

模拟与数字电路

Analog and Digital Circuits



课程主页 扫一扫

第二十讲：运算放大器与负反馈原理

Lecture 20: **Operational Amplifiers**

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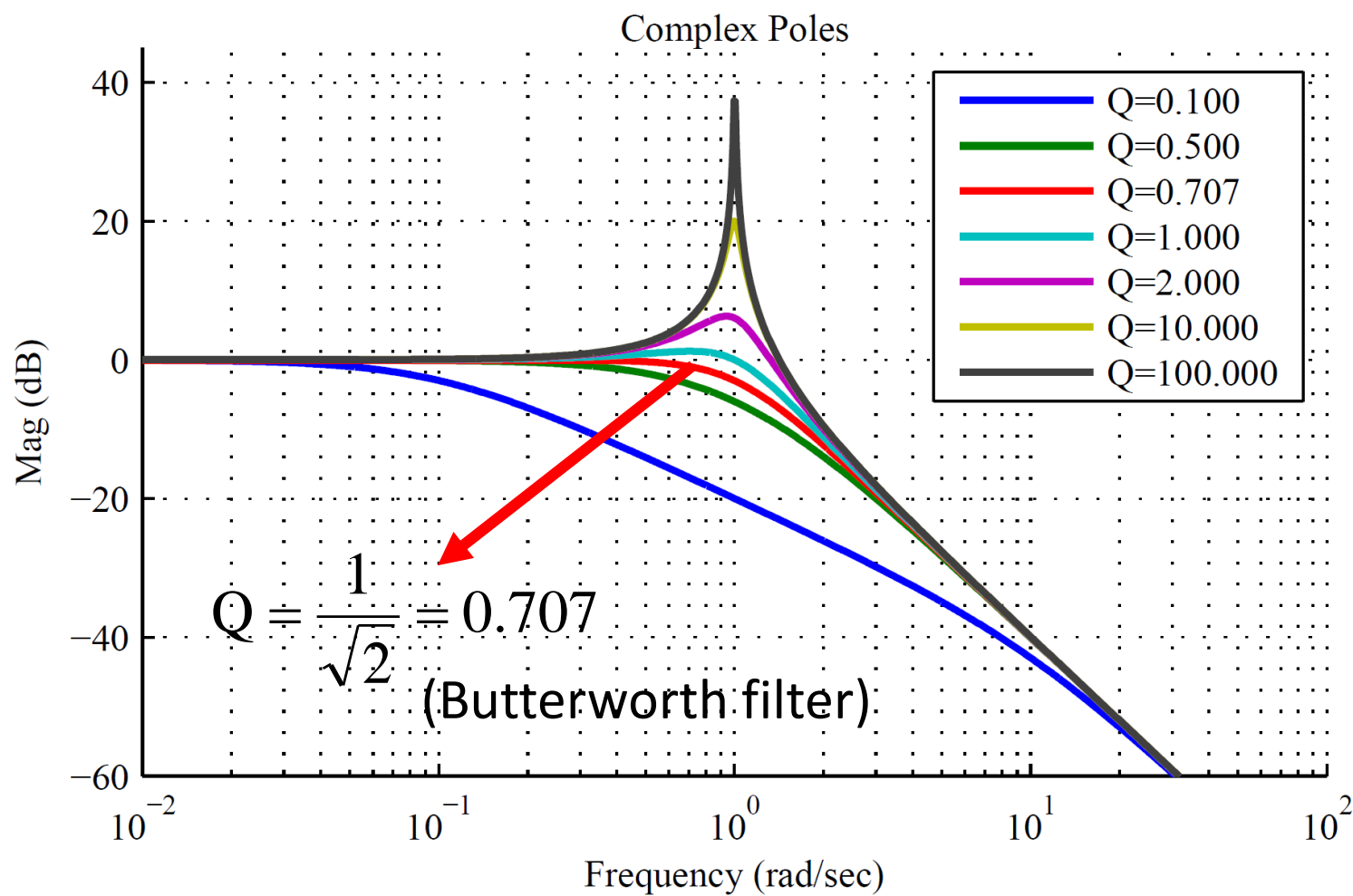
Instructor : Chixiao Chen

