

# Digital Signal Processing

## Theory & Laboratory

Chap 1. Introduction 1

## Chapter 1

### Introduction

Chap 1. Introduction 2

# SCOPE OF DIGITAL SIGNAL PROCESSING

## BACKGROUND

### ▲ Applied area of DSP

engineering, science, medicine, economics, social science, ...

### ▲ Things that we can do with DSP?

- detect trends in a signal
- extract a wanted signal from noise-contaminated signal
- assess frequency components in a signal

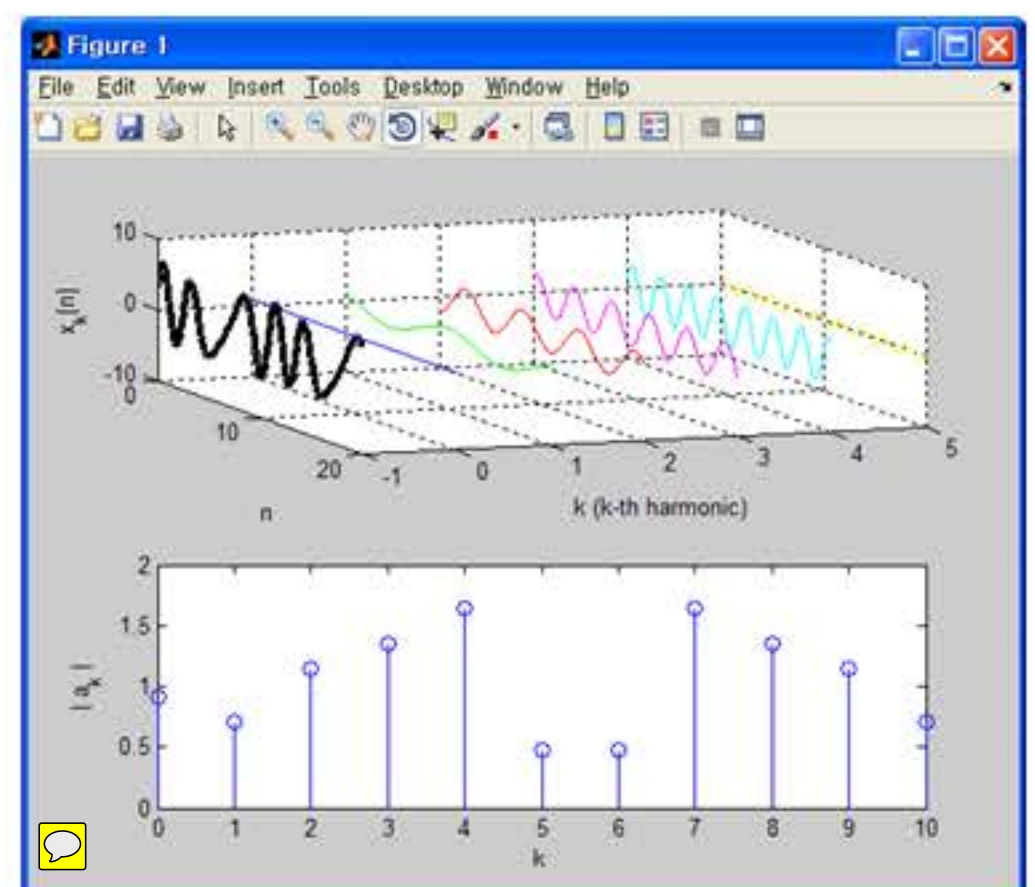
# SCOPE OF DIGITAL SIGNAL PROCESSING

### ▲ What is frequency component in a signal?

- discovery by Fourier
- every signal can be represented by combination of sinusoids

### 📁 Run m\_ex\_spectrum0.m

- shows relationship b/w time-domain and freq-domain representations of a signal
- k-th harmonic means freq component that completes k cycles in 1 period of original signal

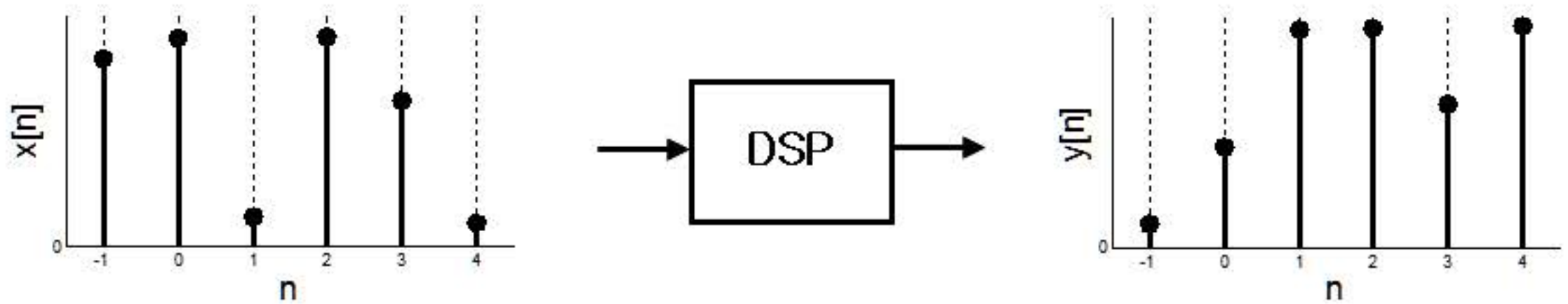




# SCOPE OF DIGITAL SIGNAL PROCESSING



## ▲ Digital signal processors



## ▲ What are they?

- difference equation

# SCOPE OF DIGITAL SIGNAL PROCESSING

## ▲ DSP scheme

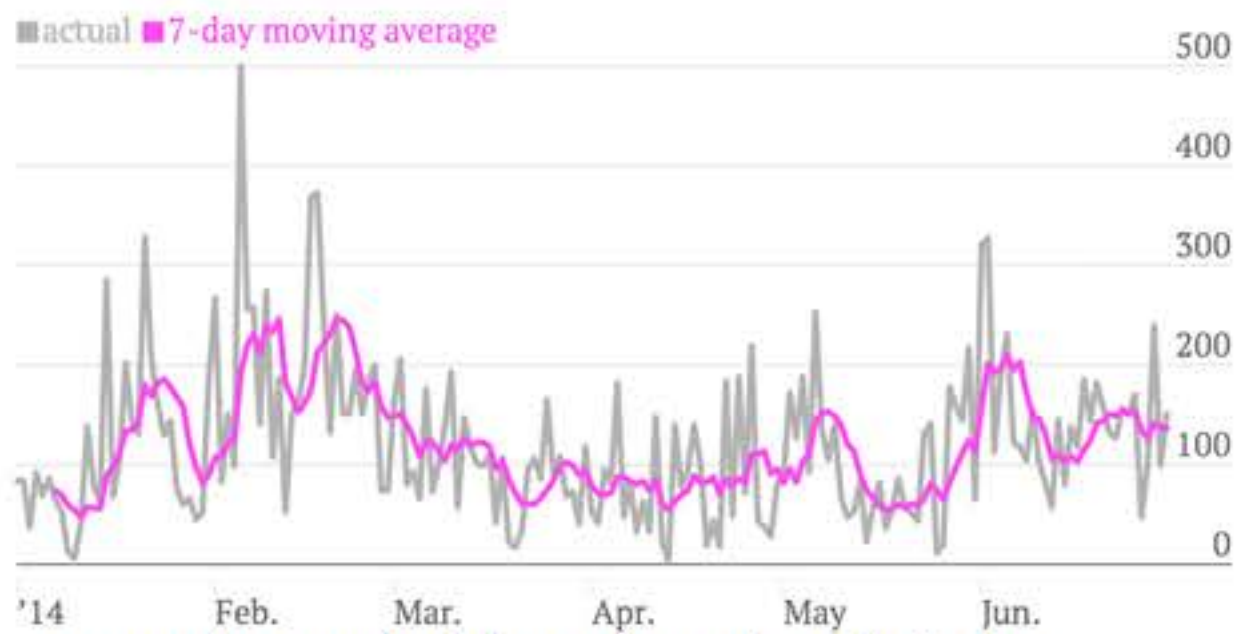


- ADC(analog-to-digital converter): sample and quantize
- DAC(digital-to-analog converter): zero-order hold
- Analog filter before ADC: limit freq. range of input signal before sampling
- Analog filter after DAC: remove sharp transitions from DAC output

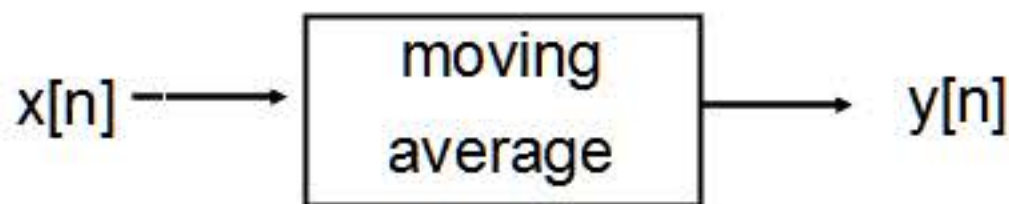
# SCOPE OF DIGITAL SIGNAL PROCESSING

## SOME PRACTICAL APPLICATIONS

### ▲ Moving Average (MA)



- reveals underlying trends of data
- MA as DSP



- low pass filter

# SCOPE OF DIGITAL SIGNAL PROCESSING

### ▲ Equations for 200-Day MA

- nonrecursive form 

$$\begin{aligned} y[n] &= \frac{1}{200} \{x[n] + x[n-1] + x[n-2] + \dots x[n-199]\} \\ &= 0.005 \sum_{k=0}^{199} x[n-k] \end{aligned}$$

- recursive form 

$$\begin{aligned} y[n+1] &= 0.005 \{x[n+1] + x[n] + x[n-1] + \dots x[n-198]\} \\ &= y[n] + 0.005 \{x[n+1] - x[n-199]\} \end{aligned}$$

