

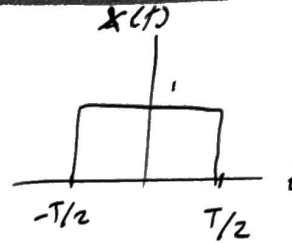
①

SES HW 8

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$= \int_{-T/2}^{T/2} 1 \cdot e^{-j\omega t} dt$$

$$= \left[ \frac{1}{-j\omega} e^{-j\omega t} \right]_{t=-T/2}^{t=T/2} = \frac{1}{-j\omega} e^{-j\omega T/2} + \frac{1}{j\omega} e^{j\omega T/2}$$



$$X(\omega) = \frac{1}{j\omega} (-e^{-j\omega T/2} + e^{j\omega T/2}) = \frac{1}{\omega} \cdot 2 \sin\left(\omega \frac{T}{2}\right) = T \frac{\sin\left(\omega \frac{T}{2}\right)}{\omega \frac{T}{2}} = \boxed{T \operatorname{sinc}\left(\omega \frac{T}{2}\right)}$$

$$T = 0.5$$

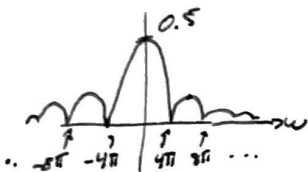
$$T = 1$$

$$T = 2$$

$$\frac{1}{2} \operatorname{sinc}\left(\frac{\omega}{4}\right)$$

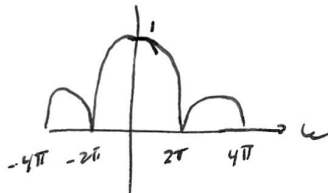
$$\operatorname{sinc}\left(\frac{\omega}{2}\right)$$

$$2 \operatorname{sinc}\left(\frac{\omega}{2}\right)$$

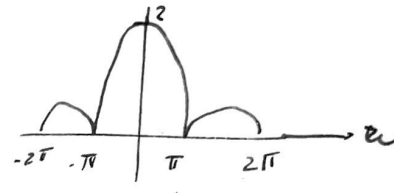


$$8\pi$$

main lobe width



$$4\pi$$



$$2\pi$$

taller and thinner

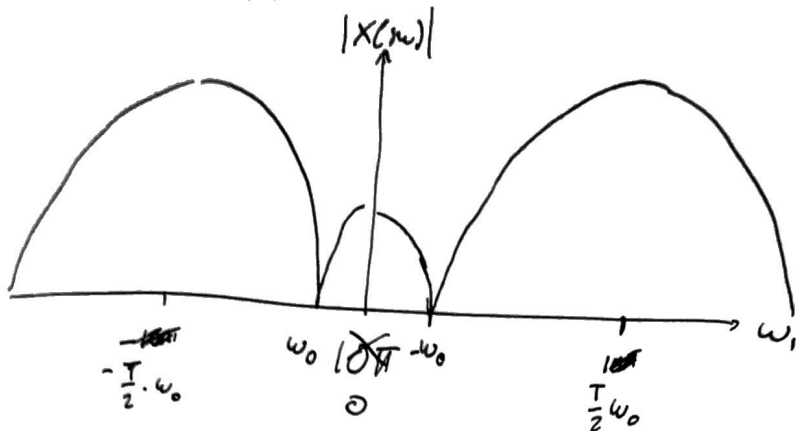
②

$$X(\omega) = \int_{-T/2}^{T/2} \cos(\omega_0 t) e^{-j\omega t} dt = \int_{-T/2}^{T/2} \frac{1}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t}) e^{-j\omega t} dt = \frac{1}{2} \int_{-T/2}^{T/2} e^{jt(\omega_0 - \omega)} + e^{-jt(\omega_0 + \omega)} dt$$

$$= \frac{1}{2} \left[ \frac{1}{j(\omega_0 - \omega)} e^{jt(\omega_0 - \omega)} + \frac{-1}{j(\omega_0 + \omega)} e^{-jt(\omega_0 + \omega)} \right]_{t=-T/2}^{t=T/2} = \frac{1}{2} \left[ \frac{e^{j\frac{T}{2}(\omega_0 - \omega)}}{j(\omega_0 - \omega)} - \frac{e^{-j\frac{T}{2}(\omega_0 + \omega)}}{j(\omega_0 + \omega)} \right]$$

$$= \frac{1}{2} \left[ \frac{e^{j\frac{T}{2}(\omega_0 - \omega)}}{j(\omega_0 - \omega)} - \frac{e^{-j\frac{T}{2}(\omega_0 + \omega)}}{j(\omega_0 + \omega)} - \frac{e^{-j\frac{T}{2}(\omega_0 - \omega)}}{j(\omega_0 - \omega)} + \frac{e^{j\frac{T}{2}(\omega_0 + \omega)}}{j(\omega_0 + \omega)} \right] = \frac{\sin\left(\frac{T}{2}(\omega_0 - \omega)\right)}{\omega_0 - \omega} + \frac{\sin\left(\frac{T}{2}(\omega_0 + \omega)\right)}{\omega_0 + \omega}$$

$$X(\omega) = \frac{T}{2} \left[ \operatorname{sinc}\left(\frac{T}{2}(\omega_0 - \omega)\right) + \operatorname{sinc}\left(\frac{T}{2}(\omega_0 + \omega)\right) \right] = \overset{0.05}{\cancel{10}} \left[ \operatorname{sinc}(10\pi - 0.05\omega) + \operatorname{sinc}(10\pi + 0.05\omega) \right]$$



Why is  $\frac{T}{2} \left( \sin\left(\frac{T}{2}(\omega_0 - \omega)\right) \right)$  equal to 0?

