#### SDM366 Optimal Control and Estimation

# Lecture Note 2 State Space Model and Stability

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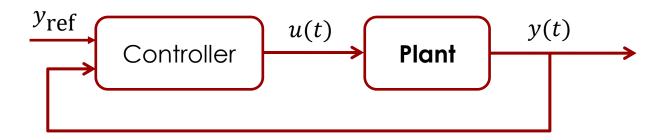
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#### Outline

- State space model: definition and examples
- From continuous-time to discrete time model
- From nonlinear to linear model
- System solution and stability

## State-space model based feedback control system:

Goal: determine control input to achieve desired output



- Controller design is based on plant model
  - Model is different from the actual plant
  - "all models are wrong, but some are useful"
- Modeling approach:
  - First principle
  - Data driven (System ID)

## Static vs. Dynamic Systems

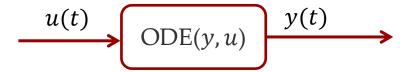
Static system

$$y = \phi(u)$$

- u(t) completely and immediately determines y(t)
- Desired output  $y_{ref}$  can be perfectly tracked (in absence of disturbance) by open-loop plant inversion

## Static vs. Dynamic Systems

**Dynamic system:** differential or difference equation



- u(t) does not fully determines y(t)
- At time  $t_0$ , the output  $y(t_0)$  does not fully captures the system "behavior"

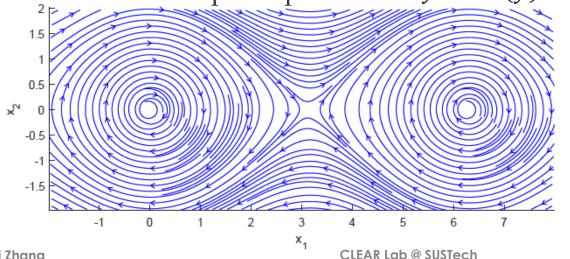
- "State": info needed for future evolution, it separates future from past
- State  $x(t_0)$  at time  $t_0$  and input u(t) over time  $[t_o, t_f]$ , completely determines the system behaviors

## General continuous-time state space model

$$\dot{x} = f(x, u)$$
$$y = h(x, u)$$

- $x \in \mathbb{R}^n$  state vector,  $u \in \mathbb{R}^m$  control input,  $y \in \mathbb{R}^p$  output,
- $f: \mathbb{R}^n \times \mathbb{R}^m \to \mathbb{R}^n$ : called **vector field**
- $h: \mathbb{R}^n \times \mathbb{R}^m \to \mathbb{R}^p$ : output function
- Called autonomous system if there is no control f(x, u) = f(x)
- For autonomous sys,  $\hat{x} \in \mathbb{R}^n$  is called **equilibrium** if  $f(\hat{x}) = 0$

Vector field example of pendulum:  $\ddot{y} + \sin(y) = 0$ 



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## General discrete-time state space model

$$x(k+1) = f(x(k), u(k))$$
$$y(k) = h(x(k), u(k))$$

- $x \in \mathbb{R}^n$  state vector,  $u \in \mathbb{R}^m$  control input,  $y \in \mathbb{R}^p$  output
- $f: \mathbb{R}^n \times \mathbb{R}^m \to \mathbb{R}^n$ : state update equation
- $h: \mathbb{R}^n \times \mathbb{R}^m \to \mathbb{R}^p$ : output function
- Called autonomous system if there is no control f(x, u) = f(x)
- For autonomous sys,  $\hat{x} \in \mathbb{R}^n$  is called **equilibrium** if  $\hat{x} = f(\hat{x})$

- Discrete-time system:
  - Some discrete-time system is obtained from continuous time model by sampling
  - Some systems naturally evolve in discrete time.

