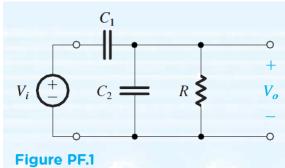


# Problem F.1 original solution

Find the transfer function  $T(s) = V_o(s) / V_i(s)$  of the circuit in Fig. PF.1. Is this an STC network? If so, of what type? For  $C_1 = C_2 = 0.5 \mu F$  and  $R = 100 k\Omega$ , find the location of the pole(s) and zero(s), and sketch Bode plots for the magnitude response and the phase response.



$$V_{o}(s) = V_{i}(s) \frac{1}{\chi_{C2} \parallel R + \chi_{C1}} \qquad \chi_{C2} \parallel R + \chi_{C1}$$

$$V_{o}(s) = V_{i}(s) \frac{\chi_{C2} \parallel R + \chi_{C1}}{\chi_{C2} \parallel R + \chi_{C1}} \qquad \chi_{C2} \parallel R + \chi_{C1}$$

esponse and the phase response. 
$$V_{o}(s) = V_{i}(s) \frac{\chi_{C2} \parallel R}{\chi_{C2} \parallel R + \chi_{C1}} \qquad \chi_{C2} \parallel R = \frac{\frac{1}{sC_{2}}R}{\frac{1}{sC_{2}} + R} = \frac{R}{1 + sC_{2}R}$$

$$T(s) \equiv \frac{V_{o}(s)}{V_{i}(s)} = \frac{\frac{s}{(1 + C_{2}/C_{1})}}{s + \frac{1}{RC_{1}}(1 + \frac{C_{2}}{C_{1}})} \qquad T(s) = \frac{Ks}{s + \omega_{0}}$$
STC highpass

$$T(s) = \frac{Ks}{s + \omega_0}$$

STC highpass

For 
$$C_1 = C_2 = 0.5 \mu F$$
 and  $R = 100 k\Omega$ 

$$K = 1/(1+0.5 \text{uF}/0.5 \text{uF}) = 0.5$$

$$\omega_o = 1/[100\text{k}\Omega \times 0.5\text{uF}(1+0.5\text{uF}/0.5\text{uF})] = 10 \text{ rad/s}$$

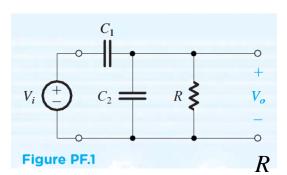
$$f_o = \omega_o / 2\pi = 1.59 \text{ Hz}$$

zeros: 
$$s = 0$$

poles: 
$$s = -10 \text{ rad/s}$$

## Problem F.1 worked through

Find the transfer function  $T(s) = V_o(s) / V_i(s)$  of the circuit in Fig. PF.1. Is this an STC network? If so, of what type? For  $C_1 = C_2 = 0.5 \mu F$  and  $R = 100 k\Omega$ , find the location of the pole(s) and zero(s), and sketch Bode plots for the magnitude response and the phase response.



Tesponse and the phase response.
$$V_{o}(s) = V_{i}(s) \frac{R \parallel \chi_{C2}}{\chi_{C1} + (R \parallel \chi_{C2})}$$

$$R \parallel \chi_{C2} = \frac{R \times \frac{1}{sC_{2}}}{R + \frac{1}{sC_{2}}} \left[ \frac{sC_{2}}{sC_{2}} \right] = \frac{R}{1 + sRC_{2}}$$

$$T(s) = \frac{V_{o}(s)}{V_{i}(s)} = \frac{R \parallel \chi_{C2}}{\chi_{C1} + (R \parallel \chi_{C2})}$$

$$T(s) = \frac{V_o(s)}{V_i(s)} = \frac{1}{\frac{1}{sC_1} + \frac{R}{1 + sRC_2}} \left[ \frac{1 + sRC_2}{1 + sRC_2} \right] = \frac{R}{R + \frac{1 + sRC_2}{sC_1}} \left[ \frac{sC_1}{sC_1} \right] = \frac{sRC_1}{1 + sRC_2 + sRC_1}$$

