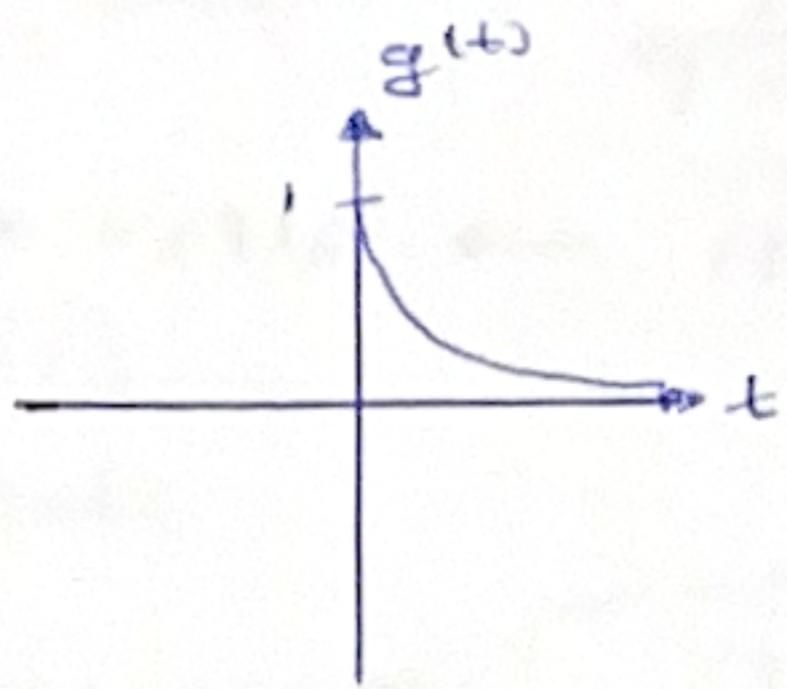


- ① An exponentially decaying pulse is denoted as $g(t) = e^{-at}u(t)$, where $u(t)$ is the unit step function.



- Ⓐ USE F.T. DEFINITION TO DERIVE THE FREQUENCY DOMAIN REPRESENTATION OF THE EXPONENTIALLY DECAYING PULSE.

$$\mathcal{F}[g(t)] = G(f) = \int_{-\infty}^{\infty} g(t) e^{-j2\pi f t} dt$$

$$\Rightarrow g(t) = e^{-at}u(t)$$

$$\Rightarrow \mathcal{F}[g(t)] = G(f) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-j2\pi f t} dt$$

$$= \int_0^{\infty} e^{-at} e^{-j2\pi f t} dt$$

$$= \int_0^{\infty} e^{-at + (-j2\pi f t)} dt$$

$$= \int_0^{\infty} e^{-(j2\pi f + a)t} dt$$

$$= \frac{-1}{j2\pi f + a} e^{-(j2\pi f + a)t} \Big|_0^{\infty}$$

$$= \frac{-1}{j2\pi f + a} \left(\lim_{t \rightarrow \infty} \frac{1}{e^{(j2\pi f + a)t}} - e^0 \right)$$

$$= \frac{-1}{j2\pi f + a} (0 - 1)$$

$$\boxed{G(f) = \frac{1}{j2\pi f + a}}$$

b/c of $u(t)$
✓

*THE INTEGRAND IS 0 FOR $(-\infty, 0)$, SO WE CAN CHANGE BOUNDS OF DOPP $u(t)$.

- Ⓑ IS THE F.T. REAL OR COMPLEX VALUES?

$G(f)$ is complex valued ($G: \mathbb{R} \rightarrow \mathbb{C}$) as it contains an imaginary component.

THE FOURIER TRANSFORMS OF REAL AND EVEN SYMMETRIC SIGNALS ARE ALSO PURELY REAL,

HOWEVER, THIS SIGNAL $g(t)$ IS ASYMMETRIC, THIS REQUIRING A COMPLEX FOURIER TRANSFORM PAIR.

① FIND THE AMPLITUDE SPECTRUM ($|G(f)|$) AND PHASE SPECTRUM $\theta(f)$ OF $g(t)$.

$$\begin{aligned}
 G(f) &= \frac{1}{a+j2\pi f} \cdot \frac{(a-j2\pi f)}{(a-j2\pi f)} \\
 &= \frac{a-j2\pi f}{a^2+4\pi^2f^2} \\
 &= \underbrace{\frac{a}{a^2+4\pi^2f^2}}_{X(f)} + j \underbrace{\frac{-2\pi f}{a^2+4\pi^2f^2}}_{Y(f)} \rightarrow G(f) = X(f) + jY(f)
 \end{aligned}$$

$|G(f)|$

$$\begin{aligned}
 |G(f)| &= \sqrt{X(f)^2 + Y(f)^2} = \sqrt{\frac{a^2}{(a^2+4\pi^2f^2)^2} + \frac{4\pi^2f^2}{(a^2+4\pi^2f^2)^2}} \\
 &= \sqrt{\frac{a^2+4\pi^2f^2}{(a^2+4\pi^2f^2)^2}} \\
 |G(f)| &= \frac{1}{\sqrt{a^2+4\pi^2f^2}}
 \end{aligned}$$

$\theta(f)$

$$\begin{aligned}
 \theta(f) &= \arctan\left(\frac{-2\pi f}{a^2+4\pi^2f^2}\right) = \arctan\left(\frac{-2\pi f}{a}\right) \\
 \Rightarrow \theta(f) &= \arctan\left(\frac{-2\pi f}{a}\right)
 \end{aligned}$$

② CODE & PLOT ATTACHED.

③ SHOW F.T. OF $g(t)$ EXHIBITS CONJUGATE SYMMETRY; AMPLITUDE EVEN, PHASE ODD.

$$\Rightarrow |G(f)| = \frac{1}{\sqrt{a^2+4\pi^2f^2}} \Rightarrow |G(-f)| = \frac{1}{\sqrt{a^2+4\pi^2(-f)^2}} = \frac{1}{\sqrt{a^2+4\pi^2f^2}} = |G(f)| \quad \blacksquare$$

\therefore AMPLITUDE SPECTRUM IS EVEN-SYMMETRIC.

$$\Rightarrow \theta(f) = \arctan\left(\frac{-2\pi f}{a}\right) \Rightarrow \theta(-f) = \arctan\left(\frac{-2\pi(-f)}{a}\right) = \arctan\left(-\left(\frac{2\pi f}{a}\right)\right) = -\arctan\left(\frac{2\pi f}{a}\right) = -\theta(f) \quad \blacksquare$$

The diagram shows two graphs side-by-side. The left graph is labeled "arctan(theta)" and shows a curve starting from negative infinity as theta approaches 0 from the left, passing through the origin (0,0), and approaching positive infinity as theta increases. The right graph is labeled "arctan(-theta) = -arctan(theta)" and shows a similar curve, but it is reflected across the x-axis, starting from positive infinity as theta approaches 0 from the left, passing through the origin (0,0), and approaching negative infinity as theta increases.

\therefore PHASE SPECTRUM IS ODD-SYMMETRIC.

\Rightarrow BOTH CONDITIONS MET, SO THE F.T. OF $g(t)$ HAS CONJUGATE SYMMETRY.

hw1-code

February 10, 2025

- (3) Plot the amplitude spectrum and phase spectrum of $G(f)$

```
[26]: import matplotlib.pyplot as plt
import numpy as np

plt.rcParams.update({
    "text.usetex": True,
    "font.family": "Helvetica"
})

# DEFINE EACH AS FUNCTIONS:
# G(f) = 1/(a + j2pi*f)
# |G(f)| = 1/sqrt(a**2 + 4pi**2f**2)
# Ø(f) = arctan(-2pi*f/a)

def G(f, a):
    return 1 / (np.complex(a, 2*np.pi*f))

def amplitude_G(f, a):
    return 1 / np.sqrt(a**2 + (2*np.pi*f)**2)

def phase_G(f, a):
    return np.arctan(-2*np.pi*f / a)

# Create range of frequencies (-10kHz to 10kHz)
f = np.linspace(-100, 100, 500)

# Temp variable for a
a = 1

# Extract the spectra
amplitude = amplitude_G(f, a)
phase = phase_G(f, a)

# Plot the spectra
plt.figure(figsize=(10,5))

plt.subplot(1,2,1)
```

