













AM3359, AM3358, AM3357, AM3356, AM3354, AM3352, AM3351

ZHCS488J-OCTOBER 2011-REVISED APRIL 2016

AM335x Sitara™ 处理器

器件概述

特性 1.1

- 高达 1GHz Sitara™ ARM® Cortex®-A8 32 位精简 指令集计算机 (RISC) 处理器
 - NEON™单指令流多数据流 (SIMD) 协处理器
 - 32KB L1 指令和 32KB 带有单位检错(奇偶校) 验)的数据缓存
 - 带有错误校正码 (ECC) 的 256KB L2 缓存
 - 176KB 片载启动 ROM
 - 64KB 专用 RAM
 - 仿真和调试 JTAG
 - 中断控制器(最多可控制 128 个中断请求)
- 片上存储器(共享 L3 RAM)
 - 64KB 通用片上存储器控制器 (OCMC) 随机存取 存储器 (RAM)
 - 可访问所有主机
 - 支持保持以实现快速唤醒
- 外部存储器接口 (EMIF)
 - mDDR(LPDDR)、DDR2、DDR3、DDR3L 控制
 - mDDR: 200MHz 时钟(400MHz 数据速率)
 - DDR2: 266MHz 时钟(532MHz 数据速率)
 - DDR3: 400MHz 时钟(800MHz 数据速率)
 - DDR3L: 400MHz 时钟(800MHz 数据速率)
 - 16 位数据总线
 - 1GB 全部可寻址空间
 - 支持一个 x16 或两个 x8 存储器件配置
 - 通用存储器控制器 (GPMC)
 - 灵活的 8 位和 16 位异步存储器接口,具有多 达七个片选(NAND、NOR、复用 NOR 和 SRAM)
 - 使用 BCH 代码,支持 4 位、8 位或 16 位 **ECC**
 - 使用海明码来支持 1 位 ECC
 - 错误定位器模块 (ELM)
 - 与 GPMC 一起使用时,可通过 BCH 算法确定 所生成的伴随多项式中数据错误的地址
 - 根据 BCH 算法, 支持 4 位、8 位和 16 位每 512 字节块错误定位
- 可编程实时单元子系统和工业通信子系统 (PRU-ICSS)
 - 支持 EtherCAT[®]、PROFIBUS、PROFINET、 EtherNet/IP™ 等协议
 - 2个可编程实时单元(PRU)
 - 32位可运行在200MHz的负载/存储RISC处理
 - 8KB 带有单位检错(奇偶校验)的指令 RAM
 - 8KB 带有单位检错(奇偶校验)的数据 RAM

- 具有64位累加器的单周期32位乘法器
- 增强型 GPIO 模块为外部信号提供移入/移出支 持以及并行锁断
- 12KB 带有单位检错(奇偶校验)的共享 RAM
- 三个 120 字节寄存器组,可被每个 PRU 访问
- 用于处理系统输入事件的中断控制器 (INTC)
- 用于将内部和外部主机连接到 PRU-ICSS 内部资 源的本地互连总线
- PRU-ICSS 内的外设:
 - 一个带有流控制引脚的通用异步收发器 (UART) 端口,支持高达 12Mbps 的数据速率
 - 一个增强型捕捉 (eCAP) 模块
 - 两个 MII 以太网端口,支持工业以太网(例如 EtherCAT)
 - 一个 MDIO 端口
- 电源、复位和时钟管理 (PRCM) 模块
 - 控制待机模式和深度休眠模式的进入和退出
 - 负责休眠排序、电源域关闭排序、唤醒排序和电 源域打开排序
 - 时钟
 - 集成了 15MHz 至 35MHz 的高频振荡器,用 于为各种系统和外设时钟生成参考时钟
 - 支持子系统和外设的单独时钟使能和禁用控 制,帮助降低功耗
 - 五个 ADPLL,用于生成系统时钟(MPU 子系 统、DDR接口、USB和外设[MMC和SD、 UART、SPI、I²C]、L3、L4、以太网、GFX [SGX530]、LCD 像素时钟)
 - 电源
 - 两个不可切换的电源域(实时时钟 [RTC] 和唤 醒逻辑 [WAKEUP])
 - 三个可切换电源域(MPU 子系统 [MPU]、 SGX530 [GFX]、外设和基础设施 [PER])
 - 执行 SmartReflex™ 2B 类,基于芯片温度、 过程变化和性能实现内核电压调节(自适应电 压调节 [AVS])
 - 动态电压频率缩放 (DVFS)
- 实时时钟 (RTC)
 - 实时日期(年、月、日和星期几)和时间(小 时、分钟和秒)信息
 - 内部 32.768kHz 振荡器, RTC 逻辑和 1.1V 内部 低压降稳压器 (LDO)
 - 独立的加电复位 (RTC_PWRONRSTn) 输入
 - 用于外部唤醒事件的专用输入引脚(EXT_ WAKEUP)
 - 可编程警报可用于生成 PRCM 内部中断 (用于唤 醒)或 Cortex-A8 内部中断(用于事件通知)



- 可编程警报可与外部输出 (PMIC_POWER_EN) 一起用来使能电源管理 IC,从而恢复非 RTC 电 源域

• 外设

- 多达两个带有集成 PHY 的 USB 2.0 高速 OTG 端口
- 多达两个工业千兆位以太网 MAC(10、100 和 1000Mbps)
 - 集成开关
 - 每个 MAC 都支持 MII、RMII、RGMII 和 MDIO 接口
 - 以太网 MAC 和交换机可独立于其它功能运行
 - IEEE 1588v2 精密时间协议 (PTP)
- 多达 2 个控制器局域网 (CAN) 端口
 - 支持 CAN 版本 2 部分 A 和 B
- 多达两个多通道音频串行端口 (McASP)
 - 高达 50MHz 的发送和接收时钟
 - 每个具有独立 TX 和 RX 时钟的 McASP 端口 对应多达四个串行数据引脚
 - 支持时分多路复用 (TDM)、内部 IC 声音 (I2S) 和类似格式
 - 支持数字音频接口传输(SPDIF、IEC60958-1 和 AES-3 格式)
 - 用于发送和接收的 FIFO 缓冲器(256 字节)
- 最多6个UART
 - 所有 UART 支持 IrDA 和 CIR 模式
 - 所有 UART 支持 RTS 和 CTS 流量控制
 - UART1 支持完整的调制解调器控制
- 多达两个主从 McSPI 串行接口
 - 最多2个芯片选择
 - 高达48 MHz
- 多达三个 MMC、SD 和 SDIO 端口
 - 1 位、4 位和 8 位 MMC、SD 和 SDIO 模式
 - MMCSD0 具有专用于 1.8V 或 3.3V 操作的电源轨
 - 高达 48MHz 的数据传输速率
 - 支持卡检测和写保护
 - 符合 MMC4.3、SD 和 SDIO 2.0 规范
- 多达三个 I2C 主从接口
 - 标准模式(高达 100kHz)
 - 快速模式(高达 400kHz)
- 多达四组通用 I/O (GPIO) 引脚
 - 每组包含 32 个 GPIO 引脚(与其他功能引脚 复用)
 - GPIO 引脚可作为中断输入(每组多达两个中断输入)
- 多达三个外部直接存储器访问 (DMA) 事件输入也可用作中断输入
- 8 个 32 位通用计时器
 - DMTIMER1 是用于操作系统 (OS) 节拍的 1ms 计时器
 - DMTIMER4-DMTIMER7 为引脚输出
- 一个安全装置计时器

- SGX530 3D 图形引擎

- 拼图架构每秒可提供最多 2000 万个多边形
- 通用可扩展着色引擎 (USSE) 是一款包含像素和顶点着色功能的多线程引擎
- 超过 Microsoft VS3.0、PS3.0 和 OGL2.0 的 高级着色功能集
- Direct3D Mobile、OGL-ES 1.1 和 2.0、
 OpenVG 1.0 以及 OpenMax 的行业标准 API 支持
- 精细的任务切换、负载均衡和电源管理
- 高级几何 DMA 驱动型操作,最大程度地减少 CPU 交互
- 可编程高质量图像防锯齿
- 用于统一存储器架构中操作系统运行的完全虚拟化存储器寻址
- LCD 控制器
 - 多达 24 位数据输出; 每像素 8 位 (RGB)
 - 分辨率最高可达 2048 × 2048 (具有最高 126MHz 的像素时钟)
 - 集成 LCD 接口显示驱动器 (LIDD) 控制器
 - 集成光栅控制器
 - 集成 DMA 引擎可通过中断或固件计时器从外部帧缓冲器获取数据,无需加重处理器的负担
 - 512 字深内部 FIFO
 - 支持的显示类型:
 - 字符显示器 使用 LIDD 控制器对这些显示 器进行编程
 - 无源矩阵 LCD 显示-使用 LCD 光栅显示控制器来为到无源显示的持续图形刷新提供定时和数据
 - 有源矩阵 LCD 显示-使用外部帧缓冲器空间和内部 DMA 引擎来驱动到控制面板的流数据
- 12 位逐次逼近寄存器 (SAR) ADC
 - 每秒采集 200K 个样本
 - 可从 8:1 模拟开关复用的八个模拟输入中任意 选择输入
 - 可配置为用作 4 线、5 线或 8 线电阻式触摸屏控制器 (TSC) 接口
- 多达三个 32 位 eCAP 模块
 - 可配置为三个捕捉输入或者三个备用 PWM 输出
- 多达三个增强型高分辨率 PWM 模块 (eHRPWM)
 - 具有时间和频率控制功能的 16 位专用时基计 数器
 - 可配置为6个单端,6个双边对称,或者3 个双边不对称输出
- 多达 3 个 32 位增强型正交编码脉冲 (eQEP) 模 块
- 器件标识
 - 包含电子熔丝组 (FuseFarm), 其中一些位厂家可编程
 - 生产 ID
 - 器件部件号(唯一的 JTAG ID)



- 设备版本(可由主机 ARM 读取)
- 调试接口支持
 - 用于 ARM(Cortex-A8 和 PRCM)和 PRU-ICSS 调试的 JTAG 和 cJTAG
 - 支持器件边界扫描
 - 支持 IEEE1500
- DMA
 - 片上增强型 DMA 控制器 (EDMA) 搭载三个第三 方传送控制器 (TPTC) 和一个第三方通道控制器 (TPCC),支持多达 64 个可编程逻辑通道和 8 个 QDMA 通道。EDMA 用于:
 - 向/从片上存储器传送
 - 向/从外部存储器(EMIF、GPMC 和从外设) 传送
- 处理器间通信 (IPC)
 - 集成了基于硬件的 IPC 邮箱,以及用于 Cortex-A8、PRCM 和 PRU-ICSS 之间进程同步的

Spinlock

- 生成中断的邮箱寄存器
 - 4 个初启程序(Cortex-A8, PRCM, PRU0, PRU1)
- 自旋锁具有 128 个软件指定的锁寄存器
- 安全性
 - 加密硬件加速器 (AES、SHA、RNG)
 - 安全启动
- 启动模式
 - 通过锁存在 PWRONRSTn 复位输入引脚上升沿的启动配置引脚来选择启动模式
- 封装:
 - 298 引脚 S-PBGA-N298 Via Channel 封装 (后缀 ZCE), 0.65mm 焊球间距
 - 324 引脚 S-PBGA-N324 封装 (后缀 ZCZ), 0.80mm 焊球间距



1.2 应用范围

- 游戏外设
- 家庭和工业自动化
- 消费类医疗器械
- 打印机
- 智能收费系统

- 联网贩售机
- 电子秤
- 教育控制台
- 高级玩具

1.3 说明

AM335x 微处理器基于 ARM Cortex-A8 处理器,在图像、图形处理、外设以及 EtherCAT 和 PROFIBUS 等工业接口选项方面得到了增强。 该器件支持高级操作系统 (HLOS)。 Linux[®]和 Android™可从德州仪器 (TI) 免费获取。

AM335x 微处理器 包含 功能框图 中显示的子系统和以下 简要 说明:

微处理器单元 (MPU) 子系统基于 ARM Cortex-A8 处理器, PowerVR SGX™图形加速器子系统提供 3D 图形加速功能以支持显示和游戏特效。

可编程实时单元子系统和工业通信子系统 (PRU-ICSS) 与 ARM 内核彼此独立,允许单独操作和计时,以实现更高的效率和灵活性。PRU-ICSS 支持更多外设接口和 EtherCAT、PROFINET、EtherNet/IP、PROFIBUS、Ethernet Powerlink、Sercos 等实时协议。此外,凭借 PRU-ICSS 的可编程特性及其对引脚、事件和所有片上系统 (SoC) 资源的访问权限,该子系统可以灵活地实现快速实时响应、专用数据处理操作以及自定义外设接口,并减轻 SoC 其他处理器内核的任务负载。

器件信息(1)

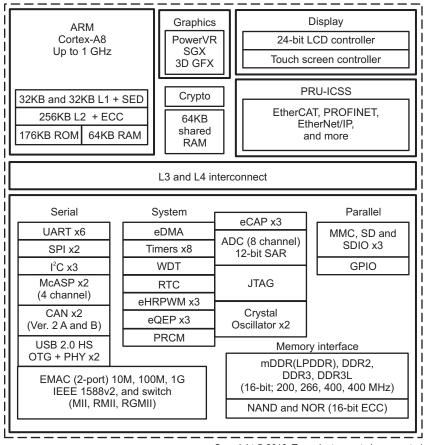
BE 11 19.0.									
器件型号	封装	封装尺寸							
AM3359ZCZ	NFBGA (324)	15.0mm x 15.0mm							
AM3358ZCZ	NFBGA (324)	15.0mm x 15.0mm							
AM3357ZCZ	NFBGA (324)	15.0mm x 15.0mm							
AM3356ZCZ、AM3356ZCE	NFBGA (324)、NFBGA (298)	15.0mm × 15.0mm \ 13.0mm × 13.0mm							
AM3354ZCZ、AM3354ZCE	NFBGA (324)、NFBGA (298)	15.0mm × 15.0mm \ 13.0mm × 13.0mm							
AM3352ZCZ、AM3352ZCE	NFBGA (324)、NFBGA (298)	15.0mm × 15.0mm \ 13.0mm × 13.0mm							
AM3351ZCE	NFBGA (298)	13.0mm x 13.0mm							

(1) 更多信息,请参见节9,机械、封装和可订购产品信息。



1.4 功能框图

图 1-1 给出了 Am335x 微处理器功能框图。



Copyright © 2016, Texas Instruments Incorporated

图 1-1. AM335x 功能框图



内容

1	器件	概述	1	7.2	Recommended Clock and Control Signal Transition	n
	1.1	特性	1		Behavior	<u>116</u>
	1.2	应用范围	4	7.3	OPP50 Support	<u>116</u>
	1.3		_	7.4	Controller Area Network (CAN)	117
	1.4	功能框图	_	7.5	DMTimer	118
2		历史记录	_	7.6	Ethernet Media Access Controller (EMAC) and	
- ว		ce Comparison	_		Switch	<u>119</u>
•	3.1	Related Products	_	7.7	External Memory Interfaces	<u>127</u>
4		ninal Configuration and Functions	_	7.8	I ² C	<u>190</u>
+		-		7.9	JTAG Electrical Data and Timing	<u>192</u>
	4.1	Pin Diagram		7.10	LCD Controller (LCDC)	<u>193</u>
	4.2	Pin Attributes		7.11	Multichannel Audio Serial Port (McASP)	209
_	4.3	Signal Descriptions		7.12	Multichannel Serial Port Interface (McSPI)	214
5	•		<u>80</u>	7.13	Multimedia Card (MMC) Interface	220
	5.1	Absolute Maximum Ratings		7.14	Programmable Real-Time Unit Subsystem and	
	5.2	ESD Ratings			Industrial Communication Subsystem (PRU-ICSS)	223
	5.3	Power-On Hours (POH)		7.15	Universal Asynchronous Receiver Transmitter	
	5.4	Operating Performance Points (OPPs)			(UART)	
	5.5	Recommended Operating Conditions	<u>85</u> 8	Devi	ce and Documentation Support	<u>235</u>
	5.6	Power Consumption Summary	<u>87</u>	8.1	Device Nomenclature	235
	5.7	DC Electrical Characteristics	<u>89</u>	8.2	Tools and Software	236
	5.8	Thermal Resistance Characteristics for ZCE and		8.3	Documentation Support	<u>240</u>
		ZCZ Packages	<u>93</u>	8.4	Related Links	<u>243</u>
	5.9	•	94	8.5	Community Resources	<u>243</u>
	5.10	Touch Screen Controller and Analog-to-Digital	07	8.6	商标	244
_	_	Subsystem Electrical Parameters		8.7	静电放电警告	244
6			<u>99</u>	8.8	术语表	244
	6.1	• •	99 9	Mec	hanical, Packaging, and Orderable	
	6.2	Clock Specifications 1			mation	<u>245</u>
7	_	oheral Information and Timings <u>1</u>		9.1	Via Channel	<u>245</u>
	7.1	Parameter Information 1	<u>16</u>	9.2	Packaging Information	245
						_



2 修订历史记录

注: 之前版本的页码可能与当前版本有所不同。

Cha	anges from Revision I (December 2015) to Revision J	Page
	• 已添加 安全启动至安全特性列表	3
	• 己添加 extended temperature range for the AM3351 device in 表 3-1	
	• 已添加 节 3.1, Related Products	9
	Reformatted and added content to 节 8, Device and Documentation Support	235



3 Device Comparison

表 3-1 lists the features supported across different AM335x devices.

表 3-1. Device Features Comparison

FUNCTION	AM3351	AM3352	AM3354	AM3356	AM3357	AM3358	AM3359
ARM Cortex-A8	Yes	Yes	Yes	Yes	Yes	Yes	Yes
Frequency ⁽¹⁾	300 MHz 600 MHz	300 MHz 600 MHz 800 MHz 1000 MHz	600 MHz 800 MHz 1000 MHz	300 MHz 600 MHz 800 MHz	300 MHz 600 MHz 800 MHz	600 MHz 800 MHz 1000 MHz	600 MHz 800 MHz
MIPS ⁽²⁾	600 1200	600 1200 1600 2000	1200 1600 2000	600 1200 1600	600 1200 1600	1200 1600 2000	1200 1600
On-chip L1 cache	64KB	64KB	64KB	64KB	64KB	64KB	64KB
On-chip L2 cache	256KB	256KB	256KB	256KB	256KB	256KB	256KB
Graphics accelerator (SGX530)	_	_	3D	_	_	3D	3D
Hardware acceleration	Crypto accelerator	Crypto accelerator	Crypto accelerator	Crypto accelerator	Crypto accelerator	Crypto accelerator	Crypto accelerator
Programmable real-time unit subsystem and industrial communication subsystem (PRU-ICSS)	_	_	_	Features including basic Industrial protocols; ZCE: Limited PRU I/Os pinned out	Features including all Industrial protocols	Features including basic Industrial protocols	Features including all Industrial protocols
On-chip memory	128KB	128KB	128KB	128KB	128KB	128KB	128KB
Display options	LCD	LCD	LCD	LCD	LCD	LCD	LCD
General-purpose memory	1 16-bit (GPMC, NAND flash, NOR flash, SRAM)	1 16-bit (GPMC, NAND flash, NOR flash, SRAM)	1 16-bit (GPMC, NAND flash, NOR flash, SRAM)	1 16-bit (GPMC, NAND flash, NOR flash, SRAM)			
DRAM ⁽³⁾	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)	1 16-bit (LPDDR-400, DDR2-532, DDR3-800)
Universal serial bus (USB)	ZCE: 1 port	ZCE: 1 port ZCZ: 2 ports	ZCE: 1 port ZCZ: 2 ports	ZCE: 1 port ZCZ: 2 ports	No ZCE Available ZCZ: 2 ports	No ZCE Available ZCZ: 2 ports	No ZCE Available ZCZ: 2 ports
Ethernet media access controller (EMAC) with 2-port switch	10/100/1000 ZCE: 1 port	10/100/1000 ZCE: 1 port ZCZ: 2 ports	10/100/1000 ZCE: 1 port ZCZ: 2 ports	10/100/1000 ZCE: 1 port ZCZ: 2 ports	10/100/1000 No ZCE Available ZCZ: 2 ports	10/100/1000 No ZCE Available ZCZ: 2 ports	10/100/1000 No ZCE Available ZCZ: 2 ports
Multimedia card (MMC)	3	3	3	3	3	3	3
Controller-area network (CAN)	_	2	2	2	2	2	2
Universal asynchronous receiver and transmitter (UART)	6	6	6	6	6	6	6
Analog-to-digital converter (ADC)	8-ch 12-bit	8-ch 12-bit	8-ch 12-bit	8-ch 12-bit	8-ch 12-bit	8-ch 12-bit	8-ch 12-bit
Enhanced high-resolution PWM modules (eHRPWM)	3	3	3	3	3	3	3
Enhanced capture modules (eCAP)	3	3	3	3	3	3	3
Enhanced quadrature encoder pulse (eQEP)	3	3	3	3	3	3	3
Real-time clock (RTC)	1	1	1	1	1	1	1
Inter-integrated circuit (I ² C)	3	3	3	3	3	3	3



表 3-1. Device Features Comparison (continued)

FUNCTION	AM3351	AM3352	AM3354	AM3356	AM3357	AM3358	AM3359
Multichannel audio serial port (McASP)	2	2	2	2	2	2	2
Multichannel serial port interface (McSPI)	2	2	2	2	2	2	2
Enhanced direct memory access (EDMA)	64-Ch	64-Ch	64-Ch	64-Ch	64-Ch	64-Ch	64-Ch
Input/output (I/O) supply	1.8 V, 3.3 V	1.8 V, 3.3 V	1.8 V, 3.3 V	1.8 V, 3.3 V	1.8 V, 3.3 V	1.8 V, 3.3 V	1.8 V, 3.3 V
Operating temperature range	0 to 90°C –40 to 105°C	-40 to 125°C ⁽⁴⁾ -40 to 105°C -40 to 90°C 0 to 90°C	-40 to 105°C -40 to 90°C 0 to 90°C	-40 to 105°C -40 to 90°C 0 to 90°C	-40 to 105°C -40 to 90°C	-40 to 105°C -40 to 90°C 0 to 90°C	-40 to 105°C -40 to 90°C

- (1) Frequencies listed correspond to silicon revision 2.x. Earlier silicon revisions support 275 MHz, 500 MHz, 600 MHz, and 720 MHz.
- (2) MIPS listed correspond to silicon revision 2.x. Earlier silicon revisions support 560, 1000, 1200, and 1440.
- (3) DRAM speeds listed are data rates.
- (4) Industrial extended temperature only supported for 300-MHz and 600-MHz frequencies.

3.1 Related Products

For information about other devices in this family of products, see the following links:

Sitara Processors Scalable processors based on ARM Cortex-A cores with flexible peripherals, connectivity and unified software support – perfect for sensors to servers.

TI's ARM Cortex-A8 Advantage The ARM Cortex-A8 core is highly-optimized by ARM for performance and power efficiency. With the ability to scale in speed from 300 MHz to 1.35 GHz, the ARM Cortex-A8-based processor can meet the requirements for power optimized devices with a power budget of less than the Cortex-A8 core a dual-issue superscalar, achieving twice the instructions executed per clock cycle at 2 DMIPS/MHz.

AM335x Sitara Processors Scalable ARM Cortex-A8-based core from 300 MHz up to 1 GHz, 3D graphics option for enhanced user interface, dual-core PRU-ICSS for industrial Ethernet protocols and position feedback control, and premium secure boot option.

Companion Products for AM335x Sitara Processors Review products that are frequently purchased or used with this product.

TI Designs for AM335x Sitara Processors The TI Designs Reference Design Library is a robust reference design library spanning analog, embedded processor and connectivity. Created by TI experts to help you jump start your system design, all TI Designs include schematic or block diagrams, BOMs and design files to speed your time to market. Search and download designs at ti.com/tidesigns.

4 Terminal Configuration and Functions

4.1 Pin Diagram

注

The terms 'ball', 'pin', and 'terminal' are used interchangeably throughout the document. An attempt is made to use 'ball' only when referring to the physical package.

4.1.1 ZCE Package Pin Maps (Top View)

The pin maps that follow show the pin assignments on the ZCE package in three sections (left, middle, and right).

Table 4-1. ZCE Pin Map [Section Left - Top View]

	Α	В	С	D	E	F
19	VSS	I2C0_SCL	UART1_TXD	UART1_RTSn	UARTO_RXD	UART0_CTSn
18	SPI0_SCLK	SPI0_D0	I2C0_SDA	UART1_RXD	ECAP0_IN_PWM0_OUT	UART0_RTSn
17	SPI0_CS0	SPI0_D1	EXTINTn	xxxx	UART1_CTSn	UART0_TXD
16	WARMRSTn	SPI0_CS1	xxxx	xxxx	XXXX	VDDS
15	EMUO	XDMA_EVENT_INTR1	XDMA_EVENT_INTR0	xxxx	PWRONRSTn	XXXX
14	TDO	тск	TMS	EMU1	XXXX	VDDSHV6
13	TRSTn	TDI	CAP_VBB_MPU	CAP_VDD_SRAM_MPU	VDDSHV6	VSS
12	AIN7	AIN5	VDDS_SRAM_MPU_BB	VDDS	VDDSHV6	VSS
11	AIN1	AIN3	xxxx	xxxx	VDDSHV6	VDD_CORE
10	AIN6	CAP_VDD_SRAM_CORE	VDDS_SRAM_CORE_BG	VSS	VSS	xxxx
9	VREFP	VREFN	xxxx	xxxx	VSS	VDD_CORE
8	AIN2	AIN0	AIN4	VSSA_ADC	VSS	VSS
7	RTC_KALDO_ENn	RTC_PWRONRSTn	PMIC_POWER_EN	VDDA_ADC	VSS	VSS
6	RTC_XTALIN	RESERVED	VDDS_RTC	CAP_VDD_RTC	XXXX	VSS
5	RTC_XTALOUT	EXT_WAKEUP	VDDS_PLL_DDR	XXXX	DDR_A4	XXXX
4	DDR_WEn	DDR_BA2	xxxx	xxxx	xxxx	DDR_A12
3	DDR_BA0	DDR_A3	DDR_A8	xxxx	DDR_A15	DDR_A0
2	DDR_A5	DDR_A9	DDR_CK	DDR_A7	DDR_A10	DDR_RASn
1	VSS	DDR_A6	DDR_CKn	DDR_A2	DDR_BA1	DDR_CASn





ZCE Pin Map [Section Middle - Top View]

	G	н	J	К	L	М
19	MMC0_CLK	MMC0_DAT3	MII1_COL	MII1_RX_ER	MII1_RX_DV	MII1_RX_CLK
18	MMC0_DAT0	MMC0_DAT2	MII1_CRS	RMII1_REF_CLK	MII1_TXD0	MII1_TXD1
17	MMC0_CMD	MMC0_DAT1	xxxx	MII1_TX_EN	xxxx	MII1_TXD3
16	USB0_DRVVBUS	VDDS_PLL_MPU	xxxx	VDD_CORE XXXX		VDDS
15	VDDSHV4	VDDSHV4	VSS	VDD_CORE VSS		VDDSHV5
14	xxxx	VDDSHV4	VSS	xxxx	VSS	VDDSHV5
13	xxxx	VDD_CORE	VDD_CORE	XXXX	VDD_CORE	VDD_CORE
12	VSS	VDD_CORE	VDD_CORE	VSS	VDD_CORE	VDD_CORE
11	VDD_CORE	VSS	VSS	VSS	vss vss	
10	xxxx	VSS	xxxx	XXXX	xxxx	VSS
9	VDD_CORE	VSS	VSS	VSS	VSS	VSS
8	VSS	VDD_CORE	VDD_CORE	VSS	VDD_CORE	VDD_CORE
7	xxxx	VDD_CORE	VDD_CORE	xxxx	VDD_CORE	VDD_CORE
6	xxxx	VDDS_DDR	VSS	xxxx	VSS	VDDS_DDR
5	VDDS_DDR	VDDS_DDR	VSS	VDDS_DDR	VSS	VDDS_DDR
4	DDR_A11	DDR_VREF	xxxx	VDDS_DDR	XXXX	DDR_D11
3	DDR_CKE	DDR_A14	xxxx	DDR_DQM1	xxxx	DDR_D10
2	DDR_RESETn	DDR_CSn0	DDR_A1	DDR_D8	DDR_DQSn1	DDR_D12
1	DDR_ODT	DDR_A13	DDR_VTP	DDR_D9	DDR_DQS1	DDR_D13





ZCE Pin Map [Section Right - Top View]

	N	Р	R	т	U	v	w
19	MII1_TX_CLK	MII1_RXD1	MDC	USB0_VBUS	USB0_DP	USB0_ID	VSS
18	MII1_TXD2	MII1_RXD0	VDDA3P3V_USB0	USB0_CE	USB0_DM	GPMC_BEn1	GPMC_WPn
17	MII1_RXD3	MDIO	VDDA1P8V_USB0	xxxx	GPMC_CSn3	GPMC_AD15	GPMC_AD14
16	MII1_RXD2	VSSA_USB	xxxx	xxxx	xxxx	GPMC_CLK	GPMC_AD9
15	VDDSHV5	xxxx	GPMC_WAIT0	xxxx	GPMC_CSn2	GPMC_AD8	GPMC_AD7
14	xxxx	VSS	xxxx	VDDS	GPMC_AD6	GPMC_CSn1	GPMC_AD5
13	xxxx	VSS	VDDSHV1	GPMC_AD13	GPMC_AD12	GPMC_AD4	GPMC_AD3
12	VSS	VSS	VDDSHV1	GPMC_AD10	GPMC_AD11	GPMC_AD2	XTALOUT
11	VDD_CORE	VDD_CORE	VDDSHV1	xxxx	xxxx	VSS_OSC	XTALIN
10	xxxx	xxxx	VSS	VSS	VDDS_OSC GPMC_ADVn_ALE		GPMC_AD0
9	VDD_CORE	VDD_CORE	VDDSHV1	xxxx	xxxx	GPMC_AD1	GPMC_OEn_REn
8	VSS	VSS	VDDSHV1	VDDS_PLL_CORE_LCD	GPMC_WEn	GPMC_BEn0_CLE	GPMC_CSn0
7	xxxx	VSS	VDDSHV6	LCD_HSYNC	LCD_VSYNC	LCD_DATA15	LCD_AC_BIAS_EN
6	xxxx	VDDSHV6	xxxx	VDDS	LCD_DATA13	LCD_DATA12	LCD_DATA14
5	VDDS_DDR	xxxx	VPP	xxxx	LCD_DATA10	LCD_DATA11	LCD_PCLK
4	DDR_D0	DDR_D1	xxxx	xxxx	xxxx	LCD_DATA8	LCD_DATA9
3	DDR_DQM0	DDR_D4	DDR_D7	xxxx	LCD_DATA7	LCD_DATA6	LCD_DATA5
2	DDR_D14	DDR_D2	DDR_DQSn0	DDR_D6	LCD_DATA1	LCD_DATA3	LCD_DATA4
1	DDR_D15	DDR_D3	DDR_DQS0	DDR_D5	LCD_DATA0	LCD_DATA2	VSS





4.1.2 ZCZ Package Pin Maps (Top View)

The pin maps that follow show the pin assignments on the ZCZ package in three sections (left, middle, and right).



ZCZ Pin Map [Section Left - Top View]

	Α	В	С	D E		F
18	VSS	EXTINTn	ECAP0_IN_PWM0_OUT	UART1_CTSn	UART0_CTSn	MMC0_DAT2
17	SPI0_SCLK	SPI0_D0	I2C0_SDA	UART1_RTSn	UART0_RTSn	MMC0_DAT3
16	SPI0_CS0	SPI0_D1	I2C0_SCL	UART1_RXD	UART0_TXD	USB0_DRVVBUS
15	XDMA_EVENT_INTR0	PWRONRSTn	SPI0_CS1	UART1_TXD	UARTO_RXD	USB1_DRVVBUS
14	MCASP0_AHCLKX	EMU1	EMU0	XDMA_EVENT_INTR1	VDDS	VDDSHV6
13	MCASP0_ACLKX	MCASP0_FSX	MCASP0_FSR	MCASP0_AXR1	VDDSHV6	VDD_MPU
12	TCK	MCASP0_ACLKR	MCASP0_AHCLKR	MCASP0_AXR0	MCASP0_AXR0 VDDSHV6	
11	TDO	TDI	TMS	CAP_VDD_SRAM_MPU	VDDSHV6	VDD_MPU
10	WARMRSTn	TRSTn	CAP_VBB_MPU	VDDS_SRAM_MPU_BB	VDDSHV6	VDD_MPU
9	VREFN	VREFP	AIN7	CAP_VDD_SRAM_CORE	VDDS_SRAM_CORE_BG	VDDS
8	AIN6	AIN5	AIN4	VDDA_ADC	VSSA_ADC	VSS
7	AIN3	AIN2	AIN1	VDDS_RTC	VDDS_PLL_DDR	VDD_CORE
6	RTC_XTALIN	AIN0	PMIC_POWER_EN	CAP_VDD_RTC	VDDS	VDD_CORE
5	VSS_RTC	RTC_PWRONRSTn	EXT_WAKEUP	DDR_A6	VDDS_DDR	VDDS_DDR
4	RTC_XTALOUT	RTC_KALDO_ENn	DDR_BA0	DDR_A8	DDR_A2	DDR_A10
3	RESERVED	DDR_BA2	DDR_A3	DDR_A15	DDR_A12	DDR_A0
2	VDD_MPU_MON	DDR_WEn	DDR_A4	DDR_CK	DDR_A7	DDR_A11
1	VSS	DDR_A5	DDR_A9	DDR_CKn	DDR_BA1	DDR_CASn





ZCZ Pin Map [Section Middle - Top View]

	G	н	J	к	L	М
18	MMC0_CMD	RMII1_REF_CLK	MII1_TXD3	MII1_TX_CLK	MII1_RX_CLK	MDC
17	MMC0_CLK	MII1_CRS	MII1_RX_DV	MII1_TXD0	MII1_RXD3	MDIO
16	MMC0_DAT0	MII1_COL	MII1_TX_EN	MII1_TXD1	MII1_RXD2	MII1_RXD0
15	MMC0_DAT1	VDDS_PLL_MPU	MII1_RX_ER	MII1_TXD2	MII1_RXD1	USB0_CE
14	VDDSHV6	VDDSHV4	VDDSHV4	VDDSHV5	VDDSHV5	VSSA_USB
13	VDD_MPU	VDD_MPU	VDD_MPU	VDDS	VSS	VDD_CORE
12	VSS	VSS	VDD_CORE	VDD_CORE	VSS	VSS
11	VSS	VDD_CORE	VSS	VSS	VSS	VDD_CORE
10	VDD_CORE	VSS	VSS	VSS	VSS	VSS
9	VSS	VSS	VSS	VSS	VDD_CORE	VSS
8	VSS	VSS	VSS	VDD_CORE	VDD_CORE	VSS
7	VDD_CORE	VSS	VSS	VSS	VDD_CORE	VSS
6	VDD_CORE	VSS	VSS	VDD_CORE	VDD_CORE	VSS
5	VDDS_DDR	VDDS_DDR	VDDS_DDR	VDDS_DDR	VDDS_DDR	VPP
4	DDR_RASn	DDR_A14	DDR_VREF	DDR_D12	DDR_D14	DDR_D1
3	DDR_CKE	DDR_A13	DDR_VTP	DDR_D11	DDR_D13	DDR_D0
2	DDR_RESETn	DDR_CSn0	DDR_DQM1	DDR_D10	DDR_DQSn1	DDR_DQM0
1	DDR_ODT	DDR_A1	DDR_D8	DDR_D9	DDR_DQS1	DDR_D15





ZCZ Pin Map [Section Right - Top View]

	N	P	R T U		U	v	
18	USB0_DM	USB1_CE	USB1_DM	USB1_VBUS	GPMC_BEn1	VSS	
17	USB0_DP	USB1_ID	USB1_DP	GPMC_WAIT0	GPMC_WPn	GPMC_A11	
16	VDDA1P8V_USB0	USB0_ID	VDDA1P8V_USB1	GPMC_A10	GPMC_A9	GPMC_A8	
15	VDDA3P3V_USB0	USB0_VBUS	VDDA3P3V_USB1	DA3P3V_USB1 GPMC_A7 GPMC_A6		GPMC_A5	
14	VSSA_USB	VDDS	GPMC_A4 GPMC_A3 GPMC_A2		GPMC_A1		
13	VDD_CORE	VDDSHV3	GPMC_A0 GPMC_CSn3 GPMC_AD15		GPMC_AD14		
12	VDD_CORE	VDDSHV3	GPMC_AD13	GPMC_AD12 GPMC_AD11		GPMC_CLK	
11	VSS	VDDSHV2	VDDS_OSC	GPMC_AD10 XTALOUT		VSS_OSC	
10	VSS	VDDSHV2	VDDS_PLL_CORE_LCD	GPMC_AD9	GPMC_AD8	XTALIN	
9	VDD_CORE	VDDS	GPMC_AD6	GPMC_AD7	GPMC_CSn1	GPMC_CSn2	
8	VDD_CORE	VDDSHV1	GPMC_AD2	GPMC_AD3	GPMC_AD4	GPMC_AD5	
7	VSS	VDDSHV1	GPMC_ADVn_ALE	GPMC_OEn_REn	GPMC_AD0	GPMC_AD1	
6	VDDS	VDDSHV6	LCD_AC_BIAS_EN	GPMC_BEn0_CLE	GPMC_WEn	GPMC_CSn0	
5	VDDSHV6	VDDSHV6	LCD_HSYNC	LCD_DATA15	LCD_VSYNC	LCD_PCLK	
4	DDR_D5	DDR_D7	LCD_DATA3	LCD_DATA7	LCD_DATA11	LCD_DATA14	
3	DDR_D4	DDR_D6	LCD_DATA2	LCD_DATA6 LCD_DATA10		LCD_DATA13	
2	DDR_D3	DDR_DQSn0	LCD_DATA1	LCD_DATA5	LCD_DATA9	LCD_DATA12	
1	DDR_D2	DDR_DQS0	LCD_DATA0	LCD_DATA4	LCD_DATA8	VSS	





4.2 Pin Attributes

The AM335x and AMIC110 Sitara Processors Technical Reference Manual and this document may reference internal signal names when discussing peripheral input and output signals because many of the AM335x package terminals can be multiplexed to one of several peripheral signals. The following table has a Pin Name column that lists all device terminal names and a Signal Name column that lists all internal signal names multiplexed to each terminal which provides a cross reference of internal signal names to terminal names. This table also identifies other important terminal characteristics.

- 1. BALL NUMBER: Package ball numbers associated with each signals.
- 2. PIN NAME: The name of the package pin or terminal.
 - Note: The table does not take into account subsystem terminal multiplexing options.
- 3. **SIGNAL NAME**: The signal name for that pin in the mode being used.
- 4. MODE: Multiplexing mode number.
 - a. Mode 0 is the primary mode; this means that when mode 0 is set, the function mapped on the terminal corresponds to the name of the terminal. There is always a function mapped on the primary mode. Notice that primary mode is not necessarily the default mode.

Note: The default mode is the mode at the release of the reset; also see the RESET REL. MODE column.

- b. Modes 1 to 7 are possible modes for alternate functions. On each terminal, some modes are effectively used for alternate functions, while some modes are not used and do not correspond to a functional configuration.
- 5. TYPE: Signal direction
 - I = Input
 - O = Output
 - I/O = Input and Output
 - D = Open drain
 - DS = Differential
 - A = Analog
 - PWR = Power
 - GND = Ground

Note: In the safe_mode, the buffer is configured in high-impedance.

- 6. BALL RESET STATE: State of the terminal while the active low PWRONRSTn terminal is low.
 - 0: The buffer drives V_{OL} (pulldown or pullup resistor not activated)
 - 0(PD): The buffer drives V_{OL} with an active pulldown resistor
 - 1: The buffer drives V_{OH} (pulldown or pullup resistor not activated)
 - 1(PU): The buffer drives V_{OH} with an active pullup resistor
 - Z: High-impedance
 - L: High-impedance with an active pulldown resistor
 - H: High-impedance with an active pullup resistor
- 7. **BALL RESET REL. STATE:** State of the terminal after the active low PWRONRSTn terminal transitions from low to high.
 - 0: The buffer drives V_{OL} (pulldown or pullup resistor not activated)
 - 0(PD): The buffer drives V_{OL} with an active pulldown resistor
 - 1: The buffer drives V_{OH} (pulldown or pullup resistor not activated)
 - 1(PU): The buffer drives V_{OH} with an active pullup resistor
 - Z: High-impedance.
 - L: High-impedance with an active pulldown resistor
 - H : High-impedance with an active pullup resistor
- 8. **RESET REL. MODE:** The mode is automatically configured after the active low PWRONRSTn terminal transitions from low to high.
- 9. **POWER:** The voltage supply that powers the I/O buffers of the terminal.

- 10. **HYS:** Indicates if the input buffer is with hysteresis.
- 11. **BUFFER STRENGTH:** Drive strength of the associated output buffer.
- 12. **PULLUP OR PULLDOWN TYPE:** Denotes the presence of an internal pullup or pulldown resistor. Pullup and pulldown resistors can be enabled or disabled via software.
- 13. I/O CELL: I/O cell information.

Note: Configuring two terminals to the same input signal is not supported as it can yield unexpected results. This can be easily prevented with the proper software configuration.

Table 4-2. Pin Attributes (ZCE and ZCZ Packages)

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]		ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
B8	B6	AIN0	AINO	0	A (22)	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	25	NA	Analog
A11	C7	AIN1	AIN1	0	A ⁽²¹⁾	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	25	NA	Analog
A8	B7	AIN2	AIN2	0	A ⁽²¹⁾	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	25	NA	Analog
B11	A7	AIN3	AIN3	0	A ⁽²⁰⁾	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	25	NA	Analog
C8	C8	AIN4	AIN4	0	A ⁽²⁰⁾	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	25	NA	Analog
B12	B8	AIN5	AIN5	0	A	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	NA	NA	Analog
A10	A8	AIN6	AIN6	0	Α	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	NA	NA	Analog
A12	C9	AIN7	AIN7	0	A	Z	Z	0	VDDA_ADC / VDDA_ADC	NA	NA	NA	Analog
C13	C10	CAP_VBB_MPU	CAP_VBB_MPU	NA	Α								
D6	D6	CAP_VDD_RTC	CAP_VDD_RTC	NA	Α								
B10	D9	CAP_VDD_SRAM_CORE	CAP_VDD_SRAM_CORE	NA	А								
D13	D11	CAP_VDD_SRAM_MPU	CAP_VDD_SRAM_MPU	NA	Α								
F3	F3	DDR_A0	ddr_a0	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
J2	H1	DDR_A1	ddr_a1	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
D1	E4	DDR_A2	ddr_a2	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
B3	C3	DDR_A3	ddr_a3	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
E5	C2	DDR_A4	ddr_a4	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
A2	B1	DDR_A5	ddr_a5	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
B1	D5	DDR_A6	ddr_a6	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
D2	E2	DDR_A7	ddr_a7	0	0	н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
C3	D4	DDR_A8	ddr_a8	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
B2	C1	DDR_A9	ddr_a9	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
E2	F4	DDR_A10	ddr_a10	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
G4	F2	DDR_A11	ddr_a11	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
F4	E3	DDR_A12	ddr_a12	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
H1	H3	DDR_A13	ddr_a13	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
НЗ	H4	DDR_A14	ddr_a14	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
E3	D3	DDR_A15	ddr_a15	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
A3	C4	DDR_BA0	ddr_ba0	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
E1	E1	DDR_BA1	ddr_ba1	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
B4	В3	DDR_BA2	ddr_ba2	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
F1	F1	DDR_CASn	ddr_casn	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
C2	D2	DDR_CK	ddr_ck	0	0	L	0	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
G3	G3	DDR_CKE	ddr_cke	0	0	L	0	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
C1	D1	DDR_CKn	ddr_nck	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
H2	H2	DDR_CSn0	ddr_csn0	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
N4	МЗ	DDR_D0	ddr_d0	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
P4	M4	DDR_D1	ddr_d1	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
P2	N1	DDR_D2	ddr_d2	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
P1	N2	DDR_D3	ddr_d3	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
P3	N3	DDR_D4	ddr_d4	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
T1	N4	DDR_D5	ddr_d5	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
T2	P3	DDR_D6	ddr_d6	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
R3	P4	DDR_D7	ddr_d7	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
K2	J1	DDR_D8	ddr_d8	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
K1	K1	DDR_D9	ddr_d9	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
M3	K2	DDR_D10	ddr_d10	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL



ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
M4	K3	DDR_D11	ddr_d11	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
M2	K4	DDR_D12	ddr_d12	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
M1	L3	DDR_D13	ddr_d13	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
N2	L4	DDR_D14	ddr_d14	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
N1	M1	DDR_D15	ddr_d15	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
N3	M2	DDR_DQM0	ddr_dqm0	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
K3	J2	DDR_DQM1	ddr_dqm1	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
R1	P1	DDR_DQS0	ddr_dqs0	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
L1	L1	DDR_DQS1	ddr_dqs1	0	I/O	L	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
R2	P2	DDR_DQSn0	ddr_dqsn0	0	I/O	Н	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
L2	L2	DDR_DQSn1	ddr_dqsn1	0	I/O	Н	Z	0	VDDS_DDR / VDDS_DDR	Yes	8	PU/PD	LVCMOS/SSTL/ HSTL
G1	G1	DDR_ODT	ddr_odt	0	0	L	0	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
F2	G4	DDR_RASn	ddr_rasn	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
G2	G2	DDR_RESETn	ddr_resetn	0	0	L	0	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
H4	J4	DDR_VREF	ddr_vref	0	A ⁽¹⁸⁾	NA	NA	NA	VDDS_DDR / VDDS_DDR	NA	NA	NA	Analog
J1	J3	DDR_VTP	ddr_vtp	0	(19)	NA	NA	NA	VDDS_DDR / VDDS_DDR	NA	NA	NA	Analog
A4	B2	DDR_WEn	ddr_wen	0	0	Н	1	0	VDDS_DDR / VDDS_DDR	NA	8	PU/PD	LVCMOS/SSTL/ HSTL
E18	C18	ECAP0_IN_PWM0_OUT	eCAP0_in_PWM0_out	0	I/O	Z	L	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			uart3_txd	1	0				VDDSHV6				
			spi1_cs1	2	I/O								
			pr1_ecap0_ecap_capin_apwm_o	3	I/O								
			spi1_sclk	4	I/O								
			mmc0_sdwp	5	ı								
			xdma_event_intr2	6	1								
			gpio0_7	7	I/O								
A15	C14	EMU0	EMU0	0	I/O	Н	Н	0	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpio3_7	7	I/O				VDDSHV6				

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
D14	B14	EMU1	EMU1	0	I/O	Н	Н	0	VDDSHV6 / VDDSHV6	Yes	6	PU/PD	LVCMOS
			gpio3_8	7	I/O				VDDSHV6				
C17	B18	EXTINTn	nNMI	0	I	Z	Н	0	VDDSHV6 / VDDSHV6	Yes	NA	PU/PD	LVCMOS
B5	C5	EXT_WAKEUP	EXT_WAKEUP	0	I	L	Z	0	VDDS_RTC / VDDS_RTC	Yes	NA	NA	LVCMOS
NA	R13	GPMC_A0	gpmc_a0	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txen	1	0								
			rgmii2_tctl	2	0								
			rmii2_txen	3	0								
			gpmc_a16	4	0								
			pr1_mii_mt1_clk	5	I								
			ehrpwm1_tripzone_input	6	I								
			gpio1_16	7	I/O								
NA	V14	GPMC_A1	gpmc_a1	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxdv	1	I								
			rgmii2_rctl	2	I								
			mmc2_dat0	3	I/O								
			gpmc_a17	4	0								
			pr1_mii1_txd3	5	0								
			ehrpwm0_synco	6	0								
			gpio1_17	7	I/O								
NA	U14	GPMC_A2	gpmc_a2	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txd3	1	0								
			rgmii2_td3	2	0								
			mmc2_dat1	3	I/O								
			gpmc_a18	4	0								
			pr1_mii1_txd2	5	0								
			ehrpwm1A	6	0								
			gpio1_18	7	I/O								
NA	T14	GPMC_A3	gpmc_a3	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txd2	1	0								
			rgmii2_td2	2	0								
			mmc2_dat2	3	I/O								
			gpmc_a19	4	0								
			pr1_mii1_txd1	5	0								
			ehrpwm1B	6	0								
			gpio1_19	7	I/O]							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
NA	R14	GPMC_A4	gpmc_a4	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txd1	1	0								
			rgmii2_td1	2	0								
			rmii2_txd1	3	0								
			gpmc_a20	4	0								
			pr1_mii1_txd0	5	0								
			eQEP1A_in	6	I								
			gpio1_20	7	I/O								
NA	V15	GPMC_A5	gpmc_a5	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txd0	1	0								
			rgmii2_td0	2	0								
			rmii2_txd0	3	0								
			gpmc_a21	4	0								
			pr1_mii1_rxd3	5	I								
			eQEP1B_in	6	I								
			gpio1_21	7	I/O								
NA	U15	GPMC_A6	gpmc_a6	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_txclk	1	I								
			rgmii2_tclk	2	0								
			mmc2_dat4	3	I/O								
			gpmc_a22	4	0								
			pr1_mii1_rxd2	5	I								
			eQEP1_index	6	I/O								
			gpio1_22	7	I/O								
NA	T15	GPMC_A7	gpmc_a7	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxclk	1	I								
			rgmii2_rclk	2	I								
			mmc2_dat5	3	I/O								
			gpmc_a23	4	0								
			pr1_mii1_rxd1	5	ı								
			eQEP1_strobe	6	I/O								
			gpio1_23	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
NA	V16	GPMC_A8	gpmc_a8	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxd3	1	I								
			rgmii2_rd3	2	I								
			mmc2_dat6	3	I/O								
			gpmc_a24	4	0								
			pr1_mii1_rxd0	5	I								
			mcasp0_aclkx	6	I/O								
			gpio1_24	7	I/O								
NA	U16	GPMC_A9 (10)	gpmc_a9	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxd2	1	ı								
			rgmii2_rd2	2	ı								
			mmc2_dat7 / rmii2_crs_dv	3	I/O								
			gpmc_a25	4	0								
			pr1_mii_mr1_clk	5	ı								
			mcasp0_fsx	6	I/O								
			gpio1_25	7	I/O								
NA	T16	GPMC_A10	gpmc_a10	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxd1	1	I								
			rgmii2_rd1	2	I								
			rmii2_rxd1	3	I								
			gpmc_a26	4	0								
			pr1_mii1_rxdv	5	I								
			mcasp0_axr0	6	I/O								
			gpio1_26	7	I/O								
NA	V17	GPMC_A11	gpmc_a11	0	0	L	L	7	NA / VDDSHV3	Yes	6	PU/PD	LVCMOS
			gmii2_rxd0	1	I								
			rgmii2_rd0	2	I								
			rmii2_rxd0	3	I								
			gpmc_a27	4	0								
			pr1_mii1_rxer	5	I								
			mcasp0_axr1	6	I/O								
			gpio1_27	7	I/O								
W10	U7	GPMC_AD0	gpmc_ad0	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat0	1	I/O				VDDSHV1				
			gpio1_0	7	I/O								
V9	V7	GPMC_AD1	gpmc_ad1	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat1	1	I/O				VDDSHV1				
			gpio1_1	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]		BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
V12	R8	GPMC_AD2	gpmc_ad2	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat2	1	I/O				VDDSHV1				
			gpio1_2	7	I/O								
W13	T8	GPMC_AD3	gpmc_ad3	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat3	1	I/O				VDDSHV1				
			gpio1_3	7	I/O								
V13	U8	GPMC_AD4	gpmc_ad4	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat4	1	I/O				VDDSHV1				
			gpio1_4	7	I/O								
W14	V8	GPMC_AD5	gpmc_ad5	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat5	1	I/O				VDDSHV1				
			gpio1_5	7	I/O								
U14	R9	GPMC_AD6	gpmc_ad6	0	I/O	L	L 7 VDDSHV1 / Yes 6 PU/PD VDDSHV1	PU/PD	LVCMOS				
			mmc1_dat6	1	I/O				VDDSHV1				
			gpio1_6	7	I/O								
W15	T9	GPMC_AD7	gpmc_ad7	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			mmc1_dat7	1	I/O				VDDSHV1				
			gpio1_7	7	I/O								
V15	U10	GPMC_AD8	gpmc_ad8	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data23	1	0				VDDSHV2				
			mmc1_dat0	2	I/O								
			mmc2_dat4	3	I/O								
			ehrpwm2A	4	0								
			pr1_mii_mt0_clk	5	I								
			gpio0_22	7	I/O								
W16	T10	GPMC_AD9	gpmc_ad9	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data22	1	0				VDDSHV2				
			mmc1_dat1	2	I/O								
			mmc2_dat5	3	I/O								
			ehrpwm2B	4	0								
			pr1_mii0_col	5	I								
			gpio0_23	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
T12	T11	GPMC_AD10	gpmc_ad10	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data21	1	0				VDDSHV2				
			mmc1_dat2	2	I/O								
			mmc2_dat6	3	I/O								
			ehrpwm2_tripzone_input	4	I								
			pr1_mii0_txen	5	0								
			gpio0_26	7	I/O								
U12	U12	GPMC_AD11	gpmc_ad11	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data20	1	0				VDDSHV2				
			mmc1_dat3	2	I/O								
			mmc2_dat7	3	I/O								
			ehrpwm0_synco	4	0								
			pr1_mii0_txd3	5	0								
			gpio0_27	7	I/O							PU/PD	
U13	T12	GPMC_AD12	gpmc_ad12	0	I/O	L	L	7	VDDSHV1 /	Yes	6		LVCMOS
			lcd_data19	1	0				VDDSHV2				
			mmc1_dat4	2	I/O								
			mmc2_dat0	3	I/O								
			eQEP2A_in	4	I								
			pr1_mii0_txd2	5	0								
			pr1_pru0_pru_r30_14	6	0								
			gpio1_12	7	I/O								
T13	R12	GPMC_AD13	gpmc_ad13	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data18	1	0				VDDSHV2				
			mmc1_dat5	2	I/O								
			mmc2_dat1	3	I/O								
			eQEP2B_in	4	I								
			pr1_mii0_txd1	5	0								
			pr1_pru0_pru_r30_15	6	0								
			gpio1_13	7	I/O								
W17	V13	GPMC_AD14	gpmc_ad14	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data17	1	0				VDDSHV2				
			mmc1_dat6	2	I/O								
			mmc2_dat2	3	I/O								
			eQEP2_index	4	I/O								
			pr1_mii0_txd0	5	0								
			pr1_pru0_pru_r31_14	6	I								
			gpio1_14	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
V17	U13	GPMC_AD15	gpmc_ad15	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_data16	1	0				VDDSHV2				
			mmc1_dat7	2	I/O								
			mmc2_dat3	3	I/O								
			eQEP2_strobe	4	I/O								
			pr1_ecap0_ecap_capin_apwm_o	5	I/O								
			pr1_pru0_pru_r31_15	6	I								
			gpio1_15	7	I/O								
V10	R7	GPMC_ADVn_ALE	gpmc_advn_ale	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			timer4	2	I/O				VDDSHV1				
			gpio2_2	7	I/O								
V8	T6	GPMC_BEn0_CLE	gpmc_be0n_cle	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			timer5	2	I/O				VDDSHV1				
			gpio2_5	7	I/O								
V18	U18	GPMC_BEn1	gpmc_be1n	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gmii2_col	1	I				VDDSHV3				
			gpmc_csn6	2	0								
			mmc2_dat3	3	I/O								
			gpmc_dir	4	0								
			pr1_mii1_rxlink	5	I								
			mcasp0_aclkr	6	I/O								
			gpio1_28	7	I/O								
V16	V12	GPMC_CLK	gpmc_clk	0	I/O	L	L	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			lcd_memory_clk	1	0				VDDSHV2				
			gpmc_wait1	2	I								
			mmc2_clk	3	I/O								
			pr1_mii1_crs	4	I								
			pr1_mdio_mdclk	5	0								
			mcasp0_fsr	6	I/O								
			gpio2_1	7	I/O								
W8	V6	GPMC_CSn0	gpmc_csn0	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gpio1_29	7	I/O				VDDSHV1				