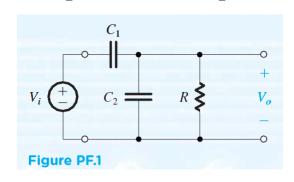
$$= \frac{sRC_1}{sR(C_2 + C_1) + 1} \left[ \frac{\frac{1}{R(C_2 + C_1)}}{\frac{1}{R(C_2 + C_1)}} \right] = \frac{\frac{sRC_1}{R(C_2 + C_1)}}{\frac{1}{R(C_2 + C_1)}} = \frac{s\frac{C_1}{C_1 + C_2}}{s + \frac{1}{R(C_2 + C_1)}}$$

$$= \frac{sRC_1}{R(C_2 + C_1)} = \frac{s\frac{C_1}{C_1 + C_2}}{s + \frac{1}{R(C_2 + C_1)}}$$
STC highpass

$$T(s) = \frac{Ks}{s + \omega_0}$$

## Problem F.1 worked through

Find the transfer function  $T(s) = V_o(s) / V_i(s)$  of the circuit in Fig. PF.1. Is this an STC network? If so, of what type? For  $C_1 = C_2 = 0.5 \mu F$  and  $R = 100 k\Omega$ , find the location of the pole(s) and zero(s), and sketch Bode plots for the magnitude response and the phase response.



$$T(s) = \frac{V_o(s)}{V_i(s)} = \frac{s \frac{C_1}{C_1 + C_2}}{s + \frac{1}{R(C_2 + C_1)}}$$

$$T(s) = \frac{Ks}{s + \omega_0}$$

STC highpass

For 
$$C_1 = C_2 = 0.5 \mu F$$
 and  $R = 100 k\Omega$ 

$$K = 0.5 \text{uF}/(0.5 \text{uF} + 0.5 \text{uF}) = 0.5$$

$$\omega_o = 1/[100\text{k}\Omega \times (0.5\text{uF} + 0.5\text{uF})] = 10 \text{ rad/s}$$

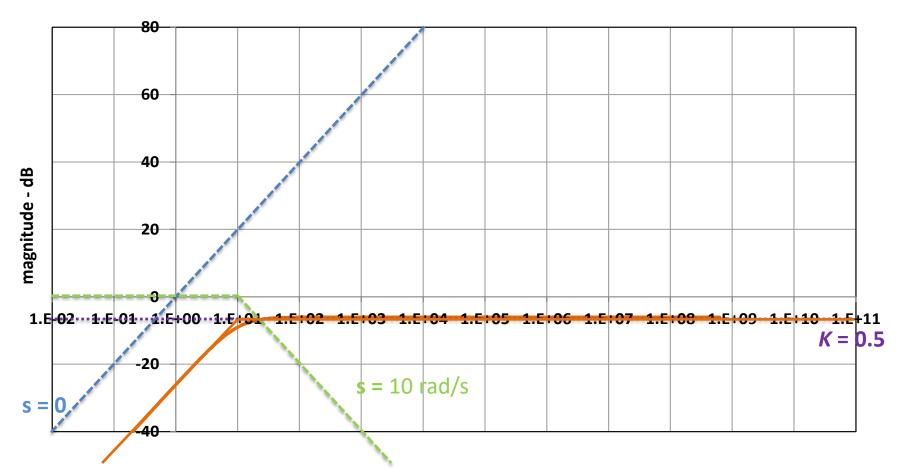
$$f_o = \omega_o / 2\pi = 1.59 \text{ Hz}$$