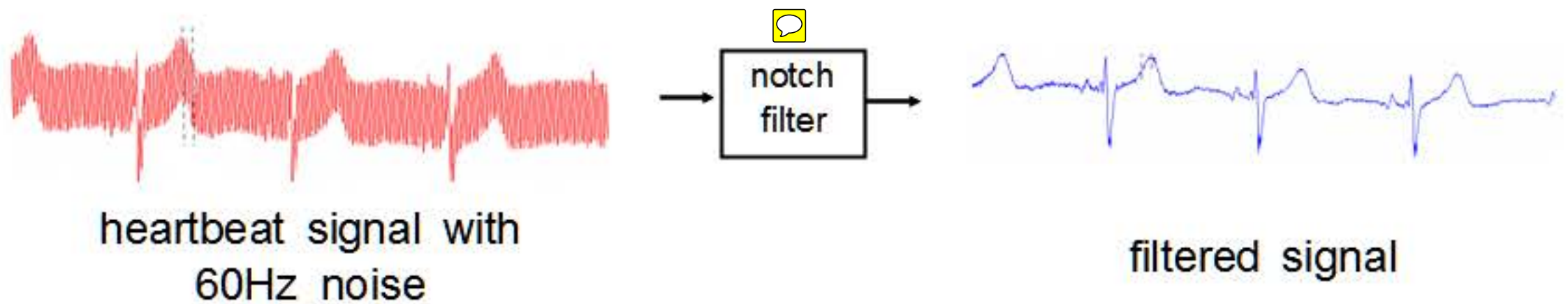
or

$$y[n] = y[n-1] + 0.005 \{x[n] - x[n-200]\}$$



# SCOPE OF DIGITAL SIGNAL PROCESSING

## ▲ Medical application

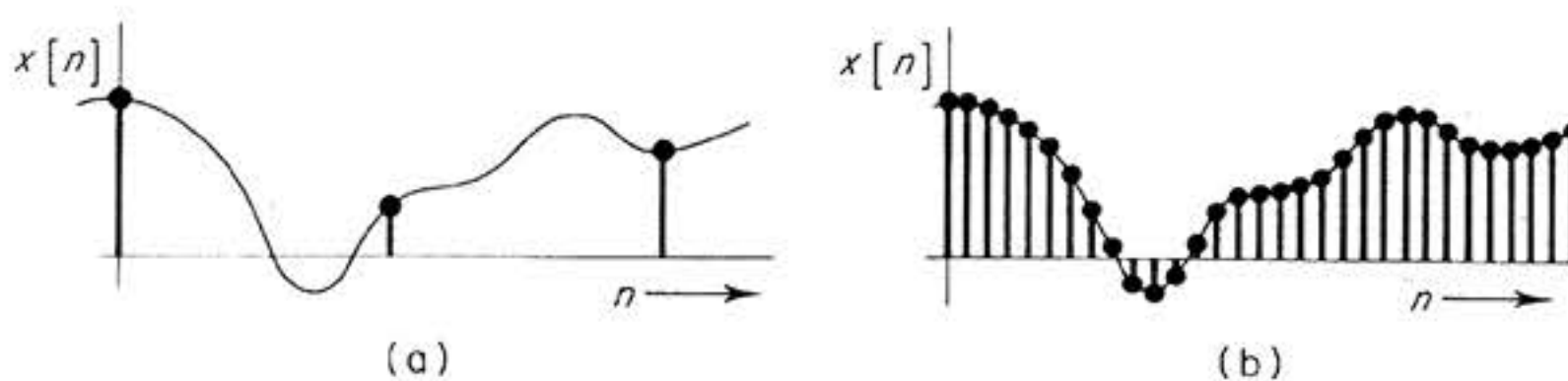


- filter equation (60Hz notch for 1.2KHz sampling)

$$y[n] = 1.8523y[n-1] - 0.94883y[n-2] + x[n] - 1.9021x[n-1] + x[n-2]$$

## SAMPLING & AD CONVERSION

### ▲ How often should we sample analog signal?



(a) too slow to pick out rapid fluctuations

(b) (maybe) too fast giving too many data

# SAMPLING & AD CONVERSION

## ▲ Sampling Theorem by Shannon 🗨

$$f \geq 2f_{\max} \quad \text{or} \quad T \leq \frac{1}{2f_{\max}}$$

$f$  [Hz],  $T$  [s]: sampling frequency and interval

$f_{\max}$  [Hz]: maximum frequency component in analog signal

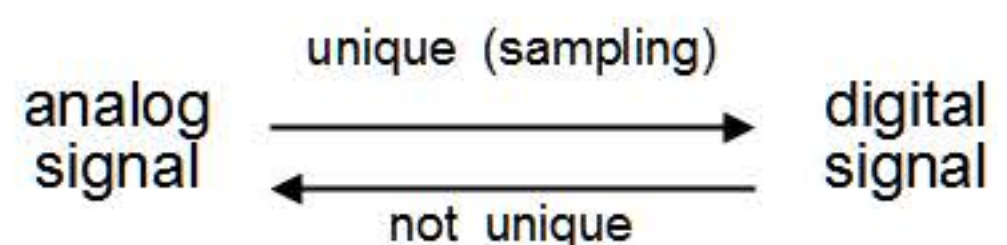
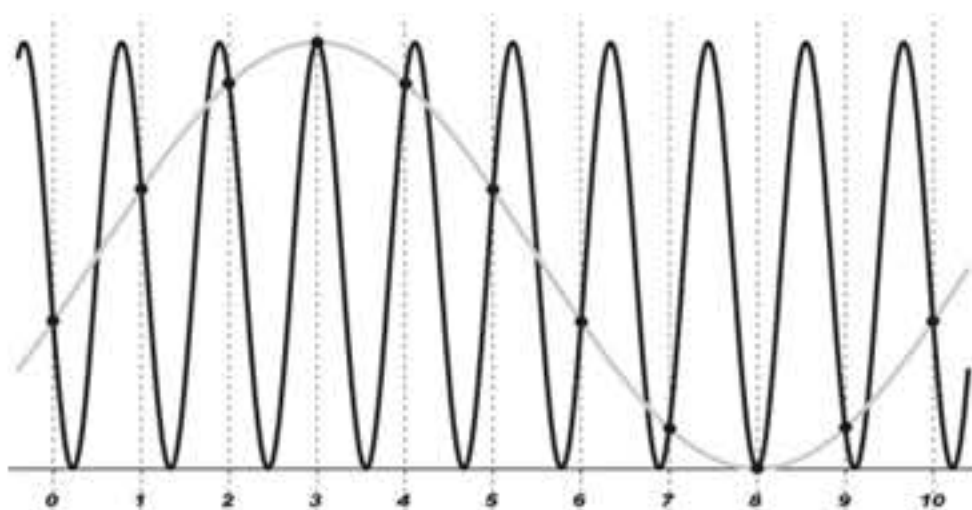
- Otherwise, aliasing occurs.

## ▲ Terminology

Nyquist frequency = half the sampling frequency ( $= \frac{f}{2}$ )

# SAMPLING & AD CONVERSION

## ▲ Ambiguity in sampled sinusoid



## 📄 Run `m_sampledsinusoid.m`

- (1)  $f = 10\text{Hz}$ ,  $f_1 = 3\text{Hz}$ ,  $f_2 = 13\text{Hz}$
- (2)  $f = 10\text{Hz}$ ,  $f_1 = 3\text{Hz}$ ,  $f_2 = 7\text{Hz}$



# SAMPLING & AD CONVERSION

## ▲ Aliasing

(Time Domain)

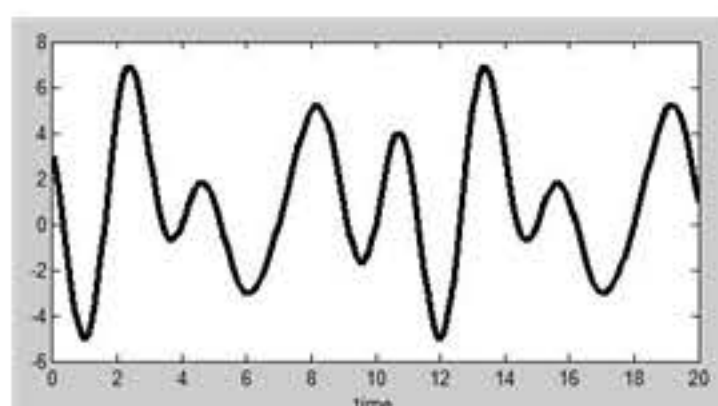
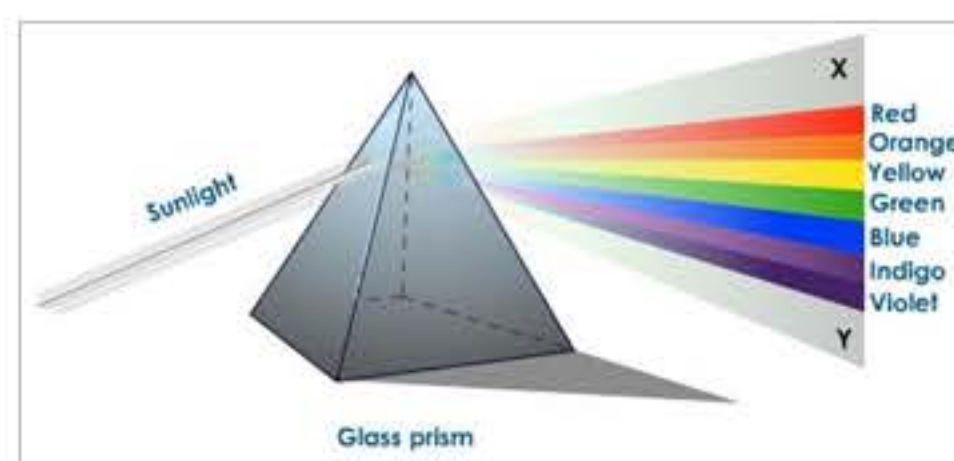
- With slow sampling, sampled signal (digital) looks slower than original signal (analog).