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
V14	U9	GPMC_CSn1	gpmc_csn1	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gpmc_clk	1	I/O				VDDSHV1				
			mmc1_clk	2	I/O								
			pr1_edio_data_in6	3	1								
			pr1_edio_data_out6	4	0								
			pr1_pru1_pru_r30_12	5	0								
			pr1_pru1_pru_r31_12	6	I								
			gpio1_30	7	I/O								
U15	V9	GPMC_CSn2	gpmc_csn2	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gpmc_be1n	1	0				VDDSHV1				
			mmc1_cmd	2	I/O								
			pr1_edio_data_in7	3	I								
			pr1_edio_data_out7	4	0								
			pr1_pru1_pru_r30_13	5	0								
			pr1_pru1_pru_r31_13	6	I								
			gpio1_31	7	I/O								
U17	T13	GPMC_CSn3 (6)	gpmc_csn3	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gpmc_a3	1	0				VDDSHV2				
			rmii2_crs_dv	2	I								
			mmc2_cmd	3	I/O								
			pr1_mii0_crs	4	I								
			pr1_mdio_data	5	I/O								
			EMU4	6	I/O								
			gpio2_0	7	I/O								
W9	T7	GPMC_OEn_REn	gpmc_oen_ren	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			timer7	2	I/O				VDDSHV1				
			gpio2_3	7	I/O								
R15	T17	GPMC_WAIT0	gpmc_wait0	0	I	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gmii2_crs	1	I				VDDSHV3				
			gpmc_csn4	2	0								
			rmii2_crs_dv	3	ı								
			mmc1_sdcd	4	ı								
			pr1_mii1_col	5	ı								
			uart4_rxd	6	ı								
			gpio0_30	7	I/O								
U8	U6	GPMC_WEn	gpmc_wen	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			timer6	2	I/O				VDDSHV1				
			gpio2_4	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
W18	U17	GPMC_WPn	gpmc_wpn	0	0	Н	Н	7	VDDSHV1 /	Yes	6	PU/PD	LVCMOS
			gmii2_rxerr	1	I				VDDSHV3				
			gpmc_csn5	2	0								
			rmii2_rxerr	3	I								
			mmc2_sdcd	4	I								
			pr1_mii1_txen	5	0								
			uart4_txd	6	0								
			gpio0_31	7	I/O								
C18	C17	I2C0_SDA	I2C0_SDA	0	I/OD	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			timer4	1	I/O				VDDSHV6				
			uart2_ctsn	2	I								
			eCAP2_in_PWM2_out	3	I/O								
			gpio3_5	7	I/O				VDDSHV6 / Ves / DII/DD				
B19	C16	I2C0_SCL	I2C0_SCL	0	I/OD	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			timer7	1	I/O	VDDCIIVC							
			uart2_rtsn	2	0								
			eCAP1_in_PWM1_out	3	I/O								
			gpio3_6	7	I/O								
W7	R6	LCD_AC_BIAS_EN	lcd_ac_bias_en	0	0	Z	L	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a11	1	0				VDDSHV6				
			pr1_mii1_crs	2	I								
			pr1_edio_data_in5	3	I								
			pr1_edio_data_out5	4	0								
			pr1_pru1_pru_r30_11	5	0								
			pr1_pru1_pru_r31_11	6	I								
			gpio2_25	7	I/O								
U1	R1	LCD_DATA0 (5)	lcd_data0	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a0	1	0				VDDSHV6				
			pr1_mii_mt0_clk	2	I								
			ehrpwm2A	3	0								
			pr1_pru1_pru_r30_0	5	0								
			pr1_pru1_pru_r31_0	6	I								
			gpio2_6	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
U2	R2	LCD_DATA1 (5)	lcd_data1	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a1	1	0				VDDSHV6				
			pr1_mii0_txen	2	0								
			ehrpwm2B	3	0								
			pr1_pru1_pru_r30_1	5	0								
			pr1_pru1_pru_r31_1	6	I								
			gpio2_7	7	I/O								
V1	R3	LCD_DATA2 (5)	lcd_data2	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a2	1	0				VDDSHV6				
			pr1_mii0_txd3	2	0								
			ehrpwm2_tripzone_input	3	ı								
			pr1_pru1_pru_r30_2	5	0								
			pr1_pru1_pru_r31_2	6	I								
			gpio2_8	7	I/O								
V2	R4	LCD_DATA3 (5)	lcd_data3	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a3	1	0				VDDSHV6				
			pr1_mii0_txd2	2	0								
			ehrpwm0_synco	3	0								
			pr1_pru1_pru_r30_3	5	0								
			pr1_pru1_pru_r31_3	6	I								
			gpio2_9	7	I/O								
W2	T1	LCD_DATA4 (5)	lcd_data4	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a4	1	0				VDDSHV6				
			pr1_mii0_txd1	2	0								
			eQEP2A_in	3	I								
			pr1_pru1_pru_r30_4	5	0								
			pr1_pru1_pru_r31_4	6	ı								
			gpio2_10	7	I/O								
W3	T2	LCD_DATA5 (5)	lcd_data5	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a5	1	0				VDDSHV6				
			pr1_mii0_txd0	2	0								
			eQEP2B_in	3	I	1							
			pr1_pru1_pru_r30_5	5	0	1							
			pr1_pru1_pru_r31_5	6	I	1							
			gpio2_11	7	I/O	1							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
V3	T3	LCD_DATA6 (5)	lcd_data6	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a6	1	0				VDDSHV6				
			pr1_edio_data_in6	2	I								
			eQEP2_index	3	I/O								
			pr1_edio_data_out6	4	0								
			pr1_pru1_pru_r30_6	5	0								
			pr1_pru1_pru_r31_6	6	I								
			gpio2_12	7	I/O								
U3	T4	LCD_DATA7 (5)	lcd_data7	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a7	1	0				VDDSHV6				
			pr1_edio_data_in7	2	I								
			eQEP2_strobe	3	I/O								
			pr1_edio_data_out7	4	0								
			pr1_pru1_pru_r30_7	5	0								
			pr1_pru1_pru_r31_7	6	I								
			gpio2_13	7	I/O								
V4	U1	LCD_DATA8 (5)	lcd_data8	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a12	1	0				VDDSHV6				
			ehrpwm1_tripzone_input	2	I								
			mcasp0_aclkx	3	I/O								
			uart5_txd	4	0								
			pr1_mii0_rxd3	5	I								
			uart2_ctsn	6	I								
			gpio2_14	7	I/O								
W4	U2	LCD_DATA9 (5)	lcd_data9	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a13	1	0				VDDSHV6				
			ehrpwm0_synco	2	0								
			mcasp0_fsx	3	I/O								
			uart5_rxd	4	I								
			pr1_mii0_rxd2	5	Ī								
			uart2_rtsn	6	0								
			gpio2_15	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
J5 U3	U3	LCD_DATA10 (5)	lcd_data10	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a14	1	0				VDDSHV6				
			ehrpwm1A	2	0								
			mcasp0_axr0	3	I/O								
			pr1_mii0_rxd1	5	I								
			uart3_ctsn	6	ı								
			gpio2_16	7	I/O								
V5	U4	LCD_DATA11 (5)	lcd_data11	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a15	1	0				VDDSHV6				
			ehrpwm1B	2	0								
			mcasp0_ahclkr	3	I/O								
			mcasp0_axr2	4	I/O								
			pr1_mii0_rxd0	5	I								
			uart3_rtsn	6	0					l			
			gpio2_17	7	I/O								
V6	V2	LCD_DATA12 (5)	lcd_data12	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a16	1	0				VDDSHV6				
			eQEP1A_in	2	I								
			mcasp0_aclkr	3	I/O								
			mcasp0_axr2	4	I/O								
			pr1_mii0_rxlink	5	I								
			uart4_ctsn	6	I								
			gpio0_8	7	I/O								
U6	V3	LCD_DATA13 (5)	lcd_data13	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a17	1	0				VDDSHV6				
			eQEP1B_in	2	I								
			mcasp0_fsr	3	I/O								
			mcasp0_axr3	4	I/O								
			pr1_mii0_rxer	5	I								
			uart4_rtsn	6	0								
			gpio0_9	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
W6	V4	LCD_DATA14 (5)	lcd_data14	0	I/O	Z	Z	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a18	1	0				VDDSHV6				
			eQEP1_index	2	I/O								
			mcasp0_axr1	3	I/O								
			uart5_rxd	4	I								
			pr1_mii_mr0_clk	5	I								
			uart5_ctsn	6	ı								
			gpio0_10	7	I/O								
V7	T5	LCD_DATA15 (5)	lcd_data15	0	I/O	Z	Z	7	VDDSHV6 / VDDSHV6	Yes	6	PU/PD	LVCMOS
			gpmc_a19	1	0								
			eQEP1_strobe	2	I/O								
			mcasp0_ahclkx	3	I/O								
			mcasp0_axr3	4	I/O								
			pr1_mii0_rxdv	5	I								
			uart5_rtsn	6	0								
			gpio0_11	7	I/O								
T7	R5	LCD_HSYNC (7)	lcd_hsync	0	0	Z	L	7	VDDSHV6 / VDDSHV6	Yes	6	PU/PD	LVCMOS
			gpmc_a9	1	0								
			gpmc_a2	2	0								
			pr1_edio_data_in3	3	I								
			pr1_edio_data_out3	4	0								
			pr1_pru1_pru_r30_9	5	0								
			pr1_pru1_pru_r31_9	6	ı								
			gpio2_23	7	I/O								
W5	V5	LCD_PCLK	lcd_pclk	0	0	Z	L	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a10	1	0				VDDSHV6				
			pr1_mii0_crs	2	I								
			pr1_edio_data_in4	3	I								
			pr1_edio_data_out4	4	0								
			pr1_pru1_pru_r30_10	5	0								
			pr1_pru1_pru_r31_10	6	I								
			gpio2_24	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
J7 U5	U5	LCD_VSYNC (7)	lcd_vsync	0	0	Z	L	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			gpmc_a8	1	0				VDDSHV6				
			gpmc_a1	2	0								
			pr1_edio_data_in2	3	I								
			pr1_edio_data_out2	4	0								
			pr1_pru1_pru_r30_8	5	0								
			pr1_pru1_pru_r31_8	6	I								
			gpio2_22	7	I/O								
NA	B13	MCASP0_FSX	mcasp0_fsx	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			ehrpwm0B	1	0								
			spi1_d0	3	I/O								
			mmc1_sdcd	4	I								
			pr1_pru0_pru_r30_1	5	0								
			pr1_pru0_pru_r31_1	6	I								
			gpio3_15	7	I/O								
NA	B12	MCASP0_ACLKR	mcasp0_aclkr	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			eQEP0A_in	1	I								
			mcasp0_axr2	2	I/O								
			mcasp1_aclkx	3	I/O								
			mmc0_sdwp	4	I								
			pr1_pru0_pru_r30_4	5	0								
			pr1_pru0_pru_r31_4	6	I								
			gpio3_18	7	I/O								
NA	C12	MCASP0_AHCLKR	mcasp0_ahclkr	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			ehrpwm0_synci	1	ı								
			mcasp0_axr2	2	I/O								
			spi1_cs0	3	I/O								
			eCAP2_in_PWM2_out	4	I/O								
			pr1_pru0_pru_r30_3	5	0								
			pr1_pru0_pru_r31_3	6	ı								
			gpio3_17	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
NA	A14	MCASP0_AHCLKX	mcasp0_ahclkx	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			eQEP0_strobe	1	I/O								
			mcasp0_axr3	2	I/O								
			mcasp1_axr1	3	I/O								
			EMU4	4	I/O								
			pr1_pru0_pru_r30_7	5	0								
			pr1_pru0_pru_r31_7	6	I								
			gpio3_21	7	I/O								
NA	A13	MCASP0_ACLKX	mcasp0_aclkx	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			ehrpwm0A	1	0								
			spi1_sclk	3	I/O								
			mmc0_sdcd	4	I								
			pr1_pru0_pru_r30_0	5	0								
			pr1_pru0_pru_r31_0	6	ı								
			gpio3_14	7	I/O								
NA	C13	MCASP0_FSR	mcasp0_fsr	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			eQEP0B_in	1	I								
			mcasp0_axr3	2	I/O								
			mcasp1_fsx	3	I/O								
			EMU2	4	I/O								
			pr1_pru0_pru_r30_5	5	0								
			pr1_pru0_pru_r31_5	6	I								
			gpio3_19	7	I/O								
NA	D12	MCASP0_AXR0	mcasp0_axr0	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			ehrpwm0_tripzone_input	1	I								
			spi1_d1	3	I/O								
			mmc2_sdcd	4	ı								
			pr1_pru0_pru_r30_2	5	0								
			pr1_pru0_pru_r31_2	6	I								
			gpio3_16	7	I/O								
NA	D13	MCASP0_AXR1	mcasp0_axr1	0	I/O	L	L	7	NA / VDDSHV6	Yes	6	PU/PD	LVCMOS
			eQEP0_index	1	I/O								
			mcasp1_axr0	3	I/O								
			EMU3	4	I/O								
			pr1_pru0_pru_r30_6	5	0								
			pr1_pru0_pru_r31_6	6	ı								
			gpio3_20	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
R19	M18	MDC	mdio_clk	0	0	Н	Н	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			timer5	1	I/O				VDDSHV5				
			uart5_txd	2	0								
			uart3_rtsn	3	0								
			mmc0_sdwp	4	I								
			mmc1_clk	5	I/O								
			mmc2_clk	6	I/O								
			gpio0_1	7	I/O								
P17	M17	MDIO	mdio_data	0	I/O	Н	Н	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			timer6	1	I/O				VDDSHV5				
			uart5_rxd	2	ı								
			uart3_ctsn	3	ı								
			mmc0_sdcd	4	I								
			mmc1_cmd	5	I/O								
			mmc2_cmd	6	I/O								
			gpio0_0	7	I/O								
L19	J17	MII1_RX_DV	gmii1_rxdv	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			lcd_memory_clk	1	0				VDDSHV5				
			rgmii1_rctl	2	ı								
			uart5_txd	3	0								
			mcasp1_aclkx	4	I/O								
			mmc2_dat0	5	I/O								
			mcasp0_aclkr	6	I/O								
			gpio3_4	7	I/O								
K17	J16	MII1_TX_EN	gmii1_txen	0	0	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_txen	1	0				VDDSHV5				
			rgmii1_tctl	2	0								
			timer4	3	I/O								
			mcasp1_axr0	4	I/O								
			eQEP0_index	5	I/O	1							
			mmc2_cmd	6	I/O	1							
			gpio3_3	7	I/O	1							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
K19	J15	MII1_RX_ER	gmii1_rxerr	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_rxerr	1	I				VDDSHV5				
			spi1_d1	2	I/O								
			I2C1_SCL	3	I/OD								
			mcasp1_fsx	4	I/O								
			uart5_rtsn	5	0								
			uart2_txd	6	0								
			gpio3_2	7	I/O								
M19	L18	MII1_RX_CLK	gmii1_rxclk	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			uart2_txd	1	0				VDDSHV5				
			rgmii1_rclk	2	I								
			mmc0_dat6	3	I/O								
			mmc1_dat1	4	I/O								
			uart1_dsrn	5	I								
			mcasp0_fsx	6	I/O								
			gpio3_10	7	I/O								
N19	K18	MII1_TX_CLK	gmii1_txclk	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			uart2_rxd	1	I				VDDSHV5				
			rgmii1_tclk	2	0								
			mmc0_dat7	3	I/O								
			mmc1_dat0	4	I/O								
			uart1_dcdn	5	I								
			mcasp0_aclkx	6	I/O								
			gpio3_9	7	I/O								
J19	H16	MII1_COL	gmii1_col	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii2_refclk	1	I/O				VDDSHV5				
			spi1_sclk	2	I/O								
			uart5_rxd	3	I								
			mcasp1_axr2	4	I/O								
			mmc2_dat3	5	I/O								
			mcasp0_axr2	6	I/O								
			gpio3_0	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
J18	H17	MII1_CRS	gmii1_crs	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_crs_dv	1	I				VDDSHV5				
			spi1_d0	2	I/O								
			I2C1_SDA	3	I/OD								
			mcasp1_aclkx	4	I/O								
			uart5_ctsn	5	l								
			uart2_rxd	6	l								
			gpio3_1	7	I/O								
P18	M16	MII1_RXD0	gmii1_rxd0	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_rxd0	1	I				VDDSHV5				
			rgmii1_rd0	2	I								
			mcasp1_ahclkx	3	I/O								
			mcasp1_ahclkr	4	I/O								
			mcasp1_aclkr	5	I/O								
			mcasp0_axr3	6	I/O								
			gpio2_21	7	I/O								
P19	L15	MII1_RXD1	gmii1_rxd1	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_rxd1	1	I				VDDSHV5				
			rgmii1_rd1	2	l								
			mcasp1_axr3	3	I/O								
			mcasp1_fsr	4	I/O								
			eQEP0_strobe	5	I/O								
			mmc2_clk	6	I/O								
			gpio2_20	7	I/O								
N16	L16	MII1_RXD2	gmii1_rxd2	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			uart3_txd	1	0				VDDSHV5				
			rgmii1_rd2	2	l								
			mmc0_dat4	3	I/O								
			mmc1_dat3	4	I/O								
			uart1_rin	5	I								
			mcasp0_axr1	6	I/O								
			gpio2_19	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]		BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
N17	L17	MII1_RXD3	gmii1_rxd3	0	I	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			uart3_rxd	1	ı				VDDSHV5				
			rgmii1_rd3	2	ı								
			mmc0_dat5	3	I/O								
			mmc1_dat2	4	I/O								
			uart1_dtrn	5	0								
			mcasp0_axr0	6	I/O								
			gpio2_18	7	I/O								
L18	K17	MII1_TXD0	gmii1_txd0	0	0	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_txd0	1	0				VDDSHV5				
			rgmii1_td0	2	0								
			mcasp1_axr2	3	I/O								
			mcasp1_aclkr	4	I/O								
			eQEP0B_in	5	I								
			mmc1_clk	6	I/O								
			gpio0_28	7	I/O								
M18	K16	MII1_TXD1	gmii1_txd1	0	0	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			rmii1_txd1	1	0				VDDSHV5				
			rgmii1_td1	2	0								
			mcasp1_fsr	3	I/O								
			mcasp1_axr1	4	I/O								
			eQEP0A_in	5	I								
			mmc1_cmd	6	I/O								
			gpio0_21	7	I/O								
N18	K15	MII1_TXD2	gmii1_txd2	0	0	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			dcan0_rx	1	I				VDDSHV5				
			rgmii1_td2	2	0								
			uart4_txd	3	0	1							
			mcasp1_axr0	4	I/O								
			mmc2_dat2	5	I/O	1							
			mcasp0_ahclkx	6	I/O	1							
			gpio0_17	7	I/O	1							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
M17	J18	MII1_TXD3	gmii1_txd3	0	0	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			dcan0_tx	1	0				VDDSHV5				
			rgmii1_td3	2	0								
			uart4_rxd	3	I								
			mcasp1_fsx	4	I/O								
			mmc2_dat1	5	I/O								
			mcasp0_fsr	6	I/O								
			gpio0_16	7	I/O								
G17	G18	MMC0_CMD	mmc0_cmd	0	I/O	Н	Н	7	VDDSHV4 /	Yes	6	PU/PD	LVCMOS
			gpmc_a25	1	0				VDDSHV4				
			uart3_rtsn	2	0								
			uart2_txd	3	0								
			dcan1_rx	4	I								
			pr1_pru0_pru_r30_13	5	0								
			pr1_pru0_pru_r31_13	6	I								
			gpio2_31	7	I/O								
G19	G17	MMC0_CLK	mmc0_clk	0	I/O	Н	Н	7	VDDSHV4 /	Yes	6	PU/PD	LVCMOS
			gpmc_a24	1	0				VDDSHV4				
			uart3_ctsn	2	I								
			uart2_rxd	3	I								
			dcan1_tx	4	0								
			pr1_pru0_pru_r30_12	5	0								
			pr1_pru0_pru_r31_12	6	I								
			gpio2_30	7	I/O								
G18	G16	MMC0_DAT0	mmc0_dat0	0	I/O	Н	Н	7	VDDSHV4 /	Yes	6	PU/PD	LVCMOS
			gpmc_a23	1	0				VDDSHV4				
			uart5_rtsn	2	0								
			uart3_txd	3	0								
			uart1_rin	4	I								
			pr1_pru0_pru_r30_11	5	0	1							
			pr1_pru0_pru_r31_11	6	I	1							
			gpio2_29	7	I/O	1							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]		BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
H17	G15	MMC0_DAT1	mmc0_dat1	0	I/O	Н	Н	7	VDDSHV4 /	Yes	6	PU/PD	LVCMOS
			gpmc_a22	1	0				VDDSHV4				
			uart5_ctsn	2	I								
			uart3_rxd	3	I								
			uart1_dtrn	4	0								
			pr1_pru0_pru_r30_10	5	0								
			pr1_pru0_pru_r31_10	6	I								
			gpio2_28	7	I/O								
H18	F18	MMC0_DAT2	mmc0_dat2	0	I/O	Н	Н	7	VDDSHV4 /	Yes	6	PU/PD	LVCMOS
			gpmc_a21	1	0				VDDSHV4				
			uart4_rtsn	2	0								
			timer6	3	I/O								
			uart1_dsrn	4	I								
			pr1_pru0_pru_r30_9	5	0								
			pr1_pru0_pru_r31_9	6	I								
			gpio2_27	7	I/O								
H19	F17	MMC0_DAT3	mmc0_dat3	0	I/O	Н	Н	7	VDDSHV4 / VDDSHV4	Yes	6	PU/PD	LVCMOS
			gpmc_a20	1	0				VDDSHV4				
			uart4_ctsn	2	I								
			timer5	3	I/O								
			uart1_dcdn	4	I								
			pr1_pru0_pru_r30_8	5	0								
			pr1_pru0_pru_r31_8	6	I								
			gpio2_26	7	I/O								
C7	C6	PMIC_POWER_EN	PMIC_POWER_EN	0	0	Н	1	0	VDDS_RTC / VDDS_RTC	NA	6	NA	LVCMOS
E15	B15	PWRONRSTn	porz	0	I	Z	Z	0	VDDSHV6 / VDDSHV6 (12)	Yes	NA	NA	LVCMOS
B6	A3	RESERVED (3)	testout	0	0	NA	NA	NA	VDDSHV6 / VDDSHV6	NA	NA	NA	Analog
K18	H18	RMII1_REF_CLK	rmii1_refclk	0	I/O	L	L	7	VDDSHV5 /	Yes	6	PU/PD	LVCMOS
			xdma_event_intr2	1	I				VDDSHV5				
			spi1_cs0	2	I/O								
			uart5_txd	3	0								
			mcasp1_axr3	4	I/O								
			mmc0_pow	5	0								
			mcasp1_ahclkx	6	I/O								
			gpio0_29	7	I/O								
A7	B4	RTC_KALDO_ENn	ENZ_KALDO_1P8V	0	I	Z	Z	0	VDDS_RTC / VDDS_RTC	NA	NA	NA	Analog

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
B7	B5	RTC_PWRONRSTn	RTC_PORz	0	I	Z	Z	0	VDDS_RTC / VDDS_RTC	Yes	NA	NA	LVCMOS
A6	A6	RTC_XTALIN	OSC1_IN	0	I	Н	Н	0	VDDS_RTC / VDDS_RTC	Yes	NA	PU ⁽¹⁾	LVCMOS
A5	A4	RTC_XTALOUT	OSC1_OUT	0	0	Z ⁽²³⁾	Z ⁽²³⁾	0	VDDS_RTC / VDDS_RTC	NA	NA ⁽¹⁵⁾	NA	LVCMOS
A18	A17	SPI0_SCLK	spi0_sclk	0	I/O	Z	Н	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			uart2_rxd	1	I				VDDSHV6				
			I2C2_SDA	2	I/OD								
			ehrpwm0A	3	0								
			pr1_uart0_cts_n	4	I								
			pr1_edio_sof	5	0								
			EMU2	6	I/O								
			gpio0_2	7	I/O								
A17	A16	SPI0_CS0	spi0_cs0	0	I/O	Z	Н	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			mmc2_sdwp	1	I				VDDSHV6				
			12C1_SCL	2	I/OD								
			ehrpwm0_synci	3	I								
			pr1_uart0_txd	4	0								
			pr1_edio_data_in1	5	I								
			pr1_edio_data_out1	6	0								
			gpio0_5	7	I/O								
B16	C15	SPI0_CS1	spi0_cs1	0	I/O	Z	Н	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			uart3_rxd	1	I				VDDSHV6				
			eCAP1_in_PWM1_out	2	I/O								
			mmc0_pow	3	0								
			xdma_event_intr2	4	I								
			mmc0_sdcd	5	I								
			EMU4	6	I/O								
			gpio0_6	7	I/O								
B18	B17	SPI0_D0	spi0_d0	0	I/O	Z	Н	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			uart2_txd	1	0				VDDSHV6				
			12C2_SCL	2	I/OD								
			ehrpwm0B	3	0								
			pr1_uart0_rts_n	4	0	1							
			pr1_edio_latch_in	5	I								
			EMU3	6	I/O	1							
			gpio0_3	7	I/O	1							

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
B17	B16	SPI0_D1	spi0_d1	0	I/O	Z	Н	7	VDDSHV6 /	Yes	6	PU/PD	LVCMOS
			mmc1_sdwp	1	I				VDDSHV6				
			I2C1_SDA	2	I/OD								
			ehrpwm0_tripzone_input	3	I								
			pr1_uart0_rxd	4	I								
			pr1_edio_data_in0	5	I								
			pr1_edio_data_out0	6	0								
			gpio0_4	7	I/O								
B14	A12	тск	TCK	0	I	Н	Н	0	VDDSHV6 / VDDSHV6	Yes	NA	PU/PD	LVCMOS
B13	B11	TDI	TDI	0	I	Н	Н	0	VDDSHV6 / VDDSHV6	Yes	NA	PU/PD	LVCMOS
A14	A11	TDO	TDO	0	0	Н	Н	0	VDDSHV6 / VDDSHV6	NA	4	PU/PD	LVCMOS
C14	C11	TMS	TMS	0	1	Н	Н	0	VDDSHV6 / VDDSHV6	Yes	NA	PU/PD	LVCMOS
A13	B10	TRSTn	nTRST	0	I	L	L	0	VDDSHV6 / VDDSHV6	Yes	NA	PU/PD	LVCMOS
F17	E16	UART0_TXD	uart0_txd	0	0	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			spi1_cs1	1	I/O				VDDSHV6				
			dcan0_rx	2	I								
			I2C2_SCL	3	I/OD								
			eCAP1_in_PWM1_out	4	I/O								
			pr1_pru1_pru_r30_15	5	0								
			pr1_pru1_pru_r31_15	6	I								
			gpio1_11	7	I/O								
F19	E18	UART0_CTSn	uart0_ctsn	0	I	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			uart4_rxd	1	I				VDDSHV6				
			dcan1_tx	2	0								
			I2C1_SDA	3	I/OD								
			spi1_d0	4	I/O								
			timer7	5	I/O	1							
			pr1_edc_sync0_out	6	0								
			gpio1_8	7	I/O								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	VO CELL [13]
E19	E15	UART0_RXD	uart0_rxd	0	I	Z	Н	7	VDDSHV6 / VDDSHV6	Yes	4	PU/PD	LVCMOS
			spi1_cs0	1	I/O				VDDSHV6				
			dcan0_tx	2	0								
			I2C2_SDA	3	I/OD								
			eCAP2_in_PWM2_out	4	I/O								
			pr1_pru1_pru_r30_14	5	0								
			pr1_pru1_pru_r31_14	6	I								
			gpio1_10	7	I/O								
F18	E17	UART0_RTSn	uart0_rtsn	0	0	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			uart4_txd	1	0				VDDSHV6				
			dcan1_rx	2	I								
			I2C1_SCL	3	I/OD								
			spi1_d1	4	I/O								
			spi1_cs0	5 1/0									
			pr1_edc_sync1_out	6	0								
			gpio1_9	7	I/O								
C19	D15	UART1_TXD	uart1_txd	0	0	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			mmc2_sdwp	1	I				VDDSHV6				
			dcan1_rx	2	I								
			I2C1_SCL	3	I/OD								
			pr1_uart0_txd	5	0								
			pr1_pru0_pru_r31_16	6	I								
			gpio0_15	7	I/O								
D18	D16	UART1_RXD	uart1_rxd	0	I	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			mmc1_sdwp	1	ı				VDDSHV6				
			dcan1_tx	2	0								
			I2C1_SDA	3	I/OD								
			pr1_uart0_rxd	5	I								
			pr1_pru1_pru_r31_16	6	I								
			gpio0_14	7	I/O								
D19	D17	UART1_RTSn	uart1_rtsn	0	0	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			timer5	1	I/O				VDDSHV6				
			dcan0_rx	2	I								
			I2C2_SCL	3	I/OD								
			spi1_cs1	4	I/O								
			pr1_uart0_rts_n	5	0								
			pr1_edc_latch1_in	6	I								
			gpio0_13	7	I/O								



ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
E17	D18	UART1_CTSn	uart1_ctsn	0	I	Z	Н	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			timer6	1	I/O				VDDSHV6				
			dcan0_tx	2	0								
			I2C2_SDA	3	I/OD								
			spi1_cs0	4	I/O								
			pr1_uart0_cts_n	5	I								
			pr1_edc_latch0_in	6	I								
			gpio0_12	7	I/O								
T18	M15	USB0_CE	USB0_CE	0	A	Z	Z	0	VDDA*_USB0 / VDDA*_USB0 (26)	NA	NA	NA	Analog
T19	P15	USB0_VBUS	USB0_VBUS	0	A	Z	Z	0	VDDA*_USB0 / VDDA*_USB0 (26)	NA	NA	NA	Analog
U18	N18	USB0_DM	USB0_DM	0	A	Z	Z	0 (13)	VDDA*_USB0 / VDDA*_USB0 (26)	Yes (16)	8 (16)	NA	Analog
G16	F16	USB0_DRVVBUS	USB0_DRVVBUS	0	0	L	0(PD)	0	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			gpio0_18	7	I/O				VDDSHV6				
V19	P16	USB0_ID	USB0_ID	0	A	Z	Z	0	VDDA*_USB0 / VDDA*_USB0 (26)	NA	NA	NA	Analog
U19	N17	USB0_DP	USB0_DP	0	A	Z	Z	0 (13)	VDDA*_USB0 / VDDA*_USB0	Yes (16)	8 (16)	NA	Analog
NA	P18	USB1_CE	USB1_CE	0	A	Z	Z	0	NA / VDDA*_USB1	NA	NA	NA	Analog
NA	P17	USB1_ID	USB1_ID	0	A	Z	Z	0	NA / VDDA*_USB1	NA	NA	NA	Analog
NA	T18	USB1_VBUS	USB1_VBUS	0	A	Z	Z	0	NA / VDDA*_USB1	NA	NA	NA	Analog
NA	R17	USB1_DP	USB1_DP	0	A	Z	Z	0 (14)	NA / VDDA*_USB1	Yes (17)	8 (17)	NA	Analog
NA	F15	USB1_DRVVBUS	USB1_DRVVBUS	0	0	L	0(PD)	0	NA / VDDSHV6	Yes	4	PU/PD	LVCMOS
			gpio3_13	7	I/O								
NA	R18	USB1_DM	USB1_DM	0	A	Z	Z	0 (14)	NA / VDDA*_USB1	Yes (17)	8 (17)	NA	Analog
R17	N16	VDDA1P8V_USB0	VDDA1P8V_USB0	NA	PWR								
NA	R16	VDDA1P8V_USB1	VDDA1P8V_USB1	NA	PWR								
R18	N15	VDDA3P3V_USB0	VDDA3P3V_USB0	NA	PWR								
NA	R15	VDDA3P3V_USB1	VDDA3P3V_USB1	NA	PWR								

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET REL. STATE [6] ⁽²⁵⁾	RESET REL. MODE [8]	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
D7	D8	VDDA_ADC	VDDA_ADC	NA	PWR							
D12, F16, M16, T6, T14	E6, E14, F9, K13, N6, P9, P14	VDDS	VDDS	NA	PWR							
R8, R9, R11, R12, R13	P7, P8	VDDSHV1	VDDSHV1	NA	PWR							
NA	P10, P11	VDDSHV2	VDDSHV2	NA	PWR							
NA	P12, P13	VDDSHV3	VDDSHV3	NA	PWR							
G15, H14, H15	H14, J14	VDDSHV4	VDDSHV4	NA	PWR							
M14, M15, N15	K14, L14	VDDSHV5	VDDSHV5	NA	PWR							
E11, E12, E13, F14, P6, R7	E10, E11, E12, E13, F14, G14, N5, P5, P6	VDDSHV6	VDDSHV6	NA	PWR							
G5, H5, H6, K4, K5, M5, M6, N5	E5, F5, G5, H5, J5, K5, L5	VDDS_DDR	VDDS_DDR	NA	PWR							
U10	R11	VDDS_OSC	VDDS_OSC	NA	PWR							
T8	R10	VDDS_PLL_CORE_LCD	VDDS_PLL_CORE_LCD	NA	PWR							
C5	E7	VDDS_PLL_DDR	VDDS_PLL_DDR	NA	PWR							
H16	H15	VDDS_PLL_MPU	VDDS_PLL_MPU	NA	PWR							
C6	D7	VDDS_RTC	VDDS_RTC	NA	PWR							
C10	E9	VDDS_SRAM_CORE_BG	VDDS_SRAM_CORE_BG	NA	PWR							
C12	D10	VDDS_SRAM_MPU_BB	VDDS_SRAM_MPU_BB	NA	PWR							
J8, J12, J13, K15, K16, L7, L8, L12, L13,	F6, F7, G6, G7, G10, H11, J12, K6, K8, K12, L6, L7, L8, L9, M11, M13, N8, N9, N12, N13	VDD_CORE	VDD_CORE	NA	PWR							
NA	F10, F11, F12, F13, G13, H13, J13	VDD_MPU	VDD_MPU ⁽³⁰⁾	NA	PWR							
NA	A2	VDD_MPU_MON	VDD_MPU_MON (31)	NA	Α							
R5	M5	VPP	VPP	NA	PWR							
B9	A9	VREFN	VREFN	0	AP	z z	0	VDDA_ADC / VDDA_ADC	NA	NA	NA	Analog
A9	B9	VREFP	VREFP	0	AP	z z	0	VDDA_ADC / VDDA_ADC	NA	NA	NA	Analog

ZCE BALL NUMBER [1]	ZCZ BALL NUMBER [1]	PIN NAME [2]	SIGNAL NAME [3]	MODE [4]	TYPE [5]	BALL RESET STATE [6] ⁽²⁵⁾	BALL RESET REL. STATE [7]	KESEI KEL.	ZCE POWER / ZCZ POWER [9]	HYS [10]	BUFFER STRENGTH (mA) [11]	PULLUP /DOWN TYPE [12]	I/O CELL [13]
F8, F12, F13, G8, G12, H9, H10, H11, J5, J6, J9, J11, J14, J15, K8, K9, K11, K12, L5, L6, L9,	G8, G9, G11, G12, H6, H7, H8, H9, H10, H12, J6, J7, J8, J9, J10, J11, K7, K9,	vss	vss	NA	GND								
M9, M10, M11, N8,	M12, N7, N10, N11, V1, V18												
D8	E8	VSSA_ADC	VSSA_ADC	NA	GND								
P16	M14, N14	VSSA_USB	VSSA_USB	NA	GND								
V11	V11	VSS_OSC	VSS_OSC (28)	NA	Α								
NA	A5	VSS_RTC	VSS_RTC (29)	NA	Α								
A16	A10	WARMRSTn	nRESETIN_OUT	0	I/OD (8)	0	0(PU) (11)	0	VDDSHV6 / VDDSHV6	Yes	4	PU/PD	LVCMOS
C15	A15	XDMA_EVENT_INTR0	xdma_event_intr0	0	I	Z	(4)	(9)	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			timer4	2	I/O				VDDSHV6				
			clkout1	3	0								
			spi1_cs1	4	I/O								
			pr1_pru1_pru_r31_16	5	I								
			EMU2	6	I/O								
			gpio0_19	7	I/O								
B15	D14	XDMA_EVENT_INTR1	xdma_event_intr1	0	I	Z	L	7	VDDSHV6 /	Yes	4	PU/PD	LVCMOS
			tclkin	2	ı				VDDSHV6				
			clkout2	3	0								
			timer7	4	I/O								
			pr1_pru0_pru_r31_16	5	I								
			EMU3	6	I/O								
			gpio0_20	7	I/O								
W11	V10	XTALIN	OSC0_IN	0	I	Z	Z	0	VDDS_OSC / VDDS_OSC	Yes	NA	PD ⁽²⁾	LVCMOS
W12	U11	XTALOUT	OSC0_OUT	0	0	(24)	(24)	0	VDDS_OSC / VDDS_OSC	NA	NA ⁽¹⁵⁾	NA	LVCMOS

- (1) An internal 10 kohm pullup is turned on when the oscillator is disabled. The oscillator is disabled by default after power is applied.
- (2) An internal 15 kohm pulldown is turned on when the oscillator is disabled. The oscillator is enabled by default after power is applied.
- (3) Do not connect anything to this terminal.
- (4) If sysboot[5] is low on the rising edge of PWRONRSTn, this terminal has an internal pulldown turned on after reset is released. If sysboot[5] is high on the rising edge or PWRONRSTn, this terminal will initially be driven low after reset is released then it begins to toggle at the same frequency of the XTALIN terminal.
- (5) LCD DATA[15:0] terminals are respectively SYSBOOT[15:0] inputs, latched on the rising edge of PWRONRSTn.
- (6) Mode1 and Mode2 signal assignments for this terminal are only available with silicon revision 2.0 or newer devices.
- (7) Mode2 signal assignment for this terminal is only available with silicon revision 2.0 or newer devices.
- (8) Refer to the External Warm Reset section of the AM335x Technical Reference Manual for more information related to the operation of this terminal.
- (9) Reset Release Mode = 7 if sysboot[5] is low. Mode = 3 if sysboot[5] is high.
- (10) Silicon revision 1.0 devices only provide the MMC2_DAT7 signal when Mode3 is selected. Silicon revision 2.0 and newer devices implement another level of pin multiplexing which provides the original MMC2_DAT7 signal or RMII2_CRS_DV signal when Mode3 is selected. This new level of pin multiplexing is selected with bit zero of the SMA2 register. For more details refer to Section 1.2 of the AM335x Technical Reference Manual.
- (11) The 0(PU) indicates that this terminal is initially low based on the description in the AM335x Technical Reference Manual. However, it is also has a weak internal pullup applied.
- (12) The input voltage thresholds for this input are not a function of VDDSHV6. Please refer to the DC Electrical Characteristics section for details related to electrical parameters associated with this input terminal.
- (13) The internal USB PHY can be configured to multiplex the UART2_TX or UART2_RX signals to this terminal. For more details refer to USB GPIO Details section of the AM335x Technical Reference Manual.
- (14) The internal USB PHY can be configured to multiplex the UART3_TX or UART3_RX signals to this terminal. For more details refer to USB GPIO Details section of the AM335x Technical Reference Manual.
- (15) This output should only be used to source the recommended crystal circuit.
- (16) This parameter only applies when this USB PHY terminal is operating in UART2 mode.
- (17) This parameter only applies when this USB PHY terminal is operating in UART3 mode.
- (18) This terminal is a analog input used to set the switching threshold of the DDR input buffers to (VDDS DDR / 2).
- (19) This terminal is a analog passive signal that connects to an external 49.9 ohm 1%, 20mW reference resistor which is used to calibrate the DDR input/output buffers.
- (20) This terminal is analog input that may also be configured as an open-drain output.
- (21) This terminal is analog input that may also be configured as an open-source or open-drain output.
- (22) This terminal is analog input that may also be configured as an open-source output.
- (23) This terminal is high-Z when the oscillator is disabled. This terminal is driven high if RTC_XTALIN is less than VIL, driven low if RTC_XTALIN is greater than VIH, and driven to a unknown value if RTC_XTALIN is between VIL and VIH when the oscillator is enabled. The oscillator is disabled by default after power is applied.
- (24) This terminal is high-Z when the oscillator is disabled. This terminal is driven high if XTALIN is less than VIL, driven low if XTALIN is greater than VIH, and driven to a unknown value if XTALIN is between VIL and VIH when the oscillator is enabled. The oscillator is enabled by default after power is applied.
- (25) For all pins with content in the Ball Reset State column of this table, the terminal is not defined until all the supplies are ramped.
- (26) This terminal requires two power supplies, VDDA3p3v_USB0 and VDDA1p8v_USB0. The "*" character in the power supply name is a wild card that represents "3p3v" and "1p8v".
- (27) This terminal requires two power supplies, VDDA3p3v USB1 and VDDA1p8v USB1. The "*" character in the power supply name is a wild card that represents "3p3v" and "1p8v".
- (28) Refer to 节 6.2.2 for additional details about VSS OSC.
- (29) Refer to 节 6.2.2 for additional details about VSS RTC.
- (30) This power rail is connected to VDD CORE in the ZCE package.
- (31) This terminal provides a Kelvin connection to VDD MPU. It can be connected to the power supply feedback input to provide remote sensing which compensates for voltage drop in the

PCB power distribution network and package. When the Kelvin connection is not used it should be connected to the same power source as VDD_MPU.

4.3 Signal Descriptions

The AM335x device contains many peripheral interfaces. In order to reduce package size and lower overall system cost while maintaining maximum functionality, many of the AM335x terminals can multiplex up to eight signal functions. Although there are many combinations of pin multiplexing that are possible, only a certain number of sets, called I/O Sets, are valid due to timing limitations. These valid I/O Sets were carefully chosen to provide many possible application scenarios for the user.

Texas Instruments has developed a Windows-based application called Pin Mux Utility that helps a system designer select the appropriate pin-multiplexing configuration for their AM335x-based product design. The Pin Mux Utility provides a way to select valid I/O Sets of specific peripheral interfaces to ensure the pin-multiplexing configuration selected for a design only uses valid I/O Sets supported by the AM335x device.



(1) SIGNAL NAME: The signal name

(2) **DESCRIPTION:** Description of the signal

(3) TYPE: Ball type for this specific function:

- I = Input

- O = \dot{O} utput

- I/O = Input/Output

D = Open drain

DS = Differential

- A = Analog

(4) BALL: Package ball location

Table 4-3. ADC Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
AIN0	Analog Input/Output	Α	B8	B6
AIN1	Analog Input/Output	Α	A11	C7
AIN2	Analog Input/Output	Α	A8	B7
AIN3	Analog Input/Output	Α	B11	A7
AIN4	Analog Input/Output	Α	C8	C8
AIN5	Analog Input	Α	B12	B8
AIN6	Analog Input	Α	A10	A8
AIN7	Analog Input	Α	A12	C9
VREFN	Analog Negative Reference Input	AP	B9	A9
VREFP	Analog Positive Reference Input	AP	A9	B9

Table 4-4. Debug Subsystem Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
EMU0	MISC EMULATION PIN	I/O	A15	C14
EMU1	MISC EMULATION PIN	I/O	D14	B14
EMU2	MISC EMULATION PIN	I/O	A18, C15	A15, A17, C13
EMU3	MISC EMULATION PIN	I/O	B15, B18	B17, D13, D14
EMU4	MISC EMULATION PIN	I/O	B16, U17	A14, C15, T13
nTRST	JTAG TEST RESET (ACTIVE LOW)	I	A13	B10
TCK	JTAG TEST CLOCK	I	B14	A12
TDI	JTAG TEST DATA INPUT	I	B13	B11
TDO	JTAG TEST DATA OUTPUT	0	A14	A11
TMS	JTAG TEST MODE SELECT	I	C14	C11

Table 4-5. LCD Controller Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
lcd_ac_bias_en	LCD AC bias enable chip select	0	W7	R6
lcd_data0	LCD data bus	I/O	U1	R1
lcd_data1	LCD data bus	I/O	U2	R2
lcd_data10	LCD data bus	I/O	U5	U3
lcd_data11	LCD data bus	I/O	V5	U4
lcd_data12	LCD data bus	I/O	V6	V2
lcd_data13	LCD data bus	I/O	U6	V3
lcd_data14	LCD data bus	I/O	W6	V4
lcd_data15	LCD data bus	I/O	V7	T5
lcd_data16	LCD data bus	0	V17	U13



Table 4-5. LCD Controller Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
lcd_data17	LCD data bus	0	W17	V13
lcd_data18	LCD data bus	0	T13	R12
lcd_data19	LCD data bus	0	U13	T12
lcd_data2	LCD data bus	I/O	V1	R3
lcd_data20	LCD data bus	0	U12	U12
lcd_data21	LCD data bus	0	T12	T11
lcd_data22	LCD data bus	0	W16	T10
lcd_data23	LCD data bus	0	V15	U10
lcd_data3	LCD data bus	I/O	V2	R4
lcd_data4	LCD data bus	I/O	W2	T1
lcd_data5	LCD data bus	I/O	W3	T2
lcd_data6	LCD data bus	I/O	V3	T3
lcd_data7	LCD data bus	I/O	U3	T4
lcd_data8	LCD data bus	I/O	V4	U1
lcd_data9	LCD data bus	I/O	W4	U2
lcd_hsync	LCD Horizontal Sync	0	T7	R5
lcd_memory_clk	LCD MCLK	0	L19, V16	J17, V12
lcd_pclk	LCD pixel clock	0	W5	V5
lcd_vsync	LCD Vertical Sync	0	U7	U5



4.3.1 External Memory Interfaces

Table 4-6. External Memory Interfaces/DDR Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
ddr_a0	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	F3	F3
ddr_a1	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	J2	H1
ddr_a10	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	E2	F4
ddr_a11	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	G4	F2
ddr_a12	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	F4	E3
ddr_a13	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	H1	H3
ddr_a14	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	H3	H4
ddr_a15	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	E3	D3
ddr_a2	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	D1	E4
ddr_a3	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	B3	C3
ddr_a4	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	E5	C2
ddr_a5	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	A2	B1
ddr_a6	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	B1	D5
ddr_a7	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	D2	E2
ddr_a8	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	C3	D4
ddr_a9	DDR SDRAM ROW/COLUMN ADDRESS OUTPUT	0	B2	C1
ddr_ba0	DDR SDRAM BANK ADDRESS OUTPUT	0	A3	C4
ddr_ba1	DDR SDRAM BANK ADDRESS OUTPUT	0	E1	E1
ddr_ba2	DDR SDRAM BANK ADDRESS OUTPUT	0	B4	B3
ddr_casn	DDR SDRAM COLUMN ADDRESS STROBE OUTPUT (ACTIVE LOW)	0	F1	F1
ddr_ck	DDR SDRAM CLOCK OUTPUT (Differential+)	0	C2	D2
ddr_cke	DDR SDRAM CLOCK ENABLE OUTPUT	0	G3	G3
ddr_csn0	DDR SDRAM CHIP SELECT OUTPUT	0	H2	H2
ddr_d0	DDR SDRAM DATA INPUT/OUTPUT	I/O	N4	M3
ddr_d1	DDR SDRAM DATA INPUT/OUTPUT	I/O	P4	M4
ddr_d10	DDR SDRAM DATA INPUT/OUTPUT	I/O	M3	K2
ddr_d11	DDR SDRAM DATA INPUT/OUTPUT	I/O	M4	K3
ddr_d12	DDR SDRAM DATA INPUT/OUTPUT	I/O	M2	K4
ddr_d13	DDR SDRAM DATA INPUT/OUTPUT	I/O	M1	L3
ddr_d14	DDR SDRAM DATA INPUT/OUTPUT	I/O	N2	L4
ddr_d15	DDR SDRAM DATA INPUT/OUTPUT	I/O	N1	M1
ddr_d2	DDR SDRAM DATA INPUT/OUTPUT	I/O	P2	N1
ddr_d3	DDR SDRAM DATA INPUT/OUTPUT	I/O	P1	N2



Table 4-6. External Memory Interfaces/DDR Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
ddr_d4	DDR SDRAM DATA INPUT/OUTPUT	I/O	P3	N3
ddr_d5	DDR SDRAM DATA INPUT/OUTPUT	I/O	T1	N4
ddr_d6	DDR SDRAM DATA INPUT/OUTPUT	I/O	T2	P3
ddr_d7	DDR SDRAM DATA INPUT/OUTPUT	I/O	R3	P4
ddr_d8	DDR SDRAM DATA INPUT/OUTPUT	I/O	K2	J1
ddr_d9	DDR SDRAM DATA INPUT/OUTPUT	I/O	K1	K1
ddr_dqm0	DDR WRITE ENABLE / DATA MASK FOR DATA[7:0]	0	N3	M2
ddr_dqm1	DDR WRITE ENABLE / DATA MASK FOR DATA[15:8]	0	K3	J2
ddr_dqs0	DDR DATA STROBE FOR DATA[7:0] (Differential+)	I/O	R1	P1
ddr_dqs1	DDR DATA STROBE FOR DATA[15:8] (Differential+)	I/O	L1	L1
ddr_dqsn0	DDR DATA STROBE FOR DATA[7:0] (Differential-)	I/O	R2	P2
ddr_dqsn1	DDR DATA STROBE FOR DATA[15:8] (Differential-)	I/O	L2	L2
ddr_nck	DDR SDRAM CLOCK OUTPUT (Differential-)	0	C1	D1
ddr_odt	ODT OUTPUT	0	G1	G1
ddr_rasn	DDR SDRAM ROW ADDRESS STROBE OUTPUT (ACTIVE LOW)	0	F2	G4
ddr_resetn	DDR3/DDR3L RESET OUTPUT (ACTIVE LOW)	0	G2	G2
ddr_vref	Voltage Reference Input	Α	H4	J4
ddr_vtp	VTP Compensation Resistor	1	J1	J3
ddr_wen	DDR SDRAM WRITE ENABLE OUTPUT (ACTIVE LOW)	0	A4	B2

Table 4-7. External Memory Interfaces/General-Purpose Memory Controller Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpmc_a0	GPMC Address	0	U1	R1, R13
gpmc_a1	GPMC Address	0	U2, U7	R2, U5, V14
gpmc_a10	GPMC Address	0	W5	T16, V5
gpmc_a11	GPMC Address	0	W7	R6, V17
gpmc_a12	GPMC Address	0	V4	U1
gpmc_a13	GPMC Address	0	W4	U2
gpmc_a14	GPMC Address	0	U5	U3
gpmc_a15	GPMC Address	0	V5	U4
gpmc_a16	GPMC Address	0	V6	R13, V2
gpmc_a17	GPMC Address	0	U6	V14, V3
gpmc_a18	GPMC Address	0	W6	U14, V4
gpmc_a19	GPMC Address	0	V7	T14, T5
gpmc_a2	GPMC Address	0	T7, V1	R3, R5, U14
gpmc_a20	GPMC Address	0	H19	F17, R14
gpmc_a21	GPMC Address	0	H18	F18, V15
gpmc_a22	GPMC Address	0	H17	G15, U15
gpmc_a23	GPMC Address	0	G18	G16, T15
gpmc_a24	GPMC Address	0	G19	G17, V16

Table 4-7. External Memory Interfaces/General-Purpose Memory Controller Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpmc_a25	GPMC Address	0	G17	G18, U16
gpmc_a26	GPMC Address	0	NA	T16
gpmc_a27	GPMC Address	0	NA	V17
gpmc_a3	GPMC Address	0	U17, V2	R4, T13, T14
gpmc_a4	GPMC Address	0	W2	R14, T1
gpmc_a5	GPMC Address	0	W3	T2, V15
gpmc_a6	GPMC Address	0	V3	T3, U15
gpmc_a7	GPMC Address	0	U3	T15, T4
gpmc_a8	GPMC Address	0	U7	U5, V16
gpmc_a9	GPMC Address	0	T7	R5, U16
gpmc_ad0	GPMC Address and Data	I/O	W10	U7
gpmc_ad1	GPMC Address and Data	I/O	V9	V7
gpmc_ad10	GPMC Address and Data	I/O	T12	T11
gpmc_ad11	GPMC Address and Data	I/O	U12	U12
gpmc_ad12	GPMC Address and Data	I/O	U13	T12
gpmc_ad13	GPMC Address and Data	I/O	T13	R12
gpmc_ad14	GPMC Address and Data	I/O	W17	V13
gpmc_ad15	GPMC Address and Data	I/O	V17	U13
gpmc_ad2	GPMC Address and Data	I/O	V12	R8
gpmc_ad3	GPMC Address and Data	I/O	W13	Т8
gpmc_ad4	GPMC Address and Data	I/O	V13	U8
gpmc_ad5	GPMC Address and Data	I/O	W14	V8
gpmc_ad6	GPMC Address and Data	I/O	U14	R9
gpmc_ad7	GPMC Address and Data	I/O	W15	Т9
gpmc_ad8	GPMC Address and Data	I/O	V15	U10
gpmc_ad9	GPMC Address and Data	I/O	W16	T10
gpmc_advn_ale	GPMC Address Valid / Address Latch Enable	0	V10	R7
gpmc_be0n_cle	GPMC Byte Enable 0 / Command Latch Enable	0	V8	T6
gpmc_be1n	GPMC Byte Enable 1	0	U15, V18	U18, V9
gpmc_clk	GPMC Clock	I/O	V14, V16	U9, V12
gpmc_csn0	GPMC Chip Select	0	W8	V6
gpmc_csn1	GPMC Chip Select	0	V14	U9
gpmc_csn2	GPMC Chip Select	0	U15	V9
gpmc_csn3	GPMC Chip Select	0	U17	T13
gpmc_csn4	GPMC Chip Select	0	R15	T17
gpmc_csn5	GPMC Chip Select	0	W18	U17
gpmc_csn6	GPMC Chip Select	0	V18	U18
gpmc_dir	GPMC Data Direction	0	V18	U18
gpmc_oen_ren	GPMC Output / Read Enable	0	W9	T7
gpmc_wait0	GPMC Wait 0	I	R15	T17
gpmc_wait1	GPMC Wait 1	I	V16	V12
gpmc_wen	GPMC Write Enable	0	U8	U6
gpmc_wpn	GPMC Write Protect	0	W18	U17



4.3.2 General-Purpose IOs

Table 4-8. General-Purpose IOs/GPIO0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio0_0	GPIO	I/O	P17	M17
gpio0_1	GPIO	I/O	R19	M18
gpio0_10	GPIO	I/O	W6	V4
gpio0_11	GPIO	I/O	V7	T5
gpio0_12	GPIO	I/O	E17	D18
gpio0_13	GPIO	I/O	D19	D17
gpio0_14	GPIO	I/O	D18	D16
gpio0_15	GPIO	I/O	C19	D15
gpio0_16	GPIO	I/O	M17	J18
gpio0_17	GPIO	I/O	N18	K15
gpio0_18	GPIO	I/O	G16	F16
gpio0_19	GPIO	I/O	C15	A15
gpio0_2	GPIO	I/O	A18	A17
gpio0_20	GPIO	I/O	B15	D14
gpio0_21	GPIO	I/O	M18	K16
gpio0_22	GPIO	I/O	V15	U10
gpio0_23	GPIO	I/O	W16	T10
gpio0_26	GPIO	I/O	T12	T11
gpio0_27	GPIO	I/O	U12	U12
gpio0_28	GPIO	I/O	L18	K17
gpio0_29	GPIO	I/O	K18	H18
gpio0_3	GPIO	I/O	B18	B17
gpio0_30	GPIO	I/O	R15	T17
gpio0_31	GPIO	I/O	W18	U17
gpio0_4	GPIO	I/O	B17	B16
gpio0_5	GPIO	I/O	A17	A16
gpio0_6	GPIO	I/O	B16	C15
gpio0_7	GPIO	I/O	E18	C18
gpio0_8	GPIO	I/O	V6	V2
gpio0_9	GPIO	I/O	U6	V3

Table 4-9. General-Purpose IOs/GPIO1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio1_0	GPIO	I/O	W10	U7
gpio1_1	GPIO	I/O	V9	V7
gpio1_10	GPIO	I/O	E19	E15
gpio1_11	GPIO	I/O	F17	E16
gpio1_12	GPIO	I/O	U13	T12
gpio1_13	GPIO	I/O	T13	R12
gpio1_14	GPIO	I/O	W17	V13
gpio1_15	GPIO	I/O	V17	U13
gpio1_16	GPIO	I/O	NA	R13
gpio1_17	GPIO	I/O	NA	V14
gpio1_18	GPIO	I/O	NA	U14

Table 4-9. General-Purpose IOs/GPIO1 Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio1_19	GPIO	I/O	NA	T14
gpio1_2	GPIO	I/O	V12	R8
gpio1_20	GPIO	I/O	NA	R14
gpio1_21	GPIO	I/O	NA	V15
gpio1_22	GPIO	I/O	NA	U15
gpio1_23	GPIO	I/O	NA	T15
gpio1_24	GPIO	I/O	NA	V16
gpio1_25	GPIO	I/O	NA	U16
gpio1_26	GPIO	I/O	NA	T16
gpio1_27	GPIO	I/O	NA	V17
gpio1_28	GPIO	I/O	V18	U18
gpio1_29	GPIO	I/O	W8	V6
gpio1_3	GPIO	I/O	W13	T8
gpio1_30	GPIO	I/O	V14	U9
gpio1_31	GPIO	I/O	U15	V9
gpio1_4	GPIO	I/O	V13	U8
gpio1_5	GPIO	I/O	W14	V8
gpio1_6	GPIO	I/O	U14	R9
gpio1_7	GPIO	I/O	W15	T9
gpio1_8	GPIO	I/O	F19	E18
gpio1_9	GPIO	I/O	F18	E17

Table 4-10. General-Purpose IOs/GPIO2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio2_0	GPIO	I/O	U17	T13
gpio2_1	GPIO	I/O	V16	V12
gpio2_10	GPIO	I/O	W2	T1
gpio2_11	GPIO	I/O	W3	T2
gpio2_12	GPIO	I/O	V3	T3
gpio2_13	GPIO	I/O	U3	T4
gpio2_14	GPIO	I/O	V4	U1
gpio2_15	GPIO	I/O	W4	U2
gpio2_16	GPIO	I/O	U5	U3
gpio2_17	GPIO	I/O	V5	U4
gpio2_18	GPIO	I/O	N17	L17
gpio2_19	GPIO	I/O	N16	L16
gpio2_2	GPIO	I/O	V10	R7
gpio2_20	GPIO	I/O	P19	L15
gpio2_21	GPIO	I/O	P18	M16
gpio2_22	GPIO	I/O	U7	U5
gpio2_23	GPIO	I/O	T7	R5
gpio2_24	GPIO	I/O	W5	V5
gpio2_25	GPIO	I/O	W7	R6
gpio2_26	GPIO	I/O	H19	F17
gpio2_27	GPIO	I/O	H18	F18
gpio2_28	GPIO	I/O	H17	G15



Table 4-10. General-Purpose IOs/GPIO2 Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio2_29	GPIO	I/O	G18	G16
gpio2_3	GPIO	I/O	W9	T7
gpio2_30	GPIO	I/O	G19	G17
gpio2_31	GPIO	I/O	G17	G18
gpio2_4	GPIO	I/O	U8	U6
gpio2_5	GPIO	I/O	V8	T6
gpio2_6	GPIO	I/O	U1	R1
gpio2_7	GPIO	I/O	U2	R2
gpio2_8	GPIO	I/O	V1	R3
gpio2_9	GPIO	I/O	V2	R4

Table 4-11. General-Purpose IOs/GPIO3 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gpio3_0	GPIO	I/O	J19	H16
gpio3_1	GPIO	I/O	J18	H17
gpio3_10	GPIO	I/O	M19	L18
gpio3_13	GPIO	I/O	NA	F15
gpio3_14	GPIO	I/O	NA	A13
gpio3_15	GPIO	I/O	NA	B13
gpio3_16	GPIO	I/O	NA	D12
gpio3_17	GPIO	I/O	NA	C12
gpio3_18	GPIO	I/O	NA	B12
gpio3_19	GPIO	I/O	NA	C13
gpio3_2	GPIO	I/O	K19	J15
gpio3_20	GPIO	I/O	NA	D13
gpio3_21	GPIO	I/O	NA	A14
gpio3_3	GPIO	I/O	K17	J16
gpio3_4	GPIO	I/O	L19	J17
gpio3_5	GPIO	I/O	C18	C17
gpio3_6	GPIO	I/O	B19	C16
gpio3_7	GPIO	I/O	A15	C14
gpio3_8	GPIO	I/O	D14	B14
gpio3_9	GPIO	I/O	N19	K18

4.3.3 Miscellaneous

Table 4-12. Miscellaneous/Miscellaneous Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
clkout1	Clock out1	0	C15	A15
clkout2	Clock out2	0	B15	D14
ENZ_KALDO_1P8V	Active low enable input for internal CAP_VDD_RTC voltage regulator	I	A7	B4
EXT_WAKEUP	EXT_WAKEUP input	I	B5	C5
nNMI	External Interrupt to ARM Cortex-A8 core	I	C17	B18
nRESETIN_OUT	Active low Warm Reset	I/OD	A16	A10
OSC0_IN	High frequency oscillator input	I	W11	V10
OSC0_OUT	High frequency oscillator output	0	W12	U11
OSC1_IN	Low frequency (32.768 kHz) Real Time Clock oscillator input	I	A6	A6
OSC1_OUT	Low frequency (32.768 kHz) Real Time Clock oscillator output	0	A5	A4
PMIC_POWER_EN	PMIC_POWER_EN output	0	C7	C6
porz	Active low Power on Reset	I	E15	B15
RTC_PORz	Active low RTC reset input	I	B7	B5
tclkin	Timer Clock In	I	B15	D14
xdma_event_intr0	External DMA Event or Interrupt 0	I	C15	A15
xdma_event_intr1	External DMA Event or Interrupt 1	I	B15	D14
xdma_event_intr2	External DMA Event or Interrupt 2	I	B16, E18, K18	C15, C18, H18



4.3.3.1 eCAP

Table 4-13. eCAP/eCAP0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eCAP0_in_PWM0_out	Enhanced Capture 0 input or Auxiliary PWM0 output	I/O	E18	C18

Table 4-14. eCAP/eCAP1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eCAP1_in_PWM1_out	Enhanced Capture 1 input or Auxiliary PWM1 output	I/O	B16, B19, F17	C15, C16, E16

Table 4-15. eCAP/eCAP2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eCAP2_in_PWM2_out	Enhanced Capture 2 input or Auxiliary PWM2 output	I/O	C18, E19	C12, C17, E15

4.3.3.2 eHRPWM

Table 4-16. eHRPWM/eHRPWM0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
ehrpwm0A	eHRPWM0 A output.	0	A18	A13, A17
ehrpwm0B	eHRPWM0 B output.	0	B18	B13, B17
ehrpwm0_synci	Sync input to eHRPWM0 module from an external pin	I	A17	A16, C12
ehrpwm0_synco	Sync Output from eHRPWM0 module to an external pin	0	U12, V2, W4	R4, U12, U2, V14
ehrpwm0_tripzone_input	eHRPWM0 trip zone input	I	B17	B16, D12

Table 4-17. eHRPWM/eHRPWM1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
ehrpwm1A	eHRPWM1 A output.	0	U5	U14, U3
ehrpwm1B	eHRPWM1 B output.	0	V5	T14, U4
ehrpwm1_tripzone_input	eHRPWM1 trip zone input	I	V4	R13, U1

Table 4-18. eHRPWM/eHRPWM2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
ehrpwm2A	eHRPWM2 A output.	0	U1, V15	R1, U10
ehrpwm2B	eHRPWM2 B output.	0	U2, W16	R2, T10
ehrpwm2_tripzone_input	eHRPWM2 trip zone input	I	T12, V1	R3, T11



4.3.3.3 eQEP

Table 4-19. eQEP/eQEP0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eQEP0A_in	eQEP0A quadrature input	I	M18	B12, K16
eQEP0B_in	eQEP0B quadrature input	I	L18	C13, K17
eQEP0_index	eQEP0 index.	I/O	K17	D13, J16
eQEP0_strobe	eQEP0 strobe.	I/O	P19	A14, L15

Table 4-20. eQEP/eQEP1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eQEP1A_in	eQEP1A quadrature input	I	V6	R14, V2
eQEP1B_in	eQEP1B quadrature input	I	U6	V15, V3
eQEP1_index	eQEP1 index.	I/O	W6	U15, V4
eQEP1_strobe	eQEP1 strobe.	I/O	V7	T15, T5

Table 4-21. eQEP/eQEP2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
eQEP2A_in	eQEP2A quadrature input	I	U13, W2	T1, T12
eQEP2B_in	eQEP2B quadrature input	I	T13, W3	R12, T2
eQEP2_index	eQEP2 index.	I/O	V3, W17	T3, V13
eQEP2_strobe	eQEP2 strobe.	I/O	U3, V17	T4, U13

4.3.3.4 Timer

Table 4-22. Timer/Timer4 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
timer4	Timer trigger event / PWM out	I/O	C15, C18, K17, V10	A15, C17, J16, R7

Table 4-23. Timer/Timer5 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
timer5	Timer trigger event / PWM out	I/O	D19, H19, R19, V8	D17, F17, M18, T6

Table 4-24. Timer/Timer6 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
timer6	Timer trigger event / PWM out	I/O	E17, H18, P17, U8	D18, F18, M17, U6

Table 4-25. Timer/Timer7 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
timer7	Timer trigger event / PWM out	I/O	B15, B19, F19, W9	C16, D14, E18, T7



4.3.4 PRU-ICSS

Table 4-26. PRU-ICSS/eCAP Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_ecap0_ecap_capin_apwm_o	Enhanced capture input or Auxiliary PWM out	I/O	E18, V17	C18, U13

Table 4-27. PRU-ICSS/ECAT Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_edc_latch0_in	Data In	I	E17	D18
pr1_edc_latch1_in	Data In	1	D19	D17
pr1_edc_sync0_out	Data Out	0	F19	E18
pr1_edc_sync1_out	Data Out	0	F18	E17
pr1_edio_data_in0	Data In	I	B17	B16
pr1_edio_data_in1	Data In	1	A17	A16
pr1_edio_data_in2	Data In	1	U7	U5
pr1_edio_data_in3	Data In	I	T7	R5
pr1_edio_data_in4	Data In	I	W5	V5
pr1_edio_data_in5	Data In	I	W7	R6
pr1_edio_data_in6	Data In	1	V14, V3	T3, U9
pr1_edio_data_in7	Data In	1	U15, U3	T4, V9
pr1_edio_data_out0	Data Out	0	B17	B16
pr1_edio_data_out1	Data Out	0	A17	A16
pr1_edio_data_out2	Data Out	0	U7	U5
pr1_edio_data_out3	Data Out	0	T7	R5
pr1_edio_data_out4	Data Out	0	W5	V5
pr1_edio_data_out5	Data Out	0	W7	R6
pr1_edio_data_out6	Data Out	0	V14, V3	T3, U9
pr1_edio_data_out7	Data Out	0	U15, U3	T4, V9
pr1_edio_latch_in	Latch In	1	B18	B17
pr1_edio_sof	Start of Frame	0	A18	A17

Table 4-28. PRU-ICSS/MDIO Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_mdio_data	MDIO Data	I/O	U17	T13
pr1_mdio_mdclk	MDIO CIk	0	V16	V12

Table 4-29. PRU-ICSS/MII0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_mii0_col	MII Collision Detect	I	W16	T10
pr1_mii0_crs	MII Carrier Sense	I	U17, W5	T13, V5
pr1_mii0_rxd0	MII Receive Data bit 0	I	V5	U4
pr1_mii0_rxd1	MII Receive Data bit 1	I	U5	U3
pr1_mii0_rxd2	MII Receive Data bit 2	I	W4	U2
pr1_mii0_rxd3	MII Receive Data bit 3	I	V4	U1
pr1_mii0_rxdv	MII Receive Data Valid	I	V7	T5
pr1_mii0_rxer	MII Receive Data Error	I	U6	V3

Table 4-29. PRU-ICSS/MII0 Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_mii0_rxlink	MII Receive Link	I	V6	V2
pr1_mii0_txd0	MII Transmit Data bit 0	0	W17, W3	T2, V13
pr1_mii0_txd1	MII Transmit Data bit 1	0	T13, W2	R12, T1
pr1_mii0_txd2	MII Transmit Data bit 2	0	U13, V2	R4, T12
pr1_mii0_txd3	MII Transmit Data bit 3	0	U12, V1	R3, U12
pr1_mii0_txen	MII Transmit Enable	0	T12, U2	R2, T11
pr1_mii_mr0_clk	MII Receive Clock	I	W6	V4
pr1_mii_mt0_clk	MII Transmit Clock	I	U1, V15	R1, U10

Table 4-30. PRU-ICSS/MII1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_mii1_col	MII Collision Detect	ı	R15	T17
pr1_mii1_crs	MII Carrier Sense	1	V16, W7	R6, V12
pr1_mii1_rxd0	MII Receive Data bit 0	1	NA	V16
pr1_mii1_rxd1	MII Receive Data bit 1	1	NA	T15
pr1_mii1_rxd2	MII Receive Data bit 2	1	NA	U15
pr1_mii1_rxd3	MII Receive Data bit 3	1	NA	V15
pr1_mii1_rxdv	MII Receive Data Valid	1	NA	T16
pr1_mii1_rxer	MII Receive Data Error	1	NA	V17
pr1_mii1_rxlink	MII Receive Link	1	V18	U18
pr1_mii1_txd0	MII Transmit Data bit 0	0	NA	R14
pr1_mii1_txd1	MII Transmit Data bit 1	0	NA	T14
pr1_mii1_txd2	MII Transmit Data bit 2	0	NA	U14
pr1_mii1_txd3	MII Transmit Data bit 3	0	NA	V14
pr1_mii1_txen	MII Transmit Enable	0	W18	U17
pr1_mii_mr1_clk	MII Receive Clock	1	NA	U16
pr1_mii_mt1_clk	MII Transmit Clock	1	NA	R13

Table 4-31. PRU-ICSS/UART0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_uart0_cts_n	UART Clear to Send	I	A18, E17	A17, D18
pr1_uart0_rts_n	UART Request to Send	0	B18, D19	B17, D17
pr1_uart0_rxd	UART Receive Data		B17, D18	B16, D16
pr1_uart0_txd	UART Transmit Data	0	A17, C19	A16, D15



4.3.4.1 PRU0

Table 4-32. PRU0/General-Purpose Inputs Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_pru0_pru_r31_0	PRU0 Data In	1	NA	A13
pr1_pru0_pru_r31_1	PRU0 Data In	1	NA	B13
pr1_pru0_pru_r31_10	PRU0 Data In	1	H17	G15
pr1_pru0_pru_r31_11	PRU0 Data In	1	G18	G16
pr1_pru0_pru_r31_12	PRU0 Data In	1	G19	G17
pr1_pru0_pru_r31_13	PRU0 Data In	1	G17	G18
pr1_pru0_pru_r31_14	PRU0 Data In	1	W17	V13
pr1_pru0_pru_r31_15	PRU0 Data In	1	V17	U13
pr1_pru0_pru_r31_16	PRU0 Data In Capture Enable	1	B15, C19	D14, D15
pr1_pru0_pru_r31_2	PRU0 Data In	1	NA	D12
pr1_pru0_pru_r31_3	PRU0 Data In	1	NA	C12
pr1_pru0_pru_r31_4	PRU0 Data In	1	NA	B12
pr1_pru0_pru_r31_5	PRU0 Data In	1	NA	C13
pr1_pru0_pru_r31_6	PRU0 Data In	1	NA	D13
pr1_pru0_pru_r31_7	PRU0 Data In	1	NA	A14
pr1_pru0_pru_r31_8	PRU0 Data In	1	H19	F17
pr1_pru0_pru_r31_9	PRU0 Data In	1	H18	F18

Table 4-33. PRU0/General-Purpose Outputs Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_pru0_pru_r30_0	PRU0 Data Out	0	NA	A13
pr1_pru0_pru_r30_1	PRU0 Data Out	0	NA	B13
pr1_pru0_pru_r30_10	PRU0 Data Out	0	H17	G15
pr1_pru0_pru_r30_11	PRU0 Data Out	0	G18	G16
pr1_pru0_pru_r30_12	PRU0 Data Out	0	G19	G17
pr1_pru0_pru_r30_13	PRU0 Data Out	0	G17	G18
pr1_pru0_pru_r30_14	PRU0 Data Out	0	U13	T12
pr1_pru0_pru_r30_15	PRU0 Data Out	0	T13	R12
pr1_pru0_pru_r30_2	PRU0 Data Out	0	NA	D12
pr1_pru0_pru_r30_3	PRU0 Data Out	0	NA	C12
pr1_pru0_pru_r30_4	PRU0 Data Out	0	NA	B12
pr1_pru0_pru_r30_5	PRU0 Data Out	0	NA	C13
pr1_pru0_pru_r30_6	PRU0 Data Out	0	NA	D13
pr1_pru0_pru_r30_7	PRU0 Data Out	0	NA	A14
pr1_pru0_pru_r30_8	PRU0 Data Out	0	H19	F17
pr1_pru0_pru_r30_9	PRU0 Data Out	0	H18	F18