提纲

- 复习
 - 从低通、高通、带通滤波器的零极点变化?
 - (续) 滤波器
 - 负反馈
 - 运算放大器
 - 运算放大器的应用



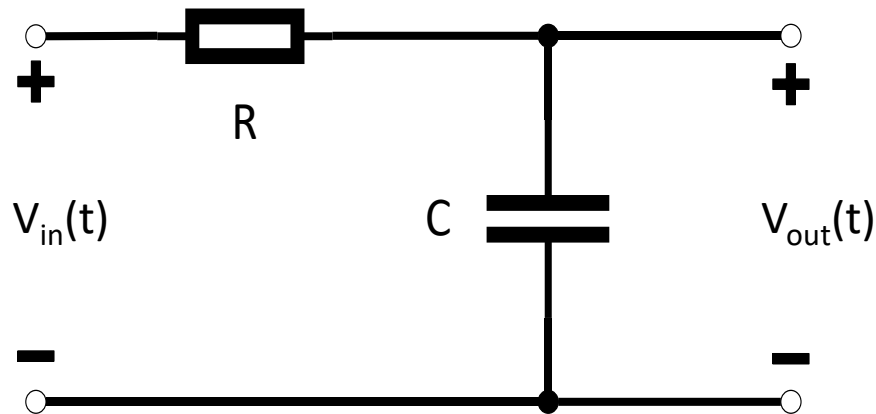
复习：二阶滤波器的品质因数



$$H(s) = \frac{K}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2}$$

$Q > 0.707$ Peaking

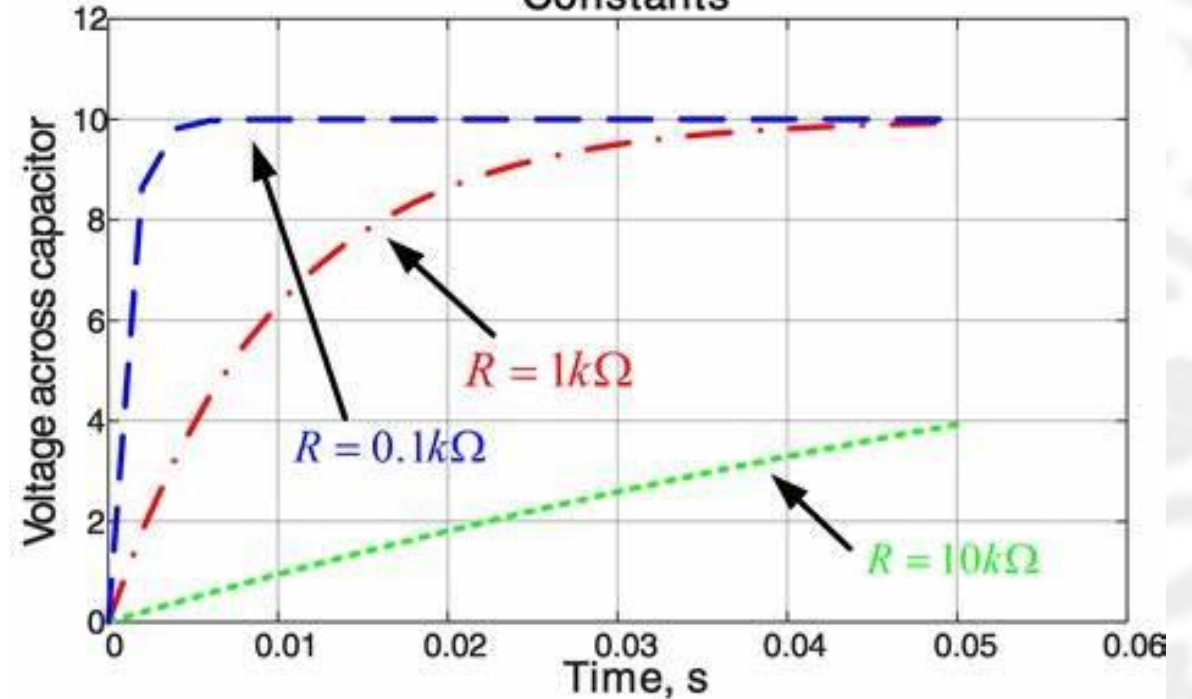
时间常数 time constant



$$C \cdot \frac{dV_{out}(t)}{dt} + \frac{1}{R} \cdot V_{out}(t) = \frac{1}{R} \cdot V_{in}(t)$$

$$V_{out}(t) = \left(1 - e^{-\frac{t}{RC}}\right) V_{in}(t)$$

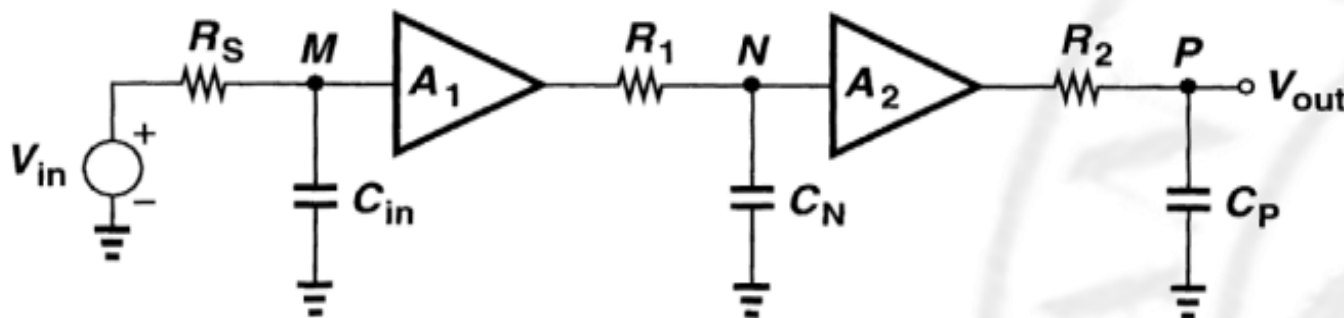
Capacitor Charging Analysis with three Time Constants



$$\frac{V_o(j\omega)}{V_i(j\omega)} = \frac{\frac{1}{j\omega C}}{R + \frac{1}{j\omega C}} = \frac{1}{1 + j\omega RC}$$

高阶有源滤波器

- 对于一个电路的结点，如果有电容和电阻连接在这个结点上，那么通常这个结点会给最终的表达式贡献一个极点：



$$\begin{aligned}\frac{V_{out}}{V_{in}}(s) &= \frac{1}{1 + R_S C_{in} s} \cdot \frac{A_1}{1 + R_1 C_N s} \cdot \frac{A_2}{1 + R_2 C_P s} \\ &= 1 \cdot A_1 A_2 \frac{1}{1 + s / \omega_{pM}} \cdot \frac{1}{1 + s / \omega_{pN}} \cdot \frac{1}{1 + s / \omega_{pP}}\end{aligned}$$

放大器的非线性

- 典型建模方案：泰勒展开

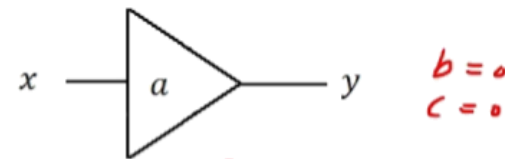
Non-Linear Amplifier

- x = input
- y = output

$$y = a.x + b.x^2 + \textcircled{c}.x^3 + \dots$$

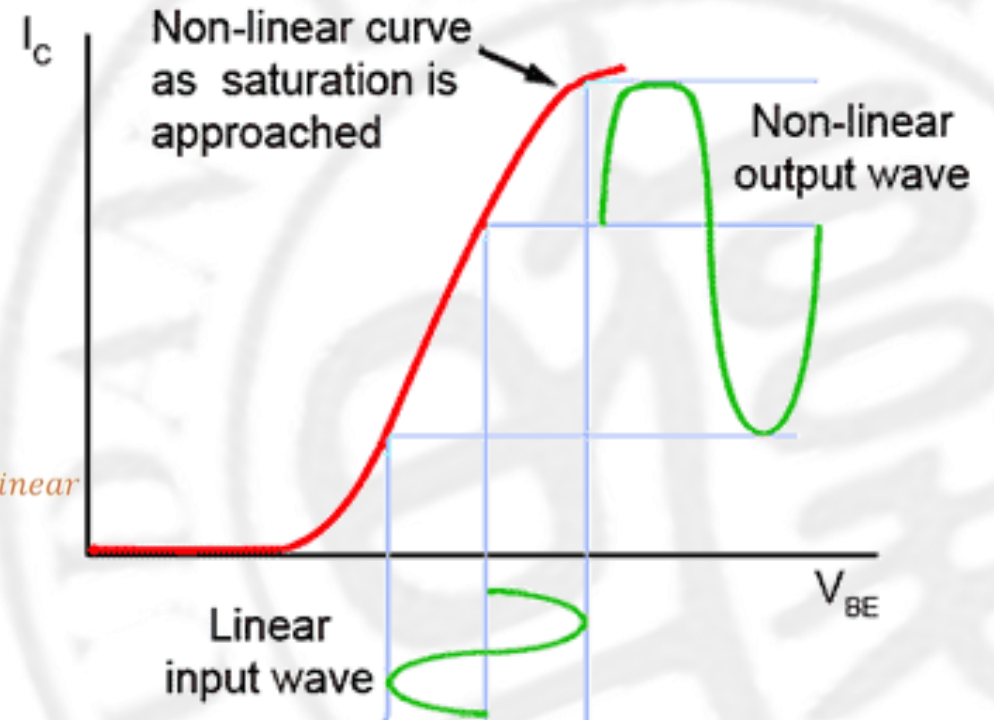
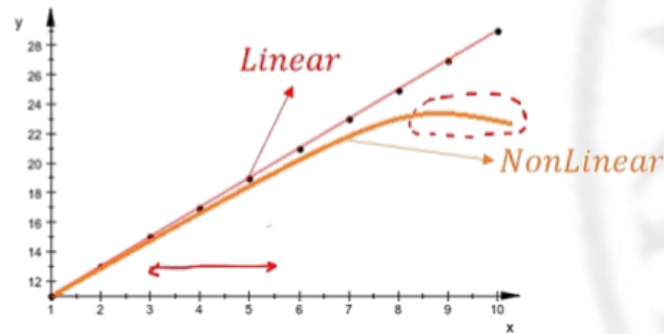
- Generally $c < 0$

