Introduction to Control

July 22, 2015

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

Outline

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

An introduction to control

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

Outline

1 Basics

- **Classical control loop**
- Demo: Inverted Pendulum
- Elements

2 Control Goals

- stability
- disturbance rejection
- Reference tracking
- Exercise

3 Closed-loop system

- Sensitivity
- Robustness
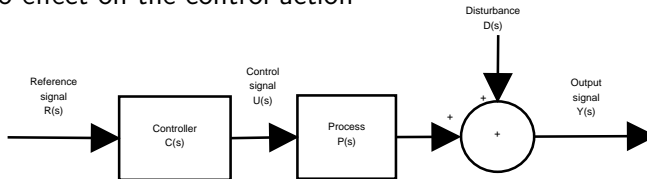- Types of systems and Steady State Error
- Noise- and disturbance rejection

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

What is control?

- The goal is to find an input (control signal U(s)) such that the process $P(s)$ produces the desired output
- Open loop control system: the actual output signal $Y(s)$ has no effect on the control action
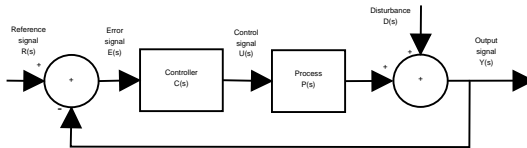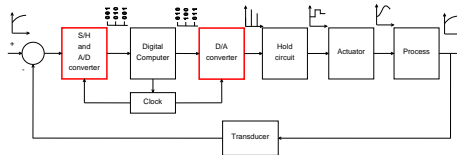

$$Y(s) = P(s)U(s) = P(s)C(s)R(s)$$

Remember the example of pouring a glass of water without looking at the glass

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

A general set-up of a closed loop system

- We will focus on <u>closed loop control systems</u>

- Example: Inverted pendulum
- Digital Control Loop:

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

Outline

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements
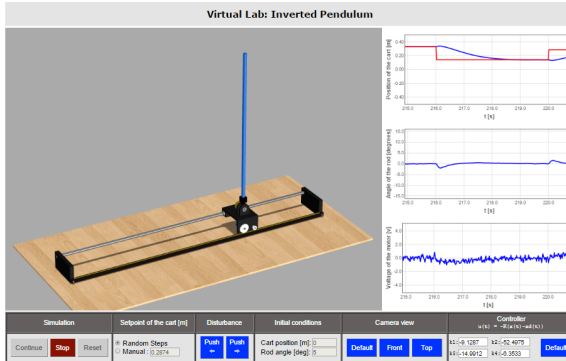
Inverted Pendulum

KU Leuven/ESAT
Oscar Mauricio Agudelo
Bart De Moor

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

Outline

Basics
Control Goals
Closed-loop system

Classical control loop
Demo: Inverted Pendulum
Elements

Concrete Controllers

- On-off controller
  - Thermostate at home
- **PID controllers, Lead and lag compensators (this course)**
  - Cruise-control in your car
- More advanced controllers
  - STATE-space feedback controllers
  - Model Predictive Controller (MPC)
  - Fuzzy Control
  - Neuro-fuzzy Control
  - ...

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Outline

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

What is good control?

- Before we will start to design control systems we will first focus on the question. What is good control?
- It depends on the application
  - Stability
  - Disturbance rejection
  - Reference tracking (speed)
  - Sensitivity to errors on model
  - Etc...

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Outline

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Examples: stability

Space shuttles are like inverted pendulums. How do you make sure they don't flip over.

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Outline

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Examples: Disturbance rejection

- Your body will try to keep the temperature in your body as constant as possible. No matter what the outside temperature is. Two people will have almost the same body temperature.

