

(BCS-II)

## Syllabus! -

### UNIT-I

Binary Codes - Weighted and Non-Weighted - Binary Arithmetic Conversion Algorithms - Error Detecting and Error Correcting Codes - Canonical and Standard Boolean Expressions - Truth Tables.

### UNIT-II

K-Map Reduction - Don't Care Condition - Adders/Subtractors - Carry Look-Ahead Adder - Code Conversions Algorithm - Design of Code Converters - Equivalence functions. Binary/Decimal Parallel Adder/Subtractor for Signed Numbers - Magnitude Comparator - Decoders/Encoders - Multiplexers/Demultiplexers - Boolean function Implementation using Multiplexers.

### UNIT-III

Sequential Logic - Basic Latch - Flip-flops (SR, D, JK, T and Master-Slave) - Triggering of flip-flops - Counters - Design Procedure - Ripple Counters - BCD and Binary - Synchronous Counters.

### UNIT-IV

Registers - Shift Registers - Registers with Parallel Load - Memory Unit - Example of RAM, ROM, PROM, EPROM - Reduction of State and flow Tables - Race-free State Assignment - Hazards.

8620 Convert this decimal no. into BCD and Excess-3 code.

8    6    2 0  
1000 0110 0010 0000

1000 0110 0010 0000  
0011 0011 0011 0011  
 $\overline{XS-3 = 1011 \ 1001 \ 0101 \ 0011}$

BCD - 1000011000100000

$XS-3 = 1011 \ 0010 \ 0100 \ 1001$

# Digital Systems and Binary Numbers

- ⇒ A sequence of instruction is called Program.
- ⇒ Any set that is restricted to a finite number of elements contains discrete information. e.g. - 10 decimal digits, the 26 letters of the alphabet, the 52 playing cards, and the 64 squares of a chessboard.

## Number System:-

1. Binary Number System - Base 2 (1, 0)
2. Octal Number System - Base 8 (0, 1, 2, 3, 4, 5, 6, 7)
3. Decimal Number System - Base 10 (0, 1, 2, 3, 4, 5, 6, 7, 8, 9, A, B, C, D, E, F)
4. Hexadecimal Number System - Base 16 (0, 1, 2, 3, 4, 5, 6, 7, 8, 9, A, B, C, D, E, F)

## Conversion:-

### 1. Binary to Octal

$$(10110)_2 = (?)_8$$

$$\begin{array}{r} 1 & 0 & 1 & 1 & 0 \\ \times & 16 & 8 & 4 & 2 & 1 \\ \hline = & 16 + 4 + 2 & & & & \\ = & 22 & & & & \\ (10110)_2 & = (22)_8 \\ \boxed{5 \quad 5} \end{array}$$

$$(10110)_2 = (55)_8$$

### 2. Octal to Binary

$$(101011100)_2$$

$$\begin{array}{r} 101011100 \\ \hline 2 \quad 1 \quad 7 \quad 4 \\ 001 \quad 010 \quad 111 \quad 100 \end{array}$$

$$(1274)_8 = (00101011100)_2$$

Octal Digit Value	Binary Equivalents
0	000
1	001
2	010
3	011
4	100
5	101
6	110
7	111

### 3. Binary to Decimal

$$(101110)_2 = (?)_{10}$$

$$\begin{array}{r} 1 \quad 0 \quad 1 \quad 1 \quad 1 \quad 0 \\ \times \quad 32 \quad 16 \quad 8 \quad 4 \quad 2 \quad 1 \\ \hline = 32 + 8 + 4 + 2 \\ = 46 \end{array}$$

$$(101110)_2 = (46)_{10}$$

### 4. Decimal to Binary

$$(46)_{10} = (?)_2$$

$$(46)_{10} = (101110)_2$$

2	46	101110
2	23	0
2	11	1
2	5	1
2	2	1
2	1	0
2	0	1

5. Binary to Hexadecimal — 6. Hexa to Binary —

$$(10010100111)_2 = (?)_{16} \quad (B2C)_{16} = (?)_2$$

$$\begin{array}{ccccccc} & & & B & 2 & C & \\ & & & \downarrow & \downarrow & \downarrow & \\ \boxed{1} & 0 & 0 & 1 & 0 & 1 & 0 & 1 & 1 \\ & 4 & 5 & 7 & & & & \\ & & & & 1011 & 0010 & 1100 & \end{array}$$

$$(10010100111)_2 = (457)_{16}$$

$$(B2C)_{16} = (101100101100)_2$$

7. Octal to Decimal —

$$(345)_8 = (?)_{10}$$

$$\begin{array}{r} 345 \\ 6481 \\ = 3 \times 64 + 4 \times 8 + 5 \times 1 \\ = 229 \end{array}$$

$$(345)_8 = (229)_{10}$$

8. Decimal to Octal —

$$(229)_{10} = (?)_8$$

$$\begin{array}{r} 8 | 229 \\ 8 | 285 \\ 8 | 34 \\ 8 | 4 \\ \hline \end{array}$$

$$(229)_{10} = (345)_8$$

9. Octal to Hexa —

$$(345)_8 = (?)_{16}$$

$$\begin{array}{ccccc} 3 & 4 & 5 \\ \downarrow & \downarrow & \downarrow \\ 011 & 100 & 101 \\ & 8 & \\ 001 & 011100101 \\ & E & 5 \\ & E5 & \end{array}$$

$$(345)_8 = (E5)_{16}$$

10. Hexa to Octal —

$$(A2D)_{16} = (?)_8$$

$$\begin{array}{r} A: 2 D \\ 1010 0010 1101 \\ \hline \end{array}$$

$$(A2D)_{16} = (5055)_8$$

11. Hexa to Decimal —

$$(A2DE)_{16} = (?)_{10}$$

$$\begin{array}{cccc} A & 2 & D & E \\ 4096 & 256 & 16 & 1 \\ \hline \end{array}$$

$$= 10 \times 4096 + 2 \times 256 + 3 \times 16 + 14 \times 1$$

$$= 41694$$

$$(A2DE)_{16} = (41694)_{10}$$

Hexa digit value Binary

	0	0000
1	1	0001
2	2	0010
3	3	0011
4	4	0100
5	5	0101
6	6	0110
7	7	0111
8	8	1000
9	9	1001
A	A	1010
B	B	1011
C	C	1000
D	D	1101
E	E	1110
F	F	1111

12. Decimal to Hexa —

$$(41694)_{10} = (?)_{16}$$

$$(41694)_{10} = (A2DE)_{16}$$

$$\begin{array}{r} 16 | 41694 \\ 16 | 2605 \dots 14 \\ 16 | 162 \dots 13 \\ 16 | 10 \dots 2 \\ \hline \end{array}$$

$$\begin{array}{r} 9 \\ 8 \\ 7 \\ 6 \\ 5 \\ 4 \\ 3 \\ 2 \\ 1 \\ 0 \end{array}$$

Binary Code :- A binary code represents text, computer processor instruction, or any other data using a two-symbol system. The two symbol system used is often "0" and "1" from the binary number system. The binary code assigns a pattern of binary digits, also known as bits, to each character, instruction etc. for ex - a binary string of eight bits can represent any of 256 possible values and can, therefore, represent a wide variety of different patterns items.

Range of an n-bit binary number :-

1-bit  $\rightarrow$  0, 1 → 0 to 1

2-bit  $\rightarrow$  0 → 00  
1 → 01

2 → 10

3 → 11

3-bit  $\rightarrow$  0 → 000

1 → 001

2 → 010

3 → 011

4 → 100

5 → 101

6 → 110

7 → 111

for n bit Range is  $[0 \text{ to } (2^n - 1)]$

for 2-bit 0 to  $2^2 - 1$   
 $0 \text{ to } 3$

for 3-bit 0 to  $2^3 - 1$   
0 to 7

Classification of Binary Codes :-

Weighted Codes

Positive Weighted Code

↓      ↓      ↓      ↓  
8421    2421    3321    4221

Decimal | 8421(Binary)

0	0000
1	0001
7	0111
9	1001

Decimal | 2421

0	0001
1	0001
7	0111
9	1111

Negative Weighted Code

↓      ↓  
84-2-1    74-2-1

Decimal | 3321

0	0000
1	0001
7	1101
9	1111

Decimal | 4221

0	0000
1	0001
7	1101
9	1111

Decimal	842-1	Excess 3 Code	Gray Code
1	0111 <sub>0</sub>	0111 <sub>0</sub>	0111 <sub>0</sub>
2	1001 <sub>0</sub>	1001 <sub>0</sub>	1001 <sub>0</sub>
3	1111 <sub>0</sub>	1111 <sub>0</sub>	1111 <sub>0</sub>
4	1011 <sub>0</sub>	1011 <sub>0</sub>	1011 <sub>0</sub>
5	0001 <sub>0</sub>	0001 <sub>0</sub>	0001 <sub>0</sub>
6	0101 <sub>0</sub>	0101 <sub>0</sub>	0101 <sub>0</sub>
7	1000 <sub>0</sub>	1000 <sub>0</sub>	1000 <sub>0</sub>
8	1011 <sub>0</sub>	1011 <sub>0</sub>	1011 <sub>0</sub>

### Gray Code

I Method - Binary to Gray Conversion II Method -

$$\begin{array}{r} \text{Binary} \\ \downarrow \\ 1011 - \text{Binary} \\ \text{MSB} \end{array}$$

$$\begin{array}{r} \text{Gray Code} \\ \Rightarrow 1110 \\ \downarrow \text{Sum of III + IV bit} \\ \downarrow \text{Sum of II + III bit} \\ \downarrow \text{Sum of I + II bit} \end{array}$$

In this process we don't calculate carry.

1011 - Binary

$$\begin{array}{r} 1011 \\ + 1011 \\ \hline 1110 \end{array}$$

Omit

Gray Code = 1110

In this method we don't calculate carry.

### # B Gray to Binary Conversion

1110 - Gray Code

1110

$\downarrow \uparrow \uparrow \uparrow$

1011 - Binary Code

Discard Carry.

0010 - B

1010 - C

0110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

1110 - Y

1110 - Z

1110 - A

1110 - B

1110 - C

1110 - D

1110 - E

1110 - F

1110 - G

1110 - H

1110 - I

1110 - J

1110 - K

1110 - L

1110 - M

1110 - N

1110 - O

1110 - P

1110 - Q

1110 - R

1110 - S

1110 - T

1110 - U

1110 - V

1110 - W

1110 - X

## Subtraction of binary numbers using 2's complement:-

$$\text{If } X = 1010100, Y = 1000011 \quad X - Y = ? \quad \text{If } Y = 1000011 \quad X = 1010100 \quad Y - X = ?$$

$[Y \text{ 2's Complement} + X]$

$$Y \text{ 1's} = 0111100$$

$$\begin{array}{r} 0111100 \\ + 1 \\ \hline 0111101 \end{array}$$

$$Y \text{ 2's} = 0111101$$

$$X = 84$$

$$Y = 67$$

$$[X - Y = 17]$$

$$X \text{ 1's} = 0101011$$

$$\begin{array}{r} 0101011 \\ + 1 \\ \hline 0101100 \end{array}$$

$$X = 84$$

$$Y = 67$$

$$Y - X = -17$$

$$X \text{ 2's} = 0101100$$

$$Y - X = 1000011$$

$$\begin{array}{r} 1000011 \\ + 0101100 \\ \hline 1101111 \end{array}$$

$$R = 1101111$$

Carry not means not desired  
Result so calculate 2's  
Complement of R.

$$R \text{ 2's} = 0010000$$

$$\begin{array}{r} 0010000 \\ + 1 \\ \hline 0010001 \end{array}$$

$$X - Y = 0111101 + 1010100$$

$$\begin{array}{r} 0111101 \\ + 1010100 \\ \hline 10010001 \end{array}$$

7bit  
7bit  
8bit

- Overflow

Then carry discard

$$X - Y = 0010001$$

$$[Y - X = -0010001]$$

Addition Rules! :-

$0+0=0$
$0+1=1$
$1+0=1$
$1+1=0 \text{ 1 Carry}$
$1+1+1=1 \text{ 1 Carry}$

Rules for Binary Subtraction :-

$0-0=0$
$1-0=1$
$0-1=1 \text{ 1 Borrow.}$
$1-1=0$

Binary Multiplication Rule! :-

$0 \times 0 = 0$
$0 \times 1 = 0$
$1 \times 0 = 0$
$1 \times 1 = 1$

Binary Division! :-

$0 \div 0 = 0$
$0 \div 1 = 0$
$1 \div 0 = \infty$
$1 \div 1 = 1$

$$\begin{array}{r} 100110 \\ \times 110 \\ \hline 110 \\ 110 \\ \hline 1001 \\ 110 \\ \hline 111 \\ 110 \\ \hline 10 \end{array}$$

## Signed Magnitude Representation:-

+5  
↓  
0101

-5  
↓  
1101

8-bit representation of +12

$$+12 = 00001100$$

Signed Magnitude Rep = 00001100

Signed 1's Complement Rep = 00001100

Signed 2's Complement Rep = 00001100

Positive No का तीनों सामें होता है।

8-bit representation of -12

$$-12 = 10001100$$

Signed Magnitude Rep = 10001100

1's of (-12) = 11110011

Signed 1's Complement Rep

10001100  
00110011  
00110011  
10001100

## Signed Binary Numbers

Positive integers can be represented as unsigned numbers.

However, to represent negative integers, we need a notation for negative values. In ordinary arithmetic, a negative number is indicated by a minus sign and a positive num. by a plus sign. Because of hardware limitations, computers must represent everything with binary digits. It is customary to represent a sign with a bit placed in the leftmost position of the number. The convention is make the sign bit 0 for positive and 1 for negative.

If it is important to realize that both signed and unsigned binary numbers consist of a string of bits when represented in a computer. If the binary number is signed, then the leftmost bit represents the sign and the leftmost bit if

most of bits represents the number. If the binary number is assumed to be unsigned, then the left-most bit is the most significant bit of the number.

Ex:- The string of bits 0100 can be considered as 9 (unsigned binary) or as +9 (signed binary) because the leftmost bit is 0. The string of bits 11001 represent the binary equivalent of 25 when considered as an unsigned number and the binary equivalent of -9 when considered as signed number. This is because the 1 that is in the leftmost position designates a negative and the other four bits represent binary 9.

### Complements:-

Complements are used in digital computers to simplify the subtraction operation and for logical manipulation. Simplifying operations leads to simpler, less expensive circuits to implement the operation. There are two types of complements for each base  $R$  system: the radix complement and the diminished radix complement. The first is referred to as the  $R$ 's complement and the second as  $(R-1)$ 's complement. When the value of the base  $R$  is substituted in the name of the two types are referred to as the 2's complement and 9's complement for binary numbers and 10's complement and 19's complement for decimal numbers.

## Signed Magnitude Representation

$$+5 = 0101$$

$$-5 = 1101$$

Signed bit or MSB (Most Significant Bit)

## 8-bit Representation of +12 :-

$$+12 = 00001100$$

Signed Magnitude Representation = 00001100

Signed 1's Complement Representation = 00001100

Signed 2's Complement Representation = 00001100

Note  $\Rightarrow$  +ve No. का तीनों Representation सame होते।

## 8-bit Representation of -12 :-

$$-12 = 00001100$$

Signed Magnitude Representation = 10001100

1's of (+12) = 11110011

Signed 1's Complement Representation = 11110011

2's of (+12) = 11110011

+1

11110100

Signed 2's Complement Representation = 11110100

Range of Signed Magnitude and Signed 1's Complement Representation  
Where  $n$  stands for no. of bits

$$[-(2^{n-1}) \text{ to } (2^{n-1}-1)] \text{ or } [(-2^{n-1}) \text{ to } (2^n-1)]$$

Range of Signed 2's Complement Representation  $[-2^{n-1} \text{ to } (2^{n-1}-1)]$

# 11110. Is signed 2's Complement Representation & its value  
decimal No: अंत कर्ता

11110

$$= -1 \times 2^5 + 1 \times 2^4 + 1 \times 2^3 + 1 \times 2^2 + 1 \times 2^1 + 0 \times 2^0$$

$$= -32 + 16 + 8 + 4 + 2$$

$$= -2$$

Ans

11110

0000

0100

1100

0110

1010

Weighted Codes:-

Weighted binary codes are those binary codes which obey the positional weight principle. Each position of the number represents a specific weight. Several systems of the codes are used to express the decimal digits 0 through 9. In these codes, each decimal digit is represented by a group of four bits.

Decimal  $\rightarrow$  A set of binary digits

Positional Weights  $\rightarrow$  8 + 4 + 2 + 1 octal. Binary place value

Code  $\rightarrow$  0010, minimum length will be unit

Non-Weighted Codes:-

In this type of binary codes, the positional weights are not assigned. Examples of non-weighted codes are Excess-3 code and Gray code.

Excess-3 Code:—The Excess-3 code is also called as XS-3 code. It is non-weighted code used to express decimal numbers. The Excess-3 code words are derived from the 8421 BCD code words adding 0011 or 3 to each code word in 8421. The Excess-3 codes are obtained as follows,—

Decimal Number  $\rightarrow$  8421 BCD  $\xrightarrow{\text{ADD } 0011}$  Excess-3

Decimal	BCD	Excess-3
	BCD + 0011	
0	0000	0011011111
1	0001	0100
2	0010	0101
3	0011	0110
4	0100	0111
5	0101	1000
6	0110	1001
7	0111	1010
8	1000	1011
9	1001	1100

Gray Code: It is the non-weighted code and it is not arithmetic codes. That means there are no specific weights assigned to the bit position. It has a very special feature that, only one bit change each time the decimal number is incremented as shown in figure. The gray code is called unit distance code. The gray code is a cyclic code. Gray code cannot be used for arithmetic operation.

Decimal	BCD	Gray
0	0000	0000
1	0001	0001
2	0010	0011
3	0011	0010
4	0100	0110
5	0101	0111
6	0110	0101
7	0111	0100
8	1000	1100
9	1001	1101

It is also known as reflected code.

Decimal	Binary unsigned	signed	$1's$ complement	$2's$ Complement
0	000	0	0	0
1	001	1	-1	1
2	010	2	-2	2
3	011	3	-3	3
4	100	4	-4	-4
5	101	5	-5	-3
6	110	6	-2	-2
7	111	7	-3	-1

In unsigned representation first bit should not be treated as sign bit and the remaining three representation (signed,  $1's$  complement &  $2's$  complement) first bit treated as sign bit.

$$n = 3 \text{ bits} ; 2^3 = 8$$

from 0 to 7		from 0 to 7		from 0 to 7	
unsigned ( $2^n$ value)	0	$+3$	$+3$	$+3$	$+3$
Signed ( $2^n$ value)	-3	$+3$	$+3$	$+3$	$+3$
$1's$ ( $2^n$ value)	-3	$+3$	$+3$	$+3$	$+3$
$2's$ ( $2^n$ value)	-4	$+3$	$+3$	$+3$	$+3$

System में हम  $2's$  Complement का use करते हैं signed &  $1's$  Complement की जगह  $2's$  Complement की use करते हैं।  $2's$  Complement में -0 & +0 को value प्राप्त होती है जबकि Mathematics में 0 माना जाता है।  $2's$  Complement में 0 असता है।

If you consider utilization of bit then unsigned and  $2's$  Complement is always better as compared to signed magnitude as well as  $1's$  Complement.  $2^3 = 8$  (2 Pt After 2 Pg)

Unsigned  $\Rightarrow$  0 to  $2^{n-1}$

Signed  $\Rightarrow$   $(2^{n-1}-1)$  to  $(2^{n-1})$

$1's$   $\Rightarrow$   $-(2^{n-1}-1)$  to  $(2^{n-1}-1)$

$2's$   $\Rightarrow$   $-(2^{n-1})$  to  $(2^{n-1})$

# Convert decimal 41 to binary.

2	41
2	20 1
2	10 0
2	5 0
2	2 1
2	1 0
2	1

$$(41)_{10} = (101001)_2$$

Conversion from decimal integers to any base- $r$  system is similar to this example, except that division is done by  $r$  instead of 2.

# Convert  $(0.6875)_{10}$  to binary.

$$0.6875 \times 2 = 1.3750 \quad \text{so } 1$$

$$0.3750 \times 2 = 0.7500 \quad \text{so } 0$$

$$0.7500 \times 2 = 1.5000 \quad \text{so } 1$$

$$0.5000 \times 2 = 1.0000 \quad \text{so } 1$$

$$(0.6875)_{10} = (0.11)_2$$

To convert a decimal fraction to a number expressed in base- $r$ , a similar procedure is used. However, multiplication is by  $r$  instead of 2, and the coefficients found from the integers may range in value from 0 to  $r-1$  instead of 0 and 1.

# Convert decimal  $(41.6875)_{10}$  to binary.

$$(41.6875)_{10} = (101001.1011)_2$$

Complement!

There are two methods for finding complement.

① Diminished Radix Complement,  $(r^k - 1)$ .

② Radix Complement ( $r^k$ ).

Complements are often used in digital computers to simplify the subtraction operation.

## ② Diminished Radix Complement ( $r'^s - 1$ ):

$$(547600)_{10}$$

$r = 10$

$$r'^s - 1 = (r^n - 1) - N$$

$n = \text{no. of total Digits}$

$n = 6$

$$N = 547600$$

$r = 10$

$$(10-1) = (10^6 - 1) - 547600$$

$$= 1000000 - 547601$$

$$9's \text{ Complement} = 999999 - 547600$$

$$9's \text{ Comp.} = 452399$$

Digit - FFFF

This complement is also known as 1's complement and 9's complement.

$$8010 = q_{10,0} 8^1 F$$

$$11-9999 = q_{10,0} 2^8$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

$$115012,$$

~~# 7's Comp. of Octal No. 5674~~

$$N = 5674, n=4, r=8$$

$$\begin{aligned} (4096-1) &= (8^4 - 1) + 5674 \\ (4095)_8 &\rightarrow (7777)_8 \\ &= 7777 \oplus 5674 \\ &= 7777 - 5674 \end{aligned}$$

$$\boxed{7'8 = 2103} \quad \text{Ans}$$

Simply we can do this

$$7'8 = 7777 - 5674$$

$$\boxed{7'8 = 2103} \quad \text{Ans}$$

$$(1000000_2 - 1101)_2 = 11001$$

$$(1000000_2 - 1101)_2 = 1111 - 1101$$

$$\begin{aligned} -1'8 &= 1111 - 1101 \\ 1'8 &= 0010 \quad \text{Ans} \end{aligned}$$

# 76.34 BT 10'8 Comp.

$$n=2(76), r=10, N=76.34$$

$$\begin{aligned} 100110101 &= N_2 = 10^2 - 76.34 \\ &= 23.66 \end{aligned}$$

Subtraction with 8's Complement

~~unsigned data :- Minuend~~ M<sub>1</sub> = 1025

~~Singed data :- M<sub>2</sub> = -370~~

~~Step 1:- Equal the length~~

$$M_1 = 1025$$

$$N_1 \Rightarrow 50 \rightarrow 0050$$

$$M_2 = -370 \rightarrow \underline{\underline{0370}} \quad N_2 = 4312$$

~~Step 2:- Represent Negative Operands by taking 8's complement of -ve operands~~

$$M_2 = 370$$

~~8's of M<sub>2</sub>~~

$$= 9630$$

$$N_2 = 4312$$

$$r=10$$

$$\begin{aligned} 1111111111111111 &= \\ -0370 &= \\ 9629 &= \\ +9630 &= \\ \hline 9630 & \end{aligned}$$

Step 3 → Complement + the subtrahend

$$M_1 = 1025$$

$$2^8 \text{ of } N_1 = 9950$$

$$\begin{array}{r} 9950 \\ - 1025 \\ \hline 8925 \\ + 1 \\ \hline 9950 \end{array}$$

$$N_2 = 1312$$

$$2^8 \text{ of } N_2 = 5688$$

$$9999$$

$$4312$$

$$5688$$

$$+ 1$$

$$5688$$

Step 4:- Addition and carry:-

$$M_1 = 1025$$

$$N_1 = 9950$$

$$D_1 = 1025 + 9950$$

$$= 10975$$

↑  
carry

$$M_2 = 9630$$

$$A_2 = 5688$$

$$D_2 = 15318$$

↑

carry

with the help of 3 bit 8 values

$(2^3 = 8)$  can be represented however because of repetition of 0 signed magnitude and  $1^8$  complement will represent one value less than the possible no. of 8.

0 has got multiple representation in SNM and  $1^8G$

0 has got its unique representation in  $2^8C$  and because of repetition you can store represent only 7 values in SNM and  $1^8C$ .

Signed Magnitude

# Binary Subtraction using 1's Complement

Step 1  $\Rightarrow$  Convert number to be subtracted to its 1's complement form.

Step 2  $\Rightarrow$  Perform the addition.

Step 3  $\Rightarrow$  If the final carry is 1, then add it to the result obtained in step 2. If final carry is 0, result obtained in step 2 is negative and its in the 1's complement form.

$$A - B = A + (-B)$$

$$\boxed{A - B = A + B \text{'s } 1\text{'s Complement}}$$

If carry is 1 then

$$A + (-B) = S$$

This process is called

+1  $\Rightarrow$  End-around carry

Result digit

$\Rightarrow$  If carry is 0

$$A + (-B) = S$$

Ex:- Perform  $(1100)_2 - (0101)_2$

$$A = 1100, B = 0101$$

$$1\text{'s of } B = 110101$$

$$A - B = A + (1\text{'s of } B) \\ = 1100 + 11010$$

$$= \underline{\underline{01110}} ; \text{ 7 } \times$$

$$\begin{array}{r} 0110 \\ + 1 \\ \hline 0111 \end{array}$$

$$A - B = 0111$$

$$A = 1100(12) \\ B = 0101(S)$$

$$12 - 5 = 7$$

7  $\rightarrow$  0111

Ex:- Perform  $(0101)_2 - (1100)_2$

$$\text{Taking } A = 0101, B = 1100$$

$$1\text{'s of } B = 110101$$

$$A - B = A + (1\text{'s of } B) \\ = 0101 + 110101$$

$$= \underline{\underline{11000}} ; \text{ 5 } \times$$

Invert this result

because it is in 1's Complement  
then,

$$\boxed{A - B = 0111 + (-7)}$$

## Decimal Subtraction Using 10's Complement:-

Case 1:- Carry is produced -

$$\text{Ex:- } (2058)_{10} - (1729)_{10} = 239$$

$$X = 2058$$

$$Y = 1729$$

$$X - Y = X + \text{9's of } Y$$

$$\begin{array}{r} 9\text{'s of } Y = \\ - 1729 \\ \hline 8270 \end{array}$$

$$10^3 \text{ of } Y = 8270$$

$$\begin{array}{r} + 1 \\ \hline 8271 \end{array}$$

$$10^3 \text{ of } Y = 8271$$

$$X - Y = 2058$$

$$+ 8271$$

$$\hline 10329$$

discard the carry then

$$X - Y = 0329$$

Case 2:- No carry -

$$\text{Ex:- } (7398)_{10} - (9207)_{10} = 239$$

$$X = 7398$$

$$Y = 9207$$

$$X - Y = X + (10^3 \text{ of } Y)$$

$$\begin{array}{r} 9\text{'s of } Y = \\ - 9207 \\ \hline 0792 \end{array}$$

$$X - Y = 7398$$

$$+ 0793$$

$$\hline 8191$$

In this step we don't have any carry so calculate:

10's complement of the sum

$$\text{10's complement of } 8191 = 1808$$

$$1808 - 9\text{'s of } 8191 = 1808 - 8191$$

$$1808 - 8191 = -1809$$

Binary Codes:- An  $n$  bit binary code is a group of  $n$  bits that assumes up to  $2^n$  distinct combinations of 1's and 0's with each combination representing one element of the set that is being coded. A set of four elements can be coded with two bits, with each element assigned one of the following bit combinations: 00, 01, 10, 11. A set of eight elements requires a three-bit code and a set of 16 elements requires a four-bit code. The bit combination of an  $n$ -bit code is determined from the count in binary from 0 to  $2^n - 1$ . Each element must be assigned a unique binary bit combination, and no two elements can have the same value; otherwise, the code assignment will be ambiguous.