## • Linear Systems: system is called linear if:

Continuous time 
$$\dot{x} = f(x, u) = Ax + Bu$$
,  $y = h(x, u) = Cx + Du$ ,

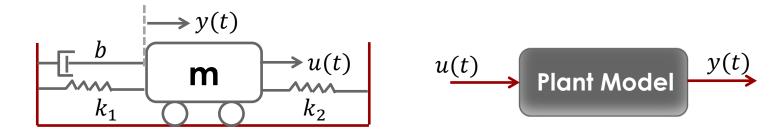
Discrete time 
$$x(k+1) = f(x(k), u(k)) = Ax(k) + Bu(k),$$
 
$$y(k) = h(x(k), u(k)) = Cx(k) + Du(k),$$

for some matrices A, B, C, D

## State-space modeling:

- Find the functions  $f(\cdot,\cdot)$ ,  $h(\cdot,\cdot)$
- Or find *A*, *B*, *C*, *D* matrices if the system is linear

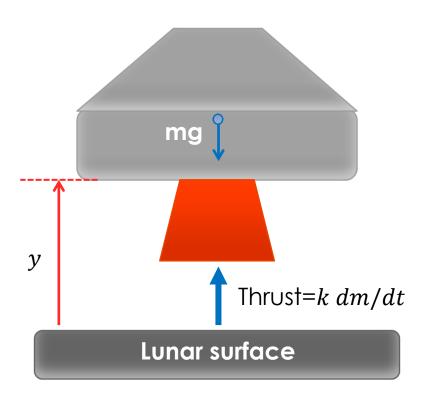
**Example 1**: Consider spring-damper cart system with zero initial conditions (initially at y = 0 and not moving). No friction



Differential equation model

State space model of Example 1 (infinitely many)

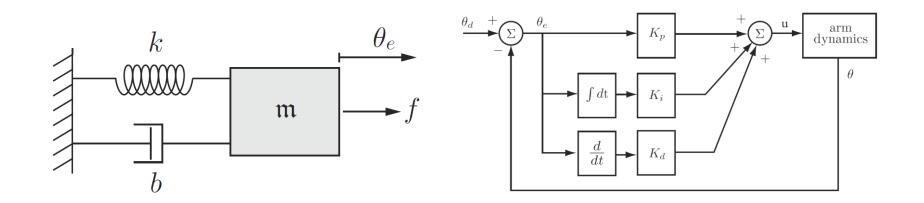
**Example 2**: soft landing of a lunar module,  $u = \frac{dm}{dt}$ 



- **Example 3**: Sensor Network
  - Each iteration, exchange measurements with neighbors
  - The updated value is the average of its own value with the neighbors



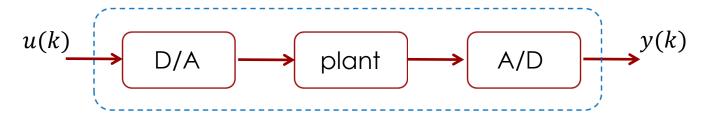
## ■ Example 4: PID for spring-damper system



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#### From continuous time to discrete time model



- Approximate differential equation with difference equation
  - Euler forward rule:

#### From continuous-time to discrete-time model

• General nonlinear case:

$$\dot{x} = f(x, u)$$
$$y = h(x, u)$$

### From continuous-time to discrete-time model

• Linear case:

$$\dot{x} = A_c x + B_c u,$$
  

$$y = C_c x + D_c u,$$

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#### From nonlinear to linear

- Given model:  $x(k+1) = f(x(k), u(k)), \ y(k) = h(x(k), u(k))$  and operating point:  $(\hat{x}, \hat{u})$
- Goal: find a linearized model around  $(\hat{x}, \hat{u})$

■ Jacobian matrix of multivariable function  $f: \mathbb{R}^n \to \mathbb{R}^m$ 

• Example of Jacobian matrix:  $f(z) = \begin{vmatrix} 2z_1 + e^{z_2} \\ \log(z_3) + \frac{1}{z_2} \end{vmatrix}$ ,  $\hat{z} = \begin{bmatrix} 1 \\ 2 \\ 3 \end{bmatrix}$ 

- Taylor expansion of multivariate function
  - General expression:  $f(z) = f(\hat{z}) + \left(\frac{\partial f}{\partial z}(z)\Big|_{z=\hat{z}}\right) \Delta z + \text{H.O.T}$

