# Signal processing presentation



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# The transfer function of a digital filter is expressed as follows:

$$H(z) = \frac{B(z)}{A(z)} = \frac{b_0 + b_1 z^{-1} + \dots + b_{n-1} z^{-(n-1)} + b_n z^{-n}}{a_0 + a_1 z^{-1} + \dots + a_{m-1} z^{-(m-1)} + a_m z^{-m}} = \frac{\sum_{i=0}^n b_i z^{-i}}{\sum_{i=0}^m a_i b^{-i}}$$

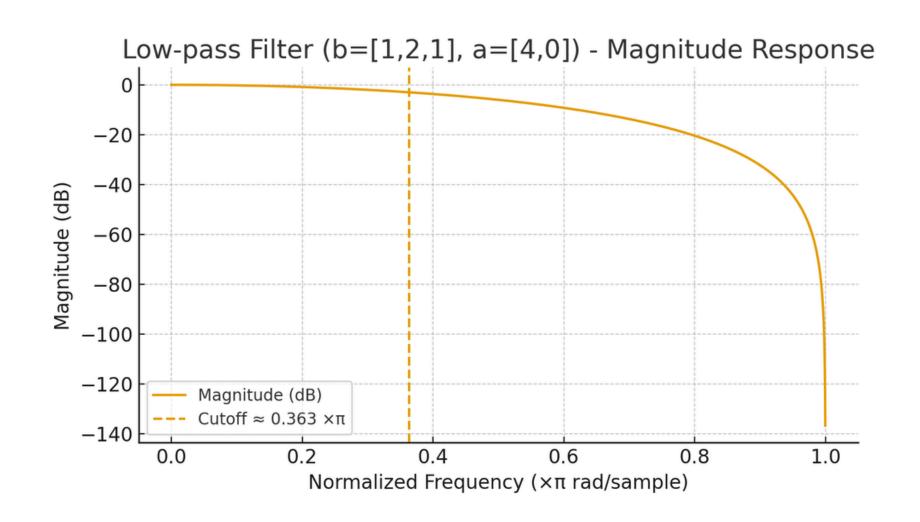
#### 1) 저역통과: b = [1, 2, 1], a = [4, 0]

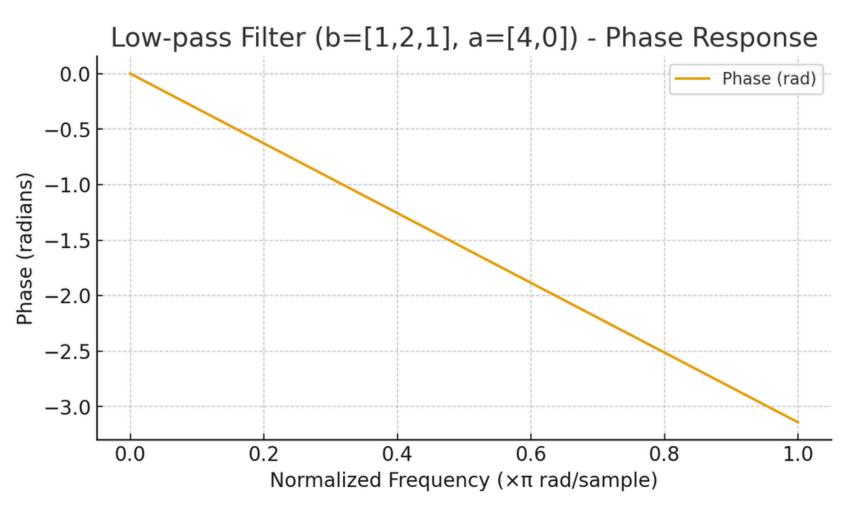
- 차분방정식:  $y[n]=rac{1}{4}\{x[n]+2x[n-1]+x[n-2]\}$  ightarrow 최근 3샘플을 부드럽게 평균(삼각 창) ightarrow 저주파(느린 변화) 통과.
- 주파수 응답(해석)

$$H_{LP}(e^{j\omega})=rac{1+2e^{-j\omega}+e^{-j2\omega}}{4}=e^{-j\omega}\,\cos^2\!\left(rac{\omega}{2}
ight)$$

- 크기  $|H|=\cos^2(\omega/2)$ :  $\omega=0$ 에서 1(=0 dB),  $\omega=\pi$ 에서 0(= $-\infty$  dB).
- 위상 ∠H ≈ −ω: 선형 위상, 군지연 ≈ 1 sample.
- 차단 주파수(-3 dB):  $\cos^2(\omega_c/2)=10^{-3/20} 
  ightarrow$   $\omega_c pprox 0.363\,\pi$  (그림의 점선). ightarrow Hz로는  $f_cpprox 0.363 imes rac{F_s}{2}pprox 0.1815\,F_s$ .

