模拟与数字电路

Analog and Digital Circuits



课程主页 扫一扫

第二十一讲: 滤波器、运算放大器

Lecture 21: Filters and Operational Amps

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提纲

- 复习
 - 从时域到频域的变化?

- 无源滤波器
- 运算放大器
- 运算放大器的应用

波特图的总结

总结:

LHP:

$$\frac{1}{1+s/\omega_0}$$

$$\frac{1}{1+j(\omega/\omega_0)}$$

RHP:

$$\frac{1}{1-s/\omega_0}$$

$$\frac{1}{1-j(\omega/\omega_0)}$$

LHZ:

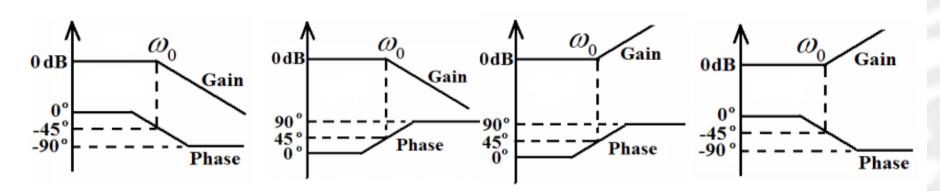
$$1+s/\omega_0$$

$$1+j(\omega/\omega_0)$$

RHZ:

$$1-s/\omega_0$$

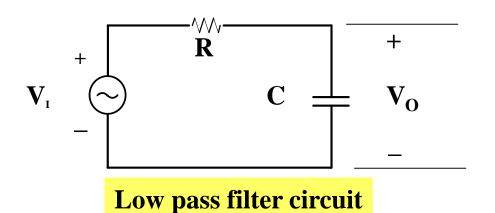
$$1+j(\omega/\omega_0)$$
 $1-j(\omega/\omega_0)$



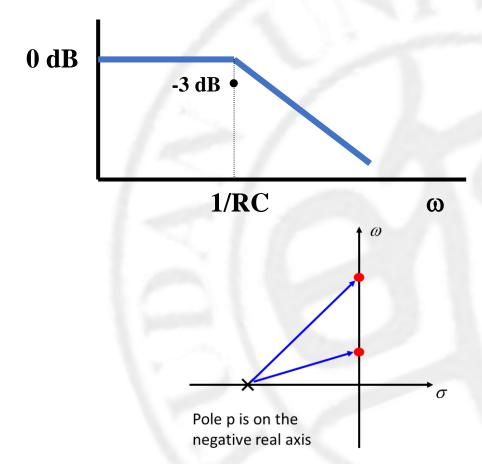
规律非常明显,即便不记得了,推一遍也并不难。

低通滤波器

• 一阶无源低通滤波器(first-order passive low-pass filter)

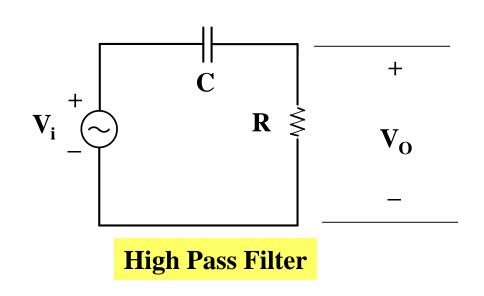


$$\frac{V_o(jw)}{V_i(jw)} = \frac{\frac{1}{jwC}}{R + \frac{1}{jwC}} = \frac{1}{1 + jwRC}$$

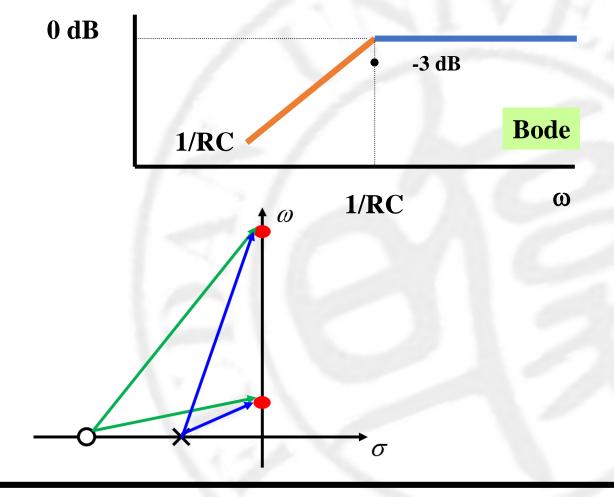


高通滤波器

• 一阶无源高通滤波器(first-order passive high-pass filter)