(Frequency Domain)

- Sampling causes original spectrum of analog signal to repeat around multiples of sampling frequency

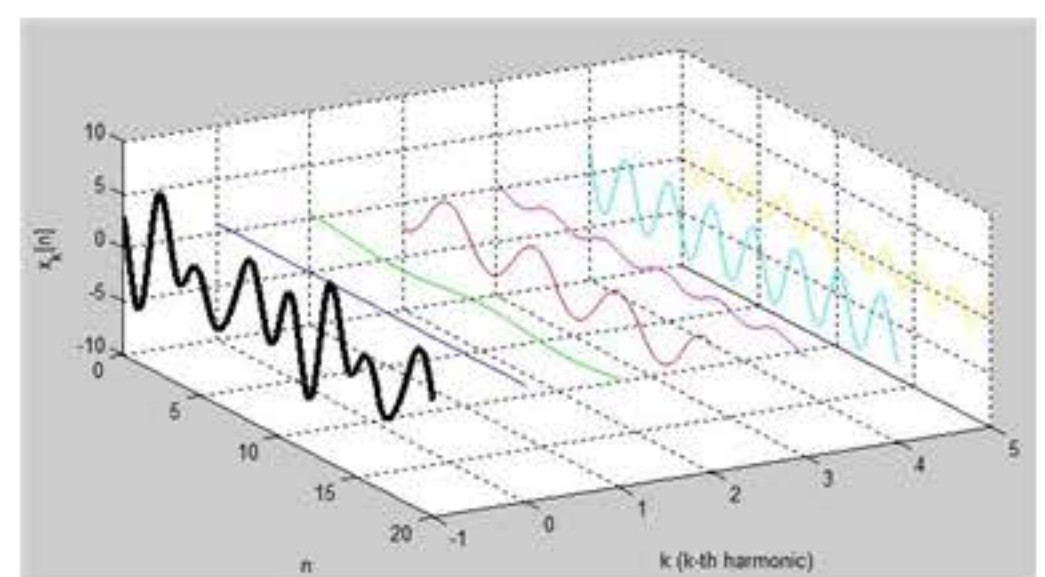
# SAMPLING & AD CONVERSION

## ▲ Spectrum



time history

Fourier  
transform

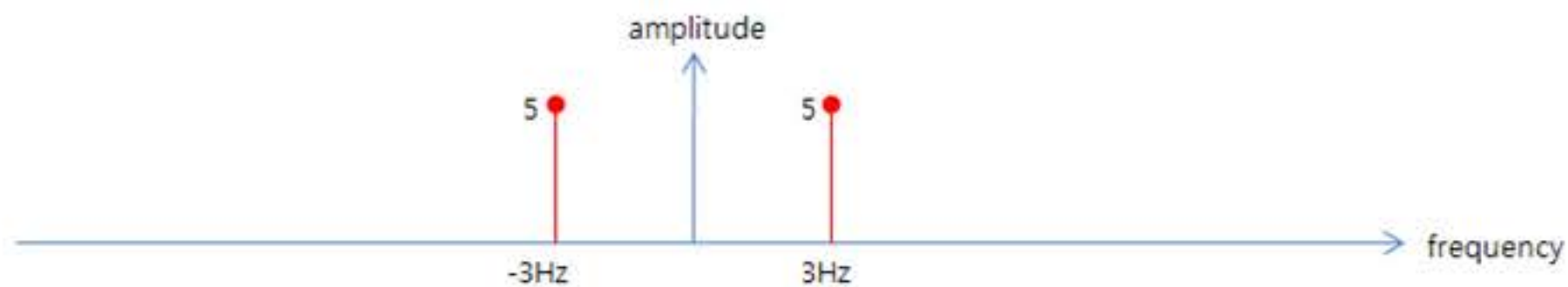


frequency spectrum

# SAMPLING & AD CONVERSION

## ▲ Effect of sampling in frequency domain

- consider pure sinusoid:  $10\cos(2\pi f_a t)$  where  $f_a = 3\text{Hz}$
- frequency spectrum



Q: Negative frequencies? Half amplitude?

A: Come from expressing sinusoid as sum of two exponentials

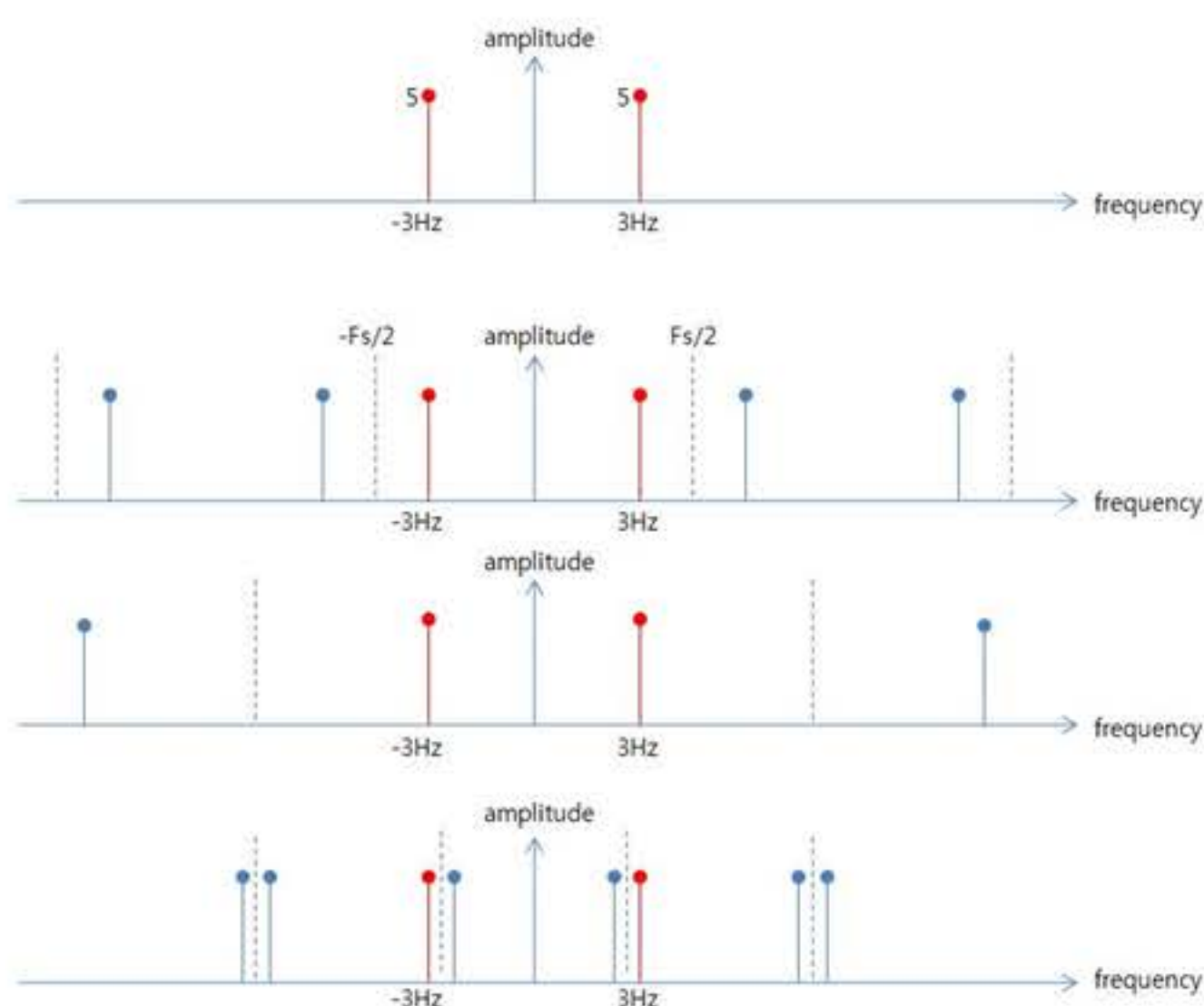
$$A\cos(\omega t) = \frac{A}{2}\exp(j\omega t) + \frac{A}{2}\exp(j(-\omega)t)$$

$\Rightarrow \{A, \omega\}$  can be represented by  $\left\{\frac{A}{2}, -\omega\right\}, \left\{\frac{A}{2}, \omega\right\}$

$\Rightarrow$  amplitude spectrum is always even

# SAMPLING & AD CONVERSION

- frequency spectrum after sampling ( $F_s$ : sampling freq)



$\rightarrow$  amplitudes repeats by sampling freq!



# SAMPLING & AD CONVERSION

## ▲ Effect of Sampling in time domain

Q: We saw that sampled sinusoid contains so many freq components.

Which one among them does sampled sinusoid look like in time domain?

A: The lowest frequency.