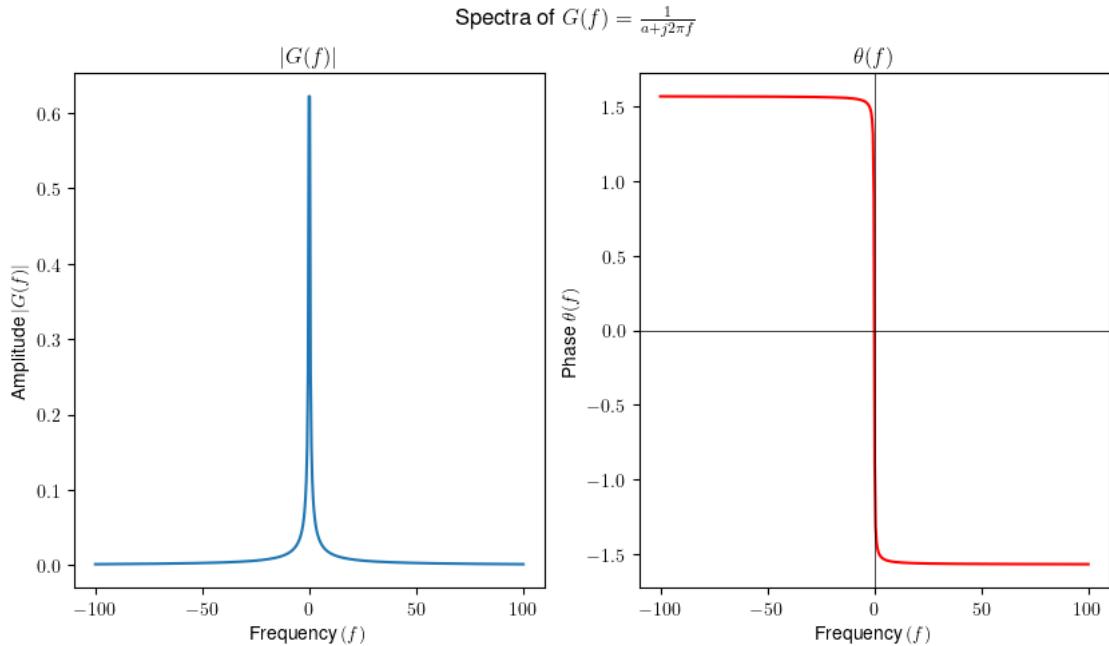
```

plt.plot(f, amplitude)
plt.title(r'$|G(f)|$')
plt.xlabel(r'Frequency $(f)$')
plt.ylabel(r'Amplitude $|G(f)|$')

plt.subplot(1,2,2)
plt.plot(f, phase, c='red')
plt.title(r'$\theta(f)$')
plt.xlabel(r'Frequency $(f)$')
plt.ylabel(r'Phase $\theta(f)$')
plt.axhline(y=0, color='black', linewidth=0.5)
plt.axvline(x=0, color='black', linewidth=0.5)

plt.suptitle(r'Spectra of $G(f)=\frac{1}{a+j2\pi f}$')
plt.show()
plt.close()

```



② CONSIDER THE TIME-DOMAIN REPRESENTATION OF A LOW-PASS FILTER:

$$g(t) = A \sin(2\pi f_0 t)$$

A) FIND F.T. OF FILTER

* TIME-SHIFT: $g(t-\tau) \Rightarrow g(t)e^{-j2\pi f_0 \tau}$

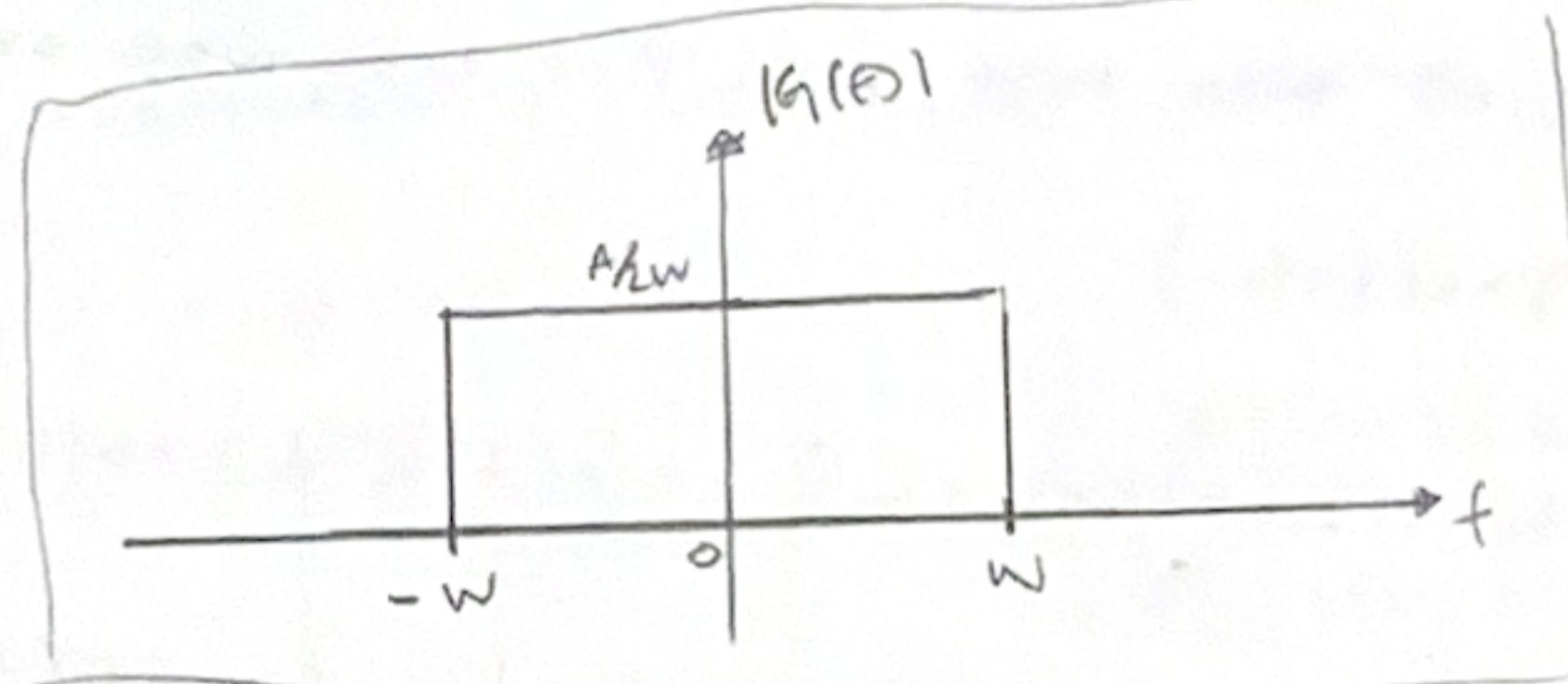
$$\begin{aligned} \Rightarrow \mathcal{F}[g(t)] &= \mathcal{F}[A \sin(2\pi f_0 t)] \\ &= \frac{A}{2} \operatorname{rect}\left(\frac{f}{2f_0}\right) e^{-j2\pi f_0 t} \\ \Rightarrow G(f) &= \frac{A}{2} e^{-j2\pi f_0 t} \operatorname{rect}\left(\frac{f}{2f_0}\right) \end{aligned}$$

B) PLOT THE AMPLITUDE RESPONSE

* $|e^{-j2\pi f_0 t}| = |\cos(2\pi f_0 t) - j \sin(2\pi f_0 t)| = \sqrt{\cos^2(2\pi f_0 t) + \sin^2(2\pi f_0 t)} = \sqrt{1} = 1.$

$$\Rightarrow |G(f)| = \frac{A}{2} \operatorname{rect}\left(\frac{f}{2f_0}\right)$$