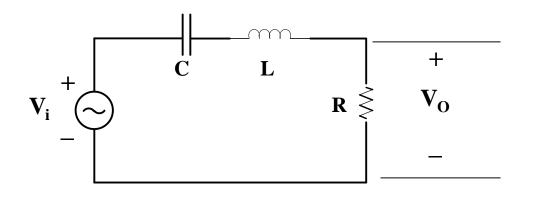


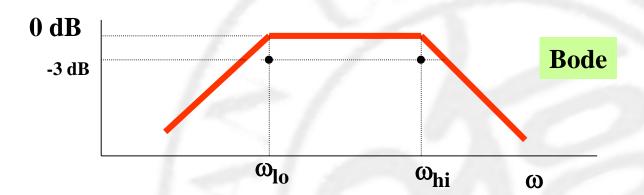
$$\frac{V_o(jw)}{V_i(jw)} = \frac{R}{R + \frac{1}{jwC}} = \frac{jwRC}{1 + jwRC}$$



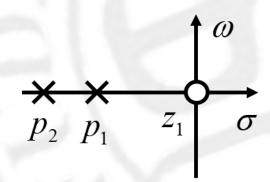
带通滤波器

• 无源带通滤波器(first-order passive band-pass filter)



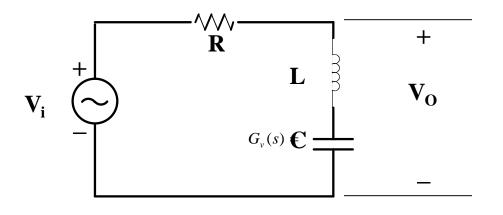


$$\frac{V_O(s)}{V_i(s)} = \frac{\frac{R}{L}s}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

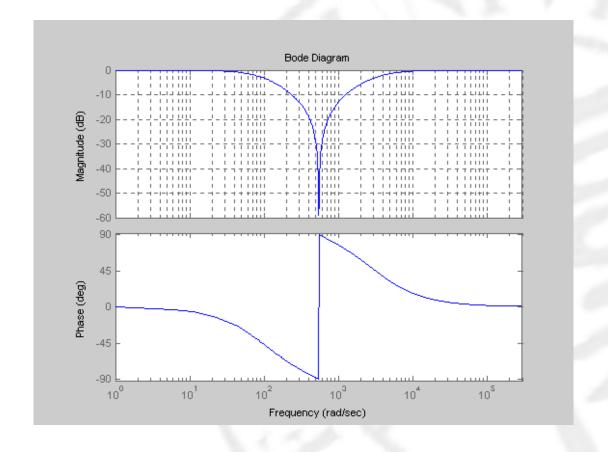


带阻滤波器

• 无源带阻滤波器(first-order passive band-stop filter)

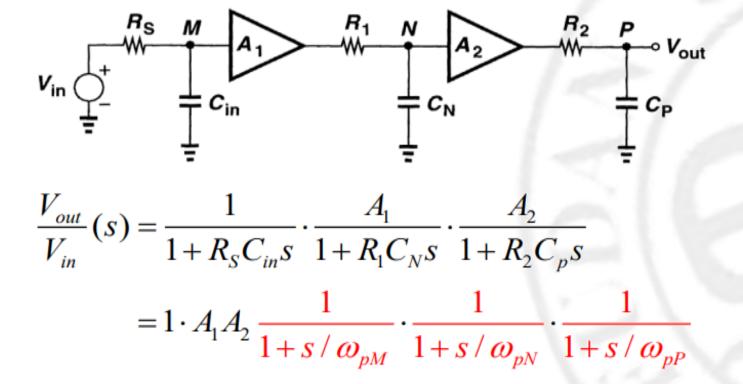


$$G_{v}(s) = \frac{s^{2} + \frac{1}{LC}}{s^{2} + \frac{R}{L}s + \frac{1}{LC}}$$



高阶有源滤波器

• 对于一个电路的结点,如果有电容和电阻连接在这个结点上,那么通常这个结点会给最终的表达式贡献一个极点:



放大器的非线性

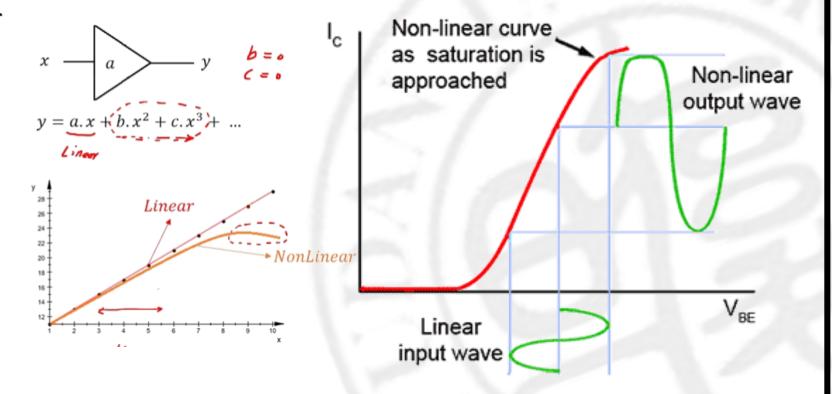
• 典型建模方案: 泰勒展开

Non-Linear Amplifier

- x = input
- y = output

$$y = a.x + b.x^2 + Cx^3 + ...$$

Generally c<0

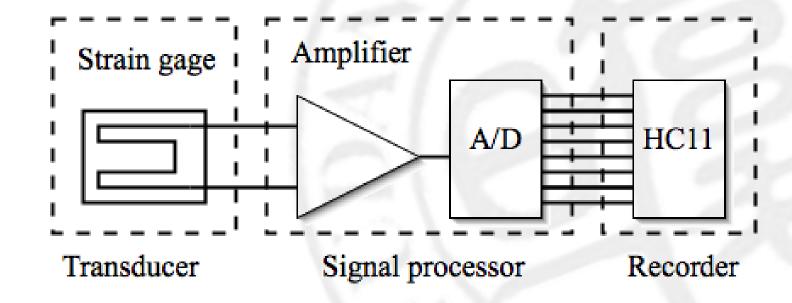


运算放大器

Sensor signals are often too weak or too noisy

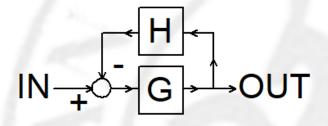
• 目标1: 以固定的比列线性放大信号

• 目标2: 将某些频率滤去



负反馈原理

- 负反馈: 将输出按一定比例反馈到输入
- 通过信号流图得到负反馈的表达式
- 当增益G不恒定,但保持远远大于1时
- 增益由H的大小决定
- 实际上当, GH>>1时 G的输入摆幅非常小



$$Gain = \frac{G}{1 + GH}$$

$$Gain = \frac{1}{H} \times \frac{GH}{1 + GH}$$

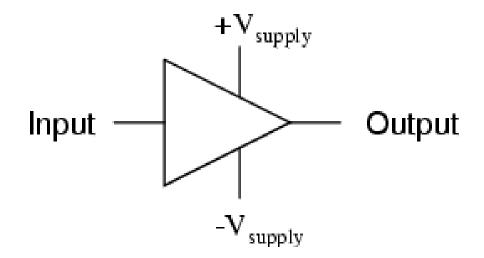
Loop gain is not GH.

Loop gain is -GH.

如何实现减法?