~~$0 = \sin(\omega_0 - \omega)$~~   
 ~~$0, \pi, 2\pi, \dots$~~   $\omega_0 - \omega = k\pi$

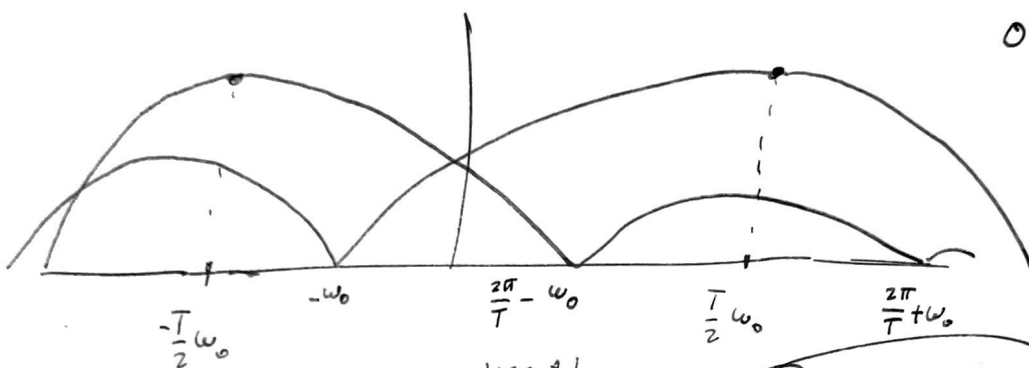
~~$\omega = \omega_0 - k\pi$~~   $k = 0, 1, 2, \dots$   
 $k = -2, -1, 0, 1, 2, \dots$

$0 = \sin\left(\frac{T}{2}(\omega_0 \pm \omega)\right)$   $k = \text{integer}$

$\frac{T}{2}\omega_0 \pm \frac{T}{2}\omega = k\pi$

$\pm \omega \frac{T}{2} = k\pi - \frac{T}{2}\omega_0$

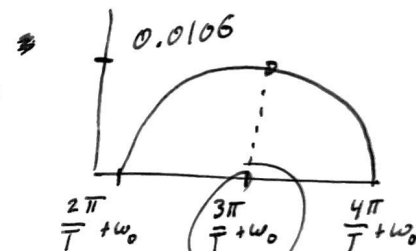
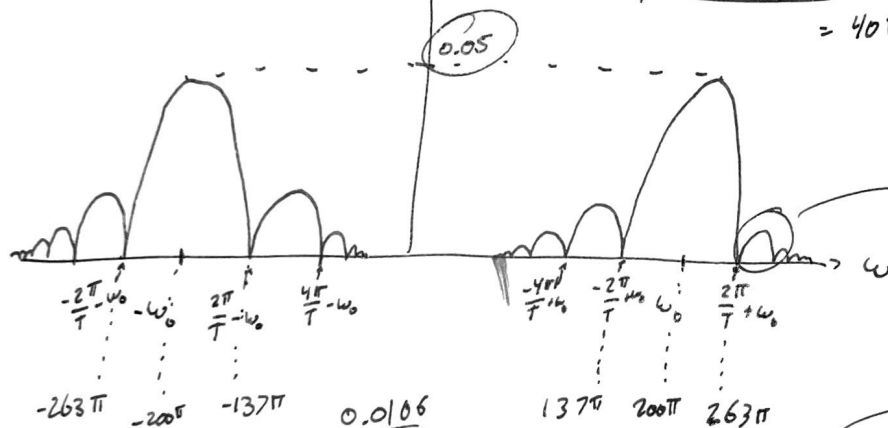
$\omega = \frac{2k\pi}{T} \pm \omega_0$



$|X(\omega)|$

width =  $\frac{4\pi}{T}$

$= 40\pi \approx 125.664 \text{ rad} = 20 \text{ kHz}$



plus in  $\omega = \dots$  to  $\sin\left(\frac{T}{2}(\omega_0 - \omega)\right)$

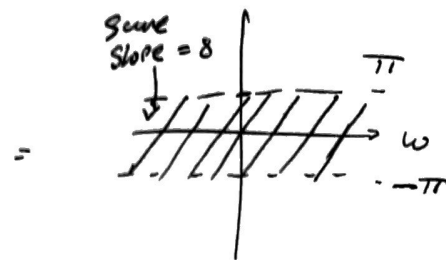
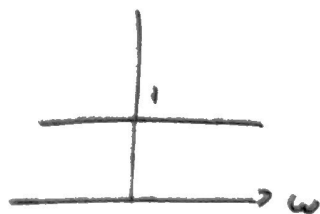
$|X(\omega)| = 0.0106$

side lobe dB =  $20 \log_{10} \left( \frac{0.05}{0.0106} \right) = \boxed{-13.465 \text{ dB}}$

$$(3) X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \delta(n-8) e^{-j\omega n} = \boxed{e^{-j\omega 8}} = \cos(8\omega) + j\sin(8\omega)$$

Magnitude = 1

phase =  $-8\omega$



delay in time  $\longrightarrow$  shift shift in phase dependent on frequency

$$a_k = \frac{1}{8} \sum_{n=0}^{7} x(n) e^{-jk \frac{2\pi}{8} n} = \frac{1}{8} \sum_{n=0}^{7} \delta(n-8) e^{-jk \frac{2\pi}{8} n} = \frac{1}{8} \sum_{k=0}^{7} 1 \cdot e^{-jk \frac{2\pi}{8} \cdot 8} = \frac{1}{8} \sum_{k=0}^{7} 1 \cdot e^{-jk 2\pi} = \frac{1}{8} \sum_{k=0}^{7} 1$$

always  $e^0$

$$\rightarrow \boxed{a_k = \frac{1}{8} \text{ for } k=0 \text{ to } 7}$$

try  $a_k = \frac{1}{N} \sum_{n=0}^{N-1} e^{-jk \frac{2\pi}{N} \cdot n_0}$  Constant  $a_k$  if  $n_0 = N$  or  $n_0 = Z \cdot N$

$Z \in \text{integers}$

Order:  $a_k = \frac{1}{8} \sum_{n=0}^{7} 1 \cdot e^0 \rightarrow \boxed{a_k = \frac{1}{8}}$

$x(0)$   
 $= x(8) = 1$

```
clear;clc;
x = [0 0 0 0 0 0 0 0 1];
N = 8;
k = 0:N-1;
```

```
n = 0:N-1;
X = 1/N*sum(x(1:N).*exp(-1j.*k*2*pi/N.*n)) % x[0] to x[7] are all zero.
```

```
X =
0
```

```
x_circ = [1 0 0 0 0 0 0 0];
X_circ = 1/N*sum(x_circ(1:N).*exp(-1j.*k*2*pi/N.*n)) % x[8] = x[0] = 1, rest are
zero
```

```
X_circ =
0.1250
```

$$(1) X(e^{j\omega}) = \sum_{n=-M}^M x[n] e^{-j\omega n} = \sum_{n=0}^{M-1} \cos(\omega_0 n) e^{-j\omega n} \quad \omega_0 = \frac{2\pi}{T_0}$$