4.3.4.2 PRU1

Table 4-34. PRU1/General-Purpose Inputs Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_pru1_pru_r31_0	PRU1 Data In	1	U1	R1
pr1_pru1_pru_r31_1	PRU1 Data In	1	U2	R2
pr1_pru1_pru_r31_10	PRU1 Data In	1	W5	V5
pr1_pru1_pru_r31_11	PRU1 Data In	1	W7	R6
pr1_pru1_pru_r31_12	PRU1 Data In	1	V14	U9
pr1_pru1_pru_r31_13	PRU1 Data In	1	U15	V9
pr1_pru1_pru_r31_14	PRU1 Data In	1	E19	E15
pr1_pru1_pru_r31_15	PRU1 Data In	1	F17	E16
pr1_pru1_pru_r31_16	PRU1 Data In Capture Enable	1	C15, D18	A15, D16
pr1_pru1_pru_r31_2	PRU1 Data In	1	V1	R3
pr1_pru1_pru_r31_3	PRU1 Data In	1	V2	R4
pr1_pru1_pru_r31_4	PRU1 Data In	1	W2	T1
pr1_pru1_pru_r31_5	PRU1 Data In	1	W3	T2
pr1_pru1_pru_r31_6	PRU1 Data In	1	V3	T3
pr1_pru1_pru_r31_7	PRU1 Data In	1	U3	T4
pr1_pru1_pru_r31_8	PRU1 Data In	1	U7	U5
pr1_pru1_pru_r31_9	PRU1 Data In	1	T7	R5

Table 4-35. PRU1/General-Purpose Outputs Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
pr1_pru1_pru_r30_0	PRU1 Data Out	0	U1	R1
pr1_pru1_pru_r30_1	PRU1 Data Out	0	U2	R2
pr1_pru1_pru_r30_10	PRU1 Data Out	0	W5	V5
pr1_pru1_pru_r30_11	PRU1 Data Out	0	W7	R6
pr1_pru1_pru_r30_12	PRU1 Data Out	0	V14	U9
pr1_pru1_pru_r30_13	PRU1 Data Out	0	U15	V9
pr1_pru1_pru_r30_14	PRU1 Data Out	0	E19	E15
pr1_pru1_pru_r30_15	PRU1 Data Out	0	F17	E16
pr1_pru1_pru_r30_2	PRU1 Data Out	0	V1	R3
pr1_pru1_pru_r30_3	PRU1 Data Out	0	V2	R4
pr1_pru1_pru_r30_4	PRU1 Data Out	0	W2	T1
pr1_pru1_pru_r30_5	PRU1 Data Out	0	W3	T2
pr1_pru1_pru_r30_6	PRU1 Data Out	0	V3	T3
pr1_pru1_pru_r30_7	PRU1 Data Out	0	U3	T4
pr1_pru1_pru_r30_8	PRU1 Data Out	0	U7	U5
pr1_pru1_pru_r30_9	PRU1 Data Out	0	T7	R5



4.3.5 Removable Media Interfaces

Table 4-36. Removable Media Interfaces/MMC0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mmc0_clk	MMC/SD/SDIO Clock	I/O	G19	G17
mmc0_cmd	MMC/SD/SDIO Command	I/O	G17	G18
mmc0_dat0	MMC/SD/SDIO Data Bus	I/O	G18	G16
mmc0_dat1	MMC/SD/SDIO Data Bus	I/O	H17	G15
mmc0_dat2	MMC/SD/SDIO Data Bus	I/O	H18	F18
mmc0_dat3	MMC/SD/SDIO Data Bus	I/O	H19	F17
mmc0_dat4	MMC/SD/SDIO Data Bus	I/O	N16	L16
mmc0_dat5	MMC/SD/SDIO Data Bus	I/O	N17	L17
mmc0_dat6	MMC/SD/SDIO Data Bus	I/O	M19	L18
mmc0_dat7	MMC/SD/SDIO Data Bus	I/O	N19	K18
mmc0_pow	MMC/SD Power Switch Control	0	B16, K18	C15, H18
mmc0_sdcd	SD Card Detect	1	B16, P17	A13, C15, M17
mmc0_sdwp	SD Write Protect	1	E18, R19	B12, C18, M18

Table 4-37. Removable Media Interfaces/MMC1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mmc1_clk	MMC/SD/SDIO Clock	I/O	L18, R19, V14	K17, M18, U9
mmc1_cmd	MMC/SD/SDIO Command	I/O	M18, P17, U15	K16, M17, V9
mmc1_dat0	MMC/SD/SDIO Data Bus	I/O	N19, V15, W10	K18, U10, U7
mmc1_dat1	MMC/SD/SDIO Data Bus	I/O	M19, V9, W16	L18, T10, V7
mmc1_dat2	MMC/SD/SDIO Data Bus	I/O	N17, T12, V12	L17, R8, T11
mmc1_dat3	MMC/SD/SDIO Data Bus	I/O	N16, U12, W13	L16, T8, U12
mmc1_dat4	MMC/SD/SDIO Data Bus	I/O	U13, V13	T12, U8
mmc1_dat5	MMC/SD/SDIO Data Bus	I/O	T13, W14	R12, V8
mmc1_dat6	MMC/SD/SDIO Data Bus	I/O	U14, W17	R9, V13
mmc1_dat7	MMC/SD/SDIO Data Bus	I/O	V17, W15	T9, U13
mmc1_sdcd	SD Card Detect	I	R15	B13, T17
mmc1_sdwp	SD Write Protect	I	B17, D18	B16, D16

Table 4-38. Removable Media Interfaces/MMC2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mmc2_clk	MMC/SD/SDIO Clock	I/O	P19, R19, V16	L15, M18, V12
mmc2_cmd	MMC/SD/SDIO Command	I/O	K17, P17, U17	J16, M17, T13
mmc2_dat0	MMC/SD/SDIO Data Bus	I/O	L19, U13	J17, T12, V14
mmc2_dat1	MMC/SD/SDIO Data Bus	I/O	M17, T13	J18, R12, U14
mmc2_dat2	MMC/SD/SDIO Data Bus	I/O	N18, W17	K15, T14, V13
mmc2_dat3	MMC/SD/SDIO Data Bus	I/O	J19, V17, V18	H16, U13, U18
mmc2_dat4	MMC/SD/SDIO Data Bus	I/O	V15	U10, U15
mmc2_dat5	MMC/SD/SDIO Data Bus	I/O	W16	T10, T15
mmc2_dat6	MMC/SD/SDIO Data Bus	I/O	T12	T11, V16
mmc2_dat7	MMC/SD/SDIO Data Bus	I/O	U12	U12
mmc2_sdcd	SD Card Detect	1	W18	D12, U17
mmc2_sdwp	SD Write Protect	1	A17, C19	A16, D15

4.3.6 Serial Communication Interfaces

4.3.6.1 CAN

Table 4-39. CAN/DCAN0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
dcan0_rx	DCAN0 Receive Data	I	D19, F17, N18	D17, E16, K15
dcan0_tx	DCAN0 Transmit Data	0	E17, E19, M17	D18, E15, J18

Table 4-40. CAN/DCAN1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
dcan1_rx	DCAN1 Receive Data	I	C19, F18, G17	D15, E17, G18
dcan1_tx	DCAN1 Transmit Data	0	D18, F19, G19	D16, E18, G17



4.3.6.2 GEMAC_CPSW

Table 4-41. GEMAC_CPSW/MDIO Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mdio_clk	MDIO CIk	0	R19	M18
mdio_data	MDIO Data	I/O	P17	M17

Table 4-42. GEMAC_CPSW/MII1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gmii1_col	MII Colision	1	J19	H16
gmii1_crs	MII Carrier Sense	1	J18	H17
gmii1_rxclk	MII Receive Clock	1	M19	L18
gmii1_rxd0	MII Receive Data bit 0	1	P18	M16
gmii1_rxd1	MII Receive Data bit 1	1	P19	L15
gmii1_rxd2	MII Receive Data bit 2	1	N16	L16
gmii1_rxd3	MII Receive Data bit 3	1	N17	L17
gmii1_rxdv	MII Receive Data Valid	1	L19	J17
gmii1_rxer	MII Receive Data Error	1	K19	J15
gmii1_txclk	MII Transmit Clock	1	N19	K18
gmii1_txd0	MII Transmit Data bit 0	0	L18	K17
gmii1_txd1	MII Transmit Data bit 1	0	M18	K16
gmii1_txd2	MII Transmit Data bit 2	0	N18	K15
gmii1_txd3	MII Transmit Data bit 3	0	M17	J18
gmii1_txen	MII Transmit Enable	0	K17	J16

Table 4-43. GEMAC_CPSW/MII2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
gmii2_col	MII Colision	1	V18	U18
gmii2_crs	MII Carrier Sense	1	R15	T17
gmii2_rxclk	MII Receive Clock	1	NA	T15
gmii2_rxd0	MII Receive Data bit 0	1	NA	V17
gmii2_rxd1	MII Receive Data bit 1	1	NA	T16
gmii2_rxd2	MII Receive Data bit 2	1	NA	U16
gmii2_rxd3	MII Receive Data bit 3	1	NA	V16
gmii2_rxdv	MII Receive Data Valid	1	NA	V14
gmii2_rxer	MII Receive Data Error	1	W18	U17
gmii2_txclk	MII Transmit Clock	1	NA	U15
gmii2_txd0	MII Transmit Data bit 0	0	NA	V15
gmii2_txd1	MII Transmit Data bit 1	0	NA	R14
gmii2_txd2	MII Transmit Data bit 2	0	NA	T14
gmii2_txd3	MII Transmit Data bit 3	0	NA	U14
gmii2_txen	MII Transmit Enable	0	NA	R13

Table 4-44. GEMAC_CPSW/RGMII1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rgmii1_rclk	RGMII Receive Clock	I	M19	L18

Table 4-44. GEMAC_CPSW/RGMII1 Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rgmii1_rctl	RGMII Receive Control	I	L19	J17
rgmii1_rd0	RGMII Receive Data bit 0	I	P18	M16
rgmii1_rd1	RGMII Receive Data bit 1	I	P19	L15
rgmii1_rd2	RGMII Receive Data bit 2	I	N16	L16
rgmii1_rd3	RGMII Receive Data bit 3	1	N17	L17
rgmii1_tclk	RGMII Transmit Clock	0	N19	K18
rgmii1_tctl	RGMII Transmit Control	0	K17	J16
rgmii1_td0	RGMII Transmit Data bit 0	0	L18	K17
rgmii1_td1	RGMII Transmit Data bit 1	0	M18	K16
rgmii1_td2	RGMII Transmit Data bit 2	0	N18	K15
rgmii1_td3	RGMII Transmit Data bit 3	0	M17	J18

Table 4-45. GEMAC_CPSW/RGMII2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rgmii2_rclk	RGMII Receive Clock	I	NA	T15
rgmii2_rctl	RGMII Receive Control	I	NA	V14
rgmii2_rd0	RGMII Receive Data bit 0	I	NA	V17
rgmii2_rd1	RGMII Receive Data bit 1	I	NA	T16
rgmii2_rd2	RGMII Receive Data bit 2	I	NA	U16
rgmii2_rd3	RGMII Receive Data bit 3	I	NA	V16
rgmii2_tclk	RGMII Transmit Clock	0	NA	U15
rgmii2_tctl	RGMII Transmit Control	0	NA	R13
rgmii2_td0	RGMII Transmit Data bit 0	0	NA	V15
rgmii2_td1	RGMII Transmit Data bit 1	0	NA	R14
rgmii2_td2	RGMII Transmit Data bit 2	0	NA	T14
rgmii2_td3	RGMII Transmit Data bit 3	0	NA	U14

Table 4-46. GEMAC_CPSW/RMII1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rmii1_crs_dv	RMII Carrier Sense / Data Valid	I	J18	H17
rmii1_refclk	RMII Reference Clock	I/O	K18	H18
rmii1_rxd0	RMII Receive Data bit 0	I	P18	M16
rmii1_rxd1	RMII Receive Data bit 1	I	P19	L15
rmii1_rxer	RMII Receive Data Error	I	K19	J15
rmii1_txd0	RMII Transmit Data bit 0	0	L18	K17
rmii1_txd1	RMII Transmit Data bit 1	0	M18	K16
rmii1_txen	RMII Transmit Enable	0	K17	J16

Table 4-47. GEMAC_CPSW/RMII2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rmii2_crs_dv	RMII Carrier Sense / Data Valid	I	R15, U17	T13, T17
rmii2_refclk	RMII Reference Clock	I/O	J19	H16
rmii2_rxd0	RMII Receive Data bit 0	I	NA	V17
rmii2_rxd1	RMII Receive Data bit 1	I	NA	T16

www.ti.com.cn

Table 4-47. GEMAC_CPSW/RMII2 Signals Description (continued)

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
rmii2_rxer	RMII Receive Data Error	I	W18	U17
rmii2_txd0	RMII Transmit Data bit 0	0	NA	V15
rmii2_txd1	RMII Transmit Data bit 1	0	NA	R14
rmii2_txen	RMII Transmit Enable	0	NA	R13

4.3.6.3 I2C

Table 4-48. I2C/I2C0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
I2C0_SCL	I2C0 Clock	I/OD	B19	C16
I2C0_SDA	I2C0 Data	I/OD	C18	C17

Table 4-49. I2C/I2C1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
I2C1_SCL	I2C1 Clock	I/OD	A17, C19, F18, K19	A16, D15, E17, J15
I2C1_SDA	I2C1 Data	I/OD	B17, D18, F19, J18	B16, D16, E18, H17

Table 4-50. I2C/I2C2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
I2C2_SCL	I2C2 Clock	I/OD	B18, D19, F17	B17, D17, E16
I2C2_SDA	I2C2 Data	I/OD	A18, E17, E19	A17, D18, E15



4.3.6.4 McASP

Table 4-51. McASP/MCASP0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mcasp0_aclkr	McASP0 Receive Bit Clock	I/O	L19, V18, V6	B12, J17, U18, V2
mcasp0_aclkx	McASP0 Transmit Bit Clock	I/O	N19, V4	A13, K18, U1, V16
mcasp0_ahclkr	McASP0 Receive Master Clock	I/O	V5	C12, U4
mcasp0_ahclkx	McASP0 Transmit Master Clock	I/O	N18, V7	A14, K15, T5
mcasp0_axr0	McASP0 Serial Data (IN/OUT)	I/O	N17, U5	D12, L17, T16, U3
mcasp0_axr1	McASP0 Serial Data (IN/OUT)	I/O	N16, W6	D13, L16, V17, V4
mcasp0_axr2	McASP0 Serial Data (IN/OUT)	I/O	J19, V5, V6	B12, C12, H16, U4, V2
mcasp0_axr3	McASP0 Serial Data (IN/OUT)	I/O	P18, U6, V7	A14, C13, M16, T5, V3
mcasp0_fsr	McASP0 Receive Frame Sync	I/O	M17, U6, V16	C13, J18, V12, V3
mcasp0_fsx	McASP0 Transmit Frame Sync	I/O	M19, W4	B13, L18, U16, U2

Table 4-52. McASP/MCASP1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
mcasp1_aclkr	McASP1 Receive Bit Clock	I/O	L18, P18	K17, M16
mcasp1_aclkx	McASP1 Transmit Bit Clock	I/O	J18, L19	B12, H17, J17
mcasp1_ahclkr	McASP1 Receive Master Clock	I/O	P18	M16
mcasp1_ahclkx	McASP1 Transmit Master Clock	I/O	K18, P18	H18, M16
mcasp1_axr0	McASP1 Serial Data (IN/OUT)	I/O	K17, N18	D13, J16, K15
mcasp1_axr1	McASP1 Serial Data (IN/OUT)	I/O	M18	A14, K16
mcasp1_axr2	McASP1 Serial Data (IN/OUT)	I/O	J19, L18	H16, K17
mcasp1_axr3	McASP1 Serial Data (IN/OUT)	I/O	K18, P19	H18, L15
mcasp1_fsr	McASP1 Receive Frame Sync	I/O	M18, P19	K16, L15
mcasp1_fsx	McASP1 Transmit Frame Sync	I/O	K19, M17	C13, J15, J18

4.3.6.5 SPI

Table 4-53. SPI/SPI0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
spi0_cs0	SPI Chip Select	I/O	A17	A16
spi0_cs1	SPI Chip Select	I/O	B16	C15
spi0_d0	SPI Data	I/O	B18	B17
spi0_d1	SPI Data	I/O	B17	B16
spi0_sclk	SPI Clock	I/O	A18	A17

Table 4-54. SPI/SPI1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
spi1_cs0	SPI Chip Select	I/O	E17, E19, F18, K18	C12, D18, E15, E17, H18
spi1_cs1	SPI Chip Select	I/O	C15, D19, E18, F17	A15, C18, D17, E16
spi1_d0	SPI Data	I/O	F19, J18	B13, E18, H17
spi1_d1	SPI Data	I/O	F18, K19	D12, E17, J15
spi1_sclk	SPI Clock	I/O	E18, J19	A13, C18, H16



4.3.6.6 UART

Table 4-55. UART/UART0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
uart0_ctsn	UART Clear to Send	I	F19	E18
uart0_rtsn	UART Request to Send	0	F18	E17
uart0_rxd	UART Receive Data	I	E19	E15
uart0_txd	UART Transmit Data	0	F17	E16

Table 4-56. UART/UART1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
uart1_ctsn	UART Clear to Send	I	E17	D18
uart1_dcdn	UART Data Carrier Detect	I	H19, N19	F17, K18
uart1_dsrn	UART Data Set Ready	I	H18, M19	F18, L18
uart1_dtrn	UART Data Terminal Ready	0	H17, N17	G15, L17
uart1_rin	UART Ring Indicator	I	G18, N16	G16, L16
uart1_rtsn	UART Request to Send	0	D19	D17
uart1_rxd	UART Receive Data	I	D18	D16
uart1_txd	UART Transmit Data	0	C19	D15

Table 4-57. UART/UART2 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
uart2_ctsn	UART Clear to Send	I	C18, V4	C17, U1
uart2_rtsn	UART Request to Send	0	B19, W4	C16, U2
uart2_rxd	UART Receive Data	I	A18, G19, J18, N19	A17, G17, H17, K18
uart2_txd	UART Transmit Data	0	B18, G17, K19, M19	B17, G18, J15, L18

Table 4-58. UART/UART3 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
uart3_ctsn	UART Clear to Send	I	G19, P17, U5	G17, M17, U3
uart3_rtsn	UART Request to Send	0	G17, R19, V5	G18, M18, U4
uart3_rxd	UART Receive Data	I	B16, H17, N17	C15, G15, L17
uart3_txd	UART Transmit Data	0	E18, G18, N16	C18, G16, L16

Table 4-59. UART/UART4 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]		ZCE BALL [4]	ZCZ BALL [4]
uart4_ctsn	UART Clear to Send	I	H19, V6	F17, V2
uart4_rtsn	UART Request to Send	0	H18, U6	F18, V3
uart4_rxd	UART Receive Data	I	F19, M17, R15	E18, J18, T17
uart4_txd	UART Transmit Data	0	F18, N18, W18	E17, K15, U17



Table 4-60. UART/UART5 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
uart5_ctsn	UART Clear to Send	I	H17, J18, W6	G15, H17, V4
uart5_rtsn	UART Request to Send	0	G18, K19, V7	G16, J15, T5
uart5_rxd	UART Receive Data	I	J19, P17, W4, W6	H16, M17, U2, V4
uart5_txd	UART Transmit Data	0	K18, L19, R19, V4	H18, J17, M18, U1



4.3.6.7 USB

Table 4-61. USB/USB0 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
USB0_CE	USB0 Active high Charger Enable output	Α	T18	M15
USB0_DM	USB0 Data minus	Α	U18	N18
USB0_DP	USB0 Data plus	Α	U19	N17
USB0_DRVVBUS	USB0 Active high VBUS control output	0	G16	F16
USB0_ID	USB0 OTG ID (Micro-A or Micro-B Plug)	Α	V19	P16
USB0_VBUS	USB0 VBUS	Α	T19	P15

Table 4-62. USB/USB1 Signals Description

SIGNAL NAME [1]	DESCRIPTION [2]	TYPE [3]	ZCE BALL [4]	ZCZ BALL [4]
USB1_CE	USB1 Active high Charger Enable output	Α	NA	P18
USB1_DM	USB1 Data minus	Α	NA	R18
USB1_DP	USB1 Data plus	Α	NA	R17
USB1_DRVVBUS	USB1 Active high VBUS control output	0	NA	F15
USB1_ID	USB1 OTG ID (Micro-A or Micro-B Plug)	Α	NA	P17
USB1_VBUS	USB1 VBUS	Α	NA	T18



5 Specifications

5.1 Absolute Maximum Ratings

over junction temperature range (unless otherwise noted)(1)(2)

		MIN	MAX	UNIT	
VDD_MPU ⁽³⁾	Supply voltage for the MPU core domain	-0.5	1.5	V	
VDD_CORE	Supply voltage for the core domain	-0.5	1.5	V	
CAP_VDD_RTC ⁽⁴⁾	Supply voltage for the RTC core domain -0.5 1.5				
VPP ⁽⁵⁾	Supply voltage for the FUSE ROM domain	-0.5	2.2	V	
VDDS_RTC	Supply voltage for the RTC domain	-0.5	2.1	V	
VDDS_OSC	Supply voltage for the System oscillator	-0.5	2.1	V	
VDDS_SRAM_CORE_BG	Supply voltage for the Core SRAM LDOs	-0.5	2.1	V	
VDDS_SRAM_MPU_BB	Supply voltage for the MPU SRAM LDOs	-0.5	2.1	V	
VDDS_PLL_DDR	Supply voltage for the DPLL DDR	-0.5	2.1	V	
VDDS_PLL_CORE_LCD	Supply voltage for the DPLL Core and LCD	-0.5	2.1	V	
VDDS_PLL_MPU	Supply voltage for the DPLL MPU	-0.5	2.1	V	
VDDS_DDR	Supply voltage for the DDR I/O domain	-0.5	2.1	V	
VDDS	Supply voltage for all dual-voltage I/O domains	-0.5	2.1	V	
VDDA1P8V_USB0	Supply voltage for USBPHY	-0.5	2.1	V	
VDDA1P8V_USB1 ⁽⁶⁾	Supply voltage for USBPHY	-0.5	2.1	V	
VDDA_ADC	Supply voltage for ADC	-0.5	2.1	V	
VDDSHV1	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDSHV2 ⁽⁶⁾	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDSHV3 ⁽⁶⁾	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDSHV4	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDSHV5	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDSHV6	Supply voltage for the dual-voltage I/O domain	-0.5	3.8	V	
VDDA3P3V_USB0	Supply voltage for USBPHY	-0.5	4	V	
VDDA3P3V_USB1 ⁽⁶⁾	Supply voltage for USBPHY	-0.5	4	V	
USB0_VBUS ⁽⁷⁾	Supply voltage for USB VBUS comparator input	-0.5	5.25	V	
USB1_VBUS ⁽⁶⁾⁽⁷⁾	Supply voltage for USB VBUS comparator input	-0.5	5.25	V	
DDR_VREF	Supply voltage for the DDR SSTL and HSTL reference voltage	-0.3	1.1	V	
Steady state max voltage at all I/O pins ⁽⁸⁾		-0.5 V to I/O suppl	ly voltage + 0.3 V		
USB0_ID ⁽⁹⁾	Steady state maximum voltage for the USB ID input	-0.5	2.1	V	
USB1_ID ⁽⁶⁾⁽⁹⁾	Steady state maximum voltage for the USB ID input	-0.5	2.1	V	
Transient overshoot and undershoot specification at I/O terminal		voltage for up to	25% of corresponding I/O supply voltage for up to 30% of signal period		
Latch-up performance ⁽¹⁰⁾	Class II (105°C)	45		mA	
Storage temperature, T _{stg} ⁽¹¹⁾		-55	155	°C	

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to their associated VSS or VSSA_x.
- (3) Not available on the ZCE package. VDD_MPU is merged with VDD_CORE on the ZCE package.
- (4) This supply is sourced from an internal LDO when RTC_KALDO_ENn is low. If RTC_KALDO_ENn is high, this supply must be sourced from an external power supply.
- (5) During functional operation, this pin is a no connect.
- (6) Not available on the ZCE package.
- (7) This terminal is connected to a fail-safe I/O and does not have a dependence on any I/O supply voltage.
- (8) This parameter applies to all I/O terminals which are not fail-safe and the requirement applies to all values of I/O supply voltage. For example, if the voltage applied to a specific I/O supply is 0 volts the valid input voltage range for any I/O powered by that supply will be -0.5 to +0.3 V. Apply special attention anytime peripheral devices are not powered from the same power sources used to power the

www.ti.com.cn

Absolute Maximum Ratings (continued)

over junction temperature range (unless otherwise noted)(1)(2)

- respective I/O supply. It is important the attached peripheral never sources a voltage outside the valid input voltage range, including power supply ramp-up and ramp-down sequences.
- (9) This terminal is connected to analog circuits in the respective USB PHY. The circuit sources a known current while measuring the voltage to determine if the terminal is connected to VSSA_USB with a resistance less than 10 Ω or greater than 100 k Ω . The terminal should be connected to ground for USB host operation or open-circuit for USB peripheral operation, and should never be connected to any external voltage source.
- (10) Based on JEDEC JESD78D [IC Latch-Up Test].
- (11) For tape and reel the storage temperature range is [-10°C; +50°C] with a maximum relative humidity of 70%. TI recommends returning to ambient room temperature before usage.

Fail-safe I/O terminals are designed such they do not have dependencies on the respective I/O power supply voltage. This allows external voltage sources to be connected to these I/O terminals when the respective I/O power supplies are turned off. The USB0_VBUS and USB1_VBUS are the only fail-safe I/O terminals. All other I/O terminals are not fail-safe and the voltage applied to them should be limited to the value defined by the **steady state max. Voltage at all I/O pins** parameter in Section 5.1.

5.2 ESD Ratings

			VALUE	UNIT
V Electrostatic discharge		Human Body Model (HBM), per ANSI/ESDA/JEDEC JS001 ⁽¹⁾	±2000	V
V _{ESD}	(ESD) performance:	Charged Device Model (CDM), per JESD22-C101 (2)	±500	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



5.3 Power-On Hours (POH)

表 5-1. Reliability Data⁽¹⁾⁽²⁾⁽³⁾⁽⁴⁾

OPERATING	COMMERCIAL		INDUSTRIAL		EXTENDED		INDUSTRIAL EXTENDED	
CONDITION	JUNCTION TEMP (T _J)	LIFETIME (POH) ⁽⁵⁾						
Nitro	0°C to 90°C	100K	-40°C to 90°C	100K	-40°C to 105°C	37K	-40°C to 125°C	_
Turbo	0°C to 90°C	100K	-40°C to 90°C	100K	-40°C to 105°C	80K	-40°C to 125°C	-
OPP120	0°C to 90°C	100K	-40°C to 90°C	100K	-40°C to 105°C	100K	-40°C to 125°C	-
OPP100	0°C to 90°C	100K	-40°C to 90°C	100K	-40°C to 105°C	100K	-40°C to 125°C	35K
OPP50	0°C to 90°C	100K	-40°C to 90°C	100K	-40°C to 105°C	100K	-40°C to 125°C	95K

- (1) The power-on hours (POH) information in this table is provided solely for your convenience and does not extend or modify the warranty provided under TI's standard terms and conditions for TI semiconductor products.
- (2) To avoid significant degradation, the device power-on hours (POH) must be limited as described in this table.
- (3) Logic functions and parameter values are not assured out of the range specified in the recommended operating conditions.
- (4) The previous notations cannot be deemed a warranty or deemed to extend or modify the warranty under TI's standard terms and conditions for TI semiconductor products.
- (5) POH = Power-on hours when the device is fully functional.

5.4 Operating Performance Points (OPPs)

Device OPPs are defined in 表 5-2 through 表 5-9.

表 5-2. VDD_CORE OPPs for ZCZ Package With Device Revision Code "Blank"⁽¹⁾

VDD_CORE	VDD_CORE						
OPP Device Rev. "Blank"	MIN	NOM	MAX	DDR3, DDR3L ⁽²⁾	DDR2 ⁽²⁾	mDDR ⁽²⁾	L3 and L4
OPP100	1.056 V	1.100 V	1.144 V	400 MHz	266 MHz	200 MHz	200 and 100 MHz
OPP50	0.912 V	0.950 V	0.988 V	_	125 MHz	90 MHz	100 and 50 MHz

- (1) Frequencies in this table indicate maximum performance for a given OPP condition.
- (2) This parameter represents the maximum memory clock frequency. Because data is transferred on both edges of the clock, double-data rate (DDR), the maximum data rate is two times the maximum memory clock frequency defined in this table.

表 5-3. VDD_MPU OPPs for ZCZ Package With Device Revision Code "Blank"⁽¹⁾

VDD_MPU OPP		ADM (AO)		
Device Rev. "Blank"	MIN	NOM	MAX	ARM (A8)
Turbo	1.210 V	1.260 V	1.326 V	720 MHz
OPP120	1.152 V	1.200 V	1.248 V	600 MHz
OPP100 ⁽²⁾	1.056 V	1.100 V	1.144 V	500 MHz
OPP100 ⁽³⁾	1.056 V	1.100 V	1.144 V	275 MHz

- (1) Frequencies in this table indicate maximum performance for a given OPP condition.
- (2) Applies to all orderable AM335__ZCZ_50 (500-MHz speed grade) or higher devices.
- (3) Applies to all orderable AM335__ZCZ_27 (275-MHz speed grade) devices.



表 5-4. Valid Combinations of VDD_CORE and VDD_MPU OPPs for ZCZ Package With Device Revision Code "Blank"

VDD_CORE	VDD_MPU
OPP50	OPP100
OPP100	OPP100
OPP100	OPP120
OPP100	Turbo

表 5-5. VDD_CORE OPPs for ZCE Package With Device Revision Code "Blank"⁽¹⁾

VDD_CORE	_		VDD_MPU ⁽²⁾					
OPP Device Rev. "Blank"	MIN	NOM	MAX	ARM (A8)	DDR3, DDR3L ⁽³⁾	DDR2 ⁽³⁾	mDDR ⁽³⁾	L3 and L4
OPP100	1.056 V	1.100 V	1.144 V	500 MHz	400 MHz	266 MHz	200 MHz	200 and 100 MHz
OPP100	1.056 V	1.100 V	1.144 V	275 MHz	400 MHz	266 MHz	200 MHz	200 and 100 MHz

- (1) Frequencies in this table indicate maximum performance for a given OPP condition.
- (2) VDD_MPU is merged with VDD_CORE on the ZCE package.
- (3) This parameter represents the maximum memory clock frequency. Because data is transferred on both edges of the clock, double-data rate (DDR), the maximum data rate is two times the maximum memory clock frequency defined in this table.

表 5-6. VDD_CORE OPPs for ZCZ Package With Device Revision Code "A" or Newer⁽¹⁾

VDD_CORE	_						
OPP Rev "A" or Newer	MIN	NOM	MAX	DDR3, DDR3L ⁽²⁾	DDR2 ⁽²⁾	mDDR ⁽²⁾	L3 and L4
OPP100	1.056 V	1.100 V	1.144 V	400 MHz	266 MHz	200 MHz	200 and 100 MHz
OPP50	0.912 V	0.950 V	0.988 V	_	125 MHz	90 MHz	100 and 50 MHz

- (1) Frequencies in this table indicate maximum performance for a given OPP condition.
- (2) This parameter represents the maximum memory clock frequency. Because data is transferred on both edges of the clock, double-data rate (DDR), the maximum data rate is two times the maximum memory clock frequency defined in this table.

表 5-7. VDD_MPU OPPs for ZCZ Package With Device Revision Code "A" or Newer⁽¹⁾

VDD MPU OPP		A DA4 (A 0)		
Rev "A" or Newer	MIN	NOM	MAX	ARM (A8)
Nitro	1.272 V	1.325 V	1.378 V	1 GHz
Turbo	1.210 V	1.260 V	1.326 V	800 MHz
OPP120	1.152 V	1.200 V	1.248 V	720 MHz
OPP100 ⁽²⁾	1.056 V	1.100 V	1.144 V	600 MHz
OPP100 ⁽³⁾	1.056 V	1.100 V	1.144 V	300 MHz
OPP50	0.912 V	0.950 V	0.988 V	300 MHz

- (1) Frequencies in this table indicate maximum performance for a given OPP condition.
- (2) Applies to all orderable AM335_ZCZ_60 (600-MHz speed grade) or higher devices.
- (3) Applies to all orderable AM335__ZCZ_30 (300-MHz speed grade) devices.



表 5-8. Valid Combinations of VDD_CORE and VDD_MPU OPPs for ZCZ Package With Device Revision Code "A" or Newer

VDD_CORE	VDD_MPU
OPP50	OPP50
OPP50	OPP100
OPP100	OPP50
OPP100	OPP100
OPP100	OPP120
OPP100	Turbo
OPP100	Nitro

表 5-9. VDD_CORE OPPs for ZCE Package With Device Revision Code "A" or Newer $^{(1)}$

VDD_CORE	VDD_MPU ⁽²⁾							
OPP Rev "A" or newer	MIN	NOM	MAX	ARM (A8)	DDR3, DDR3L ⁽³⁾	DDR2 ⁽³⁾	mDDR ⁽³⁾	L3 and L4
OPP100	1.056 V	1.100 V	1.144 V	600 MHz	400 MHz	266 MHz	200 MHz	200 and 100 MHz
OPP100	1.056 V	1.100 V	1.144 V	300 MHz	400 MHz	266 MHz	200 MHz	200 and 100 MHz
OPP50	0.912 V	0.950 V	0.988 V	300 MHz	_	125 MHz	90 MHz	100 and 50 MHz

⁽¹⁾ Frequencies in this table indicate maximum performance for a given OPP condition.

⁽²⁾ VDD_MPU is merged with VDD_CORE on the ZCE package.

⁽³⁾ This parameter represents the maximum memory clock frequency. Because data is transferred on both edges of the clock, double-data rate (DDR), the maximum data rate is two times the maximum memory clock frequency defined in this table.



5.5 Recommended Operating Conditions

over junction temperature range (unless otherwise noted)

SUPPLY NAME	DESCRIPTION	MIN	NOM	MAX	UNIT
VDD_CORE ⁽¹⁾	Supply voltage range for core domain; OPP100	1.056	1.100	1.144	V
VDD_COKE	Supply voltage range for core domain; OPP50	0.912	0.950	0.988	V
	Supply voltage range for MPU domain, Nitro	1.272	1.325	1.378	
	Supply voltage range for MPU domain; Turbo	1.210	1.260	1.326	
VDD_MPU ⁽¹⁾⁽²⁾	Supply voltage range for MPU domain; OPP120	1.152	1.200	1.248	V
	Supply voltage range for MPU domain; OPP100	1.056	1.100	1.144	
	Supply voltage range for MPU domain; OPP50	0.912	0.950	0.988	
CAP_VDD_RTC ⁽³⁾	Supply voltage range for RTC domain input	0.900	1.100	1.250	V
VDDS_RTC	Supply voltage range for RTC domain	1.710	1.800	1.890	V
	Supply voltage range for DDR I/O domain (DDR2)	1.710	1.800	1.890	
VDDS_DDR	Supply voltage range for DDR I/O domain (DDR3)	1.425	1.500	1.575	V
	Supply voltage range for DDR I/O domain (DDR3L)	1.283	1.350	1.418	
VDDS ⁽⁴⁾	Supply voltage range for all dual-voltage I/O domains	1.710	1.800	1.890	V
VDDS_SRAM_CORE_BG	Supply voltage range for Core SRAM LDOs, analog	1.710	1.800	1.890	V
VDDS_SRAM_MPU_BB	Supply voltage range for MPU SRAM LDOs, analog	1.710	1.800	1.890	V
VDDS_PLL_DDR ⁽⁵⁾	Supply voltage range for DPLL DDR, analog	1.710	1.800	1.890	V
VDDS_PLL_CORE_LCD ⁽⁵⁾	Supply voltage range for DPLL CORE and LCD, analog	1.710	1.800	1.890	V
VDDS_PLL_MPU ⁽⁵⁾	Supply voltage range for DPLL MPU, analog	1.710	1.800	1.890	V
VDDS_OSC	Supply voltage range for system oscillator I/Os, analog	1.710	1.800	1.890	V
VDDA1P8V_USB0 ⁽⁵⁾	Supply voltage range for USBPHY and PER DPLL, analog, 1.8 V	1.710	1.800	1.890	V
VDDA1P8V_USB1 ⁽⁶⁾	Supply voltage range for USB PHY, analog, 1.8 V	1.710	1.800	1.890	V
VDDA3P3V_USB0	Supply voltage range for USB PHY, analog, 3.3 V	3.135	3.300	3.465	V
VDDA3P3V_USB1 ⁽⁶⁾	Supply voltage range for USB PHY, analog, 3.3 V	3.135	3.300	3.465	V
VDDA_ADC	Supply voltage range for ADC, analog	1.710	1.800	1.890	V
VDDSHV1	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	V
VDDSHV2 ⁽⁶⁾	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	V



Recommended Operating Conditions (continued)

over junction temperature range (unless otherwise noted)

SUPPLY NAME	DESCRIPTION	MIN	NOM	MAX	UNIT
VDDSHV3 ⁽⁶⁾	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	V
VDDSHV4	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	V
VDDSHV5	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	٧
VDDSHV6	Supply voltage range for dual- voltage I/O domain (1.8-V operation)	1.710	1.800	1.890	V
VDDSHV1	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	٧
VDDSHV2 ⁽⁶⁾	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	V
VDDSHV3 ⁽⁶⁾	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	V
VDDSHV4	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	٧
VDDSHV5	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	V
VDDSHV6	Supply voltage range for dual- voltage I/O domain (3.3-V operation)	3.135	3.300	3.465	٧
DDR_VREF	Voltage range for DDR SSTL and HSTL reference input (DDR2, DDR3, DDR3L)	0.49 × VDDS_DDR	0.50 × VDDS_DDR	0.51 × VDDS_DDR	٧
USB0_VBUS	Voltage range for USB VBUS comparator input	0.000	5.000	5.250	٧
USB1_VBUS ⁽⁶⁾	Voltage range for USB VBUS comparator input	0.000	5.000	5.250	V
USB0_ID	Voltage range for the USB ID input		(7)		V
USB1_ID ⁽⁶⁾	Voltage range for the USB ID input		(7)		٧
	Commercial temperature	0		90	
Operating temperature range, T _J	Industrial temperature	-40		90	°C
90, 1,1	Extended temperature	-40		105	

- (1) The supply voltage defined by OPP100 should be applied to this power domain before the device is released from reset.
- (2) Not available on the ZCE package. VDD_MPU is merged with VDD_CORE on the ZCE package.
- (3) This supply is sourced from an internal LDO when RTC_KALDO_ENn is low. If RTC_KALDO_ENn is high, this supply must be sourced from an external power supply.
- (4) VDDS should be supplied irrespective of 1.8- or 3.3-V mode of operation of the dual-voltage I/Os.
- (5) For more details on power supply requirements, see 节 6.1.4.
- (6) Not available on the ZCE package.
- (7) This terminal is connected to analog circuits in the respective USB PHY. The circuit sources a known current while measuring the voltage to determine if the terminal is connected to VSSA_USB with a resistance less than 10 Ω or greater than 100 k Ω . The terminal should be connected to ground for USB host operation or open-circuit for USB peripheral operation, and should never be connected to any external voltage source.



5.6 Power Consumption Summary

表 5-10 summarizes the power consumption at the AM335x power terminals.

表 5-10. Maximum Current Ratings at AM335x Power Terminals⁽¹⁾

SUPPLY NAME	DESCRIPTION		MAX	UNIT
VDD 00DE(2)	Maximum current rating for the core domain; OPP100	400	1	
VDD_CORE ⁽²⁾	Maximum current rating for the core domain; OPP50		250	mA
	Maximum current rating for the MPU domain; Nitro	at 1 GHz	1000	
	Manianum aumant nation for the MDU descrip. Turks	at 800 MHz	800	
	Maximum current rating for the MPU domain; Turbo	at 720 MHz	720	
	Manifester and retire to fair the MPU descript ORD400	at 720 MHz	720	
	Maximum current rating for the MPU domain; OPP120	at 600 MHz	600	
VDD_MPU ⁽²⁾		at 600 MHz	600	mA
	Manifestory assured ration for the MPU demains ORD400	at 500 MHz	500	
	Maximum current rating for the MPU domain; OPP100	at 300 MHz	380	
		at 275 MHz	350	
	Manifestory and a street for the MPU description OPPE	at 300 MHz	330	
	Maximum current rating for the MPU domain; OPP50	at 275 MHz	300	
CAP_VDD_RTC ⁽³⁾	Maximum current rating for RTC domain input and LDO ou	itput	2	mA
VDDS_RTC	Maximum current rating for the RTC domain	5	mA	
VDDS_DDR	Maximum current rating for DDR I/O domain	250	mA	
VDDS	Maximum current rating for all dual-voltage I/O domains	50	mA	
VDDS_SRAM_CORE_BG	Maximum current rating for core SRAM LDOs	10	mA	
VDDS_SRAM_MPU_BB	Maximum current rating for MPU SRAM LDOs		10	mA
VDDS_PLL_DDR	Maximum current rating for the DPLL DDR		10	mA
VDDS_PLL_CORE_LCD	Maximum current rating for the DPLL Core and LCD		20	mA
VDDS_PLL_MPU	Maximum current rating for the DPLL MPU		10	mA
VDDS_OSC	Maximum current rating for the system oscillator I/Os		5	mA
VDDA1P8V_USB0	Maximum current rating for USBPHY 1.8 V		25	mA
VDDA1P8V_USB1 ⁽⁴⁾	Maximum current rating for USBPHY 1.8 V		25	mA
VDDA3P3V_USB0	Maximum current rating for USBPHY 3.3 V		40	mA
VDDA3P3V_USB1 ⁽⁴⁾	Maximum current rating for USBPHY 3.3 V		40	mA
VDDA_ADC	Maximum current rating for ADC		10	mA
VDDSHV1 ⁽⁵⁾	Maximum current rating for dual-voltage I/O domain		50	mA
VDDSHV2 ⁽⁴⁾	Maximum current rating for dual-voltage I/O domain		50	mA
VDDSHV3 ⁽⁴⁾	Maximum current rating for dual-voltage I/O domain		50	mA
VDDSHV4	Maximum current rating for dual-voltage I/O domain		50	mA
VDDSHV5	Maximum current rating for dual-voltage I/O domain		50	mA
VDDSHV6	Maximum current rating for dual-voltage I/O domain		100	mA

⁽¹⁾ Current ratings specified in this table are worst-case estimates. Actual application power supply estimates could be lower. For more information, see *AM335x Power Consumption Summary*.

⁽²⁾ VDD_MPU is merged with VDD_CORE and is not available separately on the ZCE package. The maximum current rating for VDD_CORE on the ZCE package is the sum of VDD_CORE and VDD_MPU shown in this table.

⁽³⁾ This supply is sourced from an internal LDO when RTC_KALDO_ENn is low. If RTC_KALDO_ENn is high, this supply must be sourced from an external power supply.

⁽⁴⁾ Not available on the ZCE package.

⁽⁵⁾ VDDSHV1 and VDDSHV2 are merged in the ZCE package. The maximum current rating for VDDSHV1 on the ZCE package is the sum of VDDSHV1 and VDDSHV2 shown in this table.



表 5-11 summarizes the power consumption of the AM335x low-power modes.

表 5-11. AM335x Low-Power Modes Power Consumption Summary

POWER MODES	APPLICATION STATE	POWER DOMAINS, CLOCKS, AND VOLTAGE SUPPLY STATES	NOM	MAX	UNIT
Standby	DDR memory is in self-refresh and contents are preserved. Wake up from any GPIO. Cortex-A8 context/register contents are lost and must be saved before entering standby. On exit, context must be restored from DDR. For wakeup, boot ROM executes and branches to system resume.	Power supplies: All power supplies are ON. VDD_MPU = 0.95 V (nom) VDD_CORE = 0.95 V (nom) Clocks: Main Oscillator (OSC0) = ON All DPLLs are in bypass. Power domains: PD_PER = ON PD_MPU = OFF PD_GFX = OFF PD_WKUP = ON DDR is in self-refresh.	16.5	22.0	mW
Deepsleep1	On-chip peripheral registers are preserved. Cortex-A8 context/registers are lost, so the application must save them to the L3 OCMC RAM or DDR before entering DeepSleep. DDR is in self-refresh. For wakeup, boot ROM executes and branches to system resume.	Power supplies: All power supplies are ON. VDD_MPU = 0.95 V (nom) VDD_CORE = 0.95 V (nom) Clocks: Main Oscillator (OSC0) = OFF All DPLLs are in bypass. Power domains: PD_PER = ON PD_MPU = OFF PD_GFX = OFF PD_WKUP = ON DDR is in self-refresh.	6.0	10.0	mW
Deepsleep0	PD_PER peripheral and Cortex-A8/MPU register information will be lost. On-chip peripheral register (context) information of PD-PER domain must be saved by application to SDRAM before entering this mode. DDR is in self-refresh. For wakeup, boot ROM executes and branches to peripheral context restore followed by system resume.	Power supplies: All power supplies are ON. VDD_MPU = 0.95 V (nom) VDD_CORE = 0.95 V (nom) Clocks: Main Oscillator (OSC0) = OFF All DPLLs are in bypass. Power domains: PD_PER = OFF PD_MPU = OFF PD_GFX = OFF PD_WKUP = ON DDR is in self-refresh.	3.0	4.3	mW



5.7 DC Electrical Characteristics

over recommended ranges of supply voltage and operating temperature (unless otherwise noted)(1)

over reco	mmended ranges of supply voltage and opera	ating temperature	(unless otherwise n	oted) ⁽¹⁾		
	PARAMETER		MIN	NOM	MAX	UNIT
2,DDR_A3 R_D1,DDR	ETn,DDR_CSn0,DDR_CKE,DDR_CK,DDR_CKn,DDR_CA,DDR_CA,DDR_CA,DDR_A8,DDR_A9,DR_A9,DR_D3,DDR_D3,DDR_D5,DDR_D6,DDR_D7,DDR_D1,DDR_D7,DDR_D8,DDR_DQS0,DDR_DQS0,DDR_DQS0,DDR_DQS0,DDR_DQS0	DDR_A10,DDR_A11, _D8,DDR_D9,DDR_D	DDR_A12,DDR_A13,DDR 10,DDR_D11,DDR_D12,D	_A14,DDR_A15,DDR	ODT,DDR	D0,DD
V _{IH}	High-level input voltage		0.65 × VDDS_DDR			V
V _{IL}	Low-level input voltage			V	0.35 x	V
V _{HYS}	Hysteresis voltage at an input		0.07		0.25	V
V _{OH}	High level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 8 mA	VDDS_DDR - 0.4			V
V _{OL}	Low level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 8 mA			0.4	٧
	Input leakage current, Receiver disabled, pullup or pulle	down inhibited			10	
l _i	Input leakage current, Receiver disabled, pullup enable	d	-240		-80	μΑ
	Input leakage current, Receiver disabled, pulldown ena	bled	80		240	
l _{oz}	Total leakage current through the terminal connection of combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.				10	μА
2,DDR_A3 R_D1,DDR	ETn,DDR_CSn0,DDR_CKE,DDR_CK,DDR_CKn,DDR_CA ,DDR_A4,DDR_A5,DDR_A6,DDR_A7,DDR_A8,DDR_A9, t_D2,DDR_D3,DDR_D4,DDR_D5,DDR_D6,DDR_D7,DDR_ tM1,DDR_DQS0,DDR_DQSn0,DDR_DQS1,DDR_DQSn1	DDR_A10,DDR_A11, _D8,DDR_D9,DDR_D	DDR_A12,DDR_A13,DDR 10,DDR_D11,DDR_D12,D	_A14,DDR_A15,DDR	ODT,DDR	D0,DD
V _{IH}	High-level input voltage		DDR_VREF + 0.125			V
V _{HYS}	Hysteresis voltage at an input			N/A		٧
V _{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 8 mA	VDDS_DDR - 0.4			V
V _{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 8 mA			0.4	V
	Input leakage current, Receiver disabled, pullup or pulle	down inhibited			10	
I _I	Input leakage current, Receiver disabled, pullup enable	d	-240		-80	μΑ
	Input leakage current, Receiver disabled, pulldown ena	bled	80		240	
l _{OZ}	Total leakage current through the terminal connection of combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.				10	μΑ
2,DDR_A3, R_D1,DDR	ETn,DDR_CSn0,DDR_CKE,DDR_CK,DDR_CKn,DDR_CA,DDR_CA,DDR_A5,DDR_A6,DDR_A7,DDR_A8,DDR_A9,t _D2,DDR_D3,DDR_D4,DDR_D5,DDR_D6,DDR_D7,DDR_M1,DDR_DQS0,DDR_DQS01,DDR_DQS01	DDR_A10,DDR_A11, _D8,DDR_D9,DDR_D	DDR_A12,DDR_A13,DDR 10,DDR_D11,DDR_D12,D - HSTL Mode)	_A14,DDR_A15,DDR	_ODT,DDR_	_D0,DD
V	High level input voltage	VDDS_DDR = 1.5 V	DDR_VREF + 0.1			V
V _{IH}	High-level input voltage	VDDS_DDR = 1.35 V	DDR_VREF + 0.09			V
V	Low level input velters	VDDS_DDR = 1.5 V		DD	R_VREF - 0.1	V
V_{IL}	Low-level input voltage	VDDS_DDR = 1.35 V		DD	R_VREF - 0.09	V
V _{HYS}	Hysteresis voltage at an input			N/A		V
V _{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 8 mA	VDDS_DDR - 0.4			V
V _{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 8 mA			0.4	V
	Input leakage current, Receiver disabled, pullup or pulle	down inhibited			10	· <u></u>
l _l	Input leakage current, Receiver disabled, pullup enable	d	-240		-80	μΑ
	Input leakage current, Receiver disabled, pulldown ena	bled	80		240	
l _{oz}	Total leakage current through the terminal connection of combination that may include a pullup or pulldown. The				10	μA



DC Electrical Characteristics (continued)

over recommended ranges of supply voltage and operating temperature (unless otherwise noted)⁽¹⁾

01011000	of interfaced ranges of supply voltage and opera	ang temperature	(driiece curiormee in			
	PARAMETER		MIN	NOM	MAX	UNIT
	_PWM0_OUT,UART0_CTSn,UART0_RTSn,UART0_RXD,\ A_EVENT_INTR0,XDMA_EVENT_INTR1,WARMRSTn,EXT)A,I2C0_
V_{IH}	High-level input voltage		0.65 × VDDSHV6			V
V_{IL}	Low-level input voltage			0.35	× VDDSHV6	V
V _{HYS}	Hysteresis voltage at an input		0.18		0.305	V
V_{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 4 mA	VDDSHV6 - 0.45			V
V _{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled $I_{\rm OL} = 4~{\rm mA}$				0.45	٧
	Input leakage current, Receiver disabled, pullup or pulld	own inhibited			8	
I _I	Input leakage current, Receiver disabled, pullup enabled	t	-161	-100	-52	μΑ
	Input leakage current, Receiver disabled, pulldown enab	oled	52	100	170	l
l _{oz}	Total leakage current through the terminal connection of combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.				8	μΑ
	N_PWM0_OUT,UART0_CTSn,UART0_RTSn,UART0_RXD, A_EVENT_INTR0,XDMA_EVENT_INTR1,WARMRSTn,EXT)A,I2C0_
V_{IH}	High-level input voltage		2			V
V_{IL}	Low-level input voltage				0.8	V
V _{HYS}	Hysteresis voltage at an input		0.265		0.44	V
V _{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 4 mA	VDDSHV6 - 0.45			V
V _{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 4 mA			0.45	٧
	Input leakage current, Receiver disabled, pullup or pulld			18		
I	Input leakage current, Receiver disabled, pullup enabled	-243	-100	-19	μΑ	
	Input leakage current, Receiver disabled, pulldown enab	Input leakage current, Receiver disabled, pulldown enabled			210	1
l _{oz}	Total leakage current through the terminal connection of combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.				18	μA
TCK (VDD	PSHV6 = 1.8 V)					
V _{IH}	High-level input voltage		1.45			V
V _{IL}	Low-level input voltage				0.46	V
V _{HYS}	Hysteresis voltage at an input		0.4			V
	Input leakage current, Receiver disabled, pullup or pulld	own inhibited			8	
I _I	Input leakage current, Receiver disabled, pullup enabled	d	-161	-100	-52	μA
	Input leakage current, Receiver disabled, pulldown enab	oled	52	100	170	l
TCK (VDD	PSHV6 = 3.3 V)					
V _{IH}	High-level input voltage		2.15			V
V _{IL}	Low-level input voltage				0.46	V
V _{HYS}	Hysteresis voltage at an input		0.4			V
	Input leakage current, Receiver disabled, pullup or pulld	own inhibited			18	
I _I	Input leakage current, Receiver disabled, pullup enabled	d	-243	-100	-19	μA
	Input leakage current, Receiver disabled, pulldown enab	oled	51	110	210	
PWRONR	STn (VDDSHV6 = 1.8 or 3.3 V) ⁽²⁾		1			
V _{IH}	High-level input voltage		1.35			V
V _{IL}	Low-level input voltage				0.5	V
V _{HYS}	Hysteresis voltage at an input		0.07			V
I _I	Input leakage current	V _I = 1.8 V			0.1	μA
	ONDET-	V _I = 3.3 V			2	
RTC_PWR	KONKOIN		0.0=			
V_{IH}	High-level input voltage		0.65 × VDDS_RTC			V



DC Electrical Characteristics (continued)

over recommended ranges of supply voltage and operating temperature (unless otherwise noted)(1)

OVET TECK	PARAMETER	ting temperature	MIN	NOM MAX	UNIT
			- IIIIV	0.35 ×	
V _{IL}	Low-level input voltage			VDDS_RTC	V
V _{HYS}	Hysteresis voltage at an input		0.065		V
I _I	Input leakage current		-1	1	μA
PMIC_PO	WER_EN				
V_{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 6 mA	VDDS_RTC - 0.45		V
V_{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 6 mA		0.45	V
	Input leakage current, Receiver disabled, pullup or pullo	lown inhibited	-1	1	
I_1	Input leakage current, Receiver disabled, pullup enable	d	-200	-40	μΑ
	Input leakage current, Receiver disabled, pulldown enal	oled	40	200	
l _{OZ}	Total leakage current through the terminal connection o combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.		-1	1	μA
EXT_WAP	KEUP				
V_{IH}	High-level input voltage		0.65 × VDDS_RTC		V
V _{IL}	Low-level input voltage			0.35 × VDDS_RTC	V
V _{HYS}	Hysteresis voltage at an input		0.15		V
	Input leakage current, Receiver disabled, pullup or pullo	lown inhibited	-1	1	
II	Input leakage current, Receiver disabled, pullup enable	d	-200	-40	μA
	Input leakage current, Receiver disabled, pulldown enal	oled	40	200	
XTALIN (OSC0)				1
V_{IH}	High-level input voltage		0.65 × VDDS_OSC		V
V _{IL}	Low-level input voltage			0.35 × VDDS_OSC	V
RTC_XTA	LIN (OSC1)				II
V _{IH}	High-level input voltage		0.65 × VDDS_RTC		V
V _{IL}	Low-level input voltage			0.35 x VDDS_RTC	V
All other I	LVCMOS pins (VDDSHVx = 1.8 V; x = 1 to 6)				1
V _{IH}	High-level input voltage		0.65 × VDDSHVx		V
V _{IL}	Low-level input voltage			0.35 × VDDSHVx	V
V _{HYS}	Hysteresis voltage at an input		0.18	0.305	V
V _{OH}	High-level output voltage, driver enabled, pullup or pulldown disabled	I _{OH} = 6 mA	VDDSHVx - 0.45		V
V _{OL}	Low-level output voltage, driver enabled, pullup or pulldown disabled	I _{OL} = 6 mA		0.45	V
	Input leakage current, Receiver disabled, pullup or pullo	lown inhibited		8	
I _I	Input leakage current, Receiver disabled, pullup enable		-161	-100 -52	-
	Input leakage current, Receiver disabled, pulldown enal		52	100 170	1 '
I _{OZ}	Total leakage current through the terminal connection o combination that may include a pullup or pulldown. The disabled and the pullup or pulldown is inhibited.	f a driver-receiver e driver output is		8	μA
All other	LVCMOS pins (VDDSHVx = 3.3 V; x = 1 to 6)				
V _{IH}	High-level input voltage		2		V
V _{IL}	Low-level input voltage			0.8	V
	Hysteresis voltage at an input		0.265	0.6	V
V _{HYS}	High-level output voltage, driver enabled, pullup or	I _{OH} = 6 mA	0.265 VDDSHVx – 0.45	0.44	V
V _{OL}	pulldown disabled Low-level output voltage, driver enabled, pullup or	I _{OL} = 6 mA		0.45	V
- UL	pulldown disabled	10L - 3 11/1		0.40	•



DC Electrical Characteristics (continued)

over recommended ranges of supply voltage and operating temperature (unless otherwise noted)(1)

	PARAMETER	MIN	NOM	MAX	UNIT
	Input leakage current, Receiver disabled, pullup or pulldown inhibited			18	
I	Input leakage current, Receiver disabled, pullup enabled	-243	-100	-19	μΑ
	Input leakage current, Receiver disabled, pulldown enabled	51	110	210	
I _{OZ}	Total leakage current through the terminal connection of a driver-receiver combination that may include a pullup or pulldown. The driver output is disabled and the pullup or pulldown is inhibited.			18	μΑ

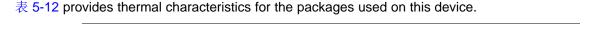
⁽¹⁾ The interfaces or signals described in this table correspond to the interfaces or signals available in multiplexing mode 0. All interfaces or signals multiplexed on the terminals described in this table have the same DC electrical characteristics.

⁽²⁾ The input voltage thresholds for this input are not a function of VDDSHV6.



5.8 Thermal Resistance Characteristics for ZCE and ZCZ Packages

Failure to maintain a junction temperature within the range specified in Section 5.5 reduces operating lifetime, reliability, and performance—and may cause irreversible damage to the system. Therefore, the product design cycle should include thermal analysis to verify the maximum operating junction temperature of the device. It is important this thermal analysis is performed using specific system use cases and conditions. TI provides an application report to aid users in overcoming some of the existing challenges of producing a good thermal design. For more information, see *AM335x Thermal Considerations*.



注

表 5-12 provides simulation data and may not represent actual use-case values.

表 5-12. Thermal Resistance Characteristics (PBGA Package) [ZCE and ZCZ]

		ZCE	E (°C/W) ⁽¹⁾	ZCZ (°C/W) ⁽¹⁾	AIR FLOW (m/s) ⁽³⁾
R _{⊕JC}	Junction-to-case		10.3	10.2	N/A
R _{⊕JB}	Junction-to-board		11.6	12.1	N/A
			24.7	24.2	0
D	lunation to force oil		20.5	20.1	1.0
$R_{\Theta JA}$	Junction-to-free air		19.7	19.3	2.0
			19.2	10.2 12.1 24.2 20.1 19.3 18.8 0.3 0.6 0.7 0.8 12.7 12.3 12.3	3.0
			0.4	0.3	0.0
	Localita de manha de la		0.6	0.6	1.0
ΨЈТ	Junction-to-package top		0.7	0.7	2.0
			0.9	0.8	3.0
			11.9	12.7	0.0
	harathar ta bassal		11.7	12.3	1.0
ΨЈВ	Junction-to-board		11.7	12.3	2.0
			11.6	12.2	3.0

⁽¹⁾ These values are based on a JEDEC-defined 2S2P system (with the exception of the theta JC [R_{⊕JC}] value, which is based on a JEDEC-defined 1S0P system) and will change based on environment as well as application. For more information, see these EIA/JEDEC standards:

- JESD51-2, Integrated Circuits Thermal Test Method Environmental Conditions Natural Convection (Still Air)
- JESD51-3, Low Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages
- JESD51-7, High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages
- JESD51-9, Test Boards for Area Array Surface Mount Package Thermal Measurements

Power dissipation of 2 W and an ambient temperature of 70°C is assumed.

- °C/W = degrees Celsius per watt.
- (3) m/s = meters per second.

5.9 External Capacitors

To improve module performance, decoupling capacitors are required to suppress the switching noise generated by high frequency and to stabilize the supply voltage. A decoupling capacitor is most effective when it is close to the device, because this minimizes the inductance of the circuit board wiring and interconnects.

5.9.1 Voltage Decoupling Capacitors

表 5-13 summarizes the Core voltage decoupling characteristics.

5.9.1.1 Core Voltage Decoupling Capacitors

To improve module performance, decoupling capacitors are required to suppress high-frequency switching noise and to stabilize the supply voltage. A decoupling capacitor is most effective when located close to the AM335x device, because this minimizes the inductance of the circuit board wiring and interconnects.

表 5-13. Core Voltage Decoupling Characteristics

PARAMETER	TYP	UNIT
C _{VDD_CORE} ⁽¹⁾	10.08	μF
C _{VDD_MPU} ⁽²⁾⁽³⁾	10.05	μF

- (1) The typical value corresponds to one capacitor of 10 μF and eight capacitors of 10 nF.
- (2) Not available on the ZCE package. VDD_MPU is merged with VDD_CORE on the ZCE package.
- (3) The typical value corresponds to one capacitor of 10 μF and five capacitors of 10 nF.

5.9.1.2 I/O and Analog Voltage Decoupling Capacitors

表 5-14 summarizes the power-supply decoupling capacitor recommendations.

表 5-14. Power-Supply Decoupling Capacitor Characteristics

PARAMETER	TYP	UNIT
C _{VDDA_ADC}	10	nF
C _{VDDA1P8V_USB0}	10	nF
C _{CVDDA3P3V_USB0}	10	nF
C _{VDDA1P8V_USB1} ⁽¹⁾	10	nF
C _{VDDA3P3V_USB1} ⁽¹⁾	10	nF
C _{VDDS} ⁽²⁾	10.04	μF
C _{VDDS_DDR}	(3)	
C _{VDDS_OSC}	10	nF
C _{VDDS_PLL_DDR}	10	nF
C _{VDDS_PLL_CORE_LCD}	10	nF
C _{VDDS_SRAM_CORE_BG} ⁽⁴⁾	10.01	μF
C _{VDDS_SRAM_MPU_BB} ⁽⁵⁾	10.01	μF
C _{VDDS_PLL_MPU}	10	nF
C _{VDDS_RTC}	10	nF
C _{VDDSHV1} ⁽⁶⁾	10.02	μF
C _{VDDSHV2} ⁽¹⁾⁽⁶⁾	10.02	μF
C _{VDDSHV3} ⁽¹⁾⁽⁶⁾	10.02	μF
C _{VDDSHV4} ⁽⁶⁾	10.02	μF
C _{VDDSHV5} ⁽⁶⁾	10.02	μF
C _{VDDSHV6} ⁽⁷⁾	10.06	μF



- (1) Not available on the ZCE package.
- (2) Typical values consist of one capacitor of 10 μF and four capacitors of 10 nF.
- (3) For more details on decoupling capacitor requirements for the mDDR(LPDDR), DDR2, DDR3, DDR3L memory interface, see 节 7.7.2.1.2.6 and 节 7.7.2.1.2.7 when using mDDR(LPDDR) memory devices, 节 7.7.2.2.2.6 and 节 7.7.2.2.2.7 when using DDR2 memory devices, or 节 7.7.2.3.3.6 and 节 7.7.2.3.3.7 when using DDR3 or DDR3L memory devices.
- (4) VDDS_SRAM_CORE_BG supply powers an internal LDO for SRAM supplies. Inrush currents could cause voltage drop on the VDDS_SRAM_CORE_BG supplies when the SRAM LDO is enabled after powering up VDDS_SRAM_CORE_BG terminals. A 10 μF is recommended to be placed close to the terminal and routed with widest traces possible to minimize the voltage drop on VDDS_SRAM_CORE_BG terminals.
- (5) VDDS_SRAM_MPU_BB supply powers an internal LDO for SRAM supplies. Inrush currents could cause voltage drop on the VDDS_SRAM_MPU_BB supplies when the SRAM LDO is enabled after powering up VDDS_SRAM_MPU_BB terminals. A 10 μF is recommended to be placed close to the terminal and routed with widest traces possible to minimize the voltage drop on VDDS_SRAM_MPU_BB terminals.
- (6) Typical values consist of one capacitor of 10 μF and two capacitors of 10 nF.
- (7) Typical values consist of one capacitor of 10 μF and six capacitors of 10 nF.