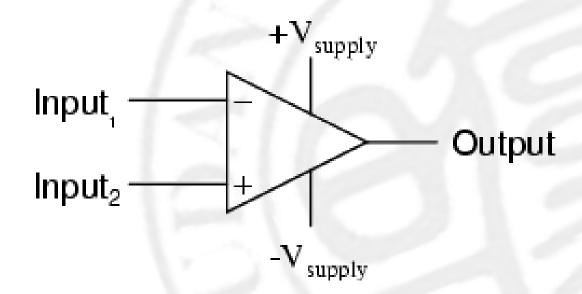
• 单端放大器

General amplifier circuit symbol



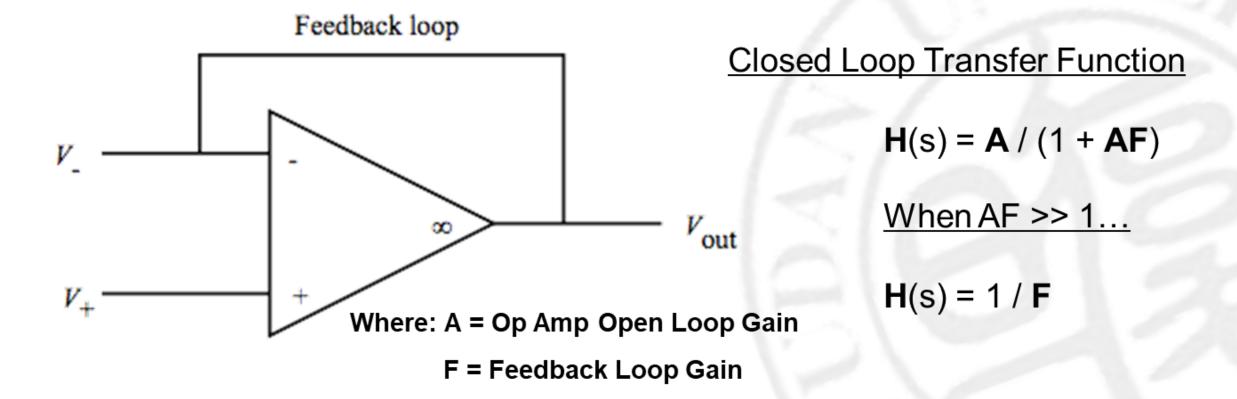
• 差分放大器

Differential amplifier



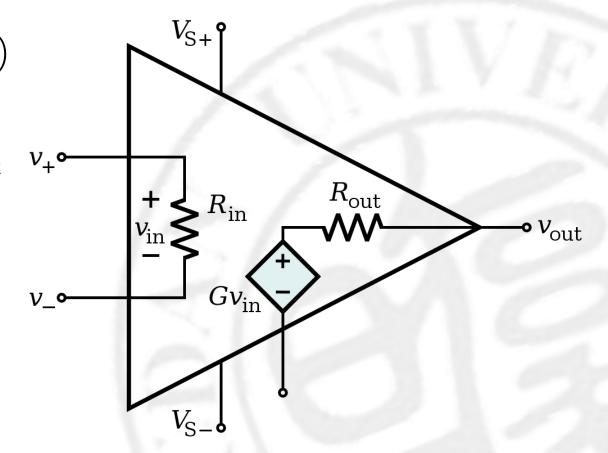
放大电路应用实例

• 将输出直接连接到负端输入



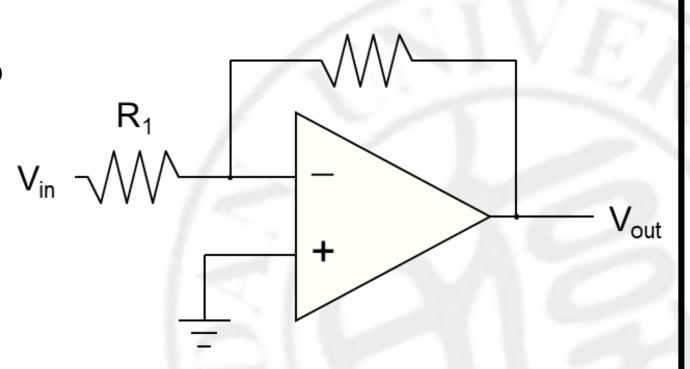
理想运算放大器参数

- 电压增益: 无限大 (大于 10000)
- 输入阻抗: 无限大 (开路)
- 输出等效电阻:零(实际上如果负载也是无线大可以不满足)
- 频率响应为固定值
- →虚短
- →虚断