BCD:- A number with  $K$  decimal digits will require  $4K$  bits in BCD. Decimal 396 is represented in BCD with 12 bits as 0011 1001 0110, with each group of 4 bits representing one decimal digit. A decimal number in BCD is the same as its equivalent binary number only when the number is between 0 and 9. A BCD number greater than 9 looks different from its equivalent binary number even though both contains 0's and 1's. Moreover, the binary combinations 1010 through 1111 are not used and have no meaning in BCD. Consider decimal 185 and its corresponding value in BCD and binary:

$$(185)_{10} = (0001 \ 0110 \ 0101)_{BCD} = (10100101)_2$$

The BCD value has 12 bit to encode 4 characters of the decimal value, but the equivalent binary number needs only 8 bit. It is obvious that the representation of a BCD number needs more bits than its equivalent binary value.

It is important to realize that BCD numbers are decimal numbers and not binary numbers, although they use bits in their representation. The only difference between a decimal number and BCD is that decimals are written with the symbols 0, 1, 2, ..., 9 and BCD numbers use the binary code 0000, 0001, 0010, ..., 1001. The decimal value is exactly the same.

ASCII Character Code:- It includes 128 characters.

ASCII - American Standard Code for Information Interchange

Uppercase letters (A through Z) - 26

Lowercase letters (a through z) - 26

Numbers (0-9) - 10

Special characters - 66

Error Detection Code:- (Only for single bit error)

Message	Odd Parity	Even Parity	Message	Parity	Message	Parity
0000	1	0	0000	0	0000	1
0001	0	1	0001	1	0001	0
0010	0	1	0010	1	0010	0
0011	1	0	0011	0	0011	1
0100	0	1	0100	1	0100	0

In single bit error only one bit of data unit is changed from 0 to 1 or 1 to 0.

Parity :- It is an extra bit that is sent with the message signal. It detects single bit error.

Boolean Al Boolean Algebra and Logic creates

Algebraic Structure :- A set associated with binary operation

Closure property not hold for natural No. It holds only for integers.

$$(N, +) \times (N, -) \times (I, -) \quad a(I, -) \checkmark, e, (I, +) \checkmark$$

Associativity :- A binary operator  $*$  on a set  $S$  is said to be associative where  $a * (b * c) = (a * b) * c$  for all  $a, b, c \in S$

$$\Rightarrow (a+b)+c = a+(b+c)$$

$$\Rightarrow (a * b) * c = a * (b * c)$$

$$\Rightarrow a - (b - c) \neq (a - b) - c$$

$$\Rightarrow a - b + c \neq a - b - c$$

Commutative Property - Existence of Identity -

$$a * b = b * a$$

$$e * x = x * e = x$$

$$\Rightarrow a + b = b + a$$

$$\Rightarrow 0 + x = x + 0 = x$$

$$\Rightarrow a * b = b * a$$

$$\Rightarrow 1 * x = x * 1 = x$$

$$\Rightarrow a - b \neq b - a$$

Distributive Law -

$$a * (b * c) = (a * b) + (a * c)$$

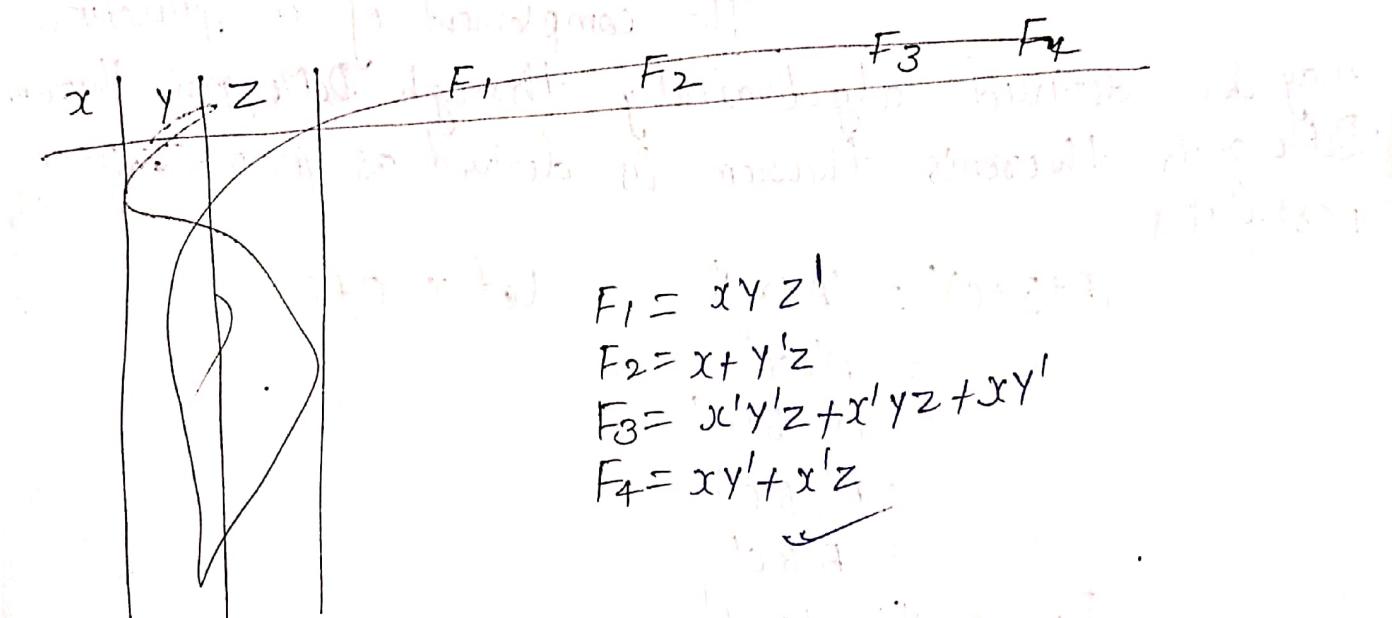
$$\Rightarrow a * (b + c) = (a * b) + (a * c)$$

$$\Rightarrow a + (b * c) \neq (a + b) * (a + c)$$

$(S, +, *)$ , field

An algebraic structure, which the properties of closure, associativity, commutative, identity, inverse and distributive hold is called a field.

Boolean function:— Boolean function is a function of boolean variable.



$x \ y \ z$	$F_1$	$F_2$	$F_3$	$F_4$
0 0 0	0	0	0	0
0 0 1	0	1	0	1
0 1 0	0	0	0	0
0 1 1	0	0	1	1
1 0 0	0	1	1	0
1 0 1	0	1	1	0
1 1 0	1	1	0	0
1 1 1	0	1	0	0

## Complement of a function:-

The complement of a function  $F$  is  $F'$  and is obtained from the interchange of 0's to 1's and 1's to 0's.

Canonical and Standard form for representing function

### Minterms and Maxterms

The complement of a function may be derived algebraically through DeMorgan's theorem. DeMorgan's theorem is derived as follows:

$$\begin{aligned}
 (A+B+C)' &= (A+x)' \\
 &= A' \oplus x' \\
 &= A' (B+C)' \\
 &= A' (B'C') \\
 &= A' B' C' \\
 \boxed{(A+B+C)' = A' B' C'}
 \end{aligned}$$

DeMorgan's theorems for any number of variables resemble the two-variable case in form and can be derived by successive substitutions similar to the method used in the preceding derivation. These theorems can be generalized as follows:-

$$\boxed{(A+B+C+D+\dots+F)' = A' B' C' D' \dots F'}$$

$$\boxed{(ABC\dots F)' = A'+B'+C'+\dots+F'}$$

Boolean Algebra - Boolean algebra is a mathematical system with a set of elements, a set of operations and a number of unproved postulates.

## Rules for Boolean Algebra

$$① x + 0 = x$$

$$② x + x' = 1$$

$$③ x + x = x$$

$$④ x \cdot 1 = x$$

$$⑤ (x')' = x$$

$$⑥ x + y = y + x$$

$$⑦ x + (y + z) = (x + y) + z$$

$$⑧ x(y + z) = xy + xz$$

$$⑨ (x + y)' = x'y'$$

$$⑩ x + xy = x$$

$$⑪ x \cdot 1 = x$$

$$⑫ x \cdot x' = 0$$

$$⑬ x \cdot x = x$$

$$⑭ x \cdot 0 = 0$$

$$⑮ xy = yx$$

$$⑯ x(yz) = (xy)z$$

$$⑰ x + yz = (x + y)(x + z)$$

$$⑱ (xy)' = x'y'$$

$$⑲ x(x+y) = x$$

$$x + x = x$$

Proof

$$(x+x) \cdot 1 = (x+x) \cdot (x+x')$$

$$= x \cdot x + x \cdot x' + x \cdot x + x \cdot x'$$

$$= x + x$$

$$= x$$

$$\boxed{x + xy = x}$$

$$x + xy = x(1+y)$$

$$= x(y+1)$$

$$= x \cdot 1$$

$$= x$$

$$\Rightarrow x(x+y)$$

$$= (x \cdot x + xy)$$

$$= x + xy$$

$$= x \cdot 1 + xy$$

$$= x \cdot (y+y') + xy$$

$$= xy + xy' + xy$$

$$= x'y + xy'$$

$$= x(y+y') = x \cdot 1 = x$$

## Canonical And Standard form for Representing function

Minterms and Maxterms! — A binary variable may appear either in its normal form ( $x$ )

or in its complement form ( $x'$ ).

Now consider two variables,  $x$  and  $y$

combined with an AND operation. Since each variable may appear in either form, there are four possible

combinations:-  $x'y'$ ,  $x'y$ ,  $xy'$  and  $xy$ . Each of these four

AND terms is called a minterm or a standard product.

In a similar manner,  $n$  variables can be combined to form  $2^n$  minterms. A symbol for each minterm is also shown in the table and is of the form  $m_j$ , where the subscript  $j$  denotes the decimal equivalent

of the binary number of the minterm designated.

In a similar fashion,  $n$  variables forming an OR term, with each variable being primed or unprimed provide  $2^n$  possible combinations, called maxterms or standard sums. for  $n$  variables total no. of maxterms will be  $2^n$  similarly. Each maxterm is obtained from an OR term of the  $n$  variables with each variable being unprimed if the corresponding bit is 0 and primed if a 1.

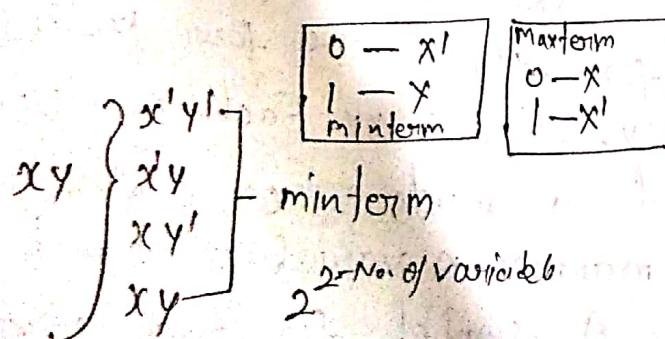
Each minterm is obtained from an AND term of the  $n$  variables, with each variable being primed if the corresponding bit of the binary number is a 0 and unprimed if a 1.

### Minterms (SOP)

$x \ y \ z$	Term	Designation
0 0 0	$x'y'z'$	$m_0$
0 0 1	$x'y'z$	$m_1$
0 1 0	$x'y'z'$	$m_2$
0 1 1	$x'yz$	$m_3$
1 0 0	$xy'z'$	$m_4$
1 0 1	$xy'z$	$m_5$
1 1 0	$xyz'$	$m_6$
1 1 1	$xyz$	$m_7$

### Maxterms (POS)

Term	Designation
$x+y+z$	$M_0$
$x+y+z'$	$M_1$
$x+y'+z$	$M_2$
$x+y'+z'$	$M_3$
$x'+y+z$	$M_4$
$x'+y+z'$	$M_5$
$x'+y'+z$	$M_6$
$x'+y'+z'$	$M_7$



Case of  $n$  variable total minterm =  $2^n$

In case of  $n$  variable  
total Maxterm =  $2^n$

## functions of Three Variables

x	y	z	function f <sub>1</sub>	function f <sub>2</sub>
0	0	0	0	0
0	0	1	1	0
0	1	0	0	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

A boolean function can be expressed algebraically from a given truth table by forming a minterm for each combination of the variables that produces a 1 in the function and then taking the OR of all those terms.

exp:- function f<sub>1</sub> is determined by expressing the combinations 001, 100 and 111 as  $x'y'z$ ,  $x'y'z'$  and  $x'yz$  respectively. Since each of the result one of these minterms result in  $f_1 = 1$ .

Sum of product form (Sum of minterm):—

$$f_1 = x'y'z + x'y'z' + x'yz$$

$$f_1 = m_1 + m_4 + m_7$$

$$= \Sigma(1, 4, 7)$$

$$= \underline{\Sigma(m_1, m_4, m_7)}$$

$$f_2 = x'y'z + x'y'z' + x'yz' + x'yz$$

$$f_2 = m_3 + m_5 + m_6 + m_7$$

$$f_2 = \underline{\Sigma(3, 5, 6, 7)}$$

Product of Sum form:-

$$f_1 = (x+y+z)(x+y'+z)(x+y+z')(x'+y+z)(x'+y'+z)$$

$$f_1 = M_0 M_2 M_3 M_5 M_6$$

$$f_1 = \Sigma(0, 2, 3, 5, 6)$$

$$f_2 = (x+y+z)(x+y+z')(x+y'+z)(x'+y+z)$$

$$f_2 = M_0 M_1 M_2 M_4$$

$$f_2 = \Sigma(0, 1, 2, 4)$$

Boolean functions expressed as a sum of minterms or product of maxterms are said to be in a Canonical form.

Sum of Minterms (Sum of Product Terms) :-

# Express the Boolean function  $F = A + B'C$  as a sum of minterms. The function has three variables : A, B and C.

$$\begin{aligned} F &= A + B'C \\ &= A \cdot 1 + 1 \cdot B'C \\ &= A \cdot (B+B') + (A+A') B'C \\ &= AB + AB' + A B'C + A'B'C \\ &= AB(C+C') + AB'(C+C') + AB'C + A'B'C \\ &= ABC + ABC' + \underline{AB'C} + \underline{AB'C'} + \underline{AB'C} + A'B'C \end{aligned}$$

If any term appears twice remove one of those,

$$\begin{aligned} &= ABC + ABC' + AB'C + AB'C' + A'B'C \\ &= m_7 + m_6 + m_5 + m_4 + m_1 \end{aligned}$$

$$F = \cancel{m_7} + m_1 + m_4 + m_5 + m_6 + m_7$$

$$F = m_1 + m_4 + m_5 + m_6 + m_7$$

$$F = \Sigma(1, 4, 5, 6, 7)$$

Product of Maxterms (Product-of-Sum Term)! -

# Express the Boolean function  $F = Y\bar{Y} + \bar{X}'Z$  as a product of maxterms.

$$\begin{aligned} F &= (\cancel{Y} + \bar{X}'Z) \\ &= \bar{A} + \bar{X}'Z \\ &= (\bar{A} + \bar{X}')(\bar{A} + Z) \\ &= (\cancel{X}Y + \bar{X}')(\bar{X}Y + Z) \\ &= (\cancel{X} + \bar{X}')(\bar{Y} + \bar{X}')(\bar{X} + Z)(Y + Z) \\ &= \underline{\underline{1}} \\ &= (Y + \bar{X}')(\bar{Y} + Z)(Y + Z) \quad \text{--- (1)} \end{aligned}$$

$$\begin{aligned} x + z \\ &= x + z + 0 \\ &= x + z + YY' \\ &= A + YY' \\ &= (A + Y)(A + Y') \\ &= (x + z + Y)(x + z + Y') \\ &= (x + Y + z)(x + Y' + z) \end{aligned}$$

$$\begin{aligned} y + z \\ &= Y + Z + 0 \\ &= Y + Z + XX' \\ &= A + XX' \\ &= (A + X)(A + X') \\ &= (Y + Z + X)(Y + Z + X') \\ &= (x + Y + z)(x' + Y + z) \end{aligned}$$

$$\begin{aligned} x' + y \\ &= x' + Y + 0 \\ &= x' + Y + ZZ' \\ &= A + ZZ' \\ &= (A + z)(A + z') \\ &= (x' + Y + z)(x' + Y + z') \\ &= \underline{\underline{1}} \end{aligned}$$

Putting these values in (1),

$$\begin{aligned} &\Rightarrow \frac{(x' + Y + z)(x' + Y + z')(x + Y + z)}{(x + Y + z)(x + Y + z')(x' + Y + z)} \\ &= (x' + Y + z)(x' + Y + z')(x + Y + z) \\ &\quad (x + Y + z') \\ &= (x' + Y + z)(x + Y' + z)(x' + Y + z) \\ &\quad (x' + Y + z') \end{aligned}$$

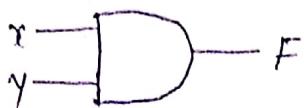
$$F = M_0 M_2 M_4 M_5$$

$$F = \Pi(0, 2, 4, 5)$$

# Digital Logic Gates:-

Name      Graphic symbol      Algebraic fun<sup>n</sup>      Truth Table

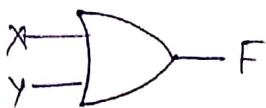
AND



$$F = xy$$

x	y	F
0	0	0
0	1	0
1	0	0
1	1	1

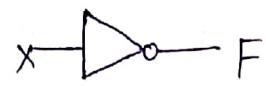
OR



$$F = x + y$$

x	y	F
0	0	0
0	1	1
1	0	1
1	1	1

Inverter



$$F = x'$$

x	F
0	1
1	0

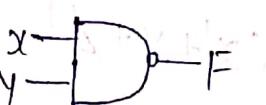
Buffer



$$F = x$$

x	F
0	0
1	1

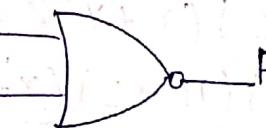
NAND



$$F = (xy)'$$

x	y	F
0	0	1
0	1	1
1	0	1
1	1	0

NOR



$$F = (x+y)'$$

x	y	F
0	0	1
0	1	0
1	0	0
1	1	0

Exclusive-OR  
(XOR)

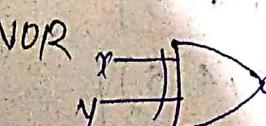


$$F = xy + x'y$$

$$= x \oplus y$$

x	y	F
0	0	0
0	1	1
1	0	1
1	1	0

Exclusive-NOR  
or  
equivalence



$$F = xy + x'y$$

$$= (x \oplus y)'$$

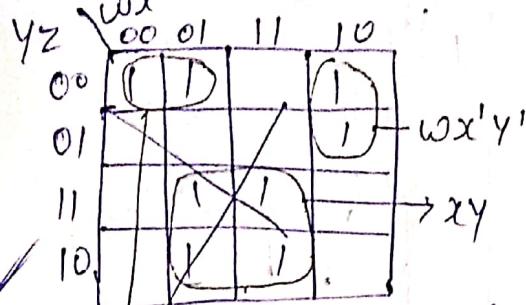
x	y	F
0	0	1
0	1	0
1	0	0
1	1	1

## Gate-Level Minimization

### Example of K-map:-

minimize -

$$\# f(w, x, y, z) = \Sigma (0, 4, 6, 7, 8, 9, 15)$$



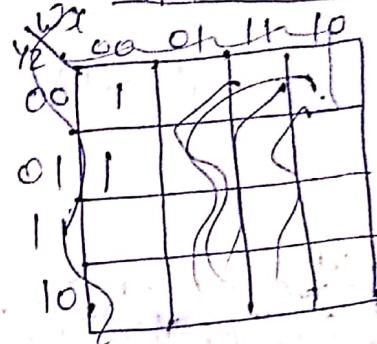
$$0 \Rightarrow w'x'y'z$$

$$4 \Rightarrow w'x'y'z$$

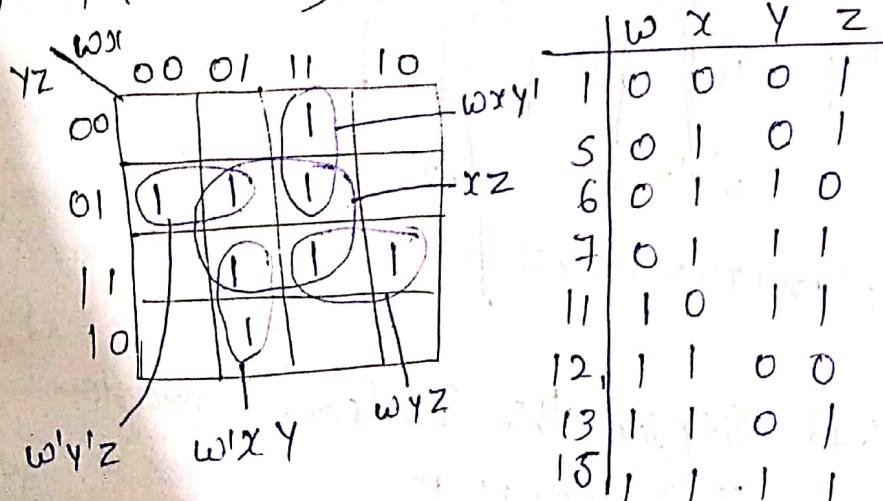
w'x'y'z	0	0	0	0
0	0	1	0	0
4	0	1	0	0
6	0	1	1	0
7	0	1	1	1
8	1	0	0	0
9	1	0	0	1
15	1	1	1	1

If any variable is change we skip that variable and if any variable not changes we just write them.

$$= w'y'z + x'y + wx'y'$$



$$\# f(w, x, y, z) = \Sigma (1, 5, 6, 7, 11, 12, 13, 15)$$



w'x'y'z	1	0	0	0	1
1	0	1	0	1	
5	0	1	1	0	
6	0	1	1	1	
7	0	1	1	1	
11	1	0	1	1	
12	1	1	0	0	
13	1	1	0	1	
15	1	1	1	1	

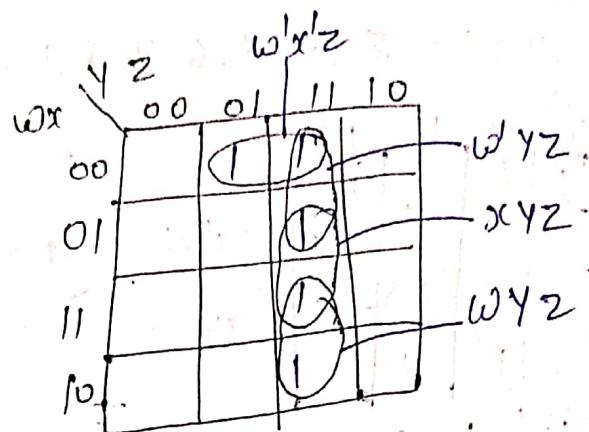
$$= w'y'z + wxy + wyz + wxy' + xz$$

The kmap is a graphical representation that provides a systematic method for simplifying the Boolean Expression.

# Simplify the Boolean function -

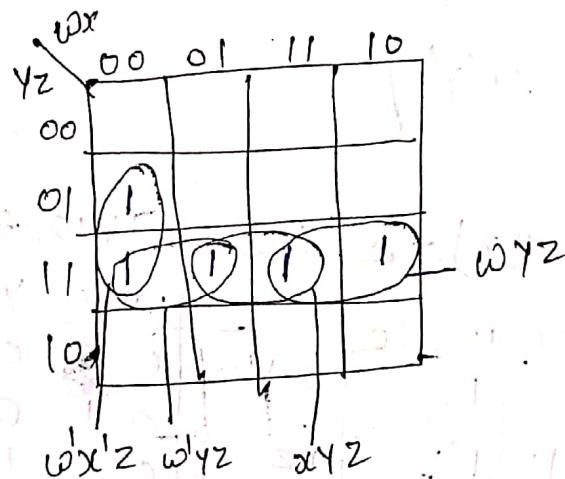
$$F(w, x, y, z) = \Sigma (1, 3, 7, 11, 15)$$

$wx$	$yz$	00	01	11	10
00	$m_0$	$m_1$	$m_3$	$m_2$	
01	$m_4$	$m_5$	$m_7$	$m_6$	
11	$m_{12}$	$m_{13}$	$m_{15}$	$m_{14}$	
10	$m_8$	$m_9$	$m_{11}$	$m_{10}$	



$$F_{(SOP)} = w'x'z + w'yz + xy'z + wz$$

$wx$	$yz$	00	01	11	10
00	$m_0$	$m_4$	$m_{12}$	$m_8$	
01	$m_1$	$m_5$	$m_{13}$	$m_9$	
11	$m_3$	$m_7$	$m_{15}$	$m_{11}$	
10	$m_2$	$m_6$	$m_{14}$	$m_{10}$	

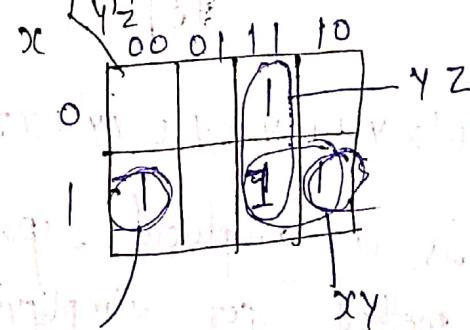


$$F_{(SOP)} = w'x'z + w'yz + xy'z + wz$$

#  $F(x, y, z) = \Sigma (3, 4, 6, 7)$

$xz$	$yz$	00	01	10	11
0	$m_0$	$m_1$	$m_3$	$m_2$	
1	$m_4$	$m_5$	$m_7$	$m_6$	

Simplify this Boolean Expression.

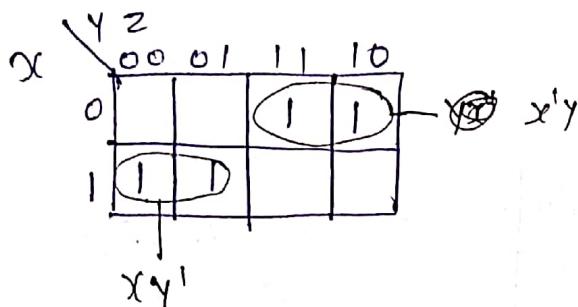


(yz + xz)

# Simplify the boolean function  $F(x,y,z) = \Sigma(2,3,4,5)$

First, 0's are marked in each minterm that represents the function

x\y	00	01	11	10
0	$m_0$	$m_1$	$m_3$	$m_2$
1	$m_4$	$m_5$	$m_7$	$m_6$



$$F_{(SOP)} = x'y + xy'$$

Karnaugh map or K-map:-

Map Method:-

The map method presented here provides a simple, straightforward procedure for minimizing Boolean function. This method may be regarded as a pictorial form of a truth table. The map method is also known as the Karnaugh map or K-map.

A K-map is a diagram made up of squares, with each square representing one minterm of the function that is to be minimized. Since any Boolean function is recognized graphically in the map from the can be expressed as a sum of minterms, it follows that a Boolean function is recognized graphically in the map form from the area enclosed by those squares whose minterms are included in the function. In fact the map presents a visual diagram

of all possible ways a function may be expressed in standard form, by recognizing various patterns.

The simplified expressions produced by the map are always in one of the two standard forms: sum of products or product of sums.

### Two-Variable map:-

$m_0$	$m_1$
$m_2$	$m_3$

(a)

	0	1
0	$m_0$ , $x'y'$	$m_1$ , $x'y$
1	$m_2$ , $xy'$	$m_3$ , $xy$

(b)

The two variable map is shown in the figure. There are four minterms for two variables. Hence the map consists of four squares, one for each minterm. The map is redrawn in (b) to show the relationship between the squares and the two variables  $x$  and  $y$ . The 0 and 1 marked in each row and column designate the value of variables. The value  $x$  appears primed in row 0 and unprimed in row 1. Similarly,  $y$  appears primed in column 0 and unprimed in column 1.

	0	1
0	$m_0$ .	$m_1$
1	$m_2$	$m_3$

$$F = \overline{xy}$$

	0	1
0	$m_0$	$m_1$
1	$m_2$	$m_3$

$$F = \overline{x} + \overline{y}$$

Note:- There will be  $2^n$  minterms for  $n$  variable.