# SAMPLING & AD CONVERSION

## ▲ Ex. Aliasing Examples

1) sinusoid freq = 120Hz, sampling freq = 100Hz

amplitudes repeats by sampling freq: ..., -180, -80, 20, 120, ...

amplitude spectrum is even: ..., -180, -120, -80, -20, 20, 120, 180, ...

lowest freq = 20Hz

2) sinusoid freq = 120Hz, sampling freq = 150Hz

$120 - 150 = -30$ , 30Hz with opposite phase

3) sinusoid freq = 430Hz, sampling freq = 150Hz

$430 - 150 - 150 - 150 = -20$ , 20Hz with opposite phase

 **Run again m\_sampledsinusoid.m for above cases.**

- signal freq can be negative



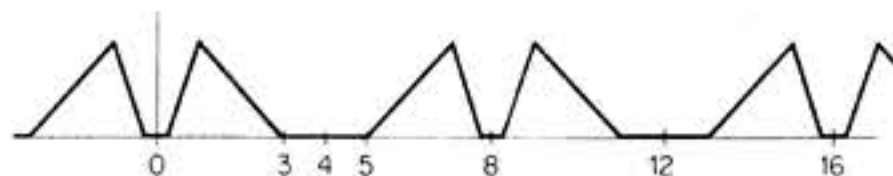
# SAMPLING & AD CONVERSION

## ▲ Aliasing for general signal

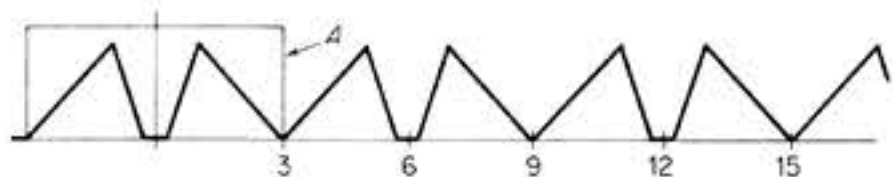
- consider analog signal with  $f_{\max} = 3\text{kHz}$



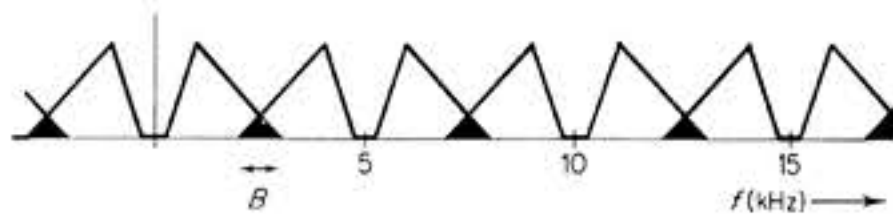
- after sampling



(a) sampled signal ( $f_s = 8\text{kHz}$ )



(b) sampled signal ( $f_s = 6\text{kHz}$ )

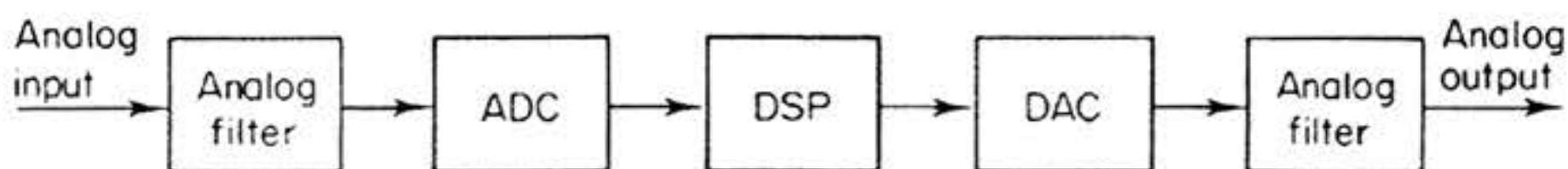


(c) sampled signal ( $f_s = 5\text{kHz}$ )  $\Rightarrow$  aliasing

# SAMPLING & AD CONVERSION

## ▲ Anti-aliasing filter

- analog filter before ADC



- ensures  $f_{\max} \leq \frac{f_s}{2}$

Q: Why must anti-aliasing filter be an analog filter?

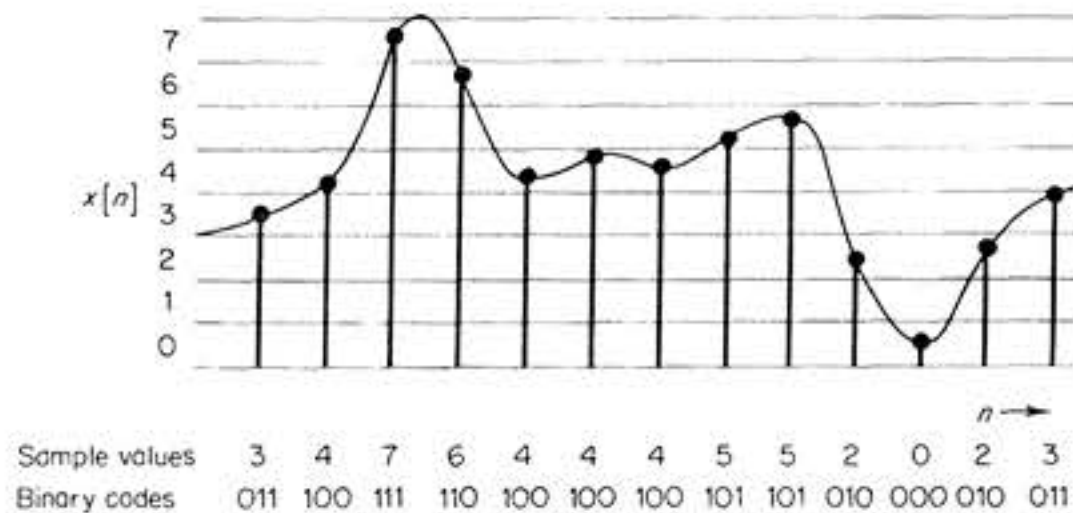
A: If it is a digital filter, sampling must come before this filter and sampling causes aliasing.



# SAMPLING & AD CONVERSION

## QUANTIZATION


continuous amplitude  $\xrightarrow{\text{quantization}}$  finitely many amplitudes (binary code)

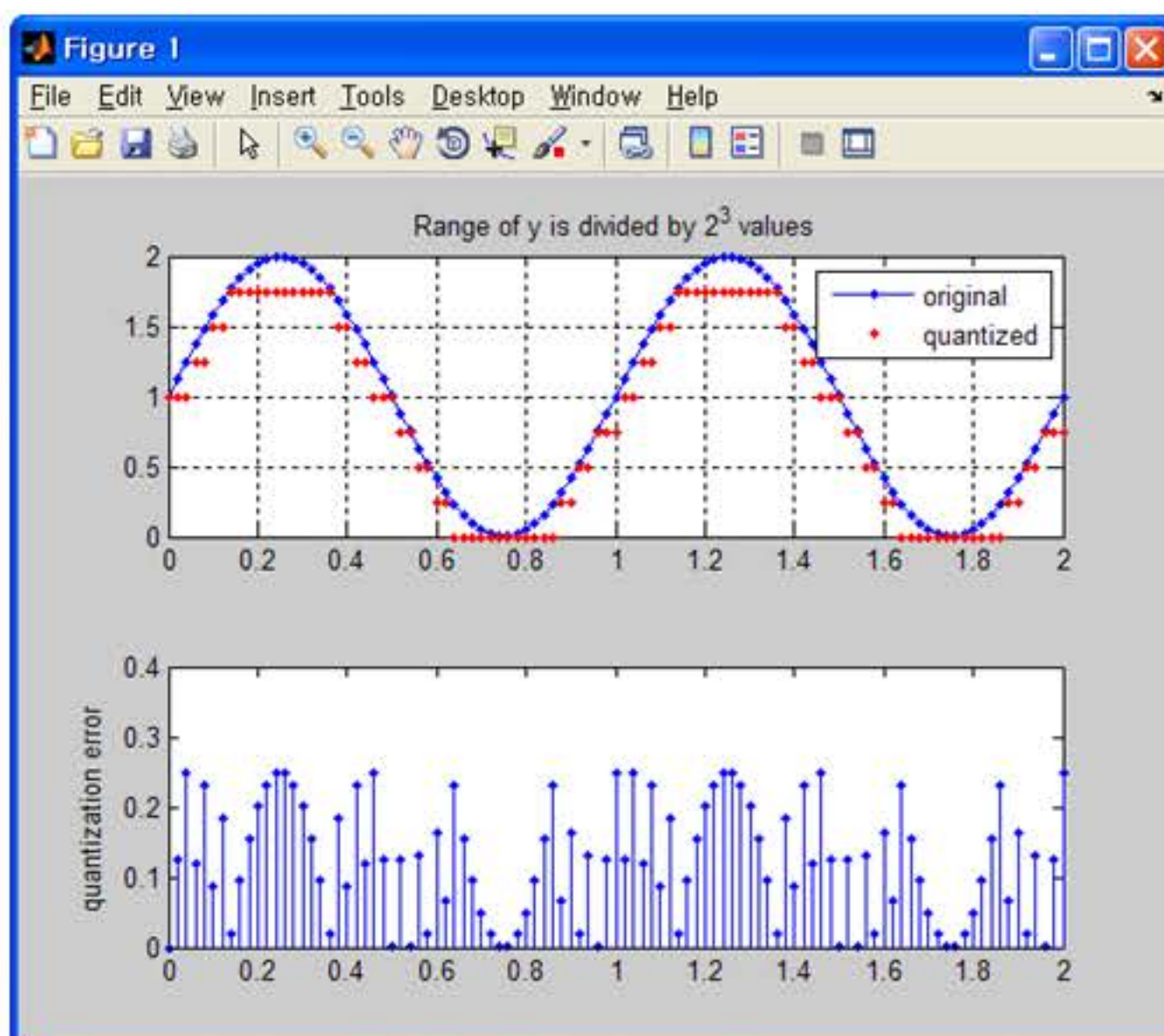


### quantization error

- difference b/w actual data and quantized data
- more bits (= better resolution) gives smaller error

# SAMPLING & AD CONVERSION

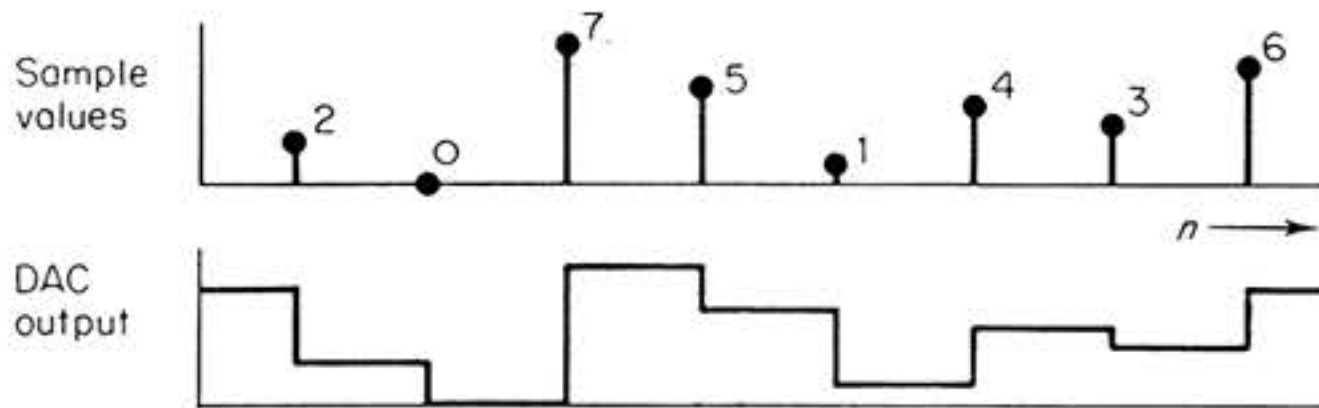
 Run `m_quantize.m` to understand quantization.