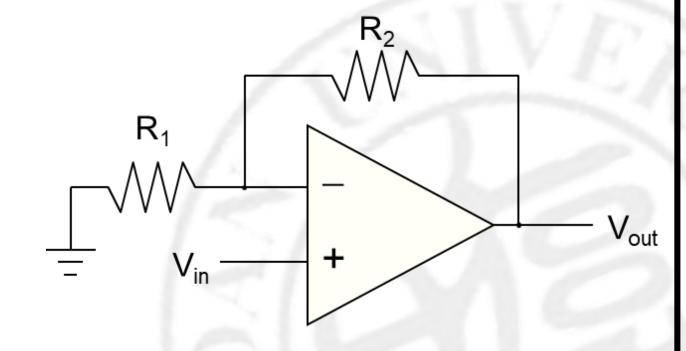
反向放大器

- Current through R_1 is $I_f = V_{in}/R_1$
- the current through R_1 must go through R_2
- Voltage drop across R_2 is then $I_f R_2 = V_{in} \times (R_2/R_1)$
- $V_{\text{out}} = 0 V_{\text{in}} \times (R_2/R_1) = -V_{\text{in}} \times (R_2/R_1)$
- Amplify V_{in} by factor $-R_2/R_1$
 - negative sign → "inverting"



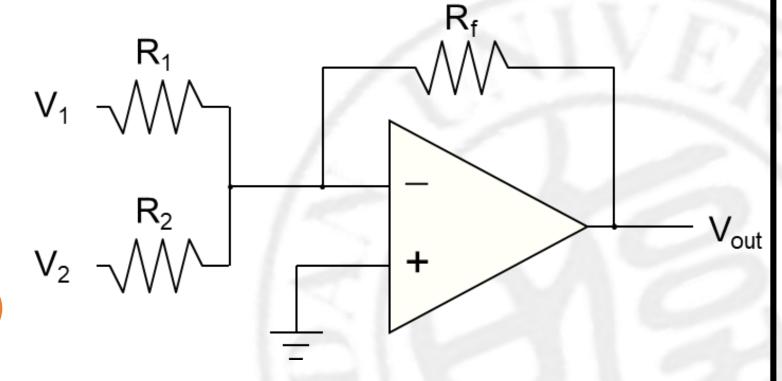
非反向放大器

- Current through R_1 is $I_f = V_{in}/R_1$ (to left, into ground)
- Voltage drop across R_2 is $I_f R_2 = V_{in} \times (R_2/R_1)$
- $V_{\text{out}} = V_{\text{in}} + V_{\text{in}} \times (R_2/R_1)$ = $V_{\text{in}} \times (1 + R_2/R_1)$
- The gain is $(1 + R_2/R_1)$, and is positive (non-inverting)



模拟加权加法器

- inverting input still held at virtual ground
- I₁ and I₂ are added together to run through R_f
- Get the (inverted) sum: $V_{\text{out}} = -R_f \times (V_1/R_1 + V_2/R_2)$



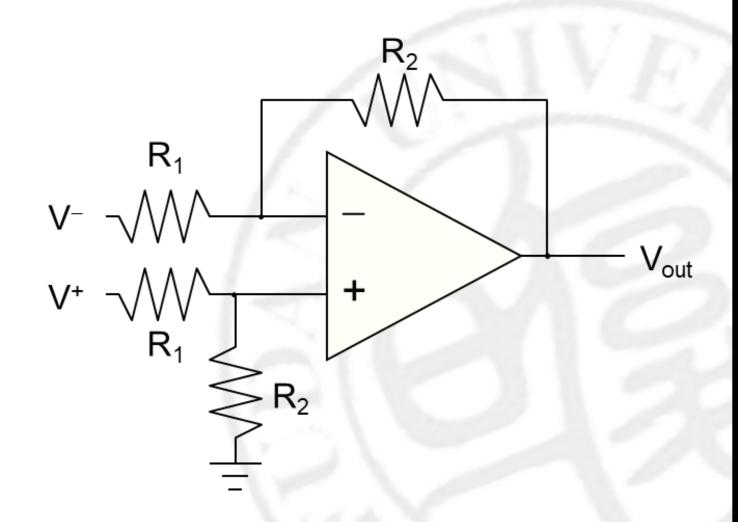
• 加权系数由电阻的比例决定

模拟加权减法器

• 运放正端输入

$$V_{\text{node}} = V^{+}R_{2}/(R_{1} + R_{2})$$

- 运放负端输入
- $I_{\rm f} = (V^- V_{\rm node})/R_1$
- $V_{\text{out}} = (R_2/R_1)(V^+ V^-)$