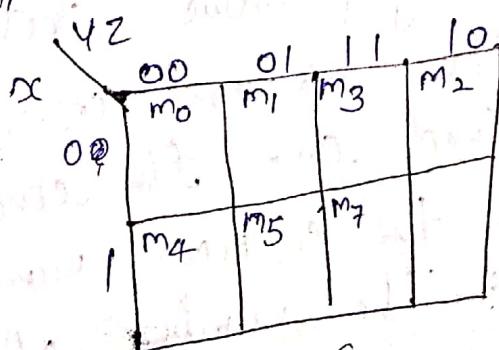
## Three-variable Map

A three variable map is shown in the figure. There are eight minterms for three binary variables; therefore, the map consists of eight squares. The minterms are arranged, not in a binary sequence, but in a sequence similar to Gray code. The characteristic of this sequence is that only one bit changes in value from one adjacent column to the next. The map diagram in part (b) is marked with numbers in each row and each column to show the relationship between the squares and the three variables.

For ex! — The square assigned to  $m_5$  corresponds to row 1 and column 01. When these two numbers are concatenated, they give the binary number 101, whose decimal equivalent is 5. To understand the usefulness of the map in simplifying Boolean functions, we must recognize the basic property possessed by adjacent squares: in the maps differ by only one variable. Any two adjacent squares differ by only one variable, which is primed in one square and unprimed in the other.

$m_0$	$m_1$	$m_3$	$m_2$
$m_4$	$m_5$	$m_7$	$m_6$

(a)



(b)

→ Let the Boolean function  $F = A'C + A'B + AB'C + BC$

- ① Express this function as a sum of minterms.
- ② Find the minimal sum-of-products expression.

BC		00	01	11	10
A		m <sub>0</sub>	m <sub>1</sub>	m <sub>3</sub>	m <sub>2</sub>
0	0		1	1	1
1	0	m <sub>4</sub>	m <sub>5</sub>	m <sub>7</sub>	m <sub>6</sub>

$$F = A'C + A'B + AB'C + BC$$

### Four Variable Map

The map for Boolean function of four binary variables is shown in the figure, are listed the 16 minterms and the squares assigned to each. In (b) the map is redrawn to show the relationship between the squares and the four variables. The rows and columns are numbered in a Gray code sequence, with only one digit changing value between two adjacent rows or columns. The minterm corresponding to each square can be obtained from the concatenation of the row number with the column number. For example, the number of the third row (11) and the second column (01), when concatenated, give the binary number 1101, the binary

equivalent of decimal 13. Thus, the square of in the third row and second column represents min term  $m_{13}$ .

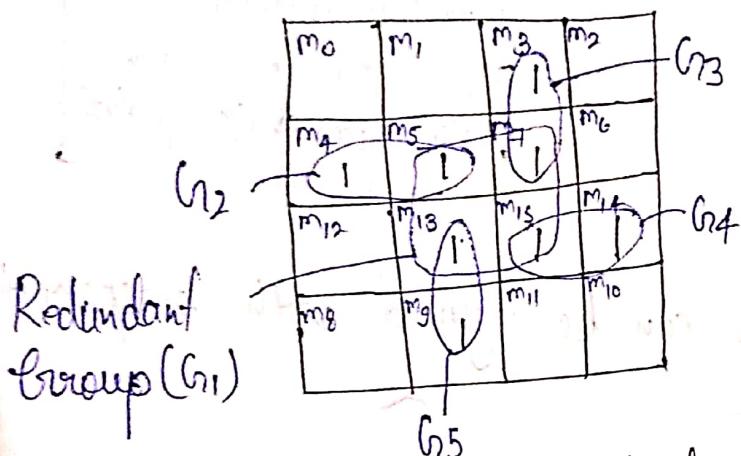
$m_0$	$m_1$	$m_3$	$m_2$
$m_4$	$m_5$	$m_7$	$m_6$
$m_{10}$	$m_{13}$	$m_{15}$	$m_{14}$
$m_8$	$m_9$	$m_{11}$	$m_{10}$

$wx \backslash yz$	00	01	11	10
00	$m_0$ $w'x'y'z'$	$m_1$ $w'x'y'z$	$m_3$ $wx'y'z$	$m_2$ $wx'y'z'$
01	$m_4$ $w'xy'z'$	$m_5$ $wxy'z$	$m_7$ $w'xyz$	$m_6$ $wxyz'$
11	$m_{10}$ $wxy'z'$	$m_{13}$ $wxy'z$	$m_{15}$ $wxyz$	$m_{14}$ $wxyz'$
10	$m_8$ $wx'y'z'$	$m_9$ $wx'y'z$	$m_{11}$ $wx'y'z$	$m_{10}$ $wx'y'z'$

The following terms can be defined with respect to K-map simplification:

- ① Pair:- Group of two adjacent minterms ( $m_4, m_5$ ). A pair eliminates one variable in output expression.
  2. Quad:- Group of four adjacent minterms ( $m_0, m_2, m_4, m_6$ ). A quad eliminates two variables in output expression.
  3. Octet:- Group of eight adjacent minterms ( $m_0, m_1, m_4, m_5, m_8, m_9, m_{12}, m_{13}$ ). An octet eliminates three variables in output expression.
  4. Redundant group:- A redundant group is a group in which all the elements in this group are covered by some other groups.
- Consider the example shown in figure. Here  $m_5, m_7, m_{13}, m_{15}$  form a quad (G, J). However

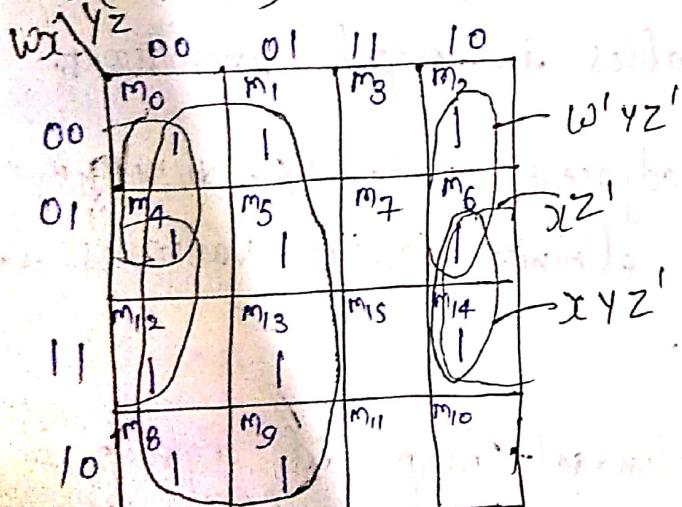
all elements of this group are covered by the adjacent groups. Minterm  $m_5$  is covered by pair  $G_2$ , similarly minterm  $m_7$  is covered by  $G_3$ , and so on. Hence,  $G_1$  becomes a redundant group and final output expression contains only four product terms instead of five terms.



Map for Redundant Group

# Simplify the Boolean function -

$$F(w,x,y,z) = \Sigma(0,1,2,4,5,6,8,9,12,13,14)$$



## Prime Implicants:

A prime implicant is a product term obtained by combining the maximum possible number of adjacent squares in the map. If a minterm is a square, it is covered by only one prime implicant. That prime implicant is said to be essential prime implicant.

The prime implicants of a function can be obtained from the map by combining all possible maximum numbers of squares. This means that a single  $1$  on a map represents a prime implicant if it is not adjacent of any other  $1$ 's. Two adjacent  $1$ 's form a prime implicant, provided that they are not within a group of four adjacent squares.

AB \ CD	00	01	11	10
00	$m_0$	$m_1$	$m_3$	$m_2$
01	$m_4$	$m_5$	$m_7$	$m_6$
11	$m_{12}$	$m_{13}$	$m_{15}$	$m_{14}$
10	$m_8$	$m_9$	$m_{11}$	$m_{10}$

AB \ CD	00	01	11	10
00	1		1	1
01		1	1	1
11		1	1	1
10	1	1	1	1

$$F(A, B, C, D) = \sum(0, 2, 3, 5, 7, 8, 9, 10, 11, 13, 15)$$

$$\text{Essential Prime Implicant} = BD + B'D'$$

$$\text{Prime Implicant} = \boxed{CD, B'C, AD, AB'}$$

$$F = BD + B'D' + CD + AD$$

$$F = BD + B'D' + CD + AB'$$

$$F = BD + B'D' + B'C + AB$$

$$F = BD + B'D' + B'C + AB'$$

## five-variable map:-

Maps more than four variables are not as simple to use as maps for four or fewer variables. A five-variable map needs  $32$  ( $2^5 = 32$ ) squares and a six-variable map needs  $64$  squares.

The five-variable map is shown in figure. It consists of two four-variable maps with variable A, B, C, D and E. Variable distinguishes between the two maps as indicated at the top of diagram. The left-hand four-variable map represents the 16 squares in which  $A=0$ , and the other four-variable map represents the squares in which  $A=1$ . Minterms 0 through 15 belong with  $A=0$  and minterms 16 through 31 with  $A=1$ .

A = 0					
BC	DE	00	01	11	10
00	m <sub>0</sub>	m <sub>1</sub>	m <sub>3</sub>	m <sub>2</sub>	0 1 3 2
01	m <sub>4</sub>	m <sub>5</sub>	m <sub>7</sub>	m <sub>6</sub>	4 5 7 6
11	m <sub>12</sub>	m <sub>18</sub>	m <sub>15</sub>	m <sub>14</sub>	12 13 15 14
10	m <sub>8</sub>	m <sub>9</sub>	m <sub>11</sub>	m <sub>10</sub>	8 9 11 10

A = 1					
BC	DE	00	01	11	10
00	m <sub>16</sub>	m <sub>17</sub>	m <sub>19</sub>	m <sub>18</sub>	16 17 19 18
01	m <sub>20</sub>	m <sub>21</sub>	m <sub>23</sub>	m <sub>22</sub>	20 21 23 22
11	m <sub>28</sub>	m <sub>29</sub>	m <sub>31</sub>	m <sub>30</sub>	28 29 31 30
10	m <sub>24</sub>	m <sub>25</sub>	m <sub>27</sub>	m <sub>26</sub>	24 25 27 26

## five-variable map

A = 0					
BC	DE	00	01	11	10
00		1			
01		1			
11		1			
10		1			

A'B'E'

A = 1					
BC	DE	00	01	11	10
00					
01		1	1		
11		1	1		
10		1	1		

ACE

$$BD'E' = A'B'E' + BD'E + ACE$$

## Product of Sum Simplification:-

If ~~sum~~ Product of sum form is given  $F(A,B,C,D) = \Sigma(0,1,2,5,8,9,10)$  then find out sum of product form.

$$F(A,B,C,D) = \Sigma(0,1,2,5,8,9,10)$$

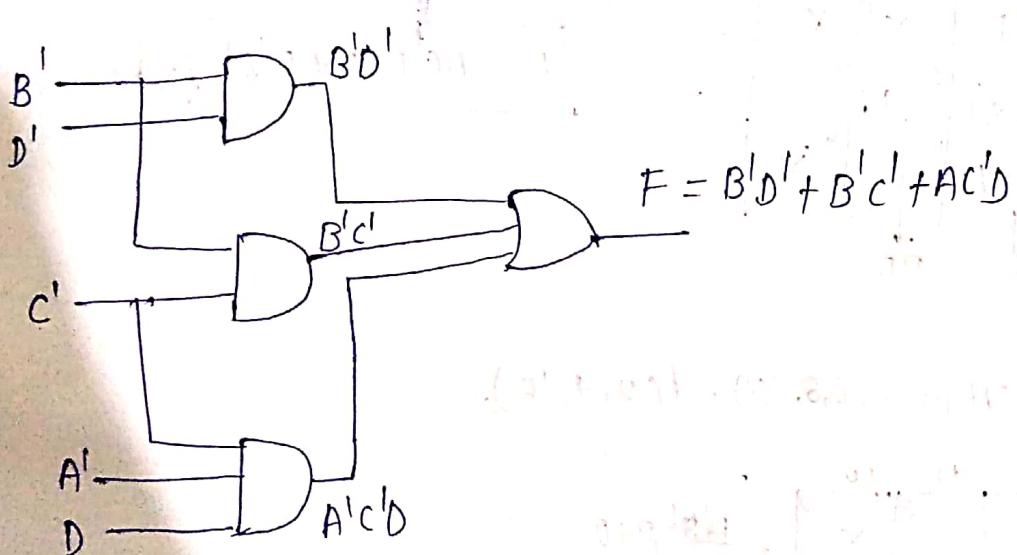
$$F(A,B,C,D) = \pi(3,4,6,7,11,12,13,14,15)$$

AB\CD	00	01	11	10	
00	1	1	0	1	$(C' + D')$
01	0	1	0	0	$(B' + D)$
11	0	0	0	0	$(A' + B')$
10	1	1	0	1	

$$F(\text{POS}) = (C' + D')(B' + D)(A' + B')$$

## Gate Implementation:-

$$F = B'D' + B'C' + A'C'D$$



## Don't-care Conditions:-

$\Sigma$ -SOP     $\Pi$ -POS

# Simplify the Boolean function:-

$$F(w, x, y, z) = \Sigma(1, 3, 7, 11, 15)$$

which has the don't-care conditions

$$d(w, x, y, z) = \Sigma(0, 2, 5)$$

w\z	00	01	11	10
00	X	1	1	X
01	4	X	1	6
11	12	13	1	14
10	8	9	11	10

$$F = YZ + w'x'$$

# Simplify the Boolean function:-

$$F(A, B, C, D) = \Sigma m(0, 2, 5, 9, 15) + d(6, 7, 8, 10, 12, 13)$$

which has

AB\CD	00	01	11	10
00	1	1	3	1
01	4	5	7	X
11	X	X	1	14
10	X	1	11	X

$$F = AC' + BD + B'D'$$

#  $F(A, B, C, D) = \Pi m(0, 6, 8, 13) \cdot d(2, 4, 10)$ .

AB\CD	00	01	11	10
00	0	1	3	X
01	X	5	7	0
11	12	13	15	14
10	0	9	11	X

$$F = (A'D') (B'D')$$

$$F = (A + D) (B + D) (A' + B')$$

$$\# F(A, B, C, D) = \sum(0, 2, 6, 8) + d(1, 3, 4, 10)$$

AB	CD	00	01	11	10
00	1	X	X	1	0
01	X			1	
11					
10	1		X		

$$F = A'B' + B'D' + A'D'$$

Ans

# The truth table represents the --- function -

- (a) x ✓
- (b) x+y
- (c) x⊕y
- (d) y

x	y	f(x,y)
0	0	0
0	1	0
1	0	1
1	1	1

# The simplified SOP of the boolean expression

$$(P+Q'+R')(P+Q'+R)(P+Q+R')$$

- (a) P'+Q+R'
- (b) P+Q'R'
- (c) P'Q+R
- (d) PQ+R

The given expression is in POS form

$$(P+Q'+R')(P+Q'+R) + (P+Q+R')$$

$\begin{matrix} 011 \\ 010 \\ 101 \\ 111 \end{matrix}$ 
  
 3      2      1

$$\prod(3, 2, 1)$$

$$\sum(0, 4, 5, 6, 7)$$

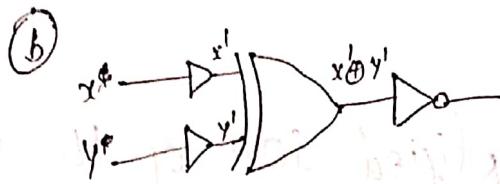
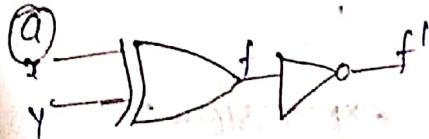
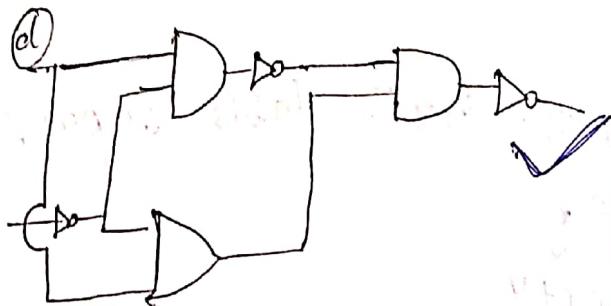
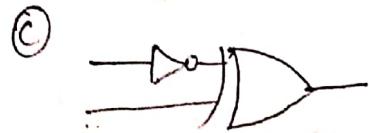
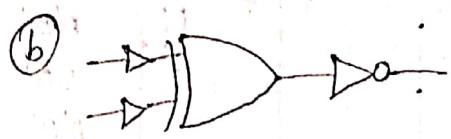
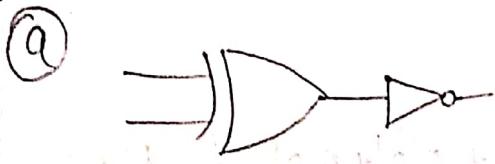
Total. 8 element

P	QR	00	01	11	10
Q'R'	00	1	1	1	0
Q'R	01	1	1	1	1
SOP	101	111	111	111	101

$$F = P + Q'R'$$

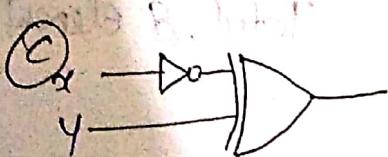
Ans

77 Which one of the following circles is not equivalent to the others?



$x$	$y$	$f$	$f'$	$-x'y'$
0	0	0	1	$-x'y'$
0	1	1	0	$-x'y'$
1	0	1	0	$-x'y'$
1	1	0	1	$-x'y$

$$f' = x'y + x'y'$$

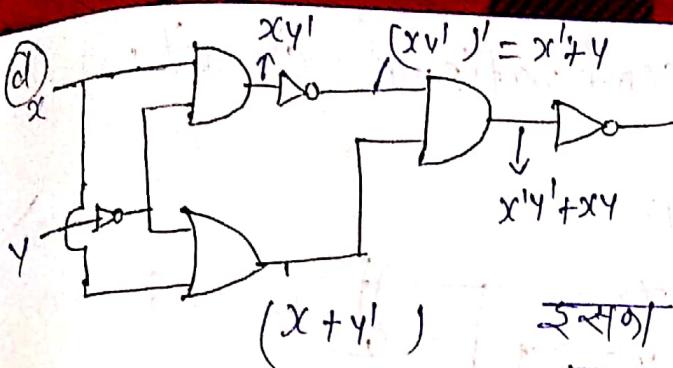


$x$	$y$	$x'$	XOR
0	0	1	1
0	1	1	0
1	0	0	0

$$F = xy + x'y'$$

					ExOR	XNOR
x	y	x'	y'	(x'y + x'y')		
0	0	1	1	0	1	-xy
0	1	1	0	1	0	
1	0	0	1	1	0	
1	1	0	0	0	1	-xy

$$F = xy + x'y'$$



$$\begin{aligned}
 & (x'+y)(x+y') \\
 & = 0 + x'y' + xy + 0 \\
 & = x'y' + xy
 \end{aligned}$$

गलत XNOR का Output है जब

इसका पुनः NOT लिया जायेगा तो  
गलत XOR Gate का जायेगा। यहाँ वह  
अब सभी से Different है।

# The minterm expansion of  $f(P, Q, R) = PQ + QR' + PR'$

(a)  $m_2 + m_4 + m_6 + m_7$

$$PQ + QR' + PR'$$

(b)  $m_0 + m_1 + m_6 + m_7$

$$= PQ(R+R') + QR'(P+P') + PR'(Q+Q')$$

(c)  $m_0 + m_1 + m_6 + m_7$

$$= PQR + \underline{PQR'} + \underline{PQR'} + P'QR' + \underline{PQR'} + \underline{PQR'}$$

(d)  $m_2 + m_3 + m_4 + m_5$

$$= PQR + PQR' + P'QR' + PQR'$$

$$= m_7 + m_6 + m_2 + m_4$$

$$= m_2 + m_4 + m_6 + m_7 \text{ Any}$$

## Logic Circuits

### Logic Circuits

#### Combinational Ckt.

Present output depends on present input only.

#### Sequential Ckt.

Present output depends on previous output as well as recent input.

$$A - B = A + (-B)$$

$$A \times B = A + A + A + \dots \text{ B Times}$$

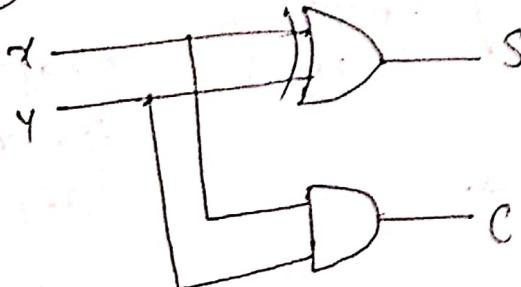
$$\frac{A}{B} = A - B$$

Half Adder:- It is a combinational circuit.

Truth Table

x	y	c	s
0	0	0	1
0	1	0	1
1	0	0	1
1	1	1	0

①



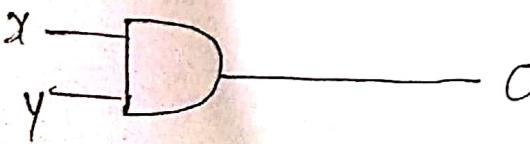
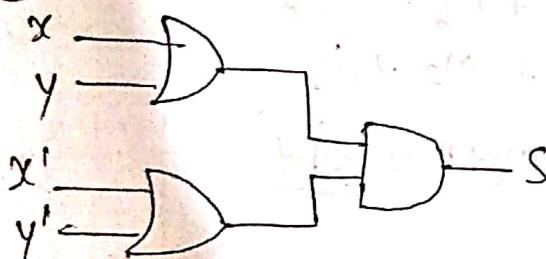
②

$$C = xy$$

$$S = x'y + xy'$$

③

②

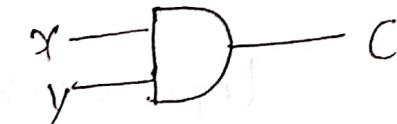
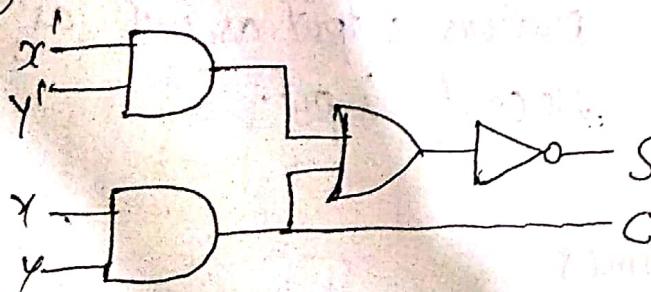


$$S = (x+y)(x'+y')$$

$$= \underline{\underline{x}n'} + \underline{\underline{xy'}} + \underline{\underline{yx'}} + \underline{\underline{yy'}}$$

$$S = xy' + x'y$$

④



$$S = xy' + x'y$$

$$S = [(x'y' + xy)]'$$

$$= (x'y')' \cdot (xy)'$$

$$= [(x')' + (y')'] \cdot [x' + y']$$

$$= (x+y)(x'+y')$$

$$= x'n' + xy' + x'y + yy'$$

$$S = xy' + x'y$$

# Full Adder!

## Truth Table

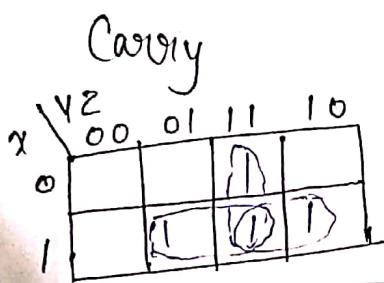
x	y	z	c	s
0	0	0	0	0 $m_0$
0	0	1	0	1 $m_1$
0	1	0	0	1 $m_2$
0	1	1	1	0 $m_3$
1	0	0	0	1 $m_4$
1	0	1	1	0 $m_5$
1	1	0	1	0 $m_6$
1	1	1	1	1 $m_7$

$$C = \Sigma(m_3, m_5, m_6, m_7)$$

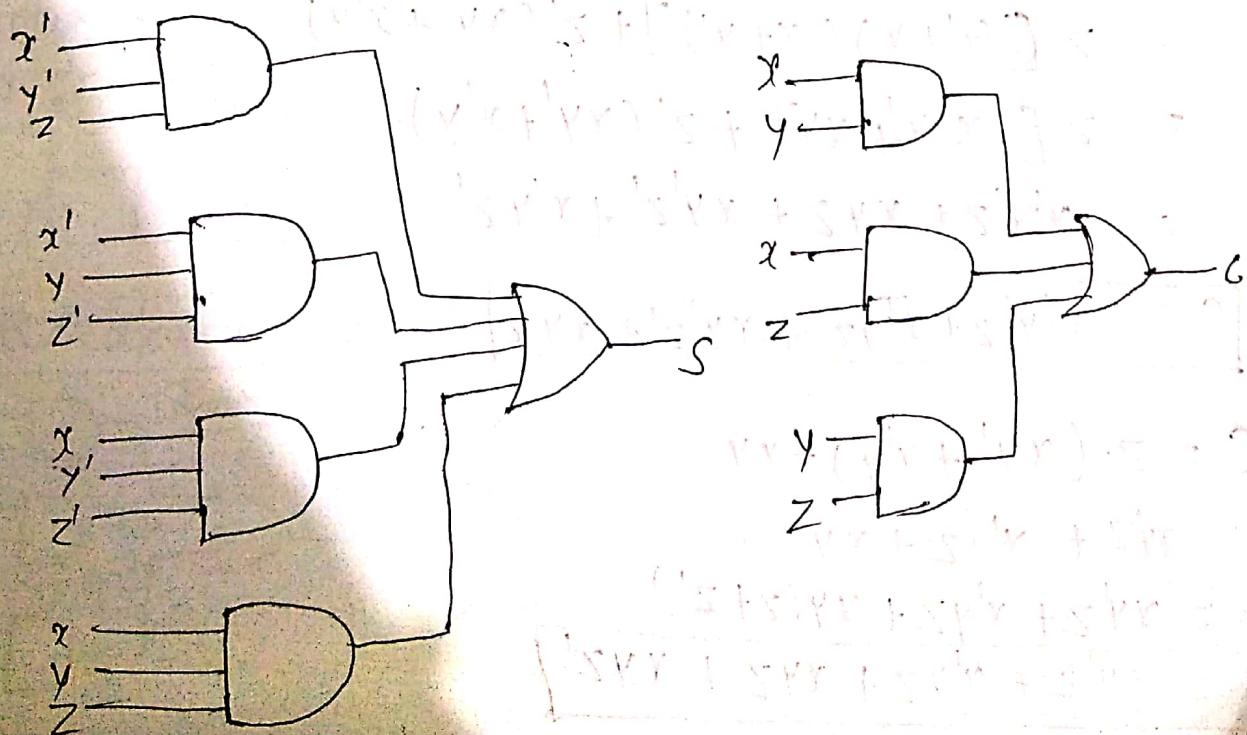
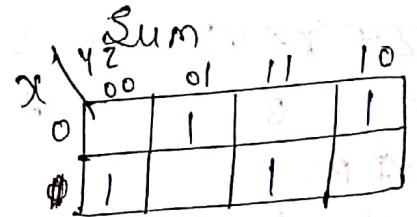
$$C = x'y'z + xy'z + xyz' + xyz$$

$$S = x'y'z + x'y'z' + xy'z' + xyz$$

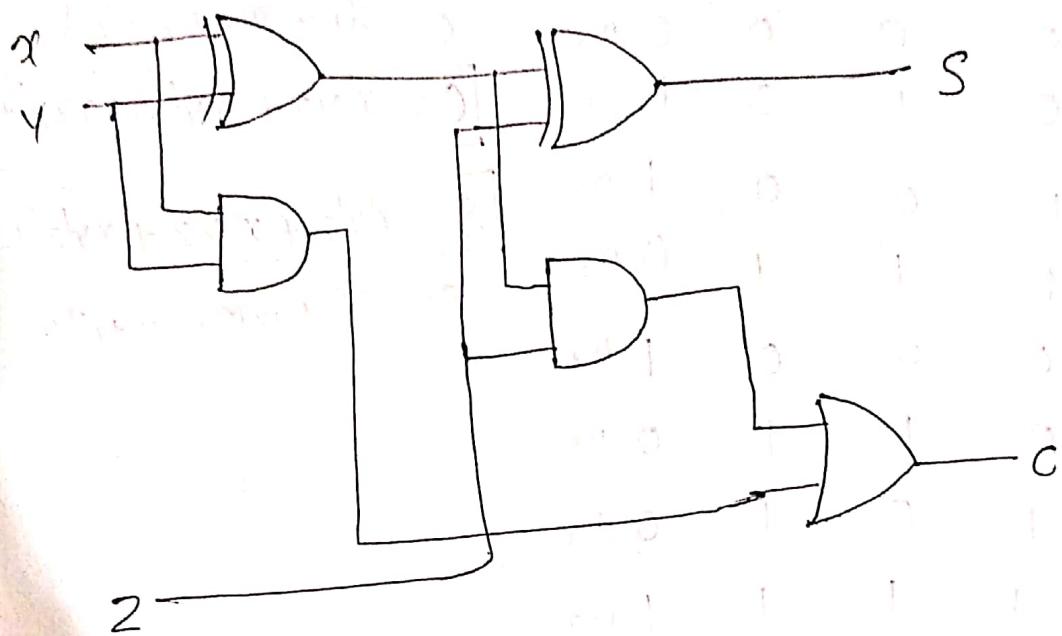
$$S = \Sigma(m_1, m_2, m_4, m_7)$$



$$C = x'y + xz + yz$$



## Implementing full Adder using half Adder:-



$$S = \underset{A}{z} \oplus \underset{\overline{B}}{(x \oplus y)} = \underset{A}{z} \oplus \underset{B}{(xy' + x'y)}$$

$$A \oplus B$$

$$AB' + A'B$$

$$= z[(xy' + x'y)]' + z'(axy' + x'y)$$

$$= \underline{z}[(x'y')' \cdot (x'y')'] + z'(xy' + x'y)$$

$$= z[(x'+y)(x+y')] + z'(xy' + x'y)$$

$$= z[x'y' + xy] + z'(xy' + x'y)$$

$$= x^4 y' z + x y z + x y' z' + x' y z'$$

$$S = x'y'z + x'y'z' + xy'z' + xyz$$

$$C = Z(x'y' + x'y) + xy$$

$$=xyz + x'yz + xy$$

$$= xy'z + x'y'z + xy(z+z')$$

$$C = xyz' + x'yz + xy'z + xy'z'$$

## Subtractor:-

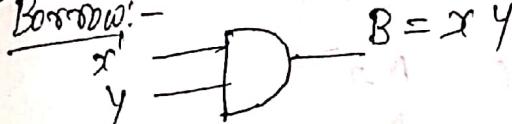
### ① Half Subtractor:-

x	y	B	D
0	0	0	0
0	1	1	1
1	0	0	1
1	1	0	0

$$B = x'y$$

$$D = x'y + xy'$$

Ckt:-



Borrow:-



### Difference

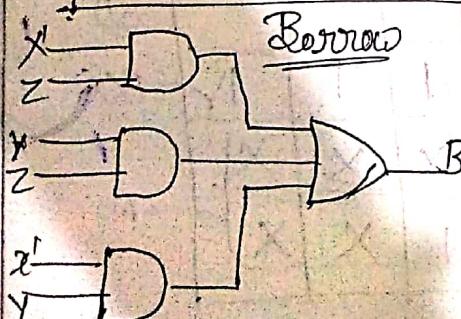
x	y	z	00	01	11	10
0	1	1	1	1	1	1
1	1	1	1	1	1	1