5.9.2 Output Capacitors

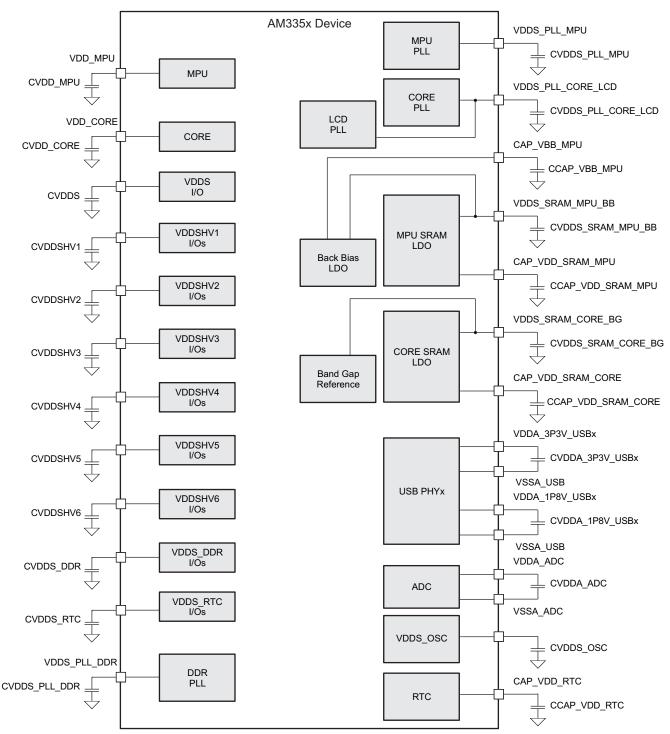
Internal low dropout output (LDO) regulators require external capacitors to stabilize their outputs. These capacitors should be placed as close as possible to the respective terminals of the AM335x device. 表 5-15 summarizes the LDO output capacitor recommendations.

表 5-15. Output Capacitor Characteristics

PARAMETER	TYP	UNIT
C _{CAP_VDD_SRAM_CORE} ⁽¹⁾	1	μF
C _{CAP_VDD_RTC} ⁽¹⁾⁽²⁾	1	μF
C _{CAP_VDD_SRAM_MPU} ⁽¹⁾	1	μF
C _{CAP_VBB_MPU} ⁽¹⁾	1	μF

- (1) LDO regulator outputs should not be used as a power source for any external components.
- (2) The CAP_VDD_RTC terminal operates as an input to the RTC core voltage domain when the RTC_KLDO_ENn terminal is high.

§ 5-1 shows an example of the external capacitors.



- Decoupling capacitors must be placed as closed as possible to the power terminal. Choose the ground closest to the power pin for each decoupling capacitor. In case of interconnecting powers, first insert the decoupling capacitor and then interconnect the powers.
- The decoupling capacitor value depends on the characteristics of the board.

图 5-1. External Capacitors



5.10 Touch Screen Controller and Analog-to-Digital Subsystem Electrical Parameters

The touch screen controller (TSC) and analog-to-digital converter (ADC) subsystem (TSC_ADC) is an 8-channel general-purpose ADC with optional support for interleaving TSC conversions for 4-wire, 5-wire, or 8-wire resistive panels. The TSC_ADC subsystem can be configured for use in one of the following applications:

- 8 general-purpose ADC channels
- 4-wire TSC with 4 general-purpose ADC channels
- 5-wire TSC with 3 general-purpose ADC channels
- 8-wire TSC.

表 5-16 summarizes the TSC_ADC subsystem electrical parameters.

表 5-16. TSC_ADC Electrical Parameters

PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
Analog Input					
VREFP ⁽¹⁾		(0.5 × VDDA_ADC) + 0.25		VDDA_ADC	V
VREFN ⁽¹⁾		0		(0.5 × VDDA_ADC) - 0.25	٧
VREFP + VREFN ⁽¹⁾			VDDA_ADC		V
Full scale input years	Internal voltage reference	0		VDDA_ADC	V
Full-scale input range	External voltage reference	VREFN		VREFP	V
Differential nonlinearity (DNL)	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP – VREFN = 1.8 V	-1	0.5	1	LSB
Integral nonlinearity (INL)	Source impedance = 50 Ω Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP – VREFN = 1.8 V	-2	±1	2	- CD
	Source impedance = $1 \text{ k}\Omega$ Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP - VREFN = 1.8 V		±1		LSB
Gain error	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP – VREFN = 1.8 V		±2		LSB
Offset error	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP – VREFN = 1.8 V		±2		LSB
Input sampling capacitance			5.5		pF
Signal-to-noise ratio (SNR)	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP - VREFN = 1.8 V Input signal: 30-kHz sine wave at -0.5-dB full scale		70		dB
Total harmonic distortion (THD)	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP - VREFN = 1.8 V Input signal: 30-kHz sine wave at -0.5-dB full scale		75		dB



表 5-16. TSC_ADC Electrical Parameters (continued)

PARAMETER	TEST CONDITIONS	MIN NOM MAX	UNIT
Spurious free dynamic range	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP - VREFN = 1.8 V Input signal: 30-kHz sine wave at -0.5-dB full scale	80	
Signal-to-noise plus distortion	Internal voltage reference: VDDA_ADC = 1.8 V External voltage reference: VREFP - VREFN = 1.8 V Input signal: 30-kHz sine wave at -0.5-dB full scale	69	dB
VREFP and VREFN input im	pedance	20	kΩ
Input impedance of AIN[7:0] ⁽²⁾	f = Input frequency	$[1 / ((65.97 \times 10^{-12}) \times f)]$	Ω
Sampling Dynamics			
Conversion time		15	ADC clock cycles
Acquisition time		2	ADC clock cycles
Sampling rate	ADC clock = 3 MHz	200	kSPS
Channel-to-channel isolation		100	dB
Touch Screen Switch Drive	ers		
Pullup and pulldown switch ON resistance (Ron)		2	Ω
Pullup and pulldown switch current leakage Ileak	Source impedance = 500 Ω	0.5	
Drive current		25	mA
Touch screen resistance		6	kΩ
Pen touch detect		2	kΩ

⁽¹⁾ VREFP and VREFN must be tied to ground if the internal voltage reference is used.

⁽²⁾ This parameter is valid when the respective AIN terminal is configured to operate as a general-purpose ADC input.

www.ti.com.cn

6 Power and Clocking

6.1 Power Supplies

6.1.1 Power Supply Slew Rate Requirement

To maintain the safe operating range of the internal ESD protection devices, TI recommends limiting the maximum slew rate for powering on the supplies to be less than 1.0E +5 V/s. For instance, as shown in 86-1, TI recommends a value greater than 18 µs for the supply ramp slew for a 1.8-V supply.

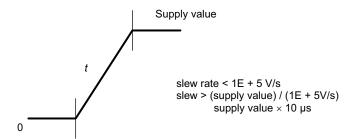
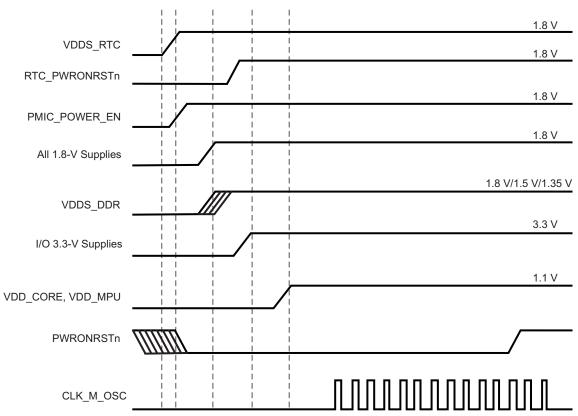


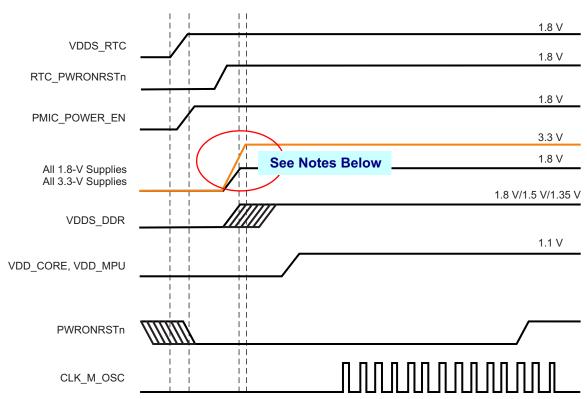
图 6-1. Power Supply Slew and Slew Rate



- A. RTC_PWRONRSTn should be asserted for at least 1 ms to provide enough time for the internal RTC LDO output to reach a valid level before RTC reset is released.
- B. When using the ZCZ package option, VDD_MPU and VDD_CORE power inputs may be powered from the same source if the application only uses operating performance points (OPPs) that define a common power supply voltage for VDD_MPU and VDD_CORE. The ZCE package option has the VDD_MPU domain merged with the VDD_CORE domain.
- C. If a USB port is not used, the respective VDDA1P8V_USB terminal may be connected to any 1.8-V power supply and the respective VDDA3P3V_USB terminal may be connected to any 3.3-V power supply. If the system does not have a 3.3-V power supply, the VDDA3P3V_USB terminal may be connected to ground.
- D. If the system uses mDDR or DDR2 memory devices, VDDS_DDR can be ramped simultaneously with the other 1.8-V I/O power supplies.
- E. VDDS_RTC can be ramped independent of other power supplies if PMIC_POWER_EN functionality is not required. If VDDS_RTC is ramped after VDD_CORE, there might be a small amount of additional leakage current on VDD_CORE. The power sequence shown provides the lowest leakage option.
- F. To configure VDDSHVx [1-6] as 1.8 V, power up the respective VDDSHVx [1-6] to 1.8 V following the recommended sequence. To configure VDDSHVx [1-6] as 3.3 V, power up the respective VDDSHVx [1-6] to 3.3 V following the recommended sequence.

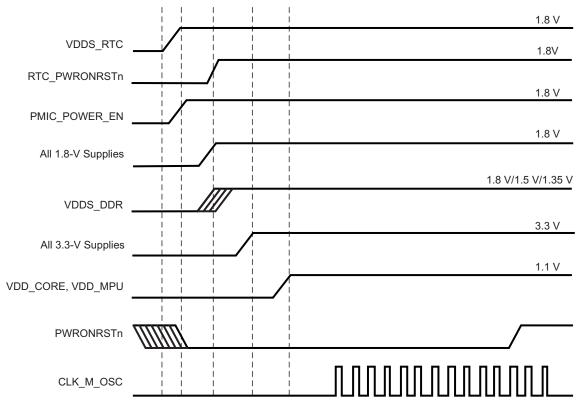
图 6-2. Preferred Power-Supply Sequencing With Dual-Voltage I/Os Configured as 3.3 V

RUMENTS



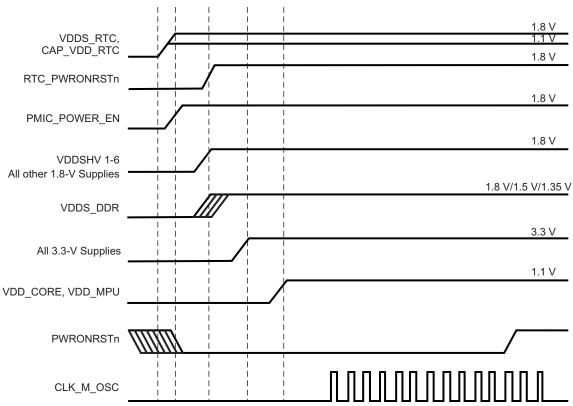
- RTC_PWRONRSTn should be asserted for at least 1 ms to provide enough time for the internal RTC LDO output to reach a valid level before RTC reset is released.
- The 3.3-V I/O power supplies may be ramped simultaneously with the 1.8-V I/O power supplies if the voltage sourced by any 3.3-V power supplies does not exceed the voltage sourced by any 1.8-V power supply by more than 2 V. Serious reliability issues may occur if the system power supply design allows any 3.3-V I/O power supplies to exceed any 1.8-V I/O power supplies by more than 2 V.
- C. When using the ZCZ package option, VDD_MPU and VDD_CORE power inputs may be powered from the same source if the application only uses operating performance points (OPPs) that define a common power supply voltage for VDD_MPU and VDD_CORE. The ZCE package option has the VDD_MPU domain merged with the VDD_CORE
- If a USB port is not used, the respective VDDA1P8V USB terminal may be connected to any 1.8-V power supply and the respective VDDA3P3V_USB terminal may be connected to any 3.3-V power supply. If the system does not have a 3.3-V power supply, the VDDA3P3V_USB terminal may be connected to ground.
- If the system uses mDDR or DDR2 memory devices, VDDS DDR can be ramped simultaneously with the other 1.8-V I/O power supplies.
- VDDS_RTC can be ramped independent of other power supplies if PMIC_POWER_EN functionality is not required. If VDDS_RTC is ramped after VDD_CORE, there might be a small amount of additional leakage current on VDD_CORE. The power sequence shown provides the lowest leakage option.
- To configure VDDSHVx [1-6] as 1.8 V, power up the respective VDDSHVx [1-6] to 1.8 V following the recommended sequence. To configure VDDSHVx [1-6] as 3.3 V, power up the respective VDDSHVx [1-6] to 3.3 V following the recommended sequence.

图 6-3. Alternate Power-Supply Sequencing With Dual-Voltage I/Os Configured as 3.3 V



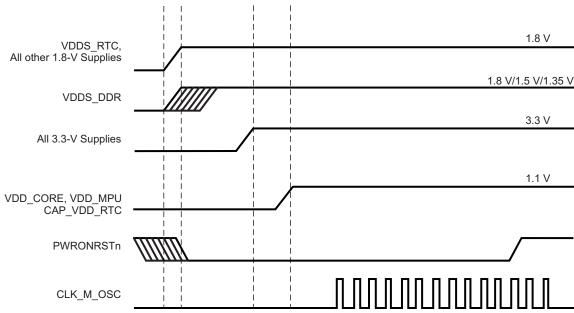
- A. RTC_PWRONRSTn should be asserted for at least 1 ms to provide enough time for the internal RTC LDO output to reach a valid level before RTC reset is released.
- B. When using the ZCZ package option, VDD_MPU and VDD_CORE power inputs may be powered from the same source if the application only uses operating performance points (OPPs) that define a common power supply voltage for VDD_MPU and VDD_CORE. The ZCE package option has the VDD_MPU domain merged with the VDD_CORE domain.
- C. If a USB port is not used, the respective VDDA1P8V_USB terminal may be connected to any 1.8-V power supply and the respective VDDA3P3V_USB terminal may be connected to any 3.3-V power supply. If the system does not have a 3.3-V power supply, the VDDA3P3V_USB terminal may be connected to ground.
- D. If the system uses mDDR or DDR2 memory devices, VDDS_DDR can be ramped simultaneously with the other 1.8-V I/O power supplies.
- E. VDDS_RTC can be ramped independent of other power supplies if PMIC_POWER_EN functionality is not required. If VDDS_RTC is ramped after VDD_CORE, there might be a small amount of additional leakage current on VDD_CORE. The power sequence shown provides the lowest leakage option.
- F. To configure VDDSHVx [1-6] as 1.8 V, power up the respective VDDSHVx [1-6] to 1.8 V following the recommended sequence. To configure VDDSHVx [1-6] as 3.3 V, power up the respective VDDSHVx [1-6] to 3.3 V following the recommended sequence.

图 6-4. Power-Supply Sequencing With Dual-Voltage I/Os Configured as 1.8 V



- A. RTC_PWRONRSTn should be asserted for at least 1 ms to provide enough time for the internal RTC LDO output to reach a valid level before RTC reset is released.
- B. The CAP_VDD_RTC terminal operates as an input to the RTC core voltage domain when the internal RTC LDO is disabled by connecting the RTC_KALDO_ENn terminal to VDDS_RTC. If the internal RTC LDO is disabled, CAP_VDD_RTC should be sourced from an external 1.1-V power supply.
- C. When using the ZCZ package option, VDD_MPU and VDD_CORE power inputs may be powered from the same source if the application only uses operating performance points (OPPs) that define a common power supply voltage for VDD_MPU and VDD_CORE. The ZCE package option has the VDD_MPU domain merged with the VDD_CORE domain.
- D. If a USB port is not used, the respective VDDA1P8V_USB terminal may be connected to any 1.8-V power supply and the respective VDDA3P3V_USB terminal may be connected to any 3.3-V power supply. If the system does not have a 3.3-V power supply, the VDDA3P3V_USB terminal may be connected to ground.
- E. If the system uses mDDR or DDR2 memory devices, VDDS_DDR can be ramped simultaneously with the other 1.8-V I/O power supplies.
- F. VDDS_RTC should be ramped at the same time or before CAP_VDD_RTC, but these power inputs can be ramped independent of other power supplies if PMIC_POWER_EN functionality is not required. If CAP_VDD_RTC is ramped after VDD_CORE, there might be a small amount of additional leakage current on VDD_CORE. The power sequence shown provides the lowest leakage option.
- G. To configure VDDSHVx [1-6] as 1.8 V, power up the respective VDDSHVx [1-6] to 1.8 V following the recommended sequence. To configure VDDSHVx [1-6] as 3.3 V, power up the respective VDDSHVx [1-6] to 3.3 V following the recommended sequence.

图 6-5. Power-Supply Sequencing With Internal RTC LDO Disabled



- A. CAP_VDD_RTC terminal operates as an input to the RTC core voltage domain when the internal RTC LDO is disabled by connecting the RTC_KALDO_ENn terminal to VDDS_RTC. If the internal RTC LDO is disabled, CAP_VDD_RTC should be sourced from an external 1.1-V power supply. The PMIC_POWER_EN output cannot be used when the RTC is disabled.
- B. When using the ZCZ package option, VDD_MPU and VDD_CORE power inputs may be powered from the same source if the application only uses operating performance points (OPPs) that define a common power supply voltage for VDD_MPU and VDD_CORE. The ZCE package option has the VDD_MPU domain merged with the VDD_CORE domain.
- C. If a USB port is not used, the respective VDDA1P8V_USB terminal may be connected to any 1.8-V power supply and the respective VDDA3P3V_USB terminal may be connected to any 3.3-V power supply. If the system does not have a 3.3-V power supply, the VDDA3P3V_USB terminal may be connected to ground.
- D. If the system uses mDDR or DDR2 memory devices, VDDS_DDR can be ramped simultaneously with the other 1.8-V I/O power supplies.
- E. VDDS_RTC should be ramped at the same time or before CAP_VDD_RTC, but these power inputs can be ramped independent of other power supplies if PMIC_POWER_EN functionality is not required. If CAP_VDD_RTC is ramped after VDD_CORE, there might be a small amount of additional leakage current on VDD_CORE. The power sequence shown provides the lowest leakage option.
- F. To configure VDDSHVx [1-6] as 1.8 V, power up the respective VDDSHVx [1-6] to 1.8 V following the recommended sequence. To configure VDDSHVx [1-6] as 3.3 V, power up the respective VDDSHVx [1-6] to 3.3 V following the recommended sequence.

图 6-6. Power-Supply Sequencing With RTC Feature Disabled

6.1.2 Power-Down Sequencing

PWRONRSTn input terminal should be taken low, which stops all internal clocks before power supplies are turned off. All other external clocks to the device should be shut off.

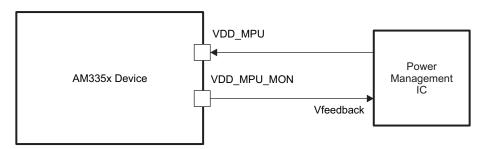
The preferred way to sequence power down is to have all the power supplies ramped down sequentially in the exact reverse order of the power-up sequencing. In other words, the power supply that has been ramped up first should be the last one that should be ramped down. This ensures there would be no spurious current paths during the power-down sequence. The VDDS power supply must ramp down after all 3.3-V VDDSHVx [1-6] power supplies.

If it is desired to ramp down VDDS and VDDSHVx [1-6] simultaneously, it should always be ensured that the difference between VDDS and VDDSHVx [1-6] during the entire power-down sequence is <2 V. Any violation of this could cause reliability risks for the device. TI recommends maintaining VDDS ≥1.5V as all the other supplies fully ramp down to minimize in-rush currents.

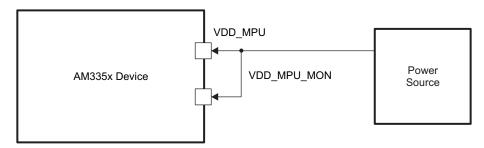


If none of the VDDSHVx [1-6] power supplies are configured as 3.3 V, the VDDS power supply may ramp down along with the VDDSHVx [1-6] supplies or after all the VDDSHVx [1-6] supplies have ramped down. TI recommends maintaining VDDS ≥1.5V as all the other supplies fully ramp down to minimize in-rush currents.

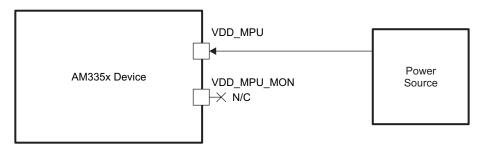
6.1.3 VDD MPU MON Connections



Connection for VDD MPU MON if voltage monitoring is used



Preferred connection for VDD_MPU_MON if voltage monitoring is NOT used



Optional connection for VDD_MPU_MON if voltage monitoring is NOT used

图 6-7. VDD_MPU_MON Connectivity



6.1.4 Digital Phase-Locked Loop Power Supply Requirements

The digital phase-locked loop (DPLL) provides all interface clocks and functional clocks to the processor of the AM335x device. The AM335x device integrates five different DPLLs—Core DPLL, Per DPLL, LCD DPLL, DDR DPLL, MPU DPLL.

图 6-8 shows the power supply connectivity implemented in the AM335x device. 表 6-1 provides the power supply requirements for the DPLL.

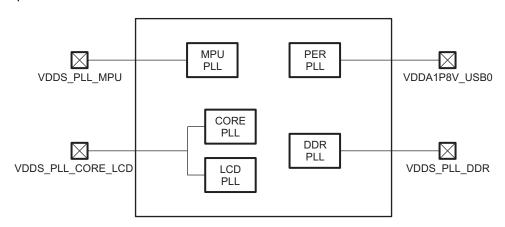


图 6-8. DPLL Power Supply Connectivity

表 6-1. DPLL Power Supply Requirements

SUPPLY NAME	DESCRIPTION	MIN	NOM	MAX	UNIT
\/DD 4 4 D0\/ 1 10 D0	Supply voltage range for USBPHY and PER DPLL, Analog, 1.8 V	1.71	1.8	1.89	V
VDDA1P8V_USB0	Max peak-to-peak supply noise			50	mV (p-p)
VDDC DII MDII	Supply voltage range for DPLL MPU, analog	1.71	1.8	1.89	V
VDDS_PLL_MPU	Max peak-to-peak supply noise			50	mV (p-p)
VDDS PLL CORE LCD	Supply voltage range for DPLL CORE and LCD, analog	1.71	1.8	1.89	V
VDD5_PLL_CORE_LCD	Max peak-to-peak supply noise	;		50	mV (p-p)
VDDC DII DDD	Supply voltage range for DPLL DDR, analog	1.71	1.8	1.89	V
VDDS_PLL_DDR	Max peak-to-peak supply noise			50	mV (p-p)



6.2 Clock Specifications

TRUMENTS

6.2.1 Input Clock Specifications

The AM335x device has two clock inputs. Each clock input passes through an internal oscillator which can be connected to an external crystal circuit (oscillator mode) or external LVCMOS square-wave digital clock source (bypass mode). The oscillators automatically operate in bypass mode when their input is connected to an external LVCMOS square-wave digital clock source. The oscillator associated with a specific clock input must be enabled when the clock input is being used in either oscillator mode or bypass mode.

The OSC1 oscillator provides a 32.768-kHz reference clock to the real-time clock (RTC) and is connected to the RTC_XTALIN and RTC_XTALOUT terminals. This clock source is referred to as the 32K oscillator (CLK 32K RTC) in the AM335x and AMIC110 Sitara Processors Technical Reference Manual. OSC1 is disabled by default after power is applied. This clock input is optional and may not be required if the RTC is configured to receive a clock from the internal 32k RC oscillator (CLK_RC32K) or peripheral PLL (CLK 32KHZ) which receives a reference clock from the OSC0 input.

The OSC0 oscillator provides a 19.2-MHz, 24-MHz, 25-MHz, or 26-MHz reference clock which is used to clock all non-RTC functions and is connected to the XTALIN and XTALOUT terminals. This clock source is referred to as the master oscillator (CLK M OSC) in the AM335x and AMIC110 Sitara Processors Technical Reference Manual. OSC0 is enabled by default after power is applied.

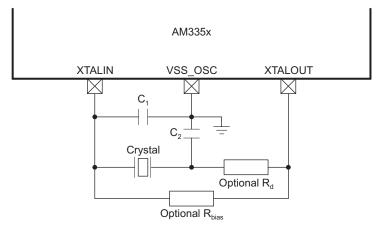
For more information related to recommended circuit topologies and crystal oscillator circuit requirements for these clock inputs, see † 6.2.2.

6.2.2 Input Clock Requirements

6.2.2.1 OSC0 Internal Oscillator Clock Source

🛚 6-9 shows the recommended crystal circuit. TI recommends that preproduction printed-circuit board (PCB) designs include the two optional resistors R_{bias} and R_d in case they are required for proper oscillator operation when combined with production crystal circuit components. In most cases, Rbias is not required and R_d is a 0- Ω resistor. These resistors may be removed from production PCB designs after evaluating oscillator performance with production crystal circuit components installed on preproduction PCBs.

The XTALIN terminal has a 15- to 40-k Ω internal pulldown resistor which is enabled when OSC0 is disabled. This internal resistor prevents the XTALIN terminal from floating to an invalid logic level which may increase leakage current through the oscillator input buffer.



Copyright © 2016, Texas Instruments Incorporated

- A. Oscillator components (Crystal, C₁, C₂, optional R_{bias} and R_d) must be located close to the AM335x package. Parasitic capacitance to the VSS_OSC and respective crystal circuit component grounds should be connected directly to the nearest PCB digital ground (VSS).
- B. C₁ and C₂ represent the total capacitance of the respective PCB trace, load capacitor, and other components (excluding the crystal) connected to each crystal terminal. The value of capacitors C₁ and C₂ should be selected to provide the total load capacitance, C_L, specified by the crystal manufacturer. The total load capacitance is C_L = [(C₁ × C₂) / (C₁ + C₂)] + C_{shunt}, where C_{shunt} is the crystal shunt capacitance (C₀) specified by the crystal manufacturer plus any mutual capacitance (C_{pkg} + C_{PCB}) seen across the AM335x XTALIN and XTALOUT signals. For recommended values of crystal circuit components, see 表 6-2.

图 6-9. OSC0 Crystal Circuit Schematic

表 6-2. OSC0 Crystal Circuit Requirements

PARAMETER			MIN	TYP MAX	UNIT
f_{xtal}	Crystal parallel resonance frequency	Fundamental mode oscillation only		19.2, 24, 25, or 26	MHz
	Crystal frequency stability and tolerance ⁽¹⁾		-50	50	ppm
0	Connecitores	C _{shunt} ≤ 5 pF	12	24	. "_
C _{C1}	C ₁ capacitance	C _{shunt} > 5 pF	18	24	pF
C	Connecitores	C _{shunt} ≤ 5 pF	12	24	pF
C _{C2}	C ₂ capacitance	C _{shunt} > 5 pF	18	24	. pr
C _{shunt}	Shunt capacitance	Shunt capacitance		7	pF
ESR	Crystal effective series resistance	$f_{ m xtal}$ = 19.2 MHz, oscillator has nominal negative resistance of 272 Ω and worst-case negative resistance of 163 Ω		54.4	
		$f_{ m xtal}$ = 24 MHz, oscillator has nominal negative resistance of 240 Ω and worst-case negative resistance of 144 Ω		48.0	
		$f_{ m xtal}$ = 25 MHz, oscillator has nominal negative resistance of 233 Ω and worst-case negative resistance of 140 Ω		46.6	Ω
		$f_{ m xtal}$ = 26 MHz, oscillator has nominal negative resistance of 227 Ω and worst-case negative resistance of 137 Ω		45.3	

(1) Initial accuracy, temperature drift, and aging effects should be combined when evaluating a reference clock for this requirement.



表 6-3. OSC0 Crystal Circuit Characteristics

NAME	DESCRIPTION		MIN	TYP	MAX	UNIT
C	Shunt capacitance of ZCE package					
C_{pkg}	package	ZCZ package		pF		
P _{xtal}	typical crystal power diss	ESR, f_{xtal} , and C_{L} should be used to yield a pation value. Using the maximum values and C_{L} parameters yields a maximum power	P	0.01 $P_{\text{xtal}} = 0.5 \text{ ESR } (2 \pi f_{\text{xtal}} + C_{\text{L}})$		
t _{sX}	Start-up time		1.5			ms

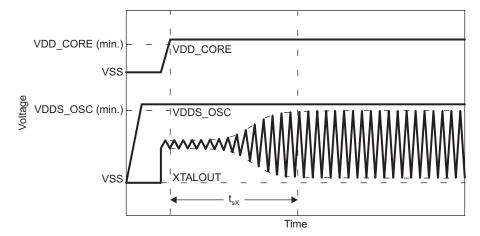


图 6-10. OSC0 Start-Up Time



6.2.2.2 OSC0 LVCMOS Digital Clock Source

wave digital clock source. The LVCMOS clock source is connected to the XTALIN terminal. The ground for the LVCMOS clock source and VSS_OSC should be connected directly to the nearest PCB digital ground (VSS). In this mode of operation, the XTALOUT terminal should not be used to source any external components. The PCB design should provide a mechanism to disconnect the XTALOUT terminal from any external components or signal traces that may couple noise into OSC0 via the XTALOUT terminal.

The XTALIN terminal has a 15- to 40-k Ω internal pulldown resistor which is enabled when OSC0 is disabled. This internal resistor prevents the XTALIN terminal from floating to an invalid logic level which may increase leakage current through the oscillator input buffer.

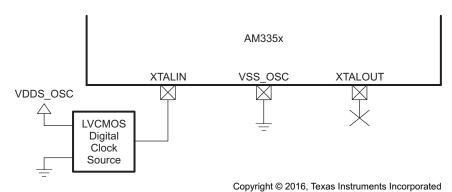


图 6-11. OSC0 LVCMOS Circuit Schematic

表 6-4. OSC0 LVCMOS Reference Clock Requirements

NAME	DESCRIPTION	MIN	TYP	MAX	UNIT
$f_{(XTALIN)}$	Frequency, LVCMOS reference clock		19.2, 24, 25, or 26		MHz
3 (XIXLIIV)	Frequency, LVCMOS reference clock stability and tolerance (1)	-50		50	ppm
t _{dc(XTALIN)}	Duty cycle, LVCMOS reference clock period	45%		55%	
t _{jpp(XTALIN)}	Jitter peak-to-peak, LVCMOS reference clock period	-1%		1%	
t _{R(XTALIN)}	Time, LVCMOS reference clock rise			5	ns
t _{F(XTALIN)}	Time, LVCMOS reference clock fall			5	ns

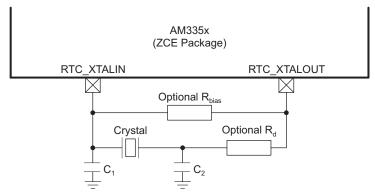
⁽¹⁾ Initial accuracy, temperature drift, and aging effects should be combined when evaluating a reference clock for this requirement.

6.2.2.3 OSC1 Internal Oscillator Clock Source

图 6-12 shows the recommended crystal circuit for OSC1 of the ZCE package and 图 6-13 shows the recommended crystal circuit for OSC1 of the ZCZ package. TI recommends that preproduction PCB designs include the two optional resistors R_{bias} and R_d in case they are required for proper oscillator operation when combined with production crystal circuit components. In most cases, Rbias is not required and R_d is a 0- Ω resistor. These resistors may be removed from production PCB designs after evaluating oscillator performance with production crystal circuit components installed on preproduction PCBs.

The RTC_XTALIN terminal has a 10- to 40-k Ω internal pullup resistor which is enabled when OSC1 is disabled. This internal resistor prevents the RTC_XTALIN terminal from floating to an invalid logic level which may increase leakage current through the oscillator input buffer.

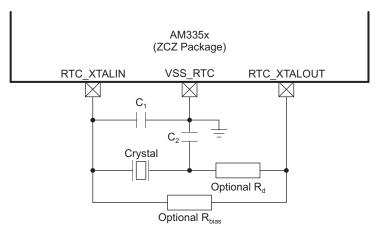




Copyright © 2016, Texas Instruments Incorporated

- Oscillator components (Crystal, C1, C2, optional Rbias and Rd) must be located close to the AM335x package. Parasitic capacitance to the PCB ground and other signals should be minimized to reduce noise coupled into the oscillator. VSS_RTC and respective crystal circuit component grounds should be connected directly to the nearest PCB digital ground (VSS)
- C1 and C2 represent the total capacitance of the respective PCB trace, load capacitor, and other components (excluding the crystal) connected to each crystal terminal. The value of capacitors C₁ and C₂ should be selected to provide the total load capacitance, C_L , specified by the crystal manufacturer. The total load capacitance is $C_L = [(C_1 \times C_2)]$ $(C_2) / (C_1 + C_2)$] + (C_{shunt}) , where (C_{shunt}) is the crystal shunt capacitance (C_0) specified by the crystal manufacturer plus any mutual capacitance (C_{pkg} + C_{PCB}) seen across the AM335x RTC_XTALIN and RTC_XTALOUT signals. For recommended values of crystal circuit components, see $\frac{1}{8}$ 6-5.

图 6-12. OSC1 (ZCE Package) Crystal Circuit Schematic



Copyright © 2016, Texas Instruments Incorporated

- Oscillator components (Crystal, C1, C2, optional Rbias and Rd) must be located close to the AM335x package. Parasitic capacitance to the PCB ground and other signals should be minimized to reduce noise coupled into the oscillator. VSS_RTC and respective crystal circuit component grounds should be connected directly to the nearest PCB digital ground (VSS).
- C₁ and C₂ represent the total capacitance of the respective PCB trace, load capacitor, and other components (excluding the crystal) connected to each crystal terminal. The value of capacitors C_1 and C_2 should be selected to provide the total load capacitance, C_L , specified by the crystal manufacturer. The total load capacitance is $C_L = [(C_1 \times C_2) / (C_1 + C_2)] + C_{shunt}$, where C_{shunt} is the crystal shunt capacitance (C_0) specified by the crystal manufacturer plus any mutual capacitance ($C_{pkg} + C_{PCB}$) seen across the AM335x RTC_XTALIN and RTC_XTALOUT signals. For recommended values of crystal circuit components, see $\frac{1}{8}$ 6-5.

图 6-13. OSC1 (ZCZ Package) Crystal Circuit Schematic

111



表 6-5. OSC1 Crystal Circuit Requirements

NAME	DESCRIPTION		MIN	TYP	MAX	UNIT
f_{xtal}	Crystal parallel resonance frequency	Fundamental mode oscillation only		32.768		kHz
	Crystal frequency stability	Maximum RTC error = 10.512 minutes per year	-20.0		20.0	ppm
	and tolerance ⁽¹⁾	Maximum RTC error = 26.28 minutes per year	-50.0		50.0	ppm
C _{C1}	C ₁ capacitance		12.0		24.0	pF
C _{C2}	C ₂ capacitance		12.0		24.0	pF
C _{shunt}	Shunt capacitance				1.5	рF
ESR	Crystal effective series resistance	$f_{ m xtal}$ = 32.768 kHz, oscillator has nominal negative resistance of 725 k Ω and worst-case negative resistance of 250 k Ω			80	kΩ

(1) Initial accuracy, temperature drift, and aging effects should be combined when evaluating a reference clock for this requirement.

表 6-6. OSC1 Crystal Circuit Characteristics

NAME	DESCRIPTION		MIN	TYP	MAX	UNIT
C .	Shunt capacitance of	ZCE package		0.17		pF
C_{pkg}	package ZCZ package		0.01			pF
P _{xtal}	The actual values of the ESR, f_{xtal} , and C_{L} should be used to yield a typical crystal power dissipation value. Using the maximum values specified for ESR, f_{xtal} , and C_{L} parameters yields a maximum power dissipation value.					
t _{sX}	Start-up time 2			s		

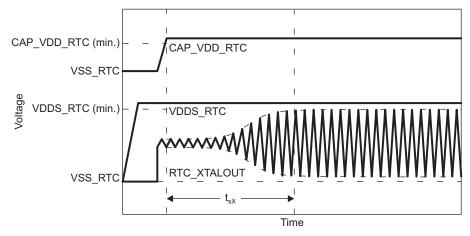


图 6-14. OSC1 Start-up Time



6.2.2.4 OSC1 LVCMOS Digital Clock Source

Is 6-15 shows the recommended oscillator connections when OSC1 of the ZCE package is connected to an LVCMOS square-wave digital clock source and Is 6-16 shows the recommended oscillator connections when OSC1 of the ZCZ package is connected to an LVCMOS square-wave digital clock source. The LVCMOS clock source is connected to the RTC_XTALIN terminal. The ground for the LVCMOS clock source and VSS_RTC of the ZCZ package should be connected directly to the nearest PCB digital ground (VSS). In this mode of operation, the RTC_XTALOUT terminal should not be used to source any external components. The PCB design should provide a mechanism to disconnect the RTC_XTALOUT terminal from any external components or signal traces that may couple noise into OSC1 through the RTC_XTALOUT terminal.

The RTC_XTALIN terminal has a 10- to 40-k Ω internal pullup resistor which is enabled when OSC1 is disabled. This internal resistor prevents the RTC_XTALIN terminal from floating to an invalid logic level which may increase leakage current through the oscillator input buffer.

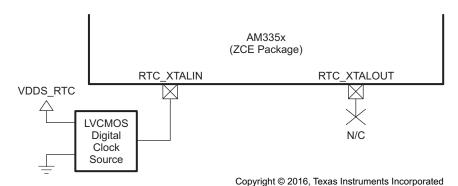
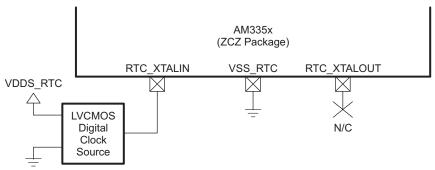


图 6-15. OSC1 (ZCE Package) LVCMOS Circuit Schematic



Copyright © 2016, Texas Instruments Incorporated

图 6-16. OSC1 (ZCZ Package) LVCMOS Circuit Schematic

表 6-7. OSC1 LVCMOS Reference Clock Requirements

NAME	DESCRIPTION		MIN	TYP MAX	UNIT
	Frequency, LVCMOS reference clock			32.768	kHz
$f_{(RTC_XTALIN)}$	Frequency, LVCMOS reference clock stability and tolerance (1)	Maximum RTC error = 10.512 minutes/year	-20	20	ppm
, _ ,	stability and tolerance ⁽¹⁾ Maximum RTC error = 26.28 minutes/year		-50	50	ppm
t _{dc(RTC_XTALIN)}	Duty cycle, LVCMOS reference clock period		45%	55%	
t _{jpp(RTC_XTALIN)}	Funnant to and I WOMOO reference deal market		-1%	1%	
t _{R(RTC_XTALIN)}	(RTC_XTALIN) Time, LVCMOS reference clock rise			5	ns
t _{F(RTC_XTALIN)}	Time, LVCMOS reference clock fall			5	ns

⁽¹⁾ Initial accuracy, temperature drift, and aging effects should be combined when evaluating a reference clock for this requirement.



6.2.2.5 OSC1 Not Used

 $\[\]$ 6-17 shows the recommended oscillator connections when OSC1 of the ZCE package is not used and $\[\]$ 6-18 shows the recommended oscillator connections when OSC1 of the ZCZ package is not used. An internal 10-k $\[\]$ pullup on the RTC_XTALIN terminal is turned on when OSC1 is disabled to prevent this input from floating to an invalid logic level which may increase leakage current through the oscillator input buffer. OSC1 is disabled by default after power is applied. Therefore, both RTC_XTALIN and RTC_XTALOUT terminals should be a no connect (NC) when OSC1 is not used.

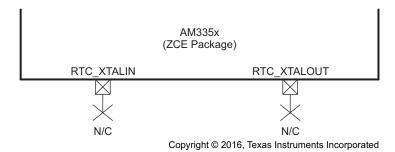


图 6-17. OSC1 (ZCE Package) Not Used Schematic

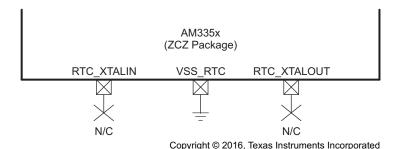


图 6-18. OSC1 (ZCZ Package) Not Used Schematic

www.ti.com.cn

6.2.3 Output Clock Specifications

The AM335x device has two clock output signals. The CLKOUT1 signal is always a replica of the OSC0 input clock which is referred to as the master oscillator (CLK_M_OSC) in the AM335x and AMIC110 Sitara Processors Technical Reference Manual. The CLKOUT2 signal can be configured to output the OSC1 input clock, which is referred to as the 32K oscillator (CLK_32K_RTC) in the AM335x and AMIC110 Sitara Processors Technical Reference Manual, or four other internal clocks. For more information related to configuring these clock output signals, see the CLKOUT Signals section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.

6.2.4 Output Clock Characteristics

注

The AM335x CLKOUT1 and CLKOUT2 clock outputs should not be used as a synchronous clock for any of the peripheral interfaces because they were not timing closed to any other signals. These clock outputs also were not designed to source any time critical external circuits that require a low jitter reference clock. The jitter performance of these outputs is unpredictable due to complex combinations of many system variables. For example, CLKOUT2 may be sourced from several PLLs with each PLL supporting many configurations that yield different jitter performance. There are also other unpredictable contributors to jitter performance such as application specific noise or crosstalk into the clock circuits. Therefore, there are no plans to specify jitter performance for these outputs.

6.2.4.1 CLKOUT1

The CLKOUT1 signal can be output on the XDMA_EVENT_INTR0 terminal. This terminal connects to one of seven internal signals via configurable multiplexers. The XDMA_EVENT_INTR0 multiplexer must be configured for Mode 3 to connect the CLKOUT1 signal to the XDMA_EVENT_INTR0 terminal.

The default reset configuration of the XDMA_EVENT_INTR0 multiplexer is selected by the logic level applied to the LCD_DATA5 terminal on the rising edge of PWRONRSTn. The XDMA_EVENT_INTR0 multiplexer is configured to Mode 7 if the LCD_DATA5 terminal is low on the rising edge of PWRONRSTn or Mode 3 if the LCD_DATA5 terminal is high on the rising edge of PWRONRSTn. This allows the CLKOUT1 signal to be output on the XDMA_EVENT_INTR0 terminal without software intervention. In this mode, the output is held low while PWRONRSTn is active and begins to toggle after PWRONRSTn is released.

6.2.4.2 CLKOUT2

The CLKOUT2 signal can be output on the XDMA_EVENT_INTR1 terminal. This terminal connects to one of seven internal signals via configurable multiplexers. The XDMA_EVENT_INTR1 multiplexer must be configured for Mode 3 to connect the CLKOUT2 signal to the XDMA_EVENT_INTR1 terminal.

The default reset configuration of the XDMA_EVENT_INTR1 multiplexer is always Mode 7. Software must configure the XDMA_EVENT_INTR1 multiplexer to Mode 3 for the CLKOUT2 signal to be output on the XDMA_EVENT_INTR1 terminal.

7 Peripheral Information and Timings

The AM335x device contains many peripheral interfaces. In order to reduce package size and lower overall system cost while maintaining maximum functionality, many of the AM335x terminals can multiplex up to eight signal functions. Although there are many combinations of pin multiplexing that are possible, only a certain number of sets, called I/O Sets, are valid due to timing limitations. These valid I/O Sets were carefully chosen to provide many possible application scenarios for the user.

Texas Instruments has developed a Windows-based application called Pin Mux Utility that helps a system designer select the appropriate pin-multiplexing configuration for their AM335x-based product design. The Pin Mux Utility provides a way to select valid I/O Sets of specific peripheral interfaces to ensure the pin-multiplexing configuration selected for a design only uses valid I/O Sets supported by the AM335x device.

7.1 Parameter Information

The data provided in the following Timing Requirements and Switching Characteristics tables assumes the device is operating within the Recommended Operating Conditions defined in # 5, unless otherwise noted.

7.1.1 Timing Parameters and Board Routing Analysis

The timing parameter values specified in this data manual do *not* include delays by board routings. As a good board design practice, such delays must *always* be taken into account. Timing values may be adjusted by increasing or decreasing such delays. TI recommends using the available I/O buffer information specification (IBIS) models to analyze the timing characteristics correctly. If needed, external logic hardware such as buffers may be used to compensate any timing differences.

The timing parameter values specified in this data manual assume the SLEWCTRL bit in each pad control register is configured for fast mode (0b).

For the mDDR(LPDDR), DDR2, DDR3, DDR3L memory interface, it is *not* necessary to use the IBIS models to analyze timing characteristics. TI provides a PCB routing rules solution that describes the routing rules to ensure the mDDR(LPDDR), DDR2, DDR3, DDR3L memory interface timings are met.

7.2 Recommended Clock and Control Signal Transition Behavior

All clocks and control signals *must* transition between V_{IH} and V_{IL} (or between V_{IL} and V_{IH}) in a monotonic manner.

7.3 OPP50 Support

Some peripherals and features have limited support when the device is operating in OPP50. A complete list of these limitations follows.

Not supported when operating in OPP50:

- CPSW
- DDR3
- DEBUGSS-Trace
- GPMC Asynchronous Mode
- LCDC LIDD Mode
- MDIO
- PRU-ICSS MII

Reduced performance when operating in OPP50:

- DDR2
- DEBUGSS-JTAG
- GPMC Synchronous Mode
- LCDC Raster Mode
- LPDDR
- McASP
- McSPI
- MMCSD



7.4 Controller Area Network (CAN)

For more information, see the Controller Area Network (CAN) section of the *AM335x and AMIC110 Sitara Processors Technical Reference Manual*.

7.4.1 DCAN Electrical Data and Timing

表 7-1. Timing Requirements for DCANx Receive

(see 图 7-1)

NO.			MIN	MAX	UNIT
	$f_{baud(baud)}$	Maximum programmable baud rate		1	Mbps
1	t _{w(RX)}	Pulse duration, receive data bit	H – 2 ⁽¹⁾	H + 2 ⁽¹⁾	ns

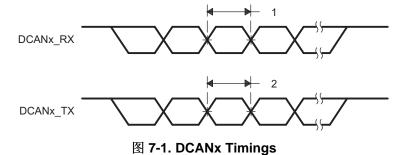
⁽¹⁾ H = Period of baud rate, 1 / programmed baud rate

表 7-2. Switching Characteristics for DCANx Transmit

(see 图 7-1)

NO.		PARAMETER		MAX	UNIT
	$f_{baud(baud)}$	Maximum programmable baud rate		1	Mbps
2	t _{w(TX)}	Pulse duration, transmit data bit	H – 2 ⁽¹⁾	H + 2 ⁽¹⁾	ns

(1) H = Period of baud rate, 1 / programmed baud rate



版权 © 2011–2016, Texas Instruments Incorporated

7.5 DMTimer

7.5.1 DMTimer Electrical Data and Timing

表 7-3. Timing Requirements for DMTimer [1-7]

(see 图 7-2)

NO.			MIN	MAX	UNIT
1	t _{c(TCLKIN)}	Cycle time, TCLKIN	4P + 1 ⁽¹⁾		ns

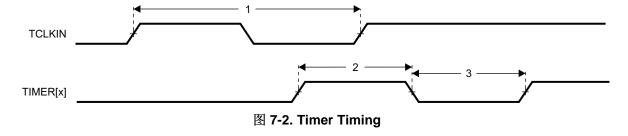
(1) P = Period of PICLKOCP (interface clock).

表 7-4. Switching Characteristics for DMTimer [4-7]

(see 图 7-2)

NO.		PARAMETER	MIN MAX	UNIT
2	t _{w(TIMERxH)}	Pulse duration, high	4P - 3 ⁽¹⁾	ns
3	t _{w(TIMERxL)}	Pulse duration, low	4P - 3 ⁽¹⁾	ns

(1) P = Period of PICLKTIMER (functional clock).





7.6 Ethernet Media Access Controller (EMAC) and Switch

7.6.1 EMAC and Switch Electrical Data and Timing

The EMAC and Switch implemented in the AM335x device supports GMII mode, but the AM335x design does not pin out 9 of the 24 GMII signals. This was done to reduce the total number of package terminals. Therefore, the AM335x device does not support GMII mode. MII mode is supported with the remaining GMII signals.

The AM335x and AMIC110 Sitara Processors Technical Reference Manual and this document may reference internal signal names when discussing peripheral input and output signals because many of the AM335x package terminals can be multiplexed to one of several peripheral signals. For example, the AM335x terminal names for port 1 of the EMAC and switch have been changed from GMII to indicate their Mode 0 function, but the internal signal is named GMII. However, documents that describe the Ethernet switch reference these signals by their internal signal name. For a cross-reference of internal signal names to terminal names, see Table 4-2.

Operation of the EMAC and switch is not supported for OPP50.

表 7-5. EMAC and Switch Timing Conditions

		•				
	PARAMETER		MIN	TYP	MAX	UNIT
Input Con	ditions				·	
t _R	Input signal rise time		1 ⁽¹⁾		5 ⁽¹⁾	ns
t _F	Input signal fall time		1 ⁽¹⁾		5 ⁽¹⁾	ns
Output Co	ondition					
C _{LOAD}	Output load capacitance		3		30	pF

⁽¹⁾ Except when specified otherwise.

7.6.1.1 EMAC/Switch MDIO Electrical Data and Timing

表 7-6. Timing Requirements for MDIO_DATA

(see 图 7-3)

NO.			MIN	TYP MAX	UNIT
1	t _{su(MDIO-MDC)}	Setup time, MDIO valid before MDC high	90		ns
2	t _{h(MDIO-MDC)}	Hold time, MDIO valid from MDC high	0		ns

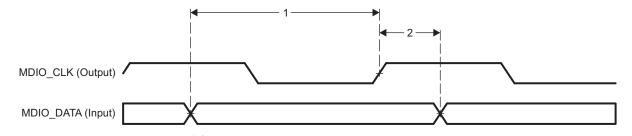


图 7-3. MDIO_DATA Timing - Input Mode

表 7-7. Switching Characteristics for MDIO_CLK

(see 图 7-4)

NO.		PARAMETER		TYP MAX	UNIT
1	t _{c(MDC)}	Cycle time, MDC	400		ns
2	t _{w(MDCH)}	Pulse duration, MDC high	160		ns
3	t _{w(MDCL)}	Pulse duration, MDC low	160		ns
4	t _{t(MDC)}	Transition time, MDC		5	ns



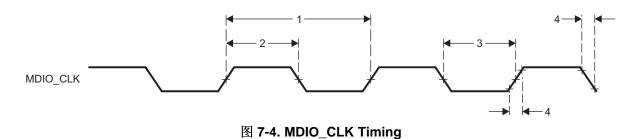
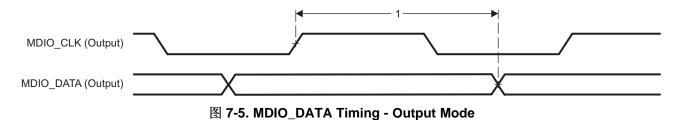


表 7-8. Switching Characteristics for MDIO_DATA

(see 图 7-5)

NO.		PARAMETER	MIN	TYP	MAX	UNIT
1	t _{d(MDC-MDIO)}	Delay time, MDC high to MDIO valid	10		390	ns



7.6.1.2 EMAC and Switch MII Electrical Data and Timing

表 7-9. Timing Requirements for GMII[x]_RXCLK - MII Mode

(see 图 7-6)

NO.			1	0 Mbps		10	UNIT		
NO.			MIN	TYP	MAX	MIN	TYP	MAX	UNII
1	t _{c(RX_CLK)}	Cycle time, RX_CLK	399.96		400.04	39.996		40.004	ns
2	t _{w(RX_CLKH)}	Pulse duration, RX_CLK high	140		260	14		26	ns
3	t _{w(RX_CLKL)}	Pulse duration, RX_CLK low	140		260	14		26	ns
4	t _{t(RX CLK)}	Transition time, RX_CLK			5			5	ns

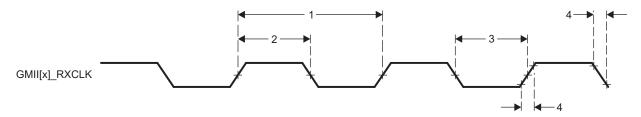


图 7-6. GMII[x]_RXCLK Timing - MII Mode



表 7-10. Timing Requirements for GMII[x]_TXCLK - MII Mode

(see 图 7-7)

NO.			1	0 Mbps		1	UNIT		
NO.			MIN	TYP	MAX	MIN	TYP	MAX	UNIT
1	t _{c(TX_CLK)}	Cycle time, TX_CLK	399.96		400.04	39.996		40.004	ns
2	t _{w(TX_CLKH)}	Pulse duration, TX_CLK high	140		260	14		26	ns
3	t _{w(TX_CLKL)}	Pulse duration, TX_CLK low	140		260	14		26	ns
4	t _{t(TX_CLK)}	Transition time, TX_CLK			5			5	ns



图 7-7. GMII[x]_TXCLK Timing - MII Mode

表 7-11. Timing Requirements for GMII[x]_RXD[3:0], GMII[x]_RXDV, and GMII[x]_RXER - MII Mode

(see 图 7-8)

NO			1	0 Mbps		10		UNIT	
			MIN	TYP	MAX	MIN	TYP	MAX	UNIT
	t _{su(RXD-RX_CLK)}	Setup time, RXD[3:0] valid before RX_CLK							
1	t _{su(RX_DV-RX_CLK)}	Setup time, RX_DV valid before RX_CLK	8			8			ns
	t _{su(RX_ER-RX_CLK)}	Setup time, RX_ER valid before RX_CLK							
	t _{h(RX_CLK-RXD)}	Hold time RXD[3:0] valid after RX_CLK							
2	t _{h(RX_CLK-RX_DV)}	Hold time RX_DV valid after RX_CLK	8			8			ns
	th(RX CLK-RX ER)	Hold time RX_ER valid after RX_CLK							

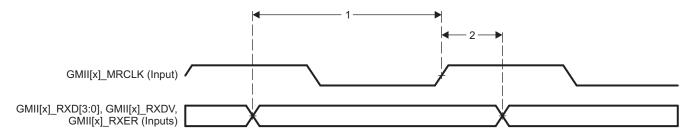


图 7-8. GMII[x]_RXD[3:0], GMII[x]_RXDV, GMII[x]_RXER Timing - MII Mode

www.ti.com.cn

表 7-12. Switching Characteristics for GMII[x]_TXD[3:0], and GMII[x]_TXEN - MII Mode

(see 图 7-9)

NO.		PARAMETER	10 Mbps			10		UNIT	
NO.		MIN	TYP	MAX	MIN	TYP	MAX	UNIT	
1	t _{d(TX_CLK-TXD)} Delay time, TX_CLK high to TXD[3:0] valid		F		25	Е		25	20
'	t _{d(TX_CLK-TX_EN)}	Delay time, TX_CLK to TX_EN valid	5		25	5		25	ns



图 7-9. $GMII[x]_TXD[3:0]$, $GMII[x]_TXEN$ Timing - MII Mode



7.6.1.3 EMAC and Switch RMII Electrical Data and Timing

表 7-13. Timing Requirements for RMII[x]_REFCLK - RMII Mode

(see **§** 7-10)

NO.			MIN	TYP MAX	UNIT
1	t _{c(REF_CLK)}	Cycle time, REF_CLK	19.999	20.001	ns
2	t _{w(REF_CLKH)}	Pulse duration, REF_CLK high	7	13	ns
3	t _{w(REF CLKL)}	Pulse duration, REF_CLK low	7	13	ns

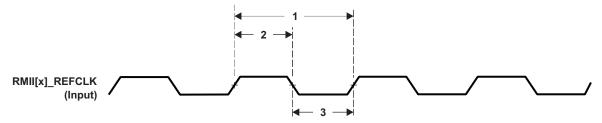


图 7-10. RMII[x]_REFCLK Timing - RMII Mode

表 7-14. Timing Requirements for RMII[x]_RXD[1:0], RMII[x]_CRS_DV, and RMII[x]_RXER - RMII Mode

(see 图 7-11)

NO.			MIN	TYP	MAX	UNIT
	t _{su(RXD-REF_CLK)}	Setup time, RXD[1:0] valid before REF_CLK				
1	t _{su(CRS_DV-REF_CLK)}	Setup time, CRS_DV valid before REF_CLK	4			ns
	t _{su(RX_ER-REF_CLK)}	Setup time, RX_ER valid before REF_CLK				
	t _{h(REF_CLK-RXD)}	Hold time RXD[1:0] valid after REF_CLK				
2	t _{h(REF_CLK-CRS_DV)}	Hold time, CRS_DV valid after REF_CLK	2			ns
	th(REF CLK-RX ER)	Hold time, RX_ER valid after REF_CLK				

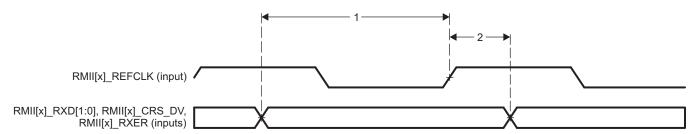


图 7-11. RMII[x]_RXD[1:0], RMII[x]_CRS_DV, RMII[x]_RXER Timing - RMII Mode

www.ti.com.cn

表 7-15. Switching Characteristics for RMII[x]_TXD[1:0], and RMII[x]_TXEN - RMII Mode

(see 图 7-12)

NO.		PARAMETER	MIN	TYP MA	λX	UNIT
1	t _{d(REF_CLK-TXD)}	Delay time, REF_CLK high to TXD[1:0] valid	2		12	no
ı	t _{d(REF_CLK-TXEN)}	Delay time, REF_CLK to TXEN valid			13	ns
2	t _{r(TXD)}	Rise time, TXD outputs	4		E	
2	t _{r(TX_EN)}	Rise time, TX_EN output	I		Э	ns
2	t _{f(TXD)}	Fall time, TXD outputs	4		E	
3	t _{f(TX EN)}	Fall time, TX_EN output	1		Э	ns

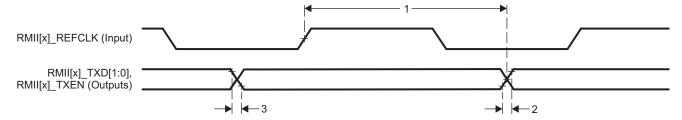


图 7-12. RMII[x]_TXD[1:0], RMII[x]_TXEN Timing - RMII Mode



7.6.1.4 EMAC and Switch RGMII Electrical Data and Timing

表 7-16. Timing Requirements for RGMII[x]_RCLK - RGMII Mode

(see 图 7-13)

NO			,	10 Mbps		100 Mbps		10	UNIT	
NO.			MIN	TYP M	X MIN	TYP	MAX	MIN	TYP MAX	UNII
1	t _{c(RXC)}	Cycle time, RXC	360	4	10 36	;	44	7.2	8.8	ns
2	t _{w(RXCH)}	Pulse duration, RXC high	160	2	10 16	;	24	3.6	4.4	ns
3	t _{w(RXCL)}	Pulse duration, RXC low	160	2	10 16	1	24	3.6	4.4	ns
4	t _{t(RXC)}	Transition time, RXC		0.	75		0.75		0.75	ns

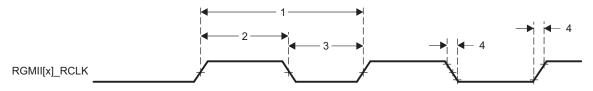
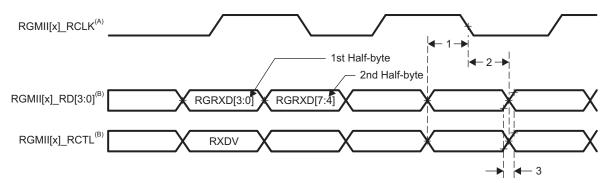


图 7-13. RGMII[x]_RCLK Timing - RGMII Mode

表 7-17. Timing Requirements for RGMII[x]_RD[3:0], and RGMII[x]_RCTL - RGMII Mode

(see 图 7-14)

NO			1	0 Mbps		10	00 Mbps	3	10	00 Mbps	s	LINUT
NO.			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
4	t _{su(RD-RXC)}	Setup time, RD[3:0] valid before RXC high or low	1			1			1			
1	t _{su(RX_CTL-RXC)}	Setup time, RX_CTL valid before RXC high or low	1			1			1			ns
	t _{h(RXC-RD)}	Hold time, RD[3:0] valid after RXC high or low	1			1			1			
2	t _{h(RXC-RX_CTL)}	Hold time, RX_CTL valid after RXC high or low	1			1			1			ns
3	t _{t(RD)}	Transition time, RD			0.75			0.75			0.75	
3	t _{t(RX_CTL)}	Transition time, RX_CTL			0.75			0.75			0.75	ns



- A. RGMII[x]_RCLK must be externally delayed relative to the RGMII[x]_RD[3:0] and RGMII[x]_RCTL signals to meet the respective timing requirements.
- B. Data and control information is received using both edges of the clocks. RGMII[x]_RD[3:0] carries data bits 3-0 on the rising edge of RGMII[x]_RCLK and data bits 7-4 on the falling edge of RGMII[x]_RCLK. Similarly, RGMII[x]_RCTL carries RXDV on rising edge of RGMII[x]_RCLK and RXERR on falling edge of RGMII[x]_RCLK.

图 7-14. RGMII[x]_RD[3:0], RGMII[x]_RCTL Timing - RGMII Mode



表 7-18. Switching Characteristics for RGMII[x]_TCLK - RGMII Mode

(see 图 7-15)

NO.		PARAMETER		0 Mbps		10	00 Mbps		10	UNIT		
NO.		PARAMETER	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
1	t _{c(TXC)}	Cycle time, TXC	360		440	36		44	7.2		8.8	ns
2	t _{w(TXCH)}	Pulse duration, TXC high	160		240	16		24	3.6		4.4	ns
3	t _{w(TXCL)}	Pulse duration, TXC low	160		240	16		24	3.6		4.4	ns
4	t _{t(TXC)}	Transition time, TXC			0.75			0.75			0.75	ns

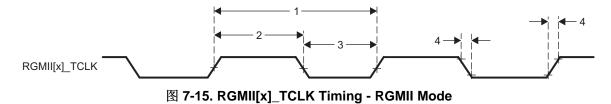
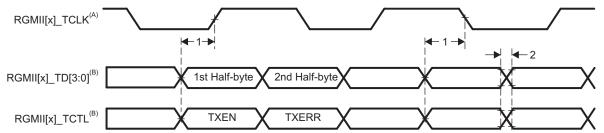


表 7-19. Switching Characteristics for RGMII[x]_TD[3:0], and RGMII[x]_TCTL - RGMII Mode

(see **8** 7-16)

NO.		PARAMETER	1	0 Mbps		10	00 Mbps		100	00 Mbps	s	UNIT
NO.		FARAMETER	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
4	t _{sk(TD-TXC)}	TD to TXC output skew	-0.5		0.5	-0.5		0.5	-0.5		0.5	
1	t _{sk(TX_CTL-TXC)}	TX_CTL to TXC output skew	-0.5		0.5	-0.5		0.5	-0.5		0.5	ns
2	t _{t(TD)}	Transition time, TD			0.75			0.75			0.75	
2	t _{t(TX CTL)}	Transition time, TX_CTL			0.75			0.75			0.75	ns



- A. The EMAC and switch implemented in the AM335x device supports internal delay mode, but timing closure was not performed for this mode of operation. Therefore, the AM335x device does not support internal delay mode.
- B. Data and control information is transmitted using both edges of the clocks. RGMII[x]_TD[3:0] carries data bits 3-0 on the rising edge of RGMII[x]_TCLK and data bits 7-4 on the falling edge of RGMII[x]_TCLK. Similarly, RGMII[x]_TCLL carries TXEN on rising edge of RGMII[x]_TCLK and TXERR of falling edge of RGMII[x]_TCLK.

图 7-16. RGMII[x]_TD[3:0], RGMII[x]_TCTL Timing - RGMII Mode

www.ti.com.cn

7.7 **External Memory Interfaces**

The device includes the following external memory interfaces:

- General-purpose memory controller (GPMC)
- mDDR(LPDDR), DDR2, DDR3, DDR3L Memory Interface (EMIF)

7.7.1 General-Purpose Memory Controller (GPMC)

注

For more information, see the Memory Subsystem and General-Purpose Memory Controller section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.