zeros: 
$$s = 0$$

poles: 
$$s = -10 \text{ rad/s}$$

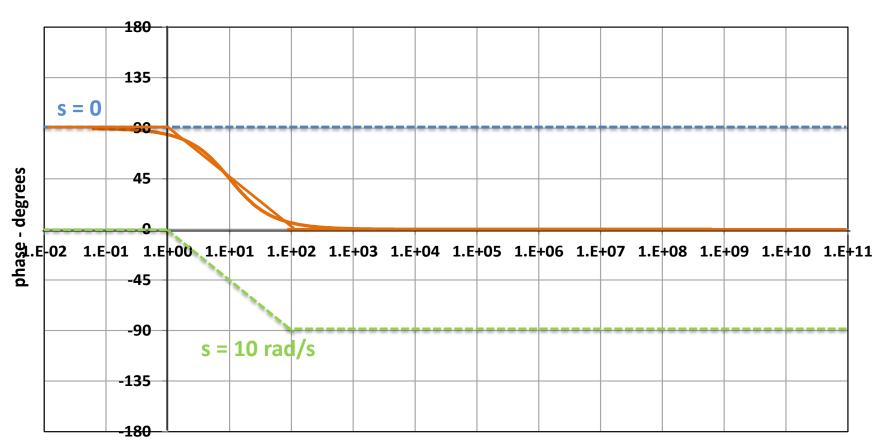
zeros: s = 0

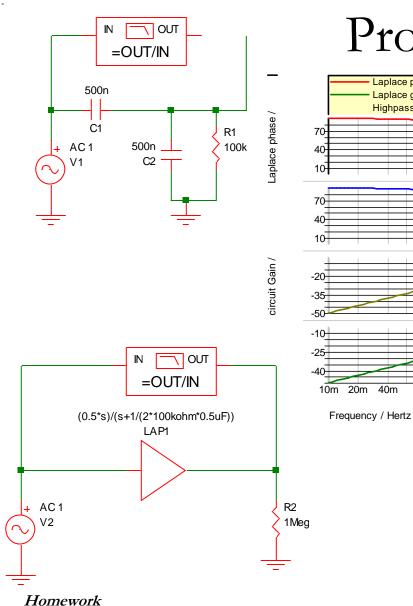
poles: s = 10 rad/s

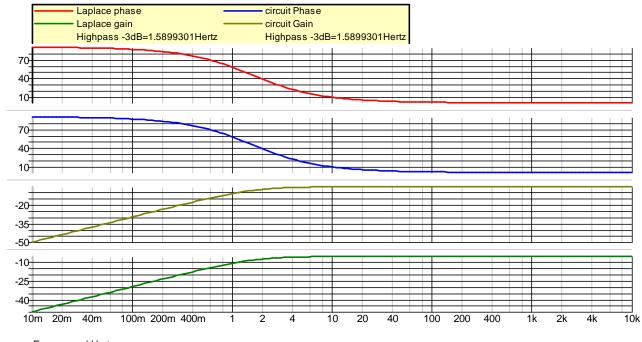


zeros: s = 0

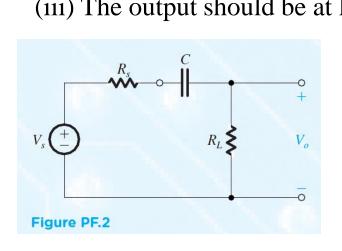
poles: s = 10 rad/s







- (a) Find the voltage transfer function  $T(s) = V_o(s) / V_i(s)$ , for the STC network shown in Fig. PF.2.
- (b) In this circuit, capacitor C is used to couple the signal source  $V_s$  having a resistance  $R_s$  to a load  $R_L$ . For  $R_s = 10 \text{ k}\Omega$ , design the circuit, specifying the values of  $R_L$  and C to only one significant digit to meet the following requirements:
- (i) The load resistance should be as small as possible.
- (ii) The output signal should be at least 70% of the input at high frequencies.
- (iii) The output should be at least 10% of the input at 10 Hz.



$$V_o = \frac{R_L}{R_L + \chi_C} V_i$$

$$T(s) \equiv \frac{V_o}{V_i} = \frac{R_L sC}{sCR_L + 1}$$

- (b) In this circuit, capacitor C is used to couple the signal source  $V_s$  having a resistance  $R_s$  to a load  $R_L$ . For  $R_s = 10 \text{ k}\Omega$ , design the circuit, specifying the values of  $R_L$  and C to only one significant digit to meet the following requirements:
- (i) The load resistance should be as small as possible.
- (ii) The output signal should be at least 70% of the input at high frequencies.
- (iii) The output should be at least 10% of the input at 10 Hz.

$$T(s) = \frac{R_L sC}{1 + R_L sC}$$

$$\frac{R_L}{R_L + R_S} = \frac{R_L}{R_L + 10k\Omega} \ge 0.70 \qquad R_L \ge 23.3k\Omega$$