$x \rightarrow 10\mu$



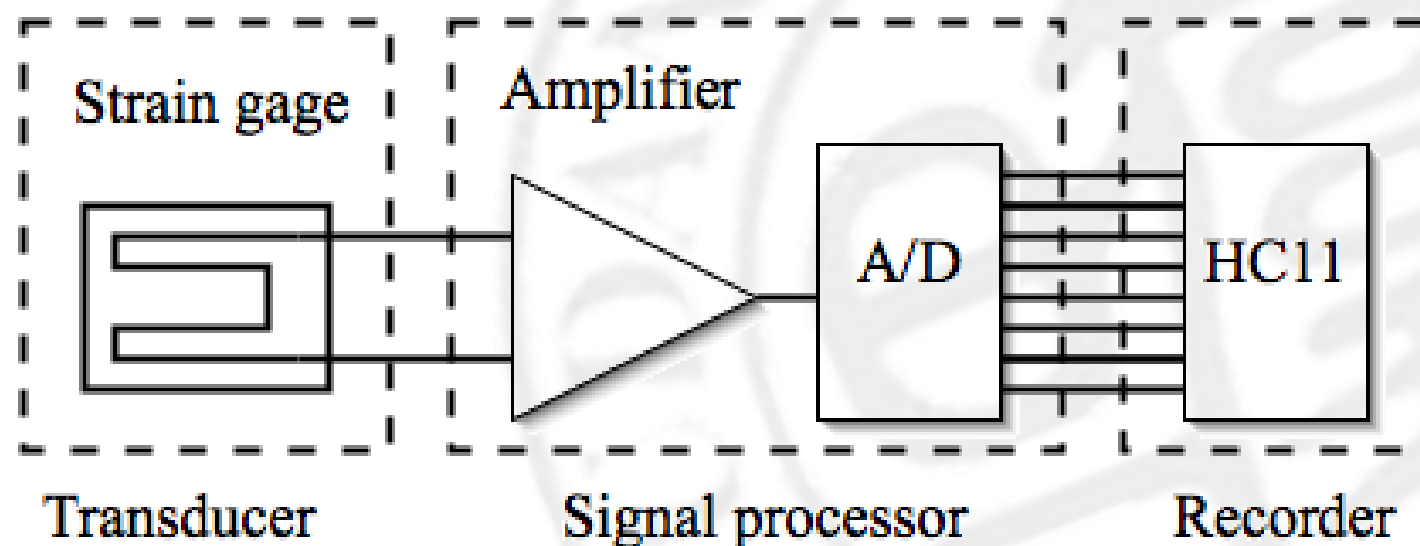
$$b = 0$$
$$c = 0$$

$$y = \underbrace{a.x}_{\text{Linear}} + \underbrace{b.x^2 + c.x^3}_{\text{NonLinear}} + \dots$$



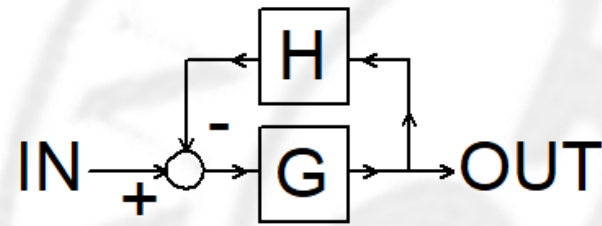
运算放大器

- Sensor signals are often too weak or too noisy
- 目标1：以固定的比例线性放大信号
- 目标2：将某些频率滤去



负反馈原理

- 负反馈：将输出按一定比例反馈到输入
- 通过信号流图得到负反馈的表达式
- 当增益G不恒定，但保持远远大于1时
- 增益由H的大小决定
- 实际上当， $GH \gg 1$ 时
G的输入摆幅非常小



$$\text{Gain} = \frac{G}{1 + GH}$$

$$\text{Gain} = \frac{1}{H} \times \frac{GH}{1 + GH}$$

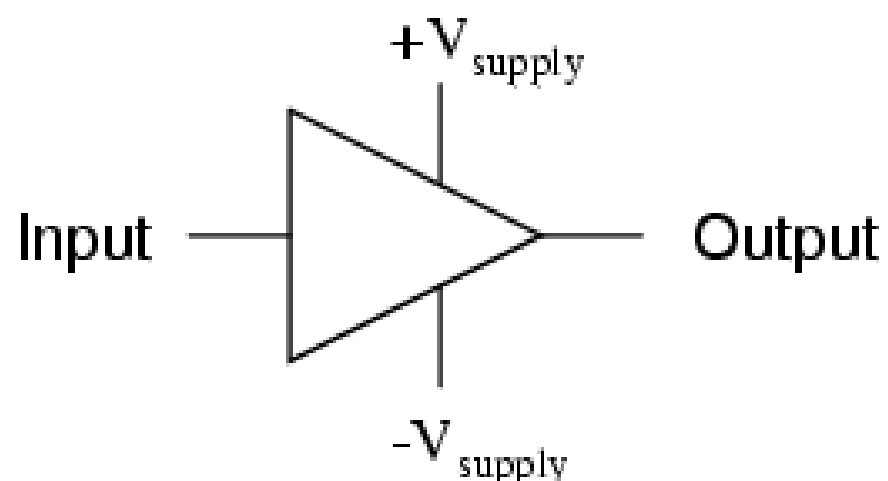
Loop gain is not GH.

Loop gain is -GH.

