b1???: out=d;
endcase
end
endmodule
```

# 5.3 parallel\_case

The parallel\_case is an instruction, which forces to use a parallel-multiplexed structure instead of a priority-encoded structure.

This instruction can only be specified in the Verilog file.

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## Description

The case statement is defined in priority order by default, executing only the first statement that matches the selected value. Priority coding allows to have input signals at several inputs at the same time, and to code only the highest priority of the several signals input at the same time in the priority order.

If the bus selected is driven from outside of the current module and the current module does not have information about the legal selected value, the software must create a logical chain of disable in order to match the tag statement to disable all subsequent statements.

However, if the legal value of the selection is known, the parallel\_case instruction can be used to eliminate the additional priority coding logic.

# **Syntax**

```
Verilog Constraint Syntax
object /* synthesis parallel_case */;
```

#### Note!

- global support: No
- object: case, casex, and casez statements.
- setting value: No value.

# Example

```
Verilog Constraint Example
module test (out, a, b, c, d, select);
output out;
input a, b, c, d;
input [3:0] select;
reg out;
always @(select or a or b or c or d)
begin
   casez (select) /* synthesis parallel case */
       4'b???1: out = a:
       4'b??1?: out = b:
       4'b?1??: out = c;
       4'b1???: out = d:
       default: out = 'bx:
endcase
end
endmodule
```

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# 5.4 syn\_black\_box

## Description

Specify a module or component as a black box. In synthesis, a black box module defines only its interface, and its content can not be accessed or optimized. The module is treated as a black box regardless of whether it is empty or not.

This attribute can only be specified in the RTL file.

#### **Syntax**

```
Verilog Constraint Syntax
object /* synthesis syn_black_box */;
VHDLConstraint Syntax
attribute syn_black_box: boolean;
attribute syn black box of object : objectType is true;
```

#### Note!

- object: It can only be sub module/entity.
- objectType: It is the type of object, usually component.

## **Examples**

```
Verilog Constraint Example
module top(clk, in1, in2, out1, out2);
input clk;
input [1:0]in1;
input [1:0]in2;
output [1:0]out1;
output [1:0]out2;
add U1 (clk, in1, in2, out1);
black_box_add U2 (in1, in2, out2);
endmodule
module add (clk, in1, in2, out1);
begin
out1 \le in1 + in2;
end
endmodule
module black_box_add(A, B, C)/* synthesis syn_black_box */;
```

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```
assign C = A + B;
endmodule
```

Before the attribute used, the content of the module black\_box\_add is visible. After the attribute used, the content of the module black\_box\_add is not visible and becomes a black box.

```
VHDL Constraint Example
library ieee;
use ieee.std_logic_1164.all;
use ieee.std_logic_unsigned.all;
entity mux2_1_top is
port(
    dina: in bit;
    dinb: in bit;
    sel : in bit;
    dout: out bit
);
end mux2_1_top;
architecture Behavioral of mux2_1_top is
    attribute syn_black_box: boolean;
    attribute syn_black_box of mux2_1 : component is true;
begin
    u_mux2_1 : mux2_1
    port map(
        dina => dina,
        dinb => dinb,
         sel => sel,
         dout => dout
    );
end Behavioral;
```

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# 5.5 syn\_dspstyle

# Description

The specified multiplier is implemented by a dedicated DSP hardware module or logic circuit, which can be applied to both specific module/entity instance and the global. This attribute can be specified in GSC and RTL files.

### **Syntax**

```
GSC Constraint Syntax

INS "object" syn_dspstyle =setting_value;

GLOBAL syn_dspstyle =setting_value;

Verilog Constraint Syntax

Verilog object /* synthesis syn_dspstyle ="setting_value" */;

VHDL Constranit Syntax

attribute syn_dspstyle:string;

attribute syn_dspstyle of object:objectType is "setting_value";
```

#### Note!

- object: It can be wire, register, module/entity name or module/entity instance name.
- setting value: Multiplier implementation currently supports DSP and logic.
- setting value is logic, inferring object to logic circuit.
- setting\_value is DSP inferring object to DSP, which is the default.