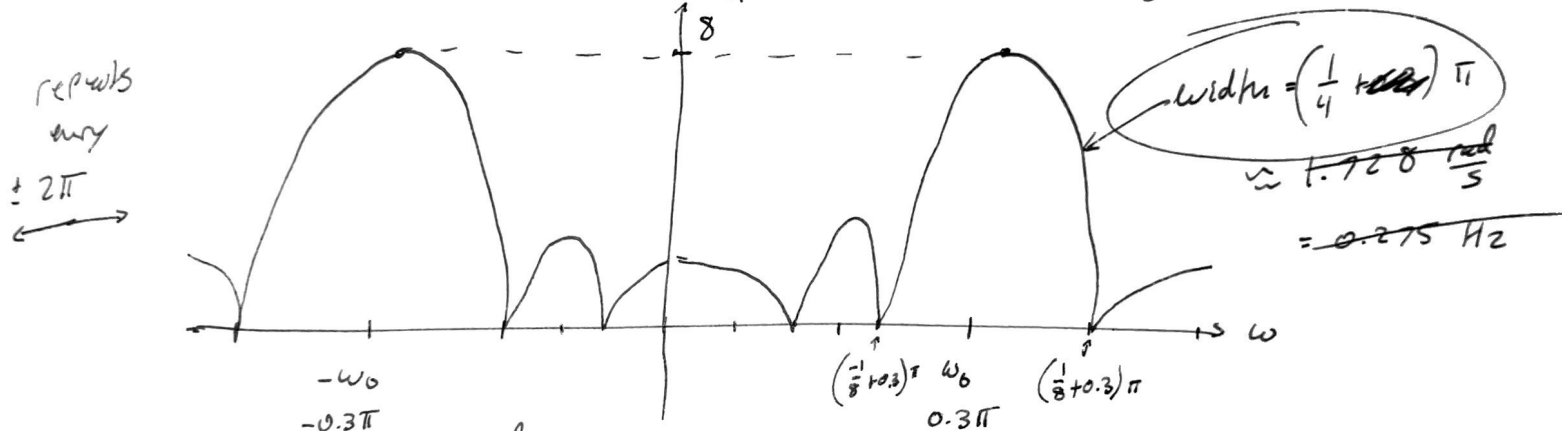
$$= \sum_{n=0}^{M-1} \frac{1}{2} [e^{j\omega_0 n} + e^{-j\omega_0 n}] e^{-j\omega n} = \frac{1}{2} \sum_{n=0}^{M-1} e^{jn(\omega_0 - \omega)} + e^{-jn(\omega_0 + \omega)}$$

$$= \frac{1}{2} \left[ \sum_{n=0}^{M-1} e^{jn(\omega_0 - \omega)} + \sum_{n=0}^{M-1} e^{-jn(\omega_0 + \omega)} \right] = \frac{1}{2} \left[ \frac{1 - e^{jM(\omega_0 - \omega)}}{1 - e^{j(\omega_0 - \omega)}} + \frac{1 - e^{-jM(\omega_0 + \omega)}}{1 - e^{-j(\omega_0 + \omega)}} \right]$$

$$= \frac{1}{2} \left[ e^{j(\omega_0 - \omega) \frac{M-1}{2}} \frac{\sin\left(\frac{M(\omega - \omega_0)}{2}\right)}{\sin\left(\frac{\omega - \omega_0}{2}\right)} + e^{-j(\omega_0 + \omega) \frac{M-1}{2}} \frac{\sin\left(\frac{M(\omega + \omega_0)}{2}\right)}{\sin\left(\frac{\omega + \omega_0}{2}\right)} \right]$$

no effect on magnitude

Dirichlet kernel identity



height:  $\frac{1}{2} \left[ \frac{\sin\left(\frac{M(\omega - \omega_0)}{2}\right)}{\sin\left(\frac{\omega - \omega_0}{2}\right)} + \frac{\sin\left(\frac{M(\omega + \omega_0)}{2}\right)}{\sin\left(\frac{\omega + \omega_0}{2}\right)} \right]$

intercepts:  $\sin\left(\frac{M(\omega \pm \omega_0)}{2}\right) = 0 \rightarrow \frac{M(\omega \pm \omega_0)}{2} = K\pi$

except when  $M =$

$\lim_{\omega \rightarrow \omega_0} \left[ \frac{\sin\left(\frac{M(\omega - \omega_0)}{2}\right)}{\sin\left(\frac{\omega - \omega_0}{2}\right)} \cdot \frac{1}{2} \right] = 8$