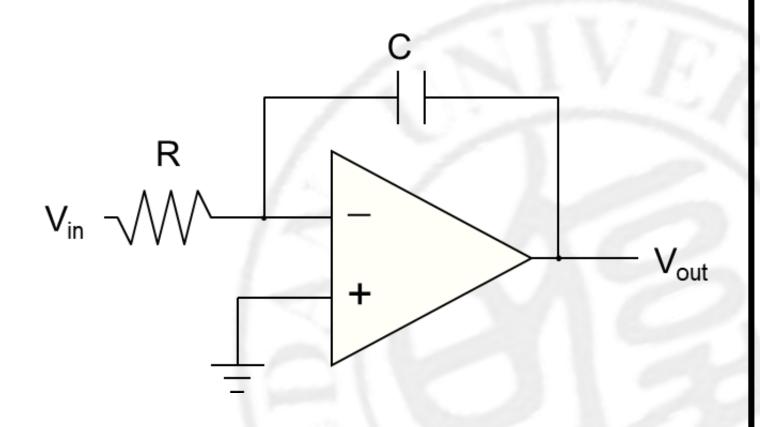
积分器 (低通滤波器)

•
$$I_f = V_{in}/R$$

• $C \cdot dV_{cap}/dt = V_{in}/R$

$$V_{
m out} = -rac{1}{RC}\int V_{in}dt$$

• 将电容的等效阻值1/sC 代入反向放大器



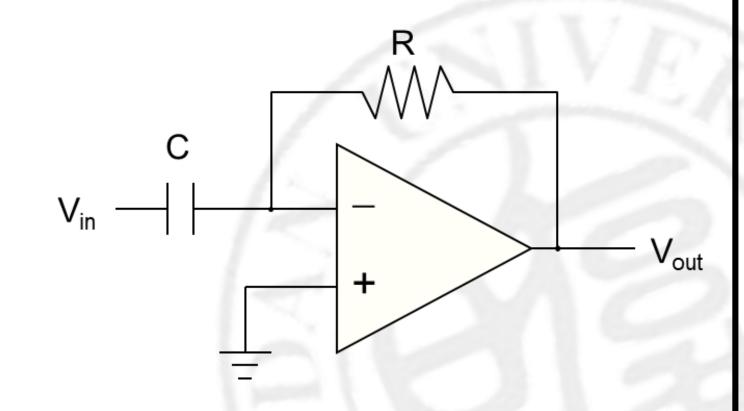
微分器 (高通滤波器)

•
$$Q = CV$$

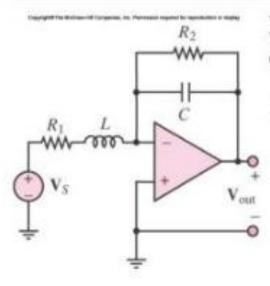
•
$$I_{cap} = dQ/dt = C \cdot dV/dt$$

•
$$V_{\text{out}} = -I_{\text{cap}}R = -RC \cdot dV/dt$$

• 将电容的等效阻值1/sC 代入反向放大器



二阶低通滤波器



 R_1 , R_2 , C and L are specially chosen so that: $\omega_0 = 1/(CR_2) = R_1/L$.

$$H_{v}(j\omega =) \frac{V_{out}}{V_{s}}(j\omega) = -\frac{R_{2}}{R_{1}(1+j\omega/\omega_{0})^{2}}$$

Above ω_0 , H_v is reduced by a factor of **100** for a ten fold increase in ω (40dB drop per decade)

First-order filter: H_v is reduced by a factor of 10 for a ten fold increase in ω (20dB drop per decade)

高阶低通滤波器

Third- Order Low pass Butterworth Filter Circuit

