1、基于简化条件,推导纵向质心运动方程简化公式

关键步骤:(1)速度坐标系下的动力学方程

$$\begin{split} m \begin{bmatrix} \dot{v} \\ v \dot{\theta} \cos \sigma \\ -v \dot{\sigma} \end{bmatrix} &= \mathbf{R}_x \big[-v \big] \Big((\mathbf{F}_P)_a + (\mathbf{F}_A)_a + m(\mathbf{g})_a - m(\omega_E)_a \times \Big((\omega_E)_a \times (\mathbf{r})_a \Big) - 2m(\omega_E)_a \times (\mathbf{v}_l)_a \Big) \\ &= F_P \mathbf{R}_x \big[-v \big] \mathbf{R}_{ab} \begin{bmatrix} \cos \left(\delta_\psi \right) \cos \left(\delta_\varphi \right) \\ \sin \left(\delta_\varphi \right) \\ -\cos \left(\delta_\varphi \right) \sin \left(\delta_\psi \right) \end{bmatrix} + \mathbf{R}_x \big[-v \big] \begin{bmatrix} -C_D q S_M \\ C_L q S_M \\ C_C q S_M \end{bmatrix} + m \mathbf{R}_x \big[-v \big] \mathbf{R}_{al} \begin{bmatrix} \mathbf{g}_r - \mathbf{g}_\phi \tan \phi \\ \mathbf{r} \end{bmatrix} \begin{bmatrix} r_{olx} + r_{lx} \\ r_{oly} + r_{ly} \\ r_{olz} + r_{lz} \end{bmatrix} + \frac{\mathbf{g}_\phi}{\omega_E \cos \phi} \begin{bmatrix} \omega_{Elx} \\ \omega_{Elx} \end{bmatrix} \\ -m \mathbf{R}_x \big[-v \big] \mathbf{R}_{al} \begin{bmatrix} \omega_{Elx}^2 - \omega_E^2 & \omega_{Elx} \omega_{Ely} & \omega_{Elx} \omega_{Elz} \\ \omega_{Elx} \omega_{Ely} & \omega_E^2 - \omega_E^2 & \omega_{Ely} \omega_{Elz} \\ \omega_{Elx} \omega_{Elz} & \omega_{Ely} \omega_{Elz} \end{bmatrix} \begin{bmatrix} r_{olx} + r_{lx} \\ r_{oly} + r_{ly} \\ r_{olz} + r_{lz} \end{bmatrix} - 2m \mathbf{R}_x \big[-v \big] \mathbf{R}_{al} \begin{bmatrix} 0 & -\omega_{Elz} & \omega_{Ely} \\ \omega_{Elz} & 0 & -\omega_{Elx} \\ \omega_{Elx} & 0 & -\omega_{Elx} \\ \omega_{Elx} \omega_{Elz} & \omega_{Ely} \omega_{Elz} \end{bmatrix} \begin{bmatrix} \dot{x} \\ \dot{y} \\ \dot{z} \end{bmatrix} \end{split}$$

- (2) 引力假设:
- (a) 地球视为均质球体,忽略地球扁率及引力分量 g_{ϕ} 的影响,且服从平方反比定律
 - (b) 引力加速度只有沿 y 轴的分量
 - (3) 地球自转假设: 忽略地球旋转的影响, 即忽略哥式加速度和牵连加速度
 - (4) 小角度假设:
- (a) 欧拉角以欧拉角 α , β , γ , ψ , σ , ν , θ φ 以及控制量 δ_{φ} , δ_{ψ} 均为小量,正弦取其角度,余弦取为 1;
 - (b) 出现这些角度值之间的乘积时,作为二阶以上项略去
- (5) 力矩瞬时平衡假设
- (6) 简化形式

$$\dot{v} = \frac{F_P - D}{m} + g \sin(\theta)$$

$$\dot{\theta} = \frac{\left(F_P + CY^\alpha\right)\alpha}{mv} + \frac{g}{v}\cos(\theta)$$

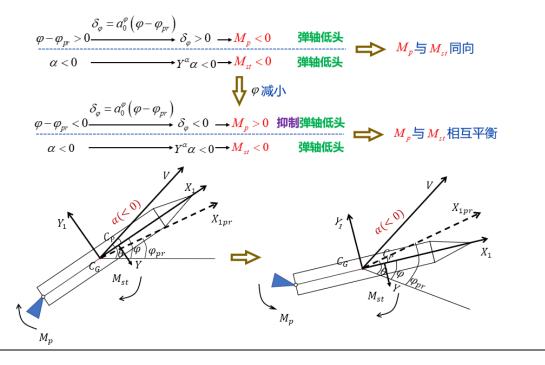
$$\dot{x} = v\cos(\theta)$$

$$\dot{y} = v\sin(\theta)$$

$$m = m_0 - \dot{m}t$$

2、画出纵向平面内的静不稳定火箭大气层内转弯受力图,并分析静不稳定火箭的(转弯过程)"三轴"变化规律。

关键步骤:



- (1) 注意程序段转弯的实际情况
- (2) 注意平衡的动态过程

3、推导关机方程与各变量的偏导关系 关键步骤:

$$\Delta L[X(t_k), t_k] = \dot{L}[\overline{X}(\overline{t_k}), \overline{t_k}](t_k - \overline{t_k}) + \delta L[X(t_k), t_k]$$

$$\delta L[X(t_k), t_k] \approx \frac{\partial L}{\partial X}\Big|_{\overline{t_k}} \delta X(t_k) + \frac{1}{2}\delta X(t_k)^T \frac{\partial^2 L}{\partial X^2}\Big|_{\overline{t_k}} \delta X(t_k)$$

$$\frac{\partial L}{\partial X_{\theta}} = \begin{bmatrix} \frac{\partial L}{\partial r_x} & \frac{\partial L}{\partial r_y} & \frac{\partial L}{\partial r_z} & \frac{\partial L}{\partial V} & \frac{\partial L}{\partial \theta} \end{bmatrix}^T$$

$$\frac{\partial^2 L}{\partial X_{\theta}^2} = \begin{bmatrix} L_{r,r_x} & L_{r,r_y} & L_{r,r_z} & L_{r,V} & L_{r,\theta} \\ L_{r,r_x} & L_{r,r_y} & L_{r,r_z} & L_{r,V} & L_{r,\theta} \\ L_{v_{r_x}} & L_{v_{r_y}} & L_{v_{r_z}} & L_{v_V} & L_{v_{\theta}} \\ L_{v_{r_x}} & L_{v_{r_y}} & L_{v_{r_z}} & L_{v_V} & L_{v_{\theta}} \end{bmatrix}$$

$$L_{r_xr_x} = \frac{\partial^2 L}{\partial r_x^2}$$

$$L_{r_xv_y} = L_{r_yr_x} = \frac{\partial^2 L}{\partial r_x \partial r_y}$$

$$L_{r_xv_y} = L_{v_{r_x}} = \frac{\partial^2 L}{\partial V \partial r_x}$$

$$L_{r_x\theta} = L_{\theta r_x} = \frac{\partial^2 L}{\partial V \partial \theta}$$

$$L_{VV} = \frac{\partial^2 L}{\partial \theta^2}$$

$$L_{V\theta} = L_{\theta V} = \frac{\partial^2 L}{\partial V \partial \theta}$$

对射程偏差系数的计算和分析表明,在二阶射程偏差系数中,以下两个偏导数对落点射程偏差影响最大:

$$\frac{\partial^2 L}{\partial \theta^2} \quad \frac{\partial^2 L}{\partial V \partial \theta}$$