$\omega = 2 \frac{K}{M} \pi \pm \omega_0$

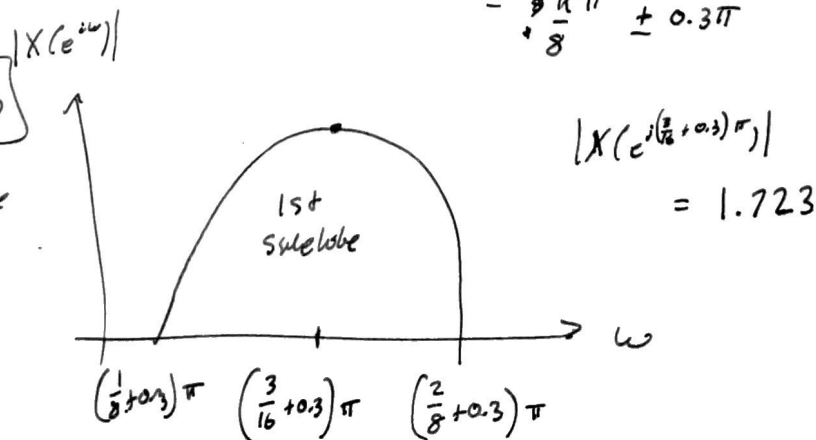
$= \frac{2}{8} K \pi \pm 0.3\pi$

$K = \text{integer}$

$100 - 200$

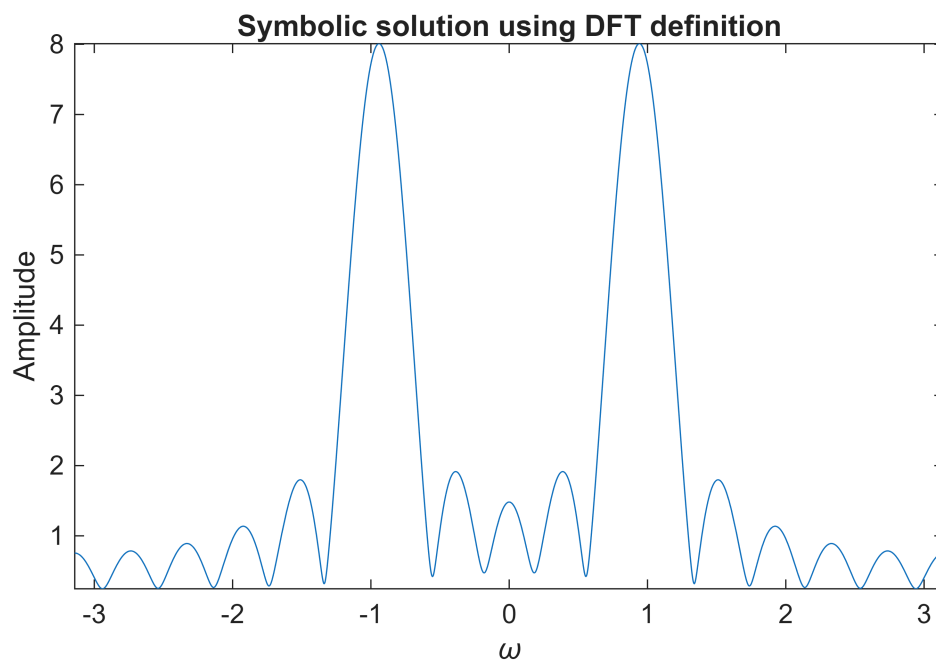
$20 \log_{10} \left( \frac{1.723}{8} \right) = -13.34 \text{ dB}$

close to -13.26. I approximate the peak of the side lobe to be at the mid point.

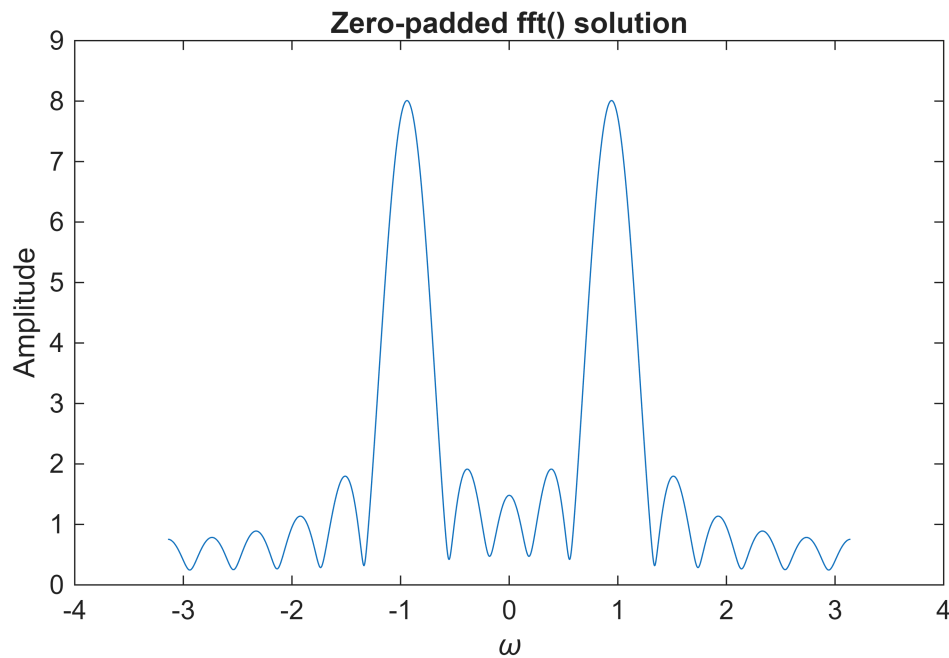


```
clear; clc;
w0 = 0.3*pi;
M = 16;
n = 0:M-1;
syms w
x = cos(w0.*n);
```

```
% solve symbolically
X = sum(x(n+1).*exp(-1j.*w.*n));
fplot(abs(X), [-pi pi])
xlabel('\omega'); ylabel('Amplitude'); title('Symbolic solution using DFT
definition');
```



```
% mucho zero padding
N = 4096;
X = fft(x, N);
plot(linspace(-pi,pi, N), abs(fftshift(X)))
xlabel('\omega'); ylabel('Amplitude'); title('Zero-padded fft() solution');
```



```
[mag, idx] = max(abs(X));
Mainlobe_height = mag
```

```
Mainlobe_height =
8.0083
```

```
min_inds = find(islocalmin(abs(X)));
right = min_inds(min_inds > idx);
left = min_inds(min_inds < idx);
Mainlobe_Width = 2*pi/N*(right(1) - left(end))
```

```
Mainlobe_Width =
0.7823
```

```
[pks, ~] = findpeaks(abs(X));
pks = sort(pks, "descend");
Sidelobe_height = pks(3)
```

```
Sidelobe_height =
1.9149
```

```
Sidelobe_dB = 20*log10(Sidelobe_height/Mainlobe_height)
```

