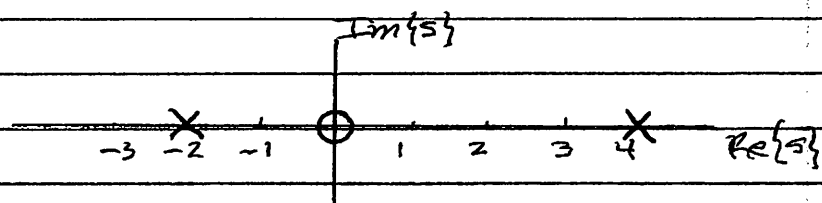


# SOLUTIONS TO EXAM #3 (5/10/18)

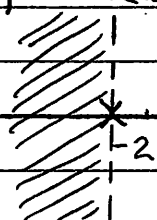
#1.

$$H(s) = \frac{6s}{s^2 - 2s - 8} = \frac{6s}{(s-4)(s+2)}$$

a.) poles ( $H(s) \rightarrow \infty$ ):  $s=4, s=-2$   
zeros ( $H(s) \rightarrow 0$ ):  $s=0, s=\infty$



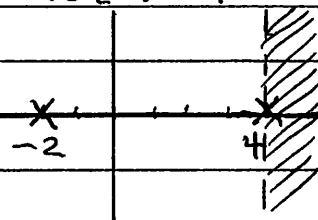
b.) (i)  $\text{Re}\{s\} < -2$



left of leftmost pole

LS signal

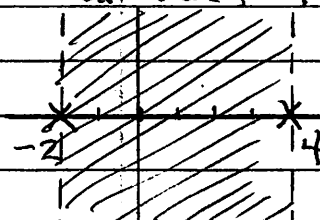
(ii)  $\text{Re}\{s\} > 4$



right of rightmost pole

RS signal (causal)

(iii)  $-2 < \text{Re}\{s\} < 4$



bounded by poles

as signal (stable)

$$c.) H(s) = \frac{6s}{(s-4)(s+2)} = \frac{A}{s-4} + \frac{B}{s+2}$$

$$A = H(s)(s-4) \Big|_{s=4} = \frac{6(4)}{(4+2)} = 4$$

$$B = H(s)(s+2) \Big|_{s=-2} = \frac{6(-2)}{(-2-4)} = 2$$

$$H(s) = \frac{4}{s-4} + \frac{2}{s+2}$$

H.c. cont'd)

iv)  $\operatorname{Re}\{s\} < -2 \Rightarrow$  both signals are left sided

$$h(t) = -4e^{4t}u(-t) - 2e^{-2t}u(-t)$$

v)  $\operatorname{Re}\{s\} > 4 \Rightarrow$  both signals are right sided

$$h(t) = 4e^{4t}u(t) + 2e^{-2t}u(t)$$

vi)  $-2 < \operatorname{Re}\{s\} < 4 \Rightarrow$  two sided signal

$$h(t) = -4e^{4t}u(-t) + 2e^{-2t}u(t)$$

$$d.) \quad H(s) = \frac{Y(s)}{X(s)} = \frac{6s}{s^2 - 2s - 8}$$

$$Y(s)(s^2 - 2s - 8) = 6sX(s)$$

$$\Downarrow \mathcal{L}^{-1}$$

$$\frac{d^2 y(t)}{dt^2} - 2 \frac{dy(t)}{dt} - 8y(t) = 6 \frac{dx(t)}{dt}$$

e.) No.

causal  $\rightarrow$  ROC is right-half planestable  $\rightarrow$  ROC includes the unit circle

iv) is not causal and not stable

v) is causal but not stable

vi) is stable but not causal

#2. LTI System

$$\text{input } x[n] = (-3)^n u[n]$$

$$\text{output } y[n] = 4(a)^n u[n] - \left(\frac{1}{2}\right)^n u[n]$$

a.) transfer function  $H(z) = \frac{Y(z)}{X(z)}$ 

$$\begin{aligned} Y(z) &= Z\{y[n]\} = \sum_{n=0}^{\infty} 4(a)^n z^{-n} - \sum_{n=0}^{\infty} \left(\frac{1}{2}\right)^n z^{-n} \\ &= 4 \sum_{n=0}^{\infty} (az^{-1})^n - \sum_{n=0}^{\infty} \left(\frac{1}{2}z^{-1}\right)^n \\ &= \frac{4}{1-az^{-1}} - \frac{1}{1-\frac{1}{2}z^{-1}} \end{aligned}$$