Flickr.com, tent86, Marathon Des Sables 046

Jack Zalium, Enduring, https://creativecommons.org/licenses/by-nd/2.0/

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Outline

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Examples: Reference tracking

Audi has a system for automatic driving in traffic jams. The audi will follow the car in front of him at an appropriate distance.
https://www.youtube.com/watch?v=Qa_ZSRj0WM0

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Outline

Basics
Control Goals
Closed-loop system

stability
disturbance rejection
Tracking
Exercise

Exercise: name the correct property

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Outline

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Transfer function of a closed-loop system

$$Y(s) - D(s) = P(s)U(s)$$
$$\text{with } U(s) = C(s)E(s)$$
$$Y(s) - D(s) = P(s)C(s)E(s)$$
$$\text{with } E(s) = R(s) - Y(s)$$
$$Y(s) - D(s) = P(s)C(s)(R(s) - Y(s))$$
$$Y(s) - D(s) = P(s)C(s)R(s) - P(s)C(s)Y(s)$$
$$\Rightarrow Y(s) = \frac{P(s)C(s)}{1 + P(s)C(s)}R(s) + \frac{1}{1 + P(s)C(s)}D(s)$$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection
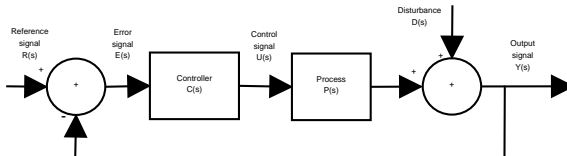
Transfer function from $R(s)$ to $Y(s)$

We define S as the transfer function from $R(s)$ to $Y(s)$

$$S(s) \triangleq \frac{P(s)C(s)}{1 + P(s)C(s)}$$

This transfer function $S(s)$ will help us to evaluate tracking
Almost perfect tracking: the output $Y(s)$ will follow $R(s)$ very
closely $\Rightarrow S(s) \approx 1$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Transfer function from $D(s)$ to $Y(s)$

We define $T(s)$ as the transfer function from $D(s)$ to $Y(s)$.

$$T(s) \triangleq \frac{1}{1 + P(s)C(s)}$$

If the disturbance rejection is very good the disturbances will have almost no effect on the output $\Rightarrow T \approx 0$

$$\begin{cases} |S(j\omega)| \cong 1 \\ |T(j\omega)| \cong 0 \end{cases}$$

$\Rightarrow |P(j\omega)C(j\omega)|$ **(open loop gain)** is very large.
! a large open loop amplification might lead to an unstable system

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example: Which controller do you prefer

The transfer function from $R(s)$ to $Y(s)$ for two different
controllers for high precision surgery

$$\frac{0.5}{s + 0.5}$$

$$\frac{10}{1.09s^2 + s + 10}$$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection
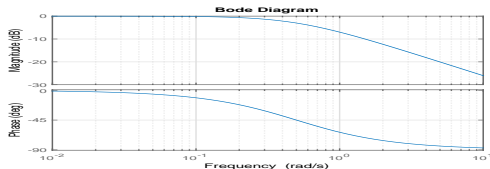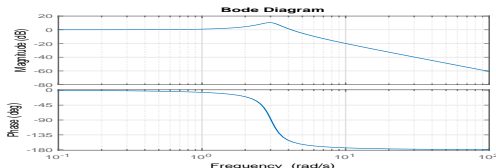
Example: step response of both controllers

The ouput $Y(s)$ when $R(s)$ is a step function.

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection
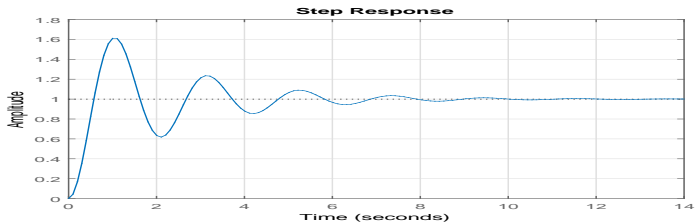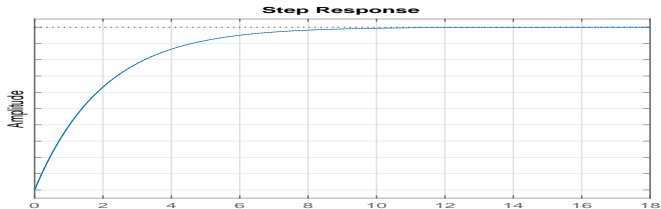
Quality of reference tracking

- Look at the step response from $R(s)$ to $Y(s)$. All the known criteria can help you determine the quality.
    - Rise-time
    - Settling time
    - Steady state error
    - Overshoot
    - ...
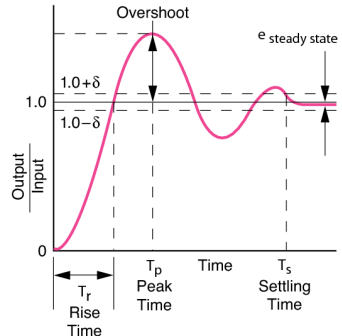
http://www.newport.com/
Control-Theory-Terminology/178319/
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Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Model errors

- In practice, the transfer function $P(S)$ is often unknown. It is important to know how model errors affect the result. Sensitivity and robustness are key concepts to evaluate these effects.
- Sensitivity
  - Quantifies the effect of small model errors on the output.
- Robustness
  - Refers to bigger changes of the model. A controller is robust if it works properly over a given set of parameters.