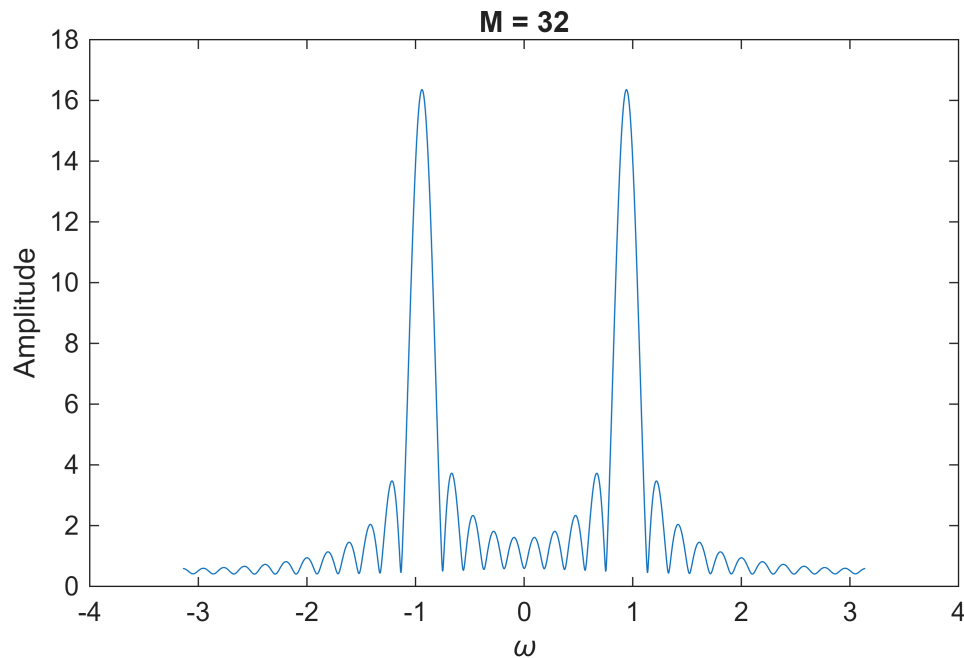
```
Sidelobe_dB =
-12.4278
```

```
% Double M
M = 2 * M;
n = 0:M-1;
x = cos(w0.*n);
```

```

X = fft(x, N);
plot(linspace(-pi,pi, N), abs(fftshift(X)))
xlabel('\omega'); ylabel('Amplitude'); title("M = " + num2str(M));

```



```

[mag, idx] = max(abs(X));
Mainlobe_height = mag

```

```

Mainlobe_height =
16.3534

```

```

min_inds = find(islocalmin(abs(X)));
right = min_inds(min_inds > idx);
left = min_inds(min_inds < idx);
Mainlobe_Width = 2*pi/N*(right(1) - left(end))

```

```

Mainlobe_Width =
0.3850

```

```

[pks, ~] = findpeaks(abs(X));
pks = sort(pks, "descend");
Sidelobe_height = pks(3)

```

```

Sidelobe_height =
3.7248

```

```

Sidelobe_dB = 20*log10(Sidelobe_height/Mainlobe_height)

```

```

Sidelobe_dB =
-12.8500

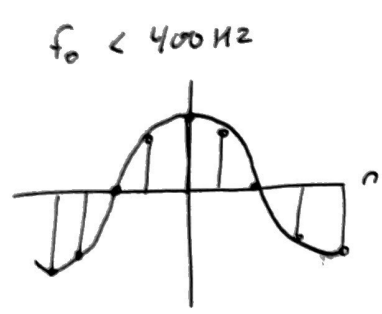
```



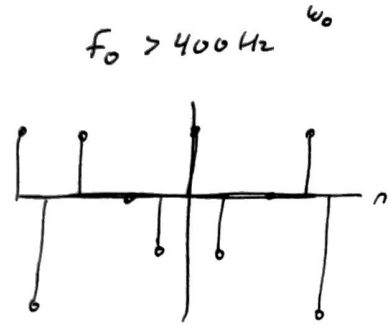
The main lobe is approximately half as wide and twice as tall. The sidelobe doubled in height as well, so the dB value remains fairly similar.

It looks like a larger  $M$  (larger window, larger number of samples) results in a better DFT.

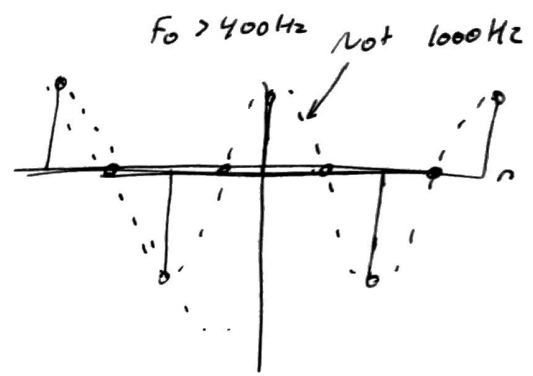
⑤  $x(t) = \cos(\omega_0 t)$   $T_s = \frac{1}{f_s}$   $x[n] = \cos(\underbrace{\omega_0}_{\omega_0} 2\pi \frac{f_0}{f_s} n) = \cos(\omega_0 n)$



100 Hz



500 Hz



800 Hz  
1000

Aliasing occurs because the original 500 Hz and 1000 Hz signals cannot be seen in just the samples. Instead, we see lower frequency signals that are Aliases of the original  $f_0$ .

---

$f_{\text{alias}} = |f_0 - K f_s|$   $K$  is smallest integer such that

$f_{\text{alias}} < \frac{f_s}{2}$  ← Nyquist

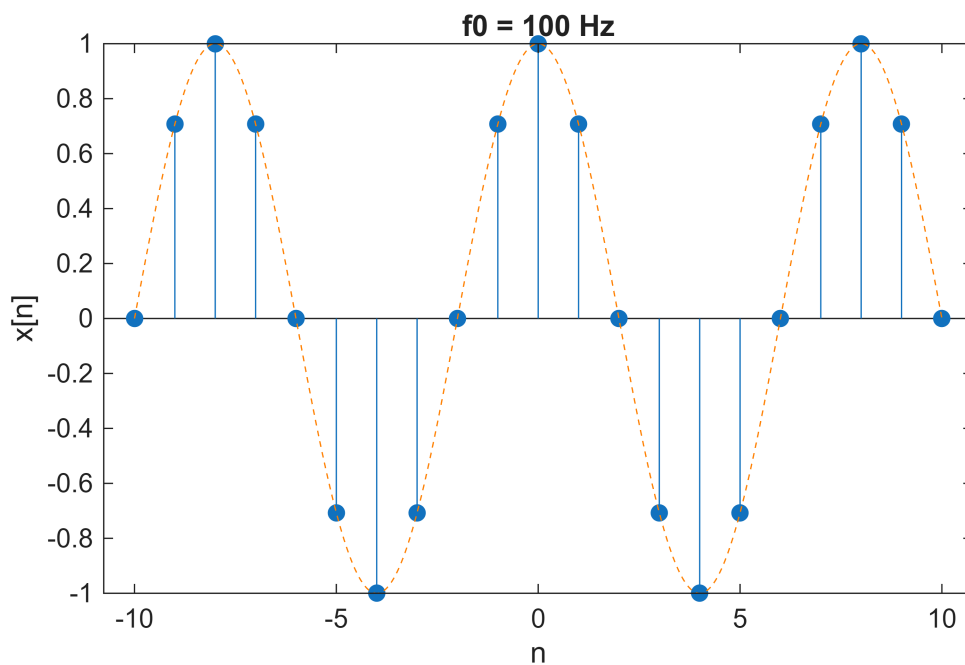
---

```
clc; clear;
n = -10:10;
t = -10:0.01:10;
```

```
w0 = 2*pi*1/8;

xn = cos(w0*n);
xt = cos(w0*t);
stem(n,xn,"filled"); hold on;
plot(t,xt,'LineStyle','--','Color',[1 0.5 0 0.5]); hold off;

xlabel('n'); ylabel('x[n]'); title('f0 = 100 Hz');
```



