Design of Over Sampling Audio DAC

Digital Signal Processing

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The Problem

Reproduction stage of digital audio system includes amplifier and speaker(s).

Preceded by digital-to-analogue conversion (DAC) typically involving:

- The DAC:
 - Takes in clocked sequence of parallel words and
 - Outputs a series of voltage levels which are constant for the duration of the sampling period;
- A low-pass filter:
 - Eliminates spectral copies inherent in any sampled signal
 - Passes only the original signal, replacing the staircase-like output from the DAC by a much smoother waveform.

Digital Source	7	DAC	k P	Filter	k V	Amplifier	7	Speaker(s)
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The Problem

- Each part has its own particular problems but not address the DAC in this unit.
- For the low-pass filter, the problem is with:

The guard bands between the original signal spectrum and its first sampling-induced duplicate

- These are very small in typical digital audio systems such as
 - The top of the audio band is at 20 kHz and
 - With a sampling frequency of 44.1 kHz,
 - The bottom end of first duplicate spectrum is only at 24.1 kHz.

This is an extremely small guard band.

A very narrow guard bands means:

high order analogue reconstruction filters

 Leading to pass-band amplitude ripple & nonlinear phase (and generally temperature and time stability issues).

Votes			

The Solution: Oversampling

Solution is to:

Numerically re-sample at a higher rate (oversampling), which eases the constraint on the analogue filter because of the larger guard band.

Interpolation: two possibilities

- Re-sampling by inserting additional zero-valued samples between original signal samples does not eliminate frequency components about the harmonics of original sample rate.
 - It just increases the sample clock rate.
- Therefore need to explicitly interpolate i.e. replace inserted zeros with values that "smoothly join the dots".

The Solution: Oversampling

Goal:

Replace difficult-to-design-and-make high order analogue filter with combination of a digital and analogue filter.

Advantage:

- Low-order analogue filter
- lacksquare ightarrow much easier to design,
- $lue{}$ ightarrow cheaper to make and more stable etc.

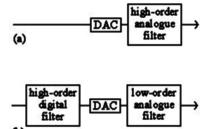
Disadvantage:

digital filter and DAC at much higher sampling frequency.

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The Solution: Oversampling

Replacing difficult-to-design-and-make high order analogue filter with combination of a digital and analogue filter.



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The Solution: Oversampling

Formally, we may define oversampling in a $\ensuremath{\mathsf{D}}/\ensuremath{\mathsf{A}}$ context as:

"numerically synthesising a sample sequence equivalent to that which would have resulted if the original analogue signal had been sampled at a much higher rate."

As such, oversampling consists of two processes:

- up-sampling: inserting additional zeros between the existing samples to increase sample rate but not changing spectral content
- interpolation: using high-order digital filter to remove intermediate spectral copies and create a large guard-band

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The Solution: Oversampling

2-times oversampling example

(Next slide)

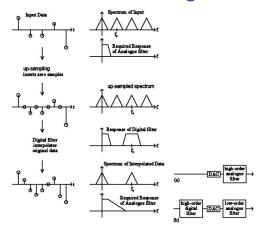
- Left-hand column: time-domain
- Right-hand column: frequency domain.

After the up-sampling process, requirement for analogue filter is unchanged. $% \begin{center} \end{center} \begin{center} \end{center}$

Only after the digital filter has removed the intermediate spectral copies that the job of the analogue filter becomers easier.

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The Solution: Oversampling



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The Solution: Oversampling

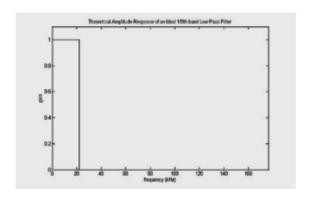
Benefits of oversampling when using realistic numbers: Example of Butterworth analogue filters with & without 8-times oversampling:

```
\mathsf{basic}\ f_s
                     44.1 kHz
                                       \mathsf{basic}\ f_s
                                                           44.1 kHz
                                     (oversampling =
                                                              8)
(oversampling
                        1)
                                       \mathsf{New}\ f_s
   New f_s
                     44.1 \; kHz
                                                          352.8 kHz
first alias freq =
                                     first alias freq =
                     24.1 kHz
                                                          332.8 kHz
                        60
                                                              5
 <u>filter order</u>
                                      filter order
    f_{3dB}
                      20 kHz
                                         f_{3dB}
                                                            30 kHz
gain @ 20kHz
                       -3 dB
                                     gain @ 20kHz
                                                           -0.07 dB
gain @ alias
                      -97 dB
                                     gain @ alias
                                                           -104 dB
```

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The Solution: Oversampling

Illustration of frequency-domain specification of 1/8th-band digital filter (suitable for 8-times oversampling)



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The Solution: Oversampling

Learning Task

Use the MATLAB program TFA_analogue_filters.m to explore the problem of analogue filter order with varying degrees of oversampling and varying gains @ 20 kHz (typically an audio application would wish for significantly less than 1dB of signal loss at 20 kHz).