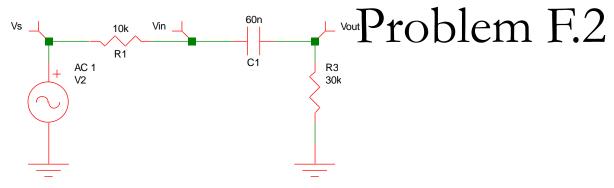
 $R_L$  to one significant digit is  $R_L = 30 \text{k}\Omega$ 

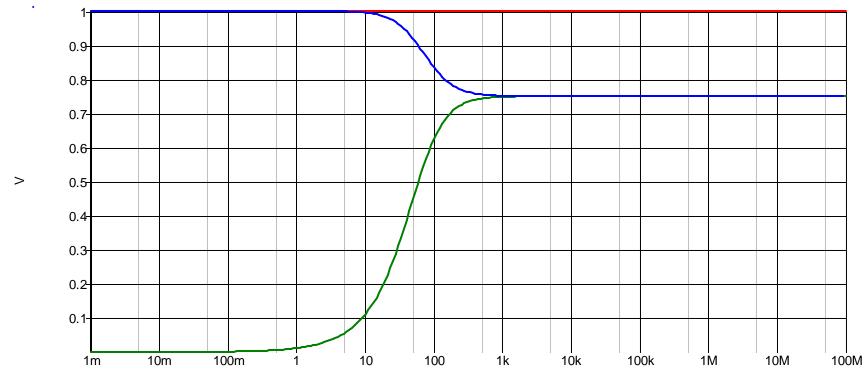
$$V_s$$
 $C$ 
 $+$ 
 $V_o$ 
 $+$ 
 $V_o$ 
 $+$ 
Figure PF.2

$$T(2\pi 10 \text{Hz}) = \frac{(30\text{k}\Omega)(2\pi 10 \text{Hz})C}{1 + (30\text{k}\Omega)(2\pi 10 \text{Hz})C} \ge 0.10$$

C > 58.9 nF

C to one significant digit is C = 60 nF





Homework

Frequency / Hertz

R. Martin

Sketch Bode plots for the magnitude and phase of the transfer function

$$T(s) = \frac{10^4 \left(1 + \frac{s}{10^5}\right)}{(1 + \frac{s}{10^3})(1 + \frac{s}{10^4})}$$

From your sketches, determine approximate values for the magnitude and phase at  $\omega = 10^6$  rad/s. What are the exact values determined from the transfer function?

zeros:  $s = -10^5 \text{ rad/s}$ ,  $\infty$ 

poles:  $s = -10^3 \text{ rad/s}, -10^4 \text{ rad/s}$ 

determine approximate values for the magnitude and phase at  $\omega = 10^6$  rad/s.

Mag from plot on next slide at  $10^6$  rad/s ~ 0 dB, Phase ~ -90deg

Solving the equations (in MathCAD) for  $s = j10^6$  rad/s yields mag = 0.043 dB and phase = -95.08 deg

**MathCAD** 

$$T(s) := \frac{10^4 \cdot \left(1 + \frac{s}{10^5}\right)}{\left(1 + \frac{s}{10^3}\right) \cdot \left(1 + \frac{s}{10^4}\right)}$$

