

# **Filtering**

**Image and Signal Processing** 

Norman Juchler



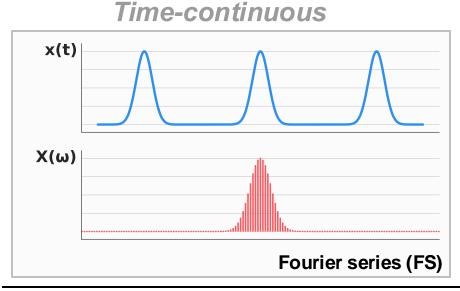
# Fourier's landscape

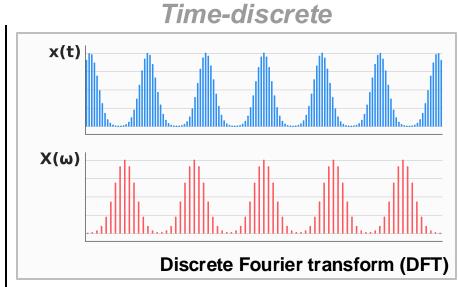
We are really wrapping this up now!



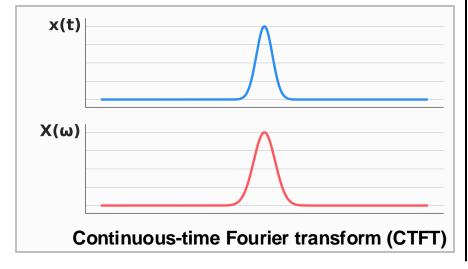
# Fourier's landscape

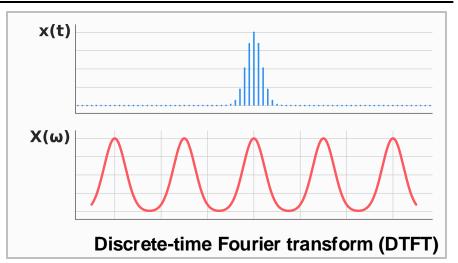
Periodic





**Aperiodic** 

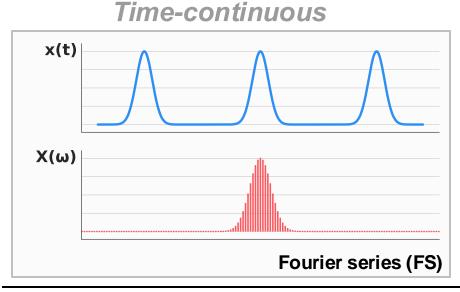




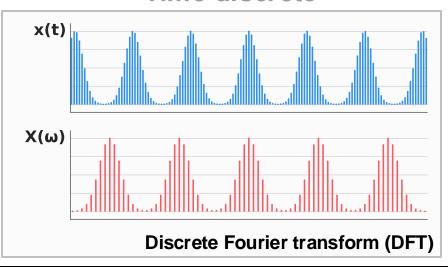


# Fourier's landscape

Periodic





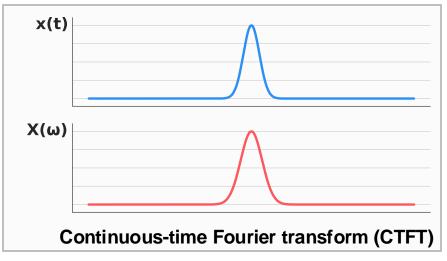


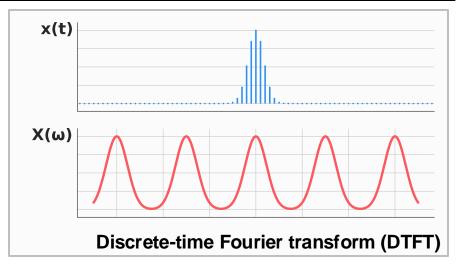
#### Related:

- Fast **Fourier** transform
- Discrete cosine transform
- Discrete sine transform

**Aperiodic** 







Generalization:

z - Transform



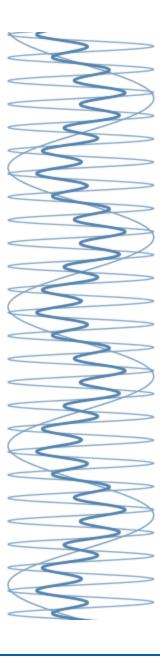
### A zoo of different transformations...