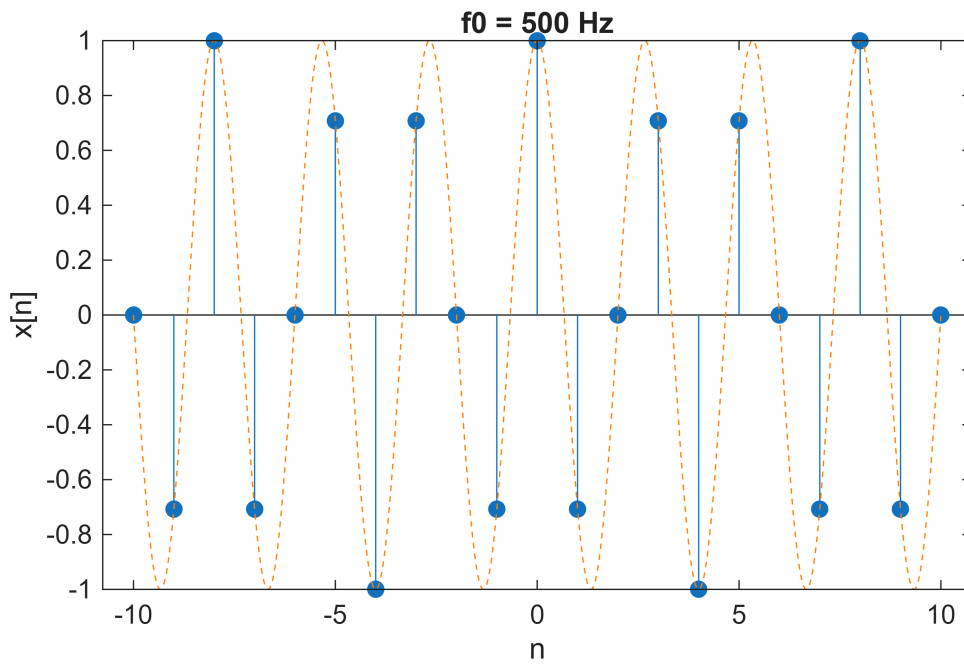
```
f0_reconstructed = abs(w0/(2*pi)*800 - 0*800)
```

```
f0_reconstructed =
100
```

```
w0 = 2*pi*5/8;

xn = cos(w0*n);
xt = cos(w0*3/5*t);
stem(n,xn,"filled"); hold on;
plot(t,xt,'LineStyle','--','Color',[1 0.5 0 0.5]); hold off;

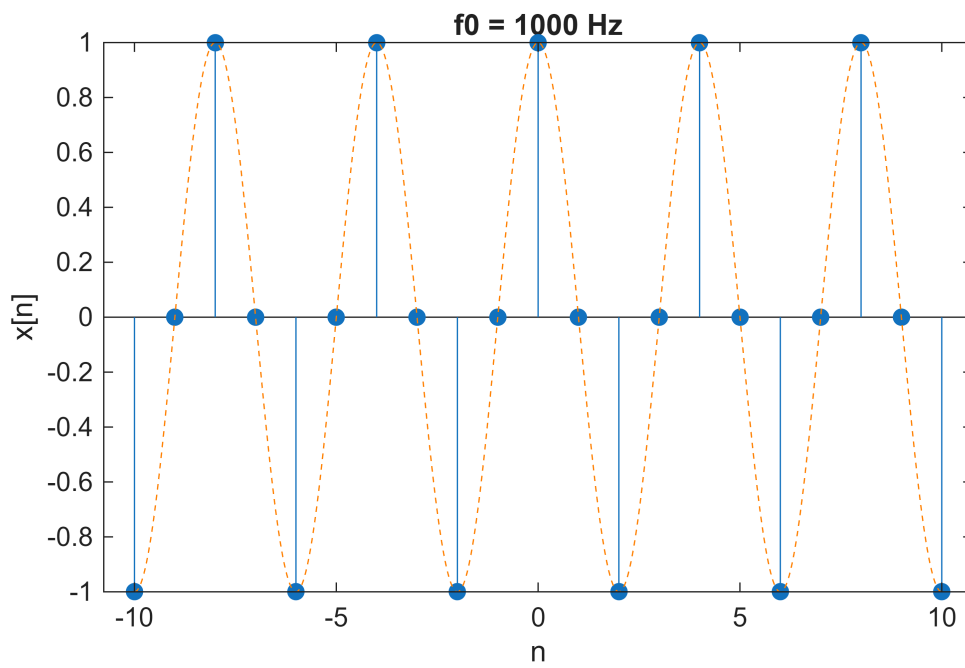
xlabel('n'); ylabel('x[n]'); title('f0 = 500 Hz');
```



```
f0_reconstructed = abs(w0/(2*pi)*800 - 1*800)
```

```
f0_reconstructed =  
300
```

```
w0 = 2*pi*10/8;  
  
xn = cos(w0*n);  
xt = cos(w0*1/5*t);  
stem(n,xn,"filled"); hold on;  
plot(t,xt,'LineStyle','--','Color',[1 0.5 0 0.5]); hold off;  
  
xlabel('n'); ylabel('x[n]'); title('f0 = 1000 Hz');
```



```
f0_reconstructed = abs(w0/(2*pi)*800 - 1*800)
```

```
f0_reconstructed =  
200
```