The GPMC is the unified memory controller used to interface external memory devices such as:

- Asynchronous SRAM-like memories and ASIC devices
- Asynchronous page mode and synchronous burst NOR flash
- · NAND flash

7.7.1.1 GPMC and NOR Flash—Synchronous Mode

表 7-21 and 表 7-22 assume testing over the recommended operating conditions and electrical characteristic conditions shown in 表 7-20 (see 图 7-17 through 图 7-21).

表 7-20. GPMC and NOR Flash Timing Conditions—Synchronous Mode

	PARAMETER	MIN	TYP MAX	UNIT			
Input Cond	ditions						
t _R	Input signal rise time		5	ns			
t _F	Input signal fall time		5	ns			
Output Condition							
C _{LOAD}	Output load capacitance	3	30	pF			

表 7-21. GPMC and NOR Flash Timing Requirements—Synchronous Mode

NO				OPP100		OPP50		LINIT	
NO.					MAX	MIN	MAX	UNIT	
F12	t _{su(dV-clkH)}	Setup time, input data gpmc_ad[15:0] valid before output clock gpmc_clk high		3.2		13.2		ns	
F13	t _{h(clkH-dV)}	Hold time, input data gpmc_ad[15:0] valid after output clock gpmc_clk	Industrial extended temperature (-40°C to 125°C)	4.74		4.74		ns	
		high	All other temperature ranges	4.74		2.75			
F21	t _{su(waitV-clkH)}	Setup time, input wait gpmc_wait[x] ⁽¹⁾ gpmc_clk high	valid before output clock	3.2		13.2		ns	
F22	t _{h(clkH-wait} V)	Hold time, input wait gpmc_wait[x] ⁽¹⁾ valid after output clock gpmc_clk	Industrial extended temperature (-40°C to 125°C)	4.74		4.74		ns	
		high	All other temperature ranges	4.74		2.75			

⁽¹⁾ In gpmc_wait[x], x is equal to 0 or 1.



表 7-22. GPMC and NOR Flash Switching Characteristics—Synchronous Mode⁽²⁾

NO		DADAMETED	OPP100		ОР	LINUT	
NO.		PARAMETER	MIN	MAX	MIN	MAX	UNIT
F0	1 / t _{c(clk)}	Frequency ⁽¹⁸⁾ , output clock gpmc_clk		100		50	MHz
F1	t _{w(clkH)}	Typical pulse duration, output clock gpmc_clk high	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	ns
F1	t _{w(clkL)}	Typical pulse duration, output clock gpmc_clk low	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	0.5P ⁽¹⁵⁾	ns
	t _{dc(clk)}	Duty cycle error, output clock gpmc_clk	-500	500	-500	500	ps
	t _{J(clk)}	Jitter standard deviation ⁽¹⁹⁾ , output clock gpmc_clk		33.33		33.33	ps
	t _{R(clk)}	Rise time, output clock gpmc_clk		2		2	ns
	t _{F(clk)}	Fall time, output clock gpmc_clk		2		2	ns
	t _{R(do)}	Rise time, output data gpmc_ad[15:0]		2		2	ns
	t _{F(do)}	Fall time, output data gpmc_ad[15:0]		2		2	ns
F2	t _{d(clkH-csnV)}	Delay time, output clock gpmc_clk rising edge to output chip select gpmc_csn[x] ⁽¹⁴⁾ transition	F ⁽⁶⁾ - 2.2	$F^{(6)} + 4.5$	F ⁽⁶⁾ - 3.2	F ⁽⁶⁾ + 9.5	ns
F3	t _{d(clkH-csnIV)}	Delay time, output clock gpmc_clk rising edge to output chip select gpmc_csn[x] ⁽¹⁴⁾ invalid	E ⁽⁵⁾ – 2.2	$E^{(5)} + 4.5$	$E^{(5)} - 3.2$	$E^{(5)} + 9.5$	ns
F4	t _{d(aV-clk)}	Delay time, output address gpmc_a[27:1] valid to output clock gpmc_clk first edge	$B^{(2)} - 4.5$	$B^{(2)} + 2.3$	$B^{(2)} - 5.5$	B ⁽²⁾ + 12.3	ns
F5	t _{d(clkH-alV)}	Delay time, output clock gpmc_clk rising edge to output address gpmc_a[27:1] invalid	-2.3	4.5	-3.3	14.5	ns
F6	t _{d(be[x]nV-clk)}	Delay time, output lower byte enable and command latch enable gpmc_be0n_cle, output upper byte enable gpmc_be1n valid to output clock gpmc_clk first edge	B ⁽²⁾ – 1.9	B ⁽²⁾ + 2.3	B ⁽²⁾ – 2.9	B ⁽²⁾ + 12.3	ns
F7	t _{d(clkH-be[x]nIV)}	Delay time, output clock gpmc_clk rising edge to output lower byte enable and command latch enable gpmc_be0n_cle, output upper byte enable gpmc_be1n invalid ⁽¹¹⁾	D ⁽⁴⁾ – 2.3	D ⁽⁴⁾ + 1.9	D ⁽⁴⁾ – 3.3	D ⁽⁴⁾ + 6.9	ns
F7	t _{d(clkL-be[x]nIV)}	Delay time, gpmc_clk falling edge to gpmc_nbe0_cle, gpmc_nbe1 invalid ⁽¹²⁾	$D^{(4)} - 2.3$	D ⁽⁴⁾ + 1.9	$D^{(4)} - 3.3$	D ⁽⁴⁾ + 6.9	ns
F7	t _{d(clkL-be[x]nIV)}	Delay time, gpmc_clk falling edge to gpmc_nbe0_cle, gpmc_nbe1 invalid ⁽¹³⁾	$D^{(4)} - 2.3$	D ⁽⁴⁾ + 1.9	$D^{(4)} - 3.3$	D ⁽⁴⁾ + 11.9	ns
F8	t _{d(clkH-advn)}	Delay time, output clock gpmc_clk rising edge to output address valid and address latch enable gpmc_advn_ale transition	$G^{(7)} - 2.3$	G ⁽⁷⁾ + 4.5	$G^{(7)} - 3.3$	G ⁽⁷⁾ + 9.5	ns
F9	t _{d(clkH-advnIV)}	Delay time, output clock gpmc_clk rising edge to output address valid and address latch enable gpmc_advn_ale invalid	$D^{(4)} - 2.3$	D ⁽⁴⁾ + 3.5	$D^{(4)} - 3.3$	D ⁽⁴⁾ + 9.5	ns
F10	t _{d(clkH-oen)}	Delay time, output clock gpmc_clk rising edge to output enable gpmc_oen transition	$H^{(8)} - 2.3$	H ⁽⁸⁾ + 3.5	$H^{(8)} - 3.3$	H ⁽⁸⁾ + 8.5	ns
F11	t _{d(clkH-oenIV)}	Delay time, output clock gpmc_clk rising edge to output enable gpmc_oen invalid	$E^{(8)} - 2.3$	$E^{(8)} + 3.5$	$E^{(8)} - 3.3$	$E^{(8)} + 8.5$	ns
F14	t _{d(clkH-wen)}	Delay time, output clock gpmc_clk rising edge to output write enable gpmc_wen transition	$I^{(9)} - 2.3$	$I^{(9)} + 4.5$	$I^{(9)} - 3.3$	I ⁽⁹⁾ + 9.5	ns
F15	t _{d(clkH-do)}	Delay time, output clock gpmc_clk rising edge to output data gpmc_ad[15:0] transition ⁽¹¹⁾	$J^{(10)} - 2.3$	J ⁽¹⁰⁾ + 1.9	$J^{(10)} - 3.3$	J ⁽¹⁰⁾ + 6.9	ns
F15	t _{d(clkL-do)}	Delay time, gpmc_clk falling edge to gpmc_ad[15:0] data bus transition ⁽¹²⁾	$J^{(10)} - 2.3$	J ⁽¹⁰⁾ + 1.9	$J^{(10)} - 3.3$	J ⁽¹⁰⁾ + 6.9	ns
F15	t _{d(clkL-do)}	Delay time, gpmc_clk falling edge to gpmc_ad[15:0] data bus transition ⁽¹³⁾	$J^{(10)} - 2.3$	J ⁽¹⁰⁾ + 1.9	$J^{(10)} - 3.3$	J ⁽¹⁰⁾ + 11.9	ns
F17	t _{d(clkH-be[x]n)}	Delay time, output clock gpmc_clk rising edge to output lower byte enable and command latch enable gpmc_be0n_cle transition ⁽¹¹⁾	J ⁽¹⁰⁾ – 2.3	J ⁽¹⁰⁾ + 1.9	J ⁽¹⁰⁾ – 3.3	J ⁽¹⁰⁾ + 6.9	ns
F17	t _{d(clkL-be[x]n)}	Delay time, gpmc_clk falling edge to gpmc_nbe0_cle, gpmc_nbe1 transition ⁽¹²⁾	$J^{(10)} - 2.3$	J ⁽¹⁰⁾ + 1.9	$J^{(10)} - 3.3$	J ⁽¹⁰⁾ + 6.9	ns
F17	t _{d(clkL-be[x]n)}	Delay time, gpmc_clk falling edge to gpmc_nbe0_cle, gpmc_nbe1 transition ⁽¹³⁾	$J^{(10)} - 2.3$	J ⁽¹⁰⁾ + 1.9	$J^{(10)} - 3.3$	J ⁽¹⁰⁾ + 11.9	ns



表 7-22. GPMC and NOR Flash Switching Characteristics—Synchronous Mode⁽²⁾ (continued)

NO	. PARAMETER			OPP100		OPP50		UNIT
NO.				MIN	MAX	MIN	MAX	UNII
E40	F18 $t_{w(csnV)}$ Pulse duration, output chip select $gpmc_csn[x]^{(14)}$ low	Pulse duration, output chip select	Read	A ⁽¹⁾		A ⁽¹⁾		ns
F18		gpmc_csn[x] ⁽¹⁴⁾ low	Write	A ⁽¹⁾		A ⁽¹⁾		ns
	F19 $t_{w(be[x]nV)}$ Pulse duration, output lower byte enable and command latch enable gpmc_be0n_cle, output upper byte enable gpmc_be1n low	Read	C ₍₃₎		C ₍₃₎		ns	
F19		Write	C ₍₃₎		C ₍₃₎		ns	
F20	t _{w(advnV)}	Pulse duration, output address valid and address latch enable gpmc_advn_ale low	Read	K ⁽¹⁶⁾		K ⁽¹⁶⁾		ns
F20			Write	K ⁽¹⁶⁾		K ⁽¹⁶⁾		ns

- (1) For single read: A = (CSRdOffTime CSOnTime) x (TimeParaGranularity + 1) x GPMC_FCLK(17) For burst read: A = (CSRdOffTime - CSOnTime + (n - 1) x PageBurstAccessTime) x (TimeParaGranularity + 1) x GPMC_FCLK(17) For burst write: $A = (CSWrOffTime - CSOnTime + (n - 1) \times PageBurstAccessTime) \times (TimeParaGranularity + 1) \times GPMC_FCLK^{(17)}$ With n being the page burst access number.
- (2) B = ClkActivationTime × GPMC FCLK⁽¹⁷⁾
- (3) For single read: C = RdCycleTime x (TimeParaGranularity + 1) x GPMC_FCLK (17) For burst read: $C = (RdCycleTime + (n - 1) \times PageBurstAccessTime) \times (TimeParaGranularity + 1) \times GPMC_FCLK^{(17)}$ For burst write: $C = (WrCycleTime + (n - 1) \times PageBurstAccessTime) \times (TimeParaGranularity + 1) \times GPMC_FCLK^{(17)}$ With n being the page burst access number.
- (4) For single read: D = (RdCycleTime AccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK(17) For burst read: D = (RdCycleTime - AccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK(17) For burst write: D = (WrCycleTime – AccessTime) \times (TimeParaGranularity + 1) \times GPMC_FCLK⁽¹⁷⁾
- (5) For single read: E = (CSRdOffTime AccessTime) x (TimeParaGranularity + 1) x GPMC_FCLK⁽¹⁷⁾ For burst read: E = (CSRdOffTime - AccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK(17) For burst write: E = (CSWrOffTime - AccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁷⁾
- (6) For csn falling edge (CS activated):
 - Case GpmcFCLKDivider = 0:
 - $F = 0.5 \times CSExtraDelay \times GPMC_FCLK^{(17)}$
 - Case GpmcFCLKDivider = 1:
 - F = 0.5 x CSExtraDelay x GPMC FCLK⁽¹⁷⁾ if (ClkActivationTime and CSOnTime are odd) or (ClkActivationTime and CSOnTime are even)
 - $F = (1 + 0.5 \times CSExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
 - Case GpmcFCLKDivider = 2:

 - F = 0.5 × CSExtraDelay × GPMC_FCLK⁽¹⁷⁾ if ((CSOnTime ClkActivationTime) is a multiple of 3)

 F = (1 + 0.5 × CSExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((CSOnTime ClkActivationTime 1) is a multiple of 3)

 F = (2 + 0.5 × CSExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((CSOnTime ClkActivationTime 2) is a multiple of 3)
- (7) For ADV falling edge (ADV activated):
 - Case GpmcFCLKDivider = 0:
 - $G = 0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$
 - Case GpmcFCLKDivider = 1:
 - G = 0.5 x ADVExtraDelay x GPMC_FCLK⁽¹⁷⁾ if (ClkActivationTime and ADVOnTime are odd) or (ClkActivationTime and ADVOnTime are even)
 - $G = (1 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
 - Case GpmcFCLKDivider = 2:

 - G = $0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$ if ((ADVOnTime ClkActivationTime) is a multiple of 3) G = $(1 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVOnTime ClkActivationTime 1) is a multiple of 3) G = $(2 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVOnTime ClkActivationTime 2) is a multiple of 3)

For ADV rising edge (ADV deactivated) in Reading mode:

- Case GpmcFCLKDivider = 0:
 - $G = 0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$
- Case GpmcFCLKDivider = 1:
 - G = 0.5 × ADVExtraDelay × GPMC_FCLK⁽¹⁷⁾ if (ClkActivationTime and ADVRdOffTime are odd) or (ClkActivationTime and ADVRdOffTime are even)
 - $G = (1 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
- Case GpmcFCLKDivider = 2:

 - $G = 0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$ if ((ADVRdOffTime ClkActivationTime) is a multiple of 3) $G = (1 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVRdOffTime ClkActivationTime 1) is a multiple of 3)
 - $G = (2 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVRdOffTime ClkActivationTime 2) is a multiple of 3)

For ADV rising edge (ADV deactivated) in Writing mode:

- Case GpmcFCLKDivider = 0:
 - $G = 0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$
- Case GpmcFCLKDivider = 1:



- $G = 0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$ if (ClkActivationTime and ADVWrOffTime are odd) or (ClkActivationTime and ADVWrOffTime are even)
- $G = (1 + 0.5 \times ADVExtraDelav) \times GPMC FCLK^{(17)}$ otherwise
- Case GpmcFCLKDivider = 2:

 - G = $0.5 \times ADVExtraDelay \times GPMC_FCLK^{(17)}$ if ((ADVWrOffTime ClkActivationTime) is a multiple of 3) G = $(1 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVWrOffTime ClkActivationTime 1) is a multiple of 3) G = $(2 + 0.5 \times ADVExtraDelay) \times GPMC_FCLK^{(17)}$ if ((ADVWrOffTime ClkActivationTime 2) is a multiple of 3)
- (8) For OE falling edge (OE activated) and I/O DIR rising edge (Data Bus input direction):
 - Case GpmcFCLKDivider = 0:
 - $H = 0.5 \times OEExtraDelay \times GPMC_FCLK^{(17)}$
 - Case GpmcFCLKDivider = 1:
 - H = 0.5 x OEExtraDelay x GPMC FCLK⁽¹⁷⁾ if (ClkActivationTime and OEOnTime are odd) or (ClkActivationTime and OEOnTime are even)
 - $H = (1 + 0.5 \times OEExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
 - Case GpmcFCLKDivider = 2:

 - H = 0.5 × OEExtraDelay × GPMC_FCLK⁽¹⁷⁾ if ((OEOnTime ClkActivationTime) is a multiple of 3)
 H = (1 + 0.5 × OEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((OEOnTime ClkActivationTime 1) is a multiple of 3)
 H = (2 + 0.5 × OEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((OEOnTime ClkActivationTime 2) is a multiple of 3)

For OE rising edge (OE deactivated):

- Case GpmcFCLKDivider = 0:
 - $H = 0.5 \times OEExtraDelay \times GPMC_FCLK^{(17)}$
- Case GpmcFCLKDivider = 1:
 - H = 0.5 × OEExtraDelay × GPMC_FCLK⁽¹⁷⁾ if (ClkActivationTime and OEOffTime are odd) or (ClkActivationTime and OEOffTime are even)
- $H = (1 + 0.5 \times OEExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
- Case GpmcFCLKDivider = 2:

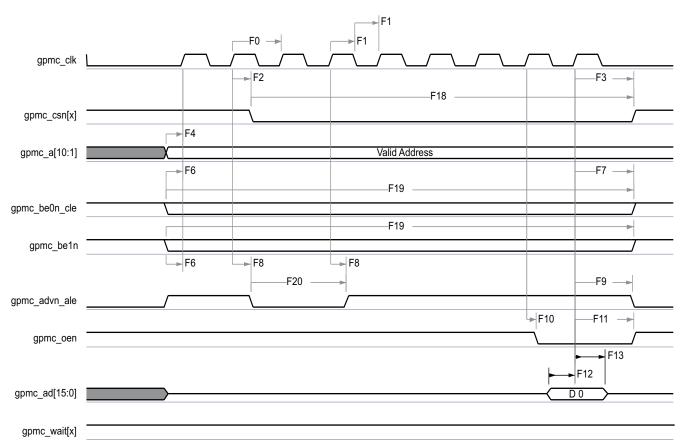
 - H = 0.5 × OEExtraDelay × GPMC_FCLK⁽¹⁷⁾ if ((OEOffTime ClkActivationTime) is a multiple of 3) H = $(1 + 0.5 \times OEExtraDelay) \times GPMC_FCLK⁽¹⁷⁾$ if ((OEOffTime ClkActivationTime 1) is a multiple of 3)
 - $H = (2 + 0.5 \times OEExtraDelay) \times GPMC_FCLK^{(17)}$ if ((OEOffTime ClkActivationTime 2) is a multiple of 3)
- (9) For WE falling edge (WE activated):
 - Case GpmcFCLKDivider = 0:
 - $I = 0.5 \times WEExtraDelay \times GPMC_FCLK^{(17)}$
 - Case GpmcFCLKDivider = 1:
 - I = 0.5 x WEExtraDelay x GPMC FCLK(17) if (ClkActivationTime and WEOnTime are odd) or (ClkActivationTime and WEOnTime are even)
 - $I = (1 + 0.5 \times WEExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
 - Case GpmcFCLKDivider = 2:

 - I = 0.5 × WEExtraDelay × GPMC_FCLK⁽¹⁷⁾ if ((WEOnTime ClkActivationTime) is a multiple of 3)
 I = (1 + 0.5 × WEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((WEOnTime ClkActivationTime 1) is a multiple of 3)
 I = (2 + 0.5 × WEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((WEOnTime ClkActivationTime 2) is a multiple of 3)

For WE rising edge (WE deactivated):

- Case GpmcFCLKDivider = 0:
 - I = 0.5 x WEExtraDelay x GPMC_FCLK (17)
- Case GpmcFCLKDivider = 1:
 - I = 0.5 × WEExtraDelay × GPMC_FCLK⁽¹⁷⁾ if (ClkActivationTime and WEOffTime are odd) or (ClkActivationTime and WEOffTime are even)
 - $I = (1 + 0.5 \times WEExtraDelay) \times GPMC_FCLK^{(17)}$ otherwise
- Case GpmcFCLKDivider = 2:
 - I = 0.5 x WEExtraDelay x GPMC_FCLK⁽¹⁷⁾ if ((WEOffTime ClkActivationTime) is a multiple of 3)
 - I = (1 + 0.5 × WEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((WEOffTime ClkActivationTime 1) is a multiple of 3)
 I = (2 + 0.5 × WEExtraDelay) × GPMC_FCLK⁽¹⁷⁾ if ((WEOffTime ClkActivationTime 2) is a multiple of 3)
- (10) $J = GPMC_FCLK^{(17)}$
- (11) First transfer only for CLK DIV 1 mode.
- (12) Half cycle; for all data after initial transfer for CLK DIV 1 mode.
- (13) Half cycle of GPMC_CLK_OUT; for all data for modes other than CLK DIV 1 mode. GPMC_CLK_OUT divide down from GPMC_FCLK.
- (14) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.
- (15) P = gpmc_clk period in ns
- (16) For read: K = (ADVRdOffTime ADVOnTime) x (TimeParaGranularity + 1) x GPMC_FCLK(17) For write: K = (ADVWrOffTime - ADVOnTime) x (TimeParaGranularity + 1) x GPMC_FCLK(17)
- (17) GPMC FCLK is general-purpose memory controller internal functional clock period in ns.
- (18) Related to the gpmc_clk output clock maximum and minimum frequencies programmable in the GPMC module by setting the GPMC CONFIG1 CSx configuration register bit field GpmcFCLKDivider.
- (19) The jitter probability density can be approximated by a Gaussian function.

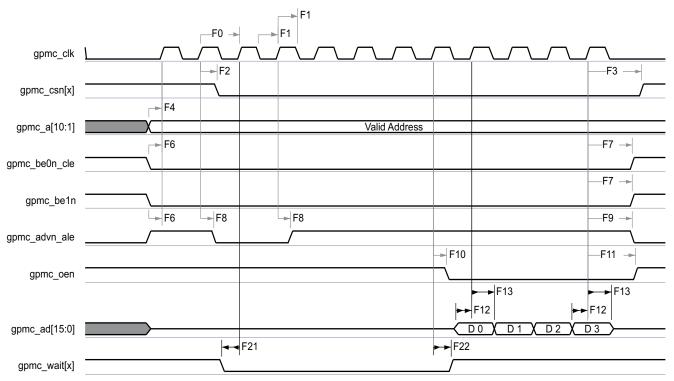




A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-17. GPMC and NOR Flash—Synchronous Single Read—(GpmcFCLKDivider = 0)

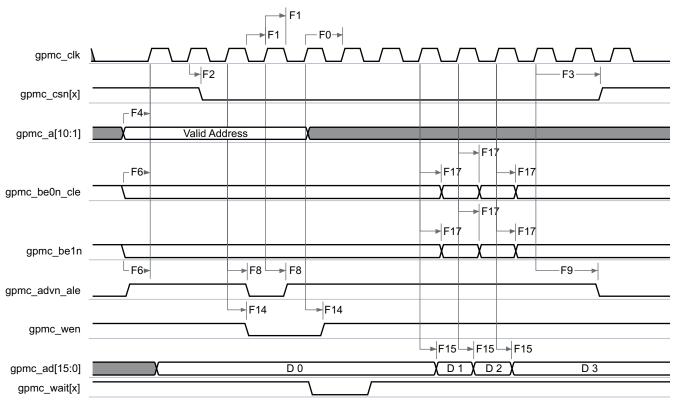
B. In gpmc_wait[x], x is equal to 0 or 1.



- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.
- B. In gpmc_wait[x], x is equal to 0 or 1.

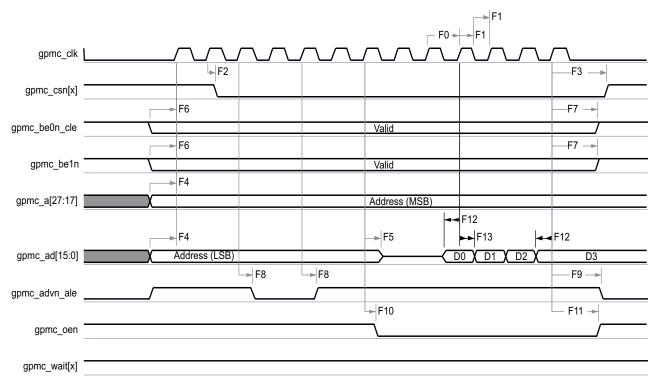
图 7-18. GPMC and NOR Flash—Synchronous Burst Read—4x16-Bit (GpmcFCLKDivider = 0)





- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.
- B. In gpmc_wait[x], x is equal to 0 or 1.

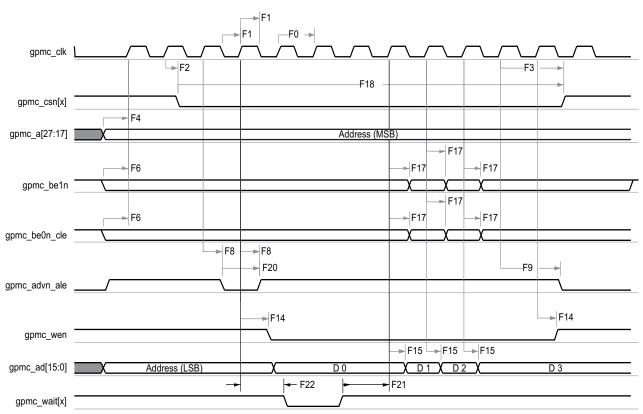
图 7-19. GPMC and NOR Flash—Synchronous Burst Write—(GpmcFCLKDivider > 0)



A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-20. GPMC and Multiplexed NOR Flash—Synchronous Burst Read

B. In gpmc_wait[x], x is equal to 0 or 1.



A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-21. GPMC and Multiplexed NOR Flash—Synchronous Burst Write

B. In gpmc_wait[x], x is equal to 0 or 1.

www.ti.com.cn

7.7.1.2 GPMC and NOR Flash—Asynchronous Mode

表 7-24 and 表 7-25 assume testing over the recommended operating conditions and electrical characteristic conditions shown in 表 7-23 (see \boxtimes 7-22 through \boxtimes 7-27).

${f \pounds}$ 7-23. GPMC and NOR Flash Timing Conditions—Asynchronous Mode

		MIN	TYP MAX	UNIT			
Input Conditi	ons						
t _R	Input signal rise time	1	5	ns			
t _F	Input signal fall time	1	5	ns			
Output Cond	Output Condition						
C _{LOAD}	Output load capacitance	3	30	pF			

表 7-24. GPMC and NOR Flash Internal Timing Requirements—Asynchronous Mode⁽¹⁾⁽²⁾

NO.		OPP100	OPP50	UNIT
NO.		MIN MAX	MIN MAX	UNII
FI1	Delay time, output data gpmc_ad[15:0] generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI2	Delay time, input data gpmc_ad[15:0] capture from internal functional clock GPMC_FCLK ⁽³⁾	4	4	ns
FI3	Delay time, output chip select gpmc_csn[x] generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI4	Delay time, output address gpmc_a[27:1] generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI5	Delay time, output address gpmc_a[27:1] valid from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI6	Delay time, output lower-byte enable and command latch enable gpmc_be0n_cle, output upper-byte enable gpmc_be1n generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI7	Delay time, output enable gpmc_oen generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI8	Delay time, output write enable gpmc_wen generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
FI9	Skew, internal functional clock GPMC_FCLK ⁽³⁾	100	100	ps

⁽¹⁾ The internal parameters table must be used to calculate data access time stored in the corresponding CS register bit field.

⁽²⁾ Internal parameters are referred to the GPMC functional internal clock which is not provided externally.

⁽³⁾ GPMC_FCLK is general-purpose memory controller internal functional clock.



表 7-25. GPMC and NOR Flash Timing Requirements—Asynchronous Mode

NO.			OPP100		OPP50		UNIT
			MIN	MAX	MIN	MAX	
FA5 ⁽¹⁾	t _{acc(d)}	Data access time		H ⁽⁵⁾		H ⁽⁵⁾	ns
FA20 ⁽²⁾	t _{acc1-pgmode(d)}	Page mode successive data access time		P ⁽⁴⁾		P ⁽⁴⁾	ns
FA21 ⁽³⁾	t _{acc2-pgmode(d)}	Page mode first data access time		H ⁽⁵⁾		H ⁽⁵⁾	ns

- (1) The FA5 parameter shows the amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA5 functional clock cycles, input data is internally sampled by active functional clock edge. FA5 value must be stored inside the AccessTime register bit field.
- (2) The FA20 parameter shows amount of time required to internally sample successive input page data. It is expressed in number of GPMC functional clock cycles. After each access to input page data, next input page data is internally sampled by active functional clock edge after FA20 functional clock cycles. The FA20 value must be stored in the PageBurstAccessTime register bit field.
- (3) The FA21 parameter shows amount of time required to internally sample first input page data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA21 functional clock cycles, first input page data is internally sampled by active functional clock edge. FA21 value must be stored inside the AccessTime register bit field.
- (4) P = PageBurstAccessTime × (TimeParaGranularity + 1) × GPMC_FCLK⁽⁶⁾
- (5) H = AccessTime × (TimeParaGranularity + 1) × GPMC_FCLK⁽⁶⁾
- (6) GPMC_FCLK is general-purpose memory controller internal functional clock period in ns.

表 7-26. GPMC and NOR Flash Switching Characteristics—Asynchronous Mode

NO	DADAMETED			OPP100		OPP50		
NO.		PARAMETER		MIN	MAX	MIN	MAX	UNIT
	t _{R(d)}	Rise time, output data gpmc_ad[15:0]			2		2	ns
	t _{F(d)}	Fall time, output data gpmc_ad[15:0]			2		2	ns
		Pulse duration, output lower-byte	Read		N ⁽¹²⁾		N ⁽¹²⁾	
FA0	t _{w(be[x]nV)}	enable and command latch enable gpmc_be0n_cle, output upper-byte enable gpmc_be1n valid time	Write		N ⁽¹²⁾		N ⁽¹²⁾	ns
FA1	•	Pulse duration, output chip select	Read		A ⁽¹⁾		A ⁽¹⁾	ns
IAI	t _{w(csnV)}	gpmc_csn[x] ⁽¹³⁾ low	Write		A ⁽¹⁾		A ⁽¹⁾	115
		Delay time, output chip select	Read	$B^{(2)} - 0.2$	$B^{(2)} + 2.0$	$B^{(2)} - 5$	B ⁽²⁾ + 5	
FA3	t _{d(csnV-advnIV)}	gpmc_csn[x] ⁽¹³⁾ valid to output address valid and address latch enable gpmc_advn_ale invalid	Write	$B^{(2)} - 0.2$	$B^{(2)} + 2.0$	$B^{(2)} - 5$	B ⁽²⁾ + 5	ns
FA4	t _{d(csnV-oenIV)}	Delay time, output chip select gpmc_c valid to output enable gpmc_oen inval read)	Delay time, output chip select gpmc_csn[x] ⁽¹³⁾ valid to output enable gpmc_oen invalid (Single read)		C ⁽³⁾ + 2.0	C ⁽³⁾ – 5	C ⁽³⁾ + 5	ns
FA9	t _{d(aV-csnV)}	Delay time, output address gpmc_a[2 to output chip select gpmc_csn[x] ⁽¹³⁾ v		$J^{(9)} - 0.2$	J ⁽⁹⁾ + 2.0	$J^{(9)} - 5$	J ⁽⁹⁾ + 5	ns
FA10	t _{d(be[x]nV-csnV)}	Delay time, output lower-byte enable a command latch enable gpmc_be0n_ci upper-byte enable gpmc_be1n valid to chip select gpmc_csn[x] ⁽¹³⁾ valid	e, output	J ⁽⁹⁾ – 0.2	J ⁽⁹⁾ + 2.0	J ⁽⁹⁾ – 5	J ⁽⁹⁾ + 5	ns
FA12	t _{d(csnV-advnV)}	Delay time, output chip select gpmc_c valid to output address valid and addrenable gpmc_advn_ale valid		K ⁽¹⁰⁾ – 0.2	K ⁽¹⁰⁾ + 2.0	K ⁽¹⁰⁾ – 5	K ⁽¹⁰⁾ + 5	ns
FA13	t _{d(csnV-oenV)}	Delay time, output chip select gpmc_c valid to output enable gpmc_oen valid		L ⁽¹¹⁾ – 0.2	L ⁽¹¹⁾ + 2.0	L ⁽¹¹⁾ – 5	L ⁽¹¹⁾ + 5	ns
FA16	t _{w(aIV)}	Pulse durationm output address gpmo invalid between 2 successive read and accesses		G ⁽⁷⁾		G ⁽⁷⁾		ns
FA18	t _{d(csnV-oenIV)}	Delay time, output chip select gpmc_c valid to output enable gpmc_oen inval read)	sn[x] ⁽¹³⁾ id (Burst	I ⁽⁸⁾ - 0.2	I ⁽⁸⁾ + 2.0	I ⁽⁸⁾ – 5	I ⁽⁸⁾ + 5	ns
FA20	t _{w(aV)}	Pulse duration, output address gpmc_valid - 2nd, 3rd, and 4th accesses	a[27:1]	D ⁽⁴⁾		D ⁽⁴⁾		ns
FA25	t _{d(csnV-wenV)}	Delay time, output chip select gpmc_c valid to output write enable gpmc_wer		E ⁽⁵⁾ - 0.2	E ⁽⁵⁾ + 2.0	E ⁽⁵⁾ – 5	E ⁽⁵⁾ + 5	ns

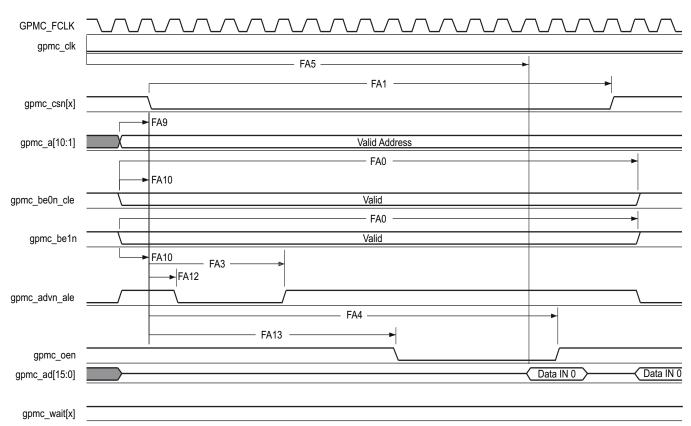


表 7-26. GPMC and NOR Flash Switching Characteristics—Asynchronous Mode (continued)

NO.	PARAMETER		OPP100		OPP50		LINUT
NO.			MIN	MAX	MIN	MAX	UNIT
FA27	t _{d(csnV-wenIV)}	Delay time, output chip select gpmc_csn[x] ⁽¹³⁾ valid to output write enable gpmc_wen invalid	F ⁽⁶⁾ - 0.2	F ⁽⁶⁾ + 2.0	F ⁽⁶⁾ – 5	F ⁽⁶⁾ + 5	ns
FA28	t _{d(wenV-dV)}	Delay time, output write enable gpmc_ wen valid to output data gpmc_ad[15:0] valid		2.0		5	ns
FA29	t _{d(dV-csnV)}	Delay time, output data gpmc_ad[15:0] valid to output chip select gpmc_csn[x] ⁽¹³⁾ valid	$J^{(9)} - 0.2$	J ⁽⁹⁾ + 2.0	J ⁽⁹⁾ – 5	J ⁽⁹⁾ + 5	ns
FA37	t _{d(oenV-aIV)}	Delay time, output enable gpmc_oen valid to output address gpmc_ad[15:0] phase end		2.0		5	ns

- (1) For single read: A = (CSRdOffTime CSOnTime) × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁴⁾
 For single write: A = (CSWrOffTime CSOnTime) × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁴⁾
 For burst read: A = (CSRdOffTime CSOnTime + (n 1) × PageBurstAccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁴⁾
 For burst write: A = (CSWrOffTime CSOnTime + (n 1) × PageBurstAccessTime) × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁴⁾
 with n being the page burst access number
- (2) For reading: B = ((ADVRdOffTime CSOnTime) \times (TimeParaGranularity + 1) + 0.5 \times (ADVExtraDelay CSExtraDelay)) \times GPMC_FCLK⁽¹⁴⁾
 For writing: B = ((ADVWrOffTime CSOnTime) \times (TimeParaGranularity + 1) + 0.5 \times (ADVExtraDelay CSExtraDelay)) \times GPMC_FCLK⁽¹⁴⁾
- (3) C = ((OEOffTime CSOnTime) x (TimeParaGranularity + 1) + 0.5 x (OEExtraDelay CSExtraDelay)) x GPMC_FCLK⁽¹⁴⁾
- (4) D = PageBurstAccessTime × (TimeParaGranularity + 1) × GPMC_FCLK⁽¹⁴⁾
- (5) E = ((WEOnTime CSOnTime) x (TimeParaGranularity + 1) + 0.5 x (WEExtraDelay CSExtraDelay)) x GPMC_FCLK(14)
- (6) F = ((WEOffTime CSOnTime) x (TimeParaGranularity + 1) + 0.5 x (WEExtraDelay CSExtraDelay)) x GPMC_FCLK(14)
- (7) G = Cycle2CycleDelay × GPMC_FCLK⁽¹⁴⁾
- (8) I = ((OEOffTime + (n 1) \times PageBurstAccessTime CSOnTime) \times (TimeParaGranularity + 1) + 0.5 \times (OEExtraDelay CSExtraDelay)) \times GPMC_FCLK⁽¹⁴⁾
- (9) J = (CSOnTime × (TimeParaGranularity + 1) + 0.5 × CSExtraDelay) × GPMC_FCLK(14)
- (10) K = ((ADVOnTime CSOnTime) × (TimeParaGranularity + 1) + 0.5 × (ADVExtraDelay CSExtraDelay)) × GPMC_FCLK⁽¹⁴⁾
- (11) L = ((OEOnTime CSOnTime) × (TimeParaGranularity + 1) + 0.5 × (OEExtraDelay CSExtraDelay)) × GPMC_FCLK⁽¹⁴⁾
- (12) For single read: N = RdCycleTime x (TimeParaGranularity + 1) x GPMC_FCLK⁽¹⁴⁾
 For single write: N = WrCycleTime x (TimeParaGranularity + 1) x GPMC_FCLK⁽¹⁴⁾
 For burst read: N = (RdCycleTime + (n 1) x PageBurstAccessTime) x (TimeParaGranularity + 1) x GPMC_FCLK⁽¹⁴⁾
 For burst write: N = (WrCycleTime + (n 1) x PageBurstAccessTime) x (TimeParaGranularity + 1) x GPMC_FCLK⁽¹⁴⁾
- (13) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.
- (14) GPMC_FCLK is general-purpose memory controller internal functional clock period in ns.

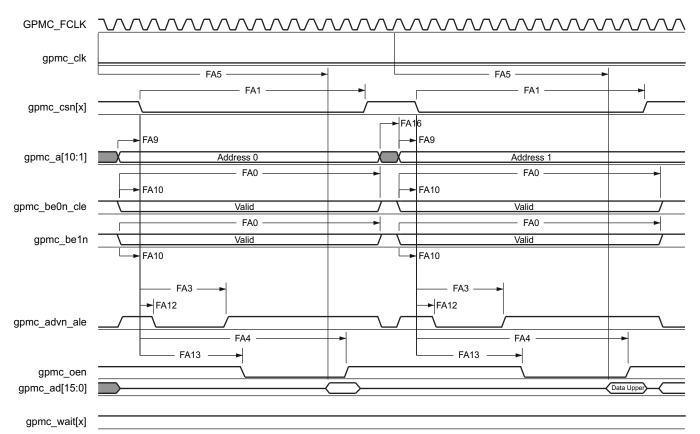




- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.
- B. FA5 parameter illustrates amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA5 functional clock cycles, input data will be internally sampled by active functional clock edge. FA5 value must be stored inside AccessTime register bits field.
- C. GPMC_FCLK is an internal clock (GPMC functional clock) not provided externally.

图 7-22. GPMC and NOR Flash—Asynchronous Read—Single Word

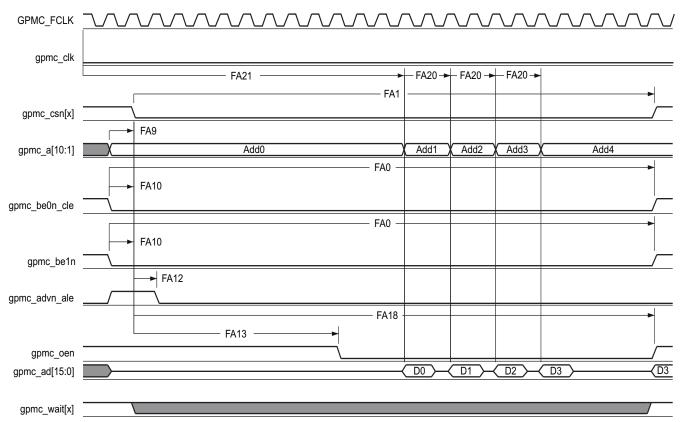




- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.
- B. FA5 parameter illustrates amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA5 functional clock cycles, input data will be internally sampled by active functional clock edge. FA5 value must be stored inside AccessTime register bits field.
- C. GPMC_FCLK is an internal clock (GPMC functional clock) not provided externally.

图 7-23. GPMC and NOR Flash—Asynchronous Read—32-Bit

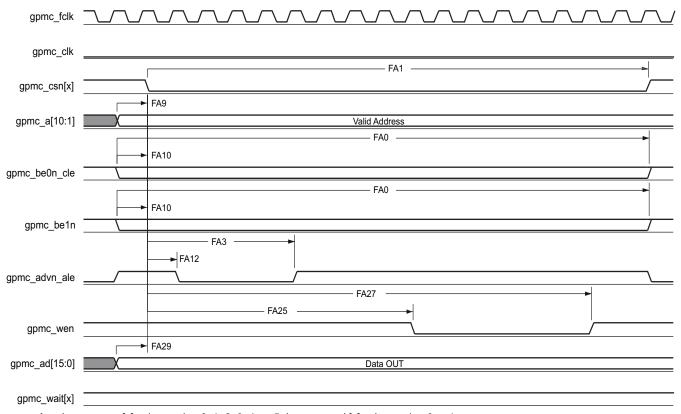




- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.
- B. FA21 parameter illustrates amount of time required to internally sample first input page data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA21 functional clock cycles, first input page data will be internally sampled by active functional clock edge. FA21 calculation must be stored inside AccessTime register bits field.
- C. FA20 parameter illustrates amount of time required to internally sample successive input page data. It is expressed in number of GPMC functional clock cycles. After each access to input page data, next input page data will be internally sampled by active functional clock edge after FA20 functional clock cycles. FA20 is also the duration of address phases for successive input page data (excluding first input page data). FA20 value must be stored in PageBurstAccessTime register bits field.
- D. GPMC_FCLK is an internal clock (GPMC functional clock) not provided externally.

图 7-24. GPMC and NOR Flash—Asynchronous Read—Page Mode 4x16-Bit

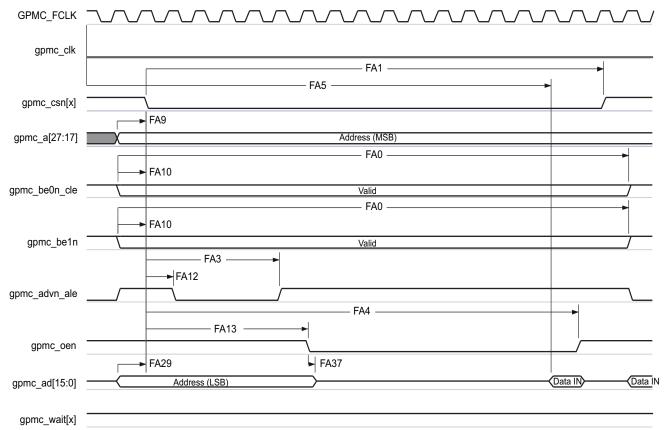




A. In $gpmc_csn[x]$, x is equal to 0, 1, 2, 3, 4, or 5. In $gpmc_wait[x]$, x is equal to 0 or 1.

图 7-25. GPMC and NOR Flash—Asynchronous Write—Single Word

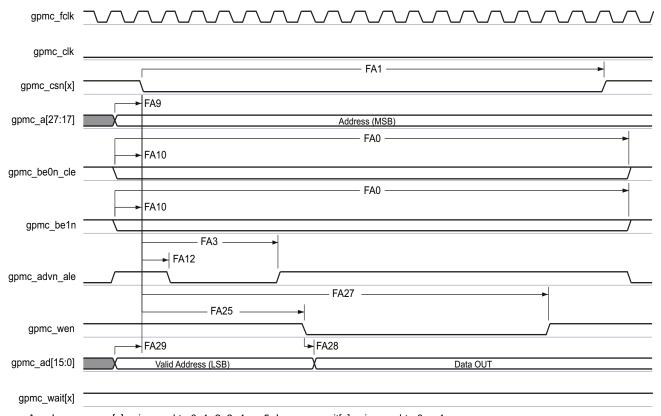




- A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.
- B. FA5 parameter illustrates amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after FA5 functional clock cycles, input data will be internally sampled by active functional clock edge. FA5 value must be stored inside AccessTime register bits field.
- C. GPMC_FCLK is an internal clock (GPMC functional clock) not provided externally.

图 7-26. GPMC and Multiplexed NOR Flash—Asynchronous Read—Single Word





A. In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.

图 7-27. GPMC and Multiplexed NOR Flash—Asynchronous Write—Single Word



7.7.1.3 GPMC and NAND Flash—Asynchronous Mode

表 7-28 and 表 7-29 assume testing over the recommended operating conditions and electrical characteristic conditions shown in 表 7-27 (see $\mathbb R$ 7-28 through $\mathbb R$ 7-31).

表 7-27. GPMC and NAND Flash Timing Conditions—Asynchronous Mode

	PARAMETER	MIN	TYP MAX	UNIT		
Input Cond	Input Conditions					
t _R	Input signal rise time	1	5	ns		
t _F	Input signal fall time	1	5	ns		
Output Condition						
C _{LOAD}	Output load capacitance	3	30	pF		

表 7-28. GPMC and NAND Flash Internal Timing Requirements—Asynchronous Mode⁽¹⁾⁽²⁾

NO.		OPP100	OPP50	UNIT
NO.		MIN MAX	MIN MAX	UNII
GNFI1	Delay time, output data gpmc_ad[15:0] generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI2	Delay time, input data gpmc_ad[15:0] capture from internal functional clock GPMC_FCLK ⁽³⁾	4.0	4.0	ns
GNFI3	Delay time, output chip select gpmc_csn[x] generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI4	Delay time, output address valid and address latch enable gpmc_advn_ale generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI5	Delay time, output lower-byte enable and command latch enable gpmc_be0n_cle generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI6	Delay time, output enable gpmc_oen generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI7	Delay time, output write enable gpmc_wen generation from internal functional clock GPMC_FCLK ⁽³⁾	6.5	6.5	ns
GNFI8	Skew, functional clock GPMC_FCLK ⁽³⁾	100	100	ps

⁽¹⁾ Internal parameters table must be used to calculate data access time stored in the corresponding CS register bit field.

表 7-29. GPMC and NAND Flash Timing Requirements—Asynchronous Mode

NO.		OPP100		OPP50		UNIT
NO.		MIN	MAX	MIN	MAX	UNII
GNF12 ⁽¹⁾	t _{acc(d)} Access time, input data gpmc_ad[15:0]		J ⁽²⁾		J ⁽²⁾	ns

⁽¹⁾ The GNF12 parameter illustrates the amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of the read cycle and after GNF12 functional clock cycles, input data is internally sampled by the active functional clock edge. The GNF12 value must be stored inside AccessTime register bit field.

- (2) J = AccessTime × (TimeParaGranularity + 1) × GPMC_FCLK⁽³⁾
- (3) GPMC_FCLK is general-purpose memory controller internal functional clock period in ns.

⁽²⁾ Internal parameters are referred to the GPMC functional internal clock which is not provided externally.

⁽³⁾ GPMC_FCLK is general-purpose memory controller internal functional clock.

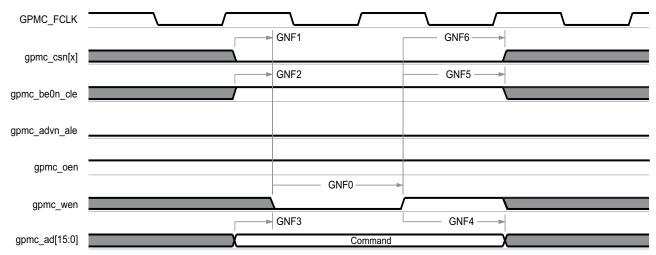


表 7-30. GPMC and NAND Flash Switching Characteristics—Asynchronous Mode

No	DADAMETED		OPP100		OPP50		UNIT
NO.		PARAMETER	MIN	MAX	MIN	MAX	UNII
	t _{R(d)}	Rise time, output data gpmc_ad[15:0]		2		2	ns
	t _{F(d)}	Fall time, output data gpmc_ad[15:0]		2		2	ns
GNF0	t _{w(wenV)}	Pulse duration, output write enable gpmc_wen valid	A ⁽¹⁾		A ⁽¹⁾		ns
GNF1	t _{d(csnV-wenV)}	Delay time, output chip select gpmc_csn[x] ⁽¹³⁾ valid to output write enable gpmc_wen valid	$B^{(2)} - 0.2$	$B^{(2)} + 2.0$	$B^{(2)} - 5$	B ⁽²⁾ + 5	ns
GNF2	t _{w(cleH-wenV)}	Delay time, output lower-byte enable and command latch enable gpmc_be0n_cle high to output write enable gpmc_wen valid	$C^{(3)} - 0.2$	C ⁽³⁾ + 2.0	C ⁽³⁾ – 5	C ⁽³⁾ + 5	ns
GNF3	t _{w(wenV-dV)}	Delay time, output data gpmc_ad[15:0] valid to output write enable gpmc_wen valid	$D^{(4)} - 0.2$	$D^{(4)} + 2.0$	$D^{(4)} - 5$	D ⁽⁴⁾ + 5	ns
GNF4	t _{w(wenIV-dIV)}	Delay time, output write enable gpmc_wen invalid to output data gpmc_ad[15:0] invalid	$E^{(5)} - 0.2$	E ⁽⁵⁾ + 5	$E^{(5)} - 5$	E ⁽⁵⁾ + 5	ns
GNF5	t _{w(wenIV-cleIV)}	Delay time, output write enable gpmc_wen invalid to output lower-byte enable and command latch enable gpmc_be0n_cle invalid	F ⁽⁶⁾ - 0.2	F ⁽⁶⁾ + 2.0	F ⁽⁶⁾ – 5	F ⁽⁶⁾ + 5	ns
GNF6	t _{w(wenIV-csnIV)}	Delay time, output write enable gpmc_wen invalid to output chip select gpmc_csn[x] ⁽¹³⁾ invalid	G ⁽⁷⁾ – 0.2	G ⁽⁷⁾ + 2.0	G ⁽⁷⁾ – 5	G ⁽⁷⁾ + 5	ns
GNF7	t _{w(aleH-wenV)}	Delay time, output address valid and address latch enable gpmc_advn_ale high to output write enable gpmc_wen valid	$C^{(3)} - 0.2$	C ⁽³⁾ + 2.0	C ⁽³⁾ – 5	C ⁽³⁾ + 5	ns
GNF8	t _{w(wenIV-aleIV)}	Delay time, output write enable gpmc_wen invalid to output address valid and address latch enable gpmc_advn_ale invalid	F ⁽⁶⁾ - 0.2	F ⁽⁶⁾ + 2.0	F ⁽⁶⁾ – 5	F ⁽⁶⁾ + 5	ns
GNF9	t _{c(wen)}	Cycle time, write		H ⁽⁸⁾		H ⁽⁸⁾	ns
GNF10	t _{d(csnV-oenV)}	Delay time, output chip select gpmc_csn[x] ⁽¹³⁾ valid to output enable gpmc_oen valid	$I^{(9)} - 0.2$	I ⁽⁹⁾ + 2.0	I ⁽⁹⁾ – 5	I ⁽⁹⁾ + 5	ns
GNF13	t _{w(oenV)}	Pulse duration, output enable gpmc_oen valid	_	K ⁽¹⁰⁾		K ⁽¹⁰⁾	ns
GNF14	t _{c(oen)}	Cycle time, read	L(11)		L ⁽¹¹⁾		ns
GNF15	t _{w(oenIV-csnIV)}	Delay time, output enable gpmc_oen invalid to output chip select gpmc_csn[x] ⁽¹³⁾ invalid	$M^{(12)} - 0.2$	$M^{(12)} + 2.0$	$M^{(12)} - 5$	$M^{(12)} + 5$	ns

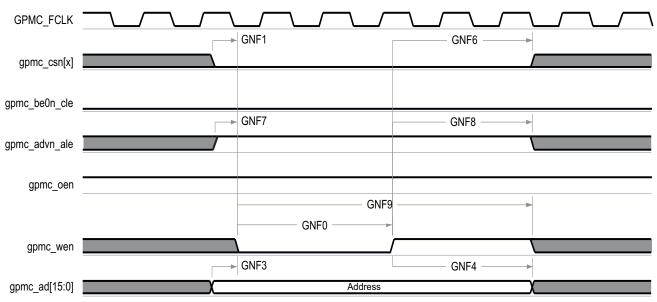
- (1) A = (WEOffTime WEOnTime) \times (TimeParaGranularity + 1) \times GPMC_FCLK⁽¹⁴⁾
- (2) B = ((WEOnTime CSOnTime) x (TimeParaGranularity + 1) + 0.5 x (WEExtraDelay CSExtraDelay)) x GPMC_FCLK⁽¹⁴⁾
- (3) C = ((WEOnTime ADVOnTime) x (TimeParaGranularity + 1) + 0.5 x (WEExtraDelay ADVExtraDelay)) x GPMC_FCLK⁽¹⁴⁾
- (4) D = (WEOnTime × (TimeParaGranularity + 1) + 0.5 × WEExtraDelay) × GPMC_FCLK⁽¹⁴⁾
- (5) E = ((WrCycleTime WEOffTime) x (TimeParaGranularity + 1) 0.5 x WEExtraDelay) x GPMC_FCLK⁽¹⁴⁾
- (6) $F = ((ADVWrOffTime WEOffTime) \times (TimeParaGranularity + 1) + 0.5 \times (ADVExtraDelay WEExtraDelay)) \times GPMC_FCLK^{(14)}$
- (7) G = ((CSWrOffTime WEOffTime) x (TimeParaGranularity + 1) + 0.5 x (CSExtraDelay WEExtraDelay)) x GPMC_FCLK(14)
- (8) H = WrCycleTime x (1 + TimeParaGranularity) x GPMC_FCLK⁽¹⁴⁾
- (9) I = ((OEOnTime CSOnTime) × (TimeParaGranularity + 1) + 0.5 × (OEExtraDelay CSExtraDelay)) × GPMC_FCLK(14)
- (10) $K = (OEOffTime OEOnTime) \times (1 + TimeParaGranularity) \times GPMC_FCLK^{(14)}$
- (11) $L = RdCycleTime \times (1 + TimeParaGranularity) \times GPMC_FCLK^{(14)}$
- (12) M = ((CSRdOffTime OEOffTime) × (TimeParaGranularity + 1) + 0.5 × (CSExtraDelay OEExtraDelay)) × GPMC_FCLK⁽¹⁴⁾
- (13) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.
- (14) GPMC_FCLK is general-purpose memory controller internal functional clock period in ns.

www.ti.com.cn



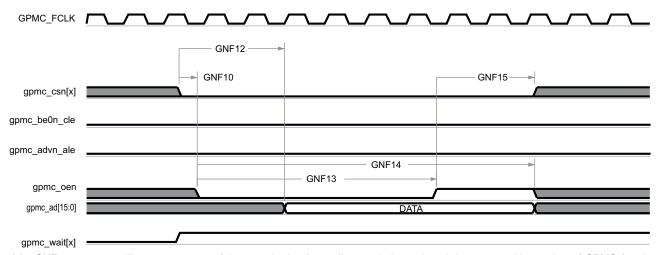
(1) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-28. GPMC and NAND Flash—Command Latch Cycle



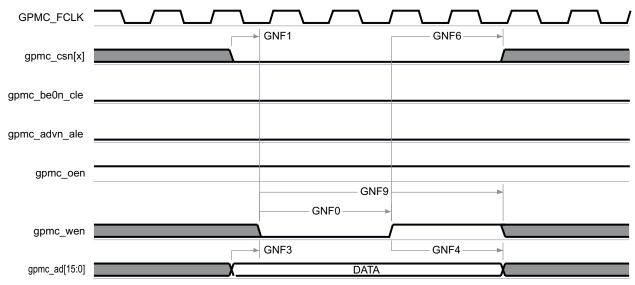
(1) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-29. GPMC and NAND Flash—Address Latch Cycle



- (1) GNF12 parameter illustrates amount of time required to internally sample input data. It is expressed in number of GPMC functional clock cycles. From start of read cycle and after GNF12 functional clock cycles, input data will be internally sampled by active functional clock edge. GNF12 value must be stored inside AccessTime register bits field.
- (2) GPMC_FCLK is an internal clock (GPMC functional clock) not provided externally.
- (3) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5. In gpmc_wait[x], x is equal to 0 or 1.

图 7-30. GPMC and NAND Flash—Data Read Cycle



(1) In gpmc_csn[x], x is equal to 0, 1, 2, 3, 4, or 5.

图 7-31. GPMC and NAND Flash—Data Write Cycle



7.7.2 mDDR(LPDDR), DDR2, DDR3, DDR3L Memory Interface

The device has a dedicated interface to mDDR(LPDDR), DDR2, DDR3, and DDR3L SDRAM. It supports JEDEC standard compliant mDDR(LPDDR), DDR2, DDR3, and DDR3L SDRAM devices with a 16-bit data path to external SDRAM memory.

For more details on the mDDR(LPDDR), DDR2, DDR3, and DDR3L memory interface, see the EMIF section of the *AM335x and AMIC110 Sitara Processors Technical Reference Manual*.

7.7.2.1 mDDR (LPDDR) Routing Guidelines

It is common to find industry references to mobile double data rate (mDDR) when discussing JEDEC defined low-power double-data rate (LPDDR) memory devices. The following guidelines use LPDDR when referencing JEDEC defined low-power double-data rate memory devices.

7.7.2.1.1 Board Designs

TI only supports board designs that follow the guidelines outlined in this document. The switching characteristics and the timing diagram for the LPDDR memory interface are shown in 表 7-31 and 图 7-32.

表 7-31. Switching Characteristics for LPDDR Memory Interface

NO.	PARAMETER		MIN	MAX	UNIT
1	$t_{c(DDR_CK)}$ $t_{c(DDR_CKn)}$	Cycle time, DDR_CK and DDR_CKn	5	(1)	ns

(1) The JEDEC JESD209B specification only defines the maximum clock period for LPDDR333 and faster speed bin LPDDR memory devices. To determine the maximum clock period, see the respective LPDDR memory data sheet.

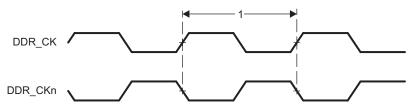


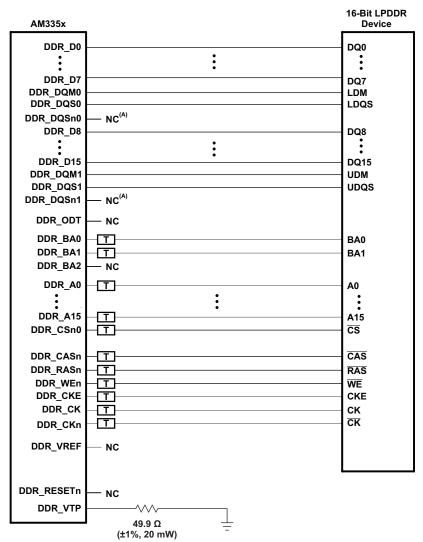
图 7-32. LPDDR Memory Interface Clock Timing

7.7.2.1.2 LPDDR Interface

This section provides the timing specification for the LPDDR interface as a PCB design and manufacturing specification. The design rules constrain PCB trace length, PCB trace skew, signal integrity, cross-talk, and signal timing. These rules, when followed, result in a reliable LPDDR memory system without the need for a complex timing closure process. For more information regarding the guidelines for using this LPDDR specification, see *Understanding TI's PCB Routing Rule-Based DDR Timing Specification*. This application report provides generic guidelines and approach. All the specifications provided in the data manual take precedence over the generic guidelines and must be adhered to for a reliable LPDDR interface operation.

7.7.2.1.2.1 LPDDR Interface Schematic

图 7-33 shows the schematic connections for 16-bit interface on the AM335x device using one x16 LPDDR device. The AM335x LPDDR memory interface only supports 16-bit-wide mode of operation. The AM335x device can only source one load connected to the DQS[x] and DQ[x] net class signals and one load connected to the CK and ADDR_CTRL net class signals. For more information related to net classes, see 节 7.7.2.1.2.8.



- A. Enable internal weak pulldown on these pins. For details, see the EMIF section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.
- B. For all the termination requirements, see 节 7.7.2.1.2.9.

图 7-33. 16-Bit LPDDR Interface Using One 16-Bit LPDDR Device



7.7.2.1.2.2 Compatible JEDEC LPDDR Devices

表 7-32 shows the parameters of the JEDEC LPDDR devices that are compatible with this interface. Generally, the LPDDR interface is compatible with x16 LPDDR400 speed grade LPDDR devices.

表 7-32. Compatible JEDEC LPDDR Devices (Per Interface)(1)

NO.	PARAMETER	MIN	MAX	UNIT
1	JEDEC LPDDR device speed grade	LPDDR400		
2	JEDEC LPDDR device bit width	x16	x16	Bits
3	JEDEC LPDDR device count		1	Devices
4	JEDEC LPDDR device terminal count		60	Terminals

⁽¹⁾ If the LPDDR interface is operated with a clock frequency less than 200 MHz, lower-speed grade LPDDR devices may be used if the minimum clock period specified for the LPDDR device is less than or equal to the minimum clock period selected for the AM335x LPDDR interface.

7.7.2.1.2.3 PCB Stackup

The minimum stackup required for routing the AM335x device is a 4-layer stackup as shown in $\frac{1}{8}$ 7-33. Additional layers may be added to the PCB stackup to accommodate other circuitry, enhance signal integrity and electromagnetic interference performance, or to reduce the size of the PCB footprint.

表 7-33. Minimum PCB Stackup(1)

LAYER	TYPE DESCRIPTION	
1	Signal	Top signal routing
2	Plane	Ground
3	Plane	Split Power Plane
4	Signal	Bottom signal routing

⁽¹⁾ All signals that have critical signal integrity requirements should be routed first on layer 1. It may not be possible to route all of these signals on layer 1, therefore requiring routing of some signals on layer 4. When this is done, the signal routes on layer 4 must not cross splits in the power plane.



Complete stackup specifications are provided in 表 7-34.

表 7-34. PCB Stackup Specifications⁽¹⁾

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	PCB routing and plane layers	4			
2	Signal routing layers	2			
3	Full ground layers under LPDDR routing region	1			
4	Number of ground plane cuts allowed within LPDDR routing region			0	
5	Full VDDS_DDR power reference layers under LPDDR routing region	1			
6	Number of layers between LPDDR routing layer and reference ground plane			0	
7	PCB routing feature size		4		mils
8	PCB trace width, w		4		mils
9	PCB BGA escape via pad size ⁽²⁾		18	20	mils
10	PCB BGA escape via hole size ⁽²⁾		10		mils
11	Single-ended impedance, Zo ⁽³⁾		50	75	Ω
12	Impedance control ⁽⁴⁾⁽⁵⁾	Zo-5	Zo	Zo+5	Ω

⁽¹⁾ For the LPDDR device BGA pad size, see the LPDDR device manufacturer documentation.

⁽²⁾ A 20-10 via may be used if enough power routing resources are available. An 18-10 via allows for more flexible power routing to the AM335x device.

⁽³⁾ Zo is the nominal singled-ended impedance selected for the PCB.

⁽⁴⁾ This parameter specifies the AC characteristic impedance tolerance for each segment of a PCB signal trace relative to the chosen Zo defined by the single-ended impedance parameter.

⁽⁵⁾ Tighter impedance control is required to ensure flight time skew is minimal.



7.7.2.1.2.4 Placement

图 7-34 shows the required placement for the LPDDR devices. The dimensions for this figure are defined in 表 7-35. The placement does not restrict the side of the PCB on which the devices are mounted. The ultimate purpose of the placement is to limit the maximum trace lengths and allow for proper routing space. For single-memory LPDDR systems, the second LPDDR device is omitted from the placement.