* $T = 2f_0$

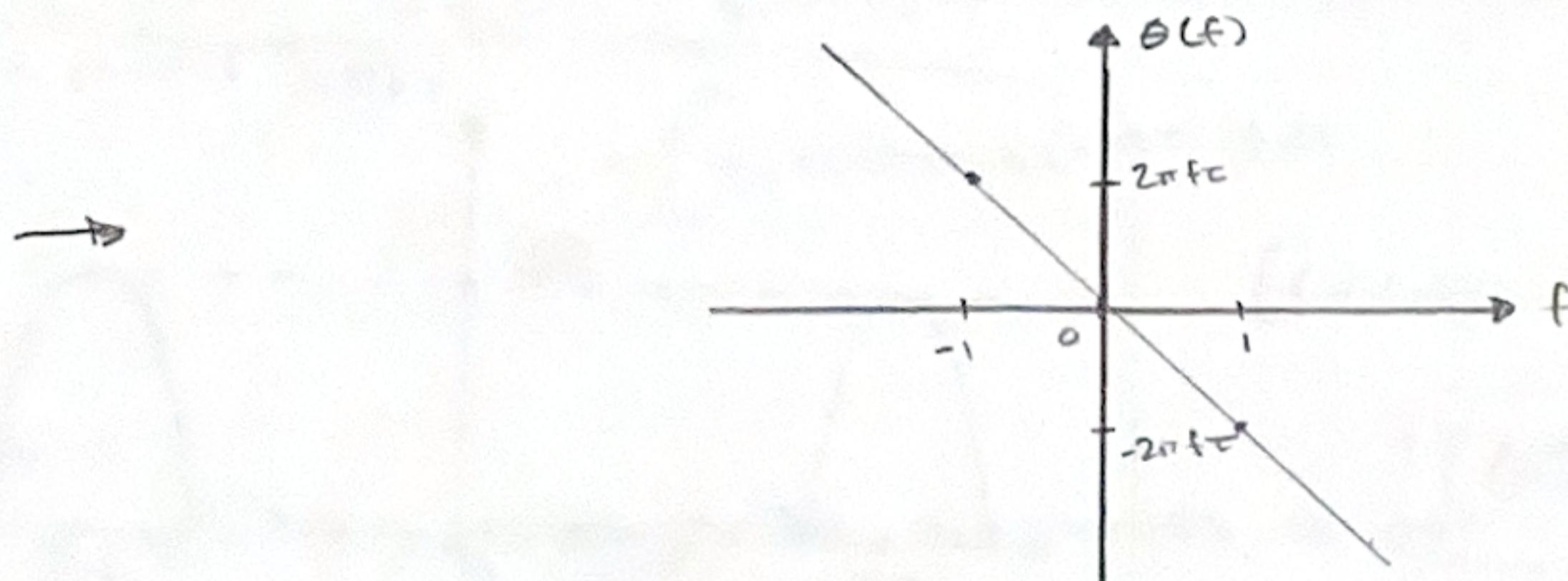


C) PLOT THE PHASE RESPONSE

For $G(f) = X(f) + jY(f) \Rightarrow \theta(f) = \arctan\left(\frac{Y(f)}{X(f)}\right)$, so we can view $G(f)$ as:

$$\rightarrow G(f) = \frac{A}{2} \operatorname{rect}\left(\frac{f}{2f_0}\right) (\cos(2\pi f_0 t) - j \sin(2\pi f_0 t))$$

$$\Rightarrow \theta(f) = \arctan\left(\frac{-\sin(2\pi f_0 t)}{\cos(2\pi f_0 t)}\right) = -\arctan(\tan(2\pi f_0 t)) = -2\pi f_0 t.$$



D) EFFECT OF TIME-SHIFT IN FREQUENCY DOMAIN?

THE TIME-DELAY CREATES A CONSTANT CHANGE IN PHASE $\forall f \in \mathbb{R}$.

E) WHAT IS THE SIGNIFICANCE IN TIME SHIFTING IN DESIGNING FILTERS?

SINCE THE STANDARD SINC PULSE IS INFINITE IN DOMAIN, AND NONCAUSAL,
THE PULSE NEEDS TO BE SHIFTED INTO POSITIVE TIME TO MAKE A
PHYSICALLY REALIZABLE CAUSAL SINC PULSE. TO DO SO, A TIME-SHIFT
IS REQUIRED.

also needs truncation

③ CONSIDER:

$$y(t) = \beta g(t) \cos(2\pi f_c t), \text{ where } \beta \in \mathbb{R}; g(t) \geq g(f).$$

④ FIND F.T OF $y(t)$

$$\mathcal{F}[w(2\pi f_c t)] = \frac{1}{2} (\delta(f-f_c) + \delta(f+f_c))$$

$$\begin{aligned} * \text{MULTIPLICATION IN} &\Rightarrow g_1(t)g_2(t) \stackrel{\text{TIME DOMAIN}}{\Leftrightarrow} G_1(f) \otimes G_2(f) \\ & \text{* SHIFTING PROPERTY} \quad f(x) \otimes S(x-a) = f(x-a) \end{aligned}$$

$$\Rightarrow \mathcal{F}[y(t)] = \beta G(f) \otimes \frac{1}{2} (\delta(f-f_c) + \delta(f+f_c))$$

$$= \frac{\beta}{2} [g(f) * \delta(f-f_c) + G(f) \otimes \delta(f+f_c)]$$

$$= \frac{\beta}{2} [G(f-f_c) + G(f+f_c)]$$

$$\Rightarrow Y(f) = \frac{\beta}{2} [G(f-f_c) + G(f+f_c)]$$

⑤ IF $g(t) = \cos(2\pi f_c t)$, THEN FIND F.T. PUT AMPLITUDE SPECTRUM FOR $f_c \gg f_0$.

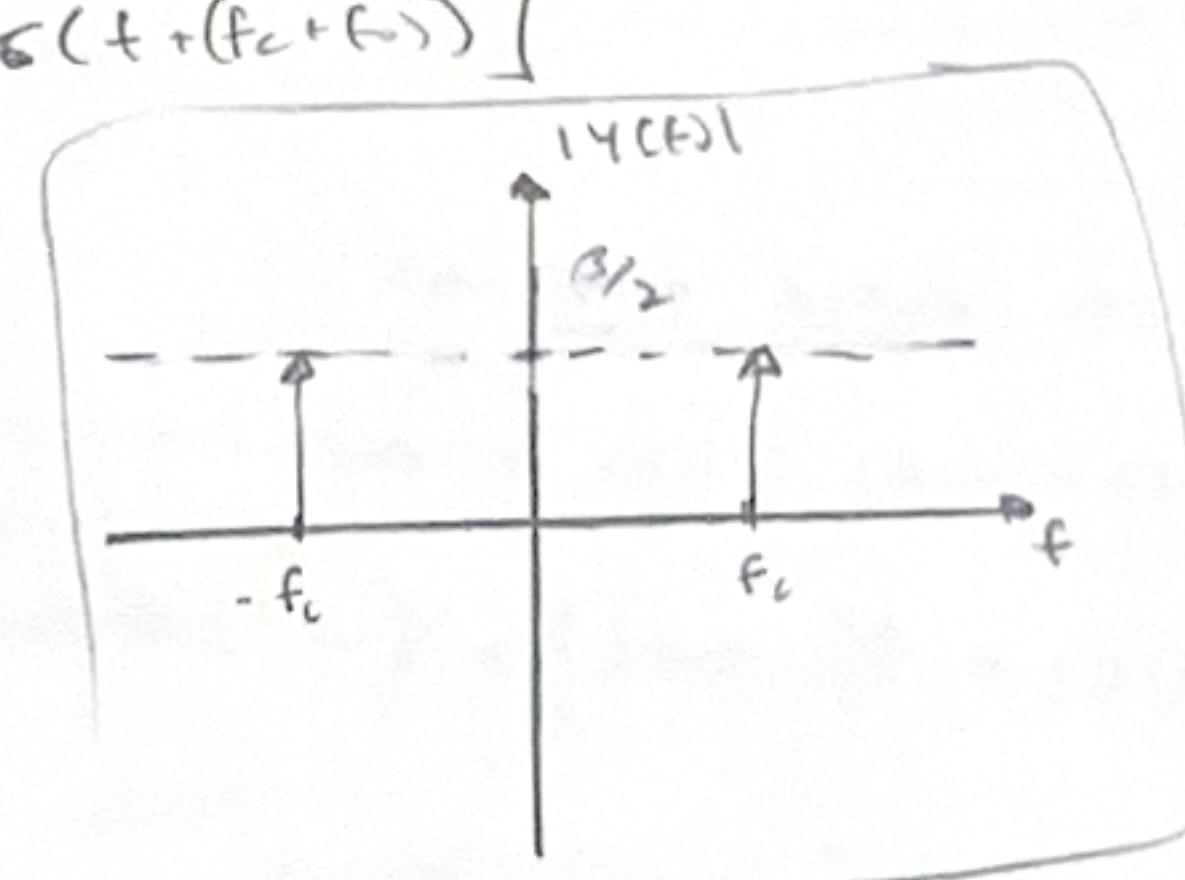
$$\Rightarrow \mathcal{F}[g(t)] = \frac{1}{2} [\delta(f-f_c) + \delta(f+f_c)]$$