如何实现减法？

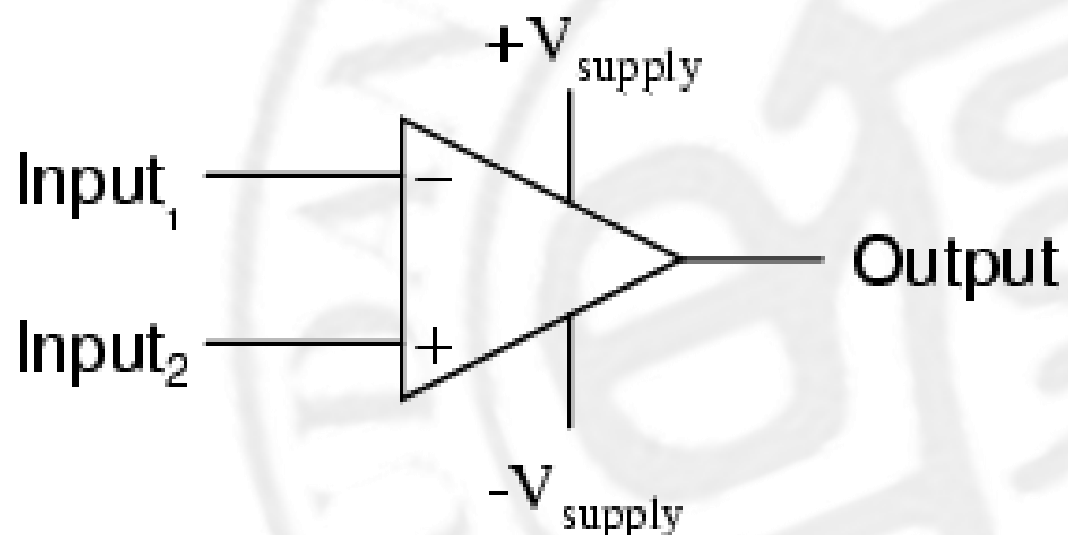
- 单端放大器

General amplifier circuit symbol



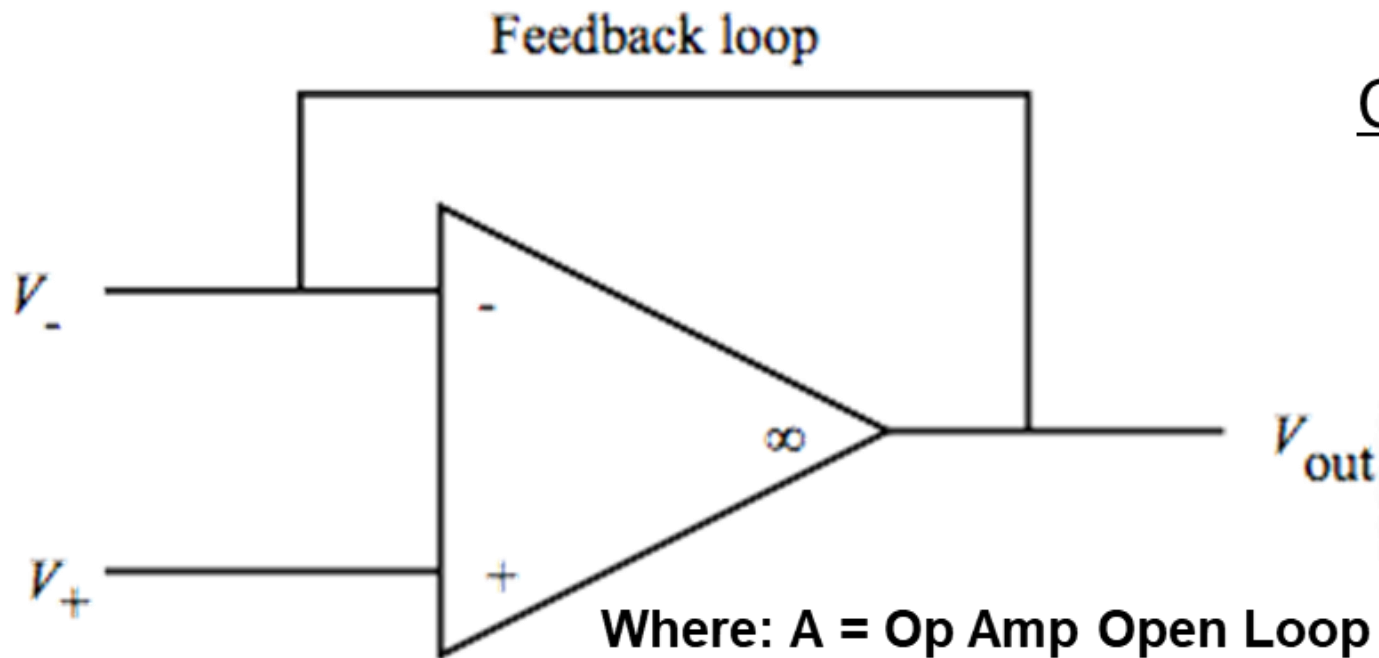
- 差分放大器

Differential amplifier



放大电路 应用实例

- 将输出直接连接到负端输入



Where: A = Op Amp Open Loop Gain
 F = Feedback Loop Gain

Closed Loop Transfer Function

$$H(s) = A / (1 + AF)$$

When $AF \gg 1$...