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Outline

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Sensitivity

- Sensitivity is a measure of the effect of a (small) disturbance on the model (e.g. variations in the process parameters)

$$Y(s) + \Delta Y(s) = \frac{(P(s) + \Delta P(s))C(s)}{1 + (P(s) + \Delta P(s))C(s)}R(s) + \frac{1}{1 + (P(s) + \Delta P(s))C(s)}D(s)$$

- Look at the effect on the system without disturbances ($D(s)=0$)

$$\begin{aligned}
\Delta Y &= \frac{(P + \Delta P)C}{1 + (P + \Delta P)C}R - \frac{PC}{1 + PC}R \\
&= \frac{(P + \Delta P)C(1 + PC) - PC - PC(P + \Delta P)C}{(1 + (P + \Delta P)C)(1 + PC)}R \\
&= \frac{\Delta PC}{(1 + (P + \Delta P)C)(1 + PC)}R \\
&= Y\frac{\Delta P}{P}\frac{1}{1 + (P + \Delta P)C}
\end{aligned}$$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Sensitivity

- Now take the relative change due to this disturbance of the model and take the limit for $\partial x \to 0$; this gives the following (measure of the) sensitivity:

$$S_P^Y(s) = \frac{\frac{\partial Y}{Y}(s)}{\frac{\partial P}{P}(s)} = \frac{1}{1 + P(s)C(s)}$$

- Again, a very large $|P(s)C(s)|$ looks like a good choice, but again there is a risk for instability!

- Note that the sensitivity can be determined for any parameter

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Outline

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example 1: robustness

- Suppose we want to stabilize the system $P(s) = \frac{1}{(s-a)}$. We only know that $a \in [0.20; 0.80]$. We propose to use a proportional controller $C(s) = K$. The closed loop transfer function becomes

$$S(s) = \frac{P(s)C(s)}{1 + P(s)C(s)} = \frac{\frac{1}{s-a}K}{1 + \frac{1}{s-a}K} = \frac{K}{s - a + K}$$

The controller stabilizes the system if $a + K > 0$. If we choose K very large the system will be robust against changes in a.

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example 2: robustness

Suppose we want to stabilize the system $P(s) = \frac{1}{(s-a)}$ with $a \in [0.20; 0.80]$. If we choose the control law $C(s) = \frac{K}{s}$ this results in the same transfer function.

$$S(s) = \frac{P(s)C(s)}{1 + P(s)C(s)} = \frac{\frac{s}{s-a}\frac{K}{s}}{1 + \frac{s}{s-a}\frac{K}{s}} = \frac{K}{s - a + K}$$

Again we choose $K > a$ to ensure stability. But how does the controller perform on the slightly perturbed system $\tilde{P}(s) = \frac{(s+\epsilon)}{(s-a)}$? The transfer function from $R(s)$ to $Y(s)$ becomes

$$S(s) = \frac{\frac{s+\epsilon}{s-a}\frac{K}{S}}{1 + \frac{s+\epsilon}{s-a}\frac{K}{s}} = \frac{(s+\epsilon)K}{s^2 + (K-a)s + \epsilon K}$$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example 2: robustness

$$\tilde{S}(s) = \frac{(s + \epsilon)K}{s^2 + (K - a)s + \epsilon K}$$

The figure on the right plots

$$s^2 + (K - a)s$$
$$s^2 + (K - a)s + \epsilon K$$

For ϵ negative the system becomes unstable. The control system is not robust and useless in practical situations. You shouldnt use control laws which rely on a pole-zero cancellation.