### ...but they share many common characteristics:

- Each has a forward and inverse transformation.
- They all exhibit duality between time-domain and frequencydomain representations.
- They share similar mathematical properties, including:
  - Linearity
  - Time/ frequency shifting
  - Convolution theorem
  - •

### Take-home messages:

- Fourier theory allows us to decompose signals into their fundamental frequency components.
- We can seamlessly switch between the time domain and the frequency domain, and vice versa.
- Different types of Fourier transforms are suited for different types of signals, ensuring the most effective representation.





### The discrete Fourier transform

- Input: Finite sequence of N equally-spaced samples of a signal
- Output: Finite sequence of N equally-spaced samples of the DTFT

$$X[m] = \sum_{n=0}^{N-1} x[n] \cdot e^{-i2\pi \frac{m}{N}n}$$

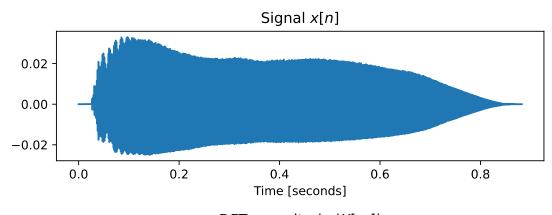
$$X[m] = \sum_{n=0}^{N-1} x[n] \cdot e^{-i2\pi \frac{m}{N}n} \qquad f_m = \begin{cases} \frac{m}{N} \cdot f_s & \text{if } 0 \le m < N/2\\ \frac{m-N}{N} \cdot f_s & \text{if } N/2 \le m < N \end{cases}$$

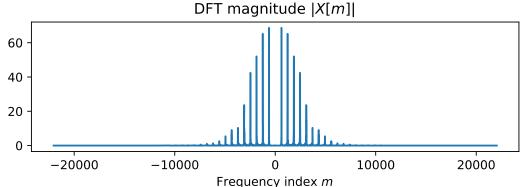
#### Observations:

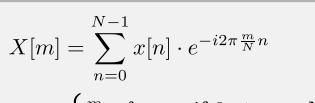
- The X[m] represents discrete samples of the (continuous) discrete-time Fourier transform
- The X[m] in general are complex numbers (with magnitude and phase)
- The DFT X[m] is periodic with period N (a fact leveraged by fftshift()...)
- Every X[m] is associated with the frequency  $f_m$ 
  - Note that we usually consider frequencies  $f_m$  that are negative for  $m \ge N/2$  (periodic extension)
  - The first value X[0] is known as the **direct current** (DC) or static component, with  $f_0 = 0$
  - X[0] always equals the sum of the sample values

# **Example: A real sound**











$$f_m = \begin{cases} \frac{m}{N} \cdot f_s & \text{if } 0 \le m < N/2\\ \frac{m-N}{N} \cdot f_s & \text{if } N/2 \le m < N \end{cases}$$

# Look at the illustrations and try to answer the following **questions**:

- What is the sampling frequency?
- How many samples?
- What is the static component about? How is it related to the mean value?
- At which frequency is the peak?

#### Answers:

- Sampling at 44'100Hz
- Samples 0.88·44'100 ≈ 38'000
- Static component: first value X[0] = sum of x[n] ≈ 0
- Corresponds to the pitch of the note played (here:  $D_5$ #=618.8Hz)



# The discrete Fourier transform – a naive implementation:

**Question**: How long does it take to compute the DFT for a signal x?

**Answer**: Line 12 needs to be calculated  $N^2$  times, or using the big-O notation: The complexity of this DFT is  $O(N^2)$ 



# The Fast Fourier Transform (FFT)

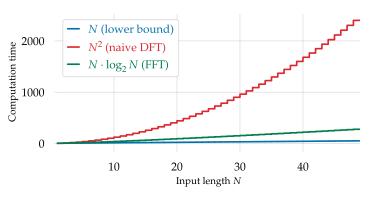
 The FFT is an efficient algorithm to compute the DFT of a discrete signal

### History:

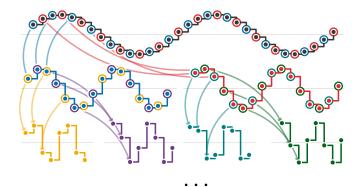
- Some ideas of FFT can be traced back to C.F. Gauss (1805).
- The algorithm was first published in 1965 by J. Cooley and J. Tukey.
- The most popular version is called radix-2 Cooley-Tukey algorithm.