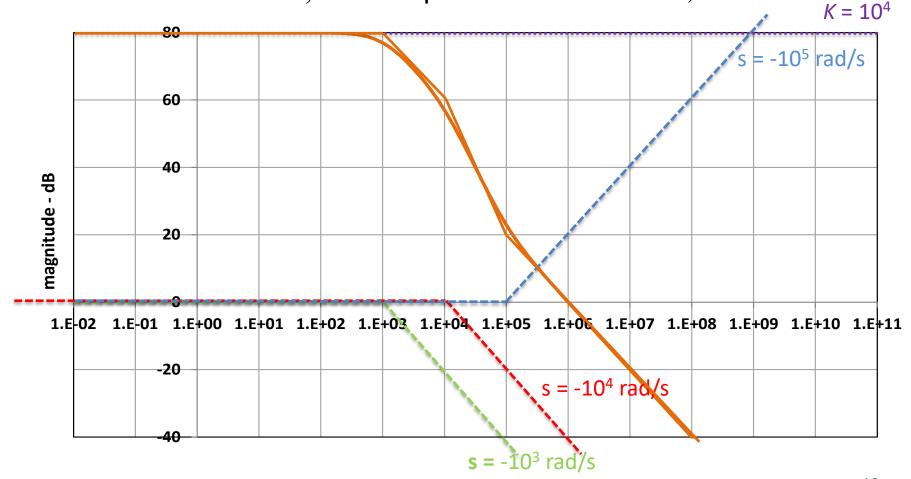
s1 := 
$$i \cdot 10^6$$
  
 $T(s1) = -0.089 - 1.001j$   
 $20 \cdot log(|T(s1)|) = 0.043$ 

arg(T(s1)) = -95.08 deg

Homework R. Martin

# Problem F.10 Mag

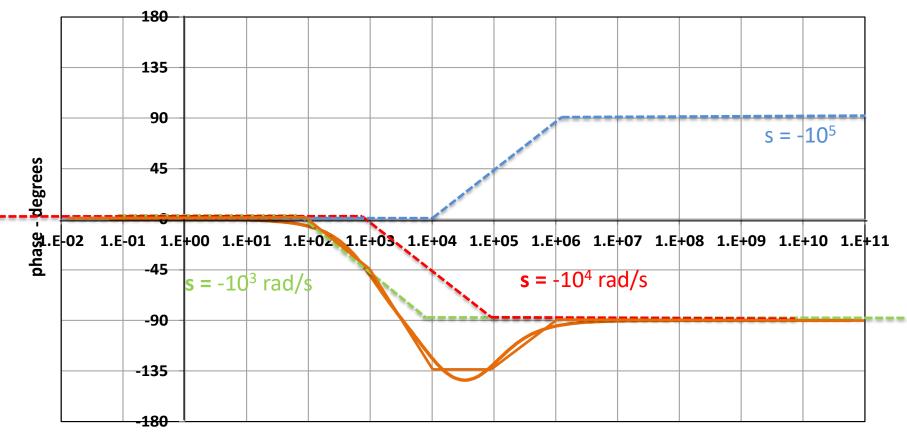
zeros:  $s = -10^5 \text{ rad/s}$ ,  $\infty$  poles:  $s = -10^3 \text{ rad/s}$ ,  $-10^4 \text{ rad/s}$ 



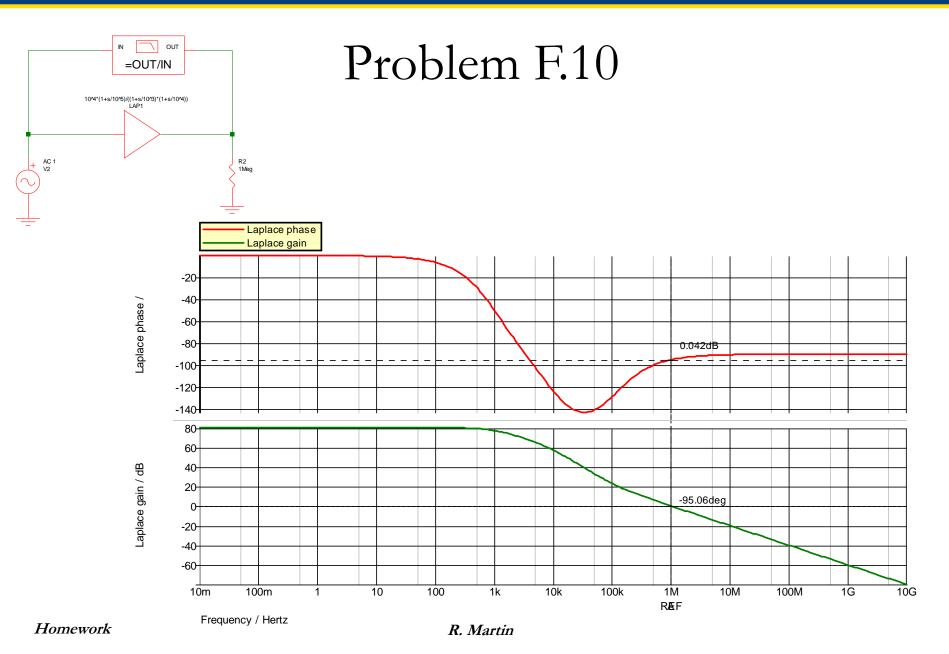
13

### Problem F.10 Phase

zeros:  $s = -10^5 \text{ rad/s}$ ,  $\infty$  poles:  $s = -10^3 \text{ rad/s}$ ,  $-10^4 \text{ rad/s}$ 