- 4、线性化动力学方程(发惯系和速度系下) 关键步骤:
- (1) 发惯系下:
- (a) 列写实际状态方程

$$\begin{split} \dot{r}_x &= v_x \\ \dot{r}_y &= v_y \\ \dot{r}_z &= v_z \\ \dot{v}_x &= \frac{1}{m} \cos \left(\psi_L \right) \cos \left(\varphi_L \right) F_p + \frac{1}{m} F_{Ax} + g_x \\ \dot{v}_y &= \frac{1}{m} \cos \left(\psi_L \right) \sin \left(\varphi_L \right) F_p + \frac{1}{m} F_{Ay} + g_y \\ \dot{v}_z &= -\frac{1}{m} \sin \left(\psi_L \right) F_p + \frac{1}{m} F_{Az} + g_z \end{split}$$

(b) 列写标称状态方程

$$\begin{split} &\dot{\overline{r}_x} = \overline{v}_x \\ &\dot{\overline{r}_y} = \overline{v}_y \\ &\dot{\overline{r}_z} = \overline{v}_z \\ &\dot{\overline{v}_x} = \frac{1}{m} \cos\left(\overline{\psi}_L\right) \cos\left(\overline{\varphi}_L\right) \overline{F}_p + \overline{g}_x + \overline{f}_x \\ &\dot{\overline{v}_y} = \frac{1}{m} \cos\left(\overline{\psi}_L\right) \sin\left(\overline{\varphi}_L\right) \overline{F}_p + \overline{g}_y + \overline{f}_y \\ &\dot{\overline{v}_z} = -\frac{1}{m} \sin\left(\overline{\psi}_L\right) \overline{F}_p + \overline{g}_z + \overline{f}_z \end{split}$$

$$\begin{bmatrix} \overline{f}_x & \overline{f}_y & \overline{f}_z \end{bmatrix}^{\mathrm{T}} = \begin{bmatrix} \frac{1}{m} \overline{F}_{Ax} & \frac{1}{m} \overline{F}_{Ay} & \frac{1}{m} \overline{F}_{Az} \end{bmatrix}^{\mathrm{T}}$$

(c)实际状态方程与标称状态方程相减,并经过泰勒一阶展开,得到线性化摄动方程

$$\begin{split} \delta \dot{r}_{x} &= \delta v_{x} \\ \delta \dot{r}_{y} &= \delta v_{y} \\ \delta \dot{r}_{z} &= \delta v_{z} \\ \delta \dot{v}_{x} &= \frac{\partial g_{x}}{\partial r_{x}} \delta r_{x} + \frac{\partial g_{x}}{\partial r_{y}} \delta r_{y} + \frac{\partial g_{x}}{\partial r_{z}} \delta r_{z} - \frac{F_{p}}{m} \cos \left(\overline{\psi}_{L}\right) \sin \left(\overline{\varphi}_{L}\right) \delta \varphi_{L} - \frac{F_{p}}{m} \sin \left(\overline{\psi}_{L}\right) \cos \left(\overline{\varphi}_{L}\right) \delta \psi_{L} + f_{x} \\ \delta \dot{v}_{y} &= \frac{\partial g_{y}}{\partial r_{x}} \delta r_{x} + \frac{\partial g_{y}}{\partial r_{y}} \delta r_{y} + \frac{\partial g_{y}}{\partial r_{z}} \delta r_{z} + \frac{F_{p}}{m} \cos \left(\overline{\psi}_{L}\right) \cos \left(\overline{\varphi}_{L}\right) \delta \varphi_{L} - \frac{F_{p}}{m} \sin \left(\overline{\psi}_{L}\right) \sin \left(\overline{\varphi}_{L}\right) \delta \psi_{L} + f_{y} \\ \delta \dot{v}_{z} &= \frac{\partial g_{z}}{\partial r_{x}} \delta r_{x} + \frac{\partial g_{z}}{\partial r_{y}} \delta r_{y} + \frac{\partial g_{z}}{\partial r_{z}} \delta r_{z} - \frac{F_{p}}{m} \cos \left(\overline{\psi}_{L}\right) \delta \psi_{L} + f_{z} \end{split}$$

(d) 列写为紧缩形势可以得到

$$\delta \dot{X}(t) = A(t)\delta X(t) + B(t)\delta U(t) + C(t)f(t)$$

$$\mathbf{A}(t) = \begin{bmatrix} 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \\ \frac{\partial g_x}{\partial r_x} & \frac{\partial g_x}{\partial r_y} & \frac{\partial g_x}{\partial r_z} & 0 & 0 & 0 \\ \frac{\partial g_y}{\partial r_x} & \frac{\partial g_y}{\partial r_y} & \frac{\partial g_y}{\partial r_z} & 0 & 0 & 0 \\ \frac{\partial g_z}{\partial r_x} & \frac{\partial g_z}{\partial r_y} & \frac{\partial g_z}{\partial r_z} & 0 & 0 & 0 \end{bmatrix}$$

$$\boldsymbol{B}(t) = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ -\frac{F_p}{m}\cos(\overline{\psi}_L)\sin(\overline{\varphi}_L) & -\frac{F_p}{m}\sin(\overline{\psi}_L)\cos(\overline{\varphi}_L) \\ \frac{F_p}{m}\cos(\overline{\psi}_L)\cos(\overline{\varphi}_L) & -\frac{F_p}{m}\sin(\overline{\psi}_L)\sin(\overline{\varphi}_L) \\ 0 & -\frac{F_p}{m}\cos(\overline{\psi}_L) \end{bmatrix}$$

$$C = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

(2) 速度系下推导过程与惯性系下相同

$$\dot{v} = \frac{F_P - D}{m} + g \sin \theta$$

$$\dot{\theta} = \frac{\left(F_P + CY^{\alpha}\right)\alpha}{mv} + \frac{g}{v}\cos(\theta) = \frac{F_P\left(\delta_{\varphi} + \alpha\right) + C_L q S_M}{mv} + \frac{g}{v}\cos\theta$$

$$\dot{y} = v \sin \theta$$

$$\dot{x} = v \cos\theta$$

$$\alpha = A\left(\varphi_{pr} - \theta\right)$$

$$\dot{\overline{v}} = \frac{F_P}{m} + \overline{g} \sin \overline{\theta}$$

$$\dot{\overline{\theta}} = \frac{1}{m\overline{v}} F_P \left(\overline{\delta}_{\varphi} + \overline{\alpha} \right) + \frac{\overline{g}}{\overline{v}} \cos \overline{\theta}$$

$$\dot{\overline{y}} = \overline{v} \sin \overline{\theta}$$

$$\dot{\overline{x}} = \overline{v} \cos \overline{\theta}$$

$$\overline{\alpha} = A \left(\varphi_{pr} - \overline{\theta} \right)$$

$$\begin{split} \delta \dot{v} &= \left[\frac{\partial g_x}{\partial x} \delta x + \frac{\partial g_x}{\partial y} \delta y \right] \sin \overline{\theta} + \overline{g} \cos \overline{\theta} \delta \theta + f_x \\ \delta \dot{\theta} &= \left[\frac{\partial g_x}{\partial x} \delta x + \frac{\partial g_x}{\partial y} \delta y \right] \cos \overline{\theta} - \left(\frac{\overline{g}}{\overline{v}} \sin \overline{\theta} + \frac{F_P}{m\overline{v}} A \right) \delta \theta - \frac{1}{\overline{v}^2} \left(\overline{g} \cos \overline{\theta} + \frac{F_P \left(\overline{\delta_{\phi}} + A \left(\phi_{pr} - \overline{\theta} \right) \right)}{m} \right) \delta v + \frac{F_P}{m\overline{v}} \delta (\delta_{\phi}) + f_y \\ \delta \dot{y} &= \overline{v} \cos \overline{\theta} \delta \theta + \sin \overline{\theta} \delta v \end{split}$$