### Key idea: Divide and conquer

- Exploit DFT symmetry to recursively break the computation into smaller sub-problems, eliminating redundant calculations.
- The roots of unity play a central role here...
- Instead of directly computing the DFT for  $N = N_1 N_2$ , it reformulates the problem in terms of for  $N_1$  smaller problems of size  $N_2$ .
- This approach reduces computational complexity from O(N²) for a naive implementation to O(N log N)!



The increase of the amount of work (that is, the number of computations) increases with the problem size



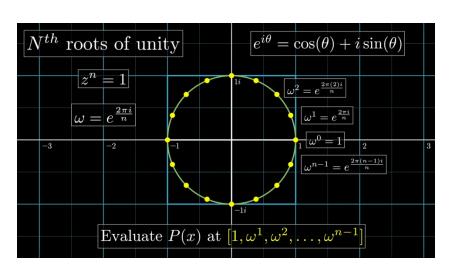
A discrete signal of N=32 samples (top row) is divided into its even and odd samples (middle-left and middle-right).

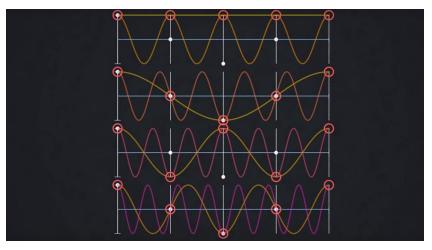


#### Recommended videos

 Reducible / 2023: The Fast Fourier Transform (FFT): Most Ingenious Algorithm Ever? <u>Link</u>

 Veritasium / 2023: The Most Important Algorithm Of All Time. <u>Link</u>







# The Fast Fourier Transform (FFT)

- The package <u>scipy.fft</u> offers fast implementations of various FFT algorithms
- See this tutorial for an overview how to use FFT.
- The main functions:

• fft(), fft2(), fftn():

ifft(), ifft2(), ifftn():

rfft(), rfft2(), rfftn():

irfft(), irfft2(), irfftn():

fftshift():

Compute the FFT on 1D, 2D, and nD input.

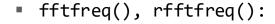
Compute the inverse DFT in 1D, 2D or nD

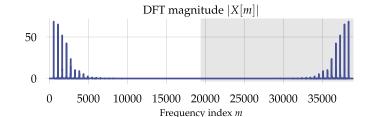
Compute the FFT on <u>real</u> input, faster than the fft\*() functions.

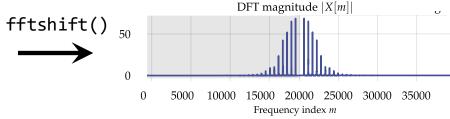
Compute the inverse real DFT in 1D, 2D or nD

fft\*() and fftfreq() return all components at positive frequencies, followed by components at negative frequencies. This function swaps the two halves so that the zero-frequency component is in the center:  $[0, 1, 2, -3, -2, -1] \rightarrow [-3, -2, -1, 0, 1, 2, 3]$ .

Get the frequencies of each bin returned by the FFT functions, such as the X-axis.

