## Arie, A. Verhoeven

## Error analysis of BDF Compound-Fast multirate method for differential-algebraic equations

HG 8 47 (Technische Universiteit Eindhoven)

Den Dolech 2

5600 MB Eindhoven

The Netherlands

averhoev@win.tue.nl

Jan, E.J.W. ter Maten

Bob, R.M.M. Mattheij

Beelen Theo, T.G.J. El Guennouni Ahmed, A. Tasić Bratislav B.

Analogue electrical circuits are usually modeled by differential-algebraic equations (DAE) of type:

$$\frac{d}{dt}\left[\mathbf{q}(t,\mathbf{x})\right] + \mathbf{j}(t,\mathbf{x}) = \mathbf{0},\tag{1}$$

where  $\mathbf{x} \in \mathbb{R}^d$  represents the state of the circuit. A common analysis is the transient analysis, which computes the solution  $\mathbf{x}(t)$  of this non-linear DAE along the time interval [0,T] for a given initial state. Often, parts of electrical circuits have latency or multirate behaviour.

For a multirate method it is necessary to partition the variables and equations into an active (A) and a latent (L) part. The active and latent parts can be expressed by  $\mathbf{x}_A = \mathbf{B}_A \mathbf{x}, \mathbf{x}_L = \mathbf{B}_L \mathbf{x}$  where  $\mathbf{B}_A, \mathbf{B}_L$  are permutation matrices. Then equation (1) is written as the following partitioned system:

$$\frac{\frac{d}{dt}\left[\mathbf{q}_{A}(t,\mathbf{x}_{A},\mathbf{x}_{L})\right] + \mathbf{j}_{A}(t,\mathbf{x}_{A},\mathbf{x}_{L}) = \mathbf{0},$$

$$\frac{d}{dt}\left[\mathbf{q}_{L}(t,\mathbf{x}_{A},\mathbf{x}_{L})\right] + \mathbf{j}_{L}(t,\mathbf{x}_{A},\mathbf{x}_{L}) = \mathbf{0}.$$
(2)

In contradiction to classical integration methods, multirate methods integrate both parts with different stepsizes. Besides the coarse time-grid  $\{T_n, 0 \le n \le N\}$  with stepsizes  $H_n = T_n - T_{n-1}$ , also a refined time-grid  $\{t_{n-1,m}, 1 \le n \le N, 0 \le m \le q_n\}$  is used with stepsizes  $h_{n,m} = t_{n,m} - t_{n,m-1}$  and multirate factors  $q_n$ . If the two time-grids are synchronized,  $T_n = t_{n,0} = t_{n-1,q_n}$  holds for all n. There are a lot of multirate approaches for partitioned systems but we will consider the Compound-Fast version of the BDF methods. This method performs the following four steps:

- 1. The complete system is integrated at the coarse time-grid.
- 2. The latent interface variables are interpolated at the refined time-grid.

- 3. The active part is integrated at the refined time-grid, using the interpolated values at the latent interface.
- 4. The active solution at the coarse time-grid is updated.

The local discretization error  $\delta^n$  of the compound phase still has the same behaviour  $\delta^n = O(H_n^{K+1})$ . Let  $\bar{\mathbf{P}}^n, \bar{\mathbf{Q}}^n$  be the Nordsieck vectors which correspond to the predictor and corrector polynomials of  $\mathbf{q}$ . Then the error  $\delta^n$  can be estimated by  $\hat{\delta}^n$ :

$$\hat{\delta}^n = \frac{-H_n}{T_n - T_{n-K-1}} \left[ \bar{\mathbf{Q}}_1^n - \bar{\mathbf{P}}_1^n \right]. \tag{3}$$

Now  $\hat{r}_C^n = \|\mathbf{B}_L \hat{\delta}^n\| + \tau \|\mathbf{B}_A \hat{\delta}^n\|$  is the used weighted error norm, which must satisfy  $\hat{r}_C^n < \text{TOL}_C$ .

The local discretization error  $\delta^{n,m}$  is defined as the residue after inserting the exact solution in the refinement BDF scheme. During the refinement instead of  $\delta^{n,m}$  the perturbed local error  $\tilde{\delta}^{n,m}$  is estimated. A tedious analysis yields the following asymptotic behaviour:

$$\mathbf{B}_{A}\delta^{n-1,m} \doteq \mathbf{B}_{A}\tilde{\delta}^{n-1,m} + \frac{1}{4}h\mathbf{K}_{n-1,m}\mathbf{B}_{L}\rho^{n-1,m}.$$
 (4)

Here  $\rho^{n-1,m}$  is the interpolation error at the refined grid and  $\mathbf{K}_{n-1,m}$  is the coupling matrix. The perturbed local discretization error  $\mathbf{B}_A \tilde{\delta}^{n,m}$  behaves as  $O(h_{n-1,m}^{k+1})$  and can be estimated in a similar way as  $\delta^n$ . Thus the active error estimate  $\mathbf{B}_A \hat{\delta}^{n-1,m}$  satisfies  $\mathbf{B}_A \hat{\delta}^{n-1,m} \doteq \mathbf{B}_A \hat{\delta}^{n-1,m} + \frac{1}{4} h \hat{\mathbf{K}}_{n-1,m} \mathbf{B}_L \hat{\rho}^{n-1,m}$ . Let L be the interpolation order, then it can be shown that  $\frac{1}{4} \|\hat{\mathbf{K}}_n \mathbf{B}_L \rho^{n-1,m}\|$  is less than

$$\hat{r}_{I}^{n} = \frac{1}{4} \frac{H_{n}}{T_{n} - T_{n-L-1}} \| \hat{\mathbf{K}}_{n} \mathbf{B}_{L} \left[ \bar{\mathbf{X}}_{1}^{n} - \bar{\mathbf{Y}}_{1}^{n} \right] \|.$$
 (5)

Here  $\bar{\mathbf{Y}}^n, \bar{\mathbf{X}}^n$  are the Nordsieck vectors which correspond to the predictor and corrector polynomials of  $\mathbf{x}$ . This error estimate  $\hat{r}_I^n$  has the asymptotic behaviour  $\hat{r}_I^n = O(H_n^{L+1})$ . It follows that  $\|\mathbf{B}_A \hat{\delta}^{n,m}\|$  satisfies:

$$\|\mathbf{B}_{A}\hat{\delta}^{n-1,m}\| \le \hat{r}_{A}^{n-1,m} + h\hat{r}_{I}^{n} =: \hat{r}_{A}^{n-1,m}.$$
 (6)

If 
$$\hat{r}_I^n \leq \text{TOL}_I = \sigma \text{TOL}_A$$
 and  $\hat{r}_A^{n-1,m} \leq \text{TOL}_A = (1 - \sigma h) \text{TOL}_A$  then  $\hat{r}_A^{n-1,m} \leq \text{TOL}_A + h \text{TOL}_I = \text{TOL}_A$ .

We tested a circuit with  $5\times 10$  inverters. The location of the active part is controlled by the connecting elements and the voltage sources. The connecting elements were chosen such that the active part consists of 3 inverters. We did an Euler Backward Compound-Fast multirate simulation on  $[0,10^{-8}]$  with  $\sigma=0.5, \tau=0$ . We get accurate results combined with a speedup factor 13.