# Given a low-pass filter with b=1,2,1 and a=4,0, plot its frequency response (with the cut-off frequency showed in the figure) and its phase response. What's <u>Response</u>?





#### Problem 1

## The transfer function of a digital filter is expressed as follows:

$$H(z) = \frac{B(z)}{A(z)} = \frac{b_0 + b_1 z^{-1} + \dots + b_{n-1} z^{-(n-1)} + b_n z^{-n}}{a_0 + a_1 z^{-1} + \dots + a_{m-1} z^{-(m-1)} + a_m z^{-m}} = \frac{\sum_{i=0}^n b_i z^{-i}}{\sum_{i=0}^m a_i b^{-i}}$$

#### 2) 고역통과: b = [-1, 2, -1], a = [4, 0]

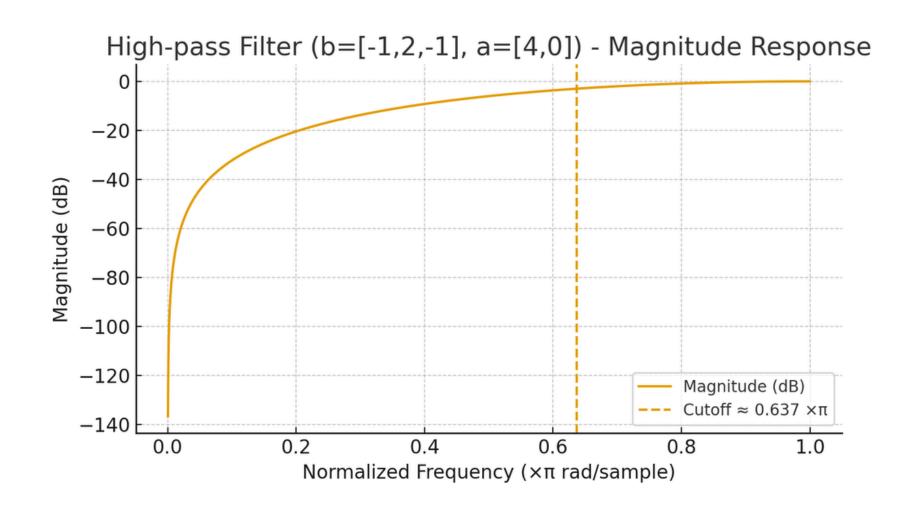
- 차분방정식:  $y[n]=\frac{1}{4}\{-x[n]+2x[n-1]-x[n-2]\}$   $\rightarrow$  \*\*이차 차분(이산 미분)\*\*에 가까움  $\rightarrow$  고주파(급격한 변화) 통과.
- 주파수 응답(해석)

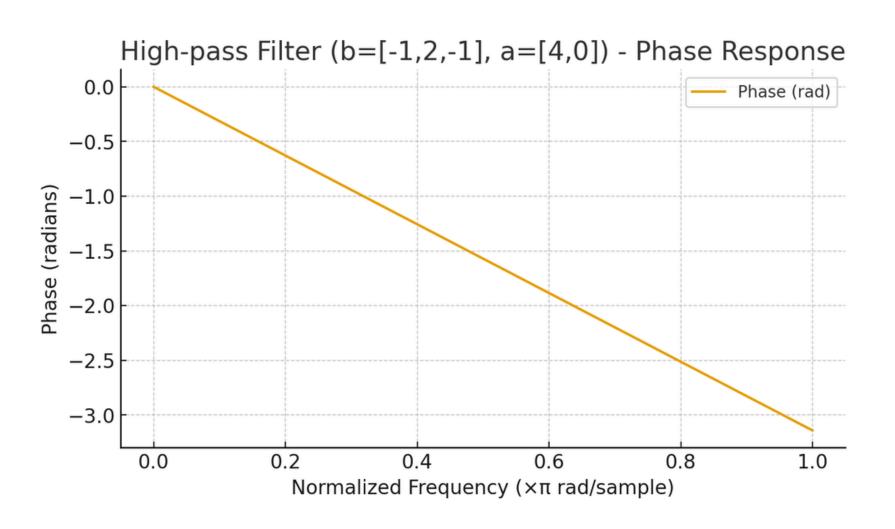
$$H_{HP}(e^{j\omega})=rac{-1+2e^{-j\omega}-e^{-j2\omega}}{4}=-e^{-j\omega}\,\sin^2\!\left(rac{\omega}{2}
ight)$$