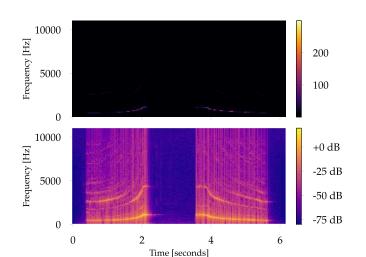




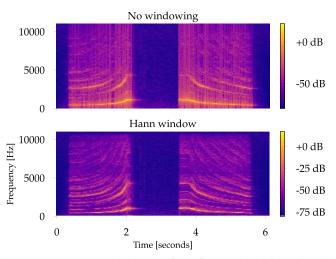


# The short-time Fourier Transform (STFT)

- The DFT makes use of all samples in the signal
- Implicitly, this is assuming that frequency content is "stationary" over the duration of the signal.
- For time-varying signals, this is often not the case
- Idea: Divide a long signal up into short pieces (socalled frames), analyze each piece separately.
- Parameters:
  - Frame length (number of samples per frame)
  - Hop length (number of samples between two frames)
- STFTs can be visualized with spectrograms
  - Constructed by stacking the frames horizontally
  - Hint 1: Use the decibel scale for better visualization
  - Hint 2: Use windowing / apodization to avoid banding artifacts



Spectrogram with linear magnitude scaling (top) and with decibel scaling:  $A_{\rm dB} = 20 \cdot \log_{10} A$ 



Spectrogram without (top) and with windowing (bottom). Use a window (here: Hann) to attenuate discontinuities at the ends of the frames.



# Convolution

Revisited

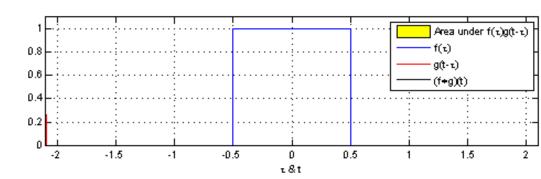


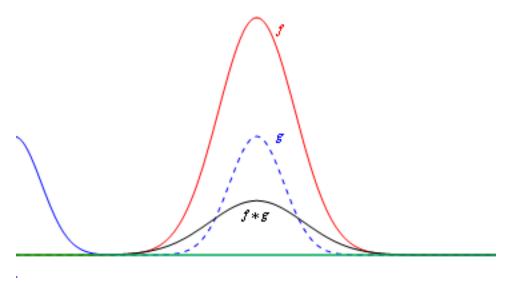
# Convolution of continuous-time signals

 The convolution is an operation that combines two functions to produce a third function.

$$y(t) = (h * x)(t) := \int_{-\infty}^{\infty} h(\tau)x(t - \tau) d\tau$$

- In words, integral computes the area under the product of  $h(\tau)$  and  $x(t-\tau)$  as t varies.
- Demo applet
- Video: 3blue1brown / 2023. But what is a convolution? <u>Link</u>





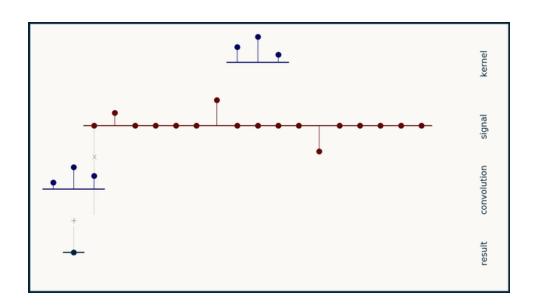


# **Convolution of discrete-time signals**

There exists also a discrete-time definition for convolution:

$$y[n] = (h * g)[n] = \sum_{m=-\infty}^{\infty} h[m]x[n-m]$$

• In the animation right is x[n] the input signal, h[n] the kernel, and y[n] the output signal.





# Naive implementation of a discrete-time convolution

```
# Compute the lengths of our input and filter
N = len(x)
K = len(h)

# Allocate an output buffer
y = np.zeros(N)

# Iterate over delay values
for k in range(K):

    # For each delay value, iterate over all sample indices
    for n in range(N):

    if n >= k:
        # No contribution from samples n < k
        y[n] += h[k] * x[n-k]</pre>
```

- scipy.signal.convolve() is a fast alternative
  - Makes use of FFT!
  - Question: Why again?