determine approximate values for the magnitude and phase at  $\omega = 10^6$  rad/s.



A particular amplifier has a voltage transfer function

$$T(s) = \frac{10s^2}{(1+s/10)(1+s/100)(1+s/10^6)}$$

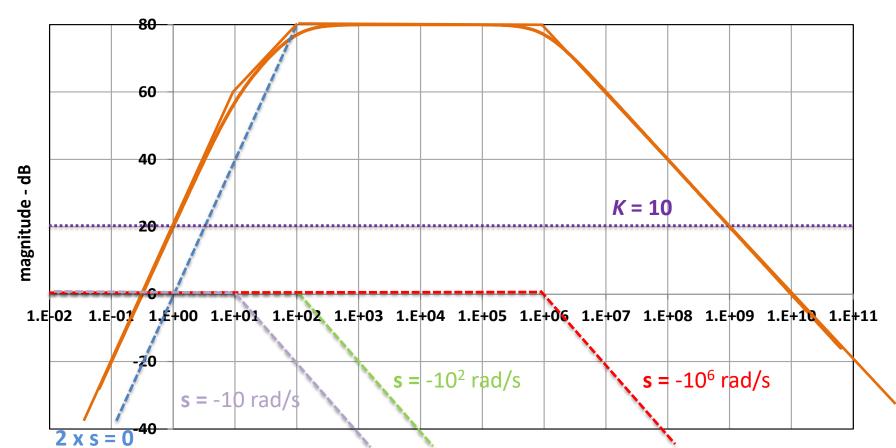
Find the poles and zeros. Sketch the magnitude of the gain in dB versus frequency on a logarithmic scale. Estimate the gain at  $10^0$ ,  $10^3$ ,  $10^5$ , and  $10^7$  rad/s.

zeros:  $s = 0, 0, \infty$ 

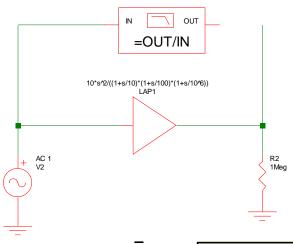
poles:  $s = -10 \text{ rad/s}, -100 \text{ rad/s}, -10^6 \text{ rad/s}$ 

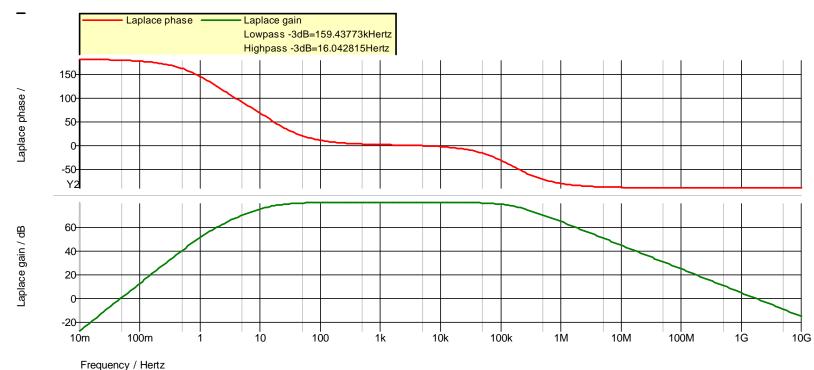
zeros: s = 0,  $s = \infty$ 

poles: s = -10 rad/s,  $-10^2 \text{ rad/s}$ ,  $s = -10^6 \text{ rad/s}$ 



Estimate the gain at 10<sup>0</sup>, 10<sup>3</sup>, 10<sup>5</sup>, and 10<sup>7</sup> rad/s





Homework