$$|z| > a \quad \cap \quad |z| > \frac{1}{2}$$

$$\begin{aligned} Y(z) &= \frac{4(1-\frac{1}{2}z^{-1}) - (1-az^{-1})}{(1-az^{-1})(1-\frac{1}{2}z^{-1})} \\ &= \frac{3}{(1-az^{-1})(1-\frac{1}{2}z^{-1})}, \quad |z| > a \end{aligned}$$

$$\begin{aligned} X(z) &= Z\{x[n]\} = \sum_{n=0}^{\infty} (-3)^n z^{-n} \\ &= \sum_{n=0}^{\infty} (-3z^{-1})^n = \frac{1}{1+3z^{-1}}, \quad |z| > 3 \end{aligned}$$

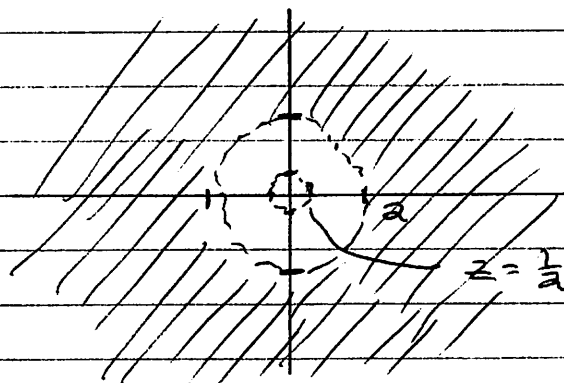
$$\therefore H(z) = \frac{Y(z)}{X(z)} = \frac{3(1+3z^{-1})}{(1-az^{-1})(1-\frac{1}{2}z^{-1})}$$

ROC:  $|z| > a$  outside outermost pole

#2 cont'd)

$$H(z) = \frac{3(1+3z^{-1})}{(1-2z^{-1})(1-\frac{1}{2}z^{-1})}$$

$$= \frac{3z(z+3)}{(z-2)(z-\frac{1}{2})}$$

poles:  $z=2, z=\frac{1}{2}$ zeros:  $z=0, z=-3 \Rightarrow$  didn't ask for these

a) impulse response

$$h[n] = \mathcal{Z}^{-1}\{H(z)\}$$

$$\leftarrow \frac{3(1+3z^{-1})}{(1-2z^{-1})(1-\frac{1}{2}z^{-1})} \quad \begin{matrix} |z| > 2 \\ \text{RS} \\ \text{sequence} \end{matrix}$$

$$H(z) = \frac{A}{1-2z^{-1}} + \frac{B}{1-\frac{1}{2}z^{-1}}$$

$$A = H(z)(1-2z^{-1}) \Big|_{z=2} = \frac{3(1+\frac{3}{2})}{1-\frac{1}{4}} = \frac{3(\frac{5}{2})}{\frac{3}{4}} = 10$$

$$B = H(z)(1-\frac{1}{2}z^{-1}) \Big|_{z=\frac{1}{2}} = \frac{3(1+6)}{1-4} = -7$$

$$H(z) = \frac{10}{1-2z^{-1}} - \frac{7}{1-\frac{1}{2}z^{-1}}$$

$$\uparrow \mathcal{Z}^{-1}$$

$$h[n] = 10(2^n)u[n] - 7\left(\frac{1}{2}\right)^n u[n]$$

#2. cont'd) d.)  $H(z) = \frac{Y(z)}{X(z)} = \frac{3(1+3z^{-1})}{(1-2z^{-1})(1-\frac{1}{2}z^{-1})} = \frac{3+9z^{-1}}{1-\frac{5}{2}z^{-1}+z^{-2}}$

$$Y(z)(1-\frac{5}{2}z^{-1}+z^{-2}) = X(z)(3+9z^{-1})$$

$$\Downarrow z^{-1}$$

$$y[n] - \frac{5}{2}y[n-1] + y[n-2] = 3x[n] + 9x[n-1]$$

e.) The system is not stable because the ROC does not contain the unit circle. (Or, you can simply compute  $\sum_n |h[n]|^2 < \infty$ )