#### **Examples**

```
GSC Constraint Example

Example 1 specifies logic to implement instance

INS 'temp" syn_dspstyle=logic;

INS "aa0/mult/c" syn_dspstyle=logic;

Example 2 specifies logic to implement global multipliers

GLOBAL syn_dspstyle=logic;

Verilog Constraint Example

Example 1 specifies logic to implement all multipliers in mult module.

module mult(...) /* synthesis syn_dspstyle = "logic" */;

......

wire [15:0] temp;

assign temp = a*b;

......

Endmodule

Example 2 specifies logic to implement the multiplier temp.

module mult(...);
```

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```
wire [15:0] temp/* synthesis syn_dspstyle = "logic" */;
assign temp = a*b;
endmodule
VHDLConstraint Example
Example 1 specifies logic to implement the multiplier result.
entity Mult is
port(
    result : out signed(23 downto 0));
    attribute syn_dspstyle:string;
    attribute syn dspstyle of result : signal is "logic";
end Mult:
architecture Behavior of Mult is
    signal x1 : signed(11 downto 0);
    signal y1 : signed(11 downto 0);
begin
    result \leq x1 * y1;
end Behavior;
```

# 5.6 syn\_encoding

#### Description

Specify the encoding mode of state machine, which can be applied to both specific objects and the global.

This attribute can only be specified in RTL files.

## **Syntax**

```
Verilog Syntax

verilog object /* synthesis syn_encoding = "setting_value" */;

VHDLConstraints Syntax

attribute syn_encoding : string;

attribute syn_encoding of object : objectType is "setting value";
```

#### Note!

- object: It can only be the name of register.
- setting\_value: One-hot code and gray code are currently supported for state

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machine.

## **Examples**

```
Verilog Example
This example specifies gray code for the state machine
module test (...);
reg [2:0] ps, ns/* synthesis syn_encoding="gray" */;
endmodule
VHDL Example
This example specifies one-hot code for the state machine.
ENTITY fsm IS
END fsm;
ARCHITECTURE behaviour OF fsm IS
TYPE state_type IS (s0,s1,s2,s3);
SIGNAL present_state,next_state : state_type;
attribute syn_encoding:string;
attribute syn encoding of present state, next state: signal is "onehot";
BEGIN
END behaviour;
```

# 5.7 syn\_insert\_pad

#### Description

Specify whether to insert an I/O buffer. Insert the I/O buffer when the attribute value is 1

This attribute can only be specified in GSC files.

## **Syntax**

```
GSC Constraints Syntax

PORT "object" syn_insert_pad=setting_value;
```

#### Note!

- setting\_value: 0 or 1. Remove I/O buffer at 0. Insert I/O buffer at 1.
- object: It can only be port. This constraint only applies to input port or output port, not inout port.

#### **Examples**

**GSC Examples** 

Example 1 specifies the inserted I/O buffer

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```
PORT "out" syn_insert_pad=1;
Example 2 specifies the removed I/O buffer
PORT "out" syn_insert_pad=0;
```

# 5.8 syn\_keep

## Description

Specify a wire, reg, port as a placeholder and leave it unoptimized This attribute can only be specified in RTL files.

# **Syntax**

```
Verilog Constraints Syntax

Verilog object /* synthesis syn_keep= setting_value */;

VHDL Constraints Syntax

attribute syn_keep: integer;

attribute syn keep of object: objectType is 1;
```

#### Note!

- object: It can only be wire, port and combination logic.
- setting\_value: The setting\_value can only be 0 or 1. When it is 1, the net is preserved without optimization.