Borrow:-

x	y	z	00	01	11	10
0	1	1	1	1	1	1
1	1	1	1	1	1	1

$$B = xz + yz + xy$$

Borrow



## Full Subtractor:-

$$x - y - z = x - (y+z)$$

x	y	z	B	D
0	0	0	0	0
0	0	1	1	1
0	1	0	1	1
0	1	1	0	0
1	0	0	1	1
1	0	1	0	0
1	1	0	0	0
1	1	1	1	1

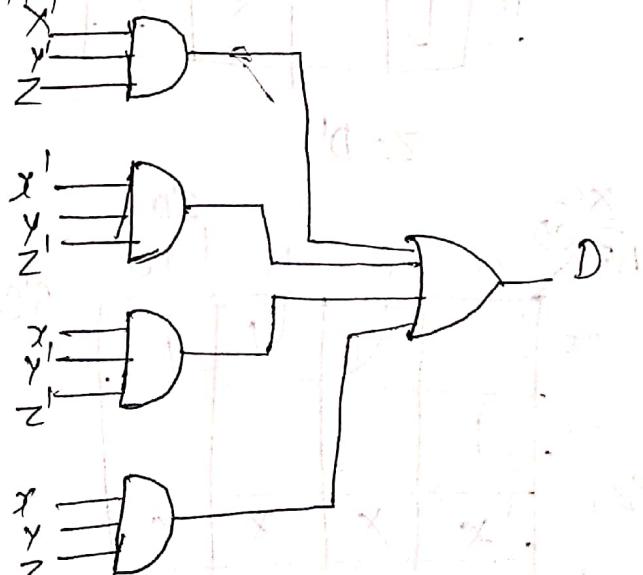
$$B = x'y'z + x'y'z' + x'y'z + x'y'z$$

$$B = m_1 + m_2 + m_3 + m_7$$

$$D = m_1 + m_2 + m_4 + m_7$$

$$D = x'y'z + x'y'z' + x'y'z + x'y'z$$

### Difference



## Code Conversion

Input BCD				Output Excess-3 Code			
A	B	C	D	W	X	Y	Z
0	0	0	0	0	0	1	1
0	0	0	1	0	0	0	$m_1$
0	0	1	0	0	1	0	$m_2$
0	0	1	1	0	1	1	$m_3$
0	1	0	0	0	1	1	$m_4$
0	1	0	1	0	1	0	$m_5$
0	1	1	0	1	0	1	$m_6$
0	1	1	1	1	0	0	$m_7$
1	0	0	0	1	0	1	$m_8$
1	0	0	1	1	0	0	$m_9$

Z'

AB	CD	00	01	11	10
00	- $m_6$	$m_1$	$m_3$	$m_4$	$m_9$
01	$m_4$	$m_5$	$m_7$	$m_6$	
11	X	X	X	X	
10	$m_8$	$m_9$	X	X	$m_{10}$

$$Z = D'$$

AB	CD	00	01	11	10
00					
01					
11	X	X	X	X	
10					

$$X = -B'C + B'D + B'C'D'$$

$$Z = m_0 + m_2 + m_6 + m_8$$

AB	CD	00	01	11	10
00					
01					
11	X	X	X	X	
10					

$$Y = CD + C'D' = CD + (C+D)'$$

W

AB	CD	00	01	11	10
00					
01					
11	X	X	X	X	
10					

$$W = A + BC + BD$$

$$z = D'$$

$$y = CD + C'D' = CD + (C+D)'$$

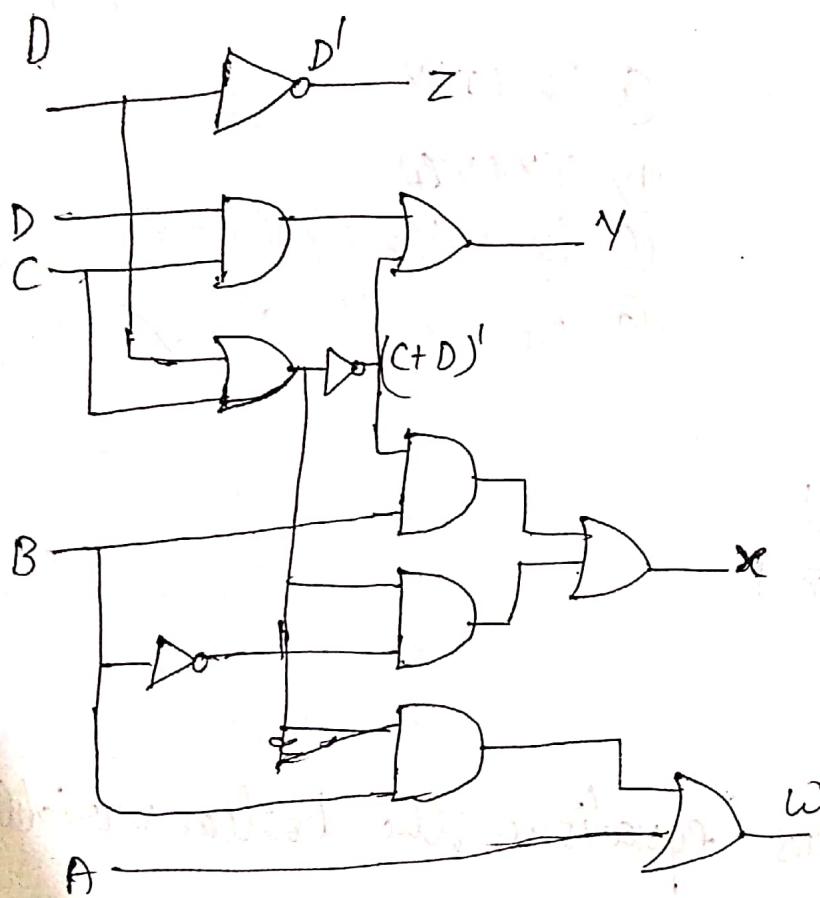
$$x = B'C + B'D + BC'D'$$

$$= B'(C+D) + BC'D'$$

$$= B'(C+D) + B(C+D)'$$

$$w = A + BC + BD$$

$$= A + B(C+D)$$



Q:- Given the following K-map which one of the following represents the minimal SOP of the map

$wz$	00	01	11	10
00	0	X	0	X
01	X	1	X	1
11		X	1	
10		1	X	

(A)  $w'y + y'z$  ✓

(B)  $w'x'y' + xy + xz$

(C)  $w'x + y'z + xy$

(D)  $xz + y$

Q:- Which function does not implement the K-map given below -

$wz$	00	01	11	10
xy	00	X	0	0
00	1	0	1	1
01	0	X	1	1
11	1	1	1	1
10	0	X	0	0

(A)  $(w+x)y$

(B)  $xy + yz$  ✓

(C)  $(w+x)(\bar{w}+y)(\bar{x}+y)$

(D) None of the above ✓

(C)  $(w+x)(\bar{w}+y)(\bar{x}+y)$

$= (w+x)(y+\bar{w})(y+\bar{x})$

$= (w+x)(y+\bar{w}\bar{x})$

Q:- The simultaneous equations the boolean variables  $x, y, z$ , and  $w$  :-

$$x+y+z=1$$

$$xy=0$$

$$xz+w=1$$

$$xy+\bar{z}\bar{w}=0$$

having the following solution for  $x, y, z$  and  $w$  respectively -

(A) 0100 ✓

(B) 1101 X

(C) 1011 ✓

(D) 1000 X

$$\sum \bar{w} = 0$$

$$\bar{z} \neq 1, \bar{w} \neq 1$$

$$z \neq 0, w \neq 0$$

Self dual function-

A boolean function is said to be self dual if and only if its dual is equivalent to the given function.

$$f(x, y, z) = xy + yz + zx$$

$$f_d(x, y, z) = (x+y)(y+z)(z+x)$$

$$xy + yz + zx$$

$$= xy(z+z') + yz(x+x') + zx(y+y')$$

$$= xyz + x(yz + x'y) + xyz + xy'z$$

$$= m_3 + m_5 + m_6 + m_7$$

$$= \Sigma(3, 5, 6, 7)$$

$$(x+y)(y+z)(z+x)$$

$$= (x+y+zz')(yz + zx + xx') + (x+yy'+z)$$

$$= (x+y+z)(x+y+z)(x+y+z)(x+y+z)$$

$$= (x+y+z)(x+y+z)(x+y+z)(x+y+z)$$

$$= M_0 M_4 M_2 M_1$$

$$= M_0 M_1 M_2 M_4$$

$$= \Pi(0, 1, 2, 4)$$

$$= \Sigma(3, 5, 6, 7)$$

Hence above function is self dual function.

A boolean function is self dual if -  $2^{n-1}$

- ① It is neutral (No. of minterms = No. of maxterms)  $\checkmark$
- ② The function does not contain two mutually exclusive terms.  $\checkmark$

$$\text{Given } f = \sum_{m_0}^{m_7} x'y'z + x'y'z' + x'yz + x'yz' + xy'z + xy'z' + xyz + xyz'$$

$\Rightarrow f = \sum_{m_0}^{m_7} m_0 + m_1 + m_2 + m_3 + m_4 + m_5 + m_6 + m_7$

इसमें से कोई ऐसा लिना है कि  $m_0 + m_7 = m_1 + m_6 = m_2 + m_5 = m_3 + m_4$

# For the 3 variable function :-  
 $f = \sum(0,1,2,4)$  its self dual.

$$n = 3$$

$$2^{n-1} = 2^2 = 4$$

This function is self dual function.

How many self dual function are possible for n boolean variables:-

n variable के case के total function =  $2^{2^n}$  possible boolean

x	y	z
0	0	0
0	0	1
0	1	0
-	-	-
1	1	1

$$\begin{aligned}
 & (m_0, m_7) \quad m_0, m_1, m_2, m_3 \\
 & (m_1, m_6) \quad m_0, m_6, m_2, m_3 \\
 & (m_2, m_5) \quad m_7, m_1, m_5, m_4 \\
 & (m_3, m_4) \\
 & = 2^4
 \end{aligned}$$

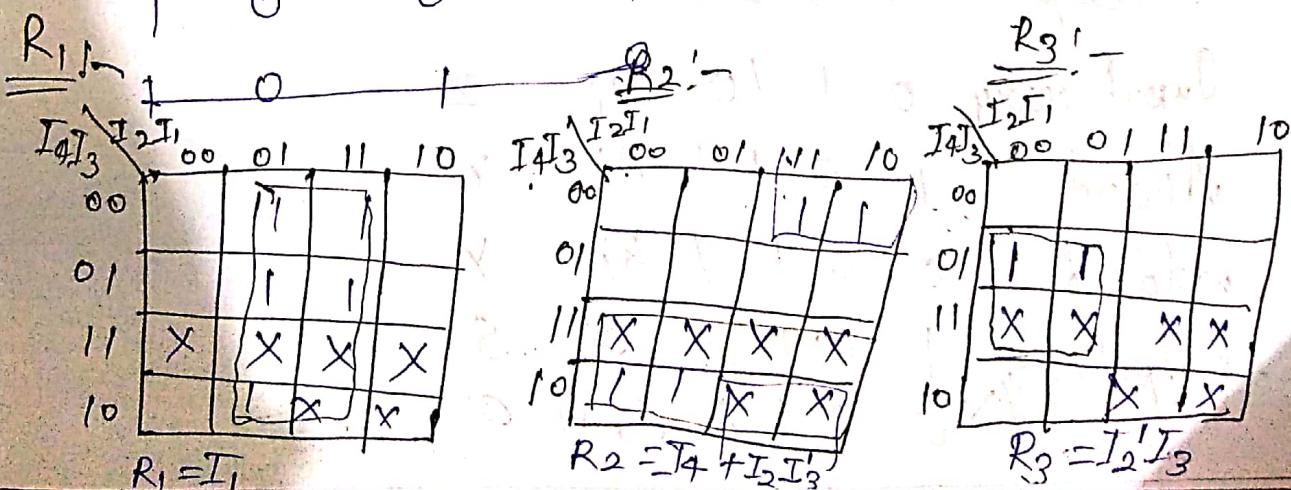
$\text{No. of self dual function} = 2^{2^{n-1}}$

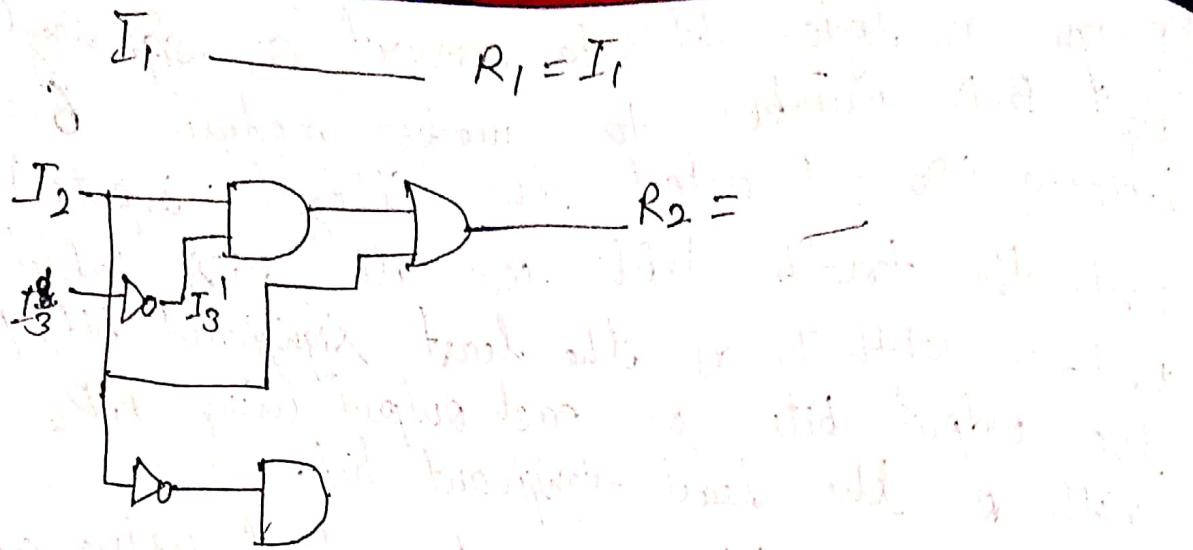
Q7 Design a logic ckt. to convert a single digit BCD number to number modulo 6 as booleans (Do not detect the illegal inputs)

- (a) Write the truth table for all bits labeled the input  $I_1, I_2, \dots$  with  $I_1$  as the least significant bit, Label the output bits for each output using  $R_1, R_2, \dots$  with  $R_1$  the least significant bit.
- (b) Draw one circuit for each output using altogether two input

modulo  $6 \Rightarrow 0-5$

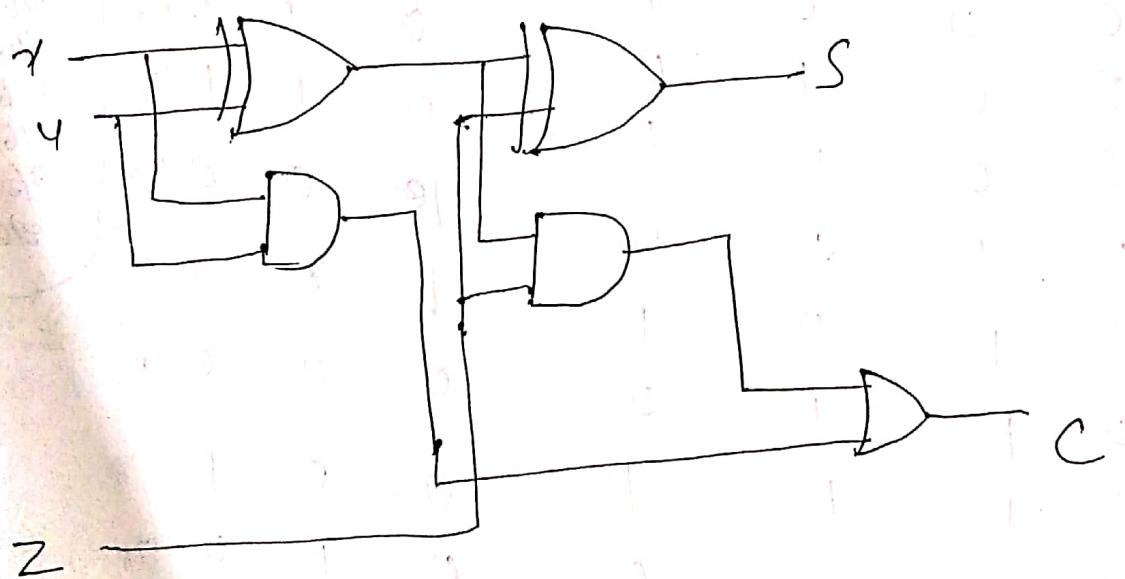
$I_4$	$I_3$	$I_2$	$I_1$	$R_3$	$R_2$	$R_1$	
0	0	0	0	0	0	0	$0/6=0$
0	0	0	1	0	0	1	$1/6=1$
0	0	1	0	0	1	0	$2/6=2$
0	0	1	1	0	1	1	$3/6=3$
0	1	0	0	1	0	0	$4/6=4$
0	1	0	1	1	0	1	$5/6=5$
0	1	1	0	0	1	1	$6/6=0$
1	0	0	0	0	0	1	$7/6=1$
1	0	0	1	0	1	0	$8/6=2$
1	0	1	0	1	0	1	$9/6=3$





## Binary Adder & Subtractor

full Adder using half Adder and OR Gate



Subscript 4 3 2 1 Full Adder

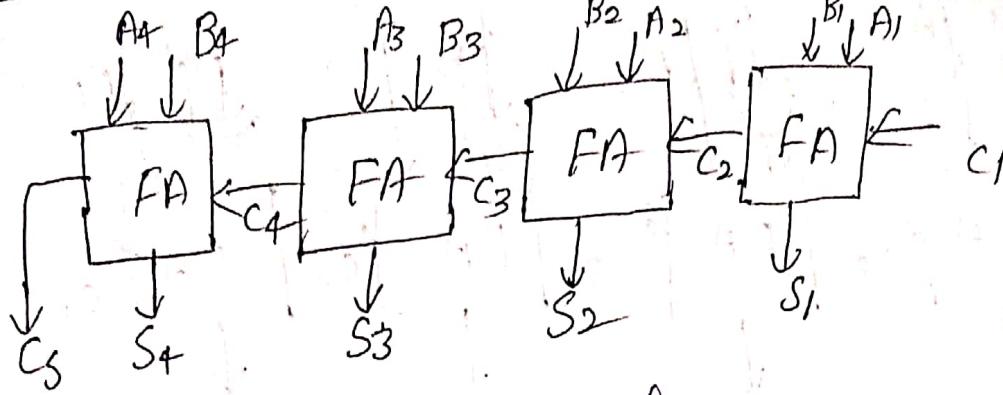
Input carry 0 1 10  $C_i$  2

Addend 1 0 1 1  $A_i$  2

Addend 0 0 1 1  $B_i$  2

Output carry 1 1 1 0  $S_i$  2

0 0 1 1  $C_i$  2

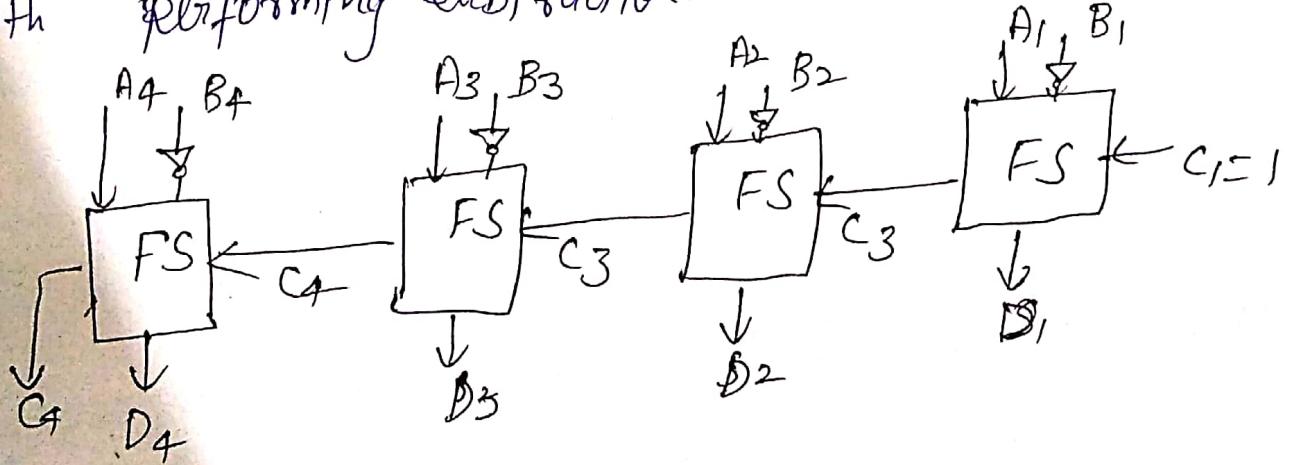


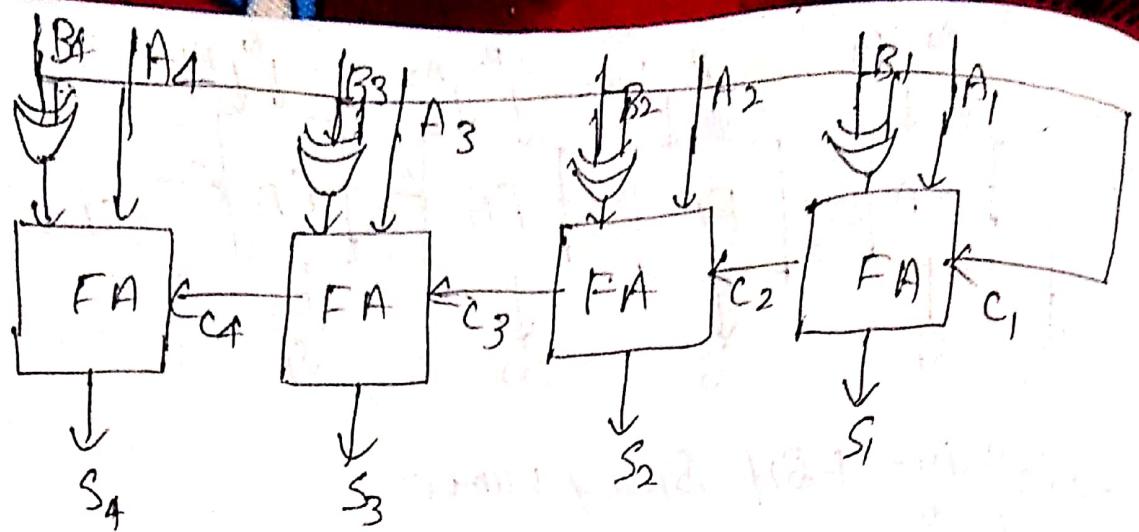
$A_4 A_3 A_2 A_1 - 4 \text{ Bit Binary number}$   
 $+ B_4 B_3 B_2 B_1$

Subtractor:-

$$\begin{aligned} & A - B \\ & A + (2^{\text{'}}\text{s complement of } B) \\ & A + (1^{\text{'}}\text{s comp.} + 1) \end{aligned}$$

The ckt. for subtractor  $A - B$  consists of a parallel adder with inverters placed between each data input  $B$ . The input carry  $C_1$  must be equal to 1 with performing subtraction.





$$\begin{array}{r}
 B_4 \quad B_3 \quad B_2 \quad B_1 \\
 \oplus \quad \oplus \quad \oplus \quad \oplus \\
 0 \quad 0 \quad 0 \quad 0 \\
 \hline
 B_4 \quad B_3 \quad B_2 \quad B_1
 \end{array}$$

$$\begin{array}{l}
 x \oplus 0 = x \\
 1 \oplus 0 = 1 \\
 0 \oplus 0 = 0
 \end{array}$$

$$\begin{array}{r}
 B_4 \quad B_3 \quad B_2 \quad B_1 \\
 \oplus \quad \oplus \quad \oplus \quad \oplus \\
 1 \quad 1 \quad 1 \quad 1 \\
 \hline
 B_4' \quad B_3' \quad B_2' \quad B_1'
 \end{array}$$

$$\begin{array}{l}
 x \oplus 1 = x' \\
 0 \oplus 1 = 1 \\
 1 \oplus 1 = 0
 \end{array}$$

$M=1$  Then Adder

$M=0$

functionally Complete Operations—

A set of operations is said to be functionally complete or universal if and only if every switching function can be expressed by means of operations in it.

The set of  $\{AND, OR, NOT\}$  is clearly functionally complete.

The set of  $\{AND, NOT\}$  is also functionally complete.

The set of  $\{OR, NOT\}$  is also functionally complete.

A set is said to be functionally complete if we can derive a set which is already functionally complete.

$\{AND, NOT\}$  is also functionally complete.

$$(A + B)' = (\overline{A} \cdot \overline{B})' = (\overline{\overline{A}}) + (\overline{\overline{B}}) = A + B$$

$\{OR, NOT\}$  is also functionally complete.

$$\overline{AB} = (\overline{A} + \overline{B})' = \overline{\overline{A} \cdot \overline{B}} = AB$$

# Show that  $f(A, B, C) = A' + BC'$  is functionally complete

$$f(A, A, A) = A' + AA' = A'$$

$$f(B, B, B) = B' + BB' = B'$$

$$f(C, C, C) = C' + CC' = C'$$

$$f(f(A, A, A), B, f(B, B, B)) = (A')' + B(B')' \\ = A + B \cdot B \\ = A + B$$

Hence it is F.C.