# SAMPLING & AD CONVERSION

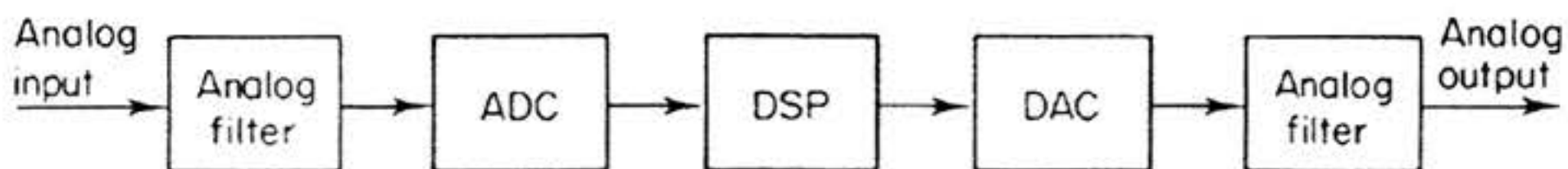
## DAC

- input: digital data
- output: zero-order-hold of input



## Smoothing Filter

- analog filter after DAC
- used when staircase waveform is not acceptable



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# DIGITAL SIGNAL

## DIGITAL SIGNALS

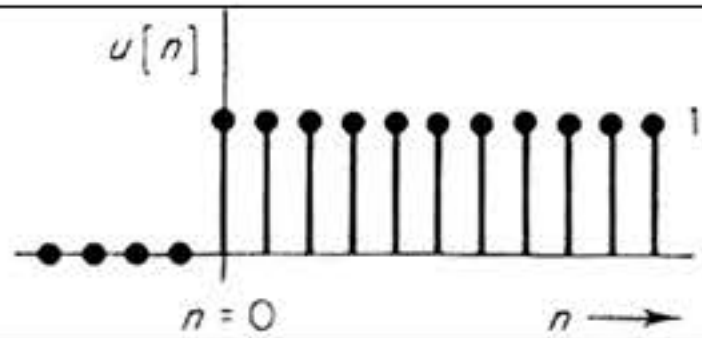
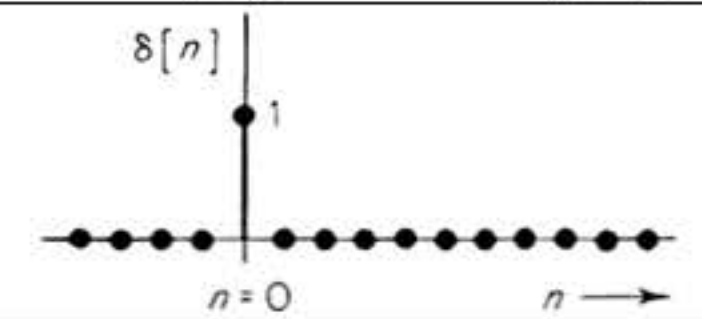
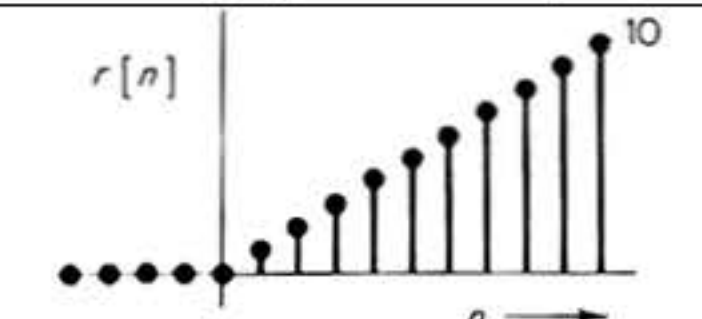
### Most important signals

- impulse
- step
- ramp
- exponential
- sinusoid



# DIGITAL SIGNAL

## STEP, IMPULSE, AND RAMP

step, $u[n]$	$u[n] = 0, n < 0$ $u[n] = 1, n \geq 0$	
impulse, $\delta[n]$	$\delta[n] = 0, n \neq 0$ $\delta[n] = 1, n = 0$	
ramp, $r[n]$	$r[n] = n u[n]$	

**Note:**  $n$  is an integer.

# DIGITAL SIGNAL

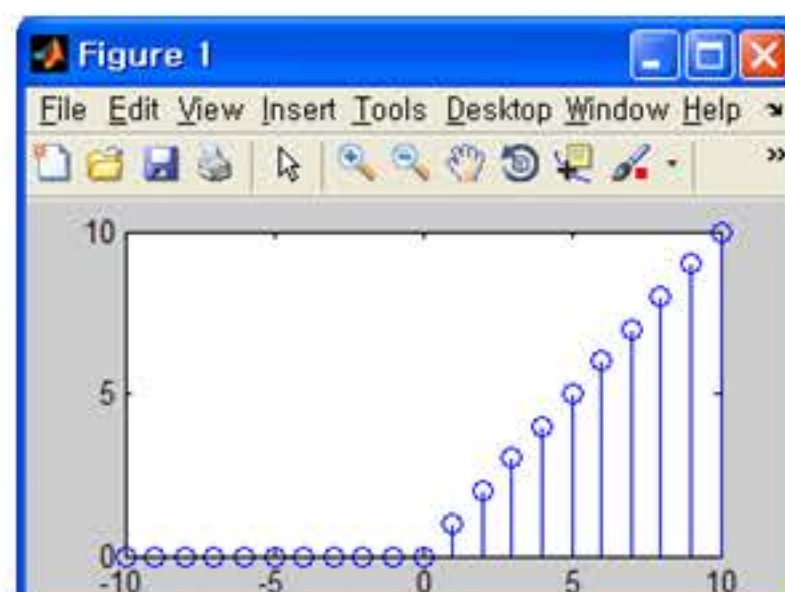
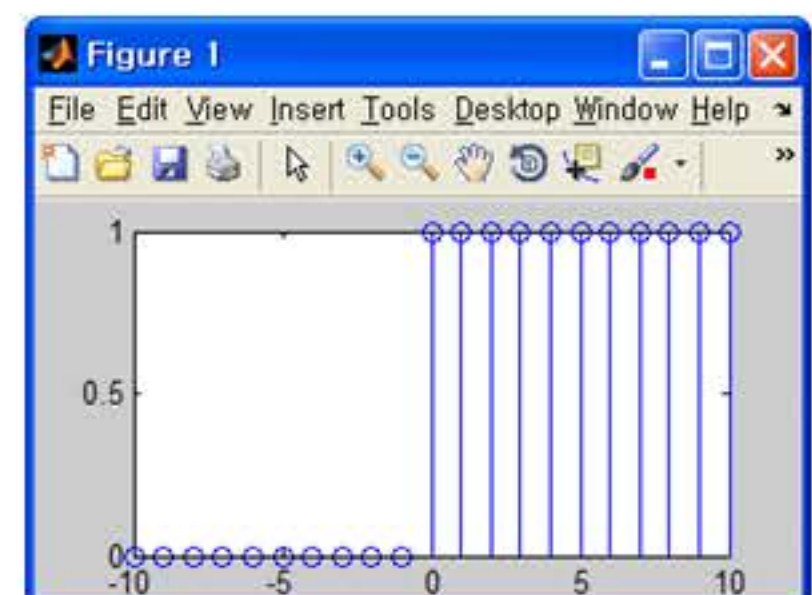
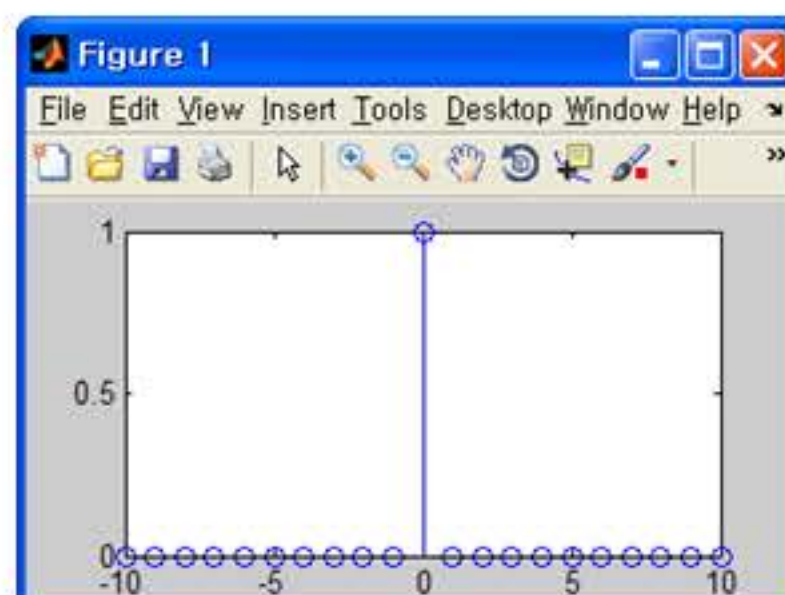
## Plot basic digital signals.

```

n = -10 : 10
n = n'
y1 = f_impulse(n)
y2 = f_step(n)
y3 = f_ramp(n)

stem(n,y1)
stem(n,y2)
stem(n,y3)