# **Examples**

```
Verilog Constraint Example

This example specifies mywire and leaves it unoptimized.

module test (...);
......

wire mywire /* synthesis syn_keep=1 */;
......

endmodule

VHDL Constraint Example

This example specifies tmp0 and leaves it unoptimized.

entity mux2_1 is

port(
......
);

end mux2_1;

architecture Behavioral of mux2_1 is

signal tmp0:bit;

signal tmp1:bit;

attribute syn_keep: integer;
```

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```
attribute syn_keep of tmp0 : signal is 1; .....
end Behavioral:
```

# 5.9 syn\_looplimit

# Description

Specify the limited value of loop iteration. The default is 2000 times. If it does not specify the time, exceeding the default, there will be an error in synthesis.

This attribute can only be set in GSC.

## **Syntax**

```
GSC Constraints Syntax
GLOBAL syn_looplimit=setting_value
```

#### Note

setting\_value: It can only be number, meaning the upper limited times in loop iteration.

#### Example

```
GSC Constraints Example

GLOBAL syn looplimit=3000
```

# 5.10 syn\_maxfan

# Description

Specify max. fanout, which can be applied to both specific objects and the global.

This attribute can be specified in GSC and RTL files.

#### **Syntax**

Note!

```
INS "object" syn_maxfan=setting_value;

NET "object" syn_maxfan=setting_value;

GLOBAL syn_maxfan=setting_value;

Verilog Constraints Syntax

Verilog object /* synthesis syn_maxfan = setting_value */;

VHDL Constraints Syntax

attribute syn_maxfan : integer;

attribute syn_maxfan of object : objectType is setting_value;
```

• object: It can be wire, register, input, output, module/entity name, module/entity instance name; it does not work for the inpin related with CLK, CE, LSR.

setting value: Integer greater than 0.

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# **Examples**

```
GSC Examples
    Example 1 specifies that the max.fanout of instance is 10.
    INS "d" syn maxfan=10;
    Example 2 specifies that the max.fanout of global is 100.
    GLOBAL syn_maxfan=100;
    Example 3 specifies that the max.fanout of instance is 10.
    INS "aa0/mult/d" syn maxfan=10;
    Example 4 specifies that the max.fanout of net is 10.
    NET "aa0/mult/d" syn_maxfan=10;
    Verilog Examples
    Example 1 specifies a maximum fanout value of 3 for all instances in
the module except CLK.
    module test (...) /* synthesis syn_maxfan = 3*/;
    endmodule
    Example 2 specifies that the max.fanout of instance is 3.
    module test (...);
    reg [7:0] d /* synthesis syn_maxfan = 3*/;
    endmodule
    VHDL Example
    entity test is
    end test:
    architecture rtl of test is
    signal d : std_logic;
    attribute syn maxfan : integer;
    attribute syn_maxfan of d : signal is 5;
    end rtl;
```

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# 5.11 syn\_netlist\_hierarchy

#### Description

Specify whether to generate hierarchy netlists. The default 1 indicates the generation of hierarchy netlists. When set to 0, the hierarchy netlists are output flattened.

This attribute can be specified in GSC and RTL files.

# **Syntax**

```
GSC Constraint Syntax

GLOBAL syn_netlist_hierarchy=setting_value;

Verilog Constraint Syntax

Verilog object /* synthesis syn_netlist_hierarchy=setting_value */;

VHDL Constraint Syntax

attribute syn_netlist_hierarchy: integer;

attribute syn_netlist_hierarchy of object: objectType is setting_value;
```

#### Note!

- object: It can only be the top module/entity.
- setting\_value: The setting\_value is 0 or 1. If it is 1, hierarchy is allowed to be generated. If it is 0, the hierarchy netlists are output flattened.

# **Examples**

```
GSC Example
GLOBAL syn_netlist_hierarchy=0;
Verilog Example
module rp_top (...) /* synthesis syn_netlist_hierarchy=1 */;
endmodule
VHDL Example
entity mux4_1_top is
port(
    dina: in bit;
    dinb: in bit;
    sel : in bit;
    dout : out bit
);
attribute syn_netlist_hierarchy: integer;
attribute syn netlist hierarchy of mux4 1 top: entity is 0;
end mux4_1_top;
```

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# 5.12 syn\_noprune

#### Description

Whether the output of module/entity instance, primitive instance or black box including primitive is optimized or not when they are all hung, both for specific objects and for the global.

This attribute can only be specified in RTL file.

# **Syntax**