# Show that  $\{I, \bar{A}B\}$  is functionally complete.

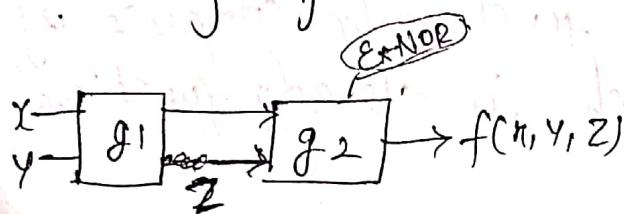
$$f(A, B) = \bar{A}\bar{B}$$

$$f(I, A) = I \cdot \bar{A} = \bar{A}$$

$$f(A, f(I, B)) = A(\bar{B}) = AB$$

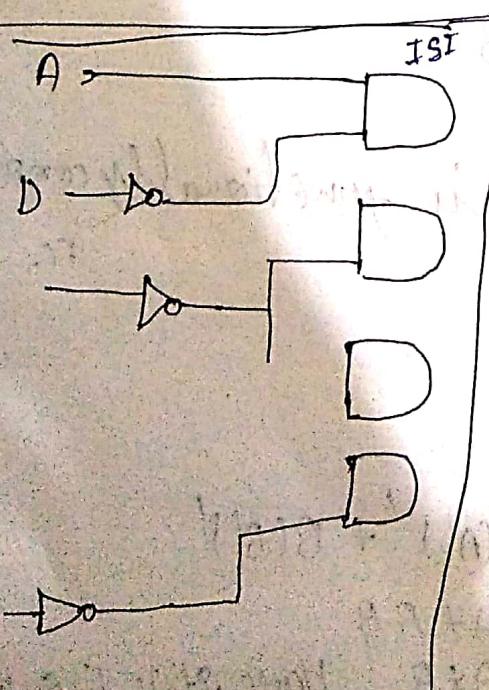
Ex-NOR	
0	0 $\rightarrow$ 1
0	1 $\rightarrow$ 0
1	0 $\rightarrow$ 0
1	1 $\rightarrow$ 1

# In the combinational ckt shown below  $g_1$  is an unknown logic gate and  $g_2(A, B) = AB + \bar{A}\bar{B}$ . A partial truth table of the boolean function specified by the circuit is given below. Complete the table with justification.

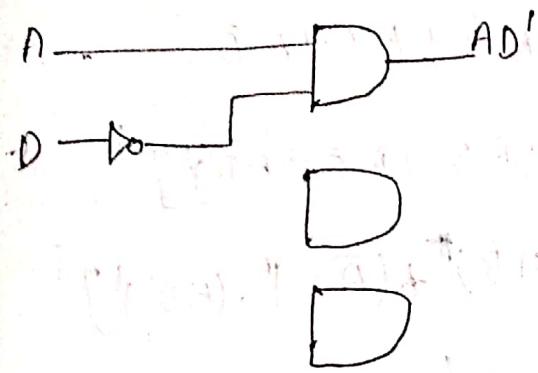


x	y	z	f
0	0	1	1
0	0	1	0
1	1	0	0
0	0	0	1

x	y	z	f
0	0	0	0
0	1	0	1
1	0	0	0
1	1	0	1
0	0	1	1
0	1	1	0
1	0	1	1
1	1	1	0



Q3 Assuming that all gates cost Rs 5 per input find the minimum cost realization F using only Inverters, AND/OR gates.



$$\begin{aligned}
 &AD' \\
 &AC \\
 &\bar{A}\bar{C}D \\
 &\bar{A}\bar{B}C
 \end{aligned}$$

D

$$F = AD + \bar{A}\bar{C} + \bar{A}\bar{C}\bar{D} + \bar{A}\bar{B}C$$

$$AD = AD(B+\bar{B}) = A\bar{D}B + A\bar{D}\bar{B} = A\bar{D}B(\cancel{A} + \cancel{\bar{A}}) + A\bar{D}\bar{B}(\cancel{C} + \cancel{\bar{C}})$$

$$= ABC\bar{D} + A\bar{B}\bar{C}\bar{D} + A\bar{B}C\bar{D} + A\bar{B}\bar{C}\bar{D}$$

$$= m_4 + m_{12} + m_{10} + m_8$$

$$\bar{A}\bar{C} = \bar{A}\bar{C}(B+\bar{B}) = \bar{A}\bar{C}B + \bar{A}\bar{C}\bar{B} = \bar{A}\bar{C}B(D+\bar{D}) + \bar{A}\bar{C}\bar{B}(D+\bar{D})$$

$$= \bar{A}\bar{B}\bar{C}\bar{D} + \bar{A}\bar{B}\bar{C}D + \bar{A}\bar{B}\bar{C}\bar{D}$$

$$= m_5 + m_4 + m_1 + m_0$$

$$\bar{A}\bar{C}\bar{D} = \bar{A}\bar{C}\bar{D}(B+\bar{B}) = \bar{A}\bar{B}\bar{B}\bar{D} + \bar{A}\bar{C}\bar{B}\bar{D}$$

$$= m_2 + m_0$$

$$\bar{A}\bar{B}C = \bar{A}\bar{B}C(D+\bar{D}) = \bar{A}\bar{B}CD + \bar{A}\bar{B}\bar{C}\bar{D}$$

$$= m_3 + m_2$$

$$F = m_4 + m_{12} + m_{10} + m_8 + m_5 + m_4 + m_1 + m_0 + m_4 + m_{10} + m_{12} + m_{14}$$

$$+ m_2$$

$$= m_0 + m_1 + m_2 + m_3 + m_4 + m_5 + m_8 + m_{10} + m_{12} + m_{14}$$

A	B	C	D	CD
00	0	0	0	00
01	0	0	1	01
11	1	1	0	11
10	1	0	1	10

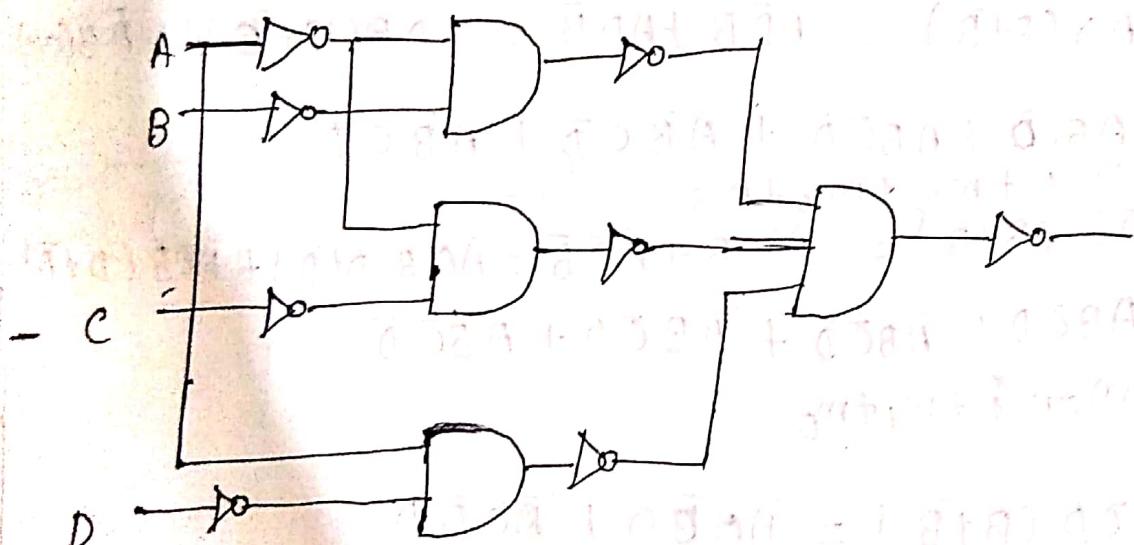
$$f = \bar{A}\bar{B} + \bar{A}\bar{C} + A\bar{D}$$

$$f = [(\bar{A}\bar{B} + \bar{A}\bar{C} + A\bar{D})']'$$

$$= [(\bar{A}\bar{B})' + (\bar{A}\bar{C})' \cdot (A\bar{D})']'$$

$$= ((\bar{A} + \bar{B})(\bar{A} + \bar{C})(\bar{A} + \bar{D})')$$

$$= (\bar{A} + \bar{B})(\bar{A} + \bar{C})(\bar{A} + \bar{D})$$



# Consider a four input hypothetical logic gate growth the following K-map.

	CD	00	01	11	10
AB	00	1			
	01		1		
	11			1	
	10				1

① Prove or disprove that  $C_4$  can be treated as a universal logic gate. A large no. of ~~and~~ functions are available to you.

② Implement  $A+B$  using minimum no. of gates.

$$\text{① } g = \bar{A}\bar{B}\bar{C}\bar{D} + \bar{A}B\bar{C}D + A\bar{B}CD + A\bar{B}\bar{C}\bar{D}$$

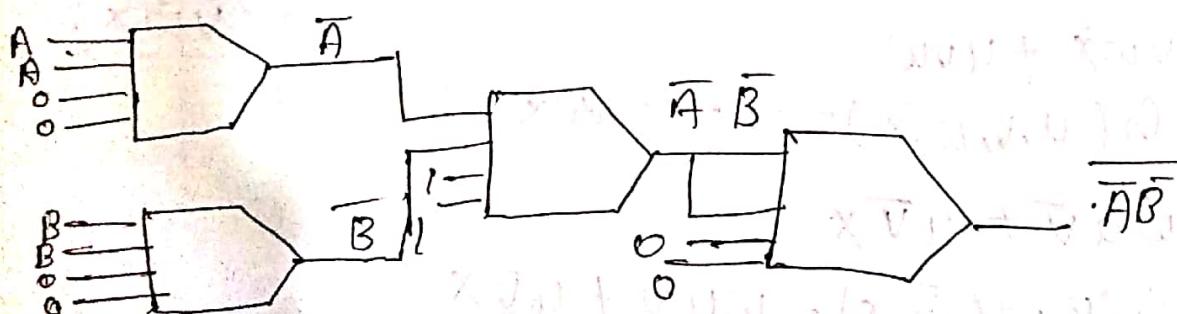
$$\begin{aligned} g(A, A, A, A) &= \overbrace{\bar{A}\bar{A}\bar{A}\bar{A}}^0 + \overbrace{A\bar{A}A\bar{A}}^0 + \overbrace{A\bar{A}A\bar{A}}^0 + \overbrace{A\bar{A}A\bar{A}}^0 \\ &= \bar{A} + 0 + A + 0 \\ &= \bar{A} + A \\ &= 1 \quad \text{Not possible} \end{aligned}$$

$$\begin{aligned} g(A, A, 0, 0) &= \bar{A}\bar{A}\cdot 1\cdot 1 + \bar{A}A\cdot 1\cdot 0 + AA\cdot 0\cdot 0 + A\bar{A}\cdot 0\cdot 1 \\ &= \bar{A} \end{aligned}$$

$$\begin{aligned} g(A, B, 1, 1) &= \bar{A}\bar{B}\cdot 0\cdot 0 + \bar{A}B\cdot 0\cdot 1 + AB\cdot 1\cdot 1 + A\bar{B}\cdot 1\cdot 0 \\ &= AB \end{aligned}$$

It is universal gate.

$$\text{② } (\overline{A \oplus B}) = \overline{\bar{A} + \bar{B}} = A + B$$



# A certain four-input gate  $G_1$  realizes the switching  $G_1(a, b, c, d) = abct + bcd$ . Assuming that the input variable  $a, b, c$  are a variable in both complemented and uncomplemented form.

① Show a realization of the function

$$f(u, v, w, x) = \sum(0, 1, 6, 9, 10, 11, 14, 15)$$

With only three  $G_1$  gates and one OR gate  
② Can all switching functions be realized with  $\{G_1, OR\}$  logic?

$uv$	$wx$	$\bar{u}\bar{v}\bar{w}$	$\bar{u}v\bar{w}$	$\bar{u}v\bar{x}$	$\bar{u}\bar{v}x$
00	00	11	10	01	11
01	11	01	00	10	01
10	10	00	11	11	10
11	01	10	01	00	11

$$f = \bar{u}\bar{v}\bar{w} + vw\bar{x} + u\bar{v}x + uw$$

$$= \bar{u}\bar{v}\bar{w} + vw\bar{x} + u\bar{v}x + u(w(v+\bar{v}))$$

$$= \bar{u}\bar{v}\bar{w} + vw\bar{x} + u\bar{v}x + \cancel{uw} + \cancel{uw} + u\bar{v}w$$

$$vw\bar{x} + uw$$

$$G_1(u, v, w, \bar{x}) = uvw + vw\bar{x}$$

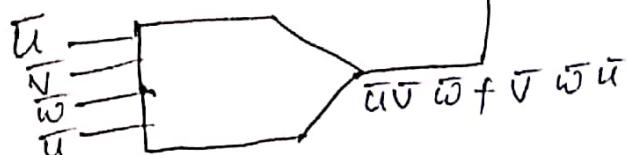
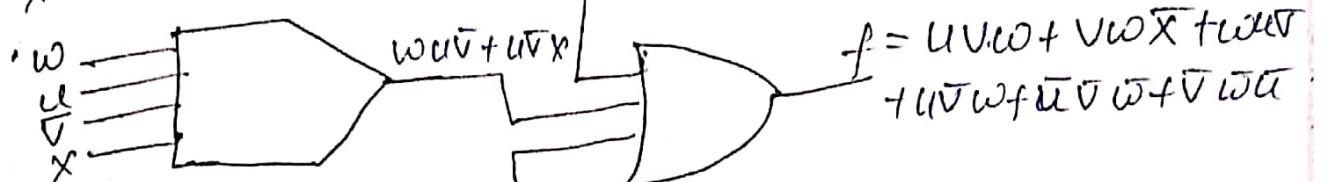
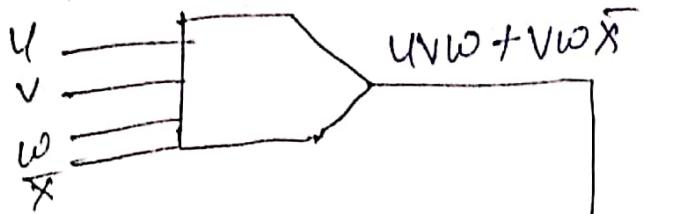
$$\boxed{vw\bar{x}}$$

$$uw + u\bar{v}x$$

$$G_1(w, u, \bar{v}, x) = uw\bar{v} + u\bar{v}x$$

$$\bar{u}\bar{v}\bar{w} + \cancel{\bar{u}\bar{v}w} \leftarrow \text{cancel}$$

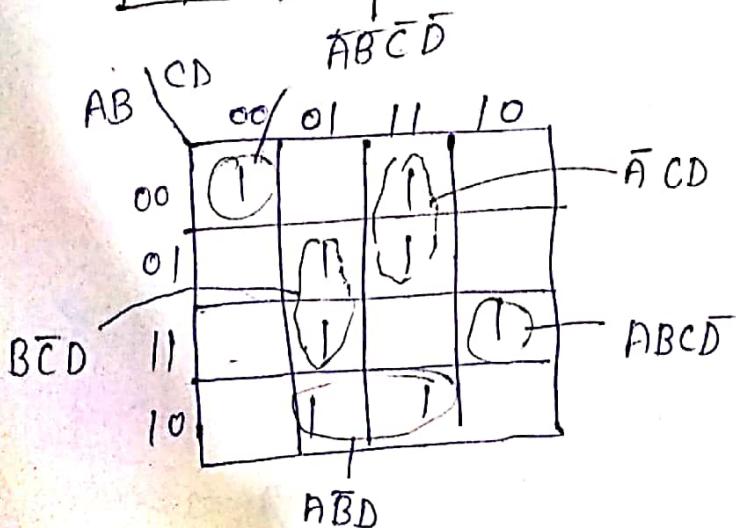
$$G_1(\bar{u}, \bar{v}, \bar{w}, \bar{x}) = \bar{u}\bar{v}\bar{w} + \bar{v}\bar{w}\bar{u}$$



$$f = UVWU + VVWX + VWXU + UVWU + UVWU + VVWX$$

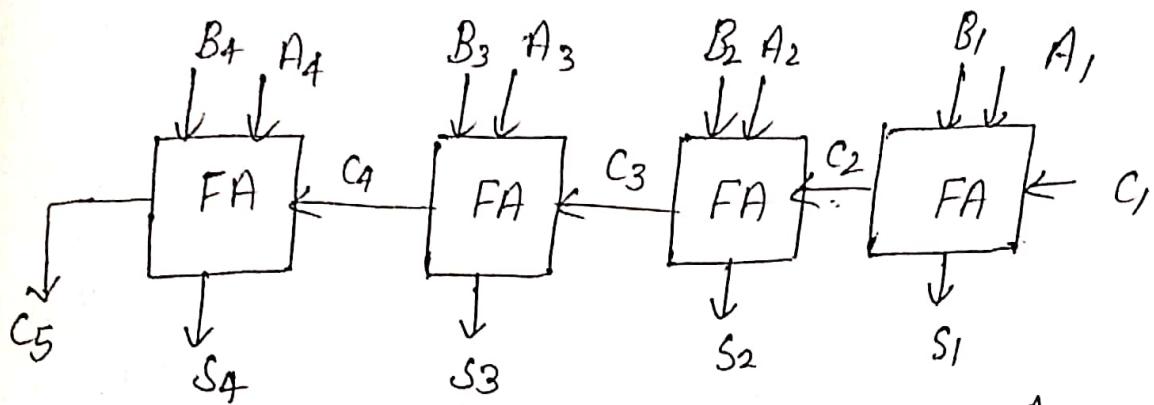
# Given a library of 2 input AND, NOT and 2 input XOR gates, synthesize the function  $f(A, B, C, D)$  as shown in the Karnaugh map using minimum no. of gates of the library.

AB \ CD	00	01	11	10
00	1		1	1
01		1	1	
11		1		1
10	1	1	1	

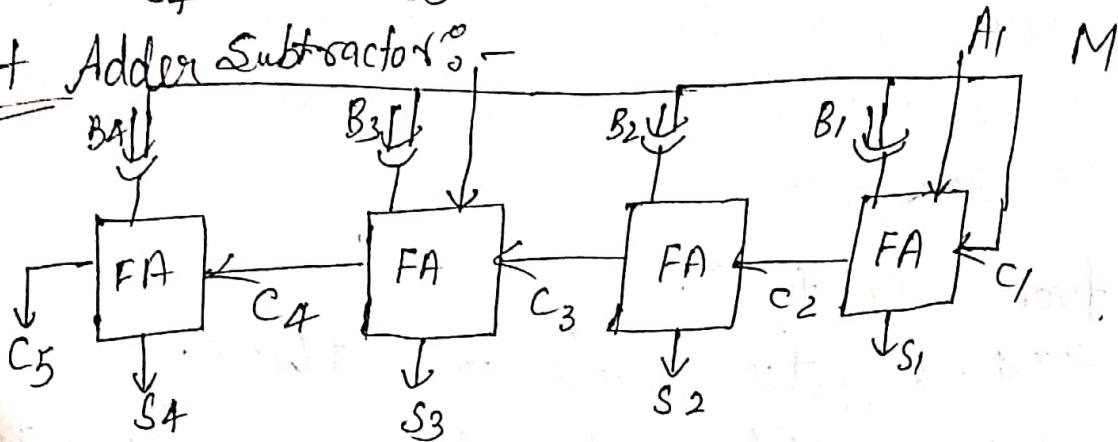


$$f = \bar{A}\bar{B}\bar{C}\bar{D} + B\bar{C}D + A\bar{B}D + \bar{A}CD + ABC\bar{D}$$

## Parallel Adder:-



## 4-bit Adder Subtractor:-



1 - Subtractor

0 - Adder

In any combinational circuit the signal must propagate through the gates before the correct output is available at the terminals.

Total propagation time is equal to the propagation delay of a typical gate times the number of gate levels in the circuits. The longest propagation delay time in a parallel adder is the time it takes the carry to propagate through the parallel adder.

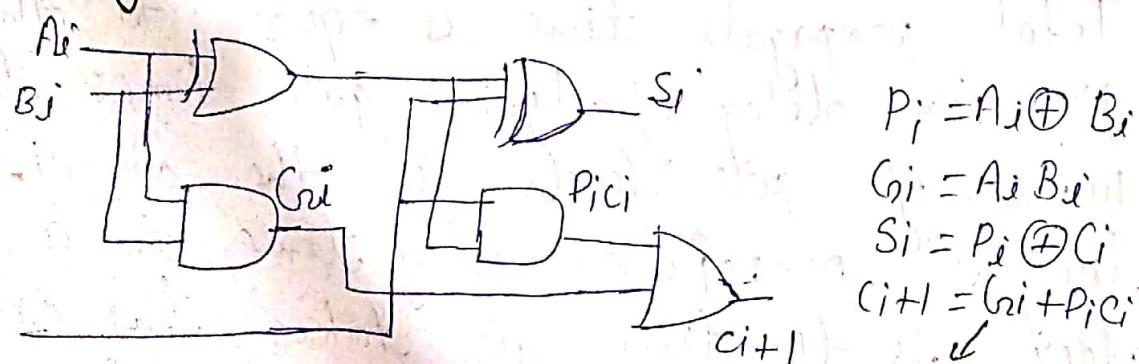
The signal from the input carry  $c_i$  total output carry  $c_{i+1}$  propagates through an AND gate and an OR gate.

from  $C_1$  to  $C_5$ .

$2 \times 4 = 8$  gate levels are there.

for  $n$  bit parallel adder there are  $2^n$  gate levels for the carry to propagate through.

→ There are several techniques for reducing the carry propagation time in a parallel adder. The most widely used technique types the principal of carry look ahead.



$G_{ij}$  is called carry generate

$G_i$  is called carry generate and it produces an output carry when both  $A_i$  and  $B_i$  are 1, regardless the sum carry  $C_i$ .

$P_i$  is called a carry propagate because it is the term associated with the propagation of carry from  $C_i$  to  $C_{i+1}$ .

$$C_2 = G_1 + P_1 C_1$$

$$C_3 = G_2 + P_2 C_2$$

$$= G_2 + P_2 (P_1 G_1 + P_1 C_1)$$

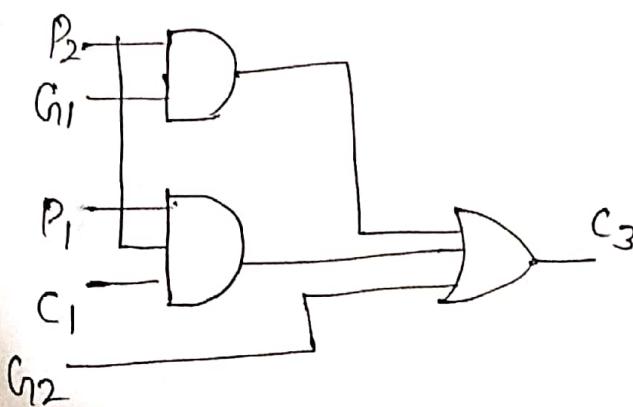
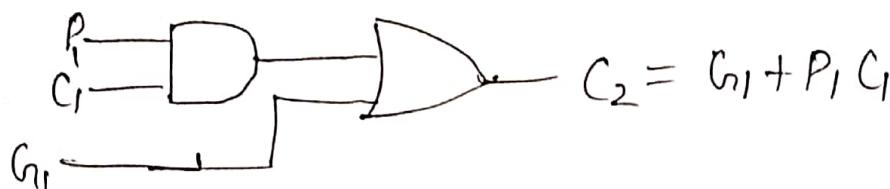
$$= G_2 + P_2 G_1 + P_2 P_1 C_1$$

$$C_4 = G_3 + P_3 C_3$$

$$= G_3 + P_3 G_2 + P_3 P_2 G_1 + P_3 P_2 P_1 C_1$$

tpw  
and

leads



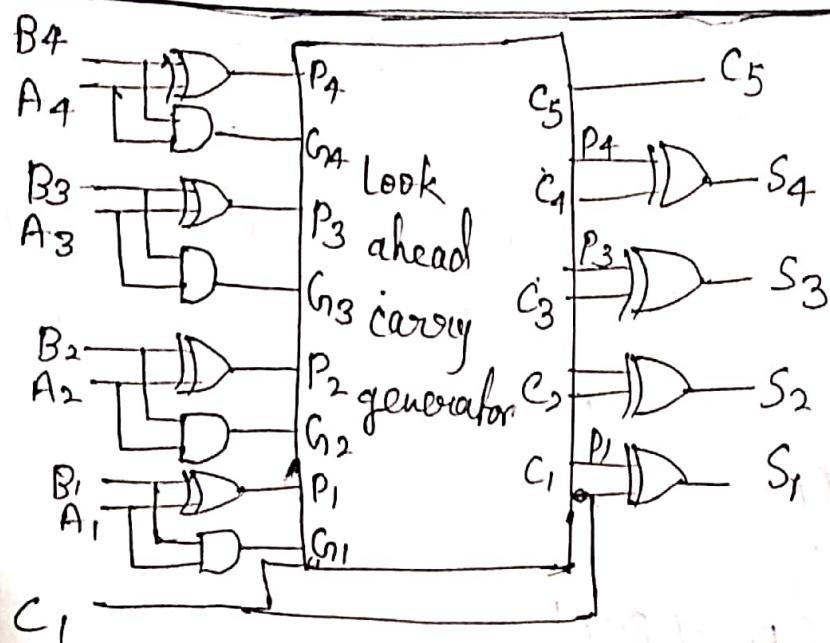
The  
principal

$B_i$

$i$   
 $c_i$

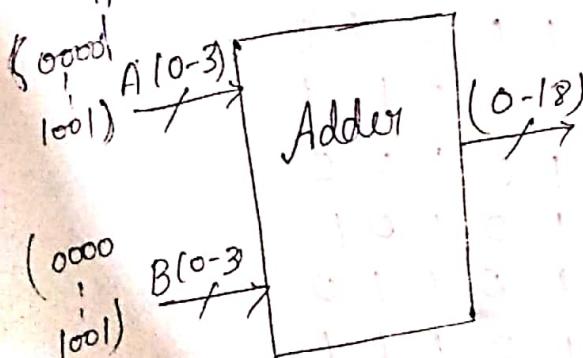
rate

4-bit full adder with look ahead carry



# BCD Adder

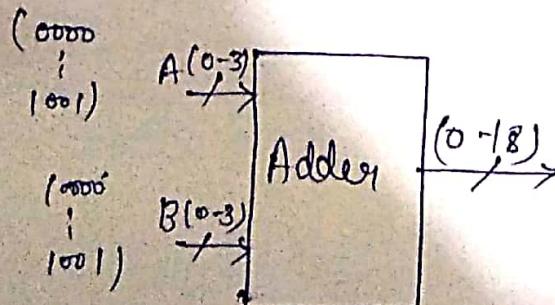
<u>Decimal</u>	<u>Binary Sum</u>					<u>BCD Sum</u>				
	C	S <sub>3</sub>	S <sub>2</sub>	S <sub>1</sub>	S <sub>0</sub>	C	S <sub>3</sub>	S <sub>2</sub>	S <sub>1</sub>	S <sub>0</sub>
0	0	0	0	0	0	0	0	0	0	0
1	0	0	0	0	1	0	0	0	0	1
2	0	0	0	1	0	0	0	0	1	0
3	0	0	0	1	1	0	0	0	1	0
4	0	0	1	0	0	0	0	1	0	1
5	0	0	1	0	1	0	0	1	1	0
6	0	0	1	1	0	0	0	1	1	1
7	0	0	1	1	1	0	0	1	0	0
8	0	1	0	0	0	0	0	0	0	0
9	0	1	0	0	1	0	1	1	1	1
10										
11										



It is used in digital clock.