#3 Q.)  $H(s) = \frac{s+2}{s+1}$ ,  $\text{Re}\{s\} > -1$   
right-sided signal

$$= 1 + \frac{1}{s+1}$$

$$\Downarrow z^{-1}$$

$$h(t) = \delta(t) + e^{-t} u(t)$$

Alternative Approach:

$$H(s) = \frac{s}{s+1} + \frac{2}{s+1}$$

$$h(t) = \frac{d}{dt}(e^{-t} u(t)) + 2e^{-t} u(t)$$

$$= -e^{-t} u(t) + e^{-t} \delta(t) + 2e^{-t} u(t)$$

$$= \delta(t) + e^{-t} u(t)$$

3 cont'd) b.) Let  $x_1(t) = e^{-t} u(t)$

$$X_1(s) = \frac{1}{s+1}, \quad \operatorname{Re}\{s\} > -1$$

Let  $x_2(t) = e^{-(t+1)} u(t+1)$ . Then,  
 $x_2(t) = x_1(t+1)$  and

$$X_2(s) = e^s \cdot \frac{1}{s+1}, \quad \operatorname{Re}\{s\} > -1 \quad (\text{Prop. 9.5.2})$$

Let  $x_3(t) = \frac{1}{2} x_2(t)$ . Then,

$$X_3(s) = \frac{1}{2} X_2(s) = e^s \frac{1}{2(s+1)}, \quad \operatorname{Re}\{s\} > -1$$

(Prop 9.5.7)

Finally,  $x(t) = e^{-t} x_3(t)$ . Using Prop. 9.5.3  
 with a shift  $s_0 = -1$

$$X(s) = X_3(s+1) = \frac{e^{s+1}}{s+2}, \quad \operatorname{Re}\{s\} > -2$$

c.)  $x[n] = x_1[n] * x_2[n]$

where  $x_1[n] = u[n-2]$

and  $x_2[n] = \left(\frac{2}{3}\right)^n u[n]$

Then,  $X(z) = X_1(z) \cdot X_2(z)$

$$X_1(z) = \sum_{n=2}^{\infty} z^{-n} = \frac{z^{-2}}{1-z^{-1}}, \quad |z| > 1$$

$$X_2(z) = \sum_{n=0}^{\infty} \left(\frac{2}{3}\right)^n z^{-n} = \sum_{n=0}^{\infty} \left(\frac{2}{3} z^{-1}\right)^n$$

$$= \frac{1}{1 - \frac{2}{3} z^{-1}}, \quad |z| > \frac{2}{3}$$

$$\therefore X(z) = \frac{z^{-2}}{(1-z^{-1})(1-\frac{2}{3}z^{-1})} = \frac{1}{(z-1)(z-\frac{2}{3})}, \quad |z| > 1$$

\*3. cont'd)

d.) (4) Initial Value Theorem

$$\begin{aligned}
 x(0^+) &= \lim_{s \rightarrow \infty} s X(s) = \lim_{s \rightarrow \infty} e^{-5s} \left( \frac{-2(s+3)}{s+2} \right) \\
 &= \lim_{s \rightarrow \infty} \underbrace{e^{-5s}}_0 \lim_{s \rightarrow \infty} \underbrace{\left( \frac{-2(1+3/s)}{(1+2/s)} \right)}_{-2} \\
 &= 0
 \end{aligned}$$

(4) Final Value Theorem

$$\begin{aligned}
 \lim_{t \rightarrow \infty} x(t) &= \lim_{s \rightarrow 0} s X(s) \\
 &= \lim_{s \rightarrow 0} e^{-5s} \left( \frac{-2(s+3)}{s+2} \right) \\
 &= \frac{1(-2)(3)}{2} = -3
 \end{aligned}$$

$$e.) H(z) = \frac{2z(4z^2-1)}{8z^3-26z^2+5z+3}, \quad \frac{1}{2} < |z| < 3$$

Yes. This system is stable because the ROC includes the unit circle.

f.) Let  $y(t) = e^{2t} \sin t u(-t)$ . Then,  $x(t) = \int_t^\infty y(\tau) d\tau$  and  $X(s) = \frac{1}{s} Y(s)$  with  $\text{ROC}, R_y \cap \text{Re}\{s\} > 0$

$$\begin{aligned}
 Y(s) &= \int_0^\infty e^{2t} \sin t e^{-st} dt \\
 &= \frac{1}{2j} \left[ \int_0^\infty e^{(2-s+j)t} dt + \int_0^\infty e^{(2-s-j)t} dt \right] \\
 &= \frac{1}{2j} \left[ \frac{e^{(2-s+j)t}}{(2-s+j)} - \frac{e^{(2-s-j)t}}{(2-s-j)} \right]_{t=0}^\infty, \quad \text{Re}\{s\} < 2
 \end{aligned}$$