```
Verilog Constraints Syntax

Verilog object /* synthesis syn_noprune = setting_value */;

VHDL Constraints Syntax

attribute syn_noprune : integer;

attribute syn_noprune of object: objectType is 1;
```

#### Note!

- object: It can be module/entity instance name, primitive instance name or black box.
- setting\_value: It can only be 0 or 1. If it is 1, retain instance and black box. If it is 0, optimize the corresponding instance and black box as needed.

# Example

```
Verilog Constraint Example
module test (out1,out2,clk,in1,in2);
noprune_bb u1(out1,in1)/*synthesis syn_noprune=1*/;
endmodule
module noprube_bb(din,dout);
input din;
output dout;
endmodule
VHDL Constraint Example
library ieee;
use ieee.std logic 1164.all;
entity top is
end entity top;
architecture arch of top is
component noprune_bb
port(
```

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```
din: in std_logic;
dout: out std_logic);
end component noprune_bb;
signal o1_noprunereg: std_logic;
signal o2_reg: std_logic;
attribute syn_noprune: integer;
attribute syn_noprune of U1: label is 1;
attribute syn_noprune of o1_noprunereg: signal is 1;
.....
end architecture arch;
```

# 5.13 syn\_preserve

# Description

Specify register or whether to optimize the register logic, which can be applied to both specific objects and the global.

This attribute can be specified in RTL and GSC files.

# **Syntax**

```
GSC Constraint Syntax

INS "object" syn_preserve=setting_value;

GLOBAL syn_preserve=setting_value;

Verilog Constraint Syntax

Verilog object /* synthesis syn_preserve = setting_value */;

VHDL Constraint Syntax

attribute syn_preserve : integer;

attribute syn_preserve of object : objectType is setting_value;
```

#### Note!

- Object: It can be the name of register, module/entity, module/entity instance.
- setting\_value: 0 or 1. When it is 1, the corresponding register is preserved. When it is 0, the corresponding register is optimized as needed.

#### **Examples**

```
GSC Examples

Example 1 specifies reg1and leaves it unoptimized.

INS "reg1" syn_ preserve =1;

Example 2 specifies that all registers in the design are preserved.

GLOBAL syn_ preserve =1;

Verilog Examples

Example 1 specifies that all registers in the module are preserved.
```

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```
module test (...) /* synthesis syn_preserve = 1 */;
endmodule
Example 2 specifies reg1and leaves it unoptimized.
module test (...).
reg reg1/* synthesis syn_preserve = 1 */;
endmodule
VHDL Example
Example 1 specifies to preserve reg 1.
entity syn_test is
    port (.....
    );
end syn_test;
architecture behave of syn_test is
    signal reg1 : std_logic;
    signal reg2 : std_logic;
    attribute syn_preserve : integer;
    attribute syn preserve of reg1: signal is 1;
begin
end behave;
```

# 5.14 syn\_probe

## Description

This attribute tests and debugs the internal signals in the design by inserting probe points. The specified probe points appear as ports in the top-level port list.

This attribute can only be specified in RTL files.

#### **Syntax**

```
Verilog Constraints Syntax

Verilog object /* synthesis syn_probe = setting_value */;

VHDL Constraints Syntax

attribute syn_probe: string;

attribute syn probe of object: objectType is " setting value ";
```

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#### Note!

- object: It can only be wire or register.
- setting\_value is 1: Insert the probe point and automatically get the probe port name according to the net name.
- setting value is 0: Probe is not allowed.
- setting\_value is a string: Insert a specified probe point name. When the name specified by setting\_value is bus, the number is automatically added after the inserted name.
- GowinSyn does not support the setting\_value with the same value as the object name or the port name of the module.

#### **Examples**

Verilog Constraints Example

When probe\_tmp is set, probe\_tmp is listed in output port list of the top level.