Convolution 18



### **Convolution and Fourier transform**

• Result: The Fourier transform simplifies the convolution operation to a simple multiplication in the frequency domain!

$$y(t) = h(t) * x(t) \Leftrightarrow Y(\omega) = H(\omega) \cdot X(\omega)$$
  
 $y[n] = h[h] * x[n] \Leftrightarrow Y[m] = H[m] \cdot X[m]$ 

- Computing the convolution in the frequency domain is very easy!
- We can efficiently recover the time-domain signal via inverse FFT
- Recommended procedure (for discrete signals):
  - Pad x[n] and h[n] to the same length (if necessary)
  - Use the FFT to compute DTFs X[m] and H[m]
  - Multiply X[m] and H[m] elementwise
  - Compute the inverse DFT to recover y[n]

Convolution

# Signal filtering

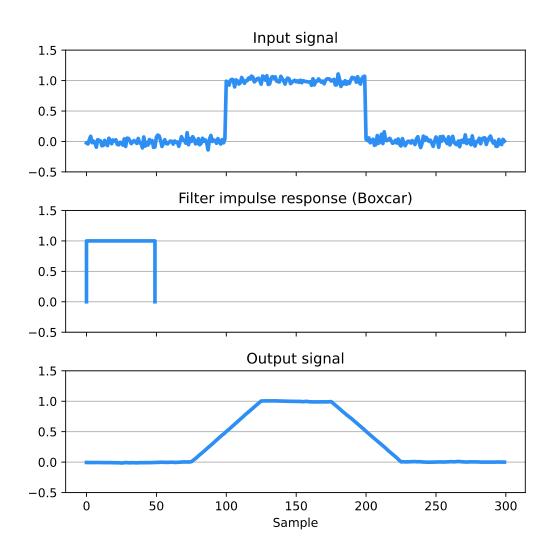
First concepts



- Input:
  - Noisy signal x(t)
  - A filter window h(t)
- Output
  - Filtered signal y(t)
- Solution:

```
import numpy as np
from scipy import signal

noise = 0.05
width = 50
win = signal.windows.boxcar(width)
x = np.repeat([0., 1., 0.], 100)
x += np.random.normal(0, noise, x.shape)
y = signal.convolve(x, win, mode='same') / sum(win)
```



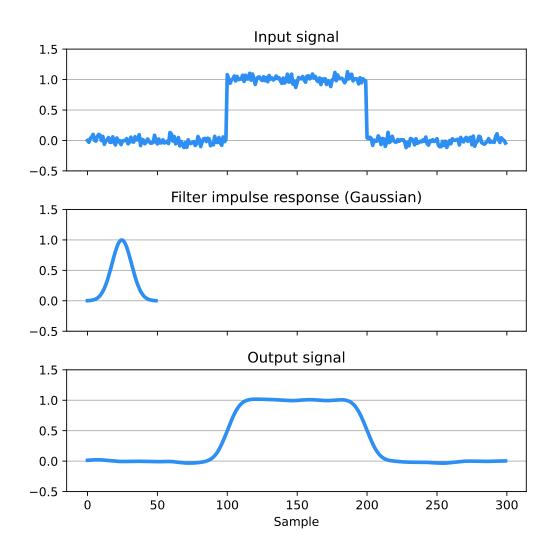


#### Observations:

- We sacrifice some of the signal, but
- We can remove the noise.

```
import numpy as np
from scipy import signal

noise = 0.05
width = 50
win = signal.windows.gaussian(width, std=7)
x = np.repeat([0., 1., 0.], 100)
x += np.random.normal(0, noise, x.shape)
y = signal.convolve(x, win, mode='same') / sum(win)
```