 $\delta \dot{x} = -\overline{v}\sin\overline{\theta}\delta\theta + \cos\overline{\theta}\delta v$

注意:1. 在制导过程中,认为控制参数为姿态角 δ_{φ} ,而非攻角 α (作业中均判正确,计算无误即可,但在考试中要注意准确性)。

2. 干扰力部分不可省略;

5、分析横向导引方程的原理

关键步骤: (1) 推导落点横向偏差

$$\Delta L[X(t_k), t_k] = \dot{L}\left[\overline{X}(\overline{t_k}), \overline{t_k}\right] \Delta t_k + \frac{\partial L}{\partial X}\Big|_{\overline{t_k}} \delta X(t_k) = 0$$

$$\Delta H[X(t_k), t_k] = \dot{H}\left[\overline{X}(\overline{t_k}), \overline{t_k}\right] \Delta t_k + \frac{\partial H}{\partial X}\Big|_{\overline{t_k}} \delta X(t_k)$$

$$\Delta t_k = -\frac{1}{\dot{L}\left[\overline{X}(\overline{t_k}), \overline{t_k}\right]} \frac{\partial L}{\partial X}\Big|_{\overline{t_k}} \delta X(t_k)$$

$$\Delta H[X(t_k), t_k] = \left[\frac{\partial H}{\partial X}\Big|_{\overline{t_k}} - \frac{\dot{H}\left[\overline{X}(\overline{t_k}), \overline{t_k}\right]}{\dot{L}\left[\overline{X}(\overline{t_k}), \overline{t_k}\right]} \frac{\partial L}{\partial X}\Big|_{\overline{t_k}}\right] \delta X(t_k) = 0$$

(2) 预测关机时间偏差

$$\begin{split} \Delta L \big[\boldsymbol{X}(t_k), t_k \big] &= \dot{L} \Big[\, \overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \, \Big] \Delta t_k + \delta L \big[\boldsymbol{X}(t_k), t_k \big] \\ &= \dot{L} \Big[\, \overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \, \Big] \Delta t_k - \boldsymbol{\Lambda}_{s2}^T(t) \delta \boldsymbol{X}(t) \Delta t_k + \boldsymbol{\Lambda}_{s1}^T(t) \delta \boldsymbol{X}(t) \\ \Delta t_k &= - \frac{\boldsymbol{\Lambda}_{s1}^T(t)}{\dot{L} \big[\, \overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \, \Big] - \boldsymbol{\Lambda}_{s2}^T(t) \delta \boldsymbol{X}(t)} \delta \boldsymbol{X}(t) \end{split}$$

(3) 预测落点横向偏差

$$\Delta H \left[X(t_k), t_k \right] = \left[\frac{\partial H}{\partial X} \Big|_{\overline{t_k}} - \frac{\dot{H} \left[\overline{X}(\overline{t_k}), \overline{t_k} \right]}{\dot{L} \left[\overline{X}(\overline{t_k}), \overline{t_k} \right]} \frac{\partial L}{\partial X} \Big|_{\overline{t_k}} \right] \delta X(t_k) = 0$$

$$\Lambda_h(t_k) = \left[\frac{\partial H}{\partial X} \Big|_{\overline{t_k}} - \frac{\dot{H} \left[\overline{X}(\overline{t_k}), \overline{t_k} \right]}{\dot{L} \left[\overline{X}(\overline{t_k}), \overline{t_k} \right]} \frac{\partial L}{\partial X} \Big|_{\overline{t_k}} \right]^{\mathrm{T}}$$

$$\Lambda_h(t) = \Lambda_{h1}(t) - \Lambda_{h2}(t) \Delta t_k$$

$$\Delta t_k = -\frac{\Lambda_{s1}^T(t)}{\dot{L} \left[\overline{X}(\overline{t_k}), \overline{t_k} \right] - \Lambda_{s2}^T(t) \delta X(t)} \delta X(t)$$

(4) 横向导引策略

$$\Delta H \left[\boldsymbol{X}(t_k), t_k \right] = \boldsymbol{\Lambda}_n^{\mathrm{T}}(t_k) \delta \boldsymbol{X}(t_k) = \boldsymbol{\Lambda}_n^{\mathrm{T}}(t) \delta \boldsymbol{X}(t)$$
$$\delta \psi_L(t) = \boldsymbol{K}_{\psi} \Delta H$$

注意:作业中此处还没有讲到伴随状态量的求解,故作业中不做要求,但考试中需要完整做答。

- 6. 推导使用导航信息的摄动制导算法(关机方程,法向导引) 关键步骤:
 - (1) 射程关机方程为:

$$\Delta L[X(t_k), t_k] = \dot{L}[\overline{X}(\overline{t_k}), \overline{t_k}] \Delta t_k + \frac{\partial L}{\partial X}\Big|_{\overline{t_k}} \delta X(t_k) = 0$$

(2) 定义伴随系统为:

$$\begin{split} \dot{\boldsymbol{\Lambda}}_{s}(t) &= -\boldsymbol{A}^{\mathrm{T}}(t)\boldsymbol{\Lambda}_{s}(t) \\ \boldsymbol{\Lambda}_{s}(t_{k}) &= \left(\frac{\partial L}{\partial \boldsymbol{X}}\Big|_{\overline{t_{k}}}\right)^{\mathrm{T}} = \left[\frac{\partial L}{\partial r_{x}}\Big|_{\overline{t_{k}}} \quad \frac{\partial L}{\partial r_{y}}\Big|_{\overline{t_{k}}} \quad \frac{\partial L}{\partial r_{z}}\Big|_{\overline{t_{k}}} \quad \frac{\partial L}{\partial v_{x}}\Big|_{\overline{t_{k}}} \quad \frac{\partial L}{\partial v_{y}}\Big|_{\overline{t_{k}}} \quad \frac{\partial L}{\partial v_{z}}\Big|_{\overline{t_{k}}} \right]^{\mathrm{T}} \end{split}$$

解得:

$$\Lambda_s(t) = \Phi(t, t_k) \Lambda_s(t_k)_k$$

(3) 泰勒展开获得:

$$\begin{split} \boldsymbol{\Lambda}_{s}(t) &= \boldsymbol{\Phi}\left(t, t_{k}\right) \boldsymbol{\Lambda}_{s}(t_{k}) \\ &= \left(\boldsymbol{\Phi}\left(t, \overline{t_{k}}\right) + \frac{d\boldsymbol{\Phi}\left(t, t_{k}\right)}{dt_{k}} \bigg|_{t_{k} = \overline{t_{k}}} \Delta t_{k}\right) \boldsymbol{\Lambda}_{s}(t_{k}) \\ &= \boldsymbol{\Phi}\left(t, \overline{t_{k}}\right) \boldsymbol{\Lambda}_{s}(t_{k}) - \left(-\boldsymbol{\Phi}\left(t, \overline{t_{k}}\right) \boldsymbol{A}^{\mathrm{T}}(\overline{t_{k}}) \boldsymbol{\Lambda}_{s}(t_{k})\right) \Delta t_{k} \end{split}$$