- 크기  $|H|=\sin^2(\omega/2)$ :  $\omega=0$ 에서  $0(=-\infty$  dB),  $\omega=\pi$ 에서 1(=0 dB).
- 위상은  $-\omega$ 에 \*\*상수  $\pi$ \*\*가 더해진 형태지만(부호 때문), unwrap하면 그림처럼 거의 직선으로 보입니다(모듈로  $2\pi$ ).
- 차단 주파수(-3 dB):  $\sin^2(\omega_c/2)=10^{-3/20} o$   $\omega_c \approx 0.637\,\pi$  (그림의 점선). o Hz로는  $f_c \approx 0.637 imes rac{F_s}{2} pprox 0.3185\,F_s$ .

참고: 두 필터의 차단 주파수는 **서로 보완**되어  $0.363\pi + 0.637\pi = \pi$ 가 됩니다.

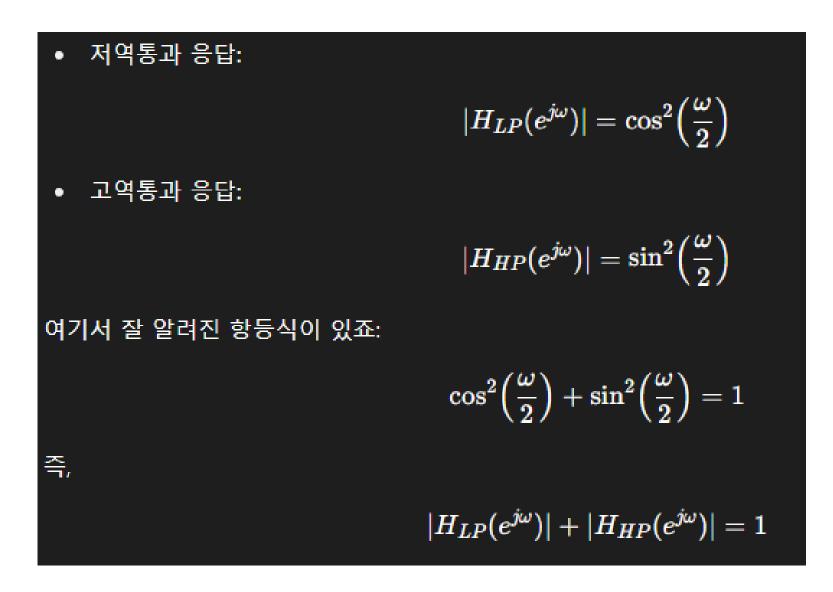
Given a high-pass filter with b=-1,2,-1 and a=4,0, plot is frequency response (with the cut-off frequency showed in the figure) and its phase response. What's <u>Response</u>?





#### **Problem 1**

### In result the solved math



### \*To recap

지금의 LPF/HPF 쌍은 수학적으로 보완 필터 (complementary filters).

- 둘의 제곱 크기 합은 항상 1 → 따라서 저역 과 고역을 더하면 중간 주파수 대역이 정확 히 이어짐
- 결국 두 개를 합치면 전 대역(All-pass) 응 답이 되며, 위상은 단순히 -ω 지연만 남음