#3 (cont'd)

$$Y(s) = \frac{-1}{(s-a)^2 + 1}, \quad \text{Re}\{s\} < a$$

$$\therefore X(s) = \frac{-1}{s[(s-a)^2 + 1]}, \quad 0 < \text{Re}\{s\} < a$$

#4.

$$y[n] = x[n] + a x[n-1], \quad |a| < 1$$

$$\Downarrow \mathcal{Z}$$

$$\begin{aligned} Y(z) &= X(z) + a z^{-1} X(z) \\ &= X(z) (1 + a z^{-1}) \end{aligned}$$

$$\therefore H(z) = \frac{Y(z)}{X(z)} = 1 + a z^{-1} \Rightarrow \text{ROC is the entire } z\text{-plane except for } z=0$$

$$\Downarrow \mathcal{Z}^{-1}$$

$$h[n] = \delta[n] + a \delta[n-1]$$

NOTE: You can get the impulse directly from the difference equation by replacing  $x[n]$  by an impulse.

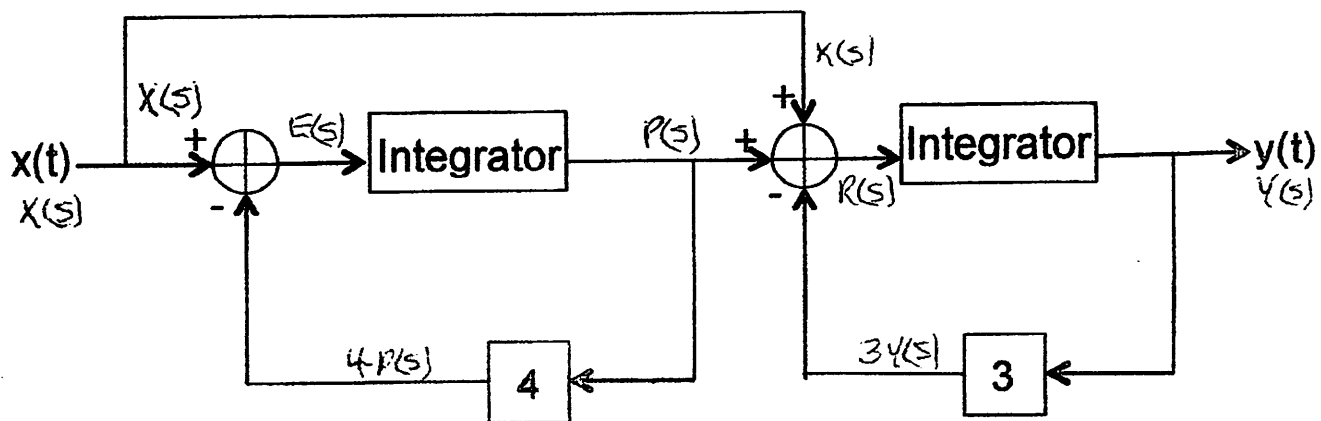
$$\text{i.e. } h[n] \stackrel{\Delta}{=} y[n] \Big|_{x[n] = \delta[n]}$$

$$= \delta[n] + a \delta[n-1]$$



Extra Credit

Label variables as shown on the figure



Then, we can write the following equations:

$$E(s) = X(s) - 4P(s)$$

$$\text{But } P(s) = \frac{1}{s} E(s) \Rightarrow E(s) = sP(s)$$

$$\therefore sP(s) = X(s) - 4P(s)$$

$$(s+4)P(s) = X(s)$$

$$P(s) = \frac{X(s)}{s+4}$$

$$R(s) = X(s) + P(s) - 3Y(s)$$

$$\text{But } Y(s) = \frac{1}{s} R(s) \Rightarrow R(s) = sY(s)$$

$$\therefore sY(s) = X(s) + P(s) - 3Y(s)$$

$$= X(s) + \frac{X(s)}{s+4} - 3Y(s)$$

$$(s+3)Y(s) = \frac{s+4+1}{s+4} X(s) = \frac{s+5}{s+4} X(s)$$

$$H(s) = \frac{Y(s)}{X(s)} = \frac{s+5}{(s+3)(s+4)} = \frac{s+5}{s^2+7s+12}$$