# BCD Adder - It is used in digital clock.

Decimal	Binary	Sum	BCD Sum
	C S <sub>3</sub> S <sub>2</sub> S <sub>1</sub> S <sub>0</sub>		C S <sub>3</sub> S <sub>2</sub> S <sub>1</sub> S <sub>0</sub>
0	0 0 0 0 0		0 0 0 0 0
1	0 0 0 0 1		0 0 0 0 1
2	0 0 0 1 0		0 0 0 1 0
3	0 0 0 1 1		0 0 0 1 1
4	0 0 1 0 0		0 0 1 0 0
5	0 0 1 0 1		0 0 1 0 1
6	0 0 1 1 0		0 0 1 1 0
7	0 0 1 1 1		0 0 1 1 1
8	0 1 0 0 0		0 1 0 0 0
9	0 1 0 0 1		0 1 0 0 1
10	0 1 0 1 0		1 0 0 0 0
11	0 1 0 1 1		1 0 0 0 1
12	0 1 1 0 0		1 0 0 1 0
13	0 1 1 0 1		1 0 0 1 1
14	0 1 1 1 0		1 0 1 0 0
15	0 1 1 1 1		1 0 1 0 1
16	1 0 0 0 0		1 0 1 0 1
17	1 0 0 0 1		1 0 1 1 0
18	1 0 0 1 0		1 0 1 1 1
19	1 0 0 1 1		1 1 0 0 0



BCD of 15 = 0001 0101  
 Q1 Q5 = 1 1111 Binary of 15

(0-9) Binary sum is same as BCD sum.

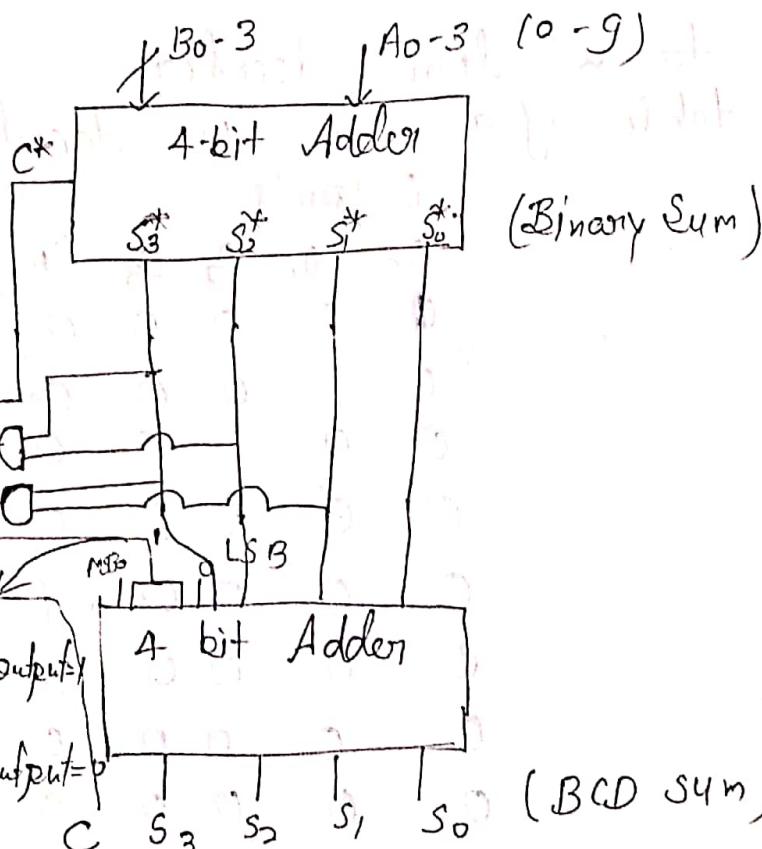
(10-19) the Binary sum is not same as BCD sum.

How we convert ~~BCD~~ sum in to Binary sum to the BCD sum!

Whenever the number is greater than 9 we have to add 6 (0110) and you will add this the output will be in BCD.

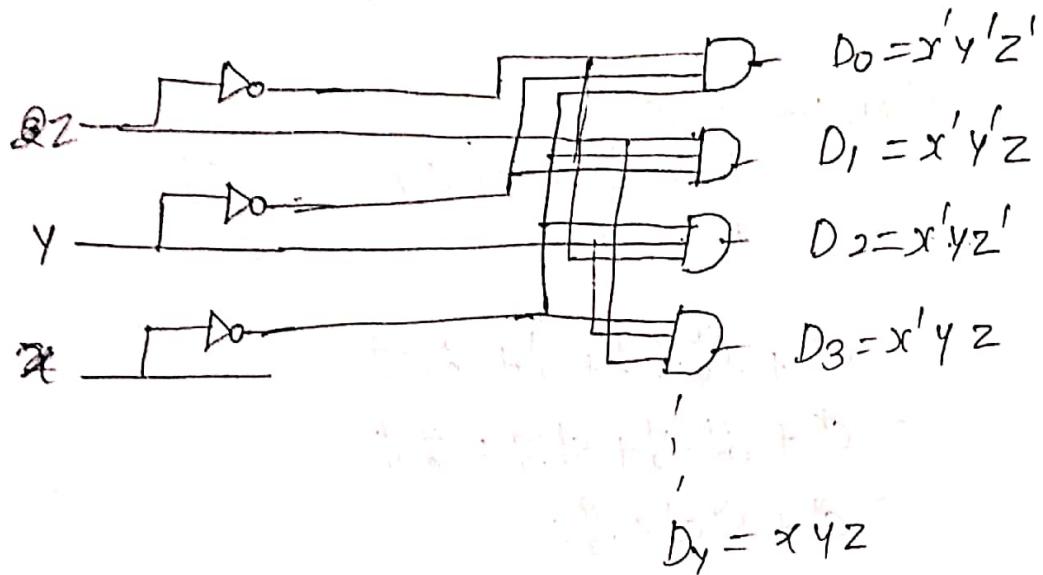
$$\text{"0110" when : - i) } C^* = 1 \quad (16 - 19) \\ \text{ii) } S_3^* \cdot (S_2^* + S_1^*) \quad (12 - 15) \\ \text{iii) } S_3^* \cdot S_1^* \quad (10, 11)$$

$$= C^* + S_3^* \cdot (S_2^* + S_1^*) + S_3^* \cdot S_1^* \\ = C^* + S_3^* \cdot S_2^* + S_3^* \cdot S_1^* + S_3^* \cdot S_1^* \\ = C^* + S_3^* S_2^* + S_3^* S_1^*$$



## Decoder

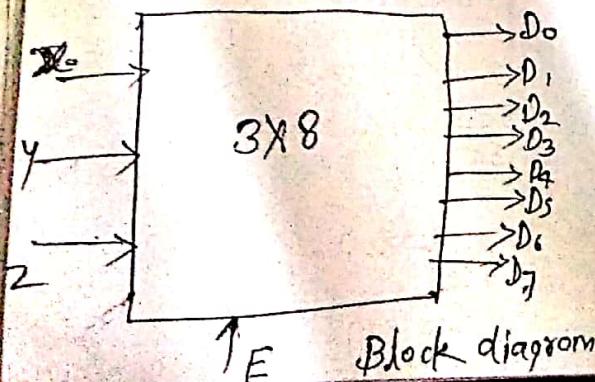
A decoder is a combinational circuit that converts Gray information from  $n$  input lines to a maximum of  $2^n$  unique output lines.



A 3 to 8 line decoder

Truth table of a 3-8 line decoder

Inputs	Outputs
$x \ y \ z$	$D_0 \ D_1 \ D_2 \ D_3 \ D_4 \ D_5 \ D_6 \ D_7$
0 0 0	0 0 0 0 0 0 0 0
0 0 1	0 1 0 0 0 0 0 0
0 1 0	0 0 1 0 0 0 0 0
0 1 1	0 0 0 1 0 0 0 0
1 0 0	0 0 0 0 1 0 0 0
1 0 1	0 0 0 0 0 1 0 0
1 1 0	0 0 0 0 0 0 1 0
1 1 1	0 0 0 0 0 0 0 1

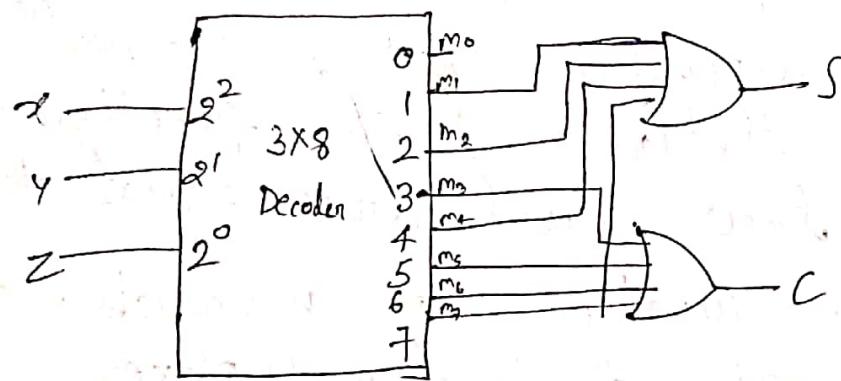


Implementation of a full adder with the help of a decoder and two OR gates!

full adder  $\Rightarrow$  3 Input and 2 Output

$$S(x, y, z) = \Sigma(1, 2, 4, 7)$$

$$C(x, y, z) = \Sigma(3, 5, 6, 7)$$



We make full adder with the decoder because it reduces the size no. of gates and also size of ckt.

Truth Table

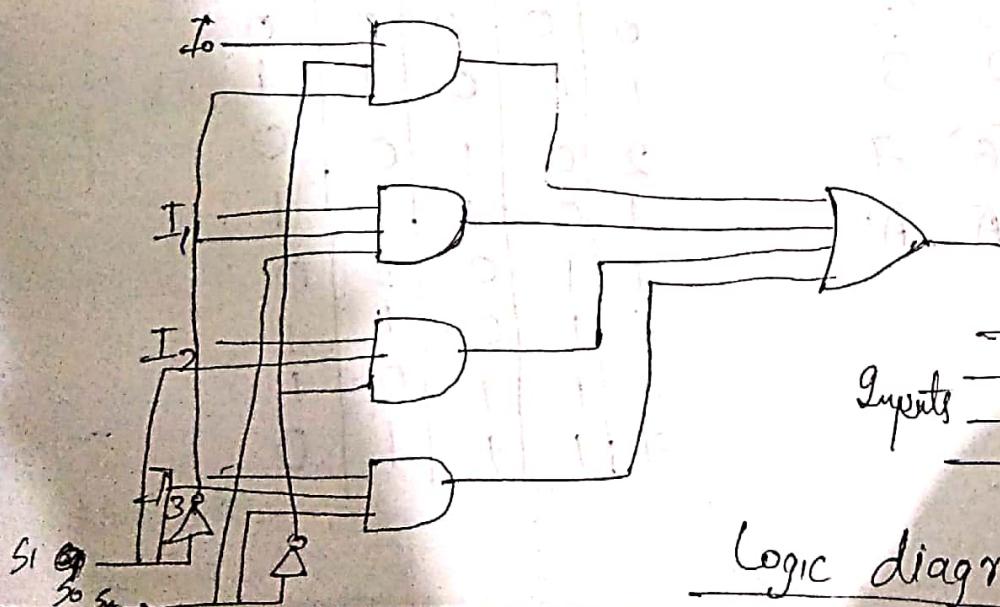
x	y	z	S	C	m <sub>0</sub>
0	0	0	0	0	m <sub>0</sub>
0	0	1	1	0	m <sub>1</sub>
0	1	0	1	0	m <sub>2</sub>
0	1	1	0	1	m <sub>3</sub>
1	0	0	1	0	m <sub>4</sub>
1	0	1	0	1	m <sub>5</sub>
1	1	0	0	1	m <sub>6</sub>
1	1	1	1	1	m <sub>7</sub>

$$S(x, y, z) = \Sigma(m_1, m_2, m_3, m_7)$$

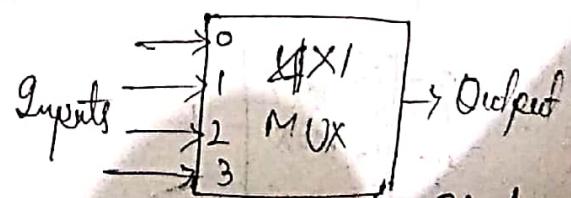
$$C(x, y, z) = \Sigma(m_4, m_5, m_6, m_7)$$

Multiplexer:-

Multiplexing means transmitting a large no. of information units over a smaller no. of channels of lines. A multiplexer is a combinational circuit that selects binary information from one of many input lines and directs it to a single output line.



S <sub>1</sub>	S <sub>0</sub>	Y
0	0	I <sub>0</sub>
0	1	I <sub>1</sub>
1	0	I <sub>2</sub>
1	1	I <sub>3</sub>



Logic diagram Select diagram

Multiplexer & its Output for TAKO in Output of 2 to 1

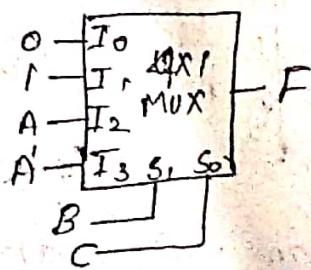
A multiplexer is essentially a decoder with the OR gate already available. The minterms out of the decoder to be chosen can be connected with the input lines. The minterms table included with the function being implemented are chosen by making their corresponding input lines equal to 1. The minterms not included in the function disabled by making their input lines equal to 0.

So any boolean function of n variables can be implemented with 2^n to 1 multiplexer.

If we have a function of n+1 boolean variables we take n of those variables and connect them to the selection lines of the multiplexer. The remaining single variable of the function is used for the inputs of the multiplexer.

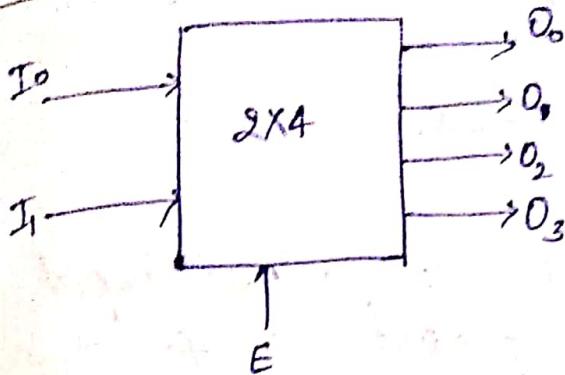
$$\# F(A, B, C) = \Sigma (1, 3, 5, 6)$$

	$I_0$	$I_1$	$I_2$	$I_3$
$A'$	0	1	2	3
A	4	5	6	7
	0	1	A	$A'$



minterm	A	B	C	F
0	0	0	0	0
1	0	0	1	1
2	0	1	0	0
3	0	1	1	1
4	1	0	0	0
5	1	0	1	1
6	1	1	0	1
7	1	1	1	0

## Decoder:-



$n=2$  input then Output =  $2^n$

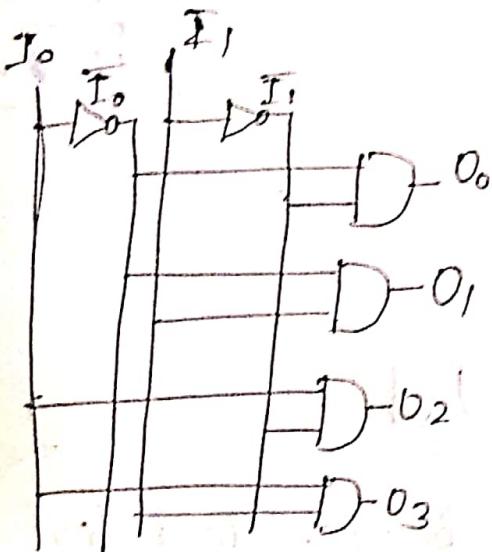
## $2 \times 4$ bit Decoder

Input		Output			
I <sub>0</sub>	I <sub>1</sub>	O <sub>0</sub>	O <sub>1</sub>	O <sub>2</sub>	O <sub>3</sub>
0	0	1	0	0	0
0	1	0	1	0	0
1	0	0	0	1	0
1	1	0	0	0	1

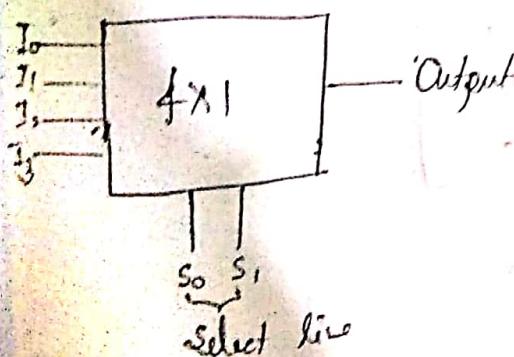
(Enable line or select line)

## Block Diagram of Decoder

→ It is a switch of decoder by which decoder is on or off.



## Multiplexer or MUX :-



How we can find, no. of select line

$$n = 2^s$$

where

$n =$  Total no. of input  
 $s =$  Total no. of select line

When input  $n=4$ , select line  
then

$$4 = 2^s$$

$$2^2 = 2^s$$

$$s = 2$$

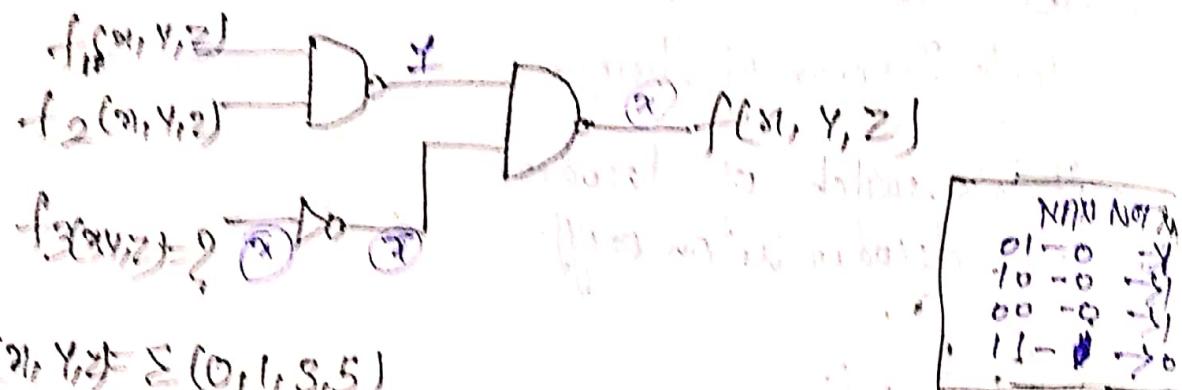
When input line  $n=8$   
 $2^3 = 2^s$   $s = 3$

There will be 3 select line

Select line degrees which input will be your output.

Multiplexer is also known as MUX.

Q:- Consider the following logic ckt whose inputs are functions  $f_1, f_2, f_3$  and output is  $f$ .



$$f_1(x,y,z) = \Sigma(0,1,5,6)$$

$$f_2(x,y,z) = \Sigma(6,7)$$

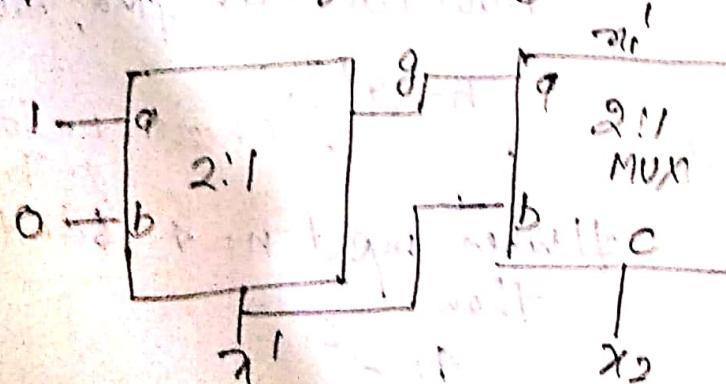
$$f_3(x,y,z) = \Sigma(1,4,5)$$

Hence  $f = f_3$

$$\therefore f_3(x,y,z) = \Sigma(1,4,5) \quad \underline{\text{Ans}}$$

NNN	Normal
01-0	0
10-0	1
00-0	0
11-1	1

Q:- Consider the circuit shown below. The output of a 2:1 MUX is given by the function  $a'bc'$ .



$$Y = I_0 S + I_1 S$$

when  $S=0$  then  $Y=I_0$

&  $I_0=1$  then  $Y=I_1$

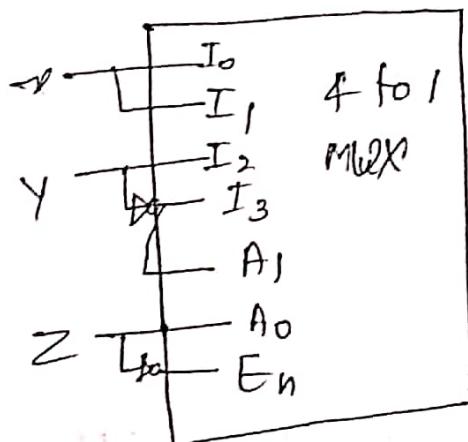
$$g = ac' + bc$$

$$= \bar{x}_1 + bx_1$$

$$= \bar{x}_1$$

$$x_1' x_2' + x_1 x_2$$

Q) Consider the following multiplexer where  $I_0, I_1, I_2, I_3$  are four data input lines selected by two address line combinations  $A_1 A_0 = 00, 01, 10, 11$  respectively and  $f$  is the output of the multiplexer.  $E_n$  is the enable input.



x	y	z	f
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	1
1	1	1	0

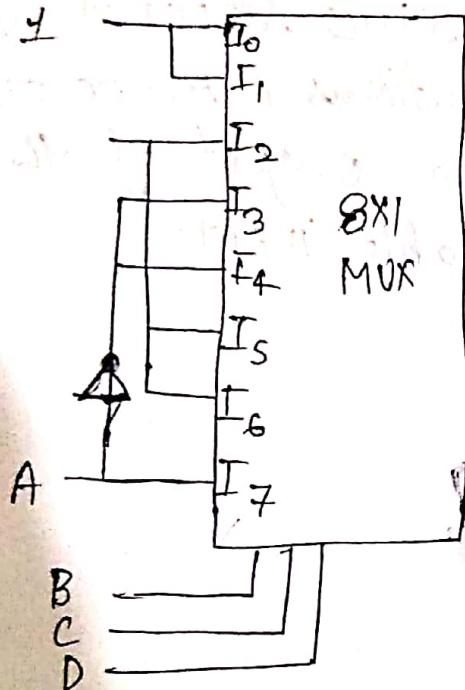
A <sub>1</sub>	A <sub>0</sub>	I <sub>0</sub>
0	0	I <sub>0</sub>
0	1	I <sub>1</sub>
1	0	I <sub>2</sub>
1	1	I <sub>3</sub>

$$f = xyz'$$

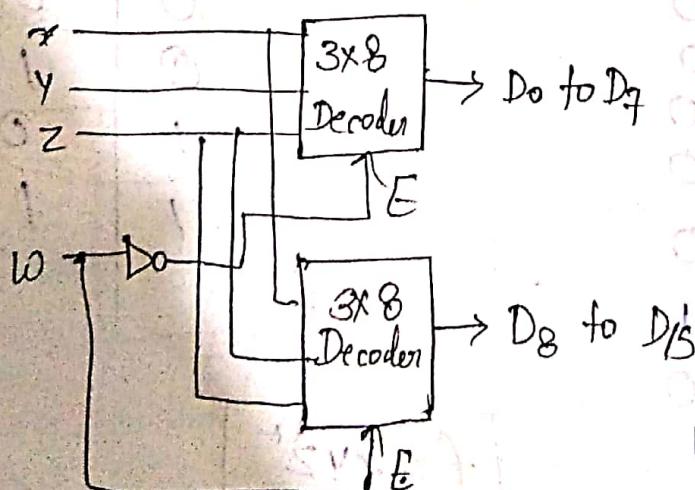
$$Q \Rightarrow F(A, B, C, D) = \Sigma(0, 1, 3, 4, 8, 9, 15)$$

	$I_0$	$I_1$	$I_2$	$I_3$	$I_4$	$I_5$	$I_6$	$I_7$
$A'$	①	②	2	③	④	5	6	7
$A$	⑧	⑨	10	11	12	13	14	15

$| \quad | \quad | \quad A' \quad | \quad A' \quad | \quad 0 \quad 0 \quad | \quad A$



# Constructing a  $4 \times 16$  decoder with two  $3 \times 8$  decoders:



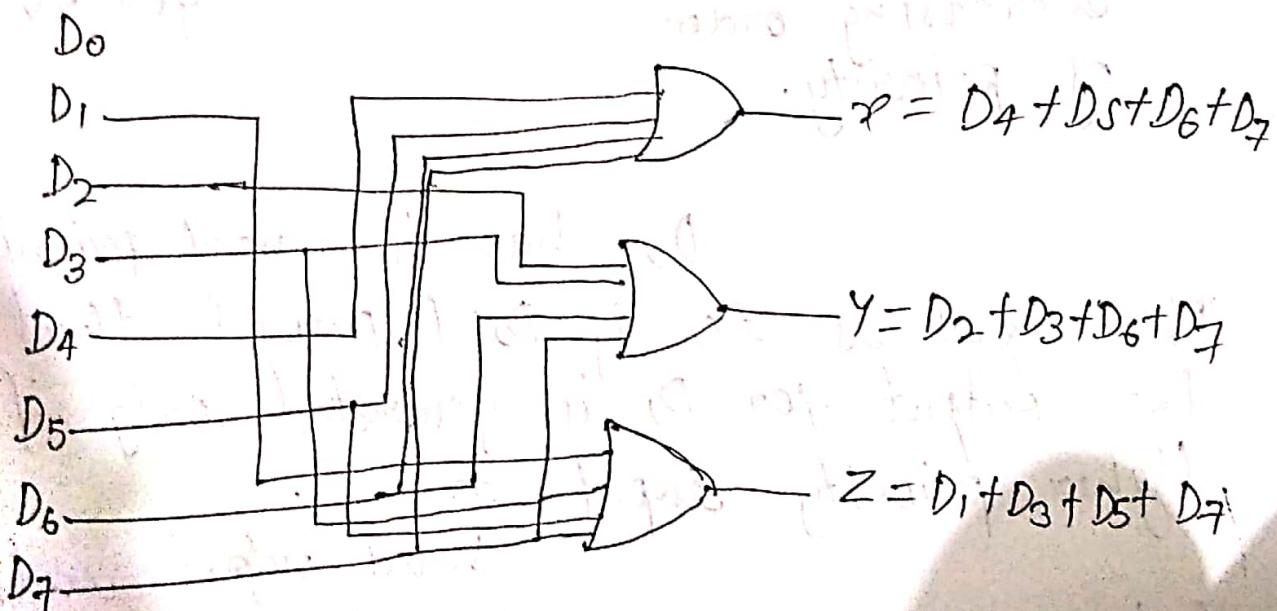
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
0	0	0	0	0	0	0	0	1	0	0	0	0	0	0	0	0
0	0	0	0	0	0	0	0	0	1	0	0	0	0	0	0	0
0	0	0	0	0	0	0	0	0	0	1	0	0	0	0	0	0
0	0	0	0	0	0	0	0	0	0	0	1	0	0	0	0	0
0	0	0	0	0	0	0	0	0	0	0	0	1	0	0	0	0
0	0	0	0	0	0	0	0	0	0	0	0	0	1	0	0	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1

## Encoders:-

The encoder is a digital circuit that performs the inverse operation of a decoder. The encoder has  $2^n$  (one fewer) input lines and  $n$  output lines.

Truth table for octal to Binary Encoder:-

Inputs								Outputs		
	$D_0$	$D_1, D_2$	$D_3$	$D_4$	$D_5$	$D_6$	$D_7$	$x$	$y$	$z$
1 0 0 0 0 0 0 0								0	0	0
0 1 0 0 0 0 0 0								0	0	1
0 0 1 0 0 0 0 0								0	1	0
0 0 0 1 0 0 0 0								0	1	1
0 0 0 0 1 0 0 0								1	0	0
0 0 0 0 0 1 0 0								1	0	1
0 0 0 0 0 0 1 0								1	1	0
0 0 0 0 0 0 0 1								1	1	1



## Priority Encoder:-

A priority encoder is an encoder circuit that includes the priority function. The operation of the priority encoder is such that if two or more inputs are equal to 1 at the same time, the input having the highest priority will take precedence.

### Truth table of a priority encoder

Inputs				Outputs	
$D_0$	$D_1$	$D_2$	$D_3$	$X$	$Y$
0	0	0	0	x	0
1	0	0	0	0	1
x	1	0	0	0	1
x	x	1	0	1	0
x	x	x	1	1	1

$D_0 \ D_1 \ D_2 \ D_3$   
Increasing order  
of priority.

Input  $D_3$  has the highest priority so regardless of themselves of other inputs. When this input is 1, the output for  $xy$  is 11.