$$\text{Hence} \Rightarrow \frac{\beta}{2} \left[\frac{1}{2} [\delta(f-f_c-f_0) + \delta(f-f_c+f_0)] + \frac{1}{2} [\delta(f+f_c-f_0) + \delta(f+f_c+f_0)] \right]$$

$$= \frac{\beta}{4} [\delta(f-(f_c+f_0)) + \delta(f+(-f_c+f_0)) + \delta(f+(f_c-f_0)) + \delta(f+(f_c+f_0))]$$

$$\text{FOR } f_c \gg f_0 \quad \hookrightarrow \quad Y(f) = \frac{\beta}{4} [\delta(f-f_c) + \delta(f-f_c) + \delta(f+f_c) + \delta(f+f_c)]$$

$$Y(f) = \frac{\beta}{2} [\delta(f-f_c) + \delta(f+f_c)] = |Y(f)|$$

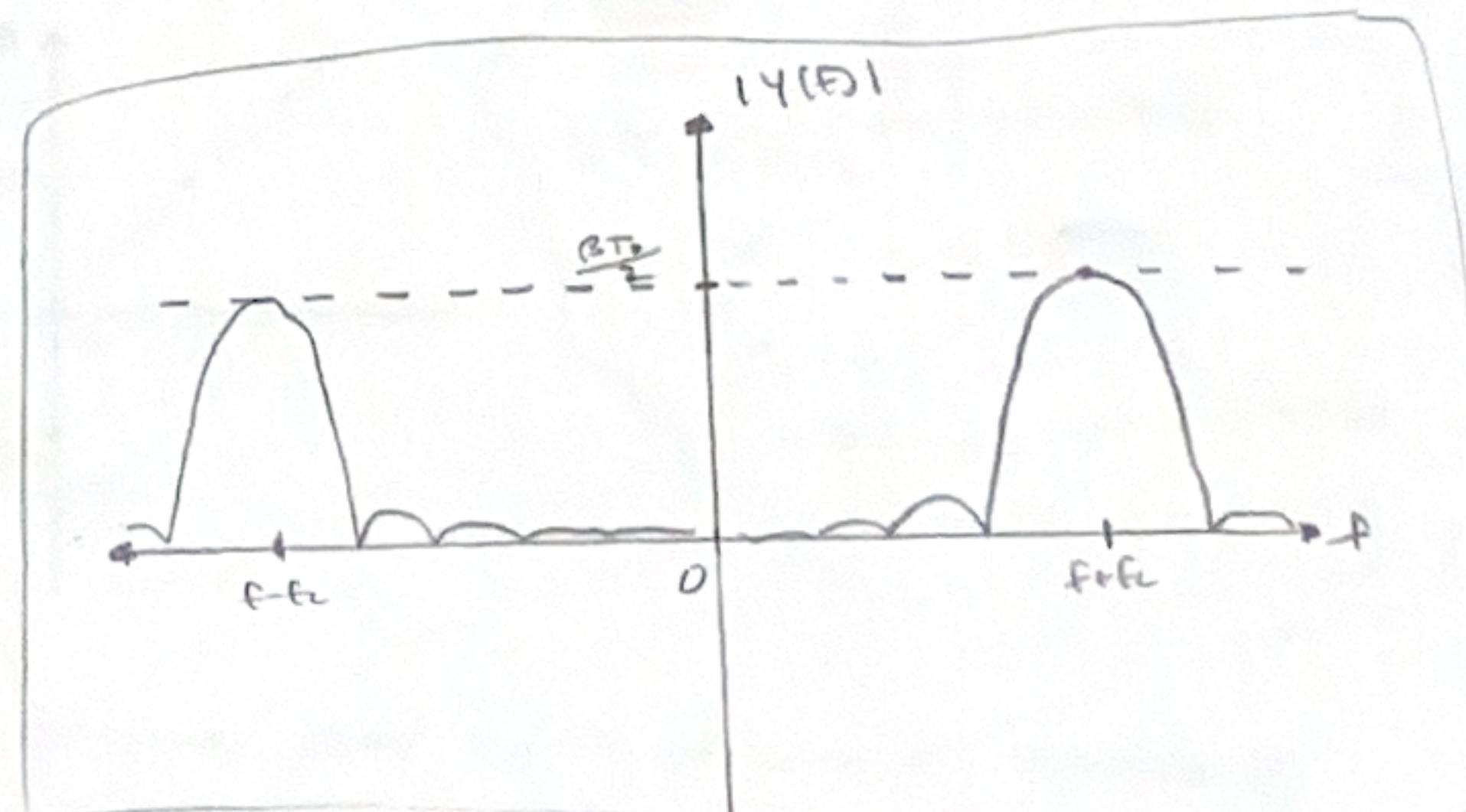


⑥ IF $g(t) = \text{rect}(t/T_0)$, FIND F.T. PUT AMPLITUDE SPECTRUM FOR $f_c \gg f_0 = 1/T_0$.

$$\mathcal{F}[g(t)] = T_0 \text{sinc}(fT_0)$$

$$\Rightarrow Y(f) = \frac{\beta T_0}{2} [\text{sinc}((f-f_c)T_0) + \text{sinc}((f+f_c)T_0)]$$

$$Y(f) = \frac{\beta T_0}{2} \left[\text{sinc}\left(\frac{f-f_c}{T_0}\right) + \text{sinc}\left(\frac{f+f_c}{T_0}\right) \right]$$

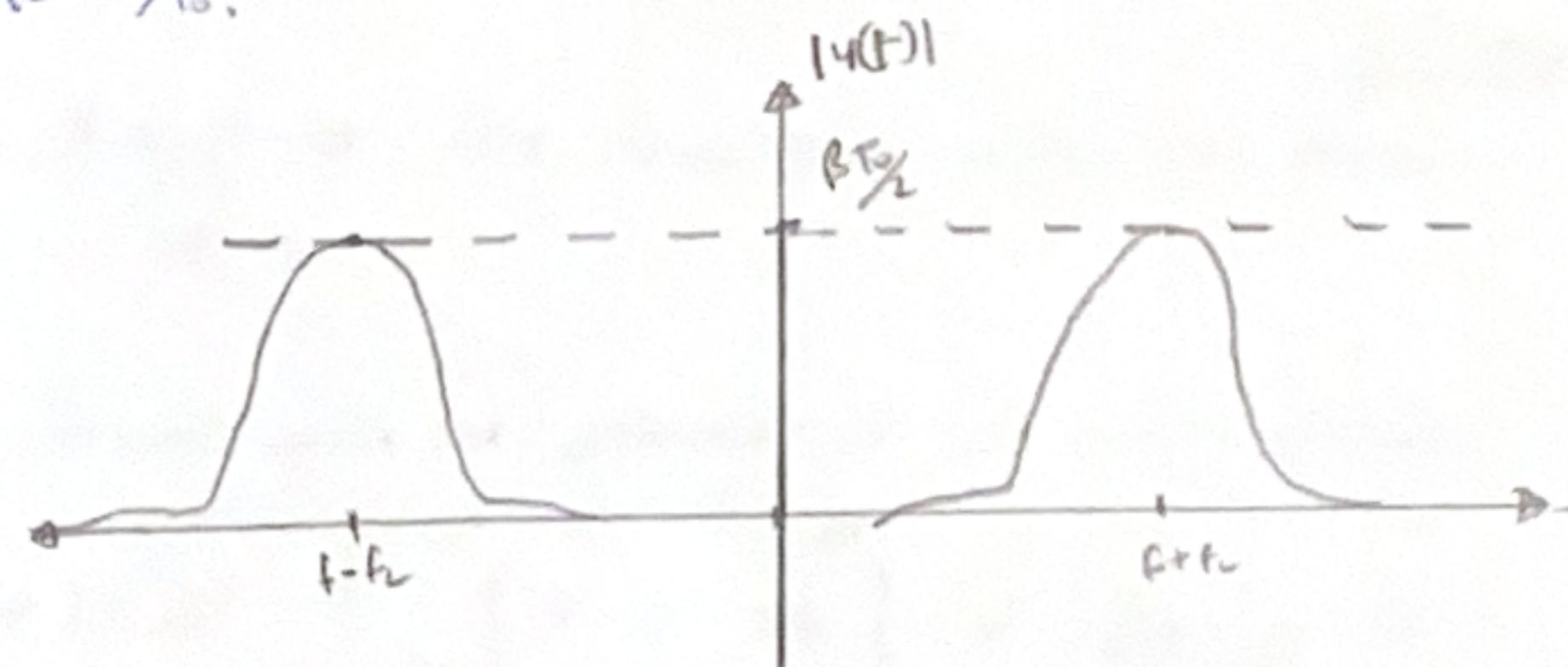


Q) $g(t) = \Delta(t/\tau_0)$, FIND F.T., PLOT TIME-SIGNAL FOR $f_c > f_0 = 1/\tau_0$.