```





# DIGITAL SIGNAL

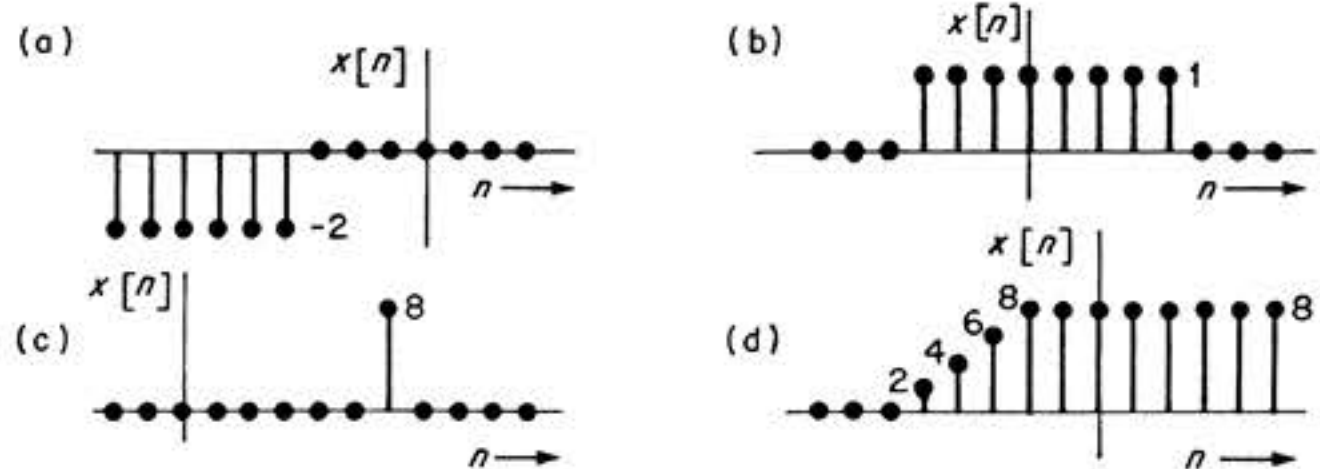
▲ Ex. Find expression.

(a)  $x[n] = -2u[-n-4]$

(b)  $x[n] = u[n+3] - u[n-5]$

(c)  $x[n] = \delta[n-8]$

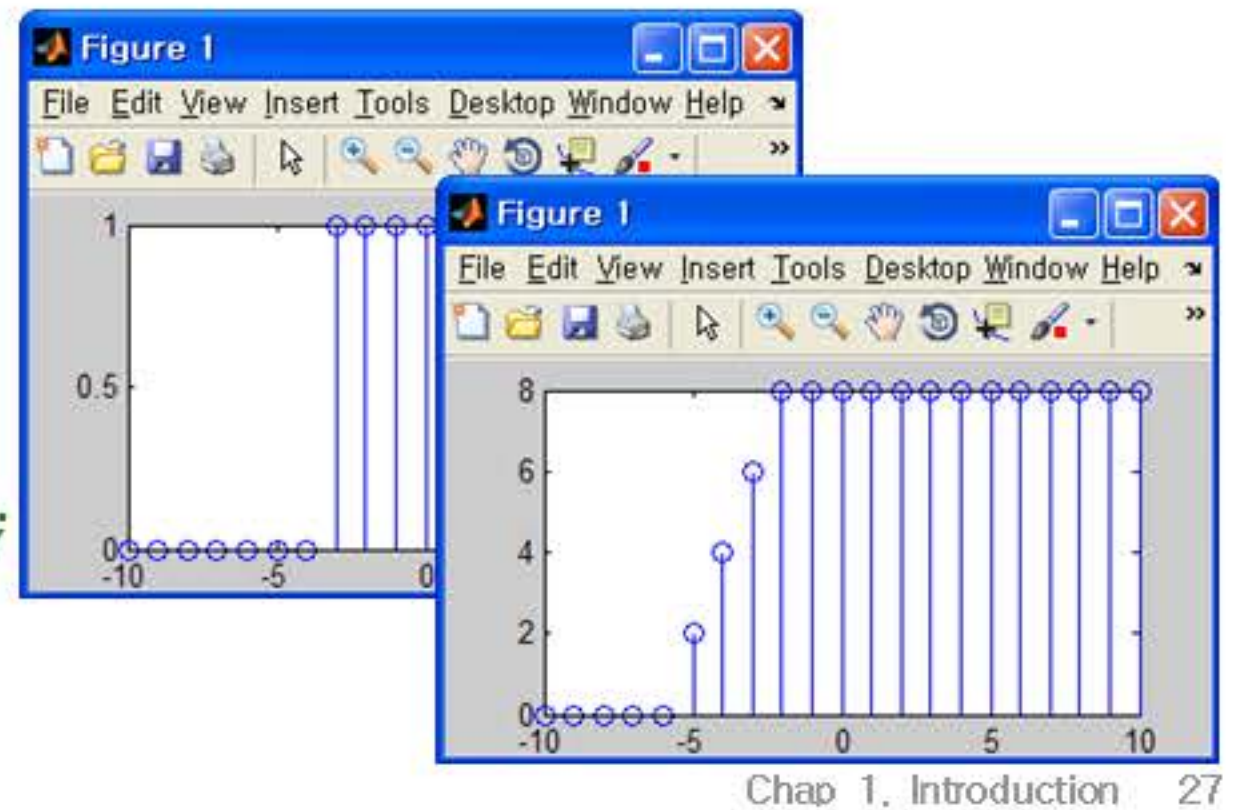
(d)  $x[n] = 2r[n+6] - 2r[n+2]$



📁 Plot (b) and (d).

```
y = f_step(n+3) - f_step(n-5);
stem(n,y)
```

```
y = 2*f_ramp(n+6) - 2*f_ramp(n+2);
stem(n,y)
```



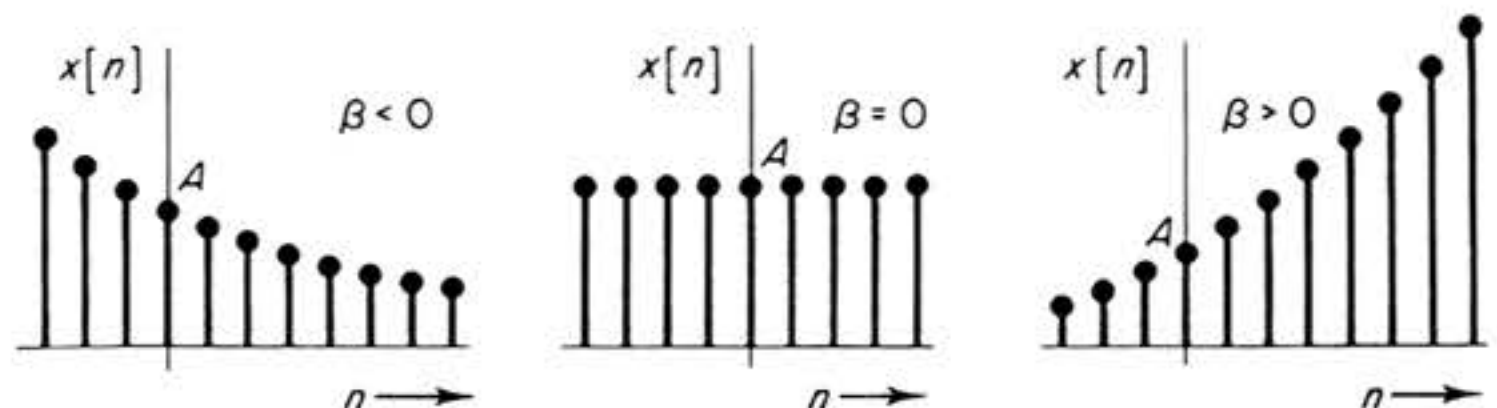
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# DIGITAL SIGNAL

## EXPONENTIAL

▲ Exponential signal

$$x[n] = A \exp(\beta n)$$



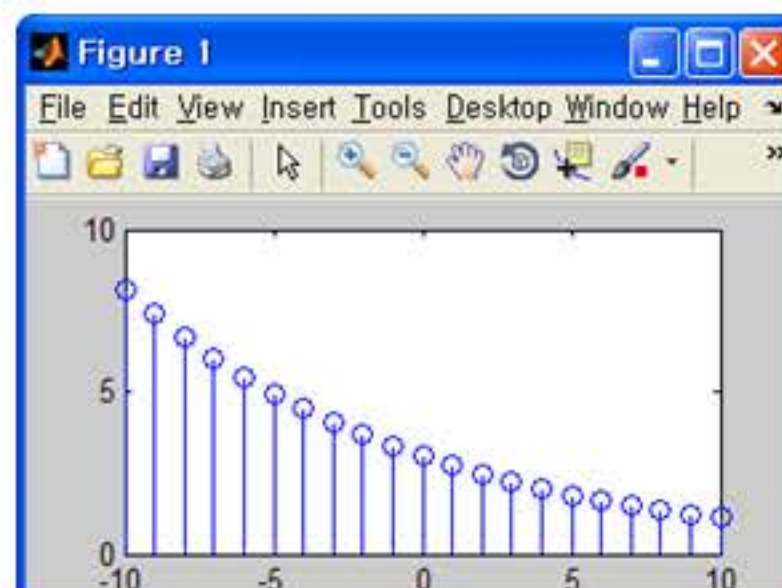
▲ Practical exponential signal

$$x[n] = A \exp(\beta n), n \geq 0 \quad \text{or} \quad x[n] = A \exp(\beta) u[n]$$

