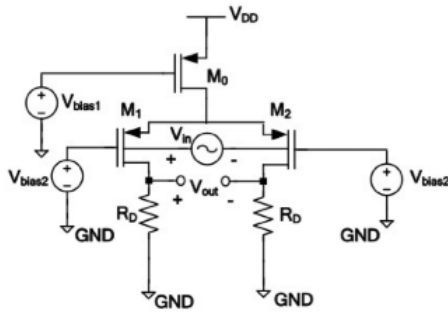


1. Consider the following differential amplifier where the small-signal input is applied to the bulk of  $M_1$  and  $M_2$ . For  $V_{bias1}=1.4V$ , and assuming that  $V_{bias2}$  is properly chosen so that all transistors are operating in their saturation region, calculate the small-signal differential gain of the amplifier.]

Recall that  $g_{mb} = \eta g_m$ .

Assume,  $\lambda = 0$ ,  $\eta = 0.2$ ,  $|V_{TH(PMOS)}| = 0.6V$  (the threshold value is in the presence of body effect),  $\mu_p C_{ox} = 100 \mu A/V^2$ ,  $R_D = 1k\Omega$ ,  $(W/L)_0 = 40$ ,  $(W/L)_1 = (W/L)_2 = 20$ , and  $V_{DD} = 3V$ .



$$I_{D1} = \frac{1}{2} \mu_p C_{ox} \cdot \left(\frac{W}{L}\right)_1 \cdot V_{eff,1}^2$$

$$\underline{M_1:} \quad V_{DS} > V_{GS} - V_{th}$$

$$V_{eff,1} = V_{GS} - V_{th}$$

$$V_D - V_S > V_G - V_S - V_{th}$$

$$V_G - V_S = 1.6$$

$$V_G = V_D + V_{th}$$

$$V_G = 1.6 + 0.8 = 2.4V$$

$$V_D = V_G - V_{th}$$

$$V_{bias2} = 2.4V$$

$$\underline{M_0:} \quad V_D = 1.4 - 0.6 = 0.8$$

Saturation

$$I_{D1} = \frac{1}{2} \cdot 100 \times 10^{-6} \cdot 20 \cdot 1^2$$

$$= 1mA$$

Recall that  $g_{mb} = \eta g_m$ .

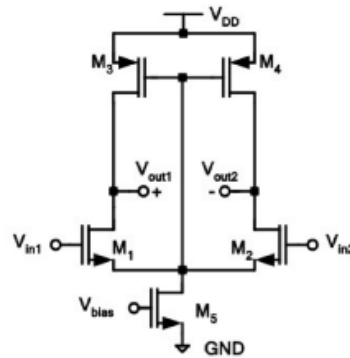
$$A_v = \frac{V_{out+} - V_{out-}}{V_{in}} = -g_{mb} \cdot R_D = -\eta \cdot g_m \cdot R_D$$

$$= -\eta \cdot \sqrt{2 \mu_p C_{ox} \left(\frac{W}{L}\right)_1 \cdot I_{D1}} \cdot R_D$$

$$= -0.2 \cdot \sqrt{2 \cdot 100 \times 10^{-6} \cdot 20 \cdot 1mA} \cdot 1k\Omega$$

$$= -0.4 V/V$$

2. In the following circuit all transistors have a W/L of  $7\mu\text{m}/0.35\mu\text{m}$  and  $M_3$  and  $M_4$  are to operate in deep triode region with an on-resistance of  $2\text{k}\Omega$ . Assume:  $I_5 = 40\mu\text{A}$  and  $\lambda = \gamma = 0$ ,  $V_{DD} = 3\text{V}$ ,  $V_{TH(NMOS)} = 0.5\text{V}$ ,  $V_{TH(PMOS)} = -0.6\text{V}$ ,  $\mu_n C_{ox} = 200\mu\text{A}/\text{V}^2$ ,  $\mu_p C_{ox} = 100\mu\text{A}/\text{V}^2$ .



- Calculate the dc level of the input (input common-mode level) that yields such on-resistance.
- Calculate the small-signal differential gain, i.e.,  $(V_{out1} - V_{out2}) / (V_{in1} - V_{in2})$ , of the circuit when the input common-mode level is equal to value calculated in part a.

$$a) I = \mu_p C_{ox} \frac{W}{L} V_{eff} \cdot V_{SD}, R_{on} = \frac{V_{SD}}{I} = \frac{1}{\mu_p C_{ox} \frac{W}{L} V_{eff}}$$

$$\hookrightarrow 2\text{k}\Omega = \frac{1}{100 \times 10^{-6} \cdot \frac{2}{0.25} \cdot V_{eff}} \rightarrow V_{eff} = 0.25\text{V}$$

$$V_{S63} = 0.6 + 0.25 = 0.85\text{V} \rightarrow V_{G3} = 3 - 0.85\text{V} = 2.15\text{V}$$

$$I_1 = \frac{I_5}{2} = \frac{40\text{nA}}{2} = 20\text{nA} = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{eff,1})^2 \rightarrow V_{eff,1} = 0.1\text{V}$$