$$Y(f) = \beta/2 [G(f-t_0) + G(f+t_0)]$$

$$\Rightarrow Y[g(t)] = \tau_0 \sin^2(f/\tau_0)$$

$$\Rightarrow Y(f) = \frac{\beta\tau_0}{2} \left[\sin^2\left(\frac{f-f_0}{\tau_0}\right) + \sin^2\left(\frac{f+f_0}{\tau_0}\right) \right]$$



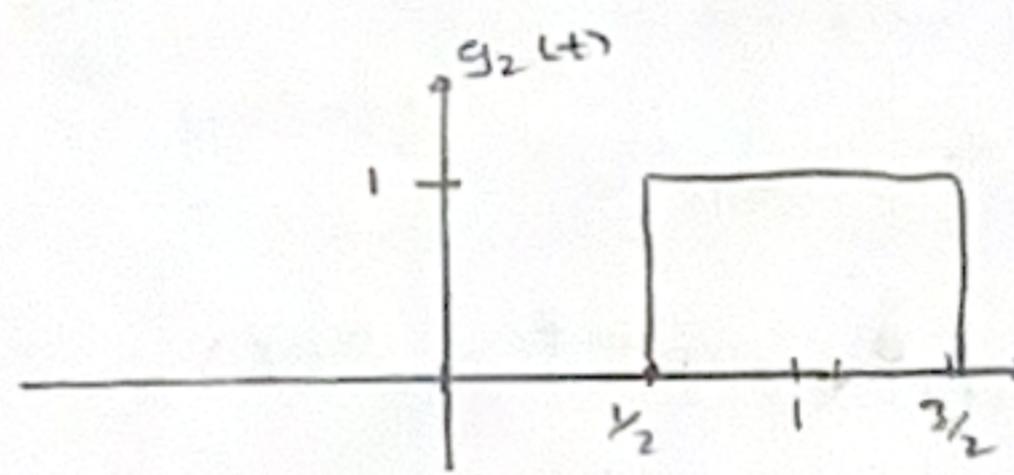
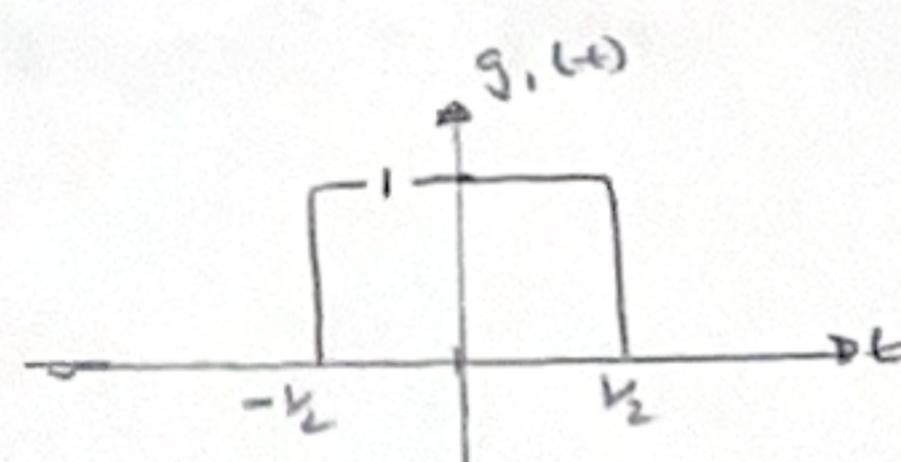
Q) WHAT ARE SOME IMPLICATIONS OF THE FREQUENCY DOMAIN ANALYSIS OF $y(t) = \beta g(t) \cos(2\pi f_0 t)$.

THIS $y(t)$ CAUSES A SLOWING BY β (ATTENUATION OR BOOSTING), AND A CENTERING OF MEAN FREQUENCY (INTENT AROUND $f = f_0 - f_c$).

THIS MODULATION COULD BE USEFUL IN EITHER CREATING A BANDPASS FILTER OR CHANNEL THE CHANNEL INFORMATION IS TRANSMITTED.

(4) CONSIDER $g_1(t) = \text{rect}(t)$ AND $g_2(t) = \text{rect}(t-1)$. LET $g_{12}(t) = g_1(t) * g_2(t) = \int_{-\infty}^{\infty} g_1(\tau) g_2(t-\tau) d\tau$.

A) PLOT $g_1(t) + g_2(t)$ IN TIME-DOMAIN. FIND F.T.S.



$$\begin{cases} Y[g_1(t)] = G_1(f) = \sin(f) \\ Y[g_2(t)] = G_2(f) = e^{-j2\pi f} \sin(f) \end{cases}$$

B) FIND F.T. OF $g_{12}(t)$.

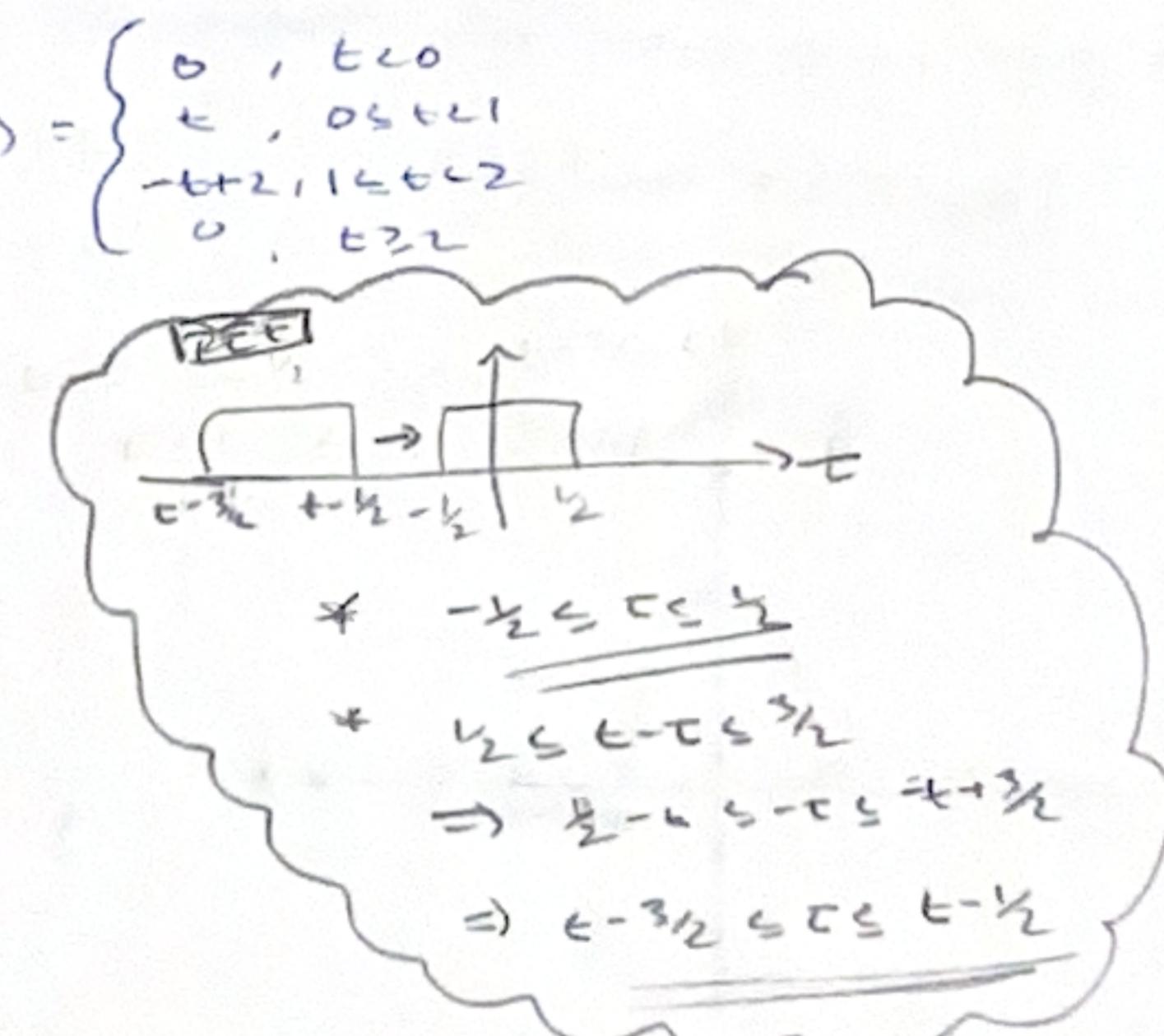
$$Y[g_{12}(t)] = Y[(g_1 * g_2)(t)] = G_1(f) G_2(f) = \left| e^{-j2\pi f} \sin^2(f) \right| = G_{12}(f)$$