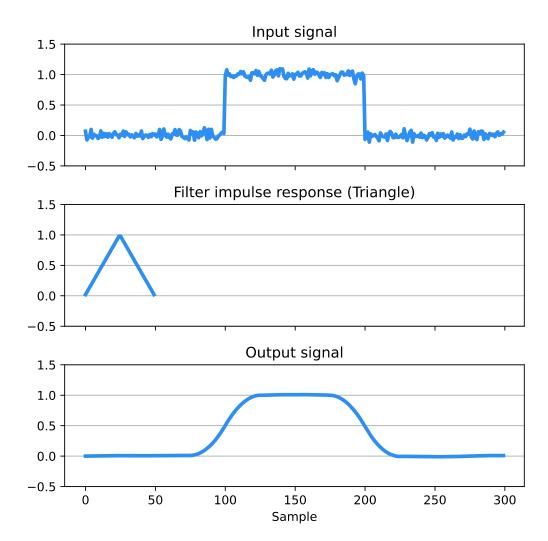




- Many different window types are possible:
  - BoxParzen
  - TriangleExponential
  - BlackmanTukey
  - HammingLanczos
  - Hann■ ...

```
import numpy as np
from scipy import signal

noise = 0.05
width = 50
win = signal.windows.triang(width)
x = np.repeat([0., 1., 0.], 100)
x += np.random.normal(0, noise, x.shape)
y = signal.convolve(x, win, mode='same') / sum(win)
```

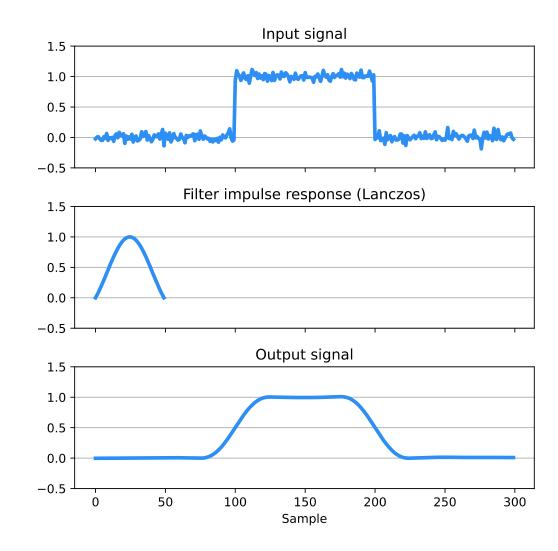




- What type of smoothing window should be used? Which parameters to select?
  - A typical filter design question!
  - Requires an understanding of signal characteristics
  - Spectral analysis of signal and filter!

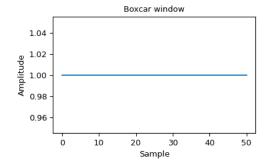
```
import numpy as np
from scipy import signal

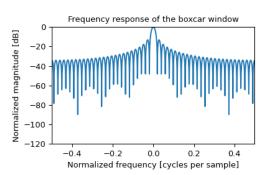
noise = 0.05
width = 50
win = signal.windows.lanczos(width)
x = np.repeat([0., 1., 0.], 100)
x += np.random.normal(0, noise, x.shape)
y = signal.convolve(x, win, mode='same') / sum(win)
```



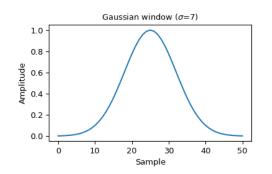


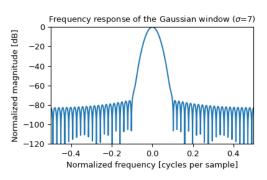
### Comparison of different windows



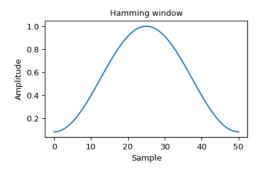


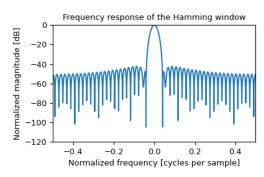
$$h_{box}[n] = 1$$



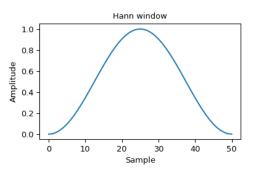


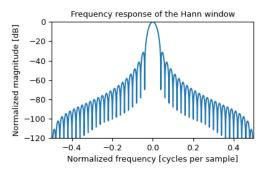
$$h_{gauss}[n] = e^{\frac{1}{2}\left(\frac{n}{\sigma}\right)^2}$$





$$h_{lanc}[n] = sinc_{\pi} \left( \frac{2n}{M-1} - 1 \right)$$





$$h_{lanc}[n] = sinc_{\pi} \left( \frac{2n}{M-1} - 1 \right)$$
  $h_{lann}[n] = 0.5 - 0.5 \cos \left( \frac{2\pi n}{M-1} \right)$ 