```
module test (...);
.....

reg [7:0] probe_tmp /* synthesis syn_probe=1*/;
.....

endmodule

VHDL Constraints Example
entity halfadd is
port(.....);
end halfadd;
architecture add of halfadd is
    signal probe_tmp: std_logic;
    attribute syn_probe: string;
    attribute syn_probe of probe_tmp: signal is "probe_string";
.....
End;
```

# 5.15 syn\_ramstyle

#### Description

Specify the implementation of memory, which can be applied to both specific instances and the global.

This attribute can be specified in GSC and RTL files.

#### **Syntax**

```
GSC Constraint Syntax
INS "object" syn_ramstyle =setting_value;
GLOBAL syn_ramstyle =setting_value;
Verilog Constraint Syntax
```

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```
Verilog object /* synthesis syn_ramstyle = "setting_value" */;

VHDL Constraint Syntax

attribute syn_ramstyle:string;

attribute syn_ramstyle of object : objectType is " setting_value";
```

#### Note!

- object: It can be module/entity name, module entity instance name, or register.
- setting\_value: The implementation of memory currently supports block\_ram, distributed\_ram, registers, rw\_check, no\_rw\_check.
- setting\_value is registers: Maps inferred RAM to registers (flip-flop and logic circuit) rather than dedicated RAM resources.
- setting\_value is block\_ram: Maps inferred RAM to dedicated memory, which uses FPGA dedicated memory resources.

## **Examples**

```
GSC Constraint Examples
```

Example 1 specifies BSRAM to implement instance.

INS "mem" syn\_ramstyle=block\_ram;

Example 2 specifies SSRAM to implement global memory.

GLOBAL syn ramstyle=distributed ram;

**Verilog Constraint Examples** 

Example 1 specifies block\_ram to implement memory in module and there are no read or write check.

```
module test (...) /* synthesis syn_ramstyle = "block_ram" */
...
endmodule
Example 2 specifies BSRAM to implement instance.
module test (...);
.....
reg [DATA_W - 1 : 0] mem [(2**ADDR_W) - 1 : 0] /* synthesis
syn_ramstyle = "block_ram" */;
.....
endmodule
VHDL Constraint Example
Example 1 specifies BSRAM to implement memory.
entity ram is
GENERIC(bits:INTEGER:=8;
words:INTEGER:=256);
PORT(.....);
end ram;
```

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```
ARCHITECTURE arch of ram IS

TYPE vector_array IS ARRAY(0 TO words-1) OF

STD_LOGIC_VECTOR(bits-1 DOWNTO 0);

SIGNAL memory:vector_array;

attribute syn_ramstyle:string;

attribute syn_ramstyle of memory: signal is " block_ram";

BEGIN

......

end arch;
```

# 5.16 syn\_romstyle

## Description

Specify the implementation of read-only memory, which can be applied to both specific objects and the global.

This attribute can be specified in GSC and RTL files.

#### **Syntax**

```
GSC Constraint Syntax

INS "object" syn_romstyle =setting_value;

GLOBAL syn_romstyle =setting_value;

Verilog Constraints Syntax

Verilog object /* synthesis syn_romstyle = "setting_value" */;

VHDL Constraint Syntax

attribute syn_romstyle:string;

attribute syn_romstyle of object : objectType is " setting_value";
```

#### Note!

- object: It can be module/entity name, module entity instance name, or register.
- setting\_value: The implementation of read-only memory currently supports block\_rom, distributed\_rom, logic.

## **Examples**