## Question 6

Answers at end of this script

### DOWNCONVERSION

```
clear; clc;

fs = 8000; % Hz
Ts = 1/fs;
N = 4096;
t = (0:N-1)*Ts;

% Pick a tone that will alias after downsampling
f0 = 0.35*fs; % 2.8 kHz
x = cos(2*pi*f0*t);

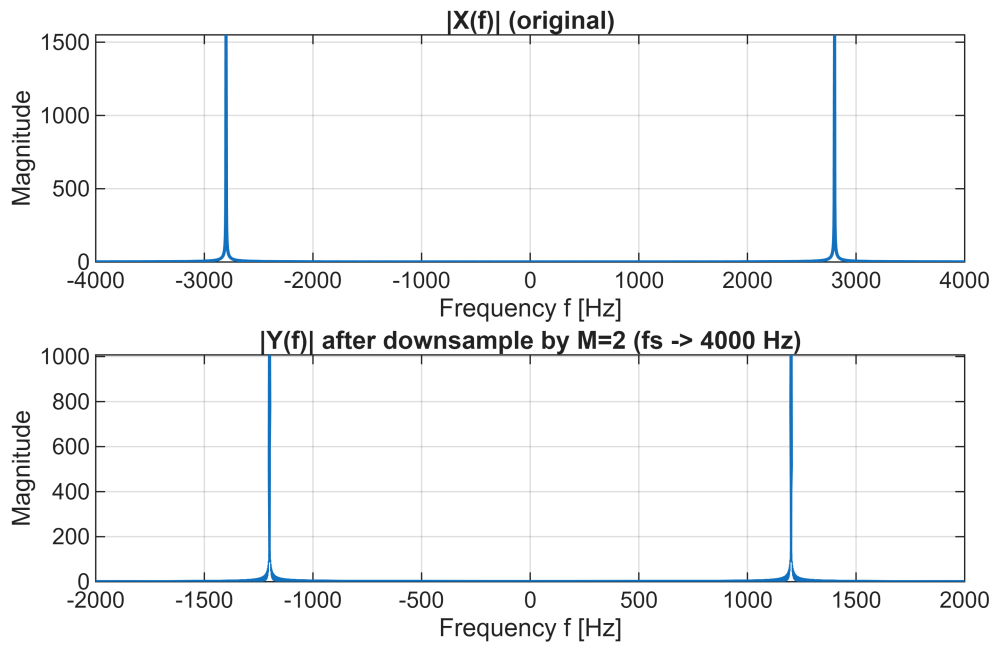
% downsample by M=2
M = 2;
y = x(1:M:end);
fs_y = fs/M;
Ty = 1/fs_y;

% Zero padding
yZ = zeros(1,N);
yZ(1:length(y)) = y;

X = fftshift(fft(x, N));
Y = fftshift(fft(yZ, N));
fx = linspace(-fs/2, fs/2, N);
fy = linspace(-fs_y/2, fs_y/2, N);

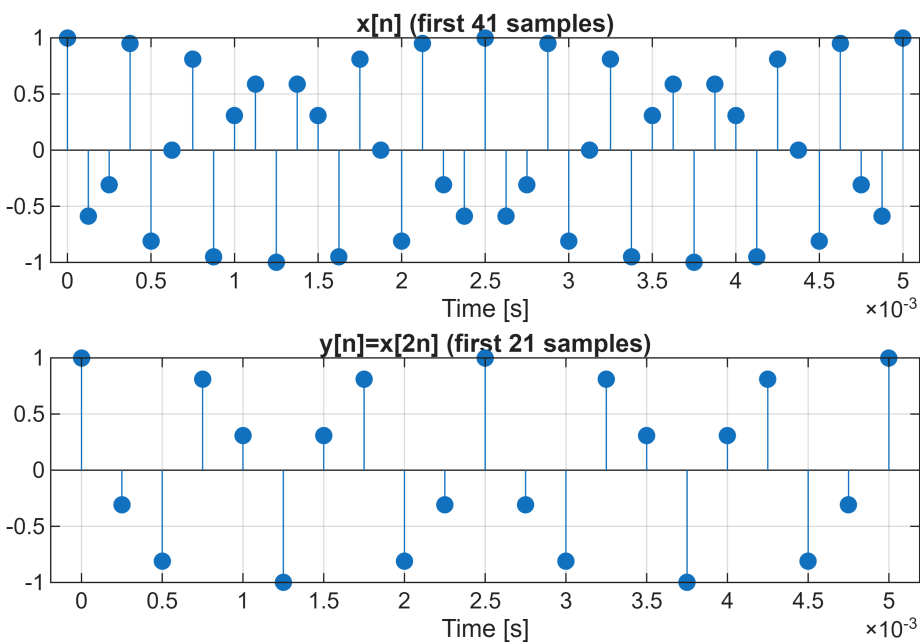
figure('Color','w');
subplot(2,1,1); plot(fx, abs(X), 'LineWidth',1.2); grid on;
title('|X(f)| (original)'); xlabel('Frequency f [Hz]'); ylabel('Magnitude');

subplot(2,1,2); plot(fy, abs(Y), 'LineWidth',1.2); grid on;
title(sprintf('|Y(f)| after downsample by M=%d (fs -> %.0f Hz)', M, fs_y));
xlabel('Frequency f [Hz]'); ylabel('Magnitude');
```



```
% time domain view
figure('Color','w');
subplot(2,1,1);
stem(t(1:41), x(1:41), 'filled'); grid on;
title('x[n] (first 41 samples)'); xlabel('Time [s]');

subplot(2,1,2);
ty = (0:length(y)-1)*Ty;
stem(ty(1:21), y(1:21), 'filled'); grid on;
title('y[n]=x[2n] (first 21 samples)'); xlabel('Time [s]');
```