# What is the theoretically best low-pass filter?

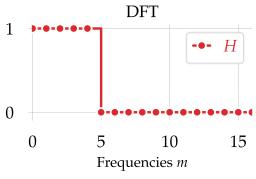
- Alternative approach: start in the frequency domain
- Goal: Get rid of all frequencies above some cut-off  $f_c$
- Idea: Let's define the ideal low-pass filter:

$$H[m] = \begin{cases} 0, & \text{if } f_m > f_c \\ 1, & \text{if } f_m \le f_c \end{cases}$$

We then can apply the filter in the frequency domain by multiplying it with the signal X[m] and recover the resulting signal y[n] in the time-domain.

 Such a filter is called ideal low-pass filter, or brick-wall filter.

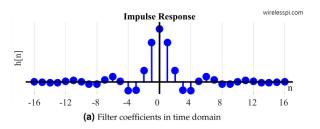


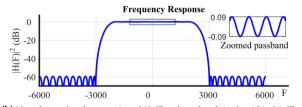


H[m]: The ideal low-pass filter in the frequency-domain



h[n]: Filter representation in the timedomain (so-called impulse response)



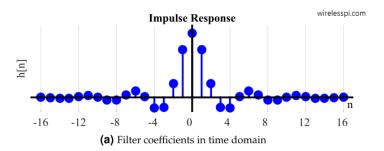


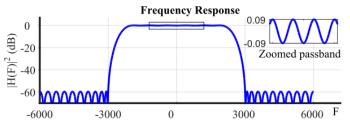
(b) Note the stopband attenuation of 60 dB and passband ripple within 0.1 dB



### Problems with the ideal low-pass filter

- Has an infinite representation in the time domain
- Cannot be handled properly with DTF, as the filter lengths are limited
- Observations:
  - We already saw that it is difficult to represent sharp edges in the time-domain using frequency components (we require infinitely high frequencies)
  - The same holds true in the frequency domain: The sharper the filter is in the frequency domain, the more time-domain samples are necessary.
  - Note that the filter requires infinitely many samples at t<0! The ideal filter therefore is a non-causal filter.</p>
- See here for an illustrative demonstration.

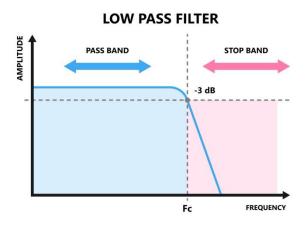


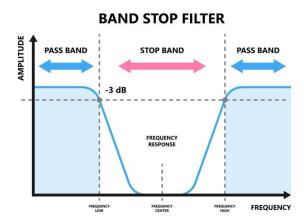


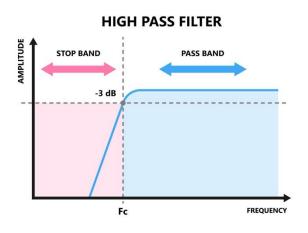
**(b)** Note the stopband attenuation of 60 dB and passband ripple within 0.1 dB

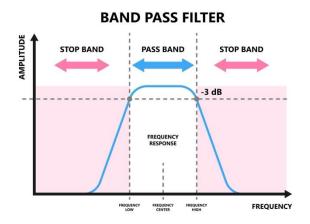


# **Overview: Types of filters**











# Frequency bands

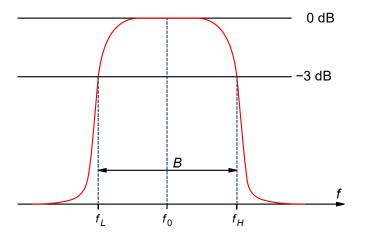
A signal is called **band-limited** if its frequency components are limited within frequencies:

$$f_L \le f \le f_H$$

The bandwidth of a signal is defined as the difference between the extreme frequencies:

$$B = \Delta f = f_H - f_L$$

- A strictly band-limited signal does not carry energy at frequencies outside the band limits
- In practice, a signal is considered band-limited if its energy outside of a frequency range is low enough to be considered negligible



Amplitude spectrum of a bandlimited signal. Source of illustration: Wikimedia