```
GSC Constraint Examples
```

Example 1 specifies BSRAM to implement instance.

INS "mem" syn\_romstyle=block\_rom;

Example 2 specifies SSRAM to implement global memory.

GLOBAL syn\_romstyle=distributed\_rom;

Verilog Constraint Example

This example specifies SSRAM to implement memory in module.

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```
module rom16_test(...)/*synthesis syn_romstyle="distributed_rom"*/;
.....

endmodule

VHDL Constraint Example

This example specifies SSRAM to implement all memories in the module.

ENTITY rom is
.....

end rom;

ARCHITECTURE rom OF rom IS

signal data_out:STD_LOGIC_VECTOR(bits-1 DOWNTO 0);
attribute syn_romstyle:string;
attribute syn_romstyle of data_out: signal is "block_rom";
.....

END rom;
```

# 5.17 syn\_srlstyle

## Description

Specify the implementation of shift registers, either acting on a specific object or the global. The shift register can be implemented by BSRAM, SSRAM, and registers. The number of registers in the shift register determines which implementation method to use by default. The default value can be changed by using syn\_srlstyle.

This attribute can be specified in GSC and RTL files.

#### **Syntax**

```
GSC Constraint Syntax

NS "object" syn_srlstyle =setting_value;

GLOBAL syn_srlstyle =setting_value;

Verilog Constraint Syntax

Verilog object /* synthesis syn_srlstyle = "setting_value" */;

VHDL Constraint Syntax

attribute syn_srlstyle:string;

attribute syn_srlstyle of object : objectType is " setting_value";
```

#### Note!

- object: It can be module/entity name, module/entityinstance name or register. Module/entity is not supported in the GSC syntax.
- setting\_value: The memory implementation supports block\_ram, distributed\_ram, and registers currently.

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# Example **GSC Constraint Examples** Example 1 specifies that instance implementation is BSRAM. INS "mem" syn sristyle=block ram; Example 2 specifies that the global memory implementation is SSRAM. GLOBAL syn\_srlstyle=distributed\_ram; Verilog Constraint Example Example 1 specifies that the register implementation in the module is block\_ram. module test (...) /\* synthesis syn\_srlstyle = "block\_ram" \*/; endmodule Example 2 specifies that instance implementation is BSRAM. module test (...); . . . . . . reg [SRL\_WIDTH-1:0] regBank[SRL\_DEPTH-1:0]/\* synthesis syn srlstyle = "block ram" \*/; endmodule VHDL Constraint Example is register. entity ram is

This example specifies that the register implementation in the module

```
GENERIC(bits:INTEGER:=8;
   words:INTEGER:=256);
PORT(.....);
attribute syn_srlstyle:string;
attribute syn_srlstyle of shiftreg : entity is "registers";
end ram:
ARCHITECTURE arch of ram IS
end ram;
```

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# 5.18 syn\_tlvds\_io/syn\_elvds\_io

#### Description

Specify the attribute of the differential I/O buffer mapping, which can be applied to both specific objects and the global.

This attribute can be specified in GSC and RTL files.

```
Syntax
```

```
GSC Constraint Syntax

PORT "object" syn_tlvds_io = setting_value;

GLOBAL syn_tlvds_io = setting_value;

PORT "object" syn_elvds_io = setting_value;

GLOBAL syn_elvds_io = setting_value;

Verilog Constraint Syntax

Verilog object /* synthesis syn_tlvds_io = setting_value */;

VHDL Constraint Syntax

attribute syn_tlvds_io: integer;

attribute syn_tlvds_io of object: objectType is setting_value;
```

#### Note!

- object: It can be module/entity name or port name.
- setting\_value: 0 or 1.

#### **Examples**

```
GSC Constraint Examples
```

Example 1 specifies that the buffer implementation is TLVDS.

```
PORT "io" syn_tlvds_io =1;
PORT "iob" syn_tlvds_io =1;
```

Example 2 specifies that the implementation of all buffers in the global is TLVDS.

```
GLOBAL syn_tlvds_io =1;

Verilog Constraint Example

module elvds_iobuf(io,iob...);

inout io/*synthesis syn_elvds_io=1*/;

inout iob/*synthesis syn_elvds_io=1*/;

.....

endmodule

VHDL Constraint Example

entity test is

port (in1_p: in std_logic;
```