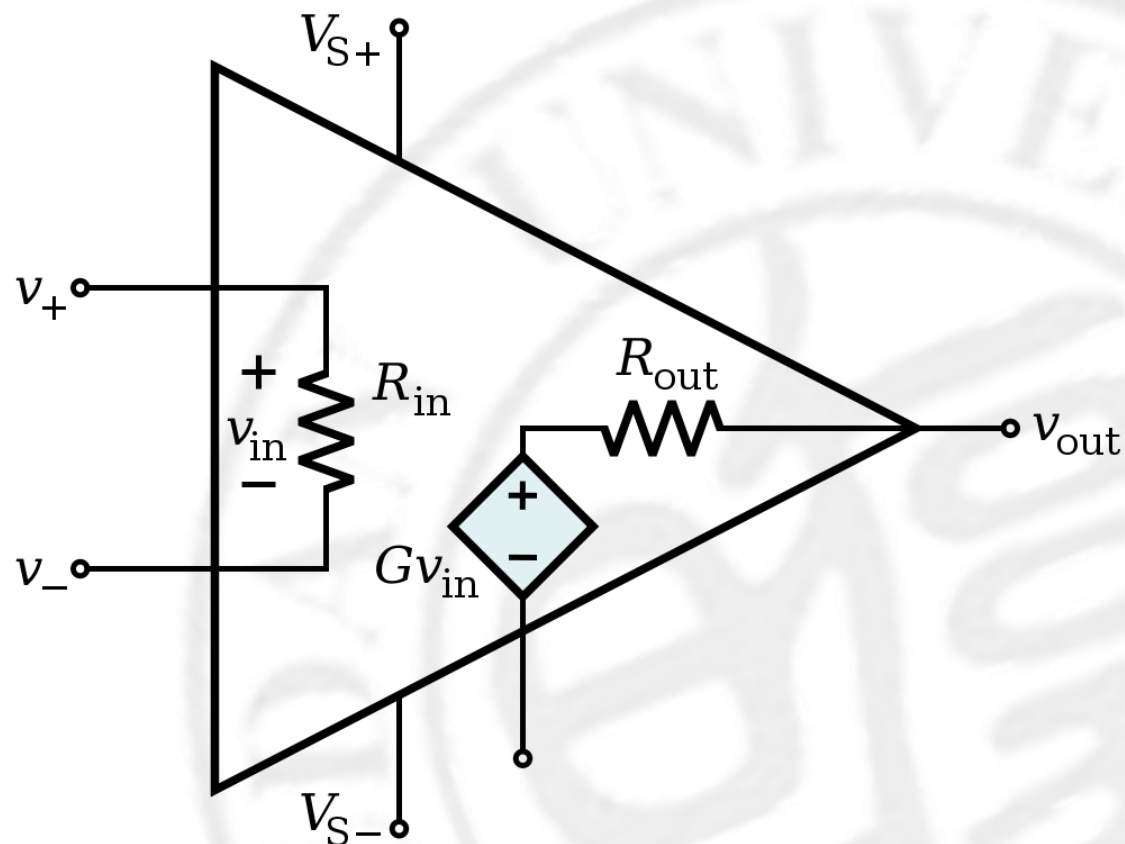
$$H(s) = 1 / F$$

理想运算放大器 参数

- 电压增益：无限大（大于 10000）
- 输入阻抗：无限大（开路）
- 输出等效电阻：零（实际上如果负载也是无限大可以不满足）
- 频率响应为固定值

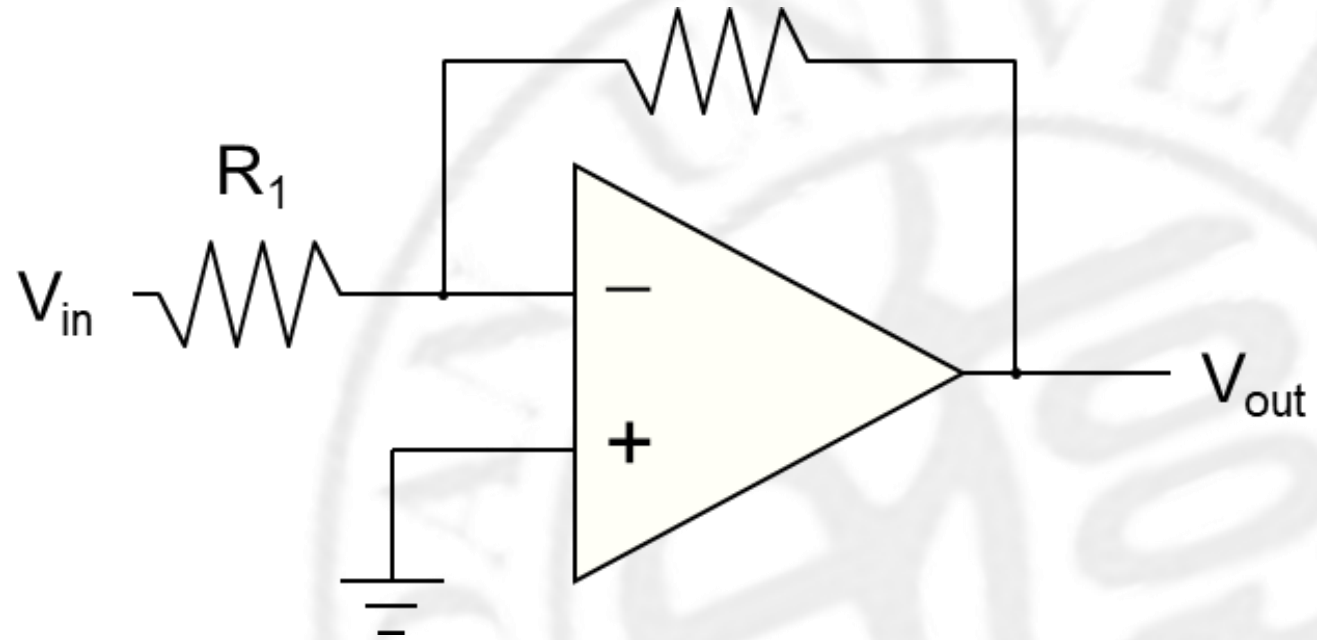
→ 虚短

→ 虚断



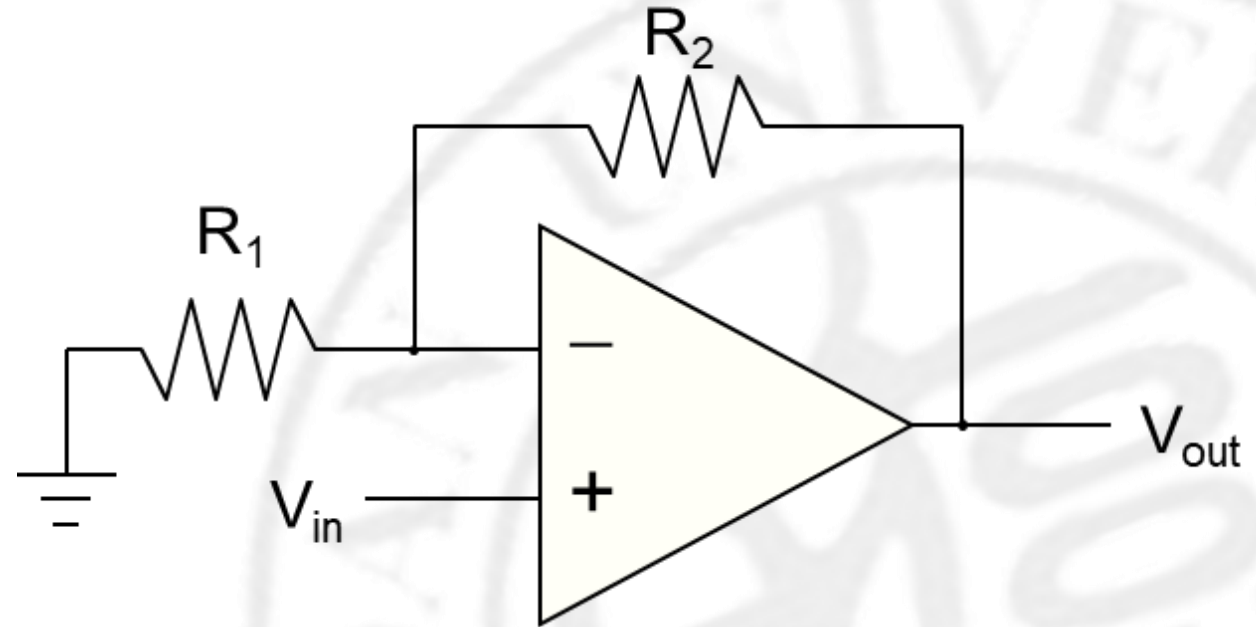
反向放大器

- Current through R_1 is $I_f = V_{in}/R_1$
- the current through R_1 must go through R_2
- Voltage drop across R_2 is then $I_f R_2 = V_{in} \times (R_2/R_1)$
- $V_{out} = 0 - V_{in} \times (R_2/R_1) = -V_{in} \times (R_2/R_1)$
- Amplify V_{in} by factor $-R_2/R_1$
 - negative sign \rightarrow “inverting”