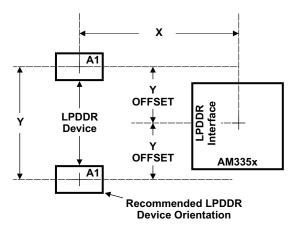


图 7-34. AM335x Device and LPDDR Device Placement

表 7-35. Placement Specifications⁽¹⁾

NO.	PARAMETER	MIN MAX	UNIT
1	X ⁽²⁾⁽³⁾	1750	mils
2	Y ⁽²⁾⁽³⁾	1280	mils
3	Y Offset ⁽²⁾⁽³⁾⁽⁴⁾	650	mils
4	Clearance from non-LPDDR signal to LPDDR keepout region ⁽⁵⁾⁽⁶⁾	4	W

- (1) LPDDR keepout region to encompass entire LPDDR routing area.
- (2) For dimension definitions, see <a>8 7-34.
- (3) Measurements from center of the AM335x device to center of LPDDR device.
- (4) For single-memory systems, TI recommends that Y offset be as small as possible.
- (5) w is defined as the signal trace width.
- (6) Non-LPDDR signals allowed within LPDDR keepout region provided they are separated from LPDDR routing layers by a ground plane.

7.7.2.1.2.5 LPDDR Keepout Region

The region of the PCB used for the LPDDR circuitry must be isolated from other signals. The LPDDR keepout region is defined for this purpose and is shown in 图 7-35. This region should encompass all LPDDR circuitry and the region size varies with component placement and LPDDR routing. Additional clearances required for the keepout region are shown in 表 7-35. Non-LPDDR signals must not be routed on the same signal layer as LPDDR signals within the LPDDR keepout region. Non-LPDDR signals may be routed in the region provided they are routed on layers separated from LPDDR signal layers by a ground layer. No breaks should be allowed in the reference ground or VDDS_DDR power plane in this region. In addition, the VDDS_DDR power plane should cover the entire keepout region.

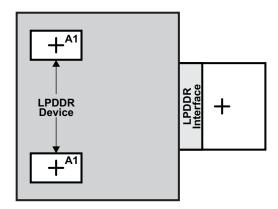


图 7-35. LPDDR Keepout Region

7.7.2.1.2.6 Bulk Bypass Capacitors

Bulk bypass capacitors are required for moderate speed bypassing of the LPDDR and other circuitry. 表 7-36 contains the minimum numbers and capacitance required for the bulk bypass capacitors. Note that this table only covers the bypass needs of the AM335x LPDDR interface and LPDDR devices. Additional bulk bypass capacitance may be needed for other circuitry.

表 7-36.	Bulk	Rypace	Cana	citore(1)
₹ /-3b.	Buik	Bybass	Caba	CITOIS

NO.	PARAMETER	MIN M	AX UNIT
1	AM335x VDDS_DDR bulk bypass capacitor count	1	Devices
2	AM335x VDDS_DDR bulk bypass total capacitance	10	μF
3	LPDDR#1 bulk bypass capacitor count	1	Devices
4	LPDDR#1 bulk bypass total capacitance	10	μF
5	LPDDR#2 bulk bypass capacitor count ⁽²⁾	1	Devices
6	LPDDR#2 bulk bypass total capacitance ⁽²⁾	10	μF

⁽¹⁾ These devices should be placed near the device they are bypassing, but preference should be given to the placement of the high-speed (HS) bypass capacitors.

⁽²⁾ Only used when two LPDDR devices are used.



7.7.2.1.2.7 High-Speed Bypass Capacitors

High-speed (HS) bypass capacitors are critical for proper LPDDR interface operation. It is particularly important to minimize the parasitic series inductance of the HS bypass capacitors, the AM335x device LPDDR power, and the AM335x device LPDDR ground connections. 表 7-37 contains the specification for the HS bypass capacitors as well as for the power connections on the PCB.

表 7-37. High-Speed Bypass Capacitors

NO.	PARAMETER	MIN	MAX	UNIT
1	HS bypass capacitor package size ⁽¹⁾		0402	10 mils
2	Distance from HS bypass capacitor to device being bypassed		250	mils
3	Number of connection vias for each HS bypass capacitor ⁽²⁾	2		Vias
4	Trace length from bypass capacitor contact to connection via		30	mils
5	Number of connection vias for each AM335x VDDS_DDR and VSS terminal	1		Vias
6	Trace length from AM335x VDDS_DDR and VSS terminal to connection via		35	mils
7	Number of connection vias for each LPDDR device power and ground terminal	1		Vias
8	Trace length from LPDDR device power and ground terminal to connection via		35	mils
9	AM335x VDDS_DDR HS bypass capacitor count ⁽³⁾	10		Devices
10	AM335x VDDS_DDR HS bypass capacitor total capacitance	0.6		μF
11	LPDDR device HS bypass capacitor count ⁽³⁾⁽⁴⁾	8		Devices
12	LPDDR device HS bypass capacitor total capacitance ⁽⁴⁾	0.4		μF

⁽¹⁾ LxW, 10-mil units; for example, a 0402 is a 40x20-mil surface-mount capacitor.

7.7.2.1.2.8 Net Classes

表 7-38 lists the clock net classes for the LPDDR interface. 表 7-39 lists the signal net classes, and associated clock net classes, for the signals in the LPDDR interface. These net classes are used for the termination and routing rules that follow.

表 7-38. Clock Net Class Definitions

CLOCK NET CLASS	AM335x PIN NAMES
CK	DDR_CK and DDR_CKn
DQS0	DDR_DQS0
DQS1	DDR_DQS1

表 7-39. Signal Net Class Definitions

SIGNAL NET CLASS	ASSOCIATED CLOCK NET CLASS	AM335x PIN NAMES
ADDR_CTRL	СК	DDR_BA[1:0], DDR_A[15:0], DDR_CSn0, DDR_CASn, DDR_RASn, DDR_WEn, DDR_CKE
DQ0	DQS0	DDR_D[7:0], DDR_DQM0
DQ1	DQS1	DDR_D[15:8], DDR_DQM1

⁽²⁾ An additional HS bypass capacitor can share the connection vias only if it is mounted on the opposite side of the board.

⁽³⁾ These devices should be placed as close as possible to the device being bypassed.

⁽⁴⁾ Per LPDDR device.



7.7.2.1.2.9 LPDDR Signal Termination

There is no specific need for adding terminations on the LPDDR interface. However, system designers may evaluate the need for serial terminators for EMI and overshoot reduction. Placement of serial terminations for DQS[x] and DQ[x] net class signals should be determined based on PCB analysis. Placement of serial terminations for ADDR_CTRL net class signals should be close to the AM335x device. 表 7-40 shows the specifications for the serial terminators in such cases.

表 7-40. LPDDR Signal Terminations

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	CK net class ⁽¹⁾	0	22	Zo ⁽²⁾	Ω
2	ADDR_CTRL net class ⁽¹⁾⁽³⁾⁽⁴⁾	0	22	Zo ⁽²⁾	Ω
3	DQS0, DQS1, DQ0, and DQ1 net classes	0	22	Zo ⁽²⁾	Ω

⁽¹⁾ Only series termination is permitted.

⁽²⁾ Zo is the LPDDR PCB trace characteristic impedance.

⁽³⁾ Series termination values larger than typical only recommended to address EMI issues.

⁽⁴⁾ Series termination values should be uniform across net class.



7.7.2.1.3 LPDDR CK and ADDR_CTRL Routing

▼ 7-36 shows the topology of the routing for the CK and ADDR_CTRL net classes. The length of signal path AB and AC should be minimized with emphasis to minimize lengths C and D such that length A is the majority of the total length of signal path AB and AC.

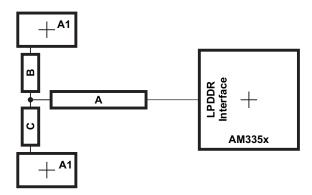


图 7-36. CK and ADDR_CTRL Routing and Topology

表 7-41. CK and ADDR_CTRL Routing Specification(1)(2)

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	Center-to-center CK spacing			2w	
2	CK differential pair skew length mismatch ⁽²⁾⁽³⁾			25	mils
3	CK B-to-CK C skew length mismatch			25	mils
4	Center-to-center CK to other LPDDR trace spacing ⁽⁴⁾	4w			
5	CK and ADDR_CTRL nominal trace length ⁽⁵⁾	CACLM-50	CACLM	CACLM+50	mils
6	ADDR_CTRL-to-CK skew length mismatch			100	mils
7	ADDR_CTRL-to-ADDR_CTRL skew length mismatch			100	mils
8	Center-to-center ADDR_CTRL to other LPDDR trace spacing ⁽⁴⁾	4w			
9	Center-to-center ADDR_CTRL to other ADDR_CTRL trace spacing ⁽⁴⁾	3w			
10	ADDR_CTRL A-to-B and ADDR_CTRL A-to-C skew length mismatch ⁽²⁾			100	mils
11	ADDR_CTRL B-to-C skew length mismatch			100	mils

⁽¹⁾ CK represents the clock net class, and ADDR_CTRL represents the address and control signal net class.

(5) CACLM is the longest Manhattan distance of the CK and ADDR_CTRL net classes.

⁽²⁾ Series terminator, if used, should be located closest to the AM335x device.

⁽³⁾ Differential impedance should be Zo x 2, where Zo is the single-ended impedance defined in $\frac{1}{8}$ 7-34.

⁽⁴⁾ Center-to-center spacing is allowed to fall to minimum (w) for up to 500 mils of routed length to accommodate BGA escape and routing congestion.

www.ti.com.cn

🛚 7-37 shows the topology and routing for the DQS[x] and DQ[x] net classes; the routes are point to point. Skew matching across bytes is not needed nor recommended.

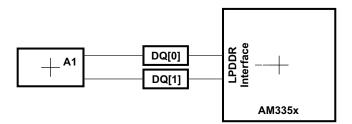


图 7-37. DQS[x] and DQ[x] Routing and Topology

表 7-42. DQS[x] and DQ[x] Routing Specification⁽¹⁾

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	Center-to-center DQS[x] spacing			2w	
2	Center-to-center DDR_DQS[x] to other LPDDR trace spacing ⁽²⁾	4w			
3	DQS[x] and DQ[x] nominal trace length ⁽³⁾	DQLM-50	DQLM	DQLM+50	mils
4	DQ[x]-to-DQS[x] skew length mismatch ⁽³⁾			100	mils
5	DQ[x]-to-DQ[x] skew length mismatch ⁽³⁾			100	mils
6	Center-to-center DQ[x] to other LPDDR trace spacing ⁽²⁾⁽⁴⁾	4w			
7	Center-to-center DQ[x] to other DQ[x] trace spacing ⁽²⁾⁽⁵⁾	3w			

- (1) DQS[x] represents the DQS0 and DQS1 clock net classes, and DQ[x] represents the DQ0 and DQ1 signal net classes.
- (2) Center-to-center spacing is allowed to fall to minimum (w) for up to 500 mils of routed length to accommodate BGA escape and routing congestion.
- (3) There is no requirement for skew matching between data bytes; that is, from net classes DQS0 and DQ0 to net classes DQS1 and DQ1.
- (4) Signals from one DQ net class should be considered other LPDDR traces to another DQ net class.
- (5) DQLM is the longest Manhattan distance of each of the DQS[x] and DQ[x] net classes.



7.7.2.2 DDR2 Routing Guidelines

7.7.2.2.1 Board Designs

TI only supports board designs that follow the guidelines outlined in this document. 表 7-43 and 图 7-38 show the switching characteristics and timing diagram for the DDR2 memory interface.

表 7-43. Switching Characteristics for DDR2 Memory Interface

NO.		PARAMETER	MIN	MAX	UNIT
1	$t_{c(DDR_CK)}$ $t_{c(DDR_CKn)}$	Cycle time, DDR_CK and DDR_CKn	3.75	8 ⁽¹⁾	ns

(1) The JEDEC JESD79-2F specification defines the maximum clock period of 8 ns for all standard-speed bin DDR2 memory devices. Therefore, all standard-speed bin DDR2 memory devices are required to operate at 125 MHz.

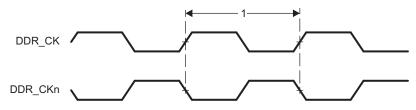


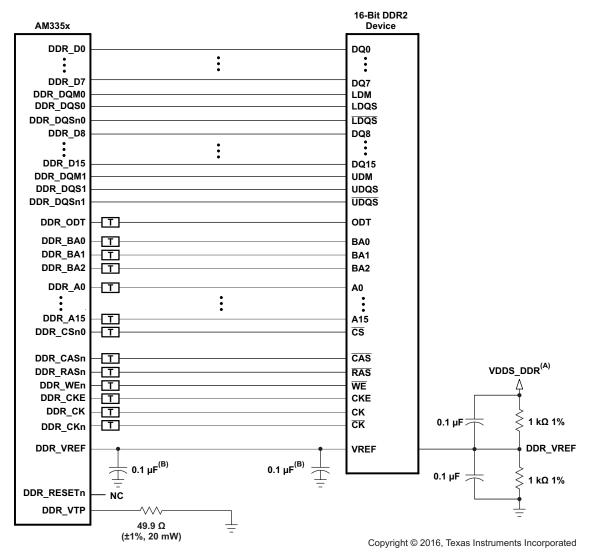
图 7-38. DDR2 Memory Interface Clock Timing

7.7.2.2.2 DDR2 Interface

This section provides the timing specification for the DDR2 interface as a PCB design and manufacturing specification. The design rules constrain PCB trace length, PCB trace skew, signal integrity, cross-talk, and signal timing. These rules, when followed, result in a reliable DDR2 memory system without the need for a complex timing closure process. For more information regarding the guidelines for using this DDR2 specification, see *Understanding TI's PCB Routing Rule-Based DDR Timing Specification*. This application report provides generic guidelines and approach. All the specifications provided in the data manual take precedence over the generic guidelines and must be adhered to for a reliable DDR2 interface operation.

7.7.2.2.2.1 DDR2 Interface Schematic

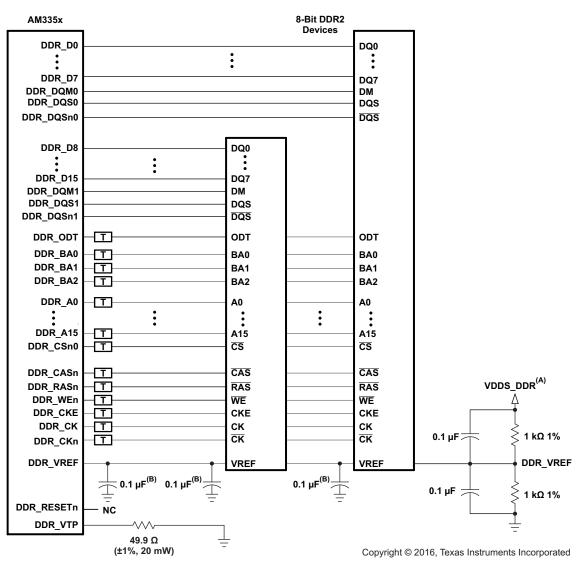
图 7-39 shows the schematic connections for 16-bit interface on the AM335x device using one x16 DDR2 device and 图 7-40 shows the schematic connections for 16-bit interface on the AM335x device using two x8 DDR2 devices. The AM335x DDR2 memory interface only supports 16-bit-wide mode of operation. The AM335x device can only source one load connected to the DQS[x] and DQ[x] net class signals and two loads connected to the CK and ADDR_CTRL net class signals. For more information related to net classes, see 节 7.7.2.2.2.8.



- A. VDDS_DDR is the power supply for the DDR2 memories and the AM335x DDR2 interface.
- B. One of these capacitors can be eliminated if the divider and its capacitors are placed near a DDR_VREF pin.
- C. For all the termination requirements, see 节 7.7.2.2.2.9.

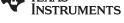
图 7-39. 16-Bit DDR2 Interface Using One 16-Bit DDR2 Device

STRUMENTS



- VDDS_DDR is the power supply for the DDR2 memories and the AM335x DDR2 interface.
- One of these capacitors can be eliminated if the divider and its capacitors are placed near a DDR_VREF pin.
- For all the termination requirements, see 节 7.7.2.2.2.9.

图 7-40. 16-Bit DDR2 Interface Using Two 8-Bit DDR2 Devices



7.7.2.2.2.2 Compatible JEDEC DDR2 Devices

表 7-44 shows the parameters of the JEDEC DDR2 devices that are compatible with this interface. Generally, the DDR2 interface is compatible with x16 or x8 DDR2-533 speed grade DDR2 devices.

表 7-44. Compatible JEDEC DDR2 Devices (Per Interface)(1)

NO.	PARAMETER	MIN	MAX	UNIT
1	JEDEC DDR2 device speed grade ⁽²⁾	DDR2-533		
2	JEDEC DDR2 device bit width	x8	x16	bits
3	JEDEC DDR2 device count	1	2	devices
4	JEDEC DDR2 device terminal count ⁽³⁾	60	84	terminals

⁽¹⁾ If the DDR2 interface is operated with a clock frequency less than 266 MHz, lower-speed grade DDR2 devices may be used if the minimum clock period specified for the DDR2 device is less than or equal to the minimum clock period selected for the AM335x DDR2 interface.

7.7.2.2.2.3 PCB Stackup

The minimum stackup required for routing the AM335x device is a 4-layer stackup as shown in 表 7-45. Additional layers may be added to the PCB stackup to accommodate other circuitry, enhance signal integrity and electromagnetic interference performance, or to reduce the size of the PCB footprint.

表 7-45. Minimum PCB Stackup(1)

LAYER	TYPE DESCRIPTION	
1	1 Signal Top signal routing	
2 Plane Ground		Ground
3	Plane Split power plane	
4	4 Signal Bottom signal routing	

(1) All signals that have critical signal integrity requirements should be routed first on layer 1. It may not be possible to route all of these signals on layer 1, therefore requiring routing of some signals on layer 4. When this is done, the signal routes on layer 4 must not cross splits in the power plane.

⁽²⁾ Higher DDR2 speed grades are supported due to inherent JEDEC DDR2 backward compatibility.

^{(3) 92-}terminal devices are also supported for legacy reasons. New designs will migrate to 84-terminal DDR2 devices. Electrically, the 92and 84-terminal DDR2 devices are the same.



Complete stackup specifications are provided in 表 7-46.

表 7-46. PCB Stackup Specifications⁽¹⁾

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	PCB routing and plane layers	4			
2	Signal routing layers	2			
3	Full ground layers under DDR2 routing region	1			
4	Number of ground plane cuts allowed within DDR2 routing region			0	
5	Full VDDS_DDR power reference layers under DDR2 routing region	1			
6	Number of layers between DDR2 routing layer and reference ground plane			0	
7	PCB routing feature size		4		mils
8	PCB trace width, w		4		mils
9	PCB BGA escape via pad size ⁽²⁾		18	20	mils
10	PCB BGA escape via hole size ⁽²⁾		10		mils
11	Single-ended impedance, Zo ⁽³⁾		50	75	Ω
12	Impedance control ⁽⁴⁾⁽⁵⁾	Zo-5	Zo	Zo+5	Ω

- (1) For the DDR2 device BGA pad size, see the DDR2 device manufacturer documentation.
- (2) A 20-10 via may be used if enough power routing resources are available. An 18-10 via allows for more flexible power routing to the AM335x device.
- (3) Zo is the nominal singled-ended impedance selected for the PCB.
- (4) This parameter specifies the AC characteristic impedance tolerance for each segment of a PCB signal trace relative to the chosen Zo defined by the single-ended impedance parameter.
- (5) Tighter impedance control is required to ensure flight time skew is minimal.

7.7.2.2.2.4 Placement

图 7-41 shows the required placement for the DDR2 devices. The dimensions for this figure are defined in 表 7-47. The placement does not restrict the side of the PCB on which the devices are mounted. The ultimate purpose of the placement is to limit the maximum trace lengths and allow for proper routing space. For single-memory DDR2 systems, the second DDR2 device is omitted from the placement.

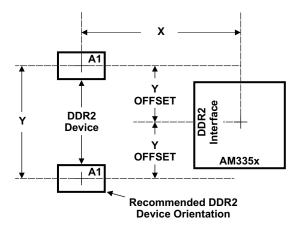


图 7-41. AM335x Device and DDR2 Device Placement

表 7-47. Placement Specifications⁽¹⁾

NO.	PARAMETER	MIN MAX	UNIT
1	χ ⁽²⁾⁽³⁾	1750	mils
2	Y ⁽²⁾⁽³⁾	1280	mils
3	Y Offset ⁽²⁾⁽³⁾⁽⁴⁾	650	mils
4	Clearance from non-DDR2 signal to DDR2 keepout region ⁽⁵⁾⁽⁶⁾	4	w

- (1) DDR2 keepout region to encompass entire DDR2 routing area.
- (2) For dimension definitions, see <a>8 7-41.
- (3) Measurements from center of the AM335x device to center of the DDR2 device.
- (4) For single-memory systems, it is recommended that Y offset be as small as possible.
- (5) w is defined as the signal trace width.
- (6) Non-DDR2 signals allowed within DDR2 keepout region provided they are separated from DDR2 routing layers by a ground plane.

7.7.2.2.2.5 DDR2 Keepout Region

The region of the PCB used for the DDR2 circuitry must be isolated from other signals. The DDR2 keepout region is defined for this purpose and is shown in 图 7-42. This region should encompass all DDR2 circuitry and the region size varies with component placement and DDR2 routing. Additional clearances required for the keepout region are shown in 表 7-47. Non-DDR2 signals must not be routed on the same signal layer as DDR2 signals within the DDR2 keepout region. Non-DDR2 signals may be routed in the region provided they are routed on layers separated from DDR2 signal layers by a ground layer. No breaks should be allowed in the reference ground or VDDS_DDR power plane in this region. In addition, the VDDS_DDR power plane should cover the entire keepout region.

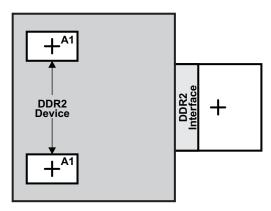


图 7-42. DDR2 Keepout Region

7.7.2.2.2.6 Bulk Bypass Capacitors

Bulk bypass capacitors are required for moderate speed bypassing of the DDR2 and other circuitry. 表 7-48 contains the minimum numbers and capacitance required for the bulk bypass capacitors. Note that this table only covers the bypass needs of the AM335x DDR2 interface and DDR2 devices. Additional bulk bypass capacitance may be needed for other circuitry.

表 7-48. Bulk Bypass Ca	pacitors	
------------------------	----------	--

NO.	PARAMETER	MIN	MAX	UNIT
1	AM335x VDDS_DDR bulk bypass capacitor count	1		devices
2	AM335x VDDS_DDR bulk bypass total capacitance	10		μF
3	DDR2 number 1 bulk bypass capacitor count	1		devices
4	DDR2 number 1 bulk bypass total capacitance	10		μF
5	DDR2 number 2 bulk bypass capacitor count ⁽²⁾	1		devices
6	DDR2 number 2 bulk bypass total capacitance ⁽²⁾	10		μF

⁽¹⁾ These devices should be placed near the device they are bypassing, but preference should be given to the placement of the high-speed (HS) bypass capacitors.

⁽²⁾ Only used when two DDR2 devices are used.



7.7.2.2.2.7 High-Speed (HS) Bypass Capacitors

HS bypass capacitors are critical for proper DDR2 interface operation. It is particularly important to minimize the parasitic series inductance of the HS bypass capacitors, the AM335x device DDR2 power, and the AM335x device DDR2 ground connections. 表 7-49 contains the specification for the HS bypass capacitors as well as for the power connections on the PCB.

表 7-49. HS Bypass Capacitors

NO.	PARAMETER	MIN	MAX	UNIT
1	HS bypass capacitor package size ⁽¹⁾		0402	10 mils
2	Distance from HS bypass capacitor to device being bypassed		250	mils
3	Number of connection vias for each HS bypass capacitor ⁽²⁾	2		vias
4	Trace length from bypass capacitor contact to connection via		30	mils
5	Number of connection vias for each AM335x VDDS_DDR and VSS terminal	1		vias
6	Trace length from AM335x VDDS_DDR and VSS terminal to connection via		35	mils
7	Number of connection vias for each DDR2 device power and ground terminal	1		vias
8	Trace length from DDR2 device power and ground terminal to connection via		35	mils
9	AM335x VDDS_DDR HS bypass capacitor count ⁽³⁾	10		devices
10	AM335x VDDS_DDR HS bypass capacitor total capacitance	0.6		μF
11	DDR2 device HS bypass capacitor count ⁽³⁾⁽⁴⁾	8		devices
12	DDR2 device HS bypass capacitor total capacitance ⁽⁴⁾	0.4		μF

⁽¹⁾ LxW, 10-mil units; for example, a 0402 is a 40x20-mil surface-mount capacitor.

7.7.2.2.2.8 Net Classes

 $\bar{\chi}$ 7-50 lists the clock net classes for the DDR2 interface. $\bar{\chi}$ 7-51 lists the signal net classes, and associated clock net classes, for the signals in the DDR2 interface. These net classes are used for the termination and routing rules that follow.

表 7-50. Clock Net Class Definitions

CLOCK NET CLASS	AM335x PIN NAMES
CK	DDR_CK and DDR_CKn
DQS0	DDR_DQS0 and DDR_DQSn0
DQS1	DDR_DQS1 and DDR_DQSn1

表 7-51. Signal Net Class Definitions

SIGNAL NET CLASS	ASSOCIATED CLOCK NET CLASS	AM335x PIN NAMES
ADDR_CTRL	СК	DDR_BA[2:0], DDR_A[15:0], DDR_CSn0, DDR_CASn, DDR_RASn, DDR_WEn, DDR_CKE, DDR_ODT
DQ0	DQS0	DDR_D[7:0], DDR_DQM0
DQ1	DQS1	DDR_D[15:8], DDR_DQM1

⁽²⁾ An additional HS bypass capacitor can share the connection vias only if it is mounted on the opposite side of the board.

⁽³⁾ These devices should be placed as close as possible to the device being bypassed.

⁽⁴⁾ Per DDR2 device.



7.7.2.2.2.9 DDR2 Signal Termination

Signal terminations are required on the CK and ADDR_CTRL net class signals. Serial terminations should be used on the CK and ADDR_CTRL lines and is the preferred termination scheme. On-device terminations (ODTs) are required on the DQS[x] and DQ[x] net class signals. They should be enabled to ensure signal integrity. $\frac{1}{8}$ 7-52 shows the specifications for the series terminators. Placement of serial terminations for ADDR_CTRL net class signals should be close to the AM335x device.

表 7-52. DDR2 Signal Terminations

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	CK net class ⁽¹⁾	0		10	Ω
2	ADDR_CTRL net class ⁽¹⁾⁽²⁾⁽³⁾	0	22	Zo ⁽⁴⁾	Ω
3	DQS0, DQS1, DQ0, and DQ1 net classes ⁽⁵⁾	N/A		N/A	Ω

- (1) Only series termination is permitted.
- (2) Series termination values larger than typical only recommended to address EMI issues.
- (3) Series termination values should be uniform across net class.
- (4) Zo is the DDR2 PCB trace characteristic impedance.
- (5) No external termination resistors are allowed and ODT must be used for these net classes.

If the DDR2 interface is operated at a lower frequency (<200-MHz clock rate), on-device terminations are not specifically required for the DQS[x] and DQ[x] net class signals and serial terminations for the CK and ADDR_CTRL net class signals are not mandatory. System designers may evaluate the need for serial terminators for EMI and overshoot reduction. Placement of serial terminations for DQS[x] and DQ[x] net class signals should be determined based on PCB analysis. Placement of serial terminations for ADDR_CTRL net class signals should be close to the AM335x device. 表 7-53 shows the specifications for the serial terminators in such cases.

表 7-53. Lower-Frequency DDR2 Signal Terminations

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	CK net class ⁽¹⁾	0	22	Zo ⁽²⁾	Ω
2	ADDR_CTRL net class ⁽¹⁾⁽³⁾⁽⁴⁾	0	22	Zo ⁽²⁾	Ω
3	DQS0, DQS1, DQ0, and DQ1 net classes	0	22	Zo ⁽²⁾	Ω

- (1) Only series termination is permitted.
- (2) Zo is the DDR2 PCB trace characteristic impedance.
- (3) Series termination values larger than typical only recommended to address EMI issues.
- (4) Series termination values should be uniform across net class.



7.7.2.2.2.10 DDR VREF Routing

DDR_VREF is used as a reference by the input buffers of the DDR2 memories as well as the AM335x device. DDR_VREF is intended to be half the DDR2 power supply voltage and should be created using a resistive divider as shown in 图 7-39 and 图 7-40. TI does not recommend other methods of creating DDR_VREF. 图 7-43 shows the layout guidelines for DDR_VREF.

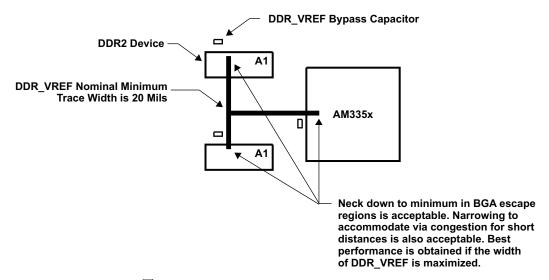


图 7-43. DDR_VREF Routing and Topology



7.7.2.2.3 DDR2 CK and ADDR_CTRL Routing

▼ 7-44 shows the topology of the routing for the CK and ADDR_CTRL net classes. The length of signal path AB and AC should be minimized with emphasis to minimize lengths C and D such that length A is the majority of the total length of signal path AB and AC.

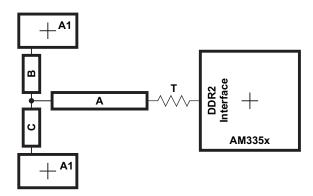


图 7-44. CK and ADDR_CTRL Routing and Topology

表 7-54. CK and ADDR_CTRL Routing Specification⁽¹⁾⁽²⁾

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	Center-to-center CK spacing			2w	
2	CK differential pair skew length mismatch ⁽²⁾⁽³⁾			25	mils
3	CK B-to-CK C skew length mismatch			25	mils
4	Center-to-center CK to other DDR2 trace spacing ⁽⁴⁾	4w			
5	CK and ADDR_CTRL nominal trace length ⁽⁵⁾	CACLM-50	CACLM	CACLM+50	mils
6	ADDR_CTRL-to-CK skew length mismatch			100	mils
7	ADDR_CTRL-to-ADDR_CTRL skew length mismatch			100	mils
8	Center-to-center ADDR_CTRL to other DDR2 trace spacing ⁽⁴⁾	4w			
9	Center-to-center ADDR_CTRL to other ADDR_CTRL trace spacing ⁽⁴⁾	3w			
10	ADDR_CTRL A-to-B and ADDR_CTRL A-to-C skew length mismatch ⁽²⁾			100	mils
11	ADDR_CTRL B-to-C skew length mismatch			100	mils

- (1) CK represents the clock net class, and ADDR_CTRL represents the address and control signal net class.
- (2) Series terminator, if used, should be located closest to the AM335x device.
- (3) Differential impedance should be Zo x 2, where Zo is the single-ended impedance defined in 表 7-46.
- (4) Center-to-center spacing is allowed to fall to minimum (w) for up to 500 mils of routed length to accommodate BGA escape and routing congestion.
- (5) CACLM is the longest Manhattan distance of the CK and ADDR_CTRL net classes.

§ 7-45 shows the topology and routing for the DQS[x] and DQ[x] net classes; the routes are point to point. Skew matching across bytes is not needed nor recommended.

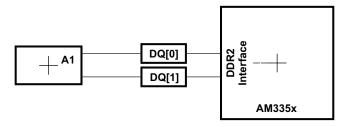


图 7-45. DQS[x] and DQ[x] Routing and Topology



表 7-55. DQS[x] and DQ[x] Routing Specification⁽¹⁾

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	Center-to-center DQS[x] spacing			2w	
2	DQS[x] differential pair skew length mismatch ⁽²⁾			25	mils
3	Center-to-center DDR_DQS[x] to other DDR2 trace spacing ⁽³⁾	4w			
4	DQS[x] and DQ[x] nominal trace length ⁽⁴⁾	DQLM-50	DQLM	DQLM+50	mils
5	DQ[x]-to-DQS[x] skew length mismatch ⁽⁴⁾			100	mils
6	DQ[x]-to-DQ[x] skew length mismatch ⁽⁴⁾			100	mils
7	Center-to-center DQ[x] to other DDR2 trace spacing ⁽³⁾⁽⁵⁾	4w			
8	Center-to-center DQ[x] to other DQ[x] trace spacing ⁽³⁾⁽⁶⁾	3w			

- (1) DQS[x] represents the DQS0 and DQS1 clock net classes, and DQ[x] represents the DQ0 and DQ1 signal net classes.
- (2) Differential impedance should be Zo x 2, where Zo is the single-ended impedance defined in 表 7-46.
- (3) Center-to-center spacing is allowed to fall to minimum (w) for up to 500 mils of routed length to accommodate BGA escape and routing congestion.
- (4) There is no requirement for skew matching between data bytes; that is, from net classes DQS0 and DQ0 to net classes DQS1 and DQ1.
- (5) Signals from one DQ net class should be considered other DDR2 traces to another DQ net class.
- (6) DQLM is the longest Manhattan distance of each of the DQS[x] and DQ[x] net classes.



7.7.2.3 DDR3 and DDR3L Routing Guidelines

注

All references to DDR3 in this section apply to DDR3 and DDR3L devices, unless otherwise noted.

7.7.2.3.1 Board Designs

TI only supports board designs using DDR3 memory that follow the guidelines in this document. The switching characteristics and timing diagram for the DDR3 memory interface are shown in 表 7-56 and 图 7-46.

表 7-56. Switching Characteristics for DDR3 Memory Interface

NO.	PARAMETER		MIN	MAX	UNIT
1	$t_{c(DDR_CK)} \ t_{c(DDR_CKn)}$	Cycle time, DDR_CK and DDR_CKn	2.5	3.3 ⁽¹⁾	ns

(1) The JEDEC JESD79-3F Standard defines the maximum clock period of 3.3 ns for all standard-speed bin DDR3 and DDR3L memory devices. Therefore, all standard-speed bin DDR3 and DDR3L memory devices are required to operate at 303 MHz.

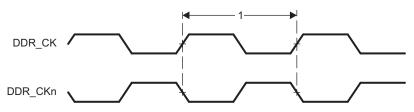


图 7-46. DDR3 Memory Interface Clock Timing

7.7.2.3.1.1 DDR3 versus DDR2

This specification only covers AM335x PCB designs that use DDR3 memory. Designs using DDR2 memory should use the DDR2 routing guidleines described in † 7.7.2.2. While similar, the two memory systems have different requirements. It is currently not possible to design one PCB that meets the requirements of both DDR2 and DDR3.

7.7.2.3.2 DDR3 Device Combinations

Because there are several possible combinations of device counts and single-side or dual-side mounting, 表 7-57 summarizes the supported device configurations.

表 7-57. Supported DDR3 Device Combinations

NUMBER OF DDR3 DEVICES	DDR3 DEVICE WIDTH (BITS)	MIRRORED?	DDR3 EMIF WIDTH (BITS)
1	16	N	16
2	8	Y ⁽¹⁾	16

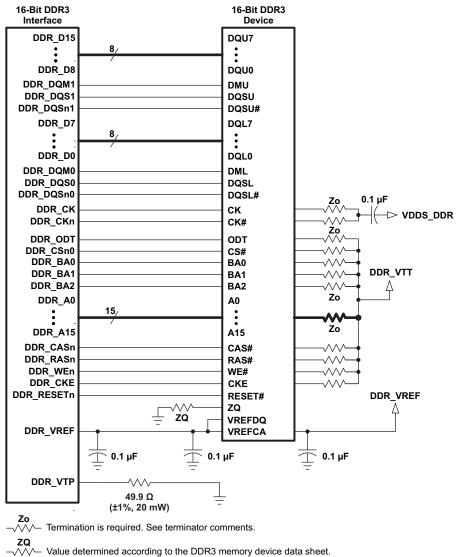
⁽¹⁾ Two DDR3 devices are mirrored when one device is placed on the top of the board and the second device is placed on the bottom of the board.

7.7.2.3.3 DDR3 Interface

This section provides the timing specification for the DDR3 interface as a PCB design and manufacturing specification. The design rules constrain PCB trace length, PCB trace skew, signal integrity, cross-talk, and signal timing. These rules, when followed, result in a reliable DDR3 memory system without the need for a complex timing closure process. For more information regarding the guidelines for using this DDR3 specification, see *Understanding TI's PCB Routing Rule-Based DDR Timing Specification*. This application report provides generic guidelines and approach. All the specifications provided in the data manual take precedence over the generic guidelines and must be adhered to for a reliable DDR3 interface operation.

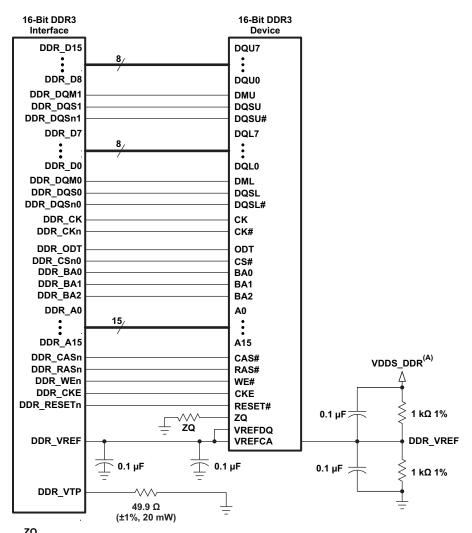
7.7.2.3.3.1 DDR3 Interface Schematic

The DDR3 interface schematic varies, depending upon the width of the DDR3 devices used. \(\begin{align*} \frac{7-47}{2} \end{align*} \) shows the schematic connections for 16-bit interface on the AM335x device using one x16 DDR3 device and 图 7-49 shows the schematic connections for 16-bit interface on the AM335x device using two x8 DDR3 devices. The AM335x DDR3 memory interface only supports 16-bit wide mode of operation. The AM335x device can only source one load connected to the DQS[x] and DQ[x] net class signals and two loads connected to the CK and ADDR_CTRL net class signals. For more information related to net classes, see † 7.7.2.3.3.8.



Copyright © 2016, Texas Instruments Incorporated

图 7-47. 16-Bit DDR3 Interface Using One 16-Bit DDR3 Device with $V_{ extsf{TT}}$ Termination



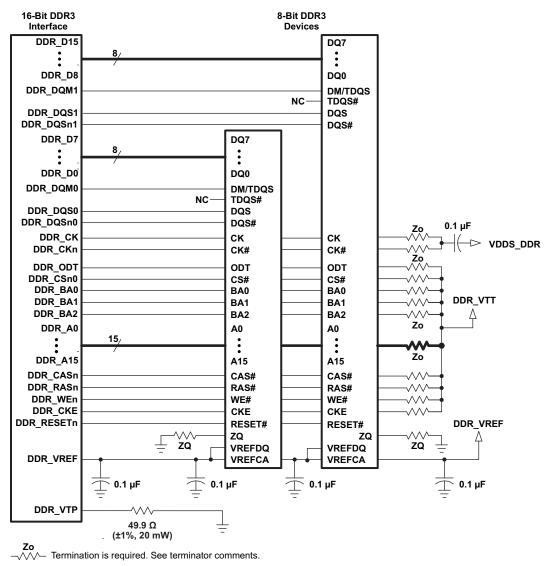
ZQ—////— Value determined according to the DDR3 memory device data sheet.

Copyright © 2016, Texas Instruments Incorporated

A. VDDS_DDR is the power supply for the DDR3 memories and the AM335x DDR3 interface.

图 7-48. 16-Bit DDR3 Interface Using One 16-Bit DDR3 Device without V_{TT} Termination





ZQ—\/\/\— Value determined according to the DDR3 memory device data sheet.

Copyright © 2016, Texas Instruments Incorporated

图 7-49. 16-Bit DDR3 Interface Using Two 8-Bit DDR3 Devices



7.7.2.3.3.2 Compatible JEDEC DDR3 Devices

表 7-58 shows the parameters of the JEDEC DDR3 devices that are compatible with this interface.

表 7-58. Compatible JEDEC DDR3 Devices (Per Interface)

NO.	PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
4	IEDEC DDD2 dovice aread aread	$t_{C(DDR_CK)}$ and $t_{C(DDR_CKn)}$ = 3.3 ns	DDR3-800		
1	JEDEC DDR3 device speed grade	$t_{C(DDR_CK)}$ and $t_{C(DDR_CKn)}$ = 2.5 ns	DDR3-1600		
2	JEDEC DDR3 device bit width		x8	x16	bits
3	JEDEC DDR3 device count ⁽¹⁾		1	2	devices

⁽¹⁾ For valid DDR3 device configurations and device counts, see 节 7.7.2.3.3.1, 图 7-47, and 图 7-49.

7.7.2.3.3.3 PCB Stackup

The minimum stackup for routing the DDR3 interface is a four-layer stack up as shown in 表 7-59. Additional layers may be added to the PCB stackup to accommodate other circuitry, enhance signal integrity and electromagnetic interference performance, or to reduce the size of the PCB footprint.

表 7-59. Minimum PCB Stackup⁽¹⁾

LAYER	TYPE	DESCRIPTION
1	Signal	Top signal routing
2	Plane	Ground
3	Plane	Split Power Plane
4	Signal	Bottom signal routing

⁽¹⁾ All signals that have critical signal integrity requirements should be routed first on layer 1. It may not be possible to route all of these signals on layer 1, therefore requiring routing of some signals on layer 4. When this is done, the signal routes on layer 4 must not cross splits in the power plane.



表 7-60. PCB Stackup Specifications(1)

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	PCB routing and plane layers	4			
2	Signal routing layers	2			
3	Full ground reference layers under DDR3 routing region ⁽²⁾	1			
4	Full VDDS_DDR power reference layers under the DDR3 routing region ⁽²⁾	1			
5	Number of reference plane cuts allowed within DDR3 routing region ⁽³⁾			0	
6	Number of layers between DDR3 routing layer and reference plane ⁽⁴⁾			0	
7	PCB routing feature size		4		mils
8	PCB trace width, w		4		mils
9	PCB BGA escape via pad size ⁽⁵⁾		18	20	mils
10	PCB BGA escape via hole size		10		mils
11	Single-ended impedance, Zo ⁽⁶⁾		50	75	Ω
12	Impedance control ⁽⁷⁾⁽⁸⁾	Zo-5	Zo	Zo+5	Ω

- (1) For the DDR3 device BGA pad size, see the DDR3 device manufacturer documentation.
- (2) Ground reference layers are preferred over power reference layers. Be sure to include bypass capacitors to accommodate reference layer return current as the trace routes switch routing layers.
- (3) No traces should cross reference plane cuts within the DDR3 routing region. High-speed signal traces crossing reference plane cuts create large return current paths which can lead to excessive crosstalk and EMI radiation.
- (4) Reference planes are to be directly adjacent to the signal plane to minimize the size of the return current loop.
- (5) An 18-mil pad assumes Via Channel is the most economical BGA escape. A 20-mil pad may be used if additional layers are available for power routing. An 18-mil pad is required for minimum layer count escape.
- (6) Zo is the nominal singled-ended impedance selected for the PCB.
- (7) This parameter specifies the AC characteristic impedance tolerance for each segment of a PCB signal trace relative to the chosen Zo defined by the single-ended impedance parameter.
- (8) Tighter impedance control is required to ensure flight time skew is minimal.



7.7.2.3.3.4 Placement

8 7-50 shows the required placement for the AM335x device as well as the DDR3 devices. The dimensions for this figure are defined in $$\frac{1}{8}$$ 7-61. The placement does not restrict the side of the PCB on which the devices are mounted. The ultimate purpose of the placement is to limit the maximum trace lengths and allow for proper routing space.

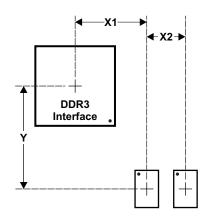


图 7-50. Placement Specifications

表 7-61. Placement Specifications⁽¹⁾

NO.	PARAMETER	MIN MAX	UNIT
1	X1 ⁽²⁾⁽³⁾⁽⁴⁾	1000	mils
2	X2 ⁽²⁾⁽³⁾	600	mils
3	Y Offset ⁽²⁾⁽³⁾⁽⁴⁾	1500	mils
4	Clearance from non-DDR3 signal to DDR3 keepout region ⁽⁵⁾⁽⁶⁾	4	W

- (1) DDR3 keepout region to encompass entire DDR3 routing area.
- (2) For dimension definitions, see 2 7-50.
- (3) Measurements from center of the AM335x device to center of the DDR3 device.
- (4) Minimizing X1 and Y improves timing margins.
- (5) w is defined as the signal trace width.
- (6) Non-DDR3 signals allowed within DDR3 keepout region provided they are separated from DDR3 routing layers by a ground plane.

7.7.2.3.3.5 DDR3 Keepout Region

The region of the PCB used for DDR3 circuitry must be isolated from other signals. The DDR3 keepout region is defined for this purpose and is shown in 图 7-51. This region should encompass all DDR3 circuitry and the region size varies with component placement and DDR3 routing. Additional clearances required for the keepout region are shown in 表 7-61. Non-DDR3 signals must not be routed on the same signal layer as DDR3 signals within the DDR3 keepout region. Non-DDR3 signals may be routed in the region provided they are routed on layers separated from DDR3 signal layers by a ground layer. No breaks should be allowed in the reference ground or VDDS_DDR power plane in this region. In addition, the VDDS_DDR power plane should cover the entire keepout region.

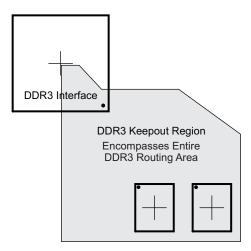


图 7-51. DDR3 Keepout Region

7.7.2.3.3.6 Bulk Bypass Capacitors

Bulk bypass capacitors are required for moderate speed bypassing of the DDR3 and other circuitry. 表 7-62 contains the minimum numbers and capacitance required for the bulk bypass capacitors. Note that this table only covers the bypass needs of the AM335x DDR3 interface and DDR3 devices. Additional bulk bypass capacitance may be needed for other circuitry.

NO.	PARAMETER	MIN	MAX	UNIT
1	AM335x VDDS_DDR bulk bypass capacitor count	2		devices
2	AM335x VDDS_DDR bulk bypass total capacitance	20		μF
3	DDR3 number 1 bulk bypass capacitor count	2		devices
4	DDR3 number 1 bulk bypass total capacitance 20		μF	
5	DDR3 number 2 bulk bypass capacitor count ⁽²⁾	2		devices
6	DDR3 number 2 bulk bypass total capacitance ⁽²⁾	20		μF

表 7-62. Bulk Bypass Capacitors⁽¹⁾

(2) Only used when two DDR3 devices are used.

⁽¹⁾ These devices should be placed near the devices they are bypassing, but preference should be given to the placement of the high-speed (HS) bypass capacitors and DDR3 signal routing.



7.7.2.3.3.7 High-Speed Bypass Capacitors

High-speed (HS) bypass capacitors are critical for proper DDR3 interface operation. It is particularly important to minimize the parasitic series inductance of the HS bypass capacitors, the AM335x device DDR3 power, and the AM335x device DDR3 ground connections. 表 7-63 contains the specification for the HS bypass capacitors as well as for the power connections on the PCB. Generally speaking, it is good to:

- Fit as many HS bypass capacitors as possible.
- Minimize the distance from the bypass capacitor to the power terminals being bypassed.
- Use the smallest physical sized capacitors possible with the highest capacitance readily available.
- Connect the bypass capacitor pads to their vias using the widest traces possible and using the largest hole size via possible.
- Minimize via sharing. Note the limits on via sharing shown in 表 7-63.

表 7-63. High-Speed Bypass Capacitors

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	HS bypass capacitor package size ⁽¹⁾	0201 0402		0402	10 mils
2	Distance, HS bypass capacitor to AM335x VDDS_DDR and VSS terminal being bypassed $^{(2)(3)(4)}$			400	mils
3	AM335x VDDS_DDR HS bypass capacitor count	20			devices
4	AM335x VDDS_DDR HS bypass capacitor total capacitance	1			μF
5	Trace length from AM335x VDDS_DDR and VSS terminal to connection via (2)		35	70	mils
6	Distance, HS bypass capacitor to DDR3 device being bypassed ⁽⁵⁾			150	mils
7	DDR3 device HS bypass capacitor count ⁽⁶⁾	12			devices
8	DDR3 device HS bypass capacitor total capacitance ⁽⁶⁾	0.85			μF
9	Number of connection vias for each HS bypass capacitor ⁽⁷⁾⁽⁸⁾	2			vias
10	Trace length from bypass capacitor connect to connection via ⁽²⁾⁽⁸⁾		35	100	mils
11	Number of connection vias for each DDR3 device power and ground terminal (9)	1		_	vias
12	Trace length from DDR3 device power and ground terminal to connection via (2)(7)		35	60	mils

- (1) LxW, 10-mil units; for example, a 0402 is a 40x20-mil surface-mount capacitor.
- (2) Closer and shorter is better.
- (3) Measured from the nearest AM335x VDDS_DDR and ground terminal to the center of the capacitor package.
- (4) Three of these capacitors should be located underneath the AM335x device, between the cluster of VDDS_DDR and ground terminals, between the DDR3 interfaces on the package.
- (5) Measured from the DDR3 device power and ground terminal to the center of the capacitor package.
- (6) Per DDR3 device.
- (7) An additional HS bypass capacitor can share the connection vias only if it is mounted on the opposite side of the board. No sharing of vias is permitted on the same side of the board.
- (8) An HS bypass capacitor may share a via with a DDR3 device mounted on the same side of the PCB. A wide trace should be used for the connection and the length from the capacitor pad to the DDR3 device pad should be less than 150 mils.
- (9) Up to two pairs of DDR3 power and ground terminals may share a via.

7.7.2.3.3.7.1 Return Current Bypass Capacitors

Use additional bypass capacitors if the return current reference plane changes due to DDR3 signals hopping from one signal layer to another. The bypass capacitor here provides a path for the return current to hop planes along with the signal. As many of these return current bypass capacitors should be used as possible. Because these are returns for signal current, the signal via size may be used for these capacitors.

www.ti.com.cn

7.7.2.3.3.8 Net Classes

 $\bar{\chi}$ 7-64 lists the clock net classes for the DDR3 interface. $\bar{\chi}$ 7-65 lists the signal net classes, and associated clock net classes, for signals in the DDR3 interface. These net classes are used for the termination and routing rules that follow.

表 7-64. Clock Net Class Definitions

CLOCK NET CLASS	AM335x PIN NAMES
CK	DDR_CK and DDR_CKn
DQS0	DDR_DQS0 and DDR_DQSn0
DQS1	DDR_DQS1 and DDR_DQSn1

表 7-65. Signal Net Class Definitions

SIGNAL NET CLASS	ASSOCIATED CLOCK NET CLASS	AM335x PIN NAMES
ADDR_CTRL	СК	DDR_BA[2:0], DDR_A[15:0], DDR_CSn0, DDR_CASn, DDR_RASn, DDR_WEn, DDR_CKE, DDR_ODT
DQ0	DQS0	DDR_D[7:0], DDR_DQM0
DQ1	DQS1	DDR_D[15:8], DDR_DQM1

7.7.2.3.3.9 DDR3 Signal Termination

Signal terminations are required for the CK and ADDR_CTRL net class signals. On-device terminations (ODTs) are required on the DQS[x] and DQ[x] net class signals. Detailed termination specifications are covered in the routing rules in the following sections.

₹ 7-48 provides an example DDR3 schematic with a single 16-bit DDR3 memory device that does not have V_{TT} termination on the address and control signals. A typical DDR3 point-to-point topology may provide acceptable signal integrity without V_{TT} termination. System performance should be verified by performing signal integrity analysis using specific PCB design details before implementing this topology.

7.7.2.3.3.10 DDR_VREF Routing

DDR_VREF is used as a reference by the input buffers of the DDR3 memories as well as the AM335x device. DDR_VREF is intended to be half the DDR3 power supply voltage and is typically generated with a voltage divider connected to the VDDS_DDR power supply. It should be routed as a nominal 20-mil wide trace with 0.1 µF bypass capacitors near each device connection. Narrowing of DDR_VREF is allowed to accommodate routing congestion.

7.7.2.3.3.11 VTT

Like DDR_VREF, the nominal value of the VTT supply is half the DDR3 supply voltage. Unlike DDR_VREF, VTT is expected to source and sink current, specifically the termination current for the ADDR_CTRL net class Thevinen terminators. VTT is needed at the end of the address bus and it should be routed as a power sub-plane. VTT should be bypassed near the terminator resistors.

7.7.2.3.4 DDR3 CK and ADDR_CTRL Topologies and Routing Definition

The CK and ADDR_CTRL net classes are routed similarly and are length matched to minimize skew between them. CK is a bit more complicated because it runs at a higher transition rate and is differential. The following subsections show the topology and routing for various DDR3 configurations for CK and ADDR_CTRL. The figures in the following subsections define the terms for the routing specification detailed in 表 7-66.



7.7.2.3.4.1 Two DDR3 Devices

Two DDR3 devices are supported on the DDR3 interface consisting of two x8 DDR3 devices arranged as one 16-bit bank. These two devices may be mounted on a single side of the PCB, or may be mirrored in a pair to save board space at a cost of increased routing complexity and parts on the backside of the PCB.

7.7.2.3.4.1.1 CK and ADDR_CTRL Topologies, Two DDR3 Devices

DDR3 Differential CK Input Buffers

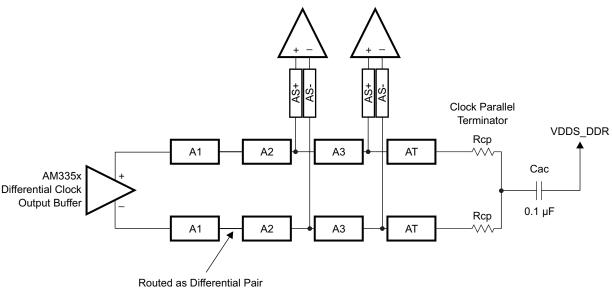


图 7-52. CK Topology for Two DDR3 Devices

DDR3 Address and Control Input Buffers

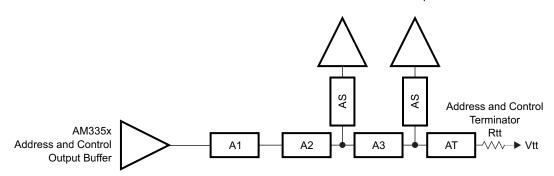


图 7-53. ADDR_CTRL Topology for Two DDR3 Devices

7.7.2.3.4.1.2 CK and ADDR_CTRL Routing, Two DDR3 Devices

图 7-54 shows the CK routing for two DDR3 devices placed on the same side of the PCB. 图 7-55 shows the corresponding ADDR_CTRL routing.

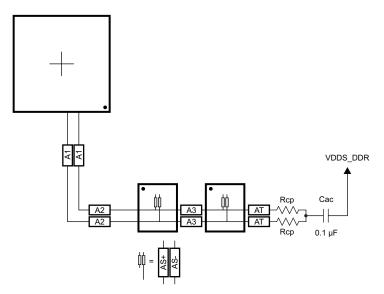


图 7-54. CK Routing for Two Single-Sided DDR3 Devices

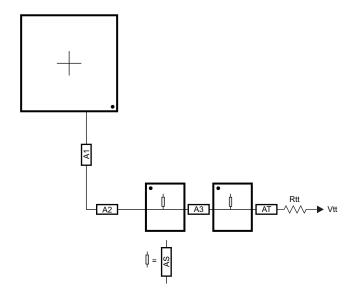


图 7-55. ADDR_CTRL Routing for Two Single-Sided DDR3 Devices



To save PCB space, the two DDR3 memories may be mounted as a mirrored pair at a cost of increased routing and assembly complexity. 图 7-56 and 图 7-57 show the routing for CK and ADDR_CTRL, respectively, for two DDR3 devices mirrored in a single-pair configuration.

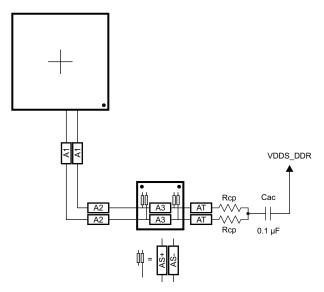


图 7-56. CK Routing for Two Mirrored DDR3 Devices

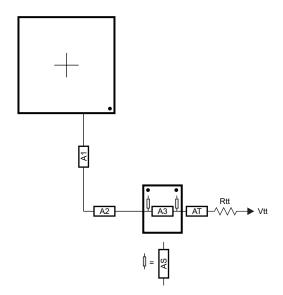


图 7-57. ADDR_CTRL Routing for Two Mirrored DDR3 Devices

7.7.2.3.4.2 One DDR3 Device

One DDR3 device is supported on the DDR3 interface consisting of one x16 DDR3 device arranged as one 16-bit bank.

7.7.2.3.4.2.1 CK and ADDR_CTRL Topologies, One DDR3 Device

§ 7-58 shows the topology of the CK net classes and
§ 7-59 shows the topology for the corresponding ADDR_CTRL net classes.

DDR3 Differential CK Input Buffer

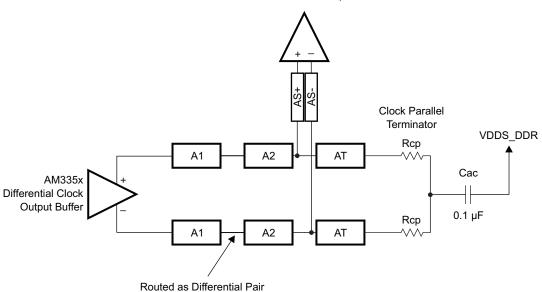


图 7-58. CK Topology for One DDR3 Device

DDR3 Address and Control Input Buffers

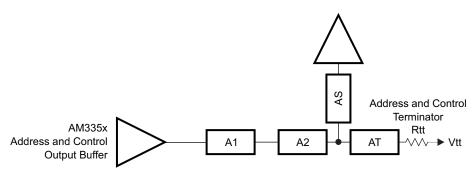


图 7-59. ADDR_CTRL Topology for One DDR3 Device



7.7.2.3.4.2.2 CK and ADDR_CTRL Routing, One DDR3 Device

图 7-60 shows the CK routing for one DDR3 device. 图 7-61 shows the corresponding ADDR_CTRL routing.

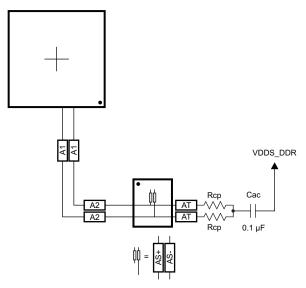


图 7-60. CK Routing for One DDR3 Device

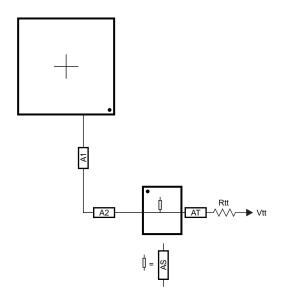


图 7-61. ADDR_CTRL Routing for One DDR3 Device

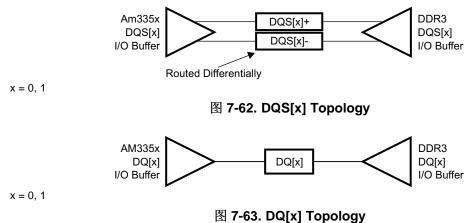
7.7.2.3.5 Data Topologies and Routing Definition

No matter the number of DDR3 devices used, the data line topology is always point to point, so its definition is simple.

7.7.2.3.5.1 DQS[x] and DQ[x] Topologies, Any Number of Allowed DDR3 Devices

DQS[x] lines are point-to-point differential, and DQ[x] lines are point-to-point single-ended. 图 7-62 and 图 7-63 show these topologies.

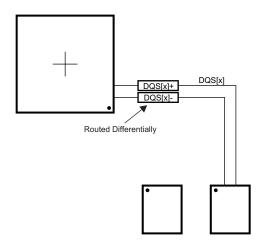




E i coi ba[x] repelegy

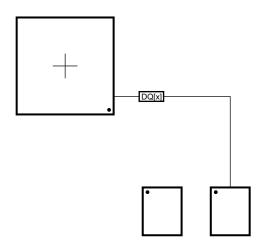
7.7.2.3.5.2 DQS[x] and DQ[x] Routing, Any Number of Allowed DDR3 Devices

图 7-64 and 图 7-65 show the DQS[x] and DQ[x] routing.



x = 0, 1

图 7-64. DQS[x] Routing With Any Number of Allowed DDR3 Devices



x = 0, 1

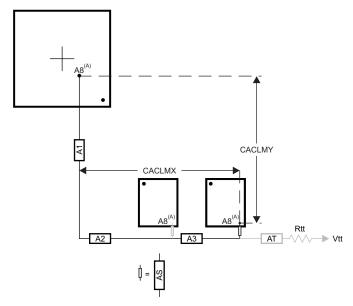
图 7-65. DQ[x] Routing With Any Number of Allowed DDR3 Devices



7.7.2.3.6 Routing Specification

7.7.2.3.6.1 CK and ADDR CTRL Routing Specification

Skew within the CK and ADDR_CTRL net classes directly reduces setup and hold margin and, thus, this skew must be controlled. The only way to practically match lengths on a PCB is to lengthen the shorter traces up to the length of the longest net in the net class and its associated clock. A metric to establish this maximum length is Manhattan distance. The Manhattan distance between two points on a PCB is the length between the points when connecting them only with horizontal or vertical segments. A reasonable trace route length is to within a percentage of its Manhattan distance. CACLM is defined as Clock Address Control Longest Manhattan distance.



A. It is very likely that the longest CK and ADDR_CTRL Manhattan distance will be for Address Input 8 (A8) on the DDR3 memories. CACLM is based on the longest Manhattan distance due to the device placement. Verify the net class that satisfies this criteria and use as the baseline for CK and ADDR_CTRL skew matching and length control.

The length of shorter CK and ADDR_CTRL stubs as well as the length of the terminator stub are not included in this length calculation. Nonincluded lengths are grayed out in the figure.

Assuming A8 is the longest, CALM = CACLMY + CACLMX + 300 mils.

The extra 300 mils allows for routing down lower than the DDR3 memories and returning up to reach A8.

图 7-66. CACLM for Two Address Loads on One Side of PCB

表 7-66. CK and ADDR_CTRL Routing Specification(1)(2)(3)

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	A1 + A2 length			2500	mils
2	A1 + A2 skew			25	mils
3	A3 length			660	mils
4	A3 skew ⁽⁴⁾			25	mils
5	A3 skew ⁽⁵⁾			125	mils
6	AS length			100	mils



表 7-66. CK and ADDR_CTRL Routing Specification(1)(2)(3) (continued)

NO.	PARAMETER	MIN	TYP	MAX	UNIT
7	AS skew			25	mils
8	AS+ and AS- length			70	mils
9	AS+ and AS- skew			5	mils
10	AT length ⁽⁶⁾		500		mils
11	AT skew ⁽⁷⁾		100		mils
12	AT skew ⁽⁸⁾			5	mils
13	CK and ADDR_CTRL nominal trace length ⁽⁹⁾	CACLM-50	CACLM	CACLM+50	mils
14	Center-to-center CK to other DDR3 trace spacing ⁽¹⁰⁾	4w			
15	Center-to-center ADDR_CTRL to other DDR3 trace spacing ⁽¹⁰⁾⁽¹¹⁾	4w			
16	Center-to-center ADDR_CTRL to other ADDR_CTRL trace spacing ⁽¹⁰⁾	3w			
17	CK center-to-center spacing ⁽¹²⁾				
18	CK spacing to other net ⁽¹⁰⁾	4w			
19	Rcp ⁽¹³⁾	Zo-1	Zo	Zo+1	Ω
20	Rtt ⁽¹³⁾⁽¹⁴⁾	Zo-5	Zo	Zo+5	Ω

- (1) CK represents the clock net class, and ADDR_CTRL represents the address and control signal net class.
- (2) The use of vias should be minimized.
- (3) Additional bypass capacitors are required when using the VDDS_DDR plane as the reference plane to allow the return current to jump between the VDDS_DDR plane and the ground plane when the net class switches layers at a via.
- (4) Mirrored configuration (one DDR3 device on top of the board and one DDR3 device on the bottom).
- (5) Nonmirrored configuration (all DDR3 memories on same side of PCB).
- (6) While this length can be increased for convenience, its length should be minimized.
- (7) ADDR_CTRL net class only (not CK net class). Minimizing this skew is recommended, but not required.
- (8) CK net class only.
- (9) CACLM is the longest Manhattan distance of the CK and ADDR_CTRL net classes + 300 mils. For definition, see 节 7.7.2.3.6.1 and 图 7-66.
- (10) Center-to-center spacing is allowed to fall to minimum (w) for up to 1250 mils of routed length.
- (11) Signals from one DQ net class should be considered other DDR3 traces to another DQ net class.
- (12) CK spacing set to ensure proper differential impedance. Differential impedance should be Z₀ x 2, where Z₀ is the single-ended impedance defined in 表 7-60.
- (13) Source termination (series resistor at driver) is specifically not allowed.
- (14) Termination values should be uniform across the net class.