(4) 将 $\Lambda_s(t)$ 近似为两部分求解

$$\begin{split} \boldsymbol{\Lambda}_{s1}(t) &= \boldsymbol{\Lambda}_{s1}(t) - \boldsymbol{\Lambda}_{s2}(t) \Delta t_k \\ \boldsymbol{\Lambda}_{s1}(t) &= \boldsymbol{\Phi}\left(t, \overline{t_k}\right) \boldsymbol{\Lambda}_{s}(t_k) \\ \boldsymbol{\Lambda}_{s1}(\overline{t_k}) &= \boldsymbol{\Lambda}_{s}(t_k) \\ \boldsymbol{\Lambda}_{s2}(t) &= -\boldsymbol{\Phi}\left(t, \overline{t_k}\right) \boldsymbol{\Lambda}^{\mathrm{T}}(\overline{t_k}) \boldsymbol{\Lambda}_{s}(t_k) = -\boldsymbol{\Phi}\left(t, \overline{t_k}\right) \boldsymbol{\Lambda}^{\mathrm{T}}(\overline{t_k}) \boldsymbol{\Lambda}_{s1}(\overline{t_k}) = \boldsymbol{\Phi}\left(t, \overline{t_k}\right) \dot{\boldsymbol{\Lambda}}_{s1}(\overline{t_k}) \\ \boldsymbol{\Lambda}_{s2}(\overline{t_k}) &= \dot{\boldsymbol{\Lambda}}_{s1}(\overline{t_k}) \end{split}$$

(5) 得到关机时间偏差

$$\begin{split} \Delta L \big[\boldsymbol{X}(t_k), t_k \big] &= \dot{L} \Big[\, \overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \, \Big] \Delta t_k + \delta L \big[\boldsymbol{X}(t_k), \overline{t_k} \, \Big] \\ &= \dot{L} \Big[\, \overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \, \Big] \Delta t_k - \boldsymbol{\Lambda}_{s2}^T(t) \delta \boldsymbol{X}(t) \Delta t_k + \boldsymbol{\Lambda}_{s1}^T(t) \delta \boldsymbol{X}(t) \\ &= 0 \end{split}$$

$$\Delta t_k = -\frac{\boldsymbol{\Lambda}_{s1}^T(t)}{\dot{L} \left[\overline{\boldsymbol{X}}(\overline{t_k}), \overline{t_k} \right] - \boldsymbol{\Lambda}_{s2}^T(t) \delta \boldsymbol{X}(t)} \delta \boldsymbol{X}(t)$$

(6) 推导关机时刻速度倾角偏差

$$\Delta\theta \left[\boldsymbol{X}(t_{k}), t_{k} \right] = \left[\frac{\partial\theta}{\partial\boldsymbol{X}} \Big|_{\overline{t_{k}}} - \frac{\dot{\theta} \left[\boldsymbol{\overline{X}}(\overline{t_{k}}), \overline{t_{k}} \right]}{\dot{L} \left[\boldsymbol{\overline{X}}(\overline{t_{k}}), \overline{t_{k}} \right]} \frac{\partial L}{\partial \boldsymbol{X}} \Big|_{\overline{t_{k}}} \right] \delta\boldsymbol{X}(t_{k})$$

$$\boldsymbol{\Lambda}_{n}(t_{k}) = \left[\frac{\partial\theta}{\partial\boldsymbol{X}} \Big|_{\overline{t_{k}}} - \frac{\dot{\theta} \left[\boldsymbol{\overline{X}}(\overline{t_{k}}), \overline{t_{k}} \right]}{\dot{L} \left[\boldsymbol{\overline{X}}(\overline{t_{k}}), \overline{t_{k}} \right]} \frac{\partial L}{\partial \boldsymbol{X}} \Big|_{\overline{t_{k}}} \right]^{\mathrm{T}}$$

$$\boldsymbol{\Lambda}_{n}(t) = \boldsymbol{\Lambda}_{n1}(t) - \boldsymbol{\Lambda}_{n2}(t) \Delta t_{k}$$

$$\Delta\theta \left[\boldsymbol{X}(t_{k}), t_{k} \right] = \boldsymbol{\Lambda}_{n}^{\mathrm{T}}(t_{k}) \delta\boldsymbol{X}(t_{k}) = \boldsymbol{\Lambda}_{n}^{\mathrm{T}}(t) \delta\boldsymbol{X}(t)$$

(7) 选择控制律并设计控制方程

$$\delta \varphi_L(t) = \pmb{K}_{\varphi} \Delta \theta$$

注意: 要求推导过程流畅严谨,逻辑清晰(不可有缺漏或跳跃)。

7. 在无导航信息解算条件下,推导使用惯性平台的摄动制导算法(关机方程, 法向导引)

关键步骤:

(1) 构建全量线性系统

$$\begin{split} \dot{r}_x &= v_x \\ \dot{r}_y &= v_y \\ \dot{r}_z &= v_z \\ \dot{v}_x &= \frac{\partial g_x}{\partial r_x} r_x + \frac{\partial g_x}{\partial r_y} r_y + \frac{\partial g_x}{\partial r_z} r_z + a_{mx} + \tilde{g}_x \\ \dot{v}_y &= \frac{\partial g_y}{\partial r_x} r_x + \frac{\partial g_y}{\partial r_y} r_y + \frac{\partial g_y}{\partial r_z} r_z + a_{my} + \tilde{g}_y \\ \dot{v}_z &= \frac{\partial g_z}{\partial r_x} r_x + \frac{\partial g_z}{\partial r_y} r_y + \frac{\partial g_z}{\partial r_z} r_z + a_{mz} + \tilde{g}_z \\ \tilde{g}_x &= \overline{g}_x - \frac{\partial g_x}{\partial r_x} \overline{r}_x - \frac{\partial g_x}{\partial r_y} \overline{r}_y - \frac{\partial g_x}{\partial r_z} \overline{r}_z \\ \tilde{g}_y &= \overline{g}_y - \frac{\partial g_y}{\partial r_x} \overline{r}_x - \frac{\partial g_y}{\partial r_y} \overline{r}_y - \frac{\partial g_y}{\partial r_z} \overline{r}_z \\ \tilde{g}_z &= \overline{g}_z - \frac{\partial g_z}{\partial r_x} \overline{r}_x - \frac{\partial g_z}{\partial r_y} \overline{r}_y - \frac{\partial g_z}{\partial r_z} \overline{r}_z \\ \tilde{g}_z &= \overline{g}_z - \frac{\partial g_z}{\partial r_x} \overline{r}_x - \frac{\partial g_z}{\partial r_y} \overline{r}_y - \frac{\partial g_z}{\partial r_z} \overline{r}_z \end{split}$$

定义状态量为:

$$\boldsymbol{X} = \begin{bmatrix} r_x & r_y & r_z & v_x & v_y & v_z \end{bmatrix}^{\mathrm{T}}$$

定义控制量为:

$$\boldsymbol{U} = \begin{bmatrix} a_{mx} + \tilde{g}_x & a_{my} + \tilde{g}_y & a_{mz} + \tilde{g}_z \end{bmatrix}^{\mathrm{T}}$$

(2) 基于无导航情况下的 ΔL 求解关机时间偏差 Δt_{k}

参照 PPT 得到

$$\Delta t_{k} \approx \frac{\frac{\partial L}{\partial t_{k}} \bigg|_{\overline{t_{k}}} - \boldsymbol{\Lambda}_{s1}^{T}(0)\boldsymbol{X}(0) - \int_{0}^{t} \boldsymbol{\Lambda}_{s1}^{T}(\tau)\boldsymbol{B}(\tau)\boldsymbol{U}(\tau) d\tau - \int_{t}^{\overline{t_{k}}} \boldsymbol{\Lambda}_{s1}^{T}(\tau)\boldsymbol{B}(\tau)\overline{\boldsymbol{U}}(\tau) d\tau}{\frac{\partial L}{\partial t_{k}} \bigg|_{\overline{t_{k}}} - \boldsymbol{\Lambda}_{s2}^{T}(0)\boldsymbol{X}(0) - \int_{0}^{t} \boldsymbol{\Lambda}_{s2}^{T}(\tau)\boldsymbol{B}(\tau)\boldsymbol{U}(\tau) d\tau - \int_{t}^{\overline{t_{k}}} \boldsymbol{\Lambda}_{s2}^{T}(\tau)\boldsymbol{B}(\tau)\overline{\boldsymbol{U}}(\tau) d\tau}$$