"1"이라는 건 통과 이득(=0 dB) 에 해당. 따라서 두 응답을 합치면 전 주파수 대역을 모두 커버하게 됨. 직관적으로는 LPF + HPF = All-pass Filter (위상은 지연만 있음) 이 된다고 볼 수 있음. The mathematical expression of the 1-D CZP(Circular Zone Plate)

signal is as follows:

$$z(x) = C_w \cdot \cos \left( \pi \frac{x^2}{T} \right) + C_{offset}$$

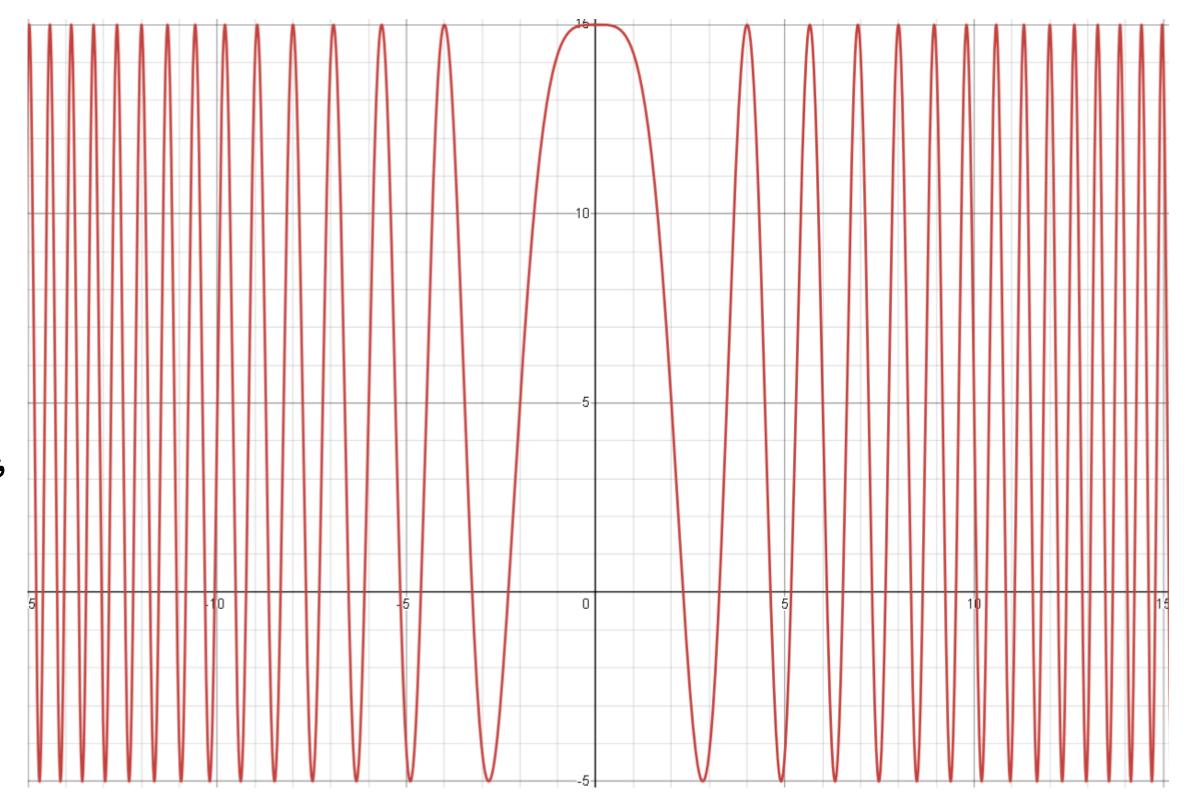
$$\cos\left(\pi\frac{x^2}{T}\right) \to \left(\pi\frac{x^2}{T}\right) = 2\pi$$

Plot the 1-D CZP signal ( $C_{w}$ ,  $C_{offset}$ , and T can be freely determined)