C) EVALUATING CONVOLUTION INTEGRAL, SHOW THAT $g_{12}(t) \Leftrightarrow \Delta(t-1) = \begin{cases} 0, t < 0 \\ 1, 0 \leq t < 1 \\ -1, 1 \leq t < 2 \\ 0, t \geq 2 \end{cases}$

$$\begin{aligned} g_1(t) &= \text{rect}(t) \rightarrow g_{12}(t) = \int_{-\infty}^{\infty} g_1(\tau) g_2(t-\tau) d\tau \\ g_2(t) &= \text{rect}(t-1) \\ &= \int_{-\infty}^{\infty} \text{rect}(\tau) \text{rect}(t-(\tau-1)) d\tau \end{aligned}$$

CASE 1

HERE $t - 1/2 < -1/2$, OR, $t < 0$, THE TWO PULSES DO NOT OVERLAP $\therefore g_{12}(t) \Big|_{t<0} = 0$.



CASE 2

HERE $t - 1/2 > -1/2 \Rightarrow t > 0$ BUT $t - 3/2 \leq -1/2 \Rightarrow t \leq 1$ SO IN TOTAL $\Rightarrow 0 < t \leq 1$

WE START TO SEE OVERLAP BETWEEN $g_1(t)$ AND $g_2(t-\tau)$.

TO INTEGRATE THIS OVERLAP IN THE τ -DOMAIN, WE WILL ACCOUNT $\forall \tau$ S.T. $-1/2 \leq \tau \leq t - 1/2$.

KNOWING HERE $\text{rect}(\tau) \text{rect}(t-\tau) = 1 \cdot 1 = 1$.

$$\Rightarrow g_{12}(t) \Big|_{0 < t \leq 1} = \int_{-1/2}^{t-1/2} d\tau = \tau \Big|_{-1/2}^{t-1/2} = (t - 1/2) - (-1/2) = t \quad \therefore g_{12}(t) \Big|_{0 < t \leq 1} = t$$

cont'd

CASE 3

THERE IS STILL OVERLAP FOR $t - \nu_2 > \frac{1}{2}$ AND $t - \nu_1 \leq \frac{1}{2} \Rightarrow \underline{1 \leq t \leq 2}$.

$$t > 1 \quad t \leq 2$$

INTEGRATING IN τ -DOMAIN, WE WILL SWIPE FROM $t - \nu_1 \leq \tau \leq \nu_2$

$$\Rightarrow g_{12}(t) \Big|_{1 \leq t \leq 2} = \int_{t-\nu_2}^{\nu_2} d\tau = \tau \Big|_{t-\nu_2}^{\nu_2} = \nu_2 - (t - \nu_2) = \underline{2-t}$$

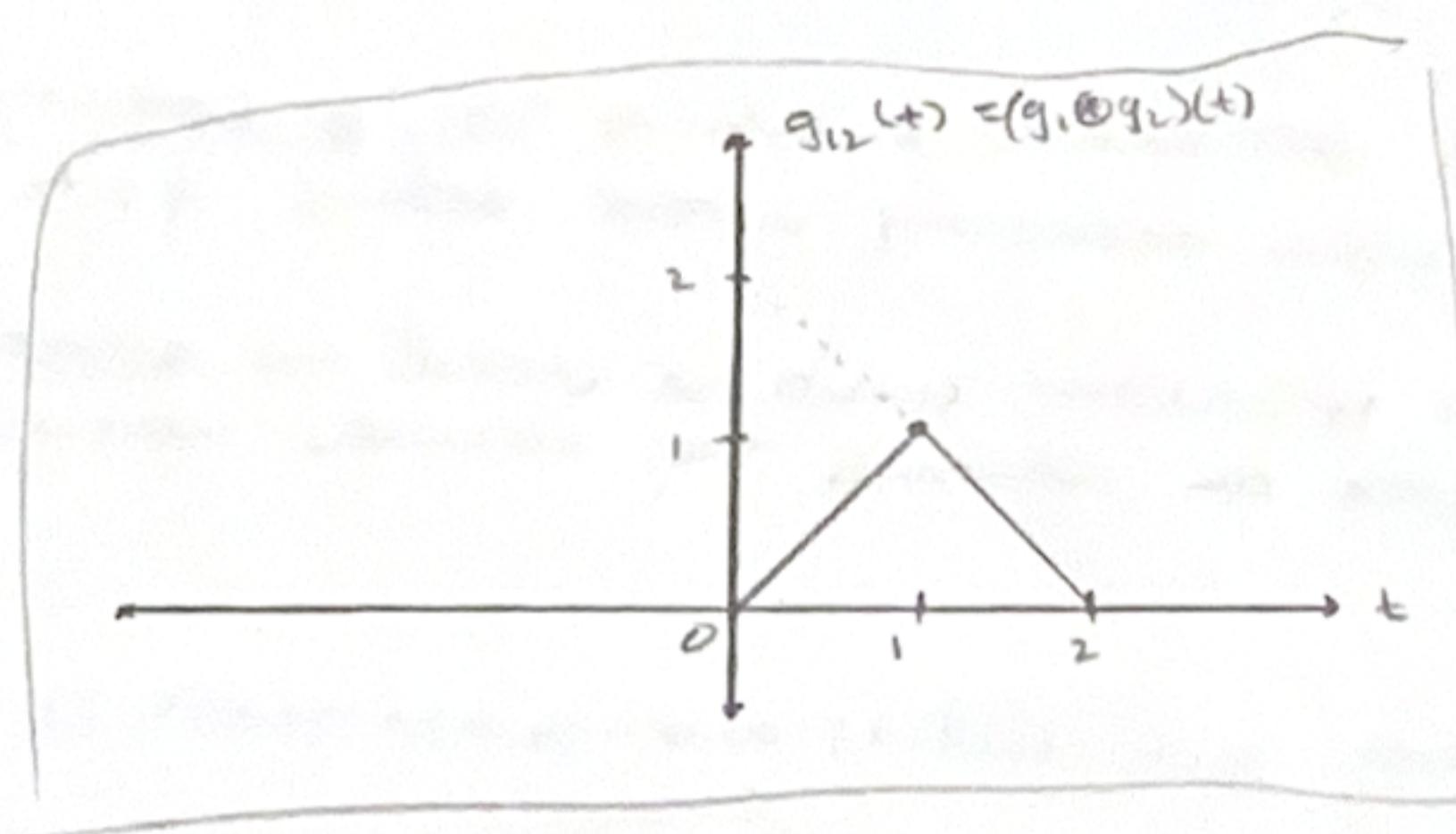
$$\therefore g_{12}(t) \Big|_{1 \leq t \leq 2} = 2-t.$$

CASE 4

WHEN $t - \nu_2 > \nu_2 \Rightarrow t > 2$, THEN WE REACH NO MORE OVERLAP $\Rightarrow g_{12}(t) \Big|_{t>2} = 0$.