$D_2$  has the next priority level. The output is 10 if  $D_2=1$ , proved that  $D_3=0$ . The output for  $D_1$  is generated only if other higher priority inputs are 0.

Valid output indicator ( $V$ ) is set to 1 only when one or more inputs are equal to 1.

Valid output indicator

K-map:-

		D <sub>2</sub>	
		00	01
D <sub>0</sub>	00	X	1
	01	1	1
D <sub>1</sub>	00	1	1
	01	1	1
D <sub>3</sub>	00	1	1
	01	1	1

		D <sub>2</sub>	
		00	01
D <sub>1</sub>	00	X	1
	01	1	1
D <sub>0</sub>	00	1	1
	01	1	1

$$y = D_3 + D_1 D_2'$$

$$x = D_2 + D_3$$

Q) Express the function  $f(x, y, z) = xy' + yz'$  with only one complement operation and one or more AND/OR operations. Draw the logic circuit implementing the expression obtained a single NOT gate and one or more AND/OR gates.

$$\begin{aligned} f(x, y, z) &= xy' + yz' \\ &= x'y'z' + xy'z + xyz' + x'y'z' \\ &= m_4 + m_5 + m_6 + m_2 \end{aligned}$$

$$f(x, y, z) = m_2 + m_4 + m_5 + m_6$$

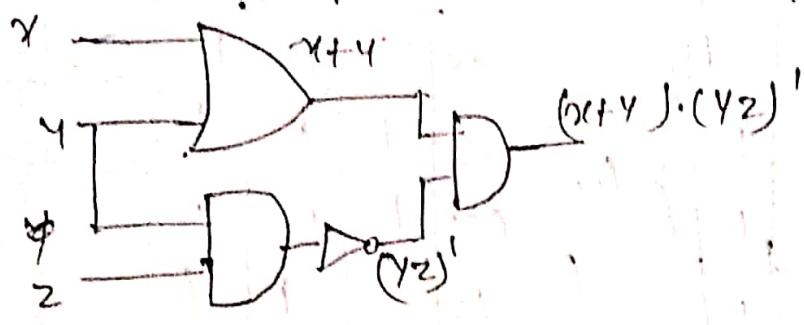
$$f'(x, y, z) = m_0 + m_1 + m_3 + m_7$$

		D <sub>2</sub>	
		00	01
D <sub>0</sub>	00	1	1
	01	1	1
D <sub>1</sub>	00	1	1
	01	1	1
D <sub>2</sub>	00	1	1
	01	1	1

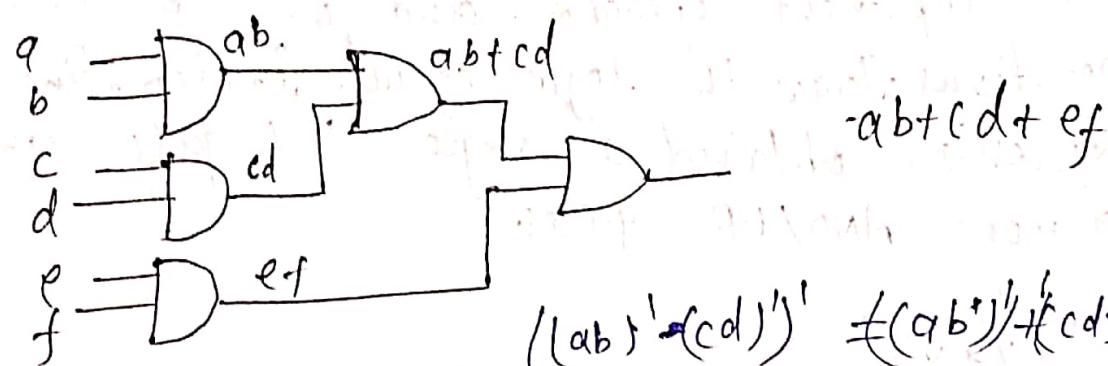
$$f'_1 = x'y' + yz$$

$$\begin{aligned} f = (f')' &= (x'y' + yz)' \\ &= (x'y')' \cdot (yz)' = (x+y)(yz)' \end{aligned}$$

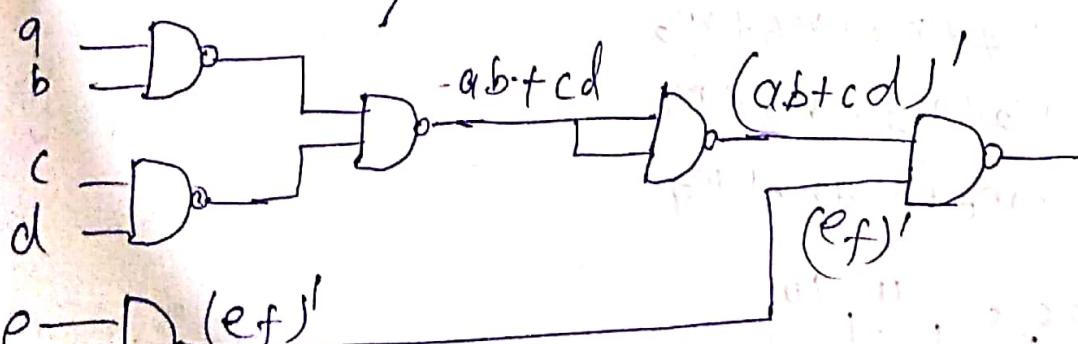
$$f = (x+y)(yz)'$$



Transform the following logic circuit (without expressing its switching function) into an equivalent logic circuit that employs only 6 NAND gates each with 2 inputs.



$$(ab)'(cd)' = (ab')' + (cd')' = ab + cd$$



$$= ((ab+cd)'(ef)')'$$

$$= ((ab+cd)')' + (ef')'$$

$$= ab + cd + ef$$

$$\text{Let } f = (w' + y)(x' + y)(w + x' + z)(w' + z)(x' + z)$$

- i) Express  $f$  in the minimal sum of products form.  
ii) If the output line is stuck at 0, for how many input combinations were the value of  $f$  be in correct.

$$f = (w' + y)(x' + y)(w + x' + z)(w' + z)(x' + z)$$

$$= (w' + x)x' + y + zz')(w' + x' + y + zz') + (w + x' + yy' + z)$$
$$(w' + x)x' + yy' + z)(w + x' + yy' + z)$$

$$f = (w' + x + y + z)(w' + x + y + z')(w' + x' + y + z)(w' + x' + y + z')$$
$$(w + x' + y + z)(w + x' + y + z')(w' + x' + y + z)(w' + x + y + z)$$
$$(w + x' + y + z)(w + x' + y' + z)(w' + x + y + z)(w' + x' + y + z)$$
$$(w' + x + y' + z)(w' + x' + y' + z)(w + x' + y + z)(w' + x + y + z)$$
$$(w + x' + y + z)(w' + x' + y' + z)$$

$$f = M_8 M_9 M_{12} M_{13} M_4 M_5 M_{12} M_{13} M_4 M_6 M_8 M_{10} M_{12}$$
$$M_{14} M_4 M_6 M_{12} M_{14}$$

$$f = M_4 M_5 M_6 M_8 M_9 M_{10} M_{12} M_{13} M_{14}$$

$$f = \pi(4, 5, 6, 8, 9, 10, 12, 13, 14)$$

$$f = \Sigma(0, 1, 2, 3, 7, 11, 15)$$

i)  $f = \Sigma(0, 1, 2, 3, 7, 11, 15)$

$$\textcircled{1} \quad f = \sum(0, 1, 2, 3, 7, 11, 15)$$

$w_2^{42}$	00	01	11	10	$w_1^1$
00	1	1	1	1	
01		1	1	1	
11			1		
10			1	1	

$f = w_1^1 + yz \underline{\underline{w_2}}$

II Express the Boolean function

$$f = AB + BC + CA$$

by using a 4:1 multiplexer.

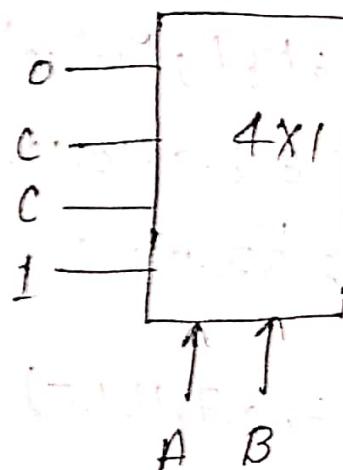
$$f = AB + BC + CA$$

$$f = ABC + ABC' + ABC + A'BC + ABC + AB'C$$

$$f = ABC + ABC + AB'C + ABC'$$

$$f = m_3 + m_5 + m_6 + m_7$$

A	B	C	$  f$
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	1

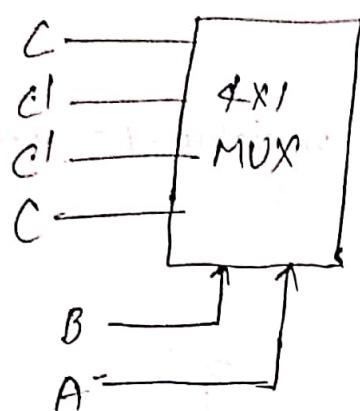


No. of variables = 3

No. of selector lines for MUX = No. of variables - 1 = 3 - 1 = 2

Multiplexer Used: 4 for

# What is the output of the following multiplexor circuit:-



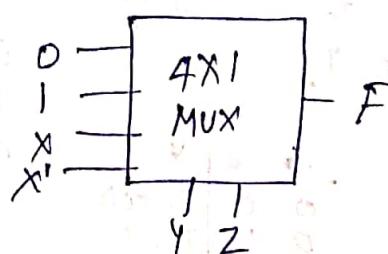
A	B	C	f
0	0	0	0
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	1
1	0	1	0
1	1	0	0
1	1	1	1

$$\begin{aligned}
 f &= A'B'C + A'BC' + AB'C' + ABC \\
 &= A'B'C' + AB'C' + A'B'C + ABC \\
 &= A'B'C' + AB'C' + (AB + A'B')C \\
 &= (A'B + AB')C' + (AB + A'B')C \\
 &= \underbrace{(A \oplus B)}_{X} C' + \underbrace{(A \oplus B)}_{X} C \\
 &= X C' + X' C \\
 &= X \oplus C \\
 &= A \oplus B \oplus C
 \end{aligned}$$

# Implementing Digital functions by using a Multiplexer

Truth table for  $F(x,y,z) = \Sigma(1,3,5,6)$

x	y	z	F
0	0	0	0
0	0	1	1
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	0



# Implementing Digital functions by using a Multiplexer

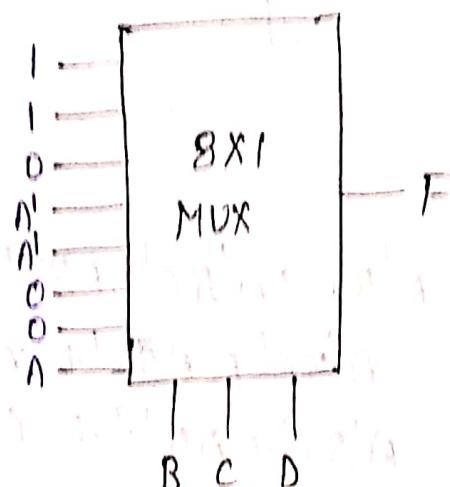
$$F(A, B, C, D) = \Sigma(0, 4, 5, 7, 8, 9, 15)$$

No. of Variables = 4

No. of selector lines for  $MUX = \text{No. of variables} - 1 = 4 - 1 = 3$

Multiplexer used: 8 to 1

A	B	C	D	F
0	0	0	0	1
0	0	0	1	1
0	0	1	0	0
0	0	1	1	1
0	1	0	0	1
0	1	0	1	0
0	1	1	0	0
0	1	1	1	0
1	0	0	0	1
1	0	0	1	1
1	0	1	0	0
1	0	1	1	0
1	1	0	0	0
1	1	0	1	0
1	1	1	0	0
1	1	1	1	1



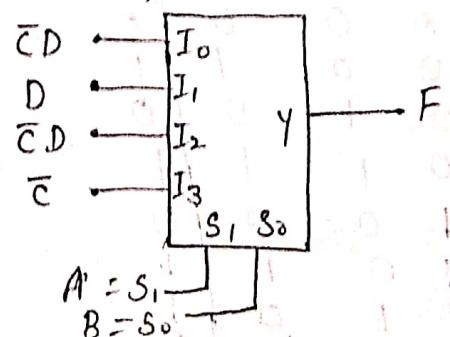
# Implementing Boolean function using Multiplexors

Implement!:-  $F(A, B, C, D) = \Sigma(1, 4, 5, 7, 9, 12, 13)$  using 4x1 mux.

AB	CD	I0	I1	I2	I3
S <sub>1</sub> S <sub>0</sub> = 00	00	0	1	0	0
S <sub>1</sub> S <sub>0</sub> = 01	01	1	1	1	0
S <sub>1</sub> S <sub>0</sub> = 11	11	1	1	1	1
S <sub>1</sub> S <sub>0</sub> = 10	10	1	1	0	0

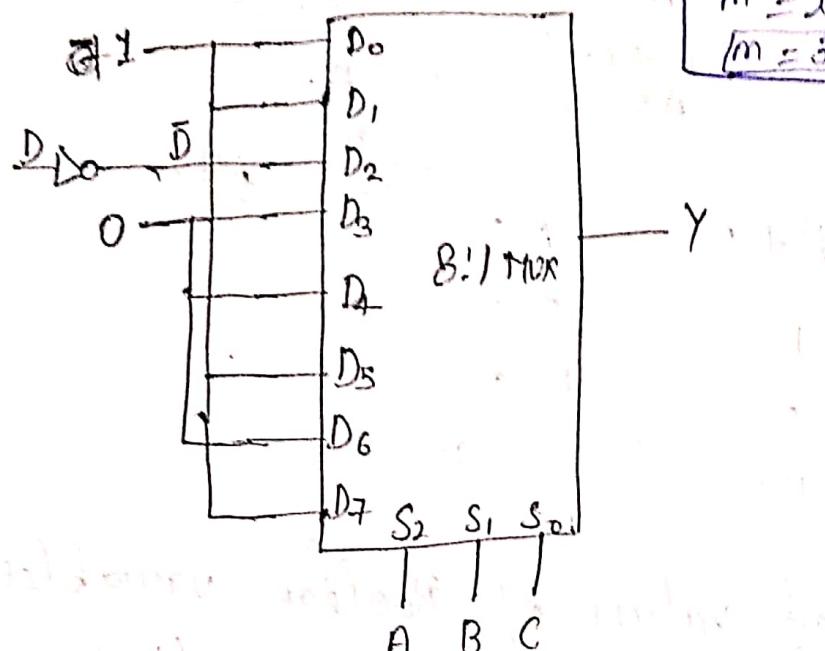
No. of Variable = 4      No. of Select line = 2 = 4  
No. of selector lines for MUX = 4 - 1 = 3      n = 2

S <sub>1</sub>	S <sub>0</sub>	Y
0	0	I <sub>0</sub> = CD
0	1	I <sub>1</sub> = D
1	0	I <sub>2</sub> = CD
1	1	I <sub>3</sub> = C



# Implement  $\Sigma m(0, 4, 2, 3, 4, 10, 11, 14, 15)$  using 8:1 MUX.

$S_2$	$S_1$	$S_0$	A	B	C	D	F
0	0	0	0	0	0	0	1 } $F=1$
0	0	0	0	0	0	1 } $F=1$	$S_2 = A, S_1 = B, S_0 = C$
0	0	1	0	0	0	1 } $F=1$	
0	0	1	1	0	0	1 } $F=1$	
0	1	0	0	0	0	1 } $F=D$	
0	1	0	0	0	0	0 } $F=0$	
0	1	1	0	0	0	0 } $F=0$	
0	1	1	1	0	0	0 } $F=0$	
1	0	0	0	0	0	0 } $F=0$	
1	0	0	1	0	0	0 } $F=1$	
1	0	1	0	1	0	0 } $F=1$	
1	0	1	1	1	0	0 } $F=0$	
1	1	0	0	0	0	0 } $F=0$	
1	1	0	1	0	0	0 } $F=0$	
1	1	1	0	0	0	1 } $F=1$	
1	1	1	1	1	1	1 } $F=1$	



1 MUX

$n=4$

$n=2$

# find values of Boolean variables  $A, B, C$  which satisfy the following equation:-

$$A + B = 1$$

$$AC = BC$$

$$A + C = 1$$

$$AB = 0$$

$$\begin{aligned} &\Rightarrow C = 0 \\ &\Rightarrow A = 1 \\ &\Rightarrow B = 0 \end{aligned}$$

check by  
7 combination  
also but it is  
very complex  
but we can do it

$$A + B$$

$$0 \ 1$$

$$1 \ 0$$

~~1~~

$$AB$$

$$0 \ 1$$

$$1 \ 0$$

~~0 0~~

# find values of Boolean variables  $A, B, C, D$  which satisfy the following equation:-

$$A' + AB = 0$$

$$AB = AC$$

$$AB + AC' + CD \neq C'D$$

$$A = 1, B = 0, C = 0$$

$$0 + 1 + 0 \times 1 = 1 \times 1$$

$$1 = 1$$

$$D = 1$$

$$A = 1, B = 0, C = 0, D = 1$$

$$A = 1, B = 0$$

$$\begin{matrix} X_0 & -C \\ A & 1 \\ B & 0 \end{matrix}$$

$$\begin{aligned} A' + AB &= (A' + A)(A' + B) = \\ &1 \cdot (A' + B) = 0 \end{aligned}$$

$$A' + B = 0$$

$$A' = 0 \Rightarrow A = 1$$

$$B = 0$$

$$\text{from } ②, AB = AC$$

$$\begin{aligned} A &= 1 \text{ then } B = C \\ &\Rightarrow C = 0 \end{aligned}$$

$$\text{Put } A = 1, B = 0, C = 0 \text{ in}$$

$$\textcircled{3} \text{ eqn}$$

$$D = 1$$

# Boolean Expression  $A \oplus B \oplus A$  is equivalent to

- (A)  $AB + A'B'$
- (B)  $A'B + AB'$
- (C)  $B$
- (D)  $A$

Competitive  $\nearrow$  Associative fns.

$$A \oplus B \oplus A$$

$$= (A \oplus A) \oplus B$$

$$= 0 \oplus B$$

$$= B$$

#  $F(P, Q) = ((1 \oplus P) \oplus (P \oplus Q)) \oplus ((P \oplus Q) \oplus (Q \oplus 0))$

(I)  $P+Q$

(II)  $(P+Q)'$

(III)  $P \oplus Q$

(IV)  $(P+Q)'$

$$= ((\bar{1} \oplus P + \bar{P} \cdot 1) \oplus (P \oplus Q)) \oplus ((P \oplus Q) \oplus (Q\bar{0} + 0\bar{Q}))$$

$$= (\bar{P} \oplus P \oplus Q) \oplus (P \oplus Q \oplus Q)$$

$$= (\bar{P}P + \bar{P}\bar{P}) \oplus Q \oplus P \oplus (P \oplus Q \oplus Q)$$

$$= (\cancel{PP} + \cancel{\bar{P}\bar{P}}) \oplus Q \oplus P \oplus (P \oplus Q \oplus Q)$$

$$= (\cancel{P} + \cancel{\bar{P}}) \oplus Q \oplus P \oplus (P \oplus Q \oplus Q)$$

$$= \cancel{1} \oplus Q \oplus P$$

$$= \cancel{1} \oplus \bar{Q} + \bar{Q} \cdot \cancel{1} \oplus P$$

$$= \cancel{\bar{Q} \oplus P}$$

$$(P \oplus Q)' \oplus 0$$

$$= (P \oplus Q)' \cancel{Auy}$$

OR  $1 \oplus (P \oplus A) \oplus Q \oplus P \oplus \bar{Q} \oplus Q \oplus 0$

$$1 \oplus 0 \oplus Q \oplus P \oplus 0 \oplus 0$$

$$1 \oplus Q \oplus P \oplus 0$$

$$1 \oplus (Q \oplus P)$$

$$= (Q \oplus P)'$$

11) Let  $\oplus$  and  $\odot$  denote the exclusive OR and exclusive NOR operation, respectively. Which one of the following is not correct.

(a)  $(P \oplus Q)' = P \odot Q$

$$(P \oplus Q)' = P \odot Q$$

$$\begin{cases} P \oplus P = 0 \\ P \oplus P' = 1 \end{cases}$$

(b)  $P'(P \oplus Q) = P \odot Q$

(c)  $P'(P \oplus Q)' = P \oplus Q$

(d)  $P \oplus P' \oplus Q = P \oplus Q$

$$\begin{cases} P \odot P = 1 \\ P \odot P' = 0 \end{cases}$$

(a)  $(P \oplus Q)' = (\bar{P}Q + P\bar{Q})' = \bar{\bar{P}}\bar{Q} \cdot \bar{\bar{P}}Q = (\bar{\bar{P}} + \bar{Q})(\bar{\bar{P}} + Q) = (\bar{P} + \bar{Q})(\bar{P} + Q) = P \oplus Q$

(b)  $P' \oplus Q = \bar{P}Q + \bar{P}\bar{Q} = P\bar{Q} + \bar{P}\bar{Q} = P \oplus Q$

(c)  $\bar{P}' \oplus \bar{Q} = \bar{\bar{P}}\bar{Q} + \bar{P}\bar{\bar{Q}} = \bar{P}\bar{Q} + \bar{P}\bar{Q} = P \oplus Q$

(d)  $P \oplus P' \oplus Q = P \oplus Q = \bar{Q}$

$$\begin{aligned} P \oplus P' \oplus Q &= (\bar{P}\bar{P} + P\bar{P}) \oplus Q \\ &= (\bar{P}P + P\bar{P}) \odot Q \\ &= 'P\bar{P} \odot Q' \\ &= \bar{O} \odot \bar{Q} \\ &= \bar{O} \bar{\bar{Q}} + O \cdot \bar{Q} \\ &= O \end{aligned}$$

$$P \oplus Q = \bar{Q}$$

If  $w, x, y$  and  $z$  are boolean variables then which one of the following is incorrect —

(a)  $wx + w(x+y) + x(xy) = x + wy$

(b)  $(wx'(y+z'))' + w'x = w' + x + y'z$

(c)  $(wx'(y+xz') + w'x)y = xy'$

(d)  $(w+y)(wx'y + wyz) = wx'y + wyz$

$$\begin{aligned} & x + wx + wy \\ & = x \end{aligned}$$

(a)  $wx + w(x+y) \neq x(x+y)$

$$= wx + wx + wy + x\cancel{x} + xy$$

$$= wx + wy + x + xy$$

$$= x(w+1) + wy + xy$$

$$= x + wy + xy$$

$$= x(1+y) + wy$$

$$= x + wy$$

(b)  $(wx'(y+z'))' + w'x$

$$= \overline{wx'} + (\overline{y+z}) + w'x$$

$$= (\overline{w} + \overline{x}) + (\overline{y} \cdot \overline{z}) + w'x$$

$$= (\overline{w} + x) + (\overline{y}z) + \overline{w}x$$

$$= \overline{w}\overline{wx} + \overline{w}xx + \overline{y}z$$

$$= \overline{w} + x + \overline{y}z$$

$$= \overline{w} + x + \overline{y}z$$

(c)  $(w+y)(wx'y + wyz)$

$$= wwxy + wwyz + wxyy + ywyz$$

$$= wx'y + wyz + wx'y + wyz$$

$$= wx'y + wyz$$

(d)  $(wx'y + w'x')y$

$$= wx'y + w'x'y$$

$$= wx'y + w'x'y$$

$$= x'y(w+w')$$

$$= x'y \cdot 1$$

$$= x'y$$

# Digital Circuits

Combinational

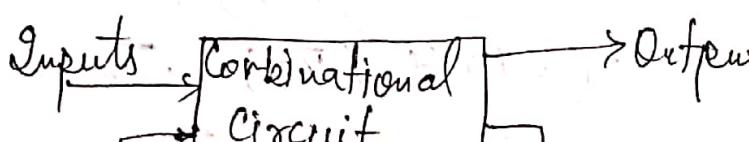
Sequential

Synchronous

Asynchronous

In combinational circuit present output depends on present inputs only.

In sequential ckt present outputs depends on previous output as well as present inputs.

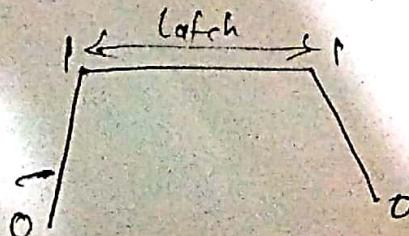


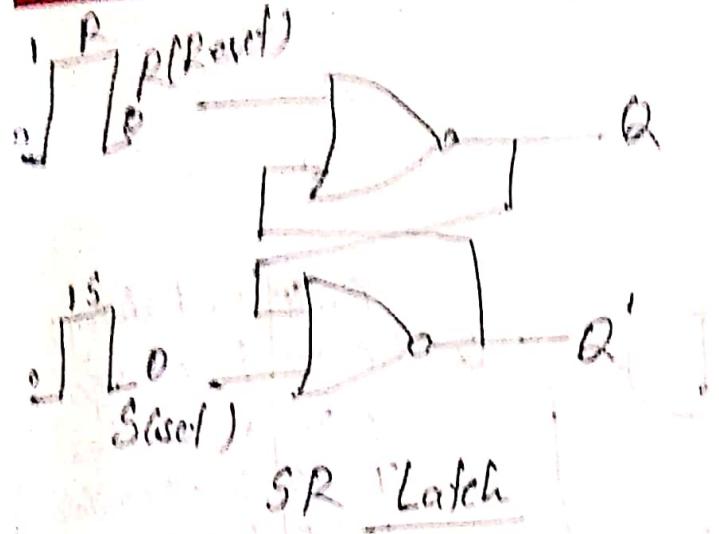
Block diagram of Sequential Circuit

flip-flop!

A flip-flop circuit can maintain a binary state ~~indefinitely~~ until directed by an input signal to change states.

Latch - If is a storage element that operate with flip-flop - If is controlled by a clock transition.





S	R	Q	Q'
1	0	1	0
0	0	0	1
0	1	0	1
1	0	0	1
1	1	0	0

This is output

RS flip-flop using NOR Gate

RS = NOR द्वारा  
SR = NAND द्वारा

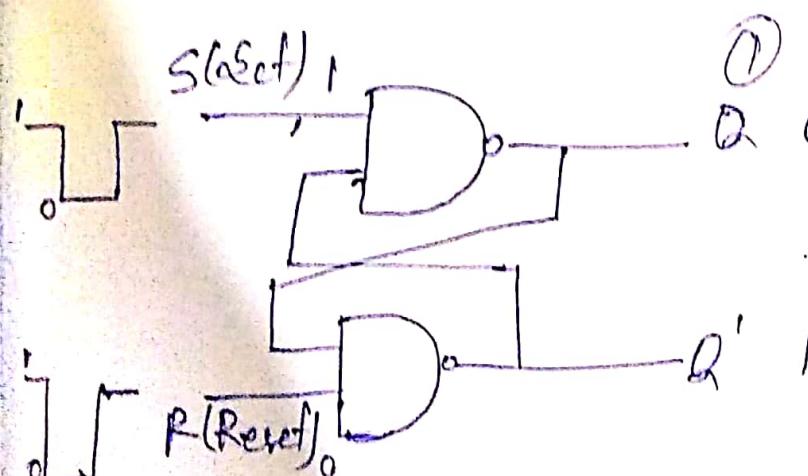
S=1 Then Set  
R=1 Then Reset

R	S	Q	Q'
0	0	0	1
0	1	1	0
1	0	1	0
1	1	0	0

"forbidden condition"

This is  
invalid  
output.