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• Linearization around  $(\hat{x}, \hat{u})$  using Taylor expansion:

$$f(x,u) \approx f(\hat{x},\hat{u}) + \left(\frac{\partial f(x,u)}{\partial x}\Big|_{x=\hat{x},u=\hat{u}}\right) \cdot (x-\hat{x}) + \left(\frac{\partial f(x,u)}{\partial u}\Big|_{x=\hat{x},u=\hat{u}}\right) \cdot (u-\hat{u})$$

$$= \hat{A} \cdot \Delta x + \hat{B} \cdot \Delta u + f(\hat{x},\hat{u})$$

$$h(x,u) \approx h(\hat{x},\hat{u}) + \left(\frac{\partial h(x,u)}{\partial x}\Big|_{x=\hat{x},u=\hat{u}}\right) \cdot (x-\hat{x}) + \left(\frac{\partial h(x,u)}{\partial u}\Big|_{x=\hat{x},u=\hat{u}}\right) \cdot (u-\hat{u})$$

$$\hat{C}$$

$$\Delta y := y - h(\hat{x}, \hat{u}) \approx \hat{C} \cdot \Delta x + \hat{D} \cdot \Delta u$$

$$y(k) = \cos(x_2(k)) + 2x_1(k) \qquad \hat{x} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}, \, \hat{u} = \begin{bmatrix} 0 \\ \frac{\pi}{2} \end{bmatrix}$$

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General linear state space model:

$$x(k+1) = A(k)x(k) + B(k)u(k),$$
  
$$y(k) = C(k)x(k) + D(k)u(k)$$

- If system matrices (A(k), B(k), C(k), D(k)) change over time k, then system is called **Linear Time Varying (LTV)** system
- If system matrices are constant w.r.t. to time, then the system is called a **Linear Time Invariant (LTI)** System

$$x(k+1) = Ax(k) + Bu(k),$$
  
$$y(k) = Cx(k) + Du(k)$$

Derivation of Solution to LTI state space system:

$$x(k+1) = Ax(k) + Bu(k),$$
  
$$y(k) = Cx(k) + Du(k)$$

• given initial state  $x(0) = \hat{x}$ , and control sequence  $u(0), ..., u(k), k \ge 0$ , we have  $x(k) = A^k \hat{x} + \sum_{j=0}^{k-1} A^{k-j-1} Bu(j)$ 

- A large portion of control applications can be transformed into a regulation problem
  - Regulation problem: keep certain function of the state x(k) or output y(k) close to a known constant reference value under disturbances and model uncertainties

- Keep inverted pendulum at upright position ( $\theta = 0$ )
- Maintain a desired attitude of spacecraft or aircraft

For example:

- Air conditioner regulate temperate close to setpoint (e.g. 75F)
- Cruise control maintain a constant speed despite uncertain road conditions
- Converter maintains a desired voltage level for different loads
- If reference  $y_{ref}(t)$  is changing, this is no longer a regulation problem (becomes a **tracking problem**)

■ Internal Stability (with  $u(k) \equiv 0$ , i.e. concerned with zero-input state response)

$$x(k+1) = Ax(k) + Bu(k),$$
  
$$y(k) = Cx(k) + Du(k)$$

- Asymptotic stable:  $||x(k)|| \to 0$ , as  $k \to \infty$ , for all initial state  $\hat{x}$
- Marginal stable:  $||x(k)|| \le M$ , for all k = 1, 2, ...
- Recall state space solution for linear systems:

$$x(k) = A^{k}\widehat{x} + \sum_{j=0}^{k-1} A^{k-j-1}Bu(j)$$

■ Therefore, or linear system, the key for stability analysis is to understand how  $A^k$  behave as  $k \to \infty$ 

■ Case 1: diagonal matrix: e.g.  $A = \begin{bmatrix} \lambda_1 & 0 & 0 \\ 0 & \lambda_2 & 0 \\ 0 & 0 & \lambda_3 \end{bmatrix}$ 

■ Case 2: diagonalizable matrix, i.e.  $\exists T$  such that  $A = TDT^{-1}$ 

- Case 3: Unfortunately, not all square matrices are diagonalizable
  - e.g.:  $\begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix}$  is **not diagonalizable**

■ Theorem (Internal stability): LTI (A, B) is asymptotically stable if all eigs of A satisfies  $|\lambda_i| < 1$ 

More discussions