Forward Laplace Transform

* Recall from 2714 that the Etgen function for et systems is est for sec.

- et at some fixed value of s.
- * H(s) to the Bilateral haplace transform of the impulse response, also called two-sided transform.

the subset of the complex plane R is the set of values for 5 where the integral converges and the transform exists, we call this the region-of-conveyore or Roc.

- Not just the impulse response. Any styral
- · Notation = LEXCHZ = I(s) or XCH ZZ > I(s)

the forward Laplace transform.

- hats look at some illustrative examples

Example #1: $x(t) = e^{-at}$ of $a \in \mathbb{R}$ Find $\underline{x}(s) = \underline{J}\{x(t)\}$ $\underline{X}(s) = \begin{pmatrix} x(t)e^{-st}dt = \begin{pmatrix} -at-st \\ -at-st \end{pmatrix} = \begin{pmatrix} -(s+a)t \\ -(s+a)t \end{pmatrix}$

$$Z(s) = \int x(t)e^{st}dt = \int e^{at}e^{-st}dt = \int e^{-(s+a)t}dt$$

$$= \frac{-1}{s+a}e^{-(s+a)t}/\infty$$

If Re 25+a3>0 Re 253>-a
or Re 253+ Re 293>0

Note this result Jige at UCHIS = 1 12 853>-a
applies to any finite real a, thus

3 {UH3= 32 = 2 (H3) = 1 R= 853 > 0

That's compare the result in Example 1 to the CTFT definition of go

The definition of the CTFT is equivalent 8=3 w to Laplace definition with 5 restricted to purely imaginary values.

What are the implications of this? OF (Juta) t dt

JEEU LIES = Seat Edut dt = Se (Juta) t dt

= -1 [lim = (Juta) T - Guta

Juta [lim éconta)T - éconta)o]

This only conveyes for a>0.

Having a real part to 5 allows us to force the integral to converge (within limits), with ROC being the values that do so.

- Recall the CTFT 7 EXCHE exists if] x 61/dt <00

When will Laplace transform exist?

IF (| XH)et | dt coo for some Real C.

if x(+) grows as t > L, et can counter the growth.

Consider however X(+)=t=, it grows too fast.

Example 2: Non-caosal pulse X(+) =
$$o(t+1) - o(t-1)$$

 $X(s) = \int_{x(+)}^{x(+)} e^{st} dt = \int_{0}^{x(+)} e^{-st} dt = -\frac{1}{5} e^{-st$

- Note: The Japlace transform of a finite length

Signal where

SIX411 d+200 L= upper bound.

will exist with ROC = entire @ plane.

Example 3: $\chi(t) = e^{t}u(t) - e^{2t}u(t)$ $\Sigma(s) = \int_{e}^{\infty} e^{-t} dt + \int_{e}^{\infty} e^{-2t} e^{-st} dt$

= 1 S+1 S+2 R, = Re \(\frac{2}{2} \fra

= 3+2+5+1 with ROC = R, UR2 = RE 553>-1

= 1 (S+1)(S+2) = 32+35+2 Re {53>-1

- Note: If X(+) is causal then

JEXLEIZ= SXLESE dt and RPC= REESE> constant.

This is the unilateral or one-sided deplace transform.

When dealing with only causal stands (and thus systems) explicitly treating the ROC IS NOT required.

-1+1 Example 4: X(+)= elt

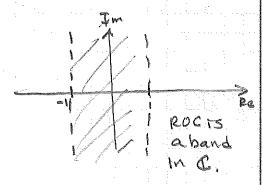
$$= \int_{-\infty}^{\infty} e^{t} e^{st} dt + \int_{-\infty}^{\infty} e^{t} e^{-st} dt \times (t) = \begin{cases} e^{t} & t < 0 \\ e^{-t} & t > 0 \end{cases}$$

$$= \int_{e}^{-\infty} e^{-(s-1)t} dt + \int_{e}^{\infty} e^{-(s+1)t} dt$$

Re 25-13 50

Re 253-120 Re£53 < 1

Re E 53 < 1 12c 553>-1



Note It we combine the expressions

$$\Sigma(s) = \frac{1}{s-1} + \frac{1}{s+1} = \frac{-s-1}{(s-1)(s+1)} = \frac{-2}{s^2-1}$$

we loose the distinction between anti-causal and causel, which is why Roc needs to be Kept in those cases. - 1 < Re 8 = 3 < 1

Lets do more examples,

Example 5: x(+)= 8(+) 0

I(s)= (x4) est d+= (s4) est d+= esta)=1 ROC = C

Example 6: $X(t) = e^{3\omega_0 t}$ $\omega_0 \in \mathbb{R}^t$ $\overline{X(s)} = \int_0^t e^{3\omega_0 t} e^{-5t} dt = \int_0^t e^{-(s-3\omega_0)t} dt$

= -(5-0'wo) + 100 5-0'wo -00

= = 1 - (5-6 mo)T - Lim
5-3-00 - (5-6 mo)T - Lim
+-00

Re[5-540] 60 Re {5-5-403>0 Refsiso No 53320

ROC does not exist,

hot 5=0, we get JEedwot 3 = 1 S(w-No)

Example 7: Compare to X(+) = & wot (+)

ICs) = Sesmot est dt = (=6-smo)t dt

-(5-3 wo)t/0

-(5-5'wo)T -(5-5'no)(0)

= -1 = 1 = 1 Re 853 > 0.

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Example 8: $X(t) = cos(cybt) u(t) = \frac{1}{2}e^{3mbt}u(t) + \frac{1}{2}e^{3mbt}u(t)$ $X(s) = \int_{0}^{\infty} X(t)e^{st} dt = \frac{1}{2}\int_{0}^{\infty} e^{3mbt} - st dt + \frac{1}{2}\int_{0}^{\infty} e^{3mot} - st dt$

Simila to previous example.

$$Z(s) = \frac{1}{2} \int_{0}^{\infty} \frac{-(s-iw_{0})t}{dt} dt + \frac{1}{2} \int_{0}^{\infty} \frac{-(s+iw_{0})t}{dt} dt$$

$$= \frac{1}{2} \frac{-1}{s-iw_{0}} \int_{0}^{\infty} \frac{-(s+iw_{0})t}{s+iw_{0}} dt$$

- Note the examples thus for cover many of the signals we are interested in.

$$S(s)$$
 $S(t)$
 $S(t)$

sm (upt) ult)

e cos(wot) u(t)

The PS for this week has you derive these.