(3) 干扰抑制原理的应用

在标称轨迹附近,将质心运动方程进行一阶线性展开,得到:

$$\begin{split} \delta \dot{r}_x &= \delta v_x \\ \delta \dot{r}_y &= \delta v_y \\ \delta \dot{r}_z &= \delta v_z \\ \delta \dot{v}_x &= \frac{\partial g_x}{\partial r_x} \delta r_x + \frac{\partial g_x}{\partial r_y} \delta r_y + \frac{\partial g_x}{\partial r_z} \delta r_z - \frac{F_p}{m} \cos(\overline{\psi}_L) \sin(\overline{\phi}_L) \delta \phi_L - \frac{F_p}{m} \sin(\overline{\psi}_L) \cos(\overline{\phi}_L) \delta \psi_L + a_{mx} - \frac{1}{m} \cos(\psi_L) \cos(\phi_L) F_p \\ \delta \dot{v}_y &= \frac{\partial g_y}{\partial r_x} \delta r_x + \frac{\partial g_y}{\partial r_y} \delta r_y + \frac{\partial g_y}{\partial r_z} \delta r_z + \frac{F_p}{m} \cos(\overline{\psi}_L) \cos(\overline{\phi}_L) \delta \phi_L - \frac{F_p}{m} \sin(\overline{\psi}_L) \sin(\overline{\phi}_L) \delta \psi_L + a_{my} - \frac{1}{m} \cos(\psi_L) \sin(\phi_L) F_p \\ \delta \dot{v}_z &= \frac{\partial g_z}{\partial r_x} \delta r_x + \frac{\partial g_z}{\partial r_y} \delta r_y + \frac{\partial g_z}{\partial r_z} \delta r_z - \frac{F_p}{m} \cos(\overline{\psi}_L) \delta \psi_L + a_{mz} + \frac{1}{m} \sin(\psi_L) F_p \end{split}$$

干扰力部分为:

$$f_x = \frac{1}{m} F_{Ax} = a_{mx} - \frac{1}{m} \cos(\psi_L) \cos(\varphi_L) F_p$$

$$f_y = \frac{1}{m} F_{Ay} = a_{my} - \frac{1}{m} \cos(\psi_L) \sin(\varphi_L) F_p$$

$$f_z = \frac{1}{m} F_{Az} = a_{mz} + \frac{1}{m} \sin(\psi_L) F_p$$

根据干扰抑制定理得到:

$$\begin{split} \delta\varphi_L(t) &= -\Big[\mathbf{\Lambda}_n^{\mathsf{T}}(t)\mathbf{B}_n(t)\Big]^{-1}\mathbf{\Lambda}_n^{\mathsf{T}}(t)\mathbf{C}_n(t)\mathbf{f}(t) \\ \delta\varphi_L(t) &= -\Big[-\Lambda_{n4}(t)\frac{F_p}{m}\cos(\overline{\psi}_L)\sin(\overline{\varphi}_L) - \Lambda_{n5}(t)\frac{F_p}{m}\cos(\overline{\psi}_L)\cos(\overline{\varphi}_L)\Big]^{-1} \times \\ & \left[\Lambda_{n4}(t)\Big(a_{mx} - \frac{1}{m}\cos(\psi_L)\cos(\varphi_L)F_p\Big) \right. \\ & \left. + \Lambda_{n5}(t)\Big(a_{my} - \frac{1}{m}\cos(\psi_L)\sin(\varphi_L)F_p\Big) \right. \\ & \left. + \Lambda_{n6}(t)\Big(a_{mz} + \frac{1}{m}\sin(\psi_L)F_p\Big)\right] \\ & \delta\psi_L(t) &= -\Big[\mathbf{\Lambda}_h^{\mathsf{T}}(t)\mathbf{B}_h(t)\Big]^{-1}\mathbf{\Lambda}_h^{\mathsf{T}}(t)\mathbf{C}_h(t)\mathbf{f}(t) \\ \delta\psi_L(t) &= -\Big[-\Lambda_{n4}(t)\frac{F_p}{m}\sin(\overline{\psi}_L)\cos(\overline{\varphi}_L) - \Lambda_{n5}(t)\frac{F_p}{m}\sin(\overline{\psi}_L)\sin(\overline{\varphi}_L) - \Lambda_{n6}(t)\frac{F_p}{m}\cos(\overline{\psi}_L)\Big]^{-1} \times \\ & \left. [\Lambda_{n4}(t)\Big(a_{mx} - \frac{1}{m}\cos(\psi_L)\cos(\varphi_L)F_p\Big) \right. \\ & \left. + \Lambda_{n5}(t)\Big(a_{mz} + \frac{1}{m}\sin(\psi_L)F_p\Big)\right] \end{split}$$

注意:要求推导过程流畅严谨,逻辑清晰(不可有缺漏或跳跃)。

8. 推导满足下面条件的显式制导算法(要求写出三通道制导指令规律)

已知: 1) 当前时刻 t_0 的位置和速度: $x(t_0) = x_0$, $\dot{x}(t_0) = \dot{x}_0$

- 2) 以推力加速度为控制量: $\ddot{x}(t) = g(r) + a_P(t)$
- 3) 终端约束包括: 终端位置、终端速度与终端加速度 终端 $t_{\rm f}$ 时刻,到达目标位置和速度: ${m x}(t_{\rm f})={m x}_{\rm f}$, ${\dot {m x}}(t_{\rm f})={\dot {m x}}_{\rm f}$

终端
$$t_f$$
时刻,推力方向:
$$\frac{\boldsymbol{a}_{Px}(t_f)}{|\boldsymbol{a}_P(t_f)|} = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}^T$$

关键步骤:

(1) 对于 y 轴和 z 轴,终端约束包括位置,速度和加速度:

以 y 轴为例, (a) 构建多项式:

$$\ddot{y}(t) = c_{y0} + c_{y1}(t_f - t) + c_{y2}(t_f - t)^2$$
$$c_{y0} = \ddot{y}_f = g_y(t_f)$$

(b) 积分计算

$$\dot{y}_{f} - \dot{y}_{0} = \int_{t_{0}}^{t_{f}} \ddot{y}(t) dt = c_{y0} t_{go} + \frac{1}{2} c_{y1} t_{go}^{2} + \frac{1}{3} c_{y2} t_{go}^{3}$$

$$y_{f} - y_{0} - \dot{y}_{0} t_{go} = \int_{t_{0}}^{t_{f}} \left[\int_{t_{0}}^{t} \ddot{y}(s) ds \right] dt = \frac{1}{2} c_{y0} t_{go}^{2} + \frac{1}{3} c_{y1} t_{go}^{3} + \frac{1}{4} c_{y2} t_{go}^{4}$$