- Use the source filter option;
- choose a line-spacing of 11025 (to force a 44.1 kHz basic sampling rate);
- choose a 16 bit DAC.

If you had to design a digital filter for a 16-bit DAC whose analogue filter was only a 3rd-order Butterworth filter, work out what oversampling ratio (power of 2) you would need to use.

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FIR Implementation Issues

FIR digital filter in an oversampling audio DAC is clocked at some integer multiple of the original audio sampling rate, giving it a shorter time in which to carry out its MACs.

If original audio sampled at 44.1 kHz and DAC running at 8 times oversampling then time available to calculate each output from the filter (per channel) is:

$$\frac{1}{44100 \times 8} = 2.834...\mu s$$

If the DSP chip MAC time is 20ns then to implement filtering for a stereo DAC (2 channels) each channel can only use:

$$\frac{1}{44100\times8\times20\times10^{-9}\times2} = 70.86... \text{MACs per output sample}.$$

Of course we cannot make use of a fraction of a MAC, so that means just $70\,$ MACs per output sample on each channel.

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FIR Implementation Issues

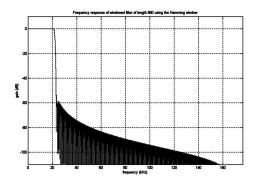
Filters of order 69 (70 coefficients), may not be enough to achieve the sharp cut-off needed.

- However, oversampled to a clock rate of 352.8 kHz means data going into the filter consists of the original audio samples with
 - seven zero-valued samples between nonzero audio samples.
- As a result 87.5% of apparent MACs are multiplying by zero,
 - which therefore contributes nothing to the output samples,
- Thus actual filter design can be order of 559 (560 coefficients) provided DSP chip (or its programmer!) is clever enough to omit the multiply-by-zero terms.

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FIR Implementation Issues

Using MATLAB program oversampling_DAC_filter.m, we can see if this gives us a good enough response. Unfortunately it doesn't:



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FIR Implementation Issues

	Filter	length	=	560
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- MACs per output sample = 70
- Window selected = Hamming (the design method used and how windows get involved is discussed later)
- \blacksquare Gain at 20kHz = 0 dB / Gain at 24.1kHz = -65.05 dB

The attenuation at 24.1 kHz (the first alias frequency) is insufficient. For a 16-bit DAC we would want an attenuation of at least 96 dB; more for a higher-precision device. This means, for this application, we need a faster DSP chip.

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FIR Implementation Issues

An order-N FIR filter obeys the finite convolution sum:

$$y[n] = \sum_{k=0}^{N} h[k]x[n-k]$$

which involves (N+1) MACs.

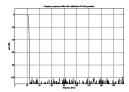
- lacksquare In a practical implementation, both input signal x[n] and filter's pulse response h[k] represented by finite precision binary numbers,
 - typically integers or a special kind of fractions where the denominator is a power of 2, e.g. 5/64 (these sort of fractions are known as dyadic rationals).
- Important to understand impact of finite precision of the coefficients:

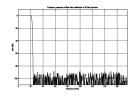
A filter may have a suitable response in simulations when its coefficients are represented as arbitrary precision floating point numbers but may fail to meet its design specification when coefficients are quantised to realistic finite precision.

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FIR Implementation Issues

Consider following two frequency responses, (FIR_coeff_precision.m) using same theoretical response (produced by oversampling_DAC_filter.m).





Left (18-bit precision coefficients), response just satisfies specification (-96 dB from 24.1 kHz upwards) but 16-bit version does not by more than 8 dB (at around 100 kHz).

This means our DSP chip must be capable of at least 18-bit multiplication, in addition to having a fast enough MAC.

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FIR Implementation Issues

Learning Task

Download the MATLAB program FIR_coeff_precision.m and save it to your MATLAB "work" directory. Use it in conjunction with oversampling_DAC_filter.m to check the above information. Then use all the MATLAB programs in this book to do a complete design to replace an 18-bit DAC in a system with a basic sampling rate of 44.1 kHz, assuming you cannot use a Butterworth filter whose order is more than 5 and you want the attenuation at 20 kHz to be no more than 0.1 dB. Remember that TFA_analogue_filters.m only allows oversampling ratios which are powers of 2 (because in that application it is related to the use of the FFT). Assume your DSP chip takes 15ns to carry out one MAC.

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Summary

- Oversampling can help ease stringent analogue filter requirements;
- Increases digital computational demand.

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