COMBINED:

$$\Rightarrow g_{12}(t) = \begin{cases} 0, & t < 0 \\ t, & 0 \leq t \leq 1 \\ 2-t, & 1 \leq t \leq 2 \\ 0, & t > 2 \end{cases}$$

(D) FIND E.T. OF $g_{12}(t)$ USING TIME-SHIFTING PROPERTY.

$$* \mathcal{F}[\delta(t)] = \text{sinc}^2(f) \Rightarrow \mathcal{F}[\delta(t-1)] = \text{sinc}^2(f) \cdot e^{-j2\pi f(1)}$$

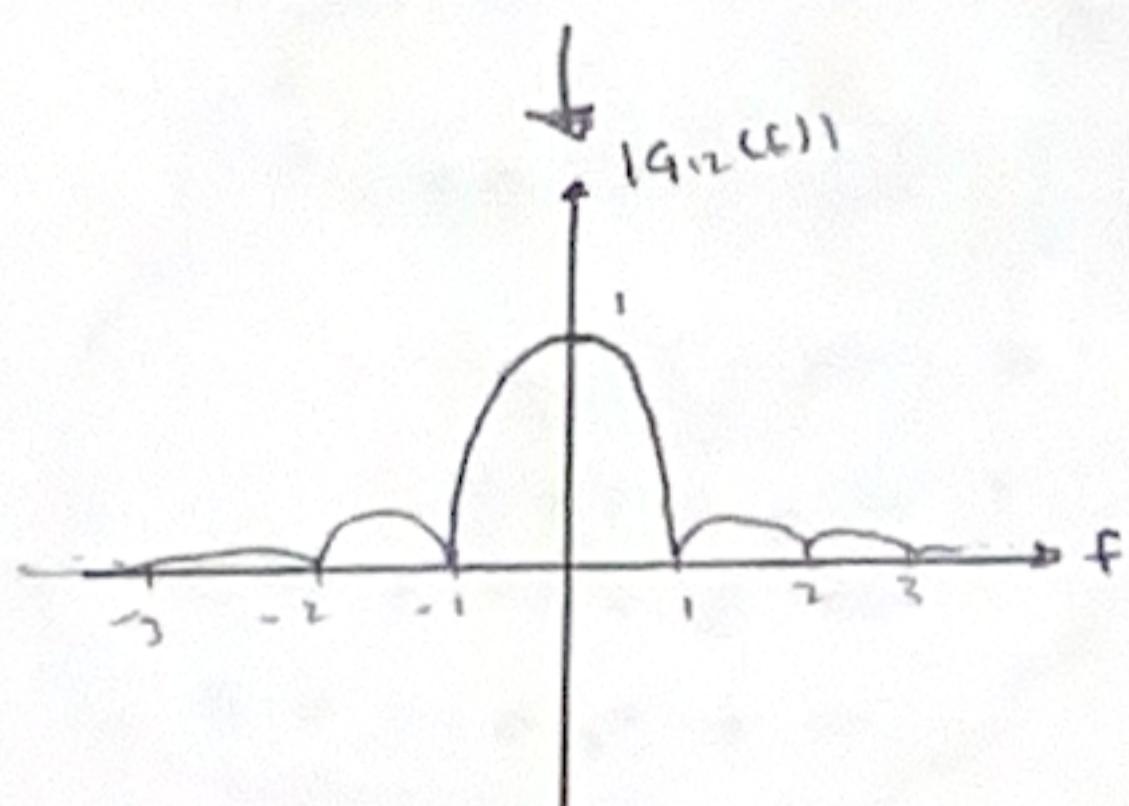
$$\Rightarrow G_{12}(f) = e^{-j2\pi f} \text{sinc}^2(f)$$

(E) VERIFY (B) & (D) EQUIV.

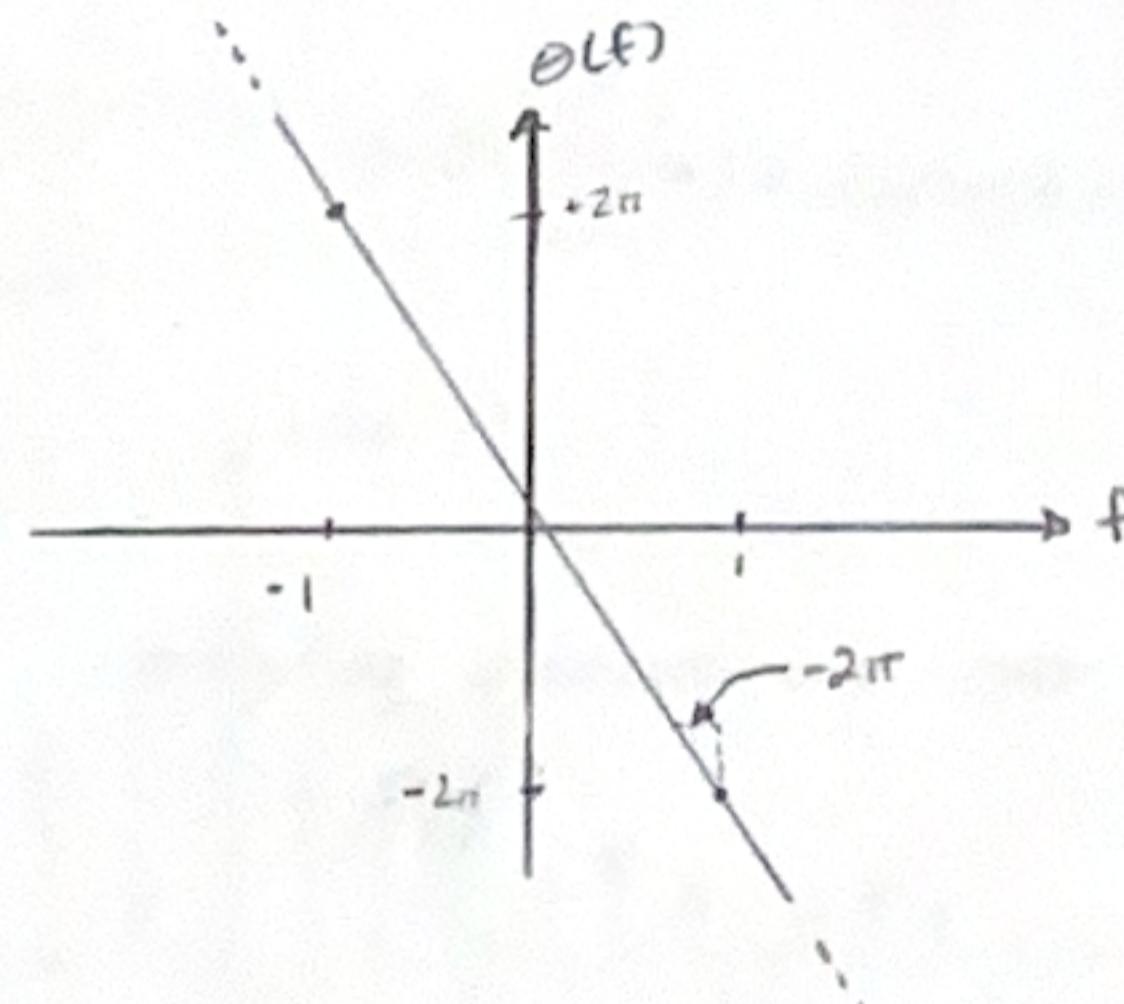
THEY'RE THE SAME. ✓

(F) PLOT AMPLITUDE AND PHASE SPECTRUM.

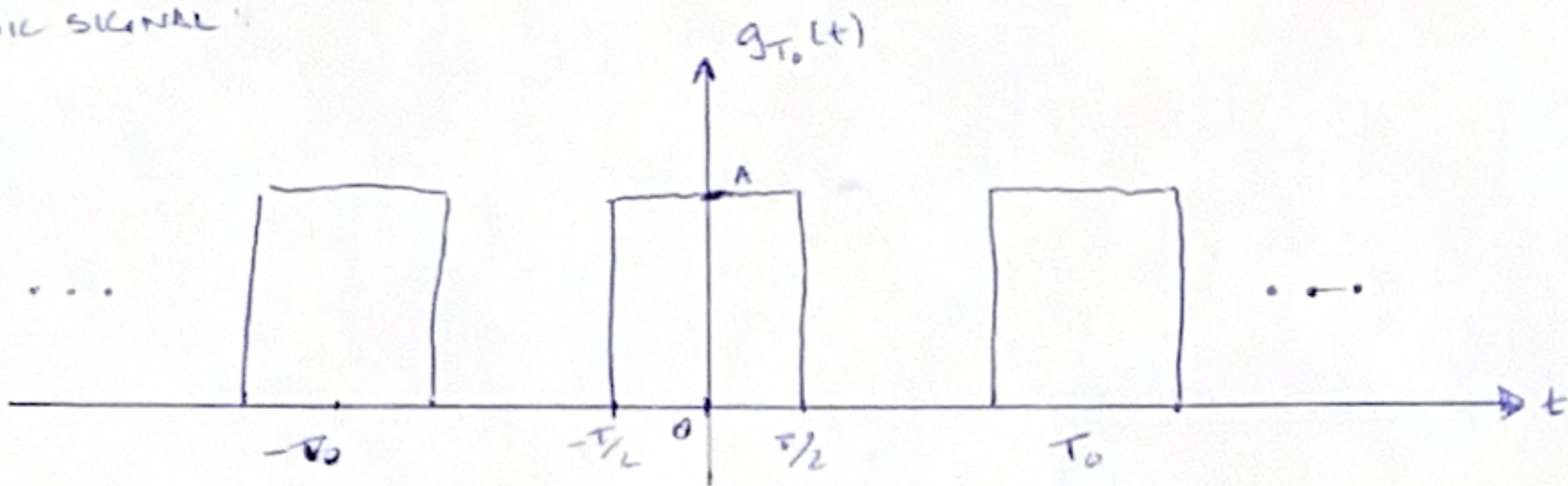
$$|G_{12}(f)| = \text{sinc}^2(f)$$



$$\begin{aligned} \theta(f) &= \arctan(\tan(-2\pi f)) \\ &= -2\pi f \end{aligned}$$