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```
in1_n: in std_logic;
clk: in std_logic;
out1: out std_logic;
out2: out std_logic);
attribute syn_tlvds_io: integer;
attribute syn_tlvds_io of in1_p,in1_n,out1,out2: signal is 1;
end test;
architecture arch of test is
.....
end arch
```

# 5.19 translate\_off/Translate\_on

# Description

translate\_off /translate\_on must occur in pairs, and statements after translate\_off are skipped during the synthesis until translate\_on occurs, which is often used to automatically mask some statements during synthesis.

This attribute can only be specified in RTL files.

# **Syntax**

```
Verilog Constraint Syntax

/* synthesis translate_off*/;

Statements ignored in the synthesis.

/* synthesis translate_on*/;

VHDL Constraint Syntax

-- synthesis translate_off

Statements ignored in the synthesis

-- synthesis translate_on

Examples

Verilog Constraint Example

The assign Nout =a*b between /*synthesis translate_off*/ and

/*synthesis translate_on*/ is ignored in the synthesis in example 1.

module test (...);

......

/*synthesis translate_off*/
assign my_ignore=a*b;
```

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/\* synthesis translate on\*/

```
endmodule

VHDL Constraint Example
entity top is
port (
.....
);
end top;
architecture rtl of top is
begin
dout <= a + b;
-- synthesis translate_off
Nout <= a * b;
-- synthesis translate_on
end rtl;
```

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6 Report Document 6.1 Synthesis Message

# 6 Report Document

The report document is the statistical report generated after synthesis. The file name is \*\_syn.rpt.html (\* is the name of specified output netlist vg file). It includes Synthesis Message, Synthesis Details, Resource, and Timing, etc.

# **6.1 Synthesis Message**

Synthesis Message refers to the basic information of Synthesis. As shown in Figure 6-1. It mainly includes design file, GowinSynthesis version, configuration information, and created time, etc.

Figure 6-1 Synthesis Message

#### **Synthesis Messages**

Report Title	GowinSynthesis Report
Design File	E:\tmp\mantis\10631\mantis10542\src\test.v
GowinSynthesis Constraints File	
GowinSynthesis Version	GowinSynthesis V1.9.8
Part Number	GW1NSR-LX2QN48PES
Device	GW1NSR-2
Created Time	Mon Aug 16 15:00:53 2021
Legal Announcement	Copyright (C)2014-2021 Gowin Semiconductor Corporation. ALL rights reserved.

# **6.2 Synthesis Details**

Synthesis Details includes the information of top level module, synthesis process, total time and memory usage, as shown in Figure 6-2.

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6 Report Document 6.3 Resource

Figure 6-2 Synthesis Details

# **Synthesis Details**

Top Level Module	top
Synthesis Process	Running parser:  CPU time = 0h 0m 0.109s, Elapsed time = 0h 0m 0.123s, Peak memory usage = 52.926MB Running netlist conversion:  CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 52.992MB Running device independent optimization:  Optimizing Phase 0: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.133MB Optimizing Phase 1: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.191MB Optimizing Phase 1: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.254MB Running inference:  Inferring Phase 0: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.344MB Inferring Phase 1: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.403MB Inferring Phase 2: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.430MB Inferring Phase 3: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.441MB Running technical mapping:  Tech-Mapping Phase 0: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.465MB Tech-Mapping Phase 1: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.465MB Tech-Mapping Phase 2: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 53.473MB Tech-Mapping Phase 3: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 3: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 4: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 4: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 4: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 4: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB Tech-Mapping Phase 4: CPU time = 0h 0m 0s, Elapsed time = 0h 0m 0s, Peak memory usage = 54.289MB