$$C_{w} = 10, C_{offset} = 5, T = 8$$

$$\rightarrow x = 4$$

$$z(x) = 10 \cdot \cos\left(\pi \frac{x^2}{8}\right) + 5$$



# The mathematical expression of the 2-D CZP signal is as follows:

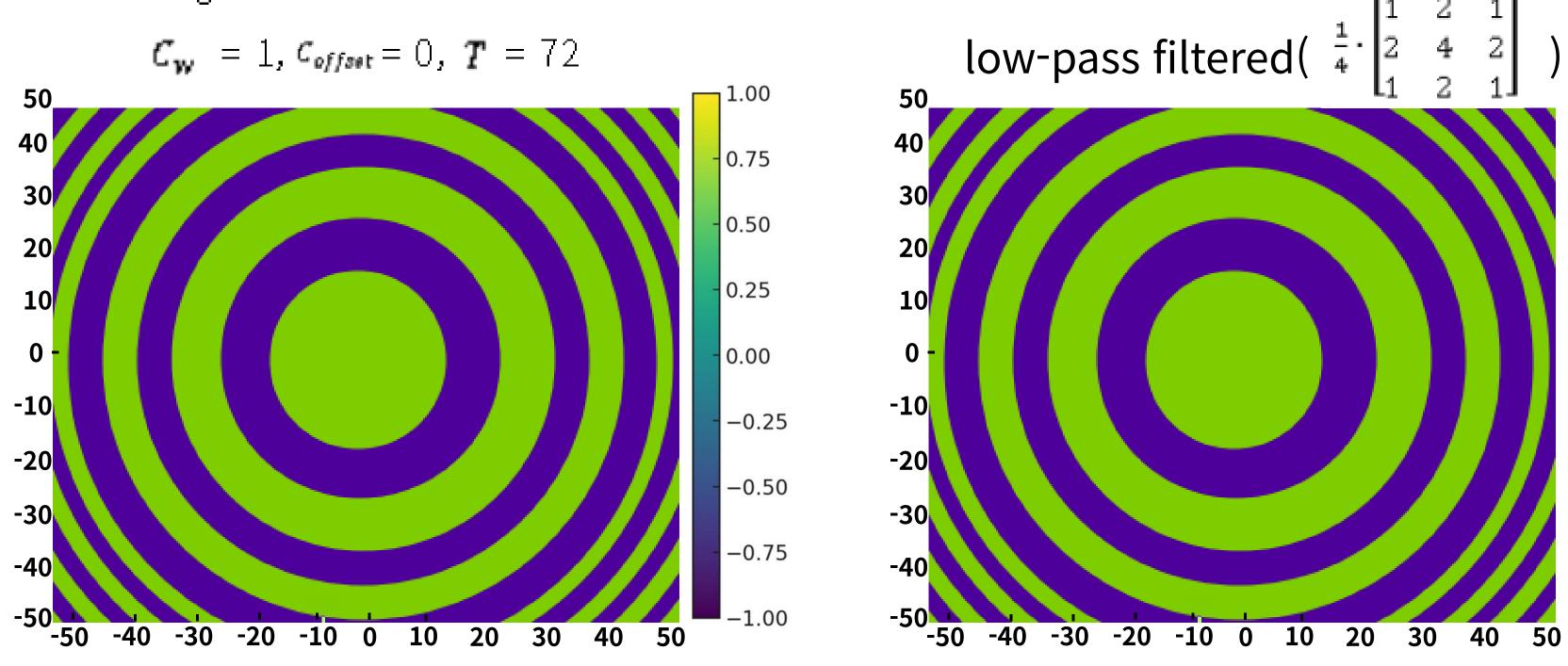
$$z(x,y) = C_w \cdot \cos\left(\pi \frac{x^2 + y^2}{T}\right) + C_{offset} + C_{offset}$$

① Plot the 2-D CZP signal ( $c_w$ ,  $c_{offset}$ , and T can be freely determined). $\leftarrow$ 

$$C_{W} = 1, C_{offset} = 0, T = 72$$

② Given a low-pass filter represented by the matrix  $\frac{1}{4} \cdot \begin{bmatrix} 1 & 2 & 1 \\ 2 & 4 & 2 \\ 1 & 2 & 1 \end{bmatrix}$ , plot the result of filtering the

CZP signal.←



③ Given a high-pass filter represented by the matrix  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$ , plot the result of filtering

the CZP signal.←

high-pass filtered ( 
$$\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$$
 ) so this phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1 \\ -2 & 4 & -2 \\ 1 & -2 & 1 \end{bmatrix}$  ) so the phase filtered (  $\frac{1}{4} \cdot \begin{bmatrix} 1 & -2 & 1$ 

#### References



- 1. Goodman, J. W. (2005). Introduction to Fourier Optics (3rd ed.). Roberts and Company Publishers.
- 2. Cao, Q., & Jahns, J. (2004). Comprehensive focusing analysis of various Fresnel zone plates. Journal of the Optical Society of America A, 21(4), 563–571.
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