(c) 反解 E 矩阵得到参数 c_1 和 c_2

$$\dot{y}_{f} - \dot{y}_{0} - c_{y0}t_{go} = \frac{1}{2}c_{y1}t_{go}^{2} + \frac{1}{3}c_{y2}t_{go}^{3}$$

$$y_{f} - y_{0} - \dot{y}_{0}t_{go} - \frac{1}{2}c_{y0}t_{go}^{2} = \frac{1}{3}c_{y1}t_{go}^{3} + \frac{1}{4}c_{y2}t_{go}^{4}$$

$$\begin{bmatrix} \dot{y}_{f} - \dot{y}_{0} - c_{0}t_{go} \\ y_{f} - y_{0} - \dot{y}_{0}t_{go} - \frac{1}{2}c_{0}t_{go}^{2} \end{bmatrix} = \begin{bmatrix} \frac{1}{2}t_{go}^{2} & \frac{1}{3}t_{go}^{3} \\ \frac{1}{3}t_{go}^{3} & \frac{1}{4}t_{go}^{4} \end{bmatrix} \begin{bmatrix} c_{y1} \\ c_{y2} \end{bmatrix}$$

$$E = \begin{bmatrix} \frac{1}{2}t_{go}^2 & \frac{1}{3}t_{go}^3 \\ \frac{1}{3}t_{go}^3 & \frac{1}{4}t_{go}^4 \end{bmatrix}^{-1} = \begin{bmatrix} \frac{18}{t_{go}^2} & -\frac{24}{t_{go}^3} \\ -\frac{24}{t_{go}^3} & -\frac{36}{t_{go}^4} \end{bmatrix}$$

$$\begin{bmatrix} c_{y1} \\ c_{y2} \end{bmatrix} = E \begin{bmatrix} \dot{y}_{f} - \dot{y}_{0} - c_{0}t_{go} \\ y_{f} - y_{0} - \dot{y}_{0}t_{go} - \frac{1}{2}c_{0}t_{go}^{2} \end{bmatrix} = \begin{bmatrix} \frac{18}{t_{go}^{2}} \left(\dot{y}_{f} - \dot{y}_{0} - c_{0}t_{go} \right) - \frac{24}{t_{go}^{3}} \left(y_{f} - y_{0} - \dot{y}_{0}t_{go} - \frac{1}{2}c_{y0}t_{go}^{2} \right) \\ -\frac{24}{t_{go}^{3}} \left(\dot{y}_{f} - \dot{y}_{0} - c_{0}t_{go} \right) - \frac{36}{t_{go}^{4}} \left(y_{f} - y_{0} - \dot{y}_{0}t_{go} - \frac{1}{2}c_{y0}t_{go}^{2} \right) \end{bmatrix}$$

$$= \begin{bmatrix} -\frac{24}{t_{go}^{3}} \left(y_{f} - y_{0} \right) + \frac{6}{t_{go}^{2}} \left(3\dot{y}_{f} + \dot{y}_{0} \right) - \frac{6c_{y0}}{t_{go}} \\ \frac{36}{t_{go}^{4}} \left(y_{f} - y_{0} \right) - \frac{12}{t_{go}^{3}} \left(2\dot{y}_{f} + \dot{y}_{0} \right) + \frac{6c_{y0}}{t_{go}} \end{bmatrix}$$

(d) 最终得到推力加速度指令为

$$a_{v}(t) = \ddot{y}(t) - g_{v}(t) = g_{v}(t_{f}) - 6(\dot{y}_{f} + \dot{y}_{0})/t_{go} + 12(y_{f} - y_{0})/t_{go}^{2} - g_{v}(t)$$

(e) z轴指令同理可得:

$$a_z(t) = \ddot{z}(t) - g_z(t) = g_z(t_{\rm f}) - 6\left(\dot{z}_{\rm f} + \dot{z}_{\rm 0}\right)/t_{\rm go} + 12\left(z_{\rm f} - z_{\rm 0}\right)/t_{\rm go}^2 - g_z(t)$$

- (1) 对于x轴,终端约束包括位置,速度:
- (a) 构建多项式:

$$\ddot{x}(t) = c_{x1} + c_{x2} \left(t_{\text{f}} - t \right)$$

(b) 积分计算

$$\dot{x}_{f} - \dot{x}_{0} = \int_{t_{0}}^{t_{f}} \ddot{x}(t) dt = c_{x1} t_{go} + \frac{1}{2} c_{x2} t_{go}^{2}$$

$$x_{f} - x_{0} - \dot{x}_{0} t_{go} = \int_{t_{0}}^{t_{f}} \left[\int_{t_{0}}^{t} \ddot{x}(s) ds \right] dt = \frac{1}{2} c_{x1} t_{go}^{2} + \frac{1}{3} c_{x2} t_{go}^{3}$$

(c) 反解 E 矩阵得到参数 c_1 和 c_2

$$\dot{x}_{f} - \dot{x}_{0} = c_{x1}t_{go} + \frac{1}{2}c_{x2}t_{go}^{2}$$

$$x_{f} - x_{0} - \dot{x}_{0}t_{go} = \frac{1}{2}c_{x1}t_{go}^{2} + \frac{1}{3}c_{x2}t_{go}^{3}$$

$$\begin{bmatrix} \dot{x}_{\rm f} - \dot{x}_{\rm 0} \\ x_{\rm f} - x_{\rm 0} - \dot{x}_{\rm 0} t_{\rm go} \end{bmatrix} = \begin{bmatrix} 1 & \frac{1}{2} t_{\rm go}^2 \\ \frac{1}{2} t_{\rm go}^2 & \frac{1}{3} t_{\rm go}^3 \end{bmatrix} \begin{bmatrix} c_{x1} \\ c_{y2} \end{bmatrix}$$

$$E = \begin{bmatrix} 1 & \frac{1}{2} t_{\rm go}^2 \\ \frac{1}{2} t_{\rm go}^2 & \frac{1}{3} t_{\rm go}^3 \end{bmatrix}^{-1} = \begin{bmatrix} \frac{4}{t_{\rm go}} & -\frac{6}{t_{\rm go}^2} \\ -\frac{6}{t_{\rm go}} & \frac{12}{t_{\rm go}} \end{bmatrix}$$

$$\begin{bmatrix} c_{x1} \\ c_{x2} \end{bmatrix} = E \begin{bmatrix} \dot{x}_{\rm f} - \dot{x}_{\rm 0} \\ x_{\rm f} - x_{\rm 0} - \dot{x}_{\rm 0} t_{\rm go} \end{bmatrix} = \begin{bmatrix} -\frac{6}{t_{\rm go}^2} (x_{\rm f} - x_{\rm 0}) + \frac{1}{t_{\rm go}} (4\dot{x}_{\rm f} + 2\dot{x}_{\rm 0}) \\ \frac{12}{t_{\rm go}^2} (x_{\rm f} - x_{\rm 0}) - \frac{6}{t_{\rm go}^2} (\dot{x}_{\rm f} + \dot{x}_{\rm 0}) \end{bmatrix}$$

(d) 最终得到推力加速度指令为

$$a_x(t) = \ddot{x}(t) - g_x = -(2\dot{x}_f + 4\dot{x}_0)/t_{go} + 6(x_f - x_0)/t_{go}^2 - g_x(t)$$

注意: 1. x轴需要单独分类讨论;

- 2. 题目中已知量为终端推力方向,终端推力大小为未知量,不可以拿未知量来进行制导计算;
- 3. y 轴终端加速度为 $g_y(t_{\rm f})$,推导制导指令时不可以等同于时间函数 $g_y(t)$; z 轴同理

9、推导偏航与滚转通道姿态运动的线性化模型 关键步骤:

(1) 列写原方程

$$\begin{cases} -mv\dot{\sigma} = -mg\sin\theta\sin\sigma - F_P\cos\alpha\sin\beta - C_z^{\beta}qS_M\beta - \frac{F_P}{2}\delta_{\psi} + F_z \\ J_y\dot{\omega}_y + (J_x - J_z)\omega_z\omega_x = -m_y^{\beta}qS_Ml_k\beta - \frac{m_y^{\omega_y}qS_Ml_k^2}{v}\omega_y - \frac{F_P}{2}(x_c - x_g)\delta_{\psi} + M_y \end{cases}$$

$$J_x \dot{\omega}_x + \left(J_z - J_y\right) \omega_y \omega_z = -\frac{m_x^{\omega_x} q S_M l_k^2}{v} \omega_x - F_P z_c \delta_\gamma + M_x$$