(5) Consider the periodic signal:



A) Find the generating function $g(t)$ of $g_{T_0}(t)$.

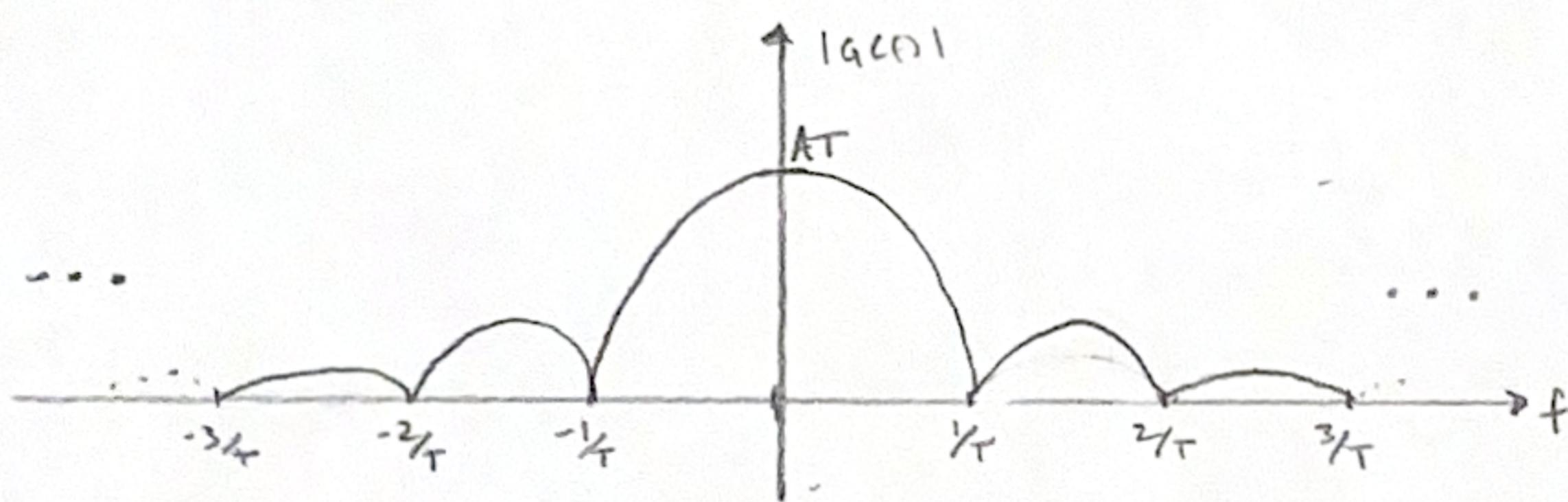
The generating function $g(t)$ by inspection is $\boxed{g(t) = \text{rect}(t/T)}$

B) Express $g_{T_0}(t)$ in terms of the its generating function $g(t)$.

$$\Rightarrow \boxed{g_{T_0}(t) = \sum_{m=-\infty}^{\infty} A \text{rect}\left(\frac{t-mT_0}{T}\right)}, \text{ where } m \in \mathbb{Z}$$

C) Find the F.T. of the generating function and plot as amplitude spectrum.

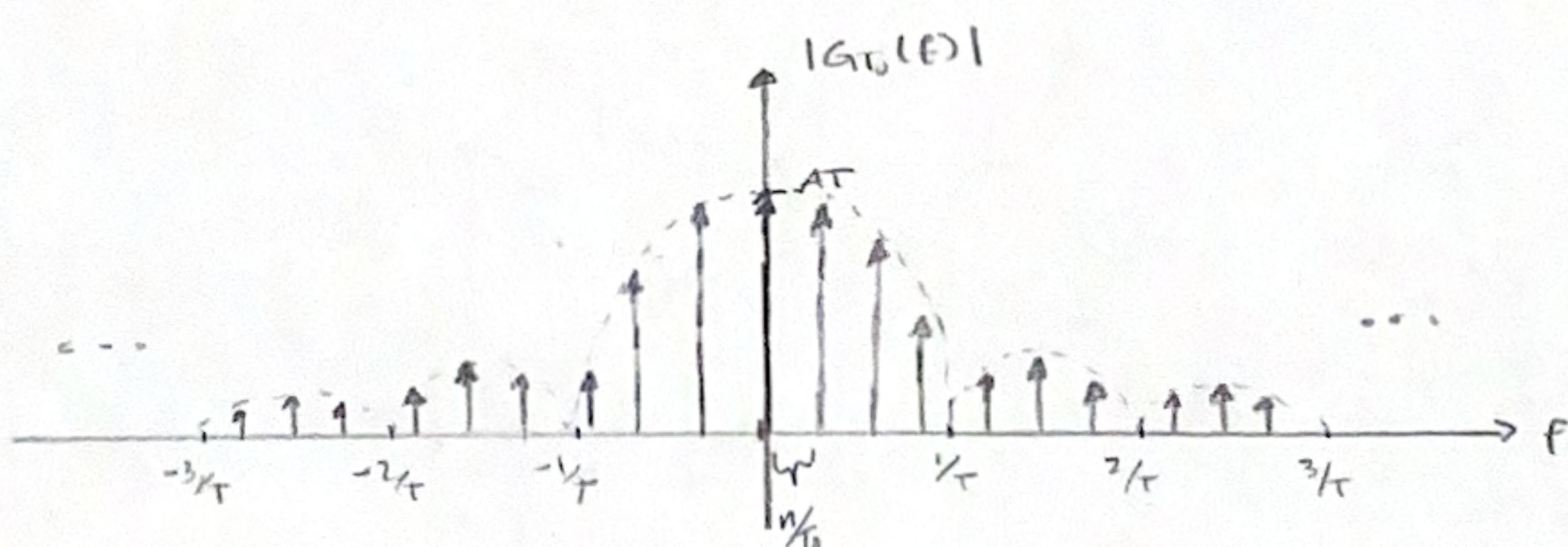
$$\Rightarrow \mathcal{F}[g(t)] = \mathcal{F}[\text{rect}(t/T)] = \frac{A}{T} \text{sinc}^2(f/T) = \boxed{AT \text{sinc}^2(Tf) = G(f)}$$



D) Find the F.T. of the periodic function $g_{T_0}(t)$ and plot its amplitude spectrum.

$$* \mathcal{F}[g_{T_0}(t)] = \frac{1}{T_0} \sum_{n=-\infty}^{\infty} G(n/T_0) \delta(f - n/T_0)$$

$$\Rightarrow \mathcal{F}[g_{T_0}(t)] = \boxed{\frac{1}{T_0} \sum_{n=-\infty}^{\infty} AT \text{sinc}^2(T \cdot n/T_0) \delta(f - n/T_0) = G_{T_0}(f)}$$



E) Insights?

PART C's plot shows the ENVELOPE over $G(f)$ follows.

THE FINAL F.T. TO GIVE $G_{T_0}(f)$ SHOWS THE TRAIN OF RECTANGULAR PULSES IN TIME-DOMAIN WITH PERIOD T_0 PRODUCES IN FREQ. DOMAIN A TRAIN OF IMPULSES WHICH FOLLOW THE $AT \text{sinc}^2(fT)$ ENVELOPE w/ $nT_0 = nf_0$ SPACING,

↑
Fundamental freq.