7.7.2.3.6.2 DQS[x] and DQ[x] Routing Specification

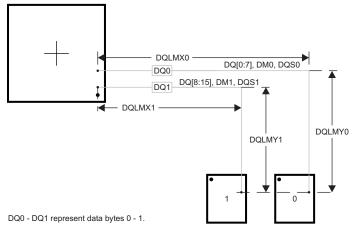
Skew within the DQS[x] and DQ[x] net classes directly reduces setup and hold margin and, thus, this skew must be controlled. The only way to practically match lengths on a PCB is to lengthen the shorter traces up to the length of the longest net in the net class and its associated clock. DQLMn is defined as DQ Longest Manhattan distance n, where n is the byte number. For a 16-bit interface, there are two DQLMs, DQLM0-DQLM1.

注

Matching the lengths across all bytes is not required, nor is it recommended. Length matching is only required within each byte.

Given the DQS[x] and DQ[x] pin locations on the AM335x device and the DDR3 memories, the maximum possible Manhattan distance can be determined given the placement. 37-67 shows this distance for a two-load case. It is from this distance that the specifications on the lengths of the transmission lines for the data bus are determined. For DQS[x] and DQ[x] routing, these specifications are contained in 57-67.





There are two DQLMs, one for each byte (16-bit interface). Each DQLM is the longest Manhattan distance of the byte;

therefore:

DQLM0 = DQLMX0 + DQLMY0 DQLM1 = DQLMX1 + DQLMY1

图 7-67. DQLM for Any Number of Allowed DDR3 Devices

表 7-67. DQS[x] and DQ[x] Routing Specification(1)(2)

NO.	PARAMETER	MIN	TYP	MAX	UNIT
1	DQ0 nominal length ⁽³⁾⁽⁴⁾			DQLM0	mils
2	DQ1 nominal length ⁽³⁾⁽⁵⁾			DQLM1	mils
3	DQ[x] skew ⁽⁶⁾			25	mils
4	DQS[x] skew			5	mils
5	DQS[x]-to-DQ[x] skew ⁽⁶⁾⁽⁷⁾			25	mils
6	Center-to-center DQ[x] to other DDR3 trace spacing ⁽⁸⁾⁽⁹⁾	4w			
7	Center-to-center DQ[x] to other DQ[x] trace spacing ⁽⁸⁾⁽¹⁰⁾	3w			
8	DQS[x] center-to-center spacing ⁽¹¹⁾				
9	DQS[x] center-to-center spacing to other net ⁽⁸⁾	4w			

- (1) DQS[x] represents the DQS0 and DQS1 clock net classes, and DQ[x] represents the DQ0 and DQ1 signal net classes.
- (2) External termination disallowed. Data termination should use built-in ODT functionality.
- (3) DQLMn is the longest Manhattan distance of a byte. For definition, see 节 7.7.2.3.6.2 and 图 7-67.
- (4) DQLM0 is the longest Manhattan length for the DQ0 net class.
- (5) DQLM1 is the longest Manhattan length for the DQ1 net class.
- (6) Length matching is only done within a byte. Length matching across bytes is not required.
- (7) Each DQS clock net class is length matched to its associated DQ signal net class.
- (8) Center-to-center spacing is allowed to fall to minimum for up to 1250 mils of routed length.
- (9) Other DDR3 trace spacing means signals that are not part of the same DQ[x] signal net class.
- (10) This applies to spacing within same DQ[x] signal net class.
- (11) DQS[x] pair spacing is set to ensure proper differential impedance. Differential impedance should be Z_o x 2, where Z_o is the single-ended impedance defined in 表 7-60.

www.ti.com.cn

7.8 I²C

For more information, see the Inter-Integrated Circuit (I²C) section of the *AM335x and AMIC110 Sitara Processors Technical Reference Manual*.

7.8.1 PC Electrical Data and Timing

表 7-68. I²C Timing Conditions – Slave Mode

	DADAMETED	STANDARD N	IODE	FAST MODE		UNIT
	PARAMETER		MAX	MIN	MAX	UNII
Output Condit	Output Condition					
C _b	Capacitive load for each bus line		400		400	pF

表 7-69. Timing Requirements for I²C Input Timings

(see \text{\begin{align*} \text{7-68} \end{align*}

NO			STANDAR	D MODE	FAST MODE	LINUT
NO.			MIN	MAX	MIN MA	UNIT
1	t _{c(SCL)}	Cycle time, SCL	10		2.5	μs
2	t _{su(SCLH-SDAL)}	Setup time, SCL high before SDA low (for a repeated START condition)	4.7		0.6	μs
3	t _{h(SDAL-SCLL)}	Hold time, SCL low after SDA low (for a START and a repeated START condition)	4		0.6	μs
4	t _{w(SCLL)}	Pulse duration, SCL low	4.7		1.3	μs
5	t _{w(SCLH)}	Pulse duration, SCL high	4		0.6	μs
6	t _{su(SDAV-SCLH)}	Setup time, SDA valid before SCL high	250		100 ⁽¹⁾	ns
7	t _{h(SCLL-SDAV)}	Hold time, SDA valid after SCL low	0 ⁽²⁾	3.45 ⁽³⁾	0 ⁽²⁾ 0.9 ⁽	³⁾ µs
8	t _{w(SDAH)}	Pulse duration, SDA high between STOP and START conditions	4.7		1.3	μs
9	t _{r(SDA)}	Rise time, SDA		1000	30) ns
10	t _{r(SCL)}	Rise time, SCL		1000	30) ns
11	t _{f(SDA)}	Fall time, SDA		300	30) ns
12	t _{f(SCL)}	Fall time, SCL		300	30) ns
13	t _{su(SCLH-SDAH)}	Setup time, high before SDA high (for STOP condition)	4		0.6	μs
14	t _{w(SP)}	Pulse duration, spike (must be suppressed)	0	50	0 5) ns

⁽¹⁾ A fast-mode I²C-bus device can be used in a standard-mode I²C-bus system, but the requirement t_{su(SDA-SCLH)}≥ 250 ns must then be met. This is automatically the case if the device does not stretch the LOW period of the SCL signal. If such a device stretches the LOW period of the SCL signal, it must output the next data bit to the SDA line t_{r max} + t_{su(SDA-SCLH)} = 1000 + 250 = 1250 ns (according to the standard-mode I²C-Bus Specification) before the SCL line is released.

⁽²⁾ A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL.

⁽³⁾ The maximum th(SDA-SCLL) has only to be met if the device does not stretch the low period [tw(SCLL)] of the SCL signal.



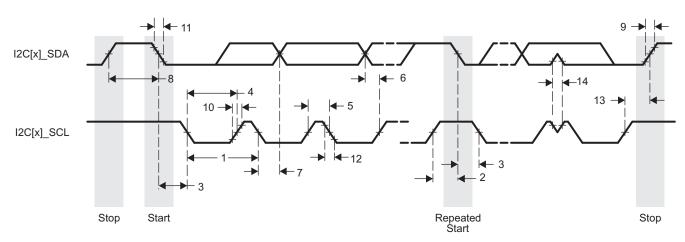


图 7-68. I²C Receive Timing

表 7-70. Switching Characteristics for I²C Output Timings

(see 图 7-69)

		DADAMETER	STANDARD	MODE	FAST MOD	E	LINUT
16 17 18 19 20 21 22 23		PARAMETER	MIN	MAX	MIN	MAX	UNIT
15	t _{c(SCL)}	Cycle time, SCL	10		2.5		μs
16	t _{su(SCLH-SDAL)}	Setup time, SCL high before SDA low (for a repeated START condition)	4.7		0.6		μs
17	t _{h(SDAL-SCLL)}	Hold time, SCL low after SDA low (for a START and a repeated START condition)	4		0.6		μs
18	t _{w(SCLL)}	Pulse duration, SCL low	4.7		1.3		μs
19	t _{w(SCLH)}	Pulse duration, SCL high	4		0.6		μs
20	t _{su(SDAV-SCLH)}	Setup time, SDA valid before SCL high	250		100		ns
21	t _{h(SCLL-SDAV)}	Hold time, SDA valid after SCL low	0	3.45	0	0.9	μs
22	t _{w(SDAH)}	Pulse duration, SDA high between STOP and START conditions	4.7		1.3		μs
23	t _{r(SDA)}	Rise time, SDA		1000		300	ns
24	t _{r(SCL)}	Rise time, SCL		1000		300	ns
25	t _{f(SDA)}	Fall time, SDA		300		300	ns
26	t _{f(SCL)}	Fall time, SCL		300		300	ns
27	t _{su(SCLH-SDAH)}	Setup time, high before SDA high (for STOP condition)	4		0.6		μs

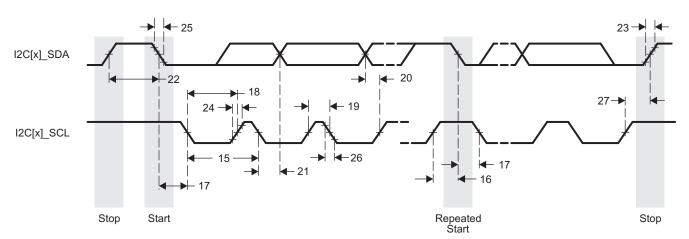


图 7-69. I²C Transmit Timing

7.9 JTAG Electrical Data and Timing

表 7-71. Timing Requirements for JTAG

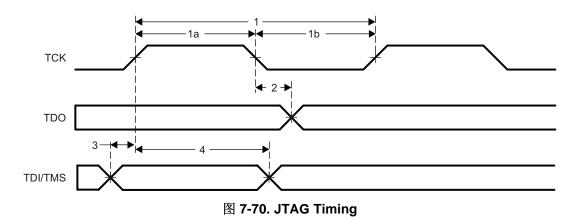
(see 图 7-70)

NO.			OPP100		OPP50		UNIT
NO.			MIN	MAX	MIN	MAX	UNII
1	t _{c(TCK)}	Cycle time, TCK	81.5		104.5		ns
1a	t _{w(TCKH)}	Pulse duration, TCK high (40% of t _c)	32.6		41.8		ns
1b	t _{w(TCKL)}	Pulse duration, TCK low (40% of t _c)	32.6		41.8		ns
3	t _{su(TDI-TCKH)}	Input setup time, TDI valid to TCK high	3		3		ns
3	t _{su(TMS-TCKH)}	Input setup time, TMS valid to TCK high	3		3	MAX	ns
4	t _{h(TCKH-TDI)}	Input hold time, TDI valid from TCK high	8.05		8.05		ns
4	t _{h(TCKH-TMS)}	Input hold time, TMS valid from TCK high	8.05		8.05		ns

表 7-72. Switching Characteristics for JTAG

(see 图 7-70)

NO	NO. PARAMETER)	OPP50	UNIT	
NO.	PARAMETER	MIN MAX MIN MA		MAX	UNII	
2	t _{d(TCKL-TDO)} Delay time, TCK low to TDO valid	3	27.6	4	36.8	ns





7.10 LCD Controller (LCDC)

The LCDC consists of two independent controllers, the raster controller and the LCD interface display driver (LIDD) controller. Each controller operates independently from the other and only one of them is active at any given time.

- The raster controller handles the synchronous LCD interface. It provides timing and data for constant graphics refresh to a passive display. It supports a wide variety of monochrome and full-color display types and sizes by use of programmable timing controls, a built-in palette, and a gray-scale and serializer. Graphics data is processed and stored in frame buffers. A frame buffer is a contiguous memory block in the system. A built-in DMA engine supplies the graphics data to the raster engine which, in turn, outputs to the external LCD device.
- The LIDD controller supports the asynchronous LCD interface. It provides full-timing programmability of control signals (CS, WE, OE, ALE) and output data.

The maximum resolution for the LCD controller is 2048×2048 pixels. The maximum frame rate is determined by the image size in combination with the pixel clock rate.

表 7-73. LCD Controller Timing Conditions

	PARAMETER		MIN	TYP MAX	UNIT
Output Condi	tion				
C _{LOAD}	Output load conscitones	LIDD mode	5	60	~F
	Output load capacitance	Raster mode	3	30	p⊦

7.10.1 LCD Interface Display Driver (LIDD Mode)

表 7-74. Timing Requirements for LCD LIDD Mode

(see 图 7-72 through 图 7-80)

NO.			OPP100		UNIT
NO.			MIN	MAX	UNII
16	t _{su(LCD_DATA-LCD_MEMORY_CLK)}	Setup time, LCD_DATA[15:0] valid before LCD_MEMORY_CLK high	18		ns
17	t _h (LCD_MEMORY_CLK-LCD_DATA)	Hold time, LCD_DATA[15:0] valid after LCD_MEMORY_CLK high	0		ns
18	t _{t(LCD_DATA)}	Transition time, LCD_DATA[15:0]	1	3	ns

表 7-75. Switching Characteristics for LCD LIDD Mode

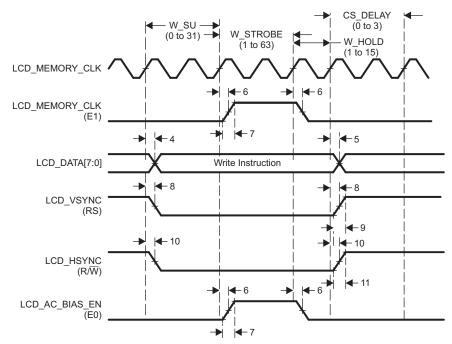
(see 图 7-72 through 图 7-80)

NO		DADAMETED	OPP100		
NO.		PARAMETER	MIN	MAX	UNIT
1	t _{c(LCD_MEMORY_CLK)}	Cycle time, LCD_MEMORY_CLK	23.7		ns
2	t _{w(LCD_MEMORY_CLKH)}	Pulse duration, LCD_MEMORY_CLK high	0.45t _c	$0.55t_{c}$	ns
3	t _{w(LCD_MEMORY_CLKL)}	Pulse duration, LCD_MEMORY_CLK low	0.45t _c	$0.55t_{c}$	ns
4	t _{d(LCD_MEMORY_CLK-LCD_DATAV)}	Delay time, LCD_MEMORY_CLK high to LCD_DATA[15:0] valid (write)		7	ns
5	t _{d(LCD_MEMORY_CLK-LCD_DATAI)}	Delay time, LCD_MEMORY_CLK high to LCD_DATA[15:0] invalid (write)	0		ns
6	t _{d(LCD_MEMORY_CLK-LCD_AC_BIAS_EN)}	Delay time, LCD_MEMORY_CLK high to LCD_AC_BIAS_EN	0	6.8	ns
7	t _t (LCD_AC_BIAS_EN)	Transition time, LCD_AC_BIAS_EN	1	10	ns
8	t _{d(LCD_MEMORY_CLK-LCD_VSYNC)}	Delay time, LCD_MEMORY_CLK high to LCD_VSYNC	0	7	ns
9	t _{t(LCD_VSYNC)}	Transition time, LCD_VSYNC	1	10	ns
10	t _{d(LCD_MEMORY_CLK-LCD_HYSNC)}	Delay time, LCD_MEMORY_CLK high to LCD_HSYNC	0	7	ns

表 7-75. Switching Characteristics for LCD LIDD Mode (continued)

(see 图 7-72 through 图 7-80)

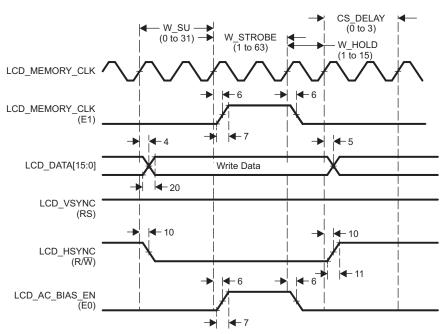
NO.		PARAMETER	OPP100		UNIT
NO.		PARAMETER	MIN	MAX 1 10 0 7 1 10 0 7	ONII
11	t _t (LCD_HSYNC)	Transition time, LCD_HYSNC	1	10	ns
12	t _d (LCD_MEMORY_CLK-LCD_PCLK)	Delay time, LCD_MEMORY_CLK high to LCD_PCLK	0	7	ns
13	t _{t(LCD_PCLK)}	Transition time, LCD_PCLK	1	10	ns
14	td(LCD_MEMORY_CLK-LCD_DATAZ)	Delay time, LCD_MEMORY_CLK high to LCD_DATA[15:0] high-Z	0	7	ns
15	td(LCD_MEMORY_CLK-LCD_DATA)	Delay time, LCD_MEMORY_CLK high to LCD_DATA[15:0] driven	0	7	ns
19	t _t (LCD_MEMORY_CLK)	Transition time, LCD_MEMORY_CLK	1	2.5	ns
20	t _{t(LCD_DATA)}	Transition time, LCD_DATA	1	10	ns



A. Hitachi mode performs asynchronous operations that do not require an external LCD_MEMORY_CLK. The first LCD_MEMORY_CLK waveform is only shown as a reference of the internal clock that sequences the other signals. The second LCD_MEMORY_CLK waveform is shown as E1 because the LCD_MEMORY_CLK signal is used to implement the E1 function in Hitachi mode.

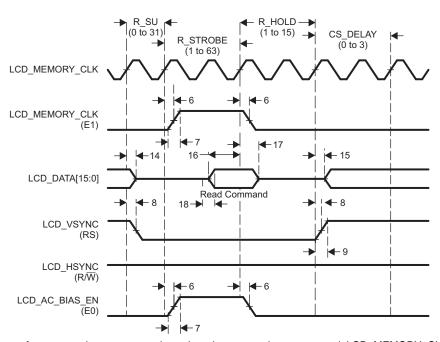
图 7-71. Command Write in Hitachi Mode





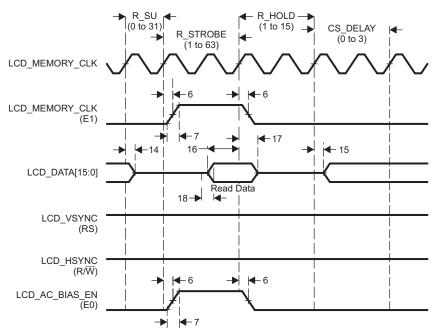
A. Hitachi mode performs asynchronous operations that do not require an external LCD_MEMORY_CLK. The first LCD_MEMORY_CLK waveform is only shown as a reference of the internal clock that sequences the other signals. The second LCD_MEMORY_CLK waveform is shown as E1 because the LCD_MEMORY_CLK signal is used to implement the E1 function in Hitachi mode.

图 7-72. Data Write in Hitachi Mode



A. Hitachi mode performs asynchronous operations that do not require an external LCD_MEMORY_CLK. The first LCD_MEMORY_CLK waveform is only shown as a reference of the internal clock that sequences the other signals. The second LCD_MEMORY_CLK waveform is shown as E1 because the LCD_MEMORY_CLK signal is used to implement the E1 function in Hitachi mode.

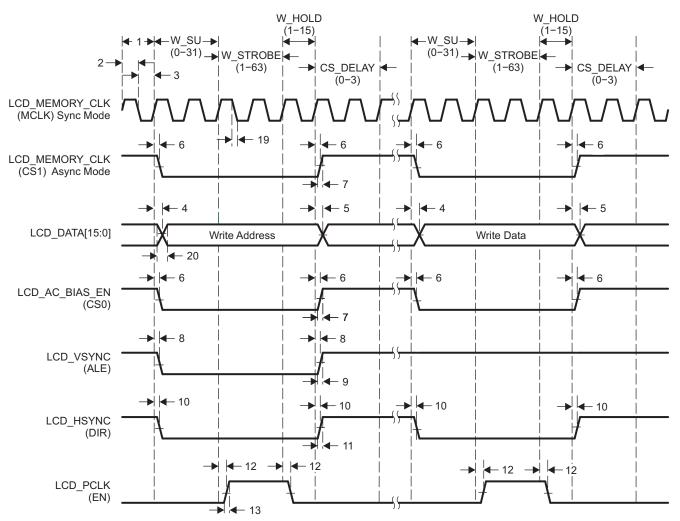
图 7-73. Command Read in Hitachi Mode



A. Hitachi mode performs asynchronous operations that do not require an external LCD_MEMORY_CLK. The first LCD_MEMORY_CLK waveform is only shown as a reference of the internal clock that sequences the other signals. The second LCD_MEMORY_CLK waveform is shown as E1 because the LCD_MEMORY_CLK signal is used to implement the E1 function in Hitachi mode.

图 7-74. Data Read in Hitachi Mode

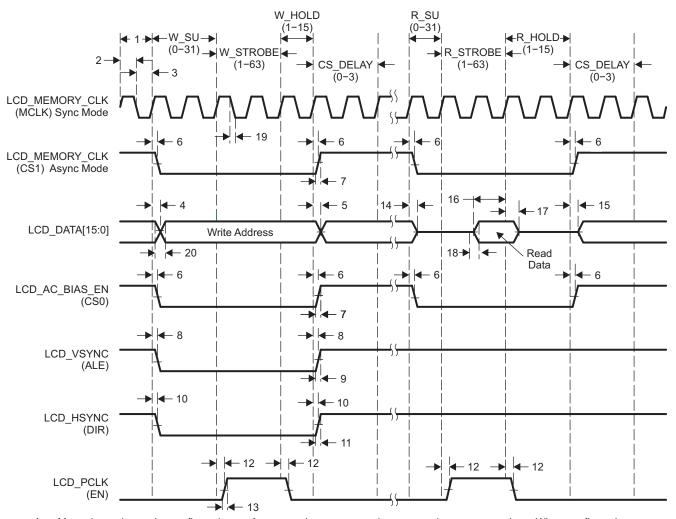




A. Motorola mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

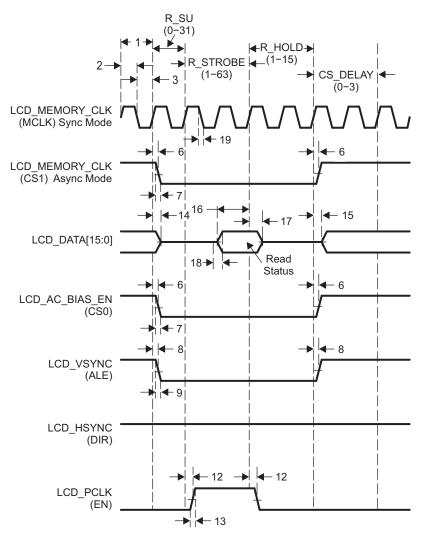
图 7-75. Micro-Interface Graphic Display Motorola Write





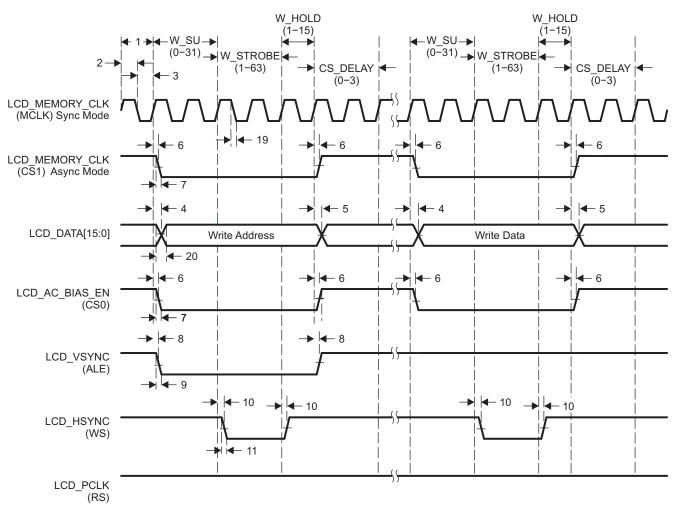
A. Motorola mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

图 7-76. Micro-Interface Graphic Display Motorola Read



A. Motorola mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

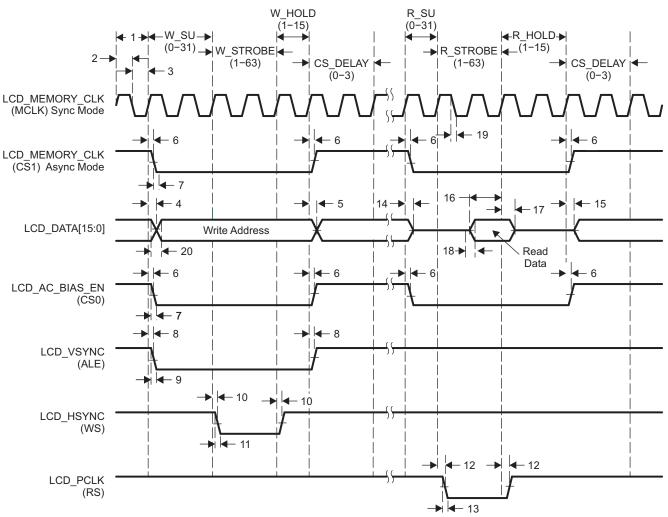
图 7-77. Micro-Interface Graphic Display Motorola Status



A. Intel mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

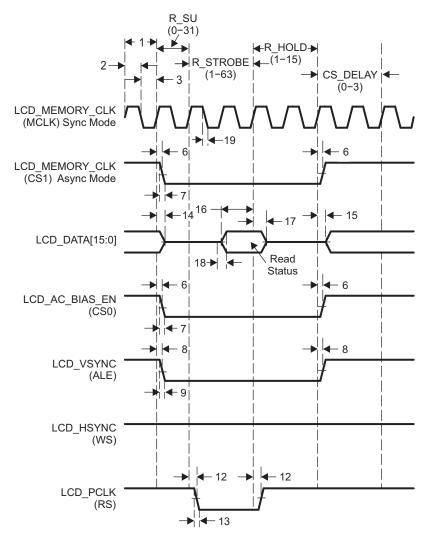
图 7-78. Micro-Interface Graphic Display Intel Write





A. Intel mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

图 7-79. Micro-Interface Graphic Display Intel Read



A. Intel mode can be configured to perform asynchronous operations or synchronous operations. When configured in asynchronous mode, LCD_MEMORY_CLK is not required, so it performs the CS1 function. When configured in synchronous mode, LCD_MEMORY_CLK performs the MCLK function. LCD_MEMORY_CLK is also shown as a reference of the internal clock that sequences the other signals.

图 7-80. Micro-Interface Graphic Display Intel Status



7.10.2 LCD Raster Mode

表 7-76. Switching Characteristics for LCD Raster Mode

(see 图 7-82 through 图 7-85)

NO	PARAMETER		OPP	50	OPP100		LINUT
NO.			MIN	MAX	MIN	MAX	UNIT
1	t _{c(LCD_PCLK)}	Cycle time, pixel clock	15.8		7.9		ns
2	t _{w(LCD_PCLKH)}	Pulse duration, pixel clock high	0.45t _c	$0.55t_{c}$	$0.45t_{c}$	$0.55t_{c}$	ns
3	t _{w(LCD_PCLKL)}	Pulse duration, pixel clock low	0.45t _c	$0.55t_{c}$	$0.45t_{c}$	$0.55t_{c}$	ns
4	t _d (LCD_PCLK-LCD_DATAV)	Delay time, LCD_PCLK to LCD_DATA[23:0] valid (write)		3.0		1.9	ns
5	t _{d(LCD_PCLK-LCD_DATAI)}	Delay time, LCD_PCLK to LCD_DATA[23:0] invalid (write)	-3.0		-1.7		ns
6	t _{d(LCD_PCLK-LCD_AC_BIAS_EN)}	Delay time, LCD_PCLK to LCD_AC_BIAS_EN	-3.0	3.0	-1.7	1.9	ns
7	t _{t(LCD_AC_BIAS_EN)}	Transition time, LCD_AC_BIAS_EN	0.5	2.4	0.5	2.4	ns
8	t _{d(LCD_PCLK-LCD_VSYNC)}	Delay time, LCD_PCLK to LCD_VSYNC	-3.0	3.0	-1.7	1.9	ns
9	t _{t(LCD_VSYNC)}	Transition time, LCD_VSYNC	0.5	2.4	0.5	2.4	ns
10	t _{d(LCD_PCLK-LCD_HSYNC)}	Delay time, LCD_PCLK to LCD_HSYNC	-3.0	3.0	-1.7	1.9	ns
11	t _{t(LCD_HSYNC)}	Transition time, LCD_HSYNC	0.5	2.4	0.5	2.4	ns
12	t _{t(LCD_PCLK)}	Transition time, LCD_PCLK	0.5	2.4	0.5	2.4	ns
13	t _{t(LCD_DATA)}	Transition time, LCD_DATA	0.5	2.4	0.5	2.4	ns

Frame-to-frame timing is derived through the following parameters in the LCD (RASTER_TIMING_1) register:

- Vertical front porch (VFP)
- Vertical sync pulse width (VSW)
- Vertical back porch (VBP)
- Lines per panel (LPP_B10 + LPP)

Line-to-line timing is derived through the following parameters in the LCD (RASTER_TIMING_0) register:

- Horizontal front porch (HFP)
- Horizontal sync pulse width (HSW)
- Horizontal back porch (HBP)
- Pixels per panel (PPLMSB + PPLLSB)

LCD_AC_BIAS_EN timing is derived through the following parameter in the LCD (RASTER_TIMING_2) register:

• AC bias frequency (ACB)

The display format produced in raster mode is shown in [8] 7-81. An entire frame is delivered one line at a time. The first line delivered starts at data pixel (1, 1) and ends at data pixel (P, 1). The last line delivered starts at data pixel (1, L) and ends at data pixel (P, L). The beginning of each new frame is denoted by the activation of I/O signal LCD_VSYNC. The beginning of each new line is denoted by the activation of I/O signal LCD_HSYNC.

www.ti.com.cn

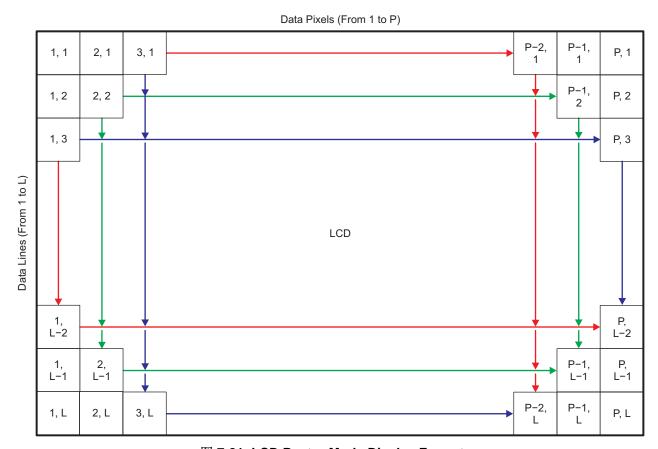


图 7-81. LCD Raster-Mode Display Format



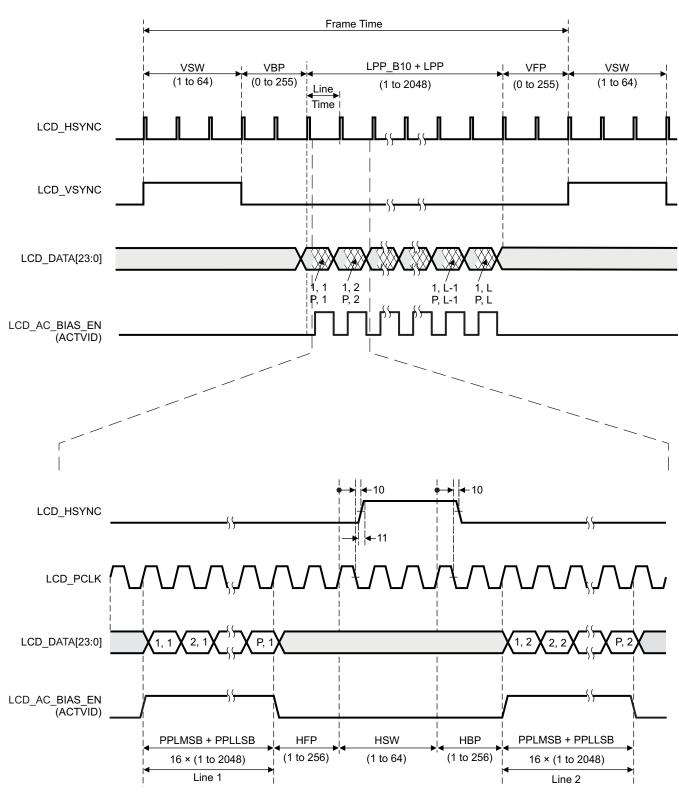
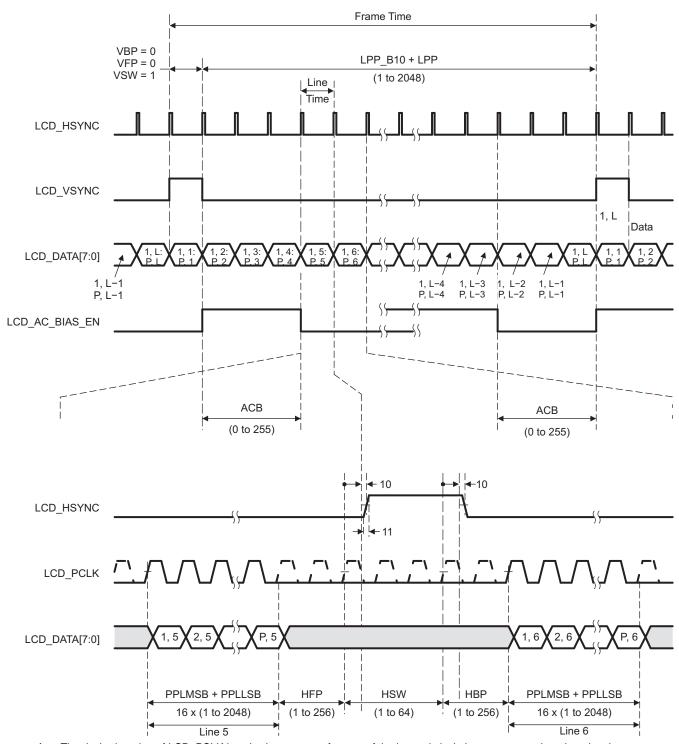


图 7-82. LCD Raster-Mode Active

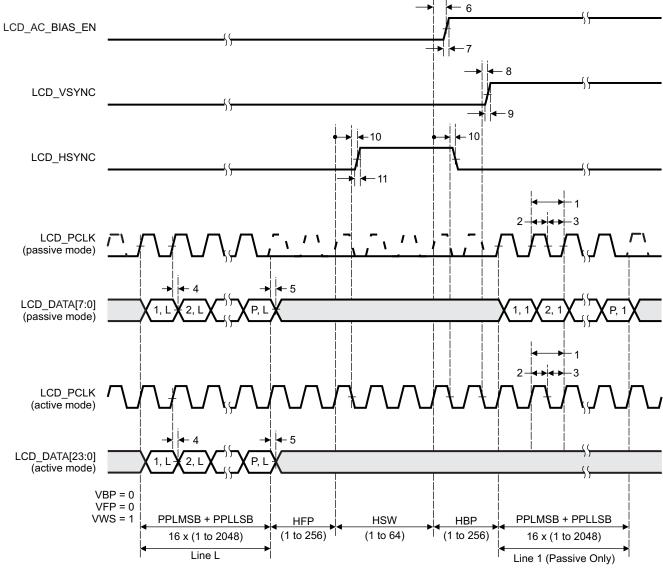




A. The dashed portion of LCD_PCLK is only shown as a reference of the internal clock that sequences the other signals.

图 7-83. LCD Raster-Mode Passive

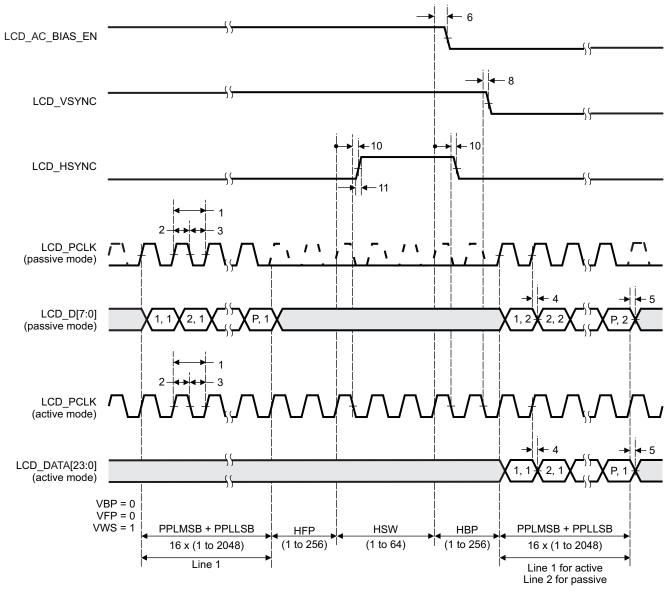




A. The dashed portion of LCD_PCLK is only shown as a reference of the internal clock that sequences the other signals.

图 7-84. LCD Raster-Mode Control Signal Activation





A. The dashed portion of LCD_PCLK is only shown as a reference of the internal clock that sequences the other signals.

图 7-85. LCD Raster-Mode Control Signal Deactivation



7.11 Multichannel Audio Serial Port (McASP)

The multichannel audio serial port (McASP) functions as a general-purpose audio serial port optimized for the needs of multichannel audio applications. The McASP is useful for time-division multiplexed (TDM) stream, Inter-Integrated Sound (I²S) protocols, and inter-component digital audio interface transmission (DIT).

7.11.1 McASP Device-Specific Information

The device includes two multichannel audio serial port (McASP) interface peripherals (McASP0 and McASP1). The McASP module consists of a transmit and receive section. These sections can operate completely independently with different data formats, separate master clocks, bit clocks, and frame syncs or, alternatively, the transmit and receive sections may be synchronized. The McASP module also includes shift registers that may be configured to operate as either transmit data or receive data.

The transmit section of the McASP can transmit data in either a time-division-multiplexed (TDM) synchronous serial format or in a digital audio interface (DIT) format where the bit stream is encoded for SPDIF, AES-3, IEC-60958, CP-430 transmission. The receive section of the McASP peripheral supports the TDM synchronous serial format.

The McASP module can support one transmit data format (either a TDM format or DIT format) and one receive format at a time. All transmit shift registers use the same format and all receive shift registers use the same format; however, the transmit and receive formats need not be the same. Both the transmit and receive sections of the McASP also support burst mode, which is useful for nonaudio data (for example, passing control information between two devices).

The McASP peripheral has additional capability for flexible clock generation and error detection/handling, as well as error management.

The device McASP0 and McASP1 modules have up to four serial data pins each. The McASP FIFO size is 256 bytes and two DMA and two interrupt requests are supported. Buffers are used transparently to better manage DMA, which can be leveraged to manage data flow more efficiently.

For more detailed information on and the functionality of the McASP peripheral, see the Multichannel Audio Serial Port (McASP) section of the *AM335x and AMIC110 Sitara Processors Technical Reference Manual*.



7.11.2 McASP Electrical Data and Timing

表 7-77. McASP Timing Conditions

	PARAMETER	MIN	TYP MAX	UNIT		
Input Con	ditions					
t _R	Input signal rise time	1 ⁽¹⁾	4 ⁽¹⁾	ns		
t _F	Input signal fall time	1 ⁽¹⁾	4 ⁽¹⁾	ns		
Output Co	Output Condition					
C _{LOAD}	Output load capacitance	15	30	pF		

⁽¹⁾ Except when specified otherwise.

表 7-78. Timing Requirements for McASP⁽¹⁾

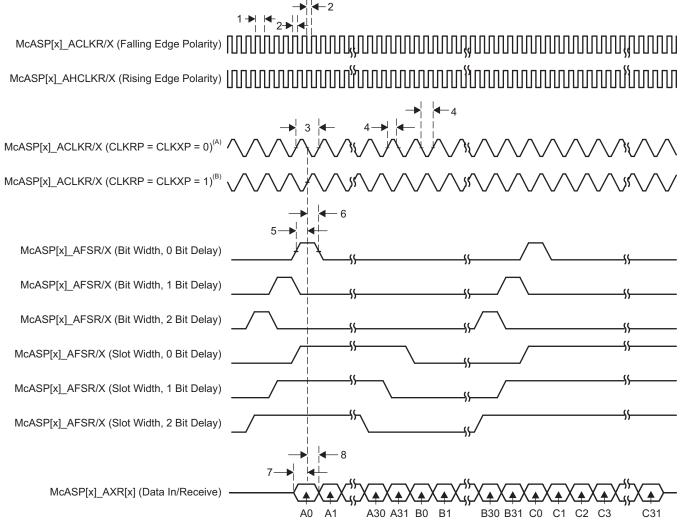
(see 图 7-86)

NO				OPP100		OPP50		LINUT
NO.				MIN	MAX	MIN	MAX	UNIT
1	t _{c(AHCLKRX)}	Cycle time, McASP[x]_AHCLKR and McASP[x]_AHCLKX		20		40		ns
2	t _{w(AHCLKRX)}	Pulse duration, McASP[x]_AHCLKR a McASP[x]_AHCLKX high or low	and	0.5P - 2.5 ⁽²⁾		0.5P - 2.5 ⁽²⁾		ns
3	t _{c(ACLKRX)}	Cycle time, McASP[x]_ACLKR and McASP[x]_ACLKX		20		40		ns
4	t _{w(ACLKRX)}	Pulse duration, McASP[x]_ACLKR an McASP[x]_ACLKX high or low	nd	0.5R - 2.5 ⁽³⁾		0.5R - 2.5 ⁽³⁾		ns
		Setup time, McASP[x]_AFSR and	ACLKR and ACLKX int	11.5		15.5		
	t _{su(AFSRX} - ACLKRX)	McASP[x]_AFSX input valid before McASP[x]_ACLKR and	ACLKR and ACLKX ext in	4		6		ns
		McASP[x]_ACLKX	ACLKR and ACLKX ext out	4		6		
	t _{h(ACLKRX-} AFSRX)		ACLKR and ACLKX int	-1		-1		ns
6			ACLKR and ACLKX ext in	0.4		0.4		
			ACLKR and ACLKX ext out	0.4		0.4		
		Setup time, McASP[x]_AXR input valid before McASP[x]_ACLKR and McASP[x]_ACLKX	ACLKR and ACLKX int	11.5		15.5		
7	t _{su(AXR-ACLKRX)}		ACLKR and ACLKX ext in	4		6		ns
		inio tei [x]_ te⊑iet	ACLKR and ACLKX ext out	4		6		
	t _{h(ACLKRX-AXR)}	Hold time, McASP[x]_AXR input valid after McASP[x]_ACLKR and McASP[x]_ACLKX	ACLKR and ACLKX int	-1		-1		
8			ACLKR and ACLKX ext in	0.4		0.4		ns
			ACLKR and ACLKX ext out	0.4		0.4		

⁽¹⁾ ACLKR internal: ACLKRCTL.CLKRM = 1, PDIR.ACLKR = 1 ACLKR external input: ACLKRCTL.CLKRM = 0, PDIR.ACLKR = 0 ACLKR external output: ACLKRCTL.CLKRM = 0, PDIR.ACLKR = 1 ACLKX internal: ACLKXCTL.CLKXM = 1, PDIR.ACLKX = 1 ACLKX external input: ACLKXCTL.CLKXM = 0, PDIR.ACLKX = 0 ACLKX external output: ACLKXCTL.CLKXM = 0, PDIR.ACLKX = 1

- (2) P = McASP[x]_AHCLKR and McASP[x]_AHCLKX period in nanoseconds (ns).
- (3) $R = McASP[x]_ACLKR$ and $McASP[x]_ACLKX$ period in ns.





- A. For CLKRP = CLKXP = 0, the McASP transmitter is configured for rising edge (to shift data out) and the McASP receiver is configured for falling edge (to shift data in).
- B. For CLKRP = CLKXP = 1, the McASP transmitter is configured for falling edge (to shift data out) and the McASP receiver is configured for rising edge (to shift data in).

图 7-86. McASP Input Timing



表 7-79. Switching Characteristics for McASP⁽¹⁾

(see 图 7-87)

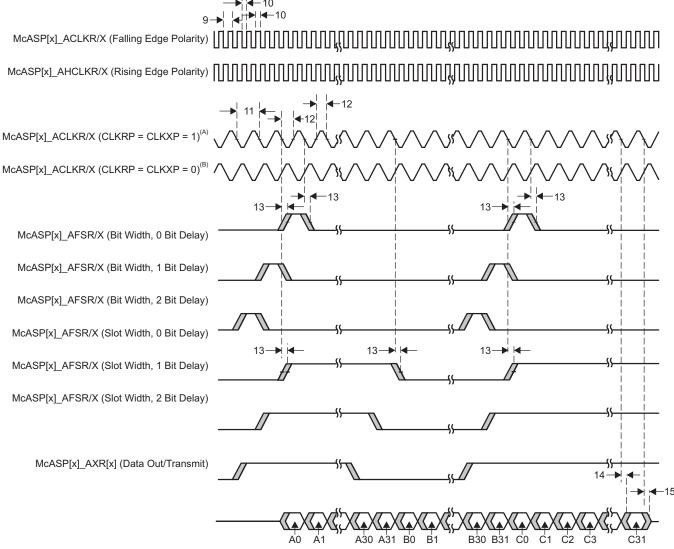
NO.	PARAMETER			OPP100		OPP50		UNIT
NO.		PARAMETER		MIN	MAX	MIN	MAX	UNII
9	t _{c(AHCLKRX)}	Cycle time, McASP[x]_AHCLKR and McASP[x]_AHCLKX		20 ⁽²⁾		40		ns
10	t _{w(AHCLKRX)}	Pulse duration, McASP[x]_AHCLKR McASP[x]_AHCLKX high or low	and	0.5P - 2.5 ⁽³⁾		0.5P - 2.5 ⁽³⁾		ns
11	t _{c(ACLKRX)}	Cycle time, McASP[x]_ACLKR and McASP[x]_ACLKX		20		40		ns
12	t _{w(ACLKRX)}	Pulse duration, McASP[x]_ACLKR an McASP[x]_ACLKX high or low	nd	0.5P - 2.5 ⁽³⁾		0.5P - 2.5 ⁽³⁾		ns
	t _{d(ACLKRX-AFSRX)}	Delay time, McASP[x]_ACLKR and McASP[x]_ACLKX transmit edge to	ACLKR and ACLKX int	0	6	0	6	
		McASP[x]_AFSR and McASP[x]_AFSX output valid	ACLKR and ACLKX ext in	2	13.5	2	18	
13		Delay time, McASP[x]_ACLKR and McASP[x]_ACLKX transmit edge to McASP[x]_AFSR and McASP[x]_AFSX output valid with Pad Loopback	ACLKR and ACLKX ext out	2	13.5	2	18	ns
		Delay time, McASP[x]_ACLKX	ACLKX int	0	6	0	6	
14		transmit edge to McASP[x]_AXR output valid	ACLKX ext in	2	13.5	2	18	
14	t _{d(ACLKX-AXR)}	Delay time, McASP[x]_ACLKX transmit edge to McASP[x]_AXR output valid with Pad Loopback	ACLKX ext	2	13.5	2	18	ns
	t _{dis(ACLKX-AXR)}	Disable time, McASP[x]_ACLKX	ACLKX int	0	6	0	6	
		transmit edge to McASP[x]_AXR output high impedance	ACLKX ext in	2	13.5	2	18	
15		Disable time, McASP[x]_ACLKX transmit edge to McASP[x]_AXR output high impedance with pad loopback	ACLKX ext out	2	13.5	2	18	ns

⁽¹⁾ ACLKR internal: ACLKRCTL.CLKRM = 1, PDIR.ACLKR = 1 ACLKR external input: ACLKRCTL.CLKRM = 0, PDIR.ACLKR = 0 ACLKR external output: ACLKRCTL.CLKRM = 0, PDIR.ACLKR = 1 ACLKX internal: ACLKXCTL.CLKXM = 1, PDIR.ACLKX = 1 ACLKX external input: ACLKXCTL.CLKXM = 0, PDIR.ACLKX = 0 ACLKX external output: ACLKXCTL.CLKXM = 0, PDIR.ACLKX = 1

^{(2) 50} MHz

⁽³⁾ P = AHCLKR and AHCLKX period.





- A. For CLKRP = CLKXP = 1, the McASP transmitter is configured for falling edge (to shift data out) and the McASP receiver is configured for rising edge (to shift data in).
- B. For CLKRP = CLKXP = 0, the McASP transmitter is configured for rising edge (to shift data out) and the McASP receiver is configured for falling edge (to shift data in).

图 7-87. McASP Output Timing

www.ti.com.cn

7.12 Multichannel Serial Port Interface (McSPI)

For more information, see the Multichannel Serial Port Interface (McSPI) section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.

7.12.1 McSPI Electrical Data and Timing

The following timings are applicable to the different configurations of McSPI in master or slave mode for any McSPI and any channel (n).

7.12.1.1 McSPI—Slave Mode

表 7-80. McSPI Timing Conditions - Slave Mode

	PARAMETER						
Input Conditio	•	·					
t _r	Input signal rise time		5 ns				
t _f	Input signal fall time		5 ns				
Output Condit	Output Condition						
C _{load}	Output load capacitance	2	0 pF				

表 7-81. Timing Requirements for McSPI Input Timings—Slave Mode

(see 图 7-88)

NO			OPP100		OPP50		UNIT
NO.			MIN	MAX	MIN	MAX	UNII
1	t _{c(SPICLK)}	Cycle time, SPI_CLK	62.5		124.8		ns
2	t _{w(SPICLKL)}	Typical pulse duration, SPI_CLK low	0.5P – 3.12 ⁽¹⁾	0.5P + 3.12 ⁽¹⁾	0.5P – 3.12 ⁽¹⁾	0.5P + 3.12 ⁽¹⁾	ns
3	t _{w(SPICLKH)}	Typical pulse duration, SPI_CLK high	0.5P – 3.12 ⁽¹⁾	0.5P + 3.12 ⁽¹⁾	0.5P – 3.12 ⁽¹⁾	0.5P + 3.12 ⁽¹⁾	ns
4	t _{su(SIMO-SPICLK)}	Setup time, SPI_D[x] (SIMO) valid before SPI_CLK active edge ⁽²⁾⁽³⁾	12.92		12.92		ns
5	t _{h(SPICLK-SIMO)}	Hold time, SPI_D[x] (SIMO) valid after SPI_CLK active edge ⁽²⁾⁽³⁾	12.92		12.92		ns
8	t _{su(CS-SPICLK)}	Setup time, SPI_CS valid before SPI_CLK first edge ⁽²⁾	12.92		12.92		ns
9	t _{h(SPICLK-CS)}	Hold time, SPI_CS valid after SPI_CLK last edge ⁽²⁾	12.92		12.92		ns

⁽¹⁾ P = SPI_CLK period.

表 7-82. Switching Characteristics for McSPI Output Timings—Slave Mode

(see **§** 7-89)

NO.	PARAMETER		OPP100		OPP50		LINUT
NO.			MIN	MAX	MIN	MAX	UNIT
6	t _d (SPICLK-SOMI)	Delay time, SPI_CLK active edge to SPI_D[x] (SOMI) transition ⁽¹⁾⁽²⁾	-4.00	17.12	-4.00	17.12	ns
7	t _{d(CS-SOMI)}	Delay time, SPI_CS active edge to SPI_D[x] (SOMI) transition ⁽¹⁾⁽²⁾		17.12		17.12	ns

⁽¹⁾ This timing applies to all configurations regardless of MCSPIX_CLK polarity and which clock edges are used to drive output data and capture input data.

(2) Pins SPIx_D0 and SPIx_D1 can function as SIMO or SOMI.

⁽²⁾ This timing applies to all configurations regardless of MCSPIX_CLK polarity and which clock edges are used to drive output data and capture input data.

⁽³⁾ Pins SPIx_D0 and SPIx_D1 can function as SIMO or SOMI.

SPI_D[x] (SIMO, In)

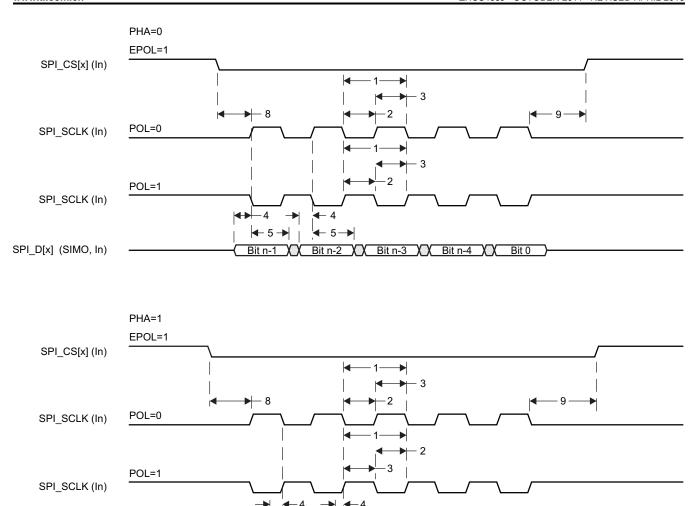


图 7-88. SPI Slave Mode Receive Timing

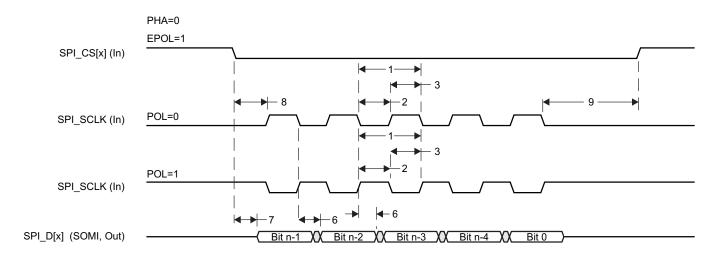
- 5 **→**

Bit n-2 () Bit n-3 ()

Bit 1

- 5 **→**





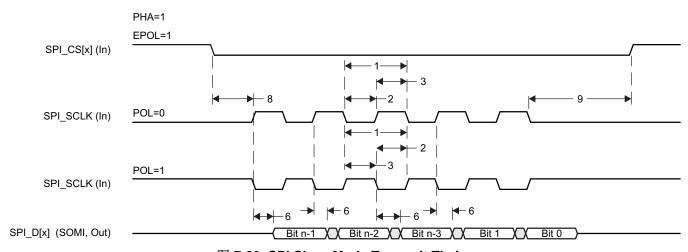


图 7-89. SPI Slave Mode Transmit Timing



7.12.1.2 McSPI—Master Mode

表 7-83. McSPI Timing Conditions - Master Mode

	PARAMETER	LOW LOA	D	HIGH LOA	AD	UNIT
	PARAMETER	MIN	MAX	MIN	MAX	UNIT
Input Condition	ns		·		·	
t _r	Input signal rise time		8		8	ns
t _f	Input signal fall time		8		8	ns
Output Conditi	on					
C _{load}	Output load capacitance		5		25	pF

表 7-84. Timing Requirements for McSPI Input Timings – Master Mode

(see 图 7-90)

				OPP100					OPI	P50		
NO.				LOW L	OAD	HIGH L	OAD.	LOW L	OAD	HIGH L	OAD	UNIT
				MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
4	t _{su(SOMI-} SPICLKH)	Setup time, SPI_D[x] (SOM SPI_CLK active edge ⁽¹⁾	II) valid before	2.29		3.02		2.29		3.02		ns
5	t _{h(SPICLKH-}	Hold time, SPI_D[x] (SOMI) valid after	Industrial extended temperature (-40°C to 125°C)	7.1		7.1		7.1		7.1		ns
	SOMI)	SPI_CLK active edge ⁽¹⁾	All other temperature ranges	4.7		4.7		4.7		4.7		

⁽¹⁾ Pins SPIx_D0 and SPIx_D1 can function as SIMO or SOMI.

表 7-85. Switching Characteristics for McSPI Output Timings – Master Mode

(see **8** 7-91)

					OP	P100			OP	P50		
NO.		PARAMETER		LOW LO	DAD	HIGH LO	AD	LOW LO	DAD	HIGH LO	AD	UNIT
				MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
1	t _{c(SPICLK)}	Cycle time, SPI_	_CLK	20.8		20.8		41.6		41.6		ns
2	t _{w(SPICLKL)}	Typical pulse du SPI_CLK low	ıration,	0.5P – 1.04 ⁽¹⁾	0.5P + 1.04 ⁽¹⁾	0.5P – 2.08 ⁽¹⁾	0.5P + 2.08 ⁽¹⁾	0.5P – 1.04 ⁽¹⁾	0.5P + 1.04 ⁽¹⁾	0.5P – 2.08 ⁽¹⁾	0.5P + 2.08 ⁽¹⁾	ns
	t _{w(SPICLKH)}	Typical pulse du SPI_CLK high	ıration,	0.5P – 1.04 ⁽¹⁾	0.5P + 1.04 ⁽¹⁾	0.5P – 2.08 ⁽¹⁾	0.5P + 2.08 ⁽¹⁾	0.5P – 1.04 ⁽¹⁾	0.5P + 1.04 ⁽¹⁾	0.5P – 2.08 ⁽¹⁾	0.5P + 2.08 ⁽¹⁾	ns
3	t _{r(SPICLK)}	Rising time, SPI	_CLK		3.82		3.82		3.82		3.82	ns
	t _{f(SPICLK)}	Falling time, SP	I_CLK		3.44		3.44		3.44		3.44	ns
6	t _{d(SPICLK-SIMO)}	Delay time, SPI active edge to S (SIMO) transition	PI_D[x]	-3.57	3.57	-4.62	4.62	-3.57	3.57	-4.62	4.62	ns
7	t _{d(CS-SIMO)}	Delay time, SPI_edge to SPI_D[x transition ⁽²⁾			3.57		4.62		3.57		4.62	ns
0		Delay time, SPI_CS active	Mode 1 and 3 ⁽³⁾	A - 4.2 ⁽⁴⁾		A – 2.54 ⁽⁴⁾		A - 4.2 ⁽⁴⁾		A – 2.54 ⁽⁴⁾		ns
8	t _d (CS-SPICLK)	to SPI_CLK first edge	Mode 0 and 2 ⁽³⁾	B – 4.2 ⁽⁵⁾		B – 2.54 ⁽⁵⁾		B - 4.2 ⁽⁵⁾		B – 2.54 ⁽⁵⁾		ns
		Delay time, SPI_CLK last	Mode 1 and 3 ⁽³⁾	B – 4.2 ⁽⁵⁾		B – 2.54 ⁽⁵⁾		B - 4.2 ⁽⁵⁾		B – 2.54 ⁽⁵⁾		ns
9	t _d (SPICLK-CS)	edge to SPI_CS inactive	Mode 0 and 2 ⁽³⁾	A - 4.2 ⁽⁴⁾		A – 2.54 ⁽⁴⁾		A - 4.2 ⁽⁴⁾		A – 2.54 ⁽⁴⁾		ns

⁽¹⁾ P = SPI_CLK period.

⁽²⁾ Pins SPIx_D0 and SPIx_D1 can function as SIMO or SOMI.

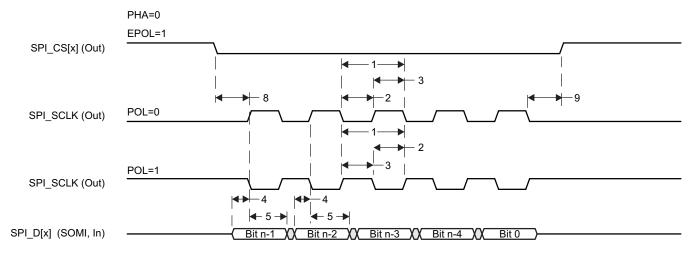
⁽³⁾ The polarity of SPIx_CLK and the active edge (rising or falling) on which mcspix_simo is driven and mcspix_somi is latched is all software configurable:

⁻ SPIx_CLK(1) phase programmable with the bit PHA of MCSPI_CH(i)CONF register: PHA = 1 (Modes 1 and 3).

⁻ SPIx_CLK(1) phase programmable with the bit PHA of MCSPI_CH(i)CONF register: PHA = 0 (Modes 0 and 2).



- (4) Case P = 20.8 ns, A = (TCS + 1) x TSPICLKREF (TCS is a bit field of MCSPI_CH(i)CONF register). Case P > 20.8 ns, A = (TCS + 0.5) x Fratio x TSPICLKREF (TCS is a bit field of MCSPI_CH(i)CONF register). Note: P = SPI_CLK clock period.
- (5) B = (TCS + 0.5) × TSPICLKREF × Fratio (TCS is a bit field of MCSPI_CH(i)CONF register, Fratio: Even ≥ 2).



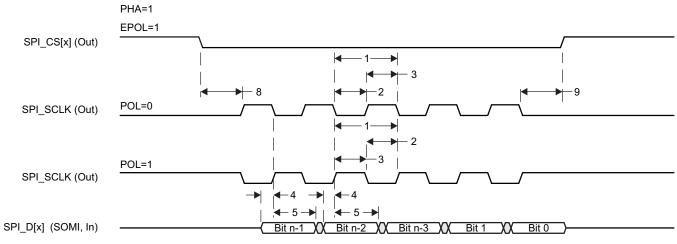
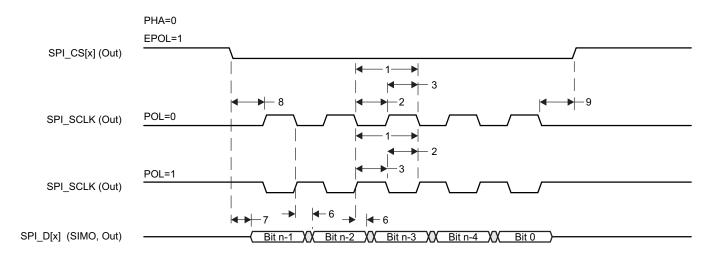


图 7-90. SPI Master Mode Receive Timing



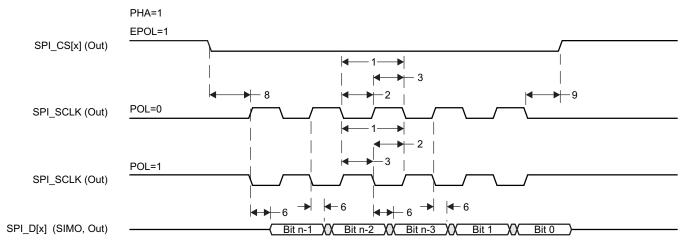


图 7-91. SPI Master Mode Transmit Timing

www.ti.com.cn

7.13 Multimedia Card (MMC) Interface

For more information, see the Multimedia Card (MMC) section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.

