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Error analysis of BDF Compound-Fast multirate method for differential-algebraic equations

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Analogue electrical circuits are usually modeled by differential-algebraic equations (DAE) of type:

$$\frac{d}{dt}\left[\mathbf{q}(t,\mathbf{x})\right] + \mathbf{j}(t,\mathbf{x}) = \mathbf{0},\tag{1}$$

where $\mathbf{x} \in \mathbb{R}^d$ represents the state of the circuit. A common analysis is the transient analysis, which computes the solution $\mathbf{x}(t)$ of this non-linear DAE along the time interval [0,T] for a given initial state. Often, parts of electrical circuits have latency or multirate behaviour.

For a multirate method it is necessary to partition the variables and equations into an active (A) and a latent (L) part. The active and latent parts can be expressed by $\mathbf{x}_A = \mathbf{B}_A \mathbf{x}, \mathbf{x}_L = \mathbf{B}_L \mathbf{x}$ where $\mathbf{B}_A, \mathbf{B}_L$ are permutation matrices. Then equation (1) is written as the following partitioned system:

$$\frac{\frac{d}{dt}\left[\mathbf{q}_{A}(t,\mathbf{x}_{A},\mathbf{x}_{L})\right] + \mathbf{j}_{A}(t,\mathbf{x}_{A},\mathbf{x}_{L}) = \mathbf{0},$$

$$\frac{d}{dt}\left[\mathbf{q}_{L}(t,\mathbf{x}_{A},\mathbf{x}_{L})\right] + \mathbf{j}_{L}(t,\mathbf{x}_{A},\mathbf{x}_{L}) = \mathbf{0}.$$
(2)

In contradiction to classical integration methods, multirate methods integrate both parts with different stepsizes. Besides the coarse time-grid $\{T_n, 0 \le n \le N\}$ with stepsizes $H_n = T_n - T_{n-1}$, also a refined time-grid $\{t_{n-1,m}, 1 \le n \le N, 0 \le m \le q_n\}$ is used with stepsizes $h_{n,m} = t_{n,m} - t_{n,m-1}$ and multirate factors q_n . If the two time-grids are synchronized, $T_n = t_{n,0} = t_{n-1,q_n}$ holds for all n. There are a lot of multirate approaches for partitioned systems but we will consider the Compound-Fast version of the BDF methods. This method performs the following four steps:

- 1. The complete system is integrated at the coarse time-grid.
- 2. The latent interface variables are interpolated at the refined time-grid.

- 3. The active part is integrated at the refined time-grid, using the interpolated values at the latent interface.
- 4. The active solution at the coarse time-grid is updated.

The local discretization error δ^n of the compound phase still has the same behaviour $\delta^n = O(H_n^{K+1})$. Let $\bar{\mathbf{P}}^n, \bar{\mathbf{Q}}^n$ be the Nordsieck vectors which correspond to the predictor and corrector polynomials of \mathbf{q} . Then th error δ^n can be estimated by $\hat{\delta}^n$:

$$\hat{\delta}^n = \frac{-H_n}{T_n - T_{n-K-1}} \left[\bar{\mathbf{Q}}_1^n - \bar{\mathbf{P}}_1^n \right]. \tag{3}$$

Now $\hat{r}_C^n = \|\mathbf{B}_L \hat{\delta}^n\| + \tau \|\mathbf{B}_A \hat{\delta}^n\|$ is the used weighted error norm, which must satisfy $\hat{r}_C^n < \text{TOL}_C$.

At the refined time-grid the DAE has been perturbed by the interpolated latent variables. The local discretization error $\delta^{n,m}$ is defined as the residue after inserting the exact solution in the BDF scheme of the refinement phase. However during the refinement instead of $\delta^{n,m}$ the perturbed local error $\tilde{\delta}^{n,m}$ is estimated. During the refinement each step $\mathbf{x}_{n}^{-1,m}$ is computed from the following scheme:

$$\alpha_{n-1,m}\mathbf{q}_{A}(t_{n-1,m},\mathbf{x}_{A}^{n-1,m},\hat{\mathbf{x}}_{L}^{n-1,m}) + h_{n-1,m}\mathbf{j}_{A}(t_{n-1,m},\mathbf{x}_{A}^{n-1,m},\hat{\mathbf{x}}_{L}^{n-1,m}) + \tilde{\beta}_{n-1,m} = \mathbf{0}.$$
(4)

Here $\tilde{\beta}_{n-1,m}$ is a constant which depends on the previous values of \mathbf{x}_A and $\hat{\mathbf{x}}_L$. A tedious analysis yields the following asymptotic behaviour:

$$\mathbf{B}_{A}\delta^{n-1,m} \doteq \mathbf{B}_{A}\tilde{\delta}^{n-1,m} + \frac{1}{4}h\mathbf{K}_{n-1,m}\mathbf{B}_{L}\rho^{n-1,m}.$$
 (5)

Here $\rho^{n-1,m}$ is the interpolation error at the refined grid and $\mathbf{K}_{n-1,m}$ is the coupling matrix. The perturbed local discretization error $\mathbf{B}_A \tilde{\delta}^{n,m}$ behaves as $O(h_{n-1,m}^{k+1})$ and can be estimated in a similar way as δ^n . Thus the active error estimate $\mathbf{B}_A \hat{\delta}^{n-1,m}$ satisfies $\mathbf{B}_A \hat{\delta}^{n-1,m} \doteq \mathbf{B}_A \hat{\delta}^{n-1,m} + \frac{1}{4} h \hat{\mathbf{K}}_{n-1,m} \mathbf{B}_L \hat{\rho}^{n-1,m}$. Let L be the interpolation order, then it can be shown that $\frac{1}{4} \|\hat{\mathbf{K}}_n \mathbf{B}_L \rho^{n-1,m}\|$ is less than

$$\hat{r}_{I}^{n} = \frac{1}{4} \frac{H_{n}}{T_{n} - T_{n-L-1}} \|\hat{\mathbf{K}}_{n} \mathbf{B}_{L} \left[\bar{\mathbf{X}}_{1}^{n} - \bar{\mathbf{Y}}_{1}^{n} \right] \|.$$
 (6)

Here $\bar{\mathbf{Y}}^n, \bar{\mathbf{X}}^n$ are the Nordsieck vectors which correspond to the predictor and corrector polynomials of \mathbf{x} . This error estimate \hat{r}_I^n has the asymptotic behaviour $\hat{r}_I^n = O(H_n^{L+1})$. It follows that $\|\mathbf{B}_A\hat{\delta}^{n,m}\|$ satisfies:

$$\|\mathbf{B}_{A}\hat{\delta}^{n-1,m}\| \le \hat{r}_{A}^{n-1,m} + h\hat{r}_{I}^{n} =: \hat{r}_{A}^{n-1,m}.$$
 (7)

If $\hat{r}_I^n \leq \text{TOL}_I = \sigma \text{TOL}_A$ and $\hat{\tilde{r}}_A^{n-1,m} \leq \text{T\tilde{O}L}_A = (1 - \sigma h) \text{TOL}_A$ then $\hat{r}_A^{n-1,m} \leq \text{T\tilde{O}L}_A + h \text{TOL}_I = \text{TOL}_A$.

We tested a circuit with 5×10 inverters. The location of the active part is controlled by the connecting elements and the voltage sources. The connecting elements were chosen such that the active part consists of 3 inverters. We did an Euler Backward Compound-Fast multirate simulation on $[0,10^{-8}]$ with $\sigma=0.5, \tau=0$. We get accurate results combined with a speedup factor 13.