(2) 小角度假设

$$\begin{split} -mv\Delta\dot{\sigma} &= -mg\sin(\theta_0 + \Delta\theta)\Delta\sigma - F_P\cos\alpha_0\Delta\beta - C_z^\beta qS_M\Delta\beta - \frac{F_P}{2}\delta_\psi + F_z \\ J_y\left(\Delta\ddot{\psi} + \dot{\phi}_0\Delta\dot{\gamma}\right) + \left(J_x - J_z\right)\dot{\phi}(\Delta\dot{\gamma} - \dot{\phi}_0\Delta\psi) &= -m_y^\beta qS_M l_k\Delta\beta - \frac{m_y^{\omega_y} qS_M l_k^2}{v} \left(\Delta\dot{\psi} + \dot{\phi}_0\Delta\gamma\right) - \frac{F_P}{2}\left(x_c - x_g\right)\delta_\psi + M_y \\ J_x\left(\Delta\ddot{\gamma} - \dot{\phi}_0\Delta\dot{\psi}\right) + \left(J_z - J_y\right)\dot{\phi}(\Delta\dot{\psi} + \dot{\phi}_0\Delta\gamma) &= -\frac{m_x^{\omega_z} qS_M l_k^2}{v} \left(\Delta\dot{\gamma} - \dot{\phi}_0\Delta\psi\right) - F_P z_c\delta_\gamma + M_x \end{split}$$

(3) 最终结果

$$\begin{split} \Delta \dot{\sigma} &= \left(\frac{F_P \cos \alpha_0}{m v} + \frac{C_z^\beta q S_M}{m v}\right) \Delta \beta + \frac{g \sin \theta_0}{v} \Delta \sigma + \frac{F_P}{2m v} \Delta \delta_\psi - \frac{F_z}{m v} = c_1 \Delta \beta + c_2 \Delta \sigma + c_3 \Delta \delta_\psi + \overline{F}_z \\ \Delta \ddot{\psi} &+ \frac{m_y^{\omega_y} q S_M l_k^2}{J_y v} \Delta \dot{\psi} + \frac{m_y^\beta q S_M l_k}{J_y} \Delta \beta + \frac{F_P}{2J_y} \Big(x_c - x_g\Big) \Delta \delta_\psi = \frac{M_y}{J_y} \\ \Delta \ddot{\gamma} &+ \frac{m_x^{\omega_x} q S_M l_k^2}{J_x v} \Delta \dot{\gamma} + \frac{F_P z_c}{J_x} \Delta \delta_\gamma = \frac{M_x}{J_x} \end{split}$$

$$\psi = \sigma + \beta$$

注意: 1. 偏航通道原方程中,推力分量进行了修正(作业中不扣分)

10、推导偏航与滚转通道姿态运动的传递函数

说明:

$$\begin{split} \psi &= \sigma + \beta \\ \Delta \dot{\sigma} &= \left(\frac{F_P}{mv} + \frac{C_z^\beta q S_M}{mv} \right) \Delta \beta + \frac{g \sin \theta_0}{v} \Delta \sigma + \frac{F_P}{2mv} \Delta \delta_\psi - \frac{F_z}{mv} = c_1 \Delta \beta + c_2 \Delta \sigma + c_3 \Delta \delta_\psi + \overline{F}_z \\ \Delta \ddot{\psi} + \frac{m_y^{\omega_y} q S_M l_k^2}{J_y v} \Delta \dot{\psi} + \frac{m_y^\beta q S_M l_k}{J_y} \Delta \beta + \frac{F_P}{2J_y} \left(x_c - x_g \right) \Delta \delta_\psi = \frac{M_y}{J_y} \\ \Delta \ddot{y} + \frac{m_x^{\omega_x} q S_M l_k^2}{J_x v} \Delta \dot{\gamma} + \frac{F_P z_c}{J_x} \Delta \delta_\gamma = \frac{M_x}{J_x} \\ \psi &= \sigma + \beta \\ \Delta \dot{\sigma} &= \left(\frac{F_P}{mv} + \frac{C_z^\beta q S_M}{mv} \right) \Delta \beta + \frac{g \sin \theta_0}{v} \Delta \sigma + \frac{F_P}{2mv} \Delta \delta_\psi - \frac{F_z}{mv} = c_1 \Delta \beta + c_2 \Delta \sigma + c_3 \Delta \delta_\psi + \overline{F}_z \\ \Delta \ddot{\psi} &= -\frac{m_y^{\omega_x} q S_M l_k^2}{J_y v} \Delta \dot{\psi} - \frac{m_y^\beta q S_M l_k}{J_y} \left(\Delta \psi - \Delta \sigma \right) - \frac{F_P}{2J_y} \left(x_c - x_g \right) \Delta \delta_\psi + \frac{M_y}{J_y} \\ \Delta \ddot{\gamma} &= -\frac{m_x^{\omega_x} q S_M l_k^2}{J_x v} \Delta \dot{\gamma} - \frac{F_P z_c}{J_x} \Delta \delta_\gamma + \frac{M_x}{J_x} \\ \Delta \dot{\phi} &= \left(c_2 - c_1 \right) \Delta \sigma + c_1 \Delta \psi + c_3 \Delta \delta_\psi + \overline{F}_z \\ \Delta \dot{\psi} &= \Delta \psi \\ \Delta \ddot{\psi} &= b_2 \Delta \sigma - b_2 \Delta \psi - b_1 \Delta \dot{\psi} - b_3 \Delta \delta_\psi + \overline{M}_y \\ \Delta \ddot{\gamma} &= -d_1 \Delta \dot{\gamma} - d_2 \Delta \delta_\gamma + \overline{M}_x \end{split}$$

忽略干扰力和力矩,则俯仰通道的状态方程可表示为:

$$\begin{bmatrix} \Delta \dot{\sigma} \\ \Delta \dot{\psi} \\ \Delta \dot{\psi} \end{bmatrix} = \begin{bmatrix} c_2 - c_1 & c_1 & 0 \\ 0 & 0 & 1 \\ b_2 & -b_2 & -b_1 \end{bmatrix} \begin{bmatrix} \Delta \sigma \\ \Delta \psi \\ \Delta \dot{\psi} \end{bmatrix} + \begin{bmatrix} c_3 \\ 0 \\ -b_3 \end{bmatrix} \Delta \delta_{\psi}$$
$$\begin{bmatrix} \Delta \dot{\gamma} \\ \Delta \ddot{\gamma} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -d_1 \end{bmatrix} \begin{bmatrix} \Delta \gamma \\ \Delta \dot{\gamma} \end{bmatrix} + \begin{bmatrix} 0 \\ -d_2 \end{bmatrix} \Delta \delta_{\gamma} = 0$$

$$\begin{bmatrix} \frac{\Delta \sigma(s)}{\Delta \delta_{\psi}(s)} \\ \frac{\Delta \psi(s)}{\Delta \delta_{\psi}(s)} \\ \frac{\Delta \psi(s)}{\Delta \delta_{\psi}(s)} \end{bmatrix} = (sI - A)^{-1}B = \begin{bmatrix} \frac{c_3 s^2 + b_1 c_3 s + b_2 c_3 - b_3 c_1}{s^3 + (b_1 + c_1 - c_2) s^2 + \left[b_2 + b_1 (c_1 - c_2)\right] s - b_2 c_2} \\ \frac{-\left[b_3 s + b_3 (c_1 - c_2) - b_2 c_3\right]}{s^3 + (b_1 + c_1 - c_2) s^2 + \left[b_2 + b_1 (c_1 - c_2)\right] s - b_2 c_2} \\ \frac{-s\left[b_3 s + b_3 (c_1 - c_2) - b_2 c_3\right]}{s^3 + (b_1 + c_1 - c_2) s^2 + \left[b_2 + b_1 (c_1 - c_2)\right] s - b_2 c_2} \end{bmatrix} := \begin{bmatrix} W_{\delta_{\psi}}^{\sigma}(s) \\ W_{\delta_{\psi}}^{\psi}(s) \end{bmatrix}$$

$$\begin{bmatrix} \frac{\Delta \gamma(s)}{\Delta \delta_{\gamma}(s)} \\ \frac{\Delta \dot{\gamma}(s)}{\Delta \delta_{\gamma}(s)} \end{bmatrix} = (sI - A)^{-1}B = \begin{bmatrix} -\frac{d_3}{s^2 + sd_1} \\ -\frac{d_3}{s + d_1} \end{bmatrix}$$