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection
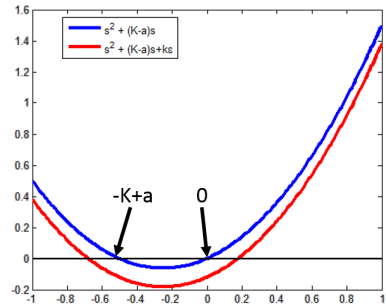
Example 3: Robustness of steady-state error

- The steady state error is defined as follows:

$$\lim_{t\to\infty} e(t) = \lim_{t\to\infty} (r(t) - y(t))$$
$$= \lim_{s\to 0} s(R(s) - Y(s)) \quad \text{(final value theorem)}$$

- A very small steady state error (preferably zero) indicates that the controller tracks the reference very well
- For the open loop system with a step reference

$$e_{ol}(\infty) = \lim_{s\to 0} s(1 - C(s)P(s))R(s) = 1 - C(0)P(0)$$

  - So the open loop controller can be free of a steady state error (for a step reference $\epsilon(t)$) by calibrating the controller such that $C(0)P(0) = 1$
  - $\Rightarrow$ a precise calibration of the DC gain

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example 3: Steady state error

- Closed loop system:

$$e_{cl} = \lim_{s \to 0} s \left(1 - \frac{C(s)P(s)}{1 + C(s)P(s)} \right) R(s)$$

$$= \lim_{s \to 0} \frac{s}{1 + C(s)P(s)} R(s) = \frac{1}{1 + C(0)P(0)}$$

- The steady state error is small if $C(0)P(0)$ is very large
- Again, calibrating the DC gain
- The difference is that here we only need a large value for it, which is far less demanding than having to make it equal to 1

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

An open loop controller is not robust

- We can now show how this results in a great advantage of the closed loop strategy over the open loop strategy:
    - If P changes slightly (for instance due to a factor that has not been taken up into the model) to $P + \Delta P$
    - Then making $e_{ol}(\infty)$ small would require to calibrate anew
    - Whereas $e_{cl}(\infty)$ would remain small, as long as $(P(0) + \Delta P(0))C(0)$ remains large
- Hence an open loop controller can't control the output **robustly** against changes in P

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Outline

1. **Basics**
   - Classical control loop
   - Demo: Inverted Pendulum
   - Elements

2. **Control Goals**
   - stability
   - disturbance rejection
   - Reference tracking
   - Exercise

3. **Closed-loop system**
   - Sensitivity
   - Robustness
   - Types of systems and Steady State Error
   - Noise- and disturbance rejection

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Type of a system

- For a step reference $r(t) = A\epsilon(t) \Rightarrow R(s) = \frac{A}{s}$

$$e_{cl}(\infty) = \frac{A}{1 + C(0)P(0)}$$

So $e_{cl}(\infty) = 0$ if C(s)P(s) has a least one pole in zero
$(C(0)P(0) \to \infty)$

- For a ramp reference function $r(t) = At \Rightarrow R(s) = \frac{A}{s^2}$, we have

$$e_{cl}(\infty) = \lim_{s \to 0} \frac{s}{1 + C(s)P(s)} \frac{A}{s^2}$$

So $e_{cl}(\infty) = 0$ if $C(s)P(s)$ has at least two poles in zero.
$\left(\lim_{s \to 0} sC(s)P(s) = \infty \right)$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Type of a system

- The type of a system is determined by the number of poles $P(s)C(s)$ has in zero, and hence it is linked to what type of references it can track perfectly

- Write $P(s)C(s)$ as $\frac{K \prod_{k=1}^{m}(s-z_k)}{s^l \prod_{k=1}^{n-l}(s-p_k)}$ with l the amount of poles in zero

- We say this system is of type l
  - It will be able to track references of shape $At^0\epsilon(t)$ up to $At^{(l-1)}\epsilon(t)$ perfectly (for $t \to \infty$, $\lim_{t \to \infty} e(t) = 0$)
  - A reference of shape $At^l\epsilon(t)$ will be missed by a finite factor (for $t \to \infty$, $\lim_{t \to \infty} e(t) = C$)
  - A reference of shape $At^{l+\cdots}\epsilon(t)$ will be missed by infinity (for $t \to \infty$, $\lim_{t \to \infty} e(t) = \infty$)