$$= 0, \quad n < 0$$

📁 Plot exponential signal.

```
y = 3*exp(-0.1*n);
stem(n,y)
```



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# DIGITAL SIGNAL

## SINUSOID

### digital sinusoid

cosine:  $A \cos(\Omega n)$ , sine:  $A \sin(\Omega n)$

### can be expressed as an exponential with imaginary power

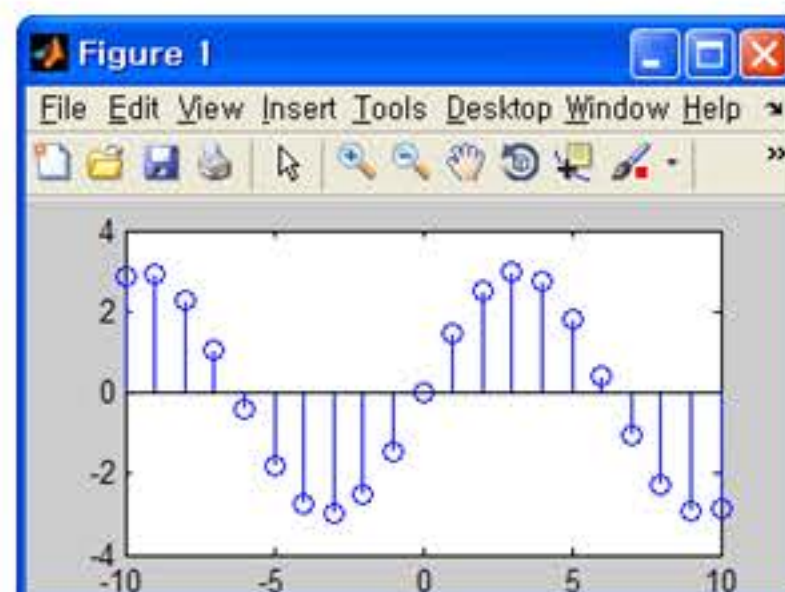
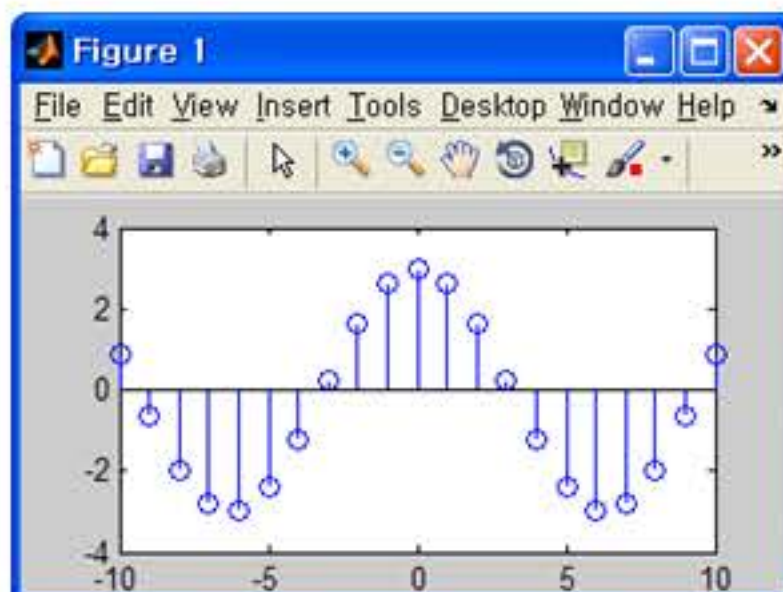
$$x[n] = A \exp(j \Omega n) = A \cos(\Omega n) + j A \sin(\Omega n)$$

$$\Rightarrow \begin{cases} A \cos(\Omega n) = \operatorname{Re}(A \exp(j \Omega n)) \\ \quad = \frac{A}{2} \exp(j \Omega n) + \frac{A}{2} \exp(-j \Omega n) \\ A \sin(\Omega n) = \operatorname{Im}(A \exp(j \Omega n)) \\ \quad = \frac{A}{2j} \exp(j \Omega n) - \frac{A}{2j} \exp(-j \Omega n) \end{cases}$$

# DIGITAL SIGNAL

## Plot $3\cos(0.5n)$ and $3\sin(0.5n)$ using exp.

```
y = 3*exp(0.5i*n)
stem(n,real(y)); stem(n,imag(y))
```





# DIGITAL SIGNAL

## ▲ Time concept in sampled sinusoid

$$x(t) = \sin(\omega_a t)$$

$$\downarrow \quad t = nT, \quad n = \dots, 0, 1, 2, \dots$$

$$x[n] = \sin(n\omega_a T) \quad \text{or} \quad x[n] = \sin\left(\frac{n 2\pi f_a}{f}\right)$$

$$\downarrow \quad \text{let } \Omega = \omega_a T$$

$$x[n] = \sin(\Omega n)$$

where

$\omega_a$  [rad/s],  $f_a$  [Hz]: analog sinusoid

$T$  [s],  $f$  [Hz]: sampling

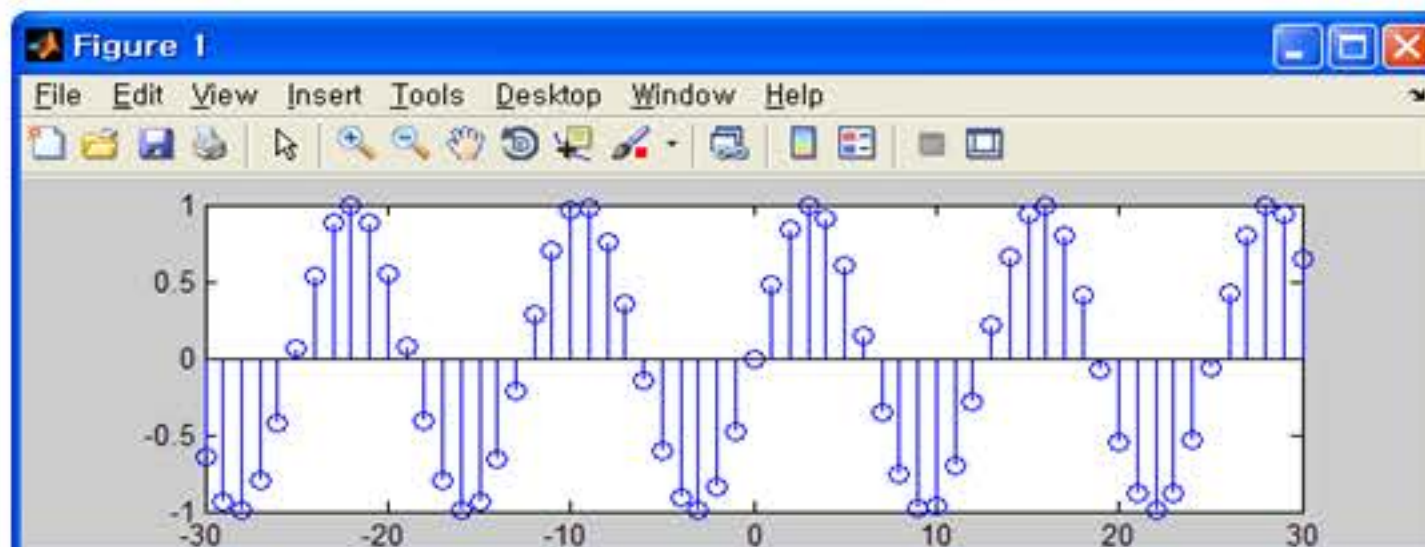
# DIGITAL SIGNAL

## ▲ Sampled sinusoid

- periodic only when  $\Omega/2\pi$  is a rational number (ratio of two integers)

$n = -30 : 30$

`stem(n, sin(0.5*n)) % sin(0.5n) is not periodic`



## ▲ Period

- period is represented by  $N$  ( = # of samples/cycle)
- $\Omega$  (frequency) has unit [rad].



# DIGITAL SIGNAL

## ▲ Ex. Find period

- 1)  $x[n] = \cos(0.2\pi n)$
- 2)  $x[n] = \sin(\frac{7\pi}{6}n)$
- 3)  $x[n] = \sin(0.5n)$
- 4)  $x[n] = \cos(\frac{2\pi}{3}n) + 0.7\sin(\frac{\pi}{2}n)$

## ▲ Sol.

- 1)  $N = 10$
- 2)  $N = 12$
- 3) aperiodic ( $0.5n$  ( $n$ : integer) never becomes a multiple of  $2\pi$ )
- 4) first term:  $N = 3$ , second term:  $N = 4 \Rightarrow N = 12$

# DIGITAL SIGNAL

## ▲ Sinusoid with exponential amplitude

$\Leftarrow$  when  $\beta$  is complex ( $= \beta_0 + j\Omega$ ) in  $A \exp(\beta n)$ ,

$$x[n] = A \exp(\{\beta_0 + j\Omega\}n) = A \exp(\beta_0 n) \exp(j\Omega n)$$

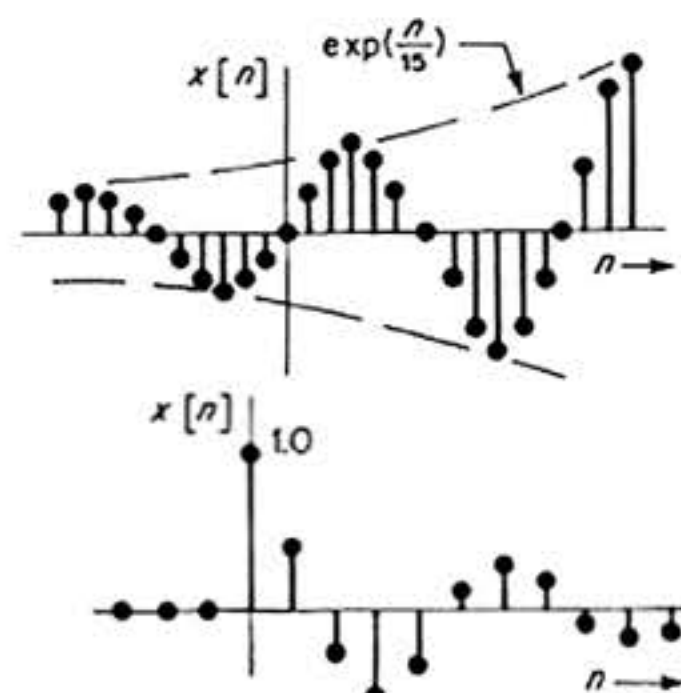
$$\Rightarrow \begin{cases} x[n] = A \exp(\beta_0 n) \sin(\Omega n) \\ x[n] = A \exp(\beta_0 n) \cos(\Omega n) \end{cases}$$