11、根据发动机布局求解相关问题

关键步骤:

(1) 冗余度为1

定义发动机中心到火箭截面圆心的距离为 z_c ,物理舵偏的推导与等效舵偏的定义分别为:

$$\begin{split} M_x &= -4pz_c \sin \delta_{\gamma} = -pz_c \left(\delta_I + \delta_{II} + \delta_{III} + \delta_{IV} \right) \\ M_y &= -4 \times \frac{\sqrt{2}}{2} p \left(x_c - x_g \right) \sin \delta_{\psi} = \frac{\sqrt{2}}{2} p \left(x_c - x_g \right) \left(\delta_I - \delta_{II} - \delta_{III} + \delta_{IV} \right) \\ M_z &= -4 \times \frac{\sqrt{2}}{2} p \left(x_c - x_g \right) \sin \delta_{\varphi} = \frac{\sqrt{2}}{2} p \left(x_c - x_g \right) \left(\delta_I + \delta_{II} - \delta_{III} - \delta_{IV} \right) \end{split}$$

辅助方程设为:

$$\delta_I + \delta_{III} = \delta_{II} + \delta_{IV}$$

经小量假设得到控制方程为:

$$\begin{split} & \delta_{I} &= \delta_{\gamma} - \delta_{\psi} - \delta_{\varphi} \\ & \delta_{II} &= \delta_{\gamma} + \delta_{\psi} - \delta_{\varphi} \\ & \delta_{III} &= \delta_{\gamma} + \delta_{\psi} + \delta_{\varphi} \\ & \delta_{IV} &= \delta_{\gamma} - \delta_{\psi} + \delta_{\varphi} \end{split}$$

(2) 冗余度为0

定义发动机中心到火箭截面圆心的距离为 z_c ,物理舵偏的推导与等效舵偏的定义分别为:

$$\begin{split} M_{x} &= -3pz_{c}\sin\delta_{\gamma} = -pz_{c}\left(\delta_{I} + \delta_{II} + \delta_{III}\right) \\ M_{y} &= -\left(p + 2\times\frac{1}{2}p\right)\left(x_{c} - x_{g}\right)\sin\delta_{\psi} = \left(x_{c} - x_{g}\right)\left(\frac{1}{2}\delta_{I} + \frac{1}{2}p\delta_{II} - p\delta_{III}\right) \\ M_{z} &= -2\times\frac{\sqrt{3}}{2}p\left(x_{c} - x_{g}\right)\sin\delta_{\varphi} = \left(x_{c} - x_{g}\right)\left(\frac{\sqrt{3}}{2}p\delta_{I} - \frac{\sqrt{3}}{2}p\delta_{II}\right) \end{split}$$

无需辅助方程;

经小量假设得到控制方程为:

$$\delta_{I} = \delta_{\gamma} - \frac{2}{3} \delta_{\psi} - \delta_{\phi}$$

$$\delta_{II} = \delta_{\gamma} - \frac{2}{3} \delta_{\psi} + \delta_{\phi}$$

$$\delta_{III} = \delta_{\gamma} + \frac{4}{3} \delta_{\psi}$$

(3) 冗余度为9, 故障后冗余度为1

定义火箭建模截面圆心到芯级发动机摆动轴的距离为 c_1 (即芯级发动机中心到火箭截面圆心的距离为 $\sqrt{2}c_1$),火箭建模截面圆心到助推发动机的距离为 c_2 ,推力大小分别为 p_1 和 p_2 ,那么物理舵偏的推导与等效舵偏的定义分别为:

$$\begin{split} M_{x} &= c_{1}p_{1}\left(\delta_{1} - \delta_{2} - \delta_{3} + \delta_{4}\right) + c_{2}p_{2}\left(\delta_{I2} + \delta_{II1} - \delta_{III2} - \delta_{IV1}\right) \\ &= -\left(2p_{1}c_{1} + 4p_{2}c_{2}\right)\sin\delta_{\gamma} \\ M_{y} &= -\left(x_{c} - x_{g}\right)\left[c_{1}p_{1}\left(\delta_{2} + \delta_{4}\right) + c_{2}p_{2}\left(\delta_{I2} + \delta_{II2} + \delta_{III2} + \delta_{IV2}\right)\right] \\ &= -\left(p_{1} + 2p_{2}\right)\left(x_{c} - x_{g}\right)\sin\delta_{\psi} \\ M_{z} &= -\left(x_{c} - x_{g}\right)\left[c_{1}p_{1}\left(\delta_{1} + \delta_{3}\right) + c_{2}p_{2}\left(\delta_{I1} + \delta_{II1} + \delta_{III1} + \delta_{IV1}\right)\right] \\ &= -\left(p_{1} + 2p_{2}\right)\left(x_{c} - x_{g}\right)\sin\delta_{\varphi} \end{split}$$

辅助方程设为:

$$\delta_I + \delta_{II} = \delta_{III} + \delta_{IV}$$

将辅助方程代入方程组,并认为助推器发动机的8个控制角为已知量,经小量假设得到:

$$\begin{split} p_{1}\left(\delta_{1}-\delta_{3}\right) &= -\left(p_{1}+2\frac{p_{2}c_{2}}{c_{1}}\right)\delta_{\gamma} - \frac{1}{2}\frac{c_{2}p_{2}}{c_{1}}\left(\delta_{I2}^{*}+\delta_{II1}^{*}-\delta_{III2}^{*}-\delta_{IV1}^{*}\right) = M_{x}^{*} \\ p_{1}\left(\delta_{2}+\delta_{4}\right) &= \left(p_{1}+2p_{2}\right)\delta_{\psi} - c_{2}p_{2}\left(\delta_{I2}^{*}+\delta_{II2}^{*}+\delta_{III2}^{*}+\delta_{IV2}^{*}\right) = M_{y}^{*} \\ p_{1}\left(\delta_{1}+\delta_{3}\right) &= \left(p_{1}+2p_{2}\right)\delta_{\varphi} - c_{2}p_{2}\left(\delta_{I1}^{*}+\delta_{II1}^{*}+\delta_{III1}^{*}+\delta_{IV1}^{*}\right) = M_{z}^{*} \end{split}$$

根据已知量 M_x^*, M_y^*, M_z^* ,得到控制方程为:

$$\delta_{1} = \frac{M_{x}^{*} + M_{z}^{*}}{2p_{1}}, \delta_{2} = \frac{M_{y}^{*} - M_{x}^{*}}{2p_{1}}$$
$$\delta_{3} = \frac{M_{z}^{*} - M_{x}^{*}}{2p_{1}}, \delta_{4} = \frac{M_{x}^{*} + M_{y}^{*}}{2p_{1}}$$

注意:引入不同的辅助方程,定义不同的等效舵偏,将得出不一样的推导结果