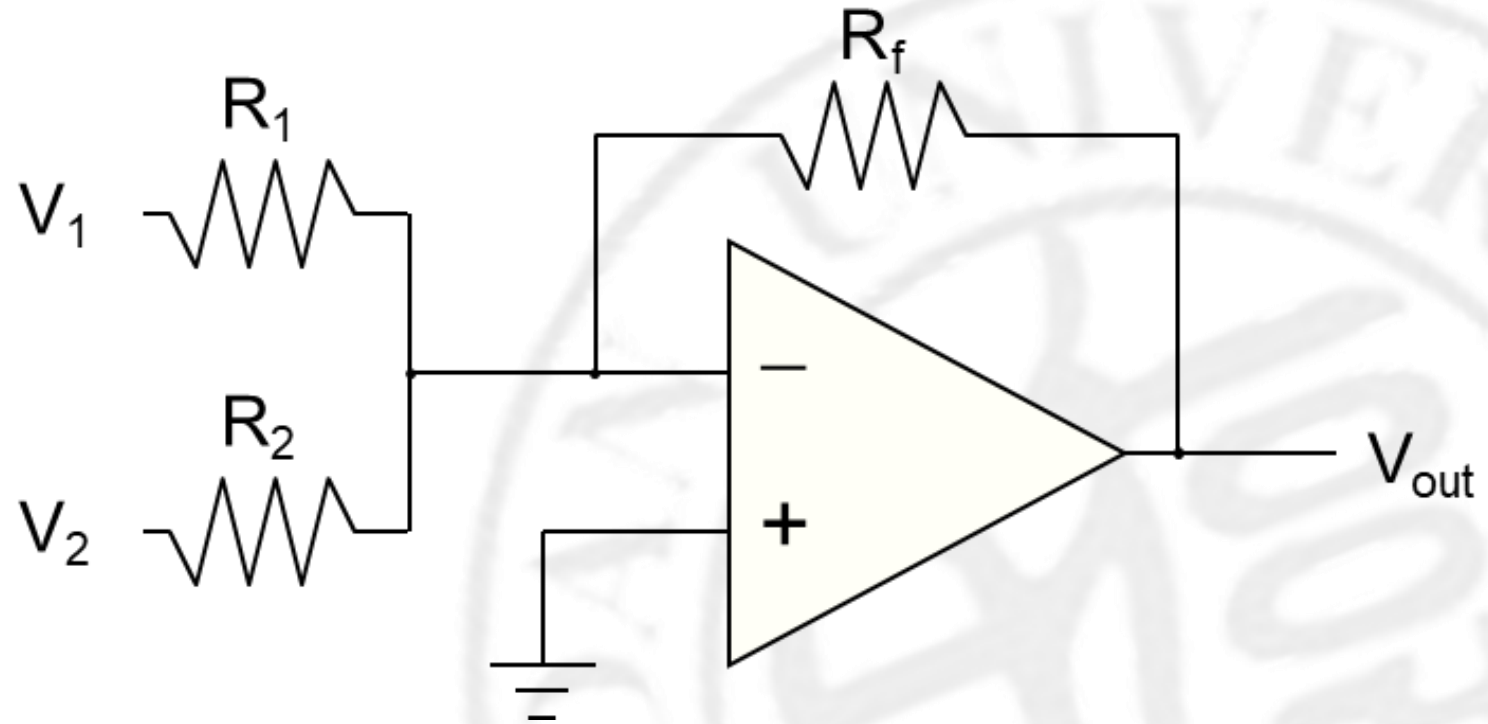
非反向放大器

- Current through R_1 is $I_f = V_{in}/R_1$ (to left, into ground)
- Voltage drop across R_2 is $I_f R_2 = V_{in} \times (R_2/R_1)$
- $V_{out} = V_{in} + V_{in} \times (R_2/R_1)$
 $= V_{in} \times (1 + R_2/R_1)$
- The gain is $(1 + R_2/R_1)$, and is positive (non-inverting)



模拟加权加法器

- inverting input still held at virtual ground
- I_1 and I_2 are added together to run through R_f
- Get the (inverted) sum:
$$V_{out} = -R_f \times (V_1/R_1 + V_2/R_2)$$
- 加权系数由电阻的比例决定