## ▲ Ex.

$$x[n] = \exp\left(\frac{n}{15}\right) \sin\left(\frac{\pi n}{6}\right)$$

```
y = exp((1/15+1/6*pi*1i)*n);
stem(n, imag(y))
```

$$x[n] = \exp\left(\frac{-n}{5}\right) \cos(n) u[n]$$



# DIGITAL SIGNAL

## ▲ Digital Sinusoid

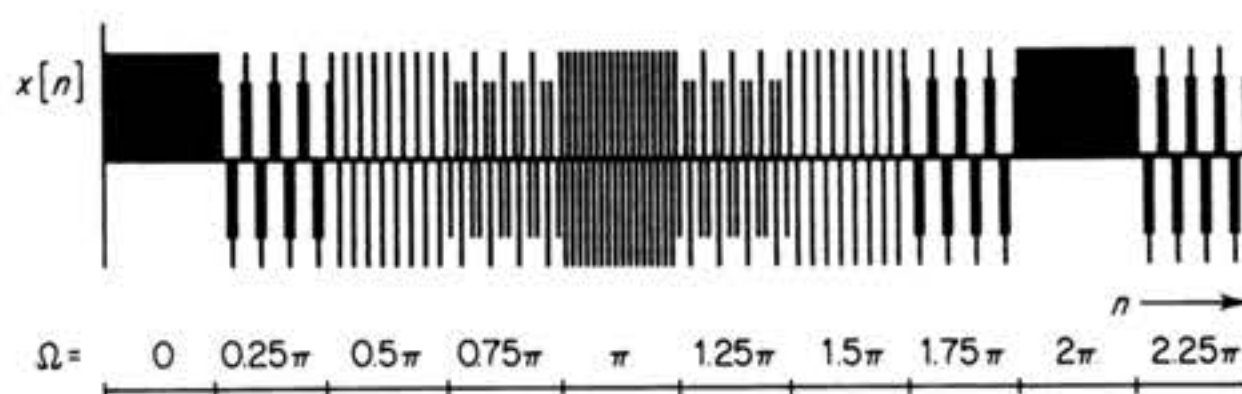
- A digital sinusoid with frequency  $\Omega$  is identical to those with  $\Omega \pm 2\pi$ ,  $\Omega \pm 4\pi$ , ...

$$x[n] = \exp(j\Omega n) = \exp(j\{\Omega + 2\pi\}n)$$

# DIGITAL SIGNAL

## ▲ Effect of frequency increase in digital sinusoid

$$x[n] = \cos(\Omega n) = \cos\left(\frac{2\pi m}{8}n\right), \quad (m = 0, 1, 2, \dots)$$



- $\Omega = 0 \sim \pi$  ( $m = 0 \sim 4$ )  $\Rightarrow$  rate of oscillation increases
- $\Omega = \pi \sim 2\pi$  ( $m = 4 \sim 8$ )  $\Rightarrow$  rate of oscillation decreases
- $\Omega > 2\pi$   $\Rightarrow$  pattern for  $\Omega \in [0, 2\pi]$  repeats

## 📄 Run m\_freqincrease.m

- shows effect when increasing  $m$  in digital sinusoid  $x[n] = \cos\left(\frac{2\pi m}{8}n\right)$



# DIGITAL SIGNAL

## ▲ Summary of digital sinusoid

- In  $x[n] = \cos(\Omega n)$ , no ambiguity when  $\Omega \leq \pi$ , but aliasing when  $\Omega > \pi$  (such effect does not happen in analog sinusoid,  $x(t) = \cos(\omega t)$ )
- $\Omega = \pi$  corresponds to  $f_a = \frac{f}{2}$  (Nyquist freq) i.e., fastest freq for a given sampling.

## ▲ Ex.

$x[n] = \cos(5\pi n)$  was obtained by sampling analog sinusoid with 10Hz freq. Find freq of original analog signal.

Sol)  $\Omega = \omega_a T \Rightarrow 5\pi = \omega_a \times 0.1 = 2\pi f_a \times 0.1 \Rightarrow f_a = 25\text{Hz}$

We sampled 25Hz sinusoid by 10Hz sampling! Aliasing!

It will look like 5Hz sinusoid after sampling. (5Hz = 25 - 10 - 10)

# DIGITAL SYSTEM

## 📁 LINEAR TIME-INVARIANT(LTI) SYSTEMS

### ▲ Form of digital processor (= system)

- CPU that is programmed for DSP task
- dedicated digital H/W

### ▲ linearity

linear system = system that obeys Principle of Superposition

known:  $x_1[n] \rightarrow \boxed{\text{linear system}} \rightarrow y_1[n]$ ,  $x_2[n] \rightarrow \boxed{\text{linear system}} \rightarrow y_2[n]$

unknown: 1)  $x_1[n] + x_2[n] \rightarrow \boxed{\text{linear system}} \rightarrow ?$

2)  $ax_1[n] \rightarrow \boxed{\text{linear system}} \rightarrow ?$

ans: 1)  $y_1[n] + y_2[n]$     2)  $ay_1[n]$



# DIGITAL SYSTEM

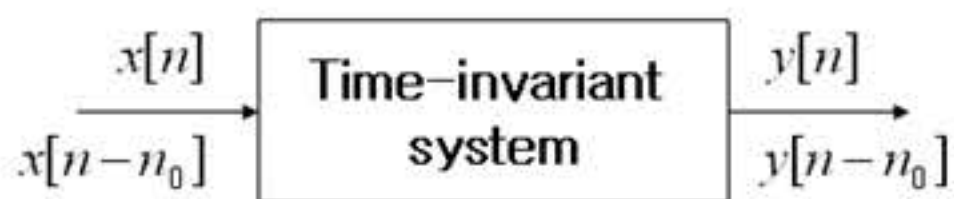
## ▲ frequency preservation

$$A \cos(\Omega n) \rightarrow \boxed{\text{linear system}} \rightarrow B \cos(\Omega (n + M))$$

output contains only those frequencies present in the input

## ▲ time invariance

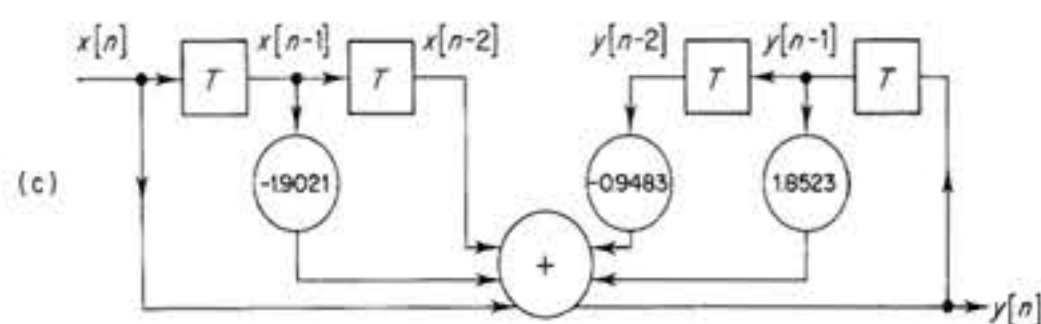
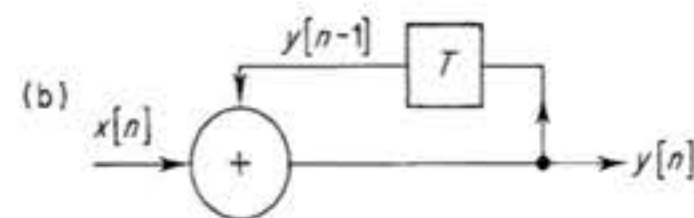
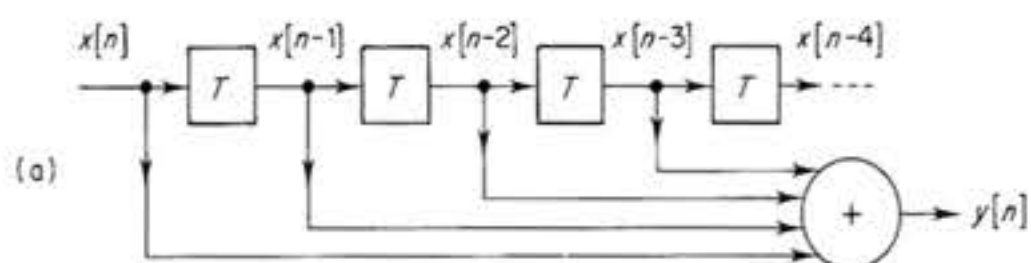
⇒ system properties don't change with time



# DIGITAL SYSTEM

## ▲ Block diagram (used for dedicated HW implementation)

- elements: summation, gain, delay  $\boxed{T}$



(a)  $y[n] = x[n] + x[n-1] + x[n-2] + \dots$

(b)  $y[n] = y[n-1] + x[n]$

(c)  $y[n] = 1.8523y[n-1] - 0.94833y[n-2] + x[n] - 1.9021x[n-1] + x[n-2]$



# DIGITAL SYSTEM

## MORE SYSTEM PROPERTIES

### ▲ causality

Output depends only on present and previous values of input.

### ▲ stability

System gives bounded output in response to bounded input.

### ▲ invertibility

If system gives  $y[n]$  in response to  $x[n]$ , its inverse gives  $x[n]$  in response to  $y[n]$ .

### ▲ memory

Output depends on previous input values.

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# DIGITAL SYSTEM

### ▲ Ex. Determine properties of following systems

(linearity, time invariance, causality, stability, invertibility, memory)

(a)  $y[n] = 3x[n] - 4x[n-1]$

⇒ all six properties hold

(b)  $y[n] = 2y[n-1] + x[n+2]$

⇒ noncausal, unstable

(c)  $y[n] = nx[n]$

⇒ time-variant, no memory

(d)  $y[n] = \cos(x[n])$

⇒ nonlinear, time-invariant, noninvertible, no memory

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