# UPSAMPLING

```
clear; clc;

fs = 8000;
Ts = 1/fs;
N = 4096;
t = (0:N-1)*Ts;
```

```
% Tone that is clear after upsampling
f0 = 0.15*fs; % 1200 Hz for fs = 8 kHz
x = cos(2*pi*f0*t);
```

```
% Upsample
L = 3;
xu = zeros(1, L*length(x));
xu(1:L:end) = x;

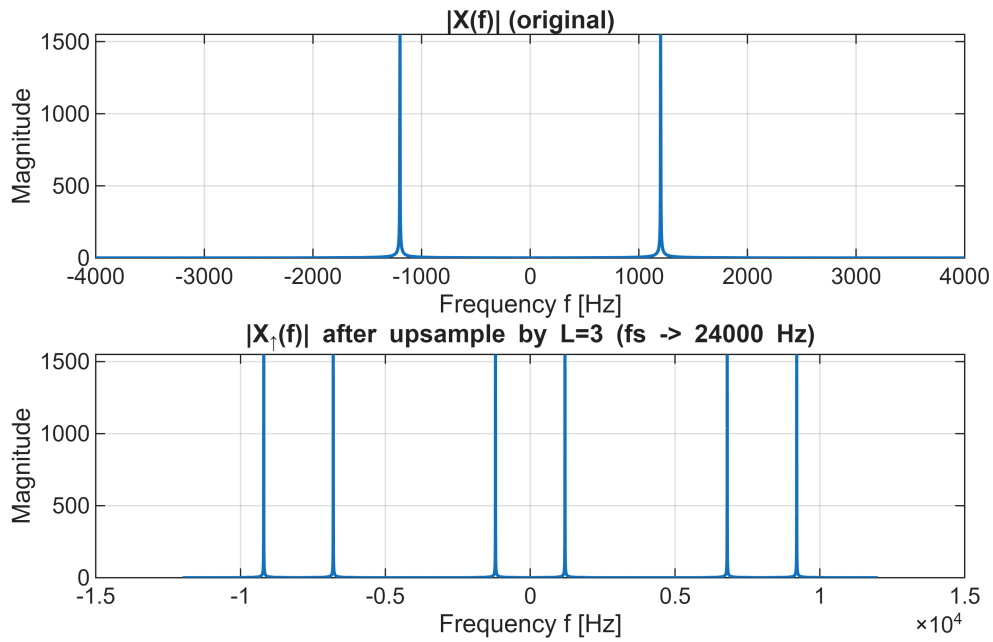
fs_u = L*fs;
Tu = 1/fs_u;
Nu = L*N;
```

```
X = fftshift(fft(x,N));
Xu = fftshift(fft(xu, Nu));
fx = linspace(-fs/2, fs/2, N);
fu = linspace(-fs_u/2, fs_u/2, Nu);
```

```
figure('Color','w');
subplot(2,1,1); plot(fx, abs(X), 'LineWidth',1.2); grid on;
title('|X(f)| (original)'); xlabel('Frequency f [Hz]'); ylabel('Magnitude');

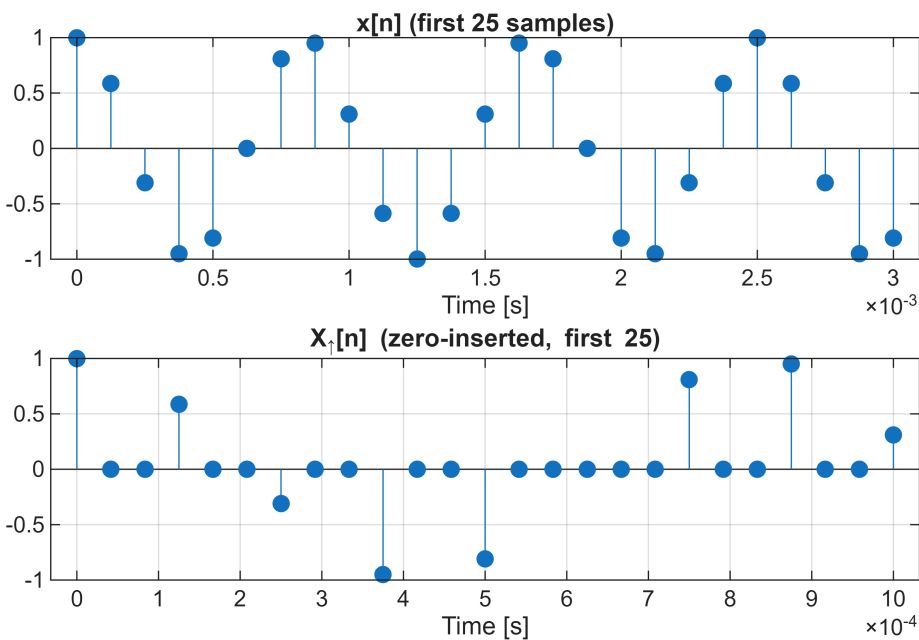
subplot(2,1,2); plot(fu, abs(Xu), 'LineWidth',1.2); grid on;
title(sprintf('|X_{\uparrow}(f)| after upsample by L=%d (fs -> %.0f Hz)', L,
fs_u));
xlabel('Frequency f [Hz]'); ylabel('Magnitude');
```





```
% time domain view
figure('Color','w');
subplot(2,1,1);
stem(t(1:25), x(1:25), 'filled'); grid on;
title('x[n] (first 25 samples)'); xlabel('Time [s]');

subplot(2,1,2);
tu = (0:length(xu)-1)*Tu;
stem(tu(1:25), xu(1:25), 'filled'); grid on;
title('X↑[n] (zero-inserted, first 25)'); xlabel('Time [s]');
```



# ANSWERS

## Downsampling

- The new apparent tone frequency is at 1200 Hz
- $|Y(f)|$  looks compressed/overlapped near the base of the tone peak because original tone has folded into the upper and lower frequencies, due to the decrease in sampling frequency when downsampling.
- The bandlimit on  $|X(f)|$  to avoid aliasing for  $M=2$  is less than  $1/4$  the original sample frequency. For original  $f_s = 8000$  Hz, the highest frequency  $|X(f)|$  is 2000 Hz. When downsampling occurs and the new sampling frequency is 4000 Hz, no aliasing will occur as  $|X(f)|$  fulfills the Nyquist-Shannon sampling theorem.

## Upsampling

- 3 images appear, although one is the same as the original  $|X(f)|$  at  $\pm 1.2$  kHz.
- The other two images are  $\pm 8000$  Hz from 0 Hz, which was the original sampling frequency  $f_s$ . Each image also contains 2 tones, each one at  $\pm 8.0$  kHz  $\pm 1.2$  kHz.
- A low-pass filter with cutoff frequency of  $\sim 1.5$  kHz would attenuate the outer images, leaving only the original.