7.13.1 MMC Electrical Data and Timing

表 7-86. MMC Timing Conditions

	PARAMETER	MIN	TYP MAX	UNIT
Input Cor	nditions			
t _r	Input signal rise time	1	5	ns
t _f	Input signal fall time	1	5	ns
Output C	ondition			•
C _{load}	Output load capacitance	3	30	pF

表 7-87. Timing Requirements for MMC[x]_CMD and MMC[x]_DAT[7:0]

(see **3 7-92**)

NO					1.8	-V MOE	ÞΕ	3.3	-V MOD	ÞΕ	LINUT
NO.					MIN	TYP	MAX	MIN	TYP	MAX	UNIT
1	t _{su(CMDV-CLKH)}	Setup time, MMC_CMD valid before MMC_	CLK rising clock edge		4.1			4.1			ns
		Hald time MMC CMD valid after	Industrial extended temperature (-40°C to 125°C)	MMC0-2	3.76			3.76			
2	$t_{h(CLKH-CMDV)}$	Hold time, MMC_CMD valid after MMC_CLK rising clock edge		MMC0	3.76			2.52			ns
			All other temperature ranges	MMC1	3.76			3.03			
			tomporature ranges	MMC2	3.76			3.0			
3	t _{su(DATV-CLKH)}	Setup time, MMC_DATx valid before MMC_	_CLK rising clock edge		4.1			4.1			ns
		Hald fire MMC DATive lide from	Industrial extended temperature (-40°C to 125°C)	MMC0-2	3.76			3.76			
4	t _{h(CLKH-DATV)}	Hold time, MMC_DATx valid after MMC_CLK rising clock edge	All other	MMC0	3.76			2.52			ns
				MMC1	3.76			3.03			
			lopo.a.aro rangoo	MMC2	3.76			3.0			

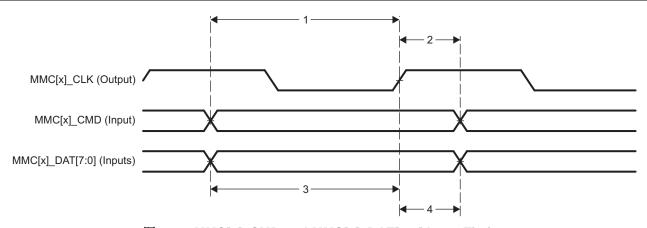


图 7-92. MMC[x]_CMD and MMC[x]_DAT[7:0] Input Timing



表 7-88. Switching Characteristics for MMC[x]_CLK

(see 图 7-93)

NO.		PARAMETER	STANDARD MODE	HIGH-SPEED MODE	UNIT
NO.		FARAINETER	MIN TYP MA	MIN TYP MAX	
	$f_{ m op(CLK)}$	Operating frequency, MMC_CLK	2	4 48	MHz
5	t _{cop(CLK)}	Operating period: MMC_CLK	41.7	20.8	ns
5	f _{id(CLK)}	Identification mode frequency, MMC_CLK	40	0 400	kHz
	t _{cid(CLK)}	Identification mode period: MMC_CLK	2500	2500	ns
6	t _{w(CLKL)}	Pulse duration, MMC_CLK low	$(0.5 \times P) - t_{f(CLK)}^{(1)}$	$ (0.5 \times P) - t_{f(CLK)}(1) $	ns
7	t _{w(CLKH)}	Pulse duration, MMC_CLK high	$ (0.5 \times P) - t_{r(CLK)}^{(1)} $	$(0.5 \times P) - t_{r(CLK)}^{(1)}$	ns
8	t _{r(CLK)}	Rise time, all signals (10% to 90%)	2.	2.2	ns
9	t _{f(CLK)} Fall time, all signals (10% to 90%)		2.	2.2	ns

(1) P = MMC_CLK period

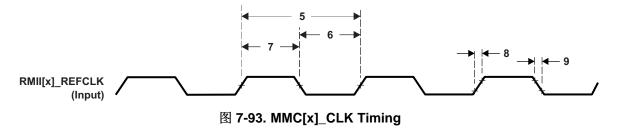


表 7-89. Switching Characteristics for MMC[x]_CMD and MMC[x]_DAT[7:0]—Standard Mode

(see 图 7-94)

<u> </u>	,								
NO	NO. PARAMETER		OPP100				OPP50		LINUT
NO.		PARAMETER	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
10	t _{d(CLKL-CMD)}	Delay time, MMC_CLK falling clock edge to MMC_CMD transition	-4		14	-4		17.5	ns
11	t _{d(CLKL-DAT)}	Delay time, MMC_CLK falling clock edge to MMC_DATx transition	-4		14	-4		17.5	ns

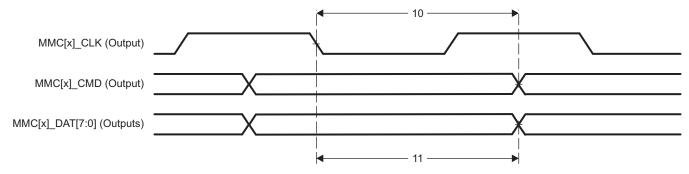


图 7-94. MMC[x]_CMD and MMC[x]_DAT[7:0] Output Timing—Standard Mode

www.ti.com.cn

表 7-90. Switching Characteristics for MMC[x]_CMD and MMC[x]_DAT[7:0]—High-Speed Mode

(see 图 7-95)

NO. PARAMETER		DADAMETED	OPP100			OPP50			UNIT
NO.	NO. PARAMETER		MIN	TYP	MAX	MIN	TYP	MAX	UNIT
12	t _{d(CLKL} -	Delay time, MMC_CLK rising clock edge to MMC_CMD transition	3		14	3		17.5	ns
13	t _{d(CLKL-DAT)}	Delay time, MMC_CLK rising clock edge to MMC_DATx transition	3		14	3		17.5	ns

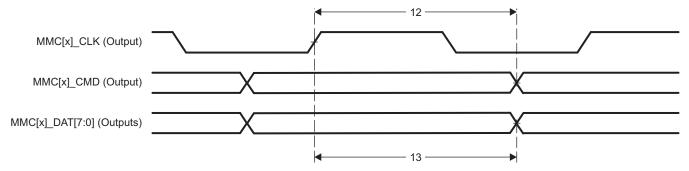


图 7-95. MMC[x]_CMD and MMC[x]_DAT[7:0] Output Timing—High-Speed Mode



7.14 Programmable Real-Time Unit Subsystem and Industrial Communication Subsystem (PRU-ICSS)

For more information, see the Programmable Real-Time Unit Subsystem and Industrial Communication Subsystem Interface (PRU-ICSS) section of the *AM335x and AMIC110 Sitara Processors Technical Reference Manual*.

7.14.1 Programmable Real-Time Unit (PRU-ICSS PRU)

表 7-91. PRU-ICSS PRU Timing Conditions

PARAMETER MIN						
Output Condition						
C _{load} Capacitive load for each bus line		30	pF			

7.14.1.1 PRU-ICSS PRU Direct Input/Output Mode Electrical Data and Timing

表 7-92. PRU-ICSS PRU Timing Requirements - Direct Input Mode

(see **图 7-96**)

NO.				MIN	MAX	UNIT
1	t _{w(GPI)}	Pulse width, GPI		2 x P ⁽¹⁾		ns
2	t _{r(GPI)}	Rise time, GPI		1.00	3.00	ns
2	t _{f(GPI)}	Fall time, GPI		1.00	3.00	ns
0		leternal alexa hatuara CDIII-01 eigenla(2)	PRU0		1.00	
3	3 t _{sk(GPI)} Internal skew between GPI[n:0] signals ⁽²⁾		PRU1		3.00	ns

- (1) P = L3_CLK (PRU-ICSS ocp clock) period.
- (2) n = 16

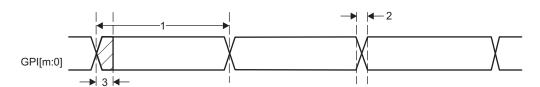


图 7-96. PRU-ICSS PRU Direct Input Timing

表 7-93. PRU-ICSS PRU Switching Requirements – Direct Output Mode

(see \bigseparty 7-69)

NO.		PARAMETER		MIN	MAX	UNIT
1	t _{w(GPO)}	Pulse width, GPO	2 x P ⁽¹⁾		ns	
2	t _{r(GPO)}	Rise time, GPO		1.00	3.00	ns
2	t _{f(GPO)}	Fall time, GPO		1.00	3.00	ns
2		Laternal allow hat was a CDO(a-0) simple (2)	PRU0		1.00	
3	t _{sk} (GPO)	Internal skew between GPO[n:0] signals ⁽²⁾	PRU1		5.00	ns

- (1) P = L3_CLK (PRU-ICSS ocp clock) period
- (2) n = 15

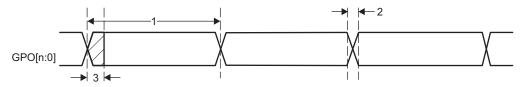


图 7-97. PRU-ICSS PRU Direct Output Timing

7.14.1.2 PRU-ICSS PRU Parallel Capture Mode Electrical Data and Timing

表 7-94. PRU-ICSS PRU Timing Requirements - Parallel Capture Mode

(see 图 7-98 and 图 7-99)

NO.			MIN	MAX	UNIT
1	t _{c(CLOCKIN)}	Cycle time, CLOCKIN	20.00		ns
2	t _{w(CLOCKIN_L)}	Pulse duration, CLOCKIN low	10.00		ns
3	t _{w(CLOCKIN_H)}	Pulse duration, CLOCKIN high	10.00		ns
4	t _{r(CLOCKIN)}	Rising time, CLOCKIN	1.00	3.00	ns
5	t _{f(CLOCKIN)}	Falling time, CLOCKIN	1.00	3.00	ns
6	t _{su(DATAIN-CLOCKIN)}	Setup time, DATAIN valid before CLOCKIN	5.00		ns
7	t _{h(CLOCKIN-DATAIN)}	Hold time, DATAIN valid after CLOCKIN	0.00		ns
8	t _{r(DATAIN)}	Rising time, DATAIN	1.00	3.00	ns
0	t _{f(DATAIN)}	Falling time, DATAIN	1.00	3.00	ns

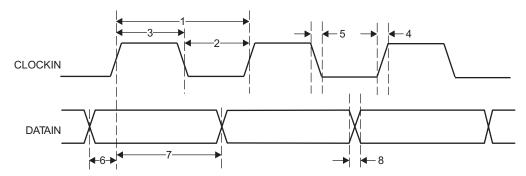


图 7-98. PRU-ICSS PRU Parallel Capture Timing - Rising Edge Mode

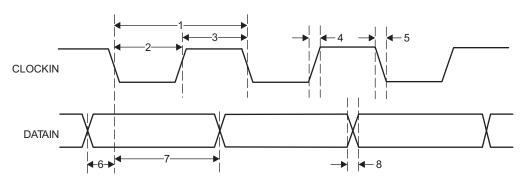


图 7-99. PRU-ICSS PRU Parallel Capture Timing - Falling Edge Mode

7.14.1.3 PRU-ICSS PRU Shift Mode Electrical Data and Timing

表 7-95. PRU-ICSS PRU Timing Requirements - Shift In Mode

(see 图 7-100)

NO.			MIN	MAX	UNIT
1	t _{c(DATAIN)}	Cycle time, DATAIN	10.00		ns
2	t _{w(DATAIN)}	Pulse width, DATAIN	$0.45 \times P^{(1)}$	$0.55 \times P^{(1)}$	ns
3	t _{r(DATAIN)}	Rising time, DATAIN	1.00	3.00	ns
4	t _{f(DATAIN)}	Falling time, DATAIN	1.00	3.00	ns

(1) P = L3_CLK (PRU-ICSS ocp clock) period.



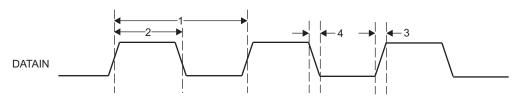


图 7-100. PRU-ICSS PRU Shift In Timing

表 7-96. PRU-ICSS PRU Switching Requirements - Shift Out Mode

(see 图 7-101)

NO.			MIN	MAX	UNIT
1	t _{c(CLOCKOUT)}	Cycle time, CLOCKOUT	10.00		ns
2	t _{w(CLOCKOUT)}	Pulse width, CLOCKOUT	$0.45 \times P^{(1)}$	$0.55 \times P^{(1)}$	ns
3	t _{r(CLOCKOUT)}	Rising time, CLOCKOUT	1.00	3.00	ns
4	t _{f(CLOCKOUT)}	Falling time, CLOCKOUT	1.00	3.00	ns
5	t _d (CLOCKOUT-DATAOUT)	Delay time, CLOCKOUT to DATAOUT valid	0.00	3.00	ns
6	t _{r(DATAOUT)}	Rising time, DATAOUT	1.00	3.00	ns
6	t _{f(DATAOUT)}	Falling time, DATAOUT	1.00	3.00	ns

(1) P = L3_CLK (PRU-ICSS ocp clock) period.

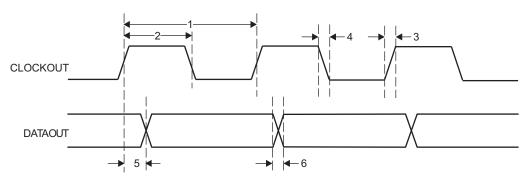


图 7-101. PRU-ICSS PRU Shift Out Timing

7.14.2 PRU-ICSS EtherCAT (PRU-ICSS ECAT)

表 7-97. PRU-ICSS ECAT Timing Conditions

	PARAMETER	MIN MAX	UNIT			
Output Co	Output Condition					
C _{load}	Capacitive load for each bus line	30	pF			

7.14.2.1 PRU-ICSS ECAT Electrical Data and Timing

表 7-98. PRU-ICSS ECAT Timing Requirements – Input Validated With LATCH_IN

(see 图 7-102)

(000	1 102)				
NO.			MIN	MAX	UNIT
1	t _{w(EDIO_LATCH_IN)}	Pulse width, EDIO_LATCH_IN	100.00		ns
2	t _{r(EDIO_LATCH_IN)}	Rising time, EDIO_LATCH_IN	1.00	3.00	ns
3	t _{f(EDIO_LATCH_IN)}	Falling time, EDIO_LATCH_IN	1.00	3.00	ns
4	t _{su(EDIO_DATA_IN-} EDIO_LATCH_IN)	Setup time, EDIO_DATA_IN valid before EDIO_LATCH_IN active edge	20.00		ns
5	t _h (EDIO_LATCH_IN- EDIO_DATA_IN)	Hold time, EDIO_DATA_IN valid after EDIO_LATCH_IN active edge	20.00		ns

表 7-98. PRU-ICSS ECAT Timing Requirements – Input Validated With LATCH_IN (continued)

(see 图 7-102)

NO.			MIN	MAX	UNIT
6	t _{r(EDIO_DATA_IN)}	Rising time, EDIO_DATA_IN	1.00	3.00	ns
6	t _{f(EDIO_DATA_IN)}	Falling time, EDIO_DATA_IN	1.00	3.00	ns

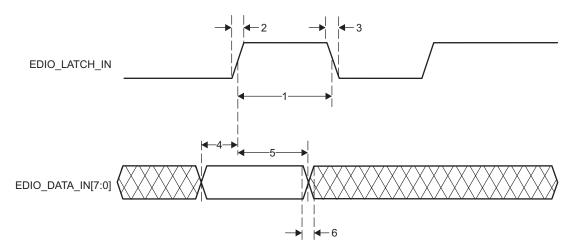


图 7-102. PRU-ICSS ECAT Input Validated With LATCH_IN Timing

表 7-99. PRU-ICSS ECAT Timing Requirements – Input Validated With SYNCx

(see 图 7-103)

NO.			MIN	MAX	UNIT
1	t _{w(EDC_SYNCx_OUT)}	Pulse width, EDC_SYNCx_OUT	100.00		ns
2	t _{r(EDC_SYNCx_OUT)}	Rising time, EDC_SYNCx_OUT	1.00	3.00	ns
3	t _{f(EDC_SYNCx_OUT)}	Falling time, EDC_SYNCx_OUT	1.00	3.00	ns
4	t _{su(EDIO_DATA_IN-} EDC_SYNCx_OUT)	Setup time, EDIO_DATA_IN valid before EDC_SYNCx_OUT active edge	20.00		ns
5	t _{h(EDC_SYNCx_OUT-EDIO_DATA_IN)}	Hold time, EDIO_DATA_IN valid after EDC_SYNCx_OUT active edge	20.00		ns
6	t _{r(EDIO_DATA_IN)}	Rising time, EDIO_DATA_IN	1.00	3.00	ns
6	t _{f(EDIO_DATA_IN)}	Falling time, EDIO_DATA_IN	1.00	3.00	ns

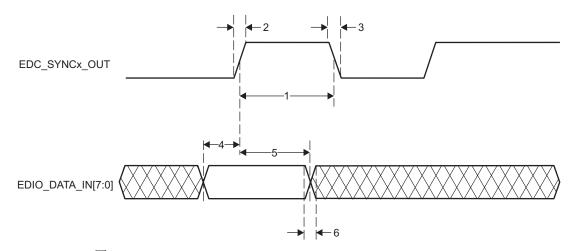


图 7-103. PRU-ICSS ECAT Input Validated With SYNCx Timing



表 7-100. PRU-ICSS ECAT Timing Requirements – Input Validated With Start of Frame (SOF)

(see 图 7-104)

NO.			MIN	MAX	UNIT
1	t _{w(EDIO_SOF)}	Pulse duration, EDIO_SOF	4 x P ⁽¹⁾	$5 \times P^{(1)}$	ns
2	t _{r(EDIO_SOF)}	Rising time, EDIO_SOF	1.00	3.00	ns
3	t _{f(EDIO_SOF)}	Falling time, EDIO_SOF	1.00	3.00	ns
4	t _{su(EDIO_DATA_IN-} EDIO_SOF)	Setup time, EDIO_DATA_IN valid before EDIO_SOF active edge	20.00		ns
5	th(EDIO_SOF-EDIO_DATA_IN)	Hold time, EDIO_DATA_IN valid after EDIO_SOF active edge	20.00		ns
6	t _{r(EDIO_DATA_IN)}	Rising time, EDIO_DATA_IN	1.00	3.00	ns
6	t _{f(EDIO_DATA_IN)}	Falling time, EDIO_DATA_IN	1.00	3.00	ns

(1) P = PRU-ICSS IEP clock source period.

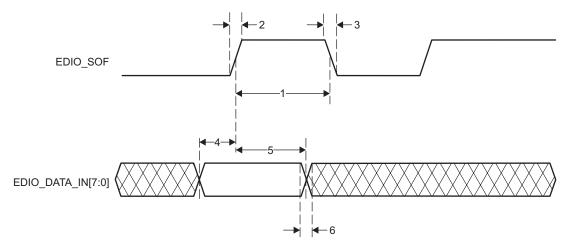


图 7-104. PRU-ICSS ECAT Input Validated With SOF

表 7-101. PRU-ICSS ECAT Timing Requirements - LATCHx_IN

(see 图 7-105)

	,				
NO.			MIN	MAX	UNIT
1	t _{w(EDC_LATCHx_IN)}	Pulse duration, EDC_LATCHx_IN	3 x P ⁽¹⁾		ns
2	t _{r(EDC_LATCHx_IN)}	Rising time, EDC_LATCHx_IN	1.00	3.00	ns
3	t _{f(EDC LATCHx IN)}	Falling time, EDC_LATCHx_IN	1.00	3.00	ns

(1) P = PRU-ICSS IEP clock source period.

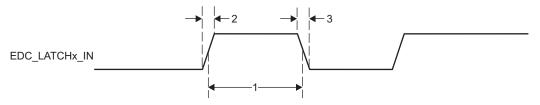


图 7-105. PRU-ICSS ECAT LATCHx_IN Timing



表 7-102. PRU-ICSS ECAT Switching Requirements - Digital I/Os

NO.		PARAMETER		MAX	UNIT
1	t _{w(EDIO_OUTVALID)}	Pulse duration, EDIO_OUTVALID	$14 \times P^{(1)}$	$32 \times P^{(1)}$	ns
2	t _{r(EDIO_OUTVALID)}	Rising time, EDIO_OUTVALID	1.00	3.00	ns
3	t _{f(EDIO_OUTVALID)}	Falling time, EDIO_OUTVALID	1.00	3.00	ns
4	t _{d(EDIO_OUTVALID-} EDIO_DATA_OUT)	Delay time, EDIO_OUTVALID to EDIO_DATA_OUT	0.00	18 × P ⁽¹⁾	ns
5	t _{r(EDIO_DATA_OUT)}	Rising time, EDIO_DATA_OUT	1.00	3.00	ns
6	t _{f(EDIO_DATA_OUT)}	Falling time, EDIO_DATA_OUT	1.00	3.00	ns
7	t _{sk(EDIO_DATA_OUT)}	EDIO_DATA_OUT skew		8.00	ns

⁽¹⁾ P = PRU-ICSS IEP clock source period.

7.14.3 PRU-ICSS MII_RT and Switch

表 7-103. PRU-ICSS MII_RT Switch Timing Conditions

	PARAMETER	MIN	TYP MAX	UNIT				
Input Conditi	ons							
t _R	Input signal rise time	1 ⁽¹⁾	3 ⁽¹⁾	ns				
t _F	Input signal fall time	1 ⁽¹⁾	3 ⁽¹⁾	ns				
Output Cond	Output Condition							
C _{LOAD}	Output load capacitance	3	20	pF				

⁽¹⁾ Except when specified otherwise.

7.14.3.1 PRU-ICSS MDIO Electrical Data and Timing

表 7-104. PRU-ICSS MDIO Timing Requirements – MDIO_DATA

(see 图 7-106)

NO.			MIN	TYP	MAX	UNIT
1	t _{su(MDIO-MDC)}	Setup time, MDIO valid before MDC high	90			ns
2	t _{h(MDIO-MDC)}	Hold time, MDIO valid from MDC high	0			ns

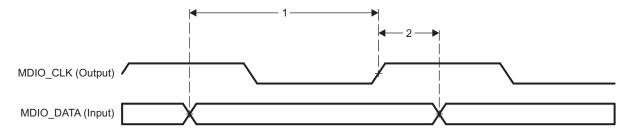


图 7-106. PRU-ICSS MDIO_DATA Timing - Input Mode

表 7-105. PRU-ICSS MDIO Switching Characteristics - MDIO_CLK

(see 图 7-107)

(000 🖂	. 101)					
NO.		PARAMETER		TYP	MAX	UNIT
1	t _{c(MDC)}	Cycle time, MDC	400			ns
2	t _{w(MDCH)}	Pulse duration, MDC high	160			ns
3	t _{w(MDCL)}	Pulse duration, MDC low	160			ns
4	t _{t(MDC)}	Transition time, MDC			5	ns



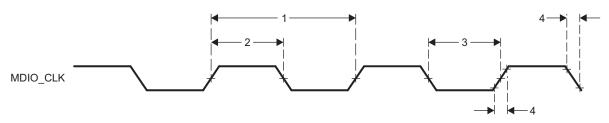
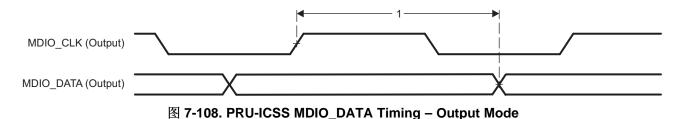


图 7-107. PRU-ICSS MDIO_CLK Timing

表 7-106. PRU-ICSS MDIO Switching Characteristics - MDIO_DATA

(see **图 7-108**)

NO.			MIN	TYP MAX	UNIT
1	t _{d(MDC-MDIO)}	Delay time, MDC high to MDIO valid	10	390	ns



7.14.3.2 PRU-ICSS MII_RT Electrical Data and Timing

注

In order to guarantee the MII_RT I/O timing values published in the device data manual, the PRU ocp_clk clock must be configured for 200 MHz (default value) and the TX_CLK_DELAY bit field in the PRUSS_MII_RT_TXCFG0/1 register must be set to 6 h (non-default value).

表 7-107. PRU-ICSS MII_RT Timing Requirements – MII_RXCLK

(see **§** 7-109)

NO	NO.		1	10 Mbps			100 Mbps		
NO.			MIN	TYP	MAX	MIN	TYP	MAX	UNIT
1	t _{c(RX_CLK)}	Cycle time, RX_CLK	399.96		400.04	39.996		40.004	ns
2	t _{w(RX_CLKH)}	Pulse duration, RX_CLK high	140		260	14		26	ns
3	t _{w(RX_CLKL)}	Pulse duration, RX_CLK low	140		260	14		26	ns
4	t _{t(RX CLK)}	Transition time, RX_CLK			3			3	ns

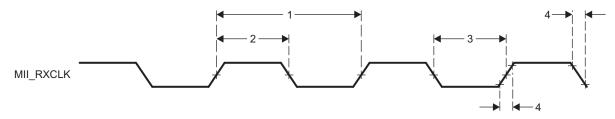


图 7-109. PRU-ICSS MII_RXCLK Timing

www.ti.com.cn

表 7-108. PRU-ICSS MII_RT Timing Requirements - MII[x]_TXCLK

(see 图 7-110)

NO.			10 Mbps		100 Mbps			UNIT	
NO.			MIN	TYP	MAX	MIN	TYP	MAX	UNII
1	t _{c(TX_CLK)}	Cycle time, TX_CLK	399.96		400.04	39.996		40.004	ns
2	t _{w(TX_CLKH)}	Pulse duration, TX_CLK high	140		260	14		26	ns
3	t _{w(TX_CLKL)}	Pulse duration, TX_CLK low	140		260	14		26	ns
4	t _{t(TX CLK)}	Transition time, TX_CLK			3			3	ns

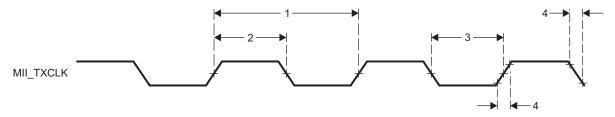


图 7-110. PRU-ICSS MII_TXCLK Timing

表 7-109. PRU-ICSS MII_RT Timing Requirements - MII_RXD[3:0], MII_RXDV, and MII_RXER

(see 图 7-111)

NO.				10 Mbps			0 Mbps		UNIT
NO.			MIN	TYP	MAX	MIN	TYP	MAX	UNII
	t _{su(RXD-RX_CLK)}	Setup time, RXD[3:0] valid before RX_CLK							
1	t _{su(RX_DV-RX_CLK)}	Setup time, RX_DV valid before RX_CLK	8			8			ns
	t _{su(RX_ER-RX_CLK)}	Setup time, RX_ER valid before RX_CLK							
	t _{h(RX_CLK-RXD)}	Hold time RXD[3:0] valid after RX_CLK							
2	t _{h(RX_CLK-RX_DV)}	Hold time RX_DV valid after RX_CLK	8			8			ns
	t _{h(RX CLK-RX ER)}	Hold time RX_ER valid after RX_CLK							

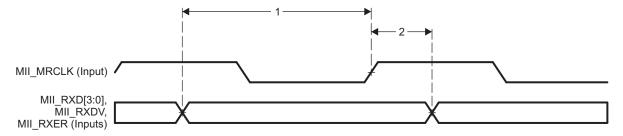


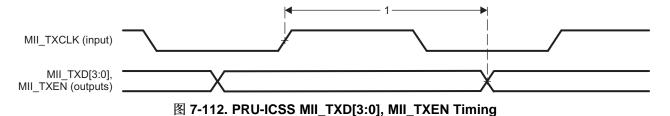
图 7-111. PRU-ICSS MII_RXD[3:0], MII_RXDV, and MII_RXER Timing



表 7-110. PRU-ICSS MII_RT Switching Characteristics - MII_TXD[3:0] and MII_TXEN

(see 图 7-112)

NO			10 Mbps			100 Mbps			UNIT
•			MIN	TYP	MAX	MIN	TYP	MAX	UNIT
1	t _{d(TX_CLK-TXD)}	Delay time, TX_CLK high to TXD[3:0] valid	E		25	-		25	
ļ	t _{d(TX_CLK-TX_EN)}	Delay time, TX_CLK to TX_EN valid	5		25	Э		25	ns



7.14.4 PRU-ICSS Universal Asynchronous Receiver Transmitter (PRU-ICSS UART)

表 7-111. Timing Requirements for PRU-ICSS UART Receive

(see 图 7-113)

NO.			MIN	MAX	UNIT
3	t _{w(RX)}	Pulse duration, receive start, stop, data bit	0.96U ⁽¹⁾	1.05U ⁽¹⁾	ns

(1) U = UART baud time = 1/programmed baud rate.

表 7-112. Switching Characteristics Over Recommended Operating Conditions for PRU-ICSS UART Transmit

(see 图 7-113)

NO.		PARAMETER	MIN	MAX	UNIT
1	$f_{baud(baud)}$	Maximum programmable baud rate	0	12	MHz
2	t _{w(TX)}	Pulse duration, transmit start, stop, data bit	U - 2 ⁽¹⁾	$U + 2^{(1)}$	ns

(1) U = UART baud time = 1/programmed baud rate.

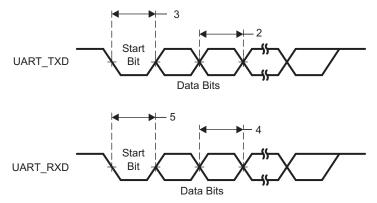


图 7-113. PRU-ICSS UART Timing

7.15 Universal Asynchronous Receiver Transmitter (UART)

For more information, see the Universal Asynchronous Receiver Transmitter (UART) section of the AM335x and AMIC110 Sitara Processors Technical Reference Manual.

7.15.1 UART Electrical Data and Timing

表 7-113. Timing Requirements for UARTx Receive

(see **§** 7-114)

NO.			MIN	MAX	UNIT
3	t _{w(RX)}	Pulse duration, receive start, stop, data bit	0.96U ⁽¹⁾	1.05U ⁽¹⁾	ns

⁽¹⁾ U = UART baud time = 1/programmed baud rate.

表 7-114. Switching Characteristics for UARTx Transmit

(see 图 7-114)

NO.		PARAMETER	MIN	MAX	UNIT
1	$f_{baud(baud)}$	Maximum programmable baud rate		3.6864	MHz
2	t _{w(TX)}	Pulse duration, transmit start, stop, data bit	U – 2 ⁽¹⁾	U + 2 ⁽¹⁾	ns

(1) U = UART baud time = 1 / programmed baud rate

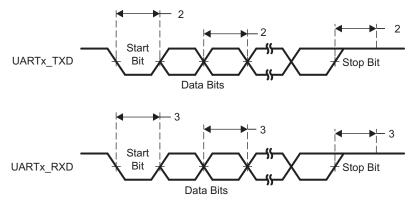


图 7-114. UART Timings



7.15.2 UART IrDA Interface

The IrDA module operates in three different modes:

- Slow infrared (SIR) (≤115.2 kbps)
- Medium infrared (MIR) (0.576 Mbps and 1.152 Mbps)
- Fast infrared (FIR) (4 Mbps).

图 7-115 shows the UART IrDA pulse parameters. 表 7-115 and 表 7-116 list the signaling rates and pulse durations for UART IrDA receive and transmit modes.

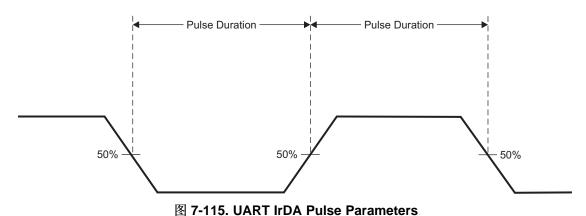


表 7-115. UART IrDA—Signaling Rate and Pulse Duration—Receive Mode

SIGNALING RATE	ELECTRICAL PULSE D	URATION	UNIT
SIGNALING RATE	MIN	MAX	UNII
SIR			
2.4 kbps	1.41	88.55	μs
9.6 kbps	1.41	22.13	μs
19.2 kbps	1.41	11.07	μs
38.4 kbps	1.41	5.96	μs
57.6 kbps	1.41	4.34	μs
115.2 kbps	1.41	2.23	μs
MIR			
0.576 Mbps	297.2	518.8	ns
1.152 Mbps	149.6	258.4	ns
FIR	·		
4 Mbps (single pulse)	67	164	ns
4 Mbps (double pulse)	190	289	ns

表 7-116. UART IrDA—Signaling Rate and Pulse Duration—Transmit Mode

SIGNALING RATE	ELECTRICAL PULSE DU	RATION	UNIT
SIGNALING RATE	MIN	MAX	UNII
SIR			
2.4 kbps	78.1	78.1	μs
9.6 kbps	19.5	19.5	μs
19.2 kbps	9.75	9.75	μs
38.4 kbps	4.87	4.87	μs
57.6 kbps	3.25	3.25	μs
115.2 kbps	1.62	1.62	μs
MIR			
0.576 Mbps	414	419	ns
1.152 Mbps	206	211	ns
FIR			
4 Mbps (single pulse)	123	128	ns
4 Mbps (double pulse)	248	253	ns



8 Device and Documentation Support

8.1 Device Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all microprocessors (MPUs) and support tools. Each device has one of three prefixes: X, P, or null (no prefix) (for example, XAM3358AZCE). Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMDX) through fully qualified production devices and tools (TMDS).

Device development evolutionary flow:

X Experimental device that is not necessarily representative of the final device's electrical specifications and may not use production assembly flow.

P Prototype device that is not necessarily the final silicon die and may not necessarily meet final electrical specifications.

null Production version of the silicon die that is fully qualified.

Support tool development evolutionary flow:

TMDX Development-support product that has not yet completed Texas Instruments internal

qualification testing.

TMDS Fully-qualified development-support product.

X and P devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

Production devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. Tl's standard warranty applies.

Predictions show that prototype devices (X or P) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

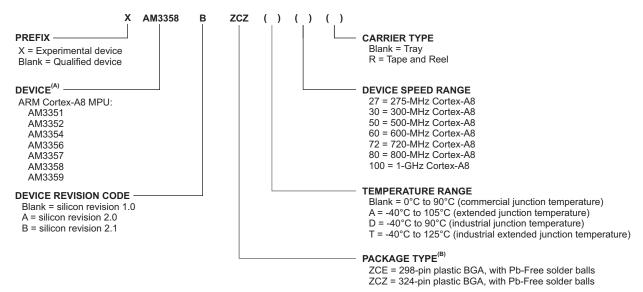
TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, ZCE), the temperature range (for example, blank is the default commercial temperature range), and the device speed range, in megahertz (for example, 27 is 275 MHz).

8-1 provides a legend for reading the complete device name for any AM335x device.

For orderable part numbers of AM335x devices in the ZCE and ZCZ package types, see the Package Option Addendum of this document, ti.com, or contact your TI sales representative.

For additional description of the device nomenclature markings on the die, see the *AM335x Sitara Processors Silicon Errata*.





- A. The AM3358 device shown in this device nomenclature example is one of several valid part numbers for the AM335x family of devices. For orderable device part numbers, see the Package Option Addendum of this document.
- B. BGA = Ball grid array

图 8-1. AM335x Device Nomenclature

8.2 Tools and Software

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

Design Kits and Evaluation Modules

- AM335x Evaluation Module Enables developers to immediately start evaluating the AM335x processor family (AM3351, AM3352, AM3354, AM3356, AM3358) and begin building applications such as portable navigation, portable gaming, home/building automation and others.
- AM335x Starter Kit Provides a stable and affordable platform to quickly start evaluation of Sitara ARM Cortex-A8 AM335x Processors (AM3351, AM3352, AM3354, AM3356, AM3358) and accelerate development for smart appliance, industrial and networking applications. It is a low-cost development platform based on the ARM Cortex-A8 processor that is integrated with options such as Dual Gigabit Ethernet, DDR3 and LCD touch screen.
- BeagleBone Black Development Board Low-cost, open source, community-supported development platform for ARM Cortex-A8 processor developers and hobbyists. Boot Linux in under 10-seconds and get started on Sitara AM335x ARM Cortex-A8 processor development in less than 5 minutes with just a single USB cable.
- BeagleBone Development Board Low-cost, community-supported development platform for ARM Cortex-A8 processor developers. Boot Linux in under 10-seconds and get started on Sitara AM335x ARM Cortex-A8 processor development in less than 5 minutes with just a single USB cable. For TI-supported hardware platforms, consider the Sitara ARM AM335x Starter Kit or AM335x Evaluation Module.
- Data Concentrator Evaluation Module Based on AM3359 as the main processor and has Power Line Communication (PLC) Module to support various OFDM PLC communication standards. TMDSDC3359 also has capability to support multiple interfaces, sub-1GHz and 2.4GHz RF, Ethernet, RS-232, and RS-485. This evaluation module is ideal development platform for smart grid infrastructure applications including data concentrator, convergent node of grid sensor network, and control equipment of power automation.
- WiLink™ 8 Dual Band 2.4 & 5 GHz Wi-Fi® + Bluetooth® COM8 Evaluation Module Enables customers to add both Wi-Fi and Bluetooth to home and building automation, smart energy, gateways, wireless audio, enterprise, wearables and many more industrial and Internet of Things (IoT) applications. TI's WiLink 8 modules are certified and offer high throughput and extended range along with Wi-Fi and Bluetooth coexistence in a power-optimized design. Drivers for the Linux and Android high-level operating systems (HLOSs) are available free of charge



from TI for the Sitara AM335x microprocessor (Linux and Android version restrictions apply).

WiLink 8 Module 2.4 GHz WiFi + Bluetooth COM8 Evaluation Module Enables customers to add Wi-Fi and Bluetooth (WL183x module only) to embedded applications based on TI's Sitara microprocessors. Tl's WiLink 8 Wi-Fi + Bluetooth modules are pre-certified and offer high throughput and extended range along with Wi-Fi and Bluetooth coexistence (WL183x modules only) in a power-optimized design. Drivers for the Linux and Android high-level operating systems (HLOSs) are available free of charge from TI for the Sitara AM335x microprocessor (Linux and Android version restrictions apply).

TI Designs

STRUMENTS

- EtherCAT Communications Development Platform Allows designers to implement real-time EtherCAT communications standards in a broad range of industrial automation equipment. It enables low foot print designs in applications such as industrial automation, factory automation or industrial communication with minimal external components and with best in class low power performance.
- PROFIBUS Communications Development Platform Allows designers to implement PROFIBUS communications standards in a broad range of industrial automation equipment. It enables low foot print designs in applications such as industrial automation, factory automation or industrial communication with minimal external components and with best in class low power performance.
- Ethernet/IP Communications Development Platform Allows designers to mplement Ethernet/IP communications standards in a broad range of industrial automation equipment. It enables low foot print designs in applications such as industrial automation, factory automation or industrial communication with minimal external components and with best in class low power performance.
- Acontis EtherCAT Master Stack Reference Design Highly portable software stack that can be used on various embedded platforms. The EC-Master supports the high performane TI Sitara MPUs. it provides a sophisticated EtherCAT Master solution which customers can use to implement EtherCAT communication interface boards, EtherCAT based PLC or motion control applications. The EC-Master architectural design does not require additional tasks to be scheduled, thus the full stack functionality is available even on an OS less platform such as TI Starterware suported on AM335x. Due to this architecture combined with the high speed Ethernet driver it is possible to implement EtherCAT master based applications on the Sitara platform with short cycle times of 100 microseconds or even below.
- Solar Inverter Gateway Development Platform Reference Design Adds communication functions to solar energy generation systems to enable system monitoring, real-time feedback, system updates, and more. The TIDEP0044 reference design describes the implementation of a solar inverter gateway using display, Ethernet, USB, and CAN on the TMDXEVM3358 featuring TI's AM335x processor.
- G3 Power Line Communications Data Concentrator on BeagleBone Black Platform simplified approach for evaluating G3-PLC utilizing Beagle Bone Black powered by the Sitara AM335x processor. Users can establish a G3-PLC network with one service node. Single phase coupling is supported.
- IEC 61850 Demonstration of Substation Bay Controller on Beaglebone Cape and Starter Kit Low-cost, simplified implementation of an IEC 61850 Substation Bay Controller is demonstrated by running the Triangle MicroWorks IEC 61850 stack efficiently on the TI AM335X platform with a Linux target layer definition. Many different substation automation applications can be built on top of the AM335X platform and 61850 stack demonstration.



- PRU Real-Time I/O Evaluation Reference Design BeagleBone Black add-on board that allows users get to know Tl's powerful Programmable Real-Time Unit (PRU) core and basic functionality. The PRU is a low-latency microcontroller subsystem integrated in the Sitara AM335x and AM437x family of devices. The PRU core is optimized for deterministic, real-time processing, direct access to I/Os and ultra-low-latency requirements. With LEDs and push buttons for GPIO, audio, a temp sensor, optional character display and more, this add-on board includes schematics, bill of materials (BOM), design files, and design guide to teach the basics of the PRU.
- Smart Home and Energy Gateway Reference Design Provides example implementation for measurement, management and communication of energy systems for smart homes and buildings. This example design is a bridge between different communication interfaces, such as WiFi, Ethernet, ZigBee or Bluetooth, that are commonly found in residential and commercial buildings. Since objects in the house and buildings are becoming more and more connected, the gateway design needs to be flexible to accommodate different RF standard, since no single RF standard is dominating the market. This example gateway addresses this problem by supporting existing legacy RF standards (WiFi, Bluetooth) and newer RF standards (ZigBee, BLE).
- Streaming Audio Reference Design Minimizes design time for customers by offering small form factor hardware and major software components, including streaming protocols and internet radio services. With this reference design, TI offers a quick and easy transition path to the AM335x and WiLink8 platform solution. This proven combo solution provides key advantages in this market category that helps bring your products to the next level.

Software

- Processor SDK for AM335X Sitara Processors Linux and TI-RTOS Support Unified software platform for TI embedded processors providing easy setup and fast out-of-the-box access to benchmarks and demos. All releases of Processor SDK are consistent across TI's broad portfolio, allowing developers to seamlessly reuse and migrate software across devices. Developing scalable platform solutions has never been easier than with the Processor SDK and TI's embedded processor solutions.
- **G3 Data Concentrator Power-Line Communication Modem** G3-PLC standard for narrowband OFDM Power Line Communications. The data concentrator solution is designed for the head-end systems which communicate with the end meters ("service node") in the neighborhood area network.
- PRIME Data Concentrator Power-Line Communication Modem PRIME standard for narrowband OFDM Power Line Communications. The data concentrator solution is designed for the head-end systems which communicate with the end meters ("service node") in the neighborhood area network.
- TI Dual-Mode Bluetooth Stack Comprised of Single-Mode and Dual-Mode offerings implementing the Bluetooth 4.0 specification. The Bluetooth stack is fully Bluetooth Special Interest Group (SIG) qualified, certified and royalty-free, provides simple command line sample applications to speed development, and upon request has MFI capability.
- Cryptography for TI Devices Enables encryption, crypto for TI devices. These files contain only cryptographic modules that were part of a TI software release. For the complete software release please search ti.com for your device part number, and download the Software Development Kit (SDK).



Development Tools

- Clock Tree Tool for Sitara ARM Processors Interactive clock tree configuration software that provides information about the clocks and modules in Sitara devices.
- Code Composer Studio (CCS) Integrated Development Environment (IDE) for Sitara ARM Processors

Integrated development environment (IDE) that supports TI's Microcontroller and Embedded Processors portfolio. Code Composer Studio comprises a suite of tools used to develop and debug embedded applications. It includes an optimizing C/C++ compiler, source code editor, project build environment, debugger, profiler, and many other features. The intuitive IDE provides a single user interface taking you through each step of the application development flow. Familiar tools and interfaces allow users to get started faster than ever before. Code Composer Studio combines the advantages of the Eclipse software framework with advanced embedded debug capabilities from TI resulting in a compelling feature-rich development environment for embedded developers.

- Pin Mux Tool Provides a Graphical User Interface for configuring pin multiplexing settings, resolving conflicts and specifying I/O cell characteristics for TI MPUs. Results are output as C header/code files that can be imported into software development kits (SDK) or used to configure customer's custom software. Version 3 of the Pin Mux utility adds the capability of automatically selecting a mux configuration that satisfies the entered requirements.
- Power Estimation Tool (PET) Provides users the ability to gain insight in to the power consumption of select TI processors. The tool includes the ability for the user to choose multiple application scenarios and understand the power consumption as well as how advanced power saving techniques can be applied to further reduce overall power consumption.
- Uniflash Standalone Flash Tool for TI Microcontrollers (MCU), Sitara Processors and SimpleLink devices

Programs on-chip flash memory on TI MCUs and onboard flash memory for Sitara processors. Uniflash has a GUI, command line, and scripting interface. CCS Uniflash is available free of charge.

- XDS200 USB Debug Probe Connects to the target board via a TI 20-pin connector (with multiple adapters for TI 14-pin, ARM 10-pin and ARM 20-pin) and to the host PC via USB2.0 High Speed (480Mbps). It also requires a license of Code Composer Studio IDE running on the host PC.
- XDS560v2 System Trace USB and Ethernet Debug Probe Adds system pin trace in its large external memory buffer. Available for selected TI devices, this external memory buffer captures device-level information that allows obtaining accurate bus performance activity and throughput, as well as power management of core and peripherals. Also, all XDS debug probes support Core and System Trace in all ARM and DSP processors that feature an Embedded Trace Buffer (ETB).
- XDS560v2 System Trace USB Debug Probe Adds system pin trace in its large external memory buffer. Available for selected TI devices, this external memory buffer captures device-level information that allows obtaining accurate bus performance activity and throughput, as well as power management of core and peripherals. Also, all XDS debug probes support Core and System Trace in all ARM and DSP processors that feature an Embedded Trace Buffer (ETB).



Models

AM335x ZCE IBIS Model ZCE package IBIS model

AM335x ZCZ IBIS Model ZCZ package IBIS model

AM335x ZCE Rev. 2.1 BSDL Model ZCE package BSDL model for the revision 2.1 TI F781962A Fixedand Floating-Point DSP with Boundary Scan

AM335x ZCZ Rev. 2.1 BSDL Model ZCZ package BSDL model for the revision 2.1 TI F781962A Fixedand Floating-Point DSP with Boundary Scan

8.3 **Documentation Support**

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

The current documentation that describes the processor, related peripherals, and other technical collateral is listed below.

Errata

AM335x Sitara Processors Silicon Errata

Describes the known exceptions to the functional specifications for the AM335x Sitara Processors.

Application Reports

Processor SDK RTOS Customization: Modifying Board Library to Change UART Instance on

Describes the procedure to modify the default UART0 example in the AM335x Processor SDK RTOS package to enable UART1. On the BeagleBone Black (BBB) P9 header, pins 24(TX) and 26(RX) are connected to UART1. This procedure shows a test to verify that UART1 is enabled on the BBB.

- High-Speed Layout Guidelines As modern bus interface frequencies scale higher, care must be taken in the printed circuit board (PCB) layout phase of a design to ensure a robust solution.
- AM335x Reliability Considerations in PLC Applications Programmable Logic Controllers (PLC) are used as the main control in an automation system with high-reliability expectations and long life in harsh environments. Processors used in these applications require an assessment of performance verses expected power on hours to achieve the optimal performance for the application.
- AM335x Thermal Considerations Discusses the thermal considerations of the AM335x devices. It offers guidance on analysis of the processor's thermal performance, suggests improvements for an end system to aid in overcoming some of the existing challenges of producing a good thermal design, and provides real power/thermal data measured with AM335x EVMs for user evaluation.

User's Guides

TPS65910Ax User's Guide for AM335x Processors User's Guide A reference for connectivity between the TPS65910Ax power-management integrated circuit (PMIC) and the AM335x processor.

AM335x and AMIC110 Sitara Processors Technical Reference Manual

Details the integration, the environment, the functional description, and the programming models for each peripheral and subsystem in the device.

G3 Power Line Communications Data Concentrator on BeagleBone Black Platform Design Guide Provide the foundation that you need including methodology, testing, and design files to quickly evaluate and customize the system. TI Designs help you accelerate your time to market.

- Powering the AM335x with the TPS65217x A reference for connectivity between the TPS65217 power management IC and the AM335x processor.
- Powering the AM335x With the TPS650250 Details a power solution for the AM335x application processor with a TPS650250 Power Management Unit (PMU) or Power Management IC (PMIC).

NSTRUMENTS

Selection and Solution Guides

Connected Sensors Building Automation Systems Guide The use of connected sensors has a wide range of uses in building automation applications, from monitoring human safety and security, controlling the environment and ambience specified by the comfort preferences of the end user, or either periodic or continuous data logging of environmental and system data to detect irregular system conditions.

White Papers

- **Building Automation for Enhanced Energy And Operational Efficiency** Discusses building automation solutions, focusing on aspects of the Building Control System. TI's Sitara processors facilitate intelligent automation of the control systems. The scalable Sitara processor portfolio offers an opportunity to build a platform solution that also spans beyond Building Control Systems.
- POWERLINK on TI Sitara Processors Supports Ethernet standard features such as cross-traffic, hotplugging and different types of network configurations such as star, ring and mixed topologies.
- EtherNet/IP on TI's Sitara AM335x Processors EtherNet/IP™ (EtherNet/Industrial Protocol) is an industrial automation networking protocol based on the IEEE 802.3 Ethernet standard that has dominated the world of IT networking for the past three decades.
- PROFINET on TI's Sitara AM335x Processors To integrate PROFINET into the Sitara AM335x processor, TI has built upon its programmable realtime unit (PRU) technology to create an industrial communication sub-system (ICSS).
- Profibus on AM335x and AM1810 Sitara ARM Microprocessor PROFIBUS, one of the most used communication technologies, is installed in more than 35 million industrial nodes worldwide and is growing at a rate of approximately 10 percent each year.
- EtherCAT on Sitara AM335x ARM Cortex-A8 Microprocessors Emerging real-time industrial Ethernet standard for industrial automation applications, such as input/output (I/O) devices, sensors and programmable logic controllers (PLCs).
- Mainline Linux Ensures Stability and Innovation Enabling and empowering the rapid development of new functionality starts at the foundational level of the system's software environment - that is, at the level of the Linux kernel – and builds upward from there.
- Complete Solutions for Next-Generation Wireless Connected Audio Robust, feature-rich and highperformance connectivity technology for Wi-Fi and Bluetooth.
- Data Concentrators: The Core of Energy and Data Management With a large install base, it is essential to establish an automated metering infrastructure (AMI). With automated meter reading (AMR) measurement, the communication of meter data to the central billing station will be seamless.
- Linaro Speeds Development in TI Linux SDKs Linaro's software is not a Linux distribution; in fact, it is distribution neutral. The focus of the organization's 120 engineers is on optimizing base-level open-source software in areas that interact directly with the silicon such as multimedia, graphics, power management, the Linux kernel and booting processes.
- Getting Started on TI ARM Embedded Processor Development Beginning with an overview of ARM technology and available processor platforms, this paper will then explore the fundamentals of embedded design that influence a system's architecture and, consequently, impact processor selection.
- Power Optimization Techniques for Energy-Efficient Systems The TI Sitara processor solutions offer the flexibility to design application-specific systems. The latest Sitara AM335x processors provide a scalable architecture with speed ranging from 300 MHz to 1 GHz.
- The Yocto Project: Changing the Way Embedded Linux Software Solutions are Developed Enabling complex silicon devices such as SoC with operating firmware and application software can be a challenge for equipment manufacturers who often are more comfortable with hardware than software issues.
- Smart Thermostats are a Cool Addition to the Connected Home Because of the pervasiveness of residential broadband connectivity and the explosion in options, the key to the connected home is – connectivity.
- BeagleBone Low-Cost Development Board Provides a Clear Path to Open-source Resources

241

Ready-to-use open-source hardware platform for rapid prototyping and firmware and software development.

Enable Security and Amp Up Chip Performance With Hardware-Accelerated Cryptography

Cryptography is one of several techniques or methodologies that are typically implemented in contemporary electronic systems to construct a secure perimeter around a device where information or digital content is being protected.

- Gesture Recognition: Enabling Natural Interactions With Electronics Enabling humans and machines to interface more easily in the home, the automobile, and at work.
- Developing Android Applications for ARM Cortex-A8 Cores The flexibility, power, versatility and ubiquity of the Android operating system (OS) and associated ecosystem have been a boon to developers of applications for ARM processor cores.

Other Documents

- Industrial Communication with Sitara AM335x ARM Cortex-A8 Microprocessors The industry's first low- power ARM Cortex-A8 devices to incorporate multiple industrial communication protocols on a single chip. The six pin-to-pin and software-compatible devices in this generation of processors, along with industrial hardware development tools, software and analog complements, provide a total industrial system solution.
- Sitara Processors Using the ARM Cortex-A series of cores, are optimized system solutions that go beyond the core, delivering products that support rich graphics capabilities, LCD displays and multiple industrial protocols.
- Industrial Communication with Sitara AM335x ARM Cortex-A8 Microprocessors Describes the key features and benefits of multiple, on-chip, production-ready industrial Ethernet and field bus communication protocols with master and slave functionality.



8.4 Related Links

表 8-1 lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

表 8-1. Related Links

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
AM3359	Click here	Click here	Click here	Click here	Click here
AM3358	Click here	Click here	Click here	Click here	Click here
AM3357	Click here	Click here	Click here	Click here	Click here
AM3356	Click here	Click here	Click here	Click here	Click here
AM3354	Click here	Click here	Click here	Click here	Click here
AM3352	Click here	Click here	Click here	Click here	Click here
AM3351	Click here	Click here	Click here	Click here	Click here

8.5 Community Resources

下列链接提供到 TI 社区资源的连接。链接的内容由各个分销商"按照原样"提供。这些内容并不构成 TI 技术规范,并且不一定反映 TI 的观点;请参阅 TI 的 《使用条款》。

TI E2E™ 在线社区为了促进工程师之间的合作,我们创建了 TI 工程师对工程师 (E2E) 社区。在 e2e.ti.com中,您可以提问、分享知识、拓展思路并与同行工程师一道帮助解决问题。

TI Embedded Processors Wiki Established to help developers get started with Embedded Processors from Texas Instruments and to foster innovation and growth of general knowledge about the hardware and software surrounding these devices.



8.6 商标

Sitara, SmartReflex, WiLink, E2E are trademarks of Texas Instruments.

NEON is a trademark of ARM Ltd or its subsidiaries.

ARM, Cortex are registered trademarks of ARM Ltd or its subsidiaries.

Bluetooth is a registered trademark of Bluetooth SIG.

EtherCAT is a registered trademark of EtherCAT Technology Group.

Android is a trademark of Google Inc.

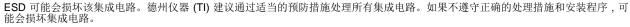
PowerVR SGX is a trademark of Imagination Technologies Limited.

Linux is a registered trademark of Linus Torvalds.

Wi-Fi is a registered trademark of Wi-Fi Alliance.

All other trademarks are the property of their respective owners.

8.7 静电放电警告





ESD 的损坏小至导致微小的性能降级,大至整个器件故障。 精密的集成电路可能更容易受到损坏,这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

8.8 术语表

TI 术语表 这份术语表列出并解释术语、缩写和定义。



www.ti.com.cn

9 Mechanical, Packaging, and Orderable Information

9.1 Via Channel

The ZCE package has been specially engineered with Via Channel technology. This allows larger than normal PCB via and trace sizes and reduced PCB signal layers to be used in a PCB design with the 0.65-mm pitch package, and substantially reduces PCB costs. It allows PCB routing in only two signal layers (four layers total) due to the increased layer efficiency of the Via Channel BGA technology.

Via Channel technology implemented on the ZCE package makes it possible to build an AM335x-based product with a 4-layer PCB, but a 4-layer PCB may not meet system performance goals. Therefore, system performance using a 4-layer PCB design must be evaluated during product design.

9.2 Packaging Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.





15-Dec-2017

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
AM3351BZCE30	ACTIVE	NFBGA	ZCE	298	160	(2) Green (RoHS & no Sb/Br)	(6) SNAGCU	Level-3-260C-168 HR	0 to 90	(4/5) AM3351BZCE30	Samples
AM3351BZCE30R	ACTIVE	NFBGA	ZCE	298	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3351BZCE30	Samples
AM3351BZCE60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3351BZCE60	Samples
AM3351BZCE60R	ACTIVE	NFBGA	ZCE	298	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3351BZCE60	Samples
AM3351BZCEA30	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3351BZCEA30	Samples
AM3351BZCEA60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3351BZCEA60	Samples
AM3352BZCE30	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3352BZCE30	Samples
AM3352BZCE30R	ACTIVE	NFBGA	ZCE	298	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR		AM3352BZCE30	Samples
AM3352BZCE60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3352BZCE60	Samples
AM3352BZCEA30	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCEA30	Samples
AM3352BZCEA30R	ACTIVE	NFBGA	ZCE	298	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCEA30	Samples
AM3352BZCEA60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCEA60	Samples
AM3352BZCEA60R	ACTIVE	NFBGA	ZCE	298	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCEA60	Samples
AM3352BZCED30	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3352BZCED30	Samples
AM3352BZCED60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3352BZCED60	Samples
AM3352BZCZ100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3352BZCZ100	Samples
AM3352BZCZ30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3352BZCZ30	Samples





www.ti.com

15-Dec-2017

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Sampl
AM3352BZCZ60	ACTIVE	NFBGA	ZCZ	324	126	(2) Green (RoHS	(6) SNAGCU	(3) Level-3-260C-168 HR	0 to 90	(4/5) AM3352BZCZ60	
	7.0				0	& no Sb/Br)		2010: 0 2000 100 1		,	Sampl
AM3352BZCZ80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3352BZCZ80	Samp
AM3352BZCZA100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCZA100	Samp
AM3352BZCZA30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCZA30	Samp
AM3352BZCZA60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCZA60	Samp
AM3352BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3352BZCZA80	Samp
AM3352BZCZD30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3352BZCZD30	Samp
AM3352BZCZD60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3352BZCZD60	Samp
AM3352BZCZD80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3352BZCZD80	Samp
AM3352BZCZT60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 125	AM3352BZCZT60	Samp
AM3352BZCZT60R	ACTIVE	NFBGA	ZCZ	324	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 125	AM3352BZCZT60	Samp
AM3354BZCE60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3354BZCE60	Samp
AM3354BZCEA60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3354BZCEA60	Samp
AM3354BZCED60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3354BZCED60	Samp
AM3354BZCZ100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3354BZCZ100	Samp
AM3354BZCZ30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3354BZCZ30	Samj
AM3354BZCZ60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3354BZCZ60	Samj
AM3354BZCZ80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3354BZCZ80	Samp



www.ti.com

15-Dec-2017

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
AM3354BZCZA100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3354BZCZA100	Samples
AM3354BZCZA60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3354BZCZA60	Samples
AM3354BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3354BZCZA80	Samples
AM3354BZCZA80R	ACTIVE	NFBGA	ZCZ	324	1000	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR		AM3354BZCZA80	Samples
AM3354BZCZD60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3354BZCZD60	Samples
AM3354BZCZD80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3354BZCZD80	Samples
AM3356BZCEA60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR		AM3356BZCEA60	Samples
AM3356BZCZ30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3356BZCZ30	Samples
AM3356BZCZ60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3356BZCZ60	Samples
AM3356BZCZ80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3356BZCZ80	Samples
AM3356BZCZA30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3356BZCZA30	Samples
AM3356BZCZA60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3356BZCZA60	Samples
AM3356BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3356BZCZA80	Samples
AM3356BZCZD30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3356BZCZD30	Samples
AM3356BZCZD60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3356BZCZD60	Samples
AM3357BZCZA30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3357BZCZA30	Samples
AM3357BZCZA60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3357BZCZA60	Samples
AM3357BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3357BZCZA80	Samples



PACKAGE OPTION ADDENDUM

15-Dec-2017

Orderable Device	Status	Package Type	_	Pins	_	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
AM3357BZCZD30	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3357BZCZD30	Samples
AM3357BZCZD60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 90	AM3357BZCZD60	Samples
AM3358BZCE60	ACTIVE	NFBGA	ZCE	298	160	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3358BZCE60	Samples
AM3358BZCZ100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3358BZCZ100	Samples
AM3358BZCZ60	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3358BZCZ60	Samples
AM3358BZCZ80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	0 to 90	AM3358BZCZ80	Samples
AM3358BZCZA100	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3358BZCZA100	Samples
AM3358BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3358BZCZA80	Samples
AM3359BZCZA80	ACTIVE	NFBGA	ZCZ	324	126	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 105	AM3359BZCZA80	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.



PACKAGE OPTION ADDENDUM

15-Dec-2017

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF AM3358:

Enhanced Product: AM3358-EP

NOTE: Qualified Version Definitions:

• Enhanced Product - Supports Defense, Aerospace and Medical Applications

ZCZ (S-PBGA-N324)

PLASTIC BALL GRID ARRAY



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. This is a Pb-free solder ball design.



ZCE (S-PBGA-N298)

PLASTIC BALL GRID ARRAY



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. This is a Pb-free solder ball design.



重要声明

德州仪器 (TI) 公司有权按照最新发布的 JESD46 对其半导体产品和服务进行纠正、增强、改进和其他修改,并不再按最新发布的 JESD48 提供任何产品和服务。买方在下订单前应获取最新的相关信息,并验证这些信息是否完整且是最新的。

TI 公布的半导体产品销售条款 (http://www.ti.com/sc/docs/stdterms.htm) 适用于 TI 已认证和批准上市的已封装集成电路产品的销售。另有其他条款可能适用于其他类型 TI 产品及服务的使用或销售。

复制 TI 数据表上 TI 信息的重要部分时,不得变更该等信息,且必须随附所有相关保证、条件、限制和通知,否则不得复制。TI 对该等复制文件不承担任何责任。第三方信息可能受到其它限制条件的制约。在转售 TI 产品或服务时,如果存在对产品或服务参数的虚假陈述,则会失去相关 TI 产品或服务的明示或暗示保证,且构成不公平的、欺诈性商业行为。TI 对此类虚假陈述不承担任何责任。

买方和在系统中整合 TI 产品的其他开发人员(总称"设计人员")理解并同意,设计人员在设计应用时应自行实施独立的分析、评价和判断,且应全权负责并确保应用的安全性,及设计人员的应用(包括应用中使用的所有 TI 产品)应符合所有适用的法律法规及其他相关要求。设计人员就自己设计的应用声明,其具备制订和实施下列保障措施所需的一切必要专业知识,能够 (1) 预见故障的危险后果,(2) 监视故障及其后果,以及 (3) 降低可能导致危险的故障几率并采取适当措施。设计人员同意,在使用或分发包含 TI 产品的任何应用前,将彻底测试该等应用和和该等应用所用 TI 产品的功能而设计。

TI 提供技术、应用或其他设计建议、质量特点、可靠性数据或其他服务或信息,包括但不限于与评估模块有关的参考设计和材料(总称"TI 资源"),旨在帮助设计人员开发整合了 TI 产品的 应用, 如果设计人员(个人,或如果是代表公司,则为设计人员的公司)以任何方式下载、访问或使用任何特定的 TI 资源,即表示其同意仅为该等目标,按照本通知的条款使用任何特定 TI 资源。

TI 所提供的 TI 资源,并未扩大或以其他方式修改 TI 对 TI 产品的公开适用的质保及质保免责声明;也未导致 TI 承担任何额外的义务或责任。TI 有权对其 TI 资源进行纠正、增强、改进和其他修改。除特定 TI 资源的公开文档中明确列出的测试外,TI 未进行任何其他测试。

设计人员只有在开发包含该等 TI 资源所列 TI 产品的 应用时,才被授权使用、复制和修改任何相关单项 TI 资源。但并未依据禁止反言原则或其他法理授予您任何TI知识产权的任何其他明示或默示的许可,也未授予您 TI 或第三方的任何技术或知识产权的许可,该等产权包括但不限于任何专利权、版权、屏蔽作品权或与使用TI产品或服务的任何整合、机器制作、流程相关的其他知识产权。涉及或参考了第三方产品或服务的信息不构成使用此类产品或服务的许可或与其相关的保证或认可。使用 TI 资源可能需要您向第三方获得对该等第三方专利或其他知识产权的许可。

TI 资源系"按原样"提供。TI 兹免除对资源及其使用作出所有其他明确或默认的保证或陈述,包括但不限于对准确性或完整性、产权保证、无屡发故障保证,以及适销性、适合特定用途和不侵犯任何第三方知识产权的任何默认保证。TI 不负责任何申索,包括但不限于因组合产品所致或与之有关的申索,也不为或对设计人员进行辩护或赔偿,即使该等产品组合已列于 TI 资源或其他地方。对因 TI 资源或其使用引起或与之有关的任何实际的、直接的、特殊的、附带的、间接的、惩罚性的、偶发的、从属或惩戒性损害赔偿,不管 TI 是否获悉可能会产生上述损害赔偿,TI 概不负责。

除 TI 己明确指出特定产品已达到特定行业标准(例如 ISO/TS 16949 和 ISO 26262)的要求外,TI 不对未达到任何该等行业标准要求而承担任何责任。

如果 TI 明确宣称产品有助于功能安全或符合行业功能安全标准,则该等产品旨在帮助客户设计和创作自己的 符合 相关功能安全标准和要求的应用。在应用内使用产品的行为本身不会 配有 任何安全特性。设计人员必须确保遵守适用于其应用的相关安全要求和 标准而设计。设计人员不可将任何 TI 产品用于关乎性命的医疗设备,除非已由各方获得授权的管理人员签署专门的合同对此类应用专门作出规定。关乎性命的医疗设备是指出现故障会导致严重身体伤害或死亡的医疗设备(例如生命保障设备、心脏起搏器、心脏除颤器、人工心脏泵、神经刺激器以及植入设备)。此类设备包括但不限于,美国食品药品监督管理局认定为 III 类设备的设备,以及在美国以外的其他国家或地区认定为同等类别设备的所有医疗设备。

TI 可能明确指定某些产品具备某些特定资格(例如 Q100、军用级或增强型产品)。设计人员同意,其具备一切必要专业知识,可以为自己的应用选择适合的 产品, 并且正确选择产品的风险由设计人员承担。设计人员单方面负责遵守与该等选择有关的所有法律或监管要求。

设计人员同意向 TI 及其代表全额赔偿因其不遵守本通知条款和条件而引起的任何损害、费用、损失和/或责任。

邮寄地址: 上海市浦东新区世纪大道 1568 号中建大厦 32 楼,邮政编码: 200122 Copyright © 2018 德州仪器半导体技术(上海)有限公司