模拟加权减法器

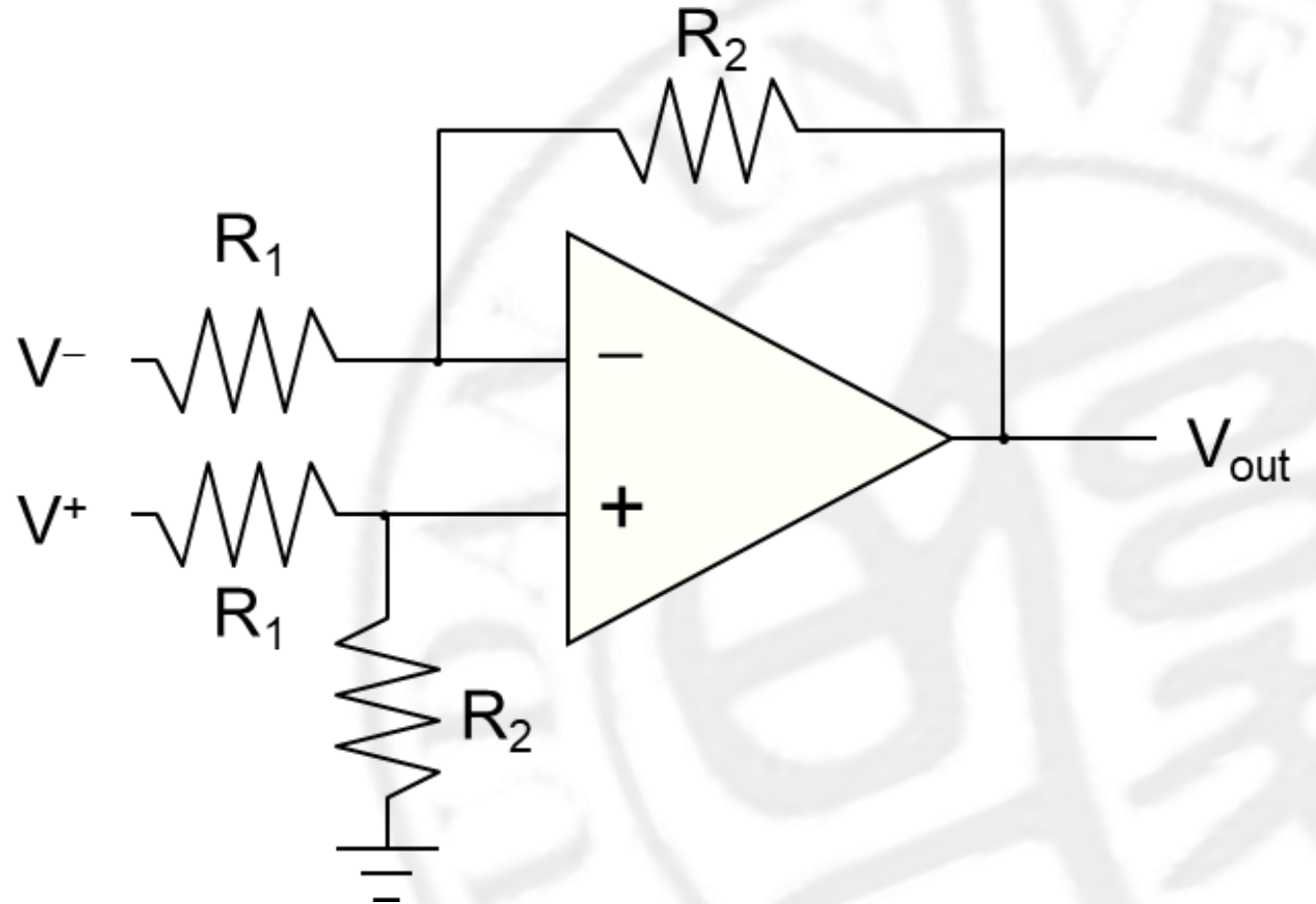
- 运放正端输入

$$V_{\text{node}} = V^+ R_2 / (R_1 + R_2)$$

- 运放负端输入

- $I_f = (V^- - V_{\text{node}}) / R_1$

- $V_{\text{out}} = (R_2 / R_1)(V^+ - V^-)$

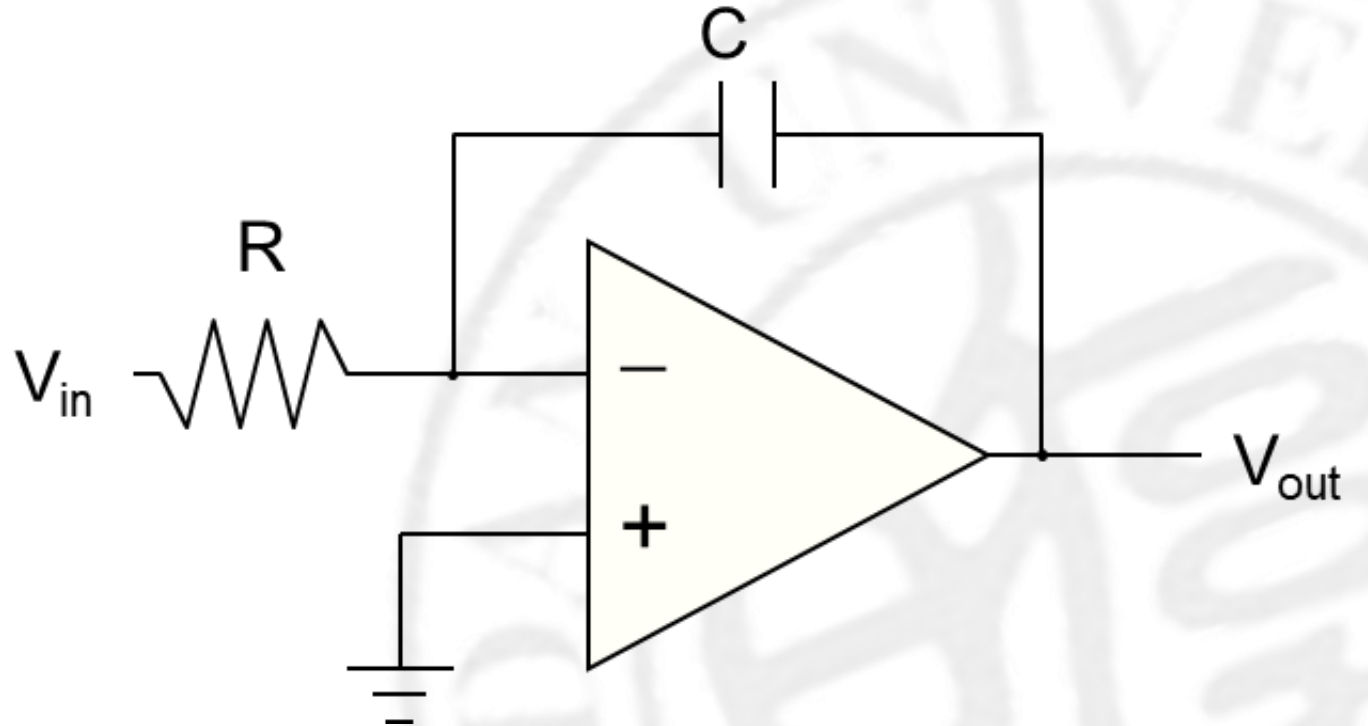


积分器（低通滤波器）

- $I_f = V_{in}/R$
- $C \cdot dV_{cap}/dt = V_{in}/R$

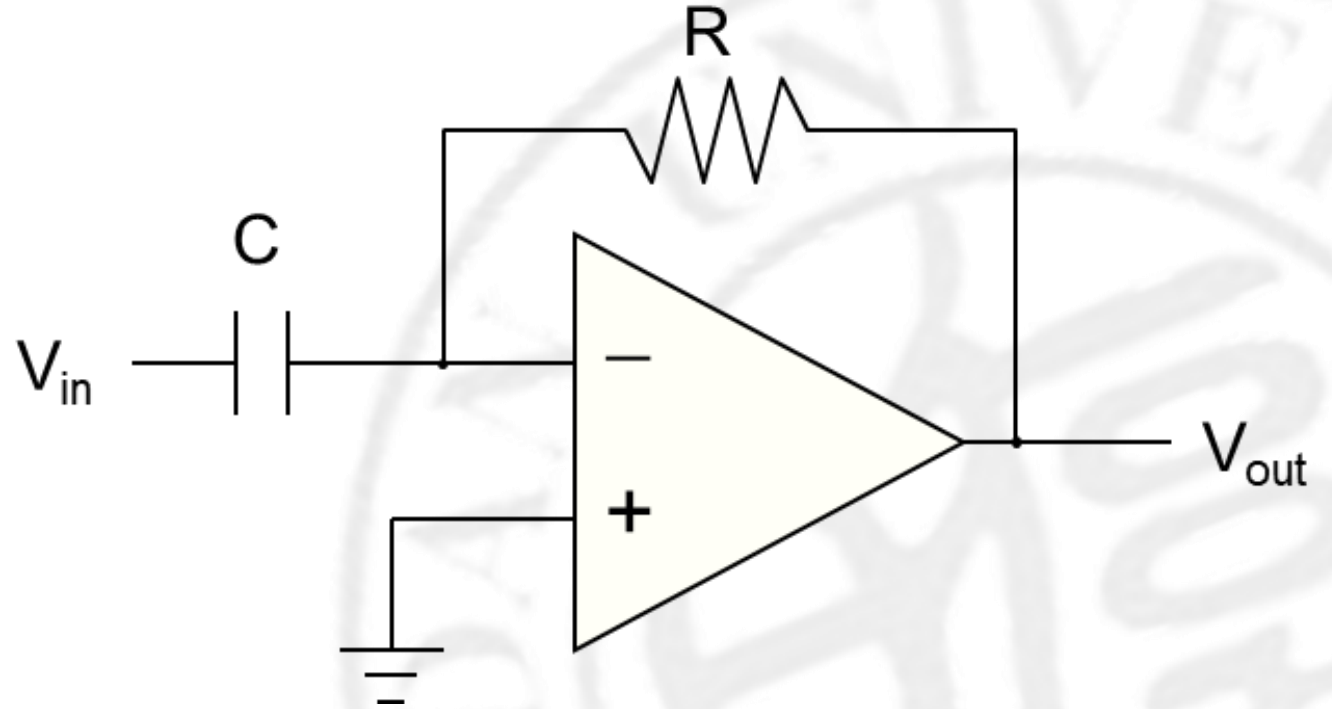
$$V_{out} = -\frac{1}{RC} \int V_{in} dt$$