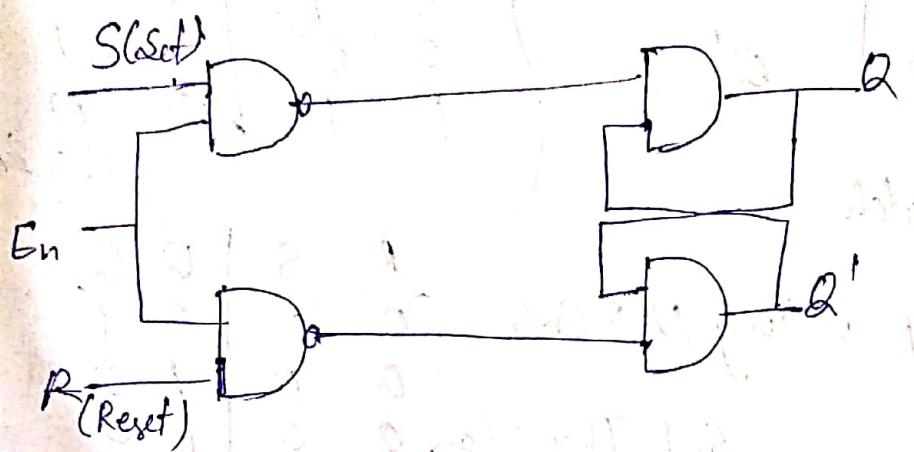
SR flip-flop using NAND Gate :-



S	R	Q	Q'
1	1	0	1
1	0	0	1
0	1	0	0
1	1	1	0
0	0	1	1

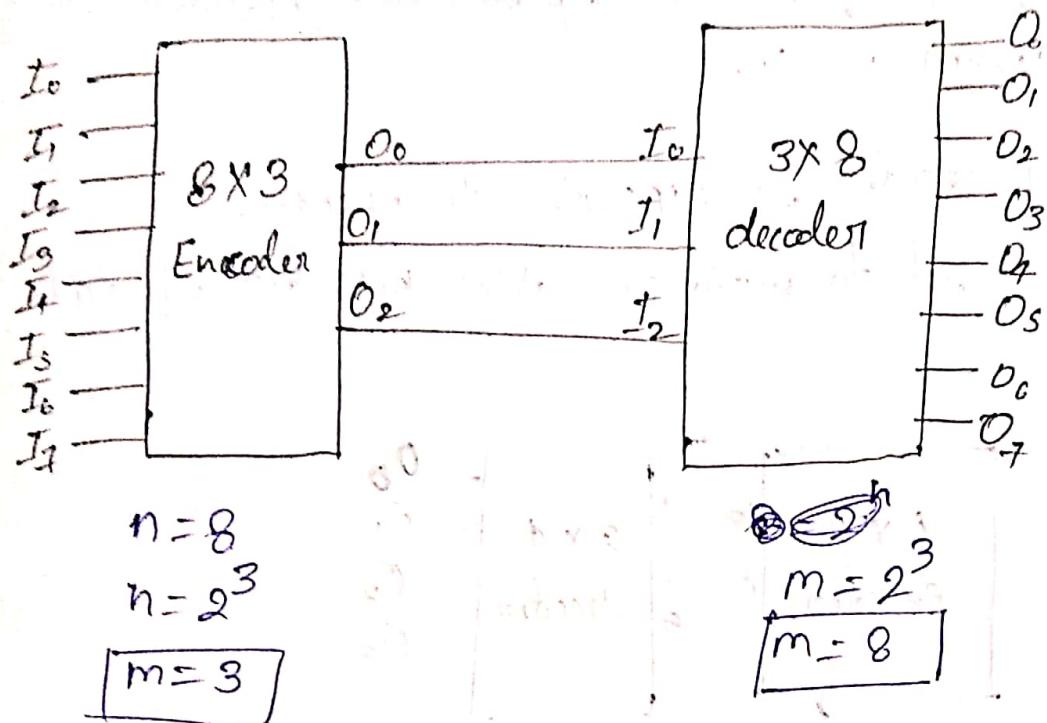
S = 0 Then Set

R = 0 Then Reset



$En SR$	Next state of $Q$
0 XX	No change
1 0 0	No change
1 0 1	$Q = 0$ Reset state
1 1 0	$Q = 1$ Set state
1 1 1	Indeterminate

Using Encoder and decoder we reduce the number of data lines since cost also be reduced.



Let  $I_5$  is high = 101 then  $O_1 = 1$ ,  $O_2 = 0$ ,  $O_3 = 1$  and again this information will be decode and  $O_5$  will be high.

### Type of Encoders:-

There are 4 type of Encoders:-

① Priority Encoders

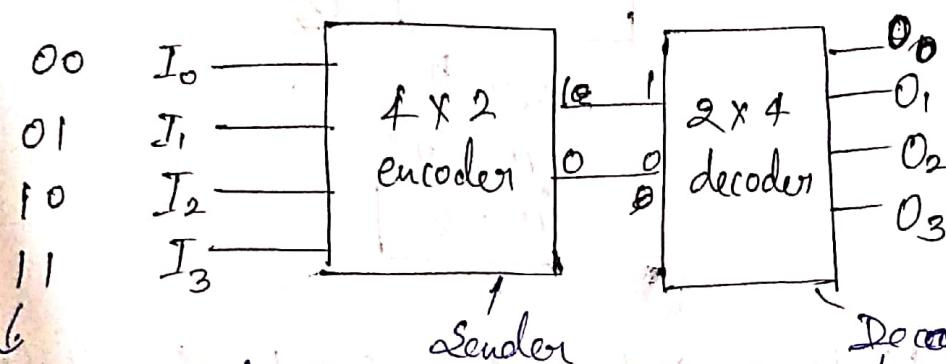
Decimal to BCD Encodes

Octal to binary Encodes

Hexadecimal to Binary encoder.

## Introduction to Encoders & Decoders

- ⇒ They are combinational circuits.
- ⇒ Encoder, decoder, and multiplexer are Medium scale integrated (MSI) circuit.
- ⇒ Function of decoder is opposite to encoder.
- ⇒ Encoder is used to minimize the number of data lines.



We see that in every case only one input is high and remaining is low then we use only two data lines one for high and other for low.

~~Encoder~~ Encoder # no. of input ( $n$ ) = 4

$$n = 4$$

$$n = 2^m$$

Relation b/w input & output.  $\boxed{h = 2^m}$

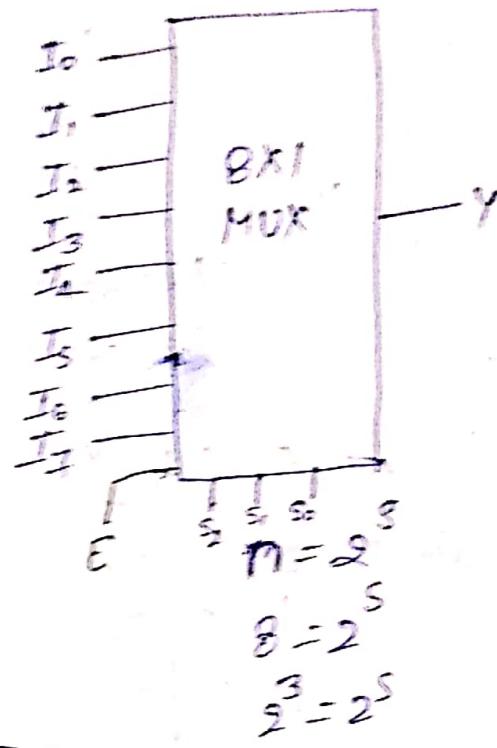
⇒ In case of decoder no. of Relation b/w no. of input and no. of output:-

$$\boxed{m = 2^n}$$

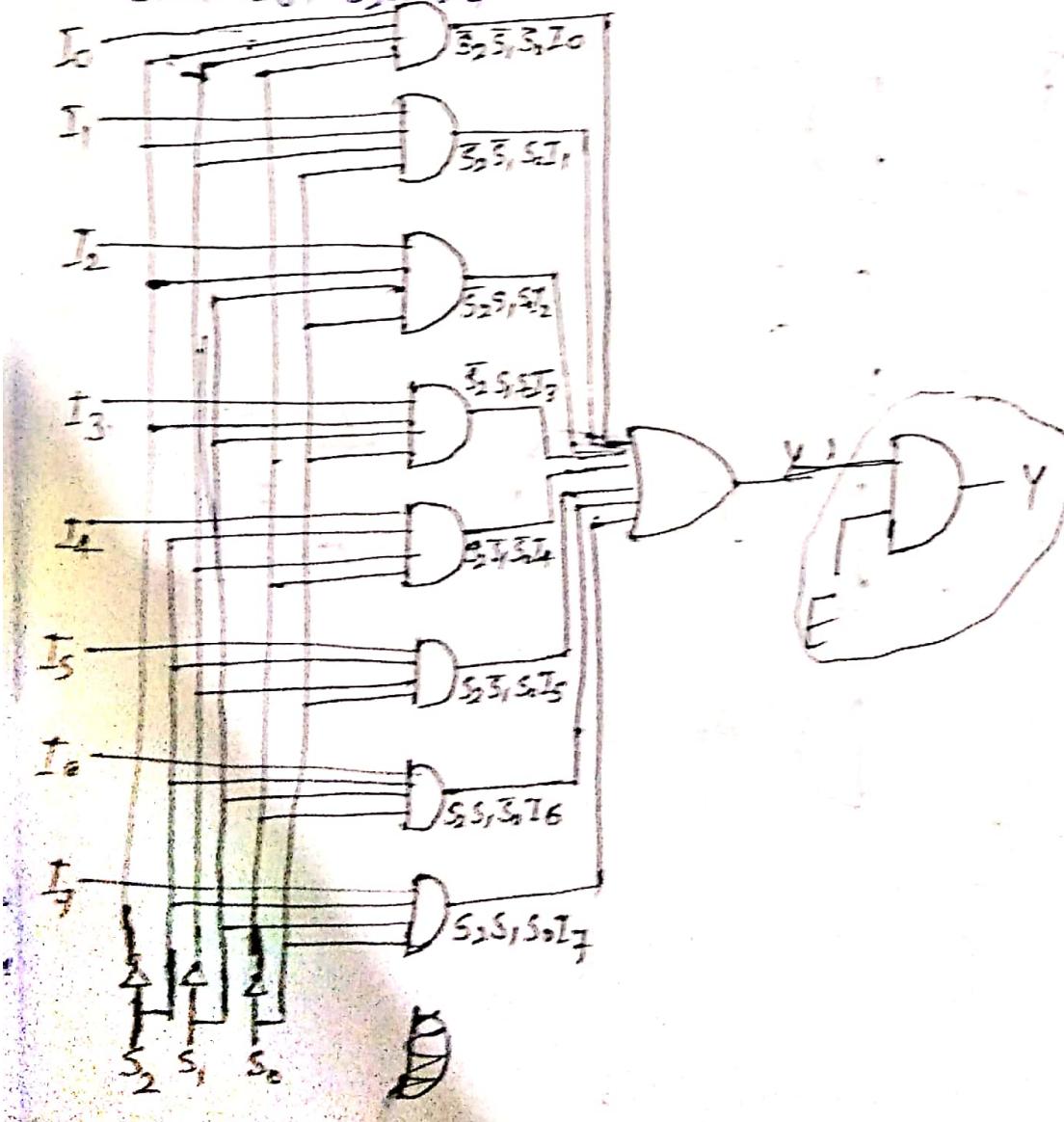
Where  $n$  = no. of input lines  
 $m$  = no. of output lines

# 8X1 MUX

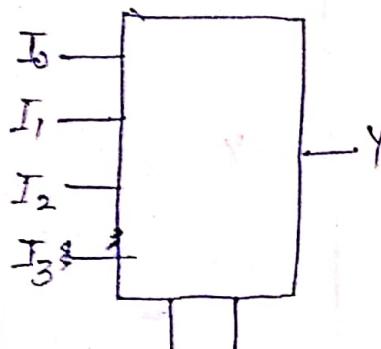
$S_2$	$S_1$	$S_0$	$y$
0	0	0	$I_0$
0	0	1	$I_1$
0	1	0	$I_2$
0	1	1	$I_3$
1	0	0	$I_4$
1	0	1	$I_5$
1	1	0	$I_6$
1	1	1	$I_7$



$$y = \bar{S}_2 \bar{S}_1 \bar{S}_0 I_0 + \bar{S}_2 \bar{S}_1 S_0 I_1 + \bar{S}_2 S_1 \bar{S}_0 I_2 + \\ \bar{S}_2 S_1 S_0 I_3 + S_2 \bar{S}_1 \bar{S}_0 I_4 + S_2 \bar{S}_1 S_0 I_5 + \\ S_2 S_1 \bar{S}_0 I_6 + S_2 S_1 S_0 I_7$$



# 4X1 Multiplexer



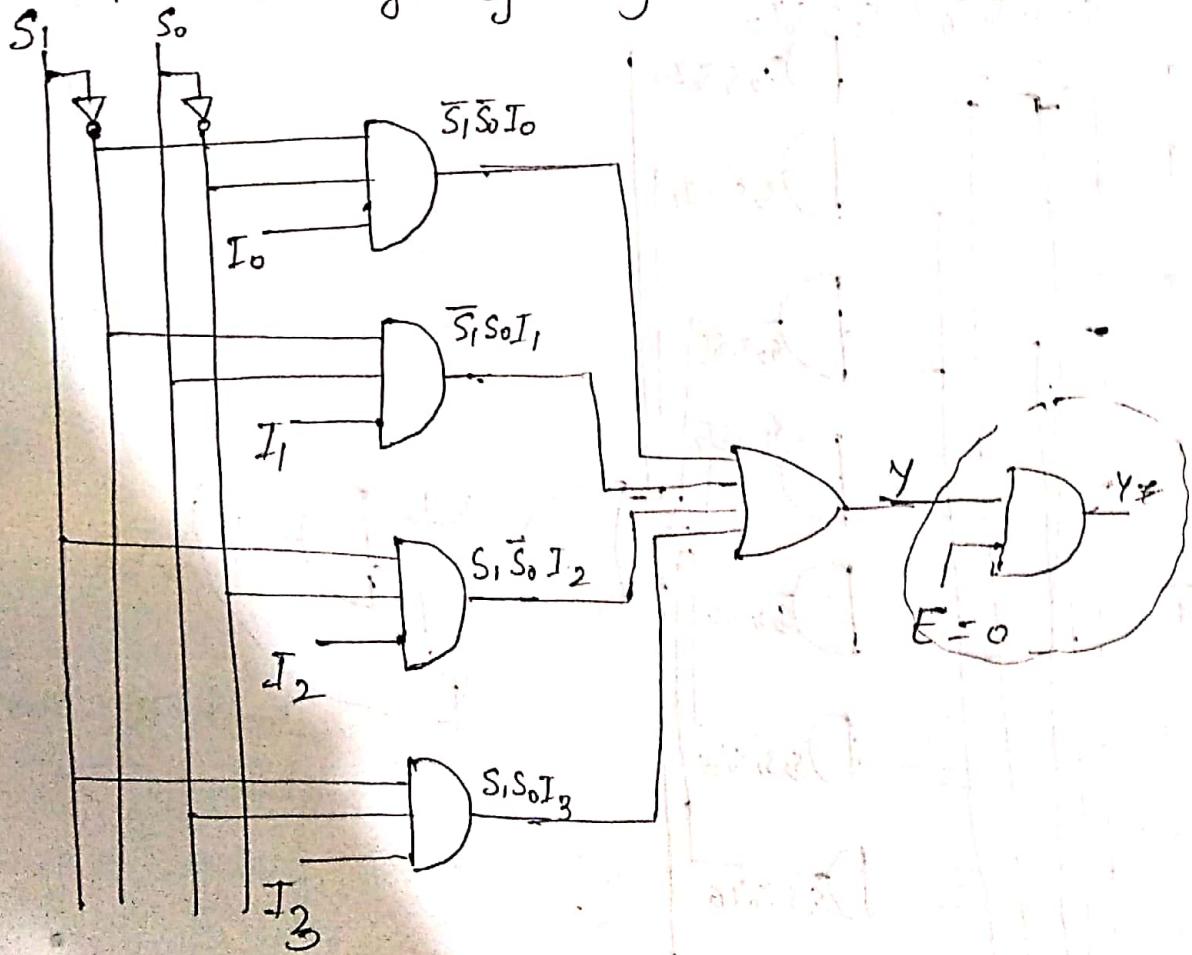
No. of selected lines  
n = 2<sup>s</sup>  
4 = 2<sup>s</sup>  
 $S = 2$

Hence in the case of 4X1 Multiplexer there will be 2 select lines.

$S_1$	$S_0$	$S_1 \quad S_0$	$Y$
0	0	00	$I_0$
0	1	01	$I_1$
1	0	10	$I_2$
1	1	11	$I_3$

$y = \bar{S}_1 \bar{S}_0 I_0 + \bar{S}_1 S_0 I_1 + S_1 \bar{S}_0 I_2 + S_1 S_0 I_3$

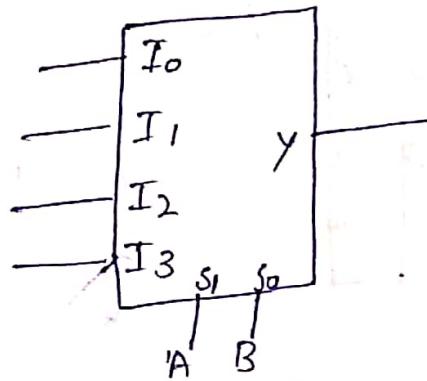
Implement this logic by using the gate:



# Implementing of Boolean function using Multiplexers

Implementation! -  $F(A, B, C, D) = \sum m(1, 4, 5, 7, 9, 12, 13)$  using  $4 \times 1$  MUX.

AB	CD	00	01	11	10
00			1		
01		1	1	1	
11		1			
10			1		



S <sub>1</sub>	S <sub>0</sub>	Y
0	0	
0	1	
1	0	
1	1	

Implementing  $8 \times 1$  MUX using  $4 \times 1$  MUX. (Special case)

$n_1 = 8$  using enable we can implement

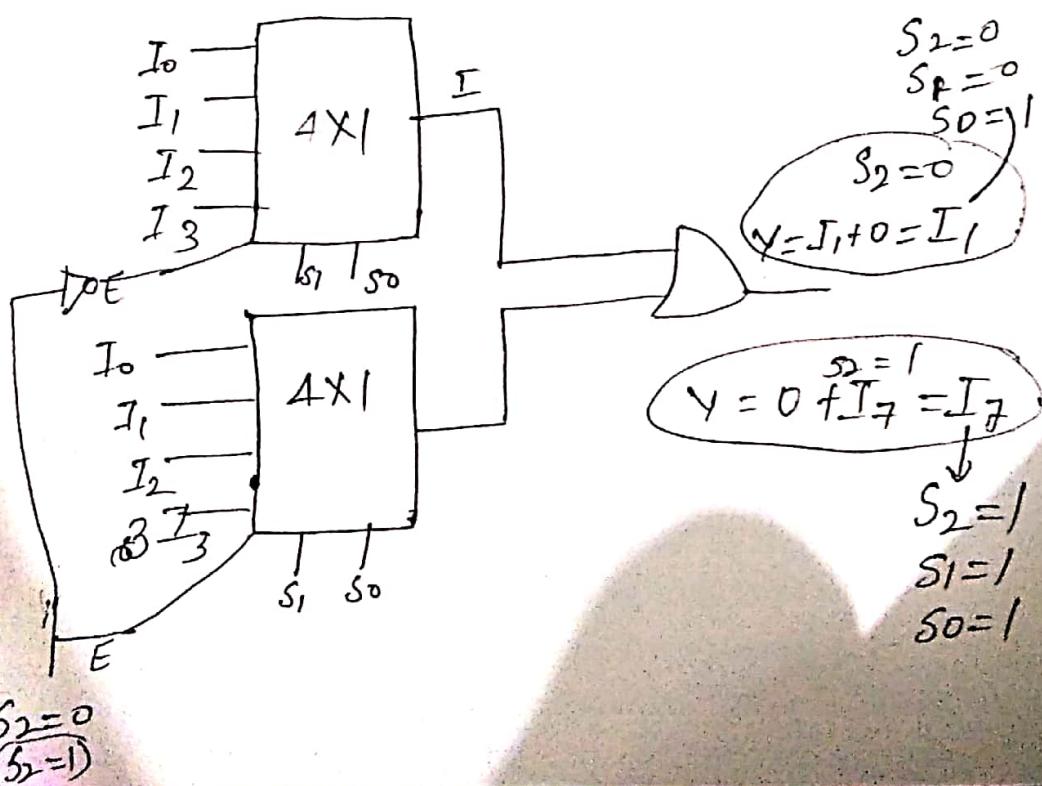
$$n_2 = 4$$

$$\frac{8}{4} = 2$$

$$\frac{2}{4} = 0.5$$

We use 2.5  $4 \times 1$  MUX to implement  $8 \times 1$  MUX

S <sub>2</sub>	S <sub>1</sub>	S <sub>0</sub>	Y
0	0	0	I <sub>0</sub>
0	0	1	I <sub>1</sub>
0	1	0	I <sub>2</sub>
0	1	1	I <sub>3</sub>
1	0	0	I <sub>4</sub>
1	0	1	I <sub>5</sub>
1	1	0	I <sub>6</sub>
1	1	1	I <sub>7</sub>



$$F(ABCD) = \sum m(1, 2, 5, 9, 13) + \sum d(3, 6, 11, 15)$$

AB	CD	00	01	11	10
00	0	1	1	X	1
01	0	4	1	5	7
11	0	12	13	X	15
10	0	8	1	9	X
		CD	AD		

$$F = AD + \bar{C}D + \bar{A}\bar{C}D$$

### Complements:-

find the 2's complement of the given number-

$$\textcircled{a} \quad 11011$$

$$N = 11011$$

$$n = 5$$

$$r = 2$$

$$\begin{aligned} 2^8 \text{ comp.} &= 2^5 - 11011 \\ &= (3^2)_10 - 11011 \\ &= 100000 - 11011 \\ &= 00101\cancel{1} \underline{011} \end{aligned}$$

$$\textcircled{b} \quad 1101.01$$

$$\begin{matrix} n = 4 \\ r = 2 \end{matrix}$$

$$\begin{aligned} 2^8 \text{ comp.} &= 2^4 - 1101.01 \\ &= (16)_2 - 1101.01 \\ &= 10000 - 1101.01 \\ &= 0010.11 \quad \underline{\text{Ans}} \end{aligned}$$

Q) Simplify  $F(AB'CD) = \overline{m}(0, 1, 3, 6, 7, 8, 9, 11, 13, 14, 15)$

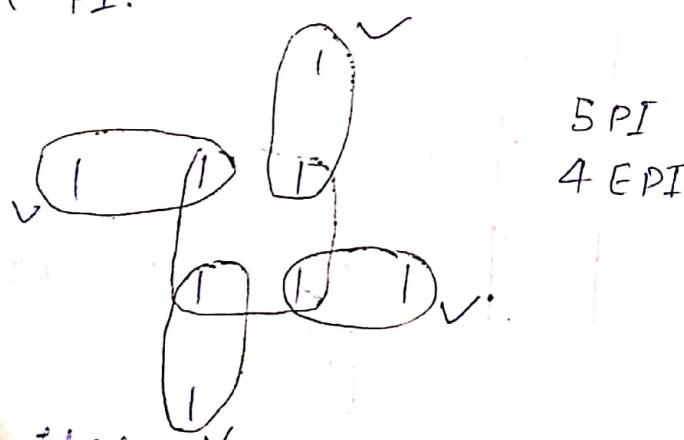
AB	CD	00	01	11	10
00	0	0	0	3	2
01	4	5	0	6	1
11	12	13	0	15	0
10	8	9	0	11	10

$$F = (\bar{B} + \bar{C}) \cdot (B + C) \cdot (C + D) \cdot (A + D)$$

## Prime Implicant and Essential Prime Implicant:-

Prime Implicant में बड़ा से बड़ा Pair होता है।

Essential Prime Implicant (EPI) :- At least one minterm which is not covered by other PI.



AB	CD	00	01	11	10
00	1	1	1	1	1
01	1	1	1	1	1
11	1	1	1	1	1
10	1	1	1	1	1

4 PI

3 EPI

POS form,  $F = \pi(0, 1, 4, 7)$

A	B	C	00	01	11	10
B	C		0	1	3	2
0			0	1	3	2
1			0	1	3	2
	1		4	5	7	6
		1	12	13	15	14
			8	9	11	10

$$F = \bar{B}C + \bar{A}\bar{B} + ABC$$

$$F = \overline{\bar{B}C + \bar{A}\bar{B} + ABC}$$

$$F = (\bar{B}\bar{C}) \cdot (\bar{A}\bar{B}) \cdot (\bar{A}BC)$$

$$\boxed{F = (B+C)(A+B)(\bar{A}+\bar{B}+\bar{C})}$$

We can write this directly.

Q → Obtain Reduced Expression for

$$F(A, B, C, D) = \sum m(0, 1, 2, 5, 7, 8, 9, 10, 13, 15)$$

AB CD

A	B	C	D	00	01	11	10
0			0	1	1	3	2
1			1	4	5	7	6
	1		1	12	13	15	14
		1	1	8	9	11	10
			1	1	1	1	1

$$F = BD + \bar{B}\bar{D} + \cancel{A\bar{B}\bar{C}\bar{D}}$$

$$\boxed{F = BD + \bar{B}\bar{D} + \bar{C}D}$$

Q → Minimize with the help of K-map:-

$$F(A, B, C, D) = \sum m(1, 3, 4, 6, 8, 9, 11, 13, 15) + \sum d(0, 2, 14)$$

A	B	C	D	00	01	11	10
0			0	1	1	1	X
1			1	4	5	7	1
	1		1	12	13	15	X
		1	1	8	9	11	10
			1	1	1	1	1

Don't care condition में यह गणना करने की वजह से 0 ताकि इसलिए यह 0 आवश्यक हो देता है कि साथ पारिं नहीं सकता है।

$$\boxed{F = A\bar{B} + A\bar{D} + \bar{B}\bar{C} + \bar{A}\bar{D}}$$

3. Groups must contain 1, 2, 4, 8 or 16 cells.

A	B	C	00	01	11	10
0	1	1	1	1	1	1
1						
X						

A	B	C	00	01	11	10
0	1	1	1	1	1	1
1						
✓						

4. Each group should be as large as possible.

A	B	C	00	01	11	10
0	1	1	1	1	1	1
1	1	1	1	1	1	1
X						

A	B	C	00	01	11	10
0	1	1	1	1	1	1
1	1	1	1	1	1	1
✓						

(This simplification does not give minimum literals)

Q) Simplify the following expression using K-map:-

$$F(ABCD) = \sum m(4, 5, 6, 7, 8, 9, 10, 11, 12, 13, 14, 15)$$

AB		CD			
		00	01	11	10
00	00	0	1	3	2
	01	1	1	1	1
	11	1	1	1	1
	10	1	1	1	1

$$F = A + B$$

Q) Simplify  $F(ABC) = \bar{A}BC + B\bar{C} + AB\bar{C} + A\bar{B}C$  using K-map in SOP form and POS form.

$\bar{B}\bar{C}$  - के लिए दोनों columns  
Consider करें।

A	B	C	$\bar{B}\bar{C}$	$\bar{B}C$	$BC$	$B\bar{C}$
$\bar{A}$	0	1	1	1	1	1
A	1	1	1	1	1	1

$$F = \sum m(2, 3, 5, 6)$$

$$F = \bar{A}B + BC + A\bar{B}C$$

This is in SOP form.

$$\text{POS form } F = \prod M(0, 1, 4, 7)$$

## Grouping of cells for simplification:-

⇒ Adjacent cells which have 1's can be grouped together.  
2's power.

i.e

$$2^0 = 1$$

$2^1 = 2$  adjacent cells can be grouped (Pair)

$2^2 = 4$ , " " " " " " (Quads)

$2^3 = 8$ , " " " " " " (Octads)

$2^4 = 16$ , " " " " " " (Hexads)

AB \ CD	00	01	11	10
00	1			
01		1		
11			1	1
10	1		1	1

AB \ CD	00	01	11	10
00	1			1
01		1	1	
11		1	1	
10	1		1	1

AB \ CD	00	01	11	10
00	1			
01		1		
11		1	1	
10	1		1	1

$$F = \bar{A}CB + BCD + A\bar{B}\bar{D}$$

$$F = \bar{D} + C$$

$$F = \cancel{\bar{B}D} + \bar{B}\bar{D}$$

## Rules followed for k-map simplification:-

1. Groups do not include only cell containing a zero.

A \ B	0	1
0	0	1
1		

A \ B	0	1
0	1	1
1	1	1

X

✓

2. Groups may be horizontal or vertical, but not diagonal.

A \ B	0	1
0	0	1
1	1	0

X

A \ B	0	1
0	0	1
1	1	1

✓