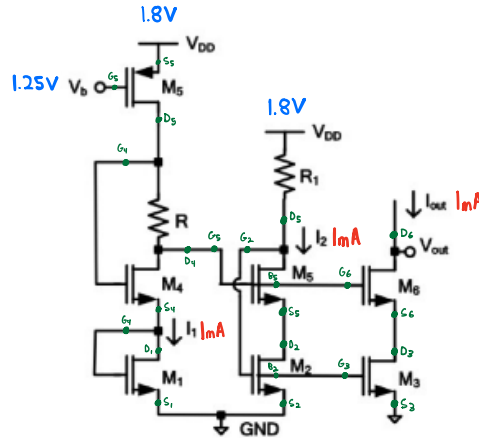
$$V_{GS1} = V_{thn} + V_{eff,1} = 0.5\text{V} + 0.1\text{V} = 0.6\text{V}$$

$$V_{i(\text{common mode})} = V_{GS1} + V_{S1} = 0.6\text{V} + 2.15\text{V} = 2.75\text{V}$$

$$b) A_v = -g_{m1} \cdot R_D = \frac{2 \cdot 20 \times 10^{-6}}{0.1} \cdot 2\text{k}\Omega = 0.8\text{V/V}$$

3. Consider the following wide-swing cascode current source. Assume that all NMOS transistors have the same size and  $I_1 = I_2 = I_{out} = 1 \text{ mA}$ . Furthermore, assume that the minimum voltage headroom required at the output node is  $0.5 \text{ V}$  (i.e., for output voltage as low as  $0.5 \text{ V}$  both transistors  $M_3$  and  $M_6$  are in saturation). The technology parameters are:

$\lambda_{(NMOS)} = \lambda_{(PMOS)} = 0 \text{ V}^{-1}$ ,  $\gamma = 0$ ,  $V_{DD} = 1.8 \text{ V}$ ,  $V_b = 1.25 \text{ V}$ ,  $V_{TH(NMOS)} = |V_{TH(PMOS)}| = 0.35 \text{ V}$ ,  $\mu_n C_{ox} = 1 \text{ mA/V}^2$ ,  $\mu_p C_{ox} = 0.5 \text{ mA/V}^2$ .



$-V_{DS} > V_{GS} - V_{th}$  SATURATION (Active)

channel is pinched off.

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})^2$$

Find the aspect ratio of the PMOS transistor and the NMOS transistors as well as the value of  $R$  and  $R_1$ .

$V_{out} = V_{DS6} + V_{DS3} \rightarrow V_{outmin} = V_{DS6min} + V_{DS3min} \rightarrow 0.5 \text{ V} = V_{DS6min} + V_{DS3min}$   
 $I_{out} = I_{D3} = I_{D6} = \frac{1}{2} \cdot 1 \times 10^{-3} \cdot \left(\frac{W}{L}\right) \cdot V_{eff}^2 = 1 \text{ mA}$   
 For Saturation:  $V_{DS} > V_{GS} - V_{th}$

$$V_{GS} = V_{DS} + V_{th}$$

$$= 0.25 + 0.35 = 0.6 \text{ V}$$

$$\therefore V_{eff} \text{ for } M_6, M_3 = 0.6 - 0.35 = 0.25 \text{ V}$$

For  $M_3, M_6$ :  $1 \text{ mA} = \frac{1}{2} \cdot 1 \times 10^{-3} \cdot \left(\frac{W}{L}\right) \cdot 0.25^2 \rightarrow \left(\frac{W}{L}\right)_{NMOS} = 32$

$M_5, M_2$ :  $V_{DS} = 0.25 \text{ V} \therefore V_{DS} = 0 + 0.25 \text{ V} + 0.25 \text{ V} = 0.5 \text{ V}$

$$\rightarrow \frac{1.8 \text{ V} - 0.5 \text{ V}}{R_1} = 1 \text{ mA} \rightarrow R_1 = 1.3 \text{ k}\Omega$$

$M_5$  in Saturation:  $V_{SG} = 1.8 \text{ V} - 1.25 \text{ V} = 0.55 \text{ V}$ ,  $V_{SD} = 0.55 - 0.35 = 0.2 \text{ V}$

$$V_{SD} = 0.2 \text{ V} = 1.8 - V_D \therefore V_D = 1.6 \text{ V} \rightarrow \frac{1.6 - 0.5}{R} = 1 \text{ mA} \rightarrow R = 1.1 \text{ k}\Omega$$

$$1 \text{ mA} = \frac{1}{2} \cdot 0.5 \times 10^{-3} \cdot \left(\frac{W}{L}\right) \cdot (0.55 - 0.35)^2 \rightarrow \left(\frac{W}{L}\right)_{PMOS} = 100$$