- 将电容的等效阻值 $1/sC$
代入反向放大器

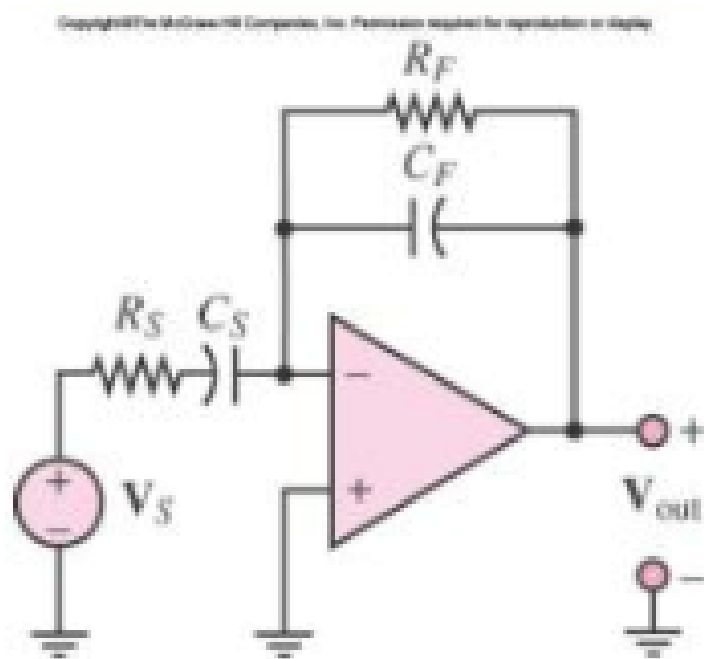


微分器（高通滤波器）

- $Q = CV$
- $I_{\text{cap}} = dQ/dt = C \cdot dV/dt$
- $V_{\text{out}} = -I_{\text{cap}}R = -RC \cdot dV/dt$
- 将电容的等效阻值 $1/sC$ 代入反向放大器



帶通濾波器



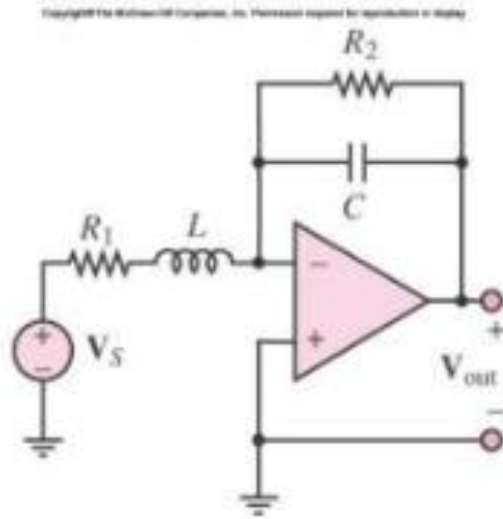
$$\frac{V_{out}}{V_S}(j\omega) = -\frac{Z_F}{Z_S}$$

$$Z_S = R_S + 1/j\omega C_S$$
$$= (1 + j\omega C_S R_S) / j\omega C_S$$

$$Z_F = R_F \parallel C_F$$
$$= \frac{R_F}{1 + j\omega C_F R_F}$$

$$\frac{V_{out}}{V_S}(j\omega) = -\frac{j\omega C_S R_F}{(1 + j\omega C_S R_S)(1 + j\omega C_F R_F)}$$

二阶低通滤波器



R_1 , R_2 , C and L are specially chosen so that:

$$\omega_0 = 1/(CR_2) = R_1/L.$$

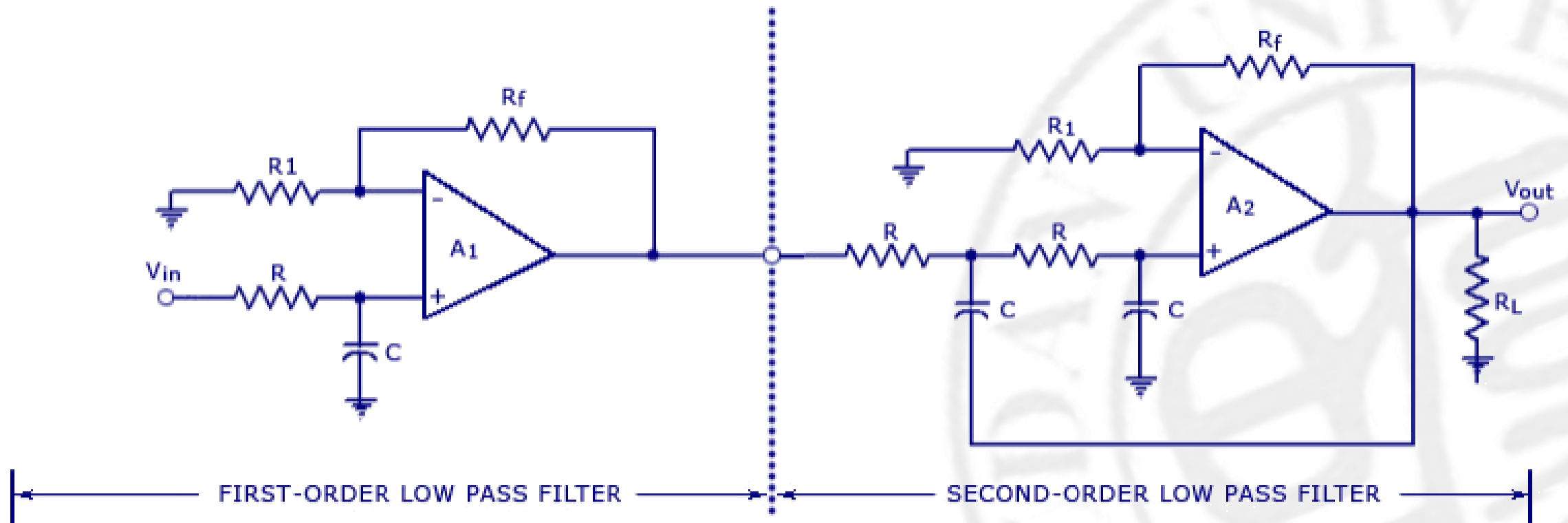
$$H_v(j\omega) = \frac{V_{out}}{V_S}(j\omega) = -\frac{R_2}{R_1(1 + j\omega/\omega_0)^2}$$

Above ω_0 , H_v is reduced by a factor of **100** for a ten fold increase in ω (40dB drop per decade)

First-order filter: H_v is reduced by a factor of **10** for a ten fold increase in ω (20dB drop per decade)

高阶低通滤波器

Third-Order Low pass Butterworth Filter Circuit



Pulse Width Modulator / 比较器

- Output changes when
 - $V_{in} \approx V_{pot}$
- Potentiometer used to vary duty cycle