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Example of a type 0 system

- Step reference $r(t) = A\epsilon(t) \Rightarrow R(s) = \frac{A}{s}$

$$e_{cl}(\infty) = \lim_{s \to 0} \frac{s}{1 + C(s)P(s)} \frac{A}{s}$$

$$= \lim_{s \to 0} \frac{A}{1 + \frac{K \prod_{k=1}^{m}(-z_k)}{\prod_{k=1}^{n-l}(-p_k)}} = \frac{A}{1 + K_p}$$

  with K_p the position constant

- Ramp reference $r(t) = At\epsilon(t) \Rightarrow R(s) = \frac{A}{s^2}$

$$e_{cl}(\infty) = \lim_{s \to 0} \frac{s}{1 + C(s)P(s)} \frac{A}{s^2}$$

$$= \lim_{s \to 0} \frac{1}{1 + \frac{K \prod_{k=1}^{m}(-z_k)}{\prod_{k=1}^{n-l}(-p_k)}} = \infty$$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Steady state errors - type of a system

- With:

  $$K_p = \lim_{s \to 0} P(s)C(s) \qquad K_p = \text{Static position constant}$$
  $$K_v = \lim_{s \to 0} sP(s)C(s) \qquad K_v = \text{Static velocity constant}$$
  $$K_a = \lim_{s \to 0} s^2 P(s)C(s) \qquad K_a = \text{Static acceleration constant}$$

- And the respective steady state errors for different system types are:

Type l	Step A$\epsilon(t)$	Ramp At$\epsilon(t)$	Parabola $\frac{At^2 \epsilon(t)}{2}$
0	$\frac{A}{1+K_p}$	∞	∞
1	0	$\frac{A}{K_v}$	∞
2	0	0	$\frac{A}{K_a}$

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Outline

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Noise rejection and disturbance rejection

- Both refer to rejection of the input D but are not the same
- Measurement noise is often modelled by white noise. (High frequency signal)
- Disturbances are actual changes to the state of the system

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection
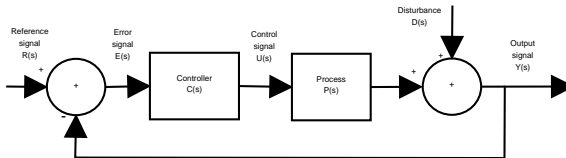
Noise versus Disturbances

- Cruise control:
  Measurement errors on the speed are noise
  A change in slope is a disturbance

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
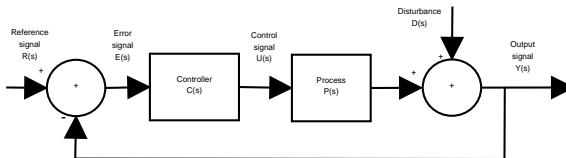Types of systems and SS-error
Noise- and disturbance rejection

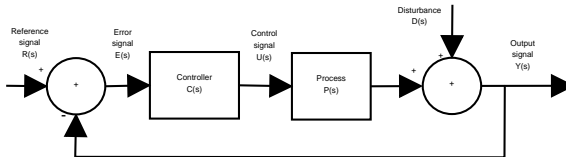## Noise versus Disturbances

- Remember from the start of the lecture

$$Y(s) = \frac{P(s)C(s)}{1 + P(s)C(s)}R(s) + \frac{1}{1 + P(s)C(s)}D(s)$$

- Choosing $T(s) = \frac{1}{1+P(s)C(s)}$ sufficiently small results in the rejection of D. This can be achieved by choosing $C(s)$ sufficiently high

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Disturbance vs noise

- But what happens to measurement noise
  - The amplified measurement noise will be applied to the input of the plant.
  - In reality measurement noise requires no control actions. Good noise rejection means that the controller ignores this noise.

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Disturbance versus noise rejection

- Good disturbance rejection requires fast action to bring the state back to the desired state
- Good noise rejection requires that measurement noise will be ignored
- Note that a controller can not see the difference between measurement noise and disturbances. Slow controllers will be less sensitive to measurement noise. Fast system will have better disturbance rejection

Basics
Control Goals
Closed-loop system

Sensitivity
Robustness
Types of systems and SS-error
Noise- and disturbance rejection

Classical trade-offs in control theory

Slow Controllers:
Not sensitive to noise
Small control inputs
Robust controllers:
Model errors will not
affect the behaviour of
the system strongly

Fast controllers:
Good disturbance rejection
Fast tracking

Aggresive controllers:
Exchanges robustness for
better performance