Total Time and Memory Usage	CPU time = 0h 0m 0.148s, Elapsed time = 0h 0m 0.159s, Peak memory usage = 57.043MB

# 6.3 Resource

Resource shows the resource information. It mainly includes resource and chip utilization statistics.

The resource usage summary table counts the number of I/O PORT, I/O BUF, REG, LUT, etc. The resource utilization summary table is used to estimate the resource utilization of CFU Logics, Register, BSRAM, DSP in the current device, as shown in Figure 6-3.

Figure 6-3 Resource

## Resource

## **Resource Usage Summary**

Resource	Usage
I/O Port	16
I/O Buf	16
IBUF	11
OBUF	5
Black Box	1
test	1

## **Resource Utilization Summary**

Resource	Usage	Utilization
Logic	0(0 LUTs, 0 ALUs) / 4608	0%
Register	0 / 4020	0%
Register as Latch	0 / 4020	0%
Register as FF	0 / 4020	0%
BSRAM	0 / 10	0%

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6 Report Document 6.4 Timing

# 6.4 Timing

Timing shows the timing statistics. It includes Clock Summary, Max. Frequency Summary, and Detail Timing Paths Information, etc.

Clock Summary mainly describes the clock signals of netlists, as shown in Figure 6-4. It shows a default clock with 100MHz and a period of 10ns, a rising edge at 0ns and a falling edge at 5ns.

Figure 6-4 Timing

# **Timing**

#### **Clock Summary:**

Clock Name	Туре	Period	Frequency(MHz)	Rise	Fall	Source	Master	Object	
clk	Base	10.000	100.0	0.000	5.000			clk_ibuf/l	

Max. Frequency Summary mainly counts the time frequency of netlist file in order to measure whether the timing meets the requirements. As shown in Figure 6-5, the requested frequency is 100MHz and the actual frequency is 747.2MHz, and which meet the timing requirements. If the requirements can not be met, the specific timing path needs to be further checked.

Figure 6-5 Max Frequency Summary

#### **Max Frequency Summary:**

No.	Clock Name	Constraint	Actual Fmax	Logic Level	Entity
1	clk	100.0(MHz)	747.2(MHz)	1	TOP

Detail Timing Paths Information shows the details of timing path. The default is 5 paths, and the unit is nanoseconds. Path Summary describes the key path, nodes and delays in the netlist file, as shown in Figure 6-6; Data Arrival Path and Data Require Path are key paths, and you can see from/to nodes and fanout in Figure 6-7. Path Statistics shows the path delay information, as shown in Figure 6-8.

**Figure 6-6 Path Summary** 

#### **Detail Timing Paths Information**

Path 1

#### **Path Summary:**

Slack	8.662
Data Arrival Time	2.283
Data Required Time	10.945
From	reg2_s0
То	out2
Launch Clk	clk[R]
Latch Clk	clk[R]

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Figure 6-7 Connection Relation, Delay and Fanout Information Data Arrival Path:

AT	DELAY	TYPE	RF	FANOUT	NODE
0.000	0.000				clk
0.000	0.000	tCL	RR	1	clk_ibuf/l
0.982	0.982	tINS	RR	3	clk_ibuf/O
1.345	0.363	tNET	RR	1	reg2_s0/CLK
1.803	0.458	tC2Q	RF	3	reg2_s0/Q
2.283	0.480	tNET	FF	1	out2/D

#### Data Required Path:

AT	DELAY	TYPE	RF	FANOUT	NODE
10.000	0.000				clk
10.000	0.000	tCL	RR	1	clk_ibuf/l
10.982	0.982	tINS	RR	3	clk_ibuf/O
11.345	0.363	tNET	RR	1	out2/CLK
10.945	-0.400	tSu		1	out2

# **Figure 6-8 Path Statistics**

#### Path Statistics:

Clock Skew:	0.000
Setup Relationship:	10.000
Logic Level:	1
Arrival Clock Path Delay:	cell: 0.982, 73.009%; route: 0.363, 26.991%
Arrival Data Path Delay:	cell: 0.000, 0.000%; route: 0.480, 51.155%; tC2Q: 0.458, 48.845%
Required Clock Path Delay:	cell: 0.982, 73.009%; route: 0.363, 26.991%

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