

Fourier Series - Joseph Fourier (1768-1830)

A periodic signal $x : \text{Reals} \rightarrow \text{Reals}$ with period $p \in \text{Reals}$ can usually be described as a constant term plus a sum of sinusoids:



$$x(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(k\omega_0 t + \phi_k)$$

- x - Fourier series
- A_k, ϕ_k - amplitude and phase of cosine function (depend on x)
- ω_0 - fundamental frequency ($\omega_0 = 2\pi/p$)
- A_0 - DC term
- cosine terms for $k \geq 2$ - harmonics

Uniqueness of Fourier series

They are unique... what does that mean?

$$x(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(k\omega_0 t + \phi_k)$$

$$x(t) = B_0 + \sum_{k=1}^{\infty} B_k \cos(k\omega_0 t + \theta_k)$$

Then $A_0 = B_0$, $A_k = B_k$, $\phi_k = \theta_k \pmod{2\pi}$.

Discrete-time signals

NOT ALL FXNS
ARE PERIODIC

$$x(n+p) = x(n),$$

EASIER TO WORK
W/ CONTINUOUS

but n is now an integer. Subtlety: not all sinusoidal discrete-time signals are periodic. Consider:

$$x(n) = \cos(2\pi fn).$$

Need

$$x(n+p) = \cos(2\pi fn + 2\pi fp) = \cos(2\pi fn) = x(n).$$

In other words, $2\pi fp$ is an integer multiple of 2π , i.e. there exist nonzero integers p and m such that $f = m/p$.

Linear time-invariant (LTI) systems

- Key property: given a sinusoidal input, output is a sinusoidal signal with the same frequency but possibly a different amplitude and a different phase
- filtering (adjusting signal) is more naturally a frequency-based operation
- for example, removing abrupt changes corresponds to removing high frequency components
- will look at “frequency response” for systems

GET RID OF SHARP SOUNDS by
REDUCING HIGH FREQS

Time Invariance - Delay Systems

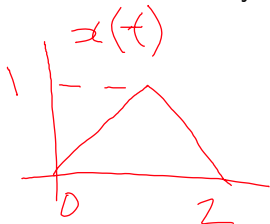
differential
equations,

$$D_2(x) = x(t - 2)$$

"-" implies
in past, "+"

We start with continuous-time systems. Consider a *delay system* (D_τ). This system takes an input x and produces an output y which is x delayed by τ time units.

$D_1(x+1)$ delay



$$y(t) = x(t - \tau)$$



Time invariant systems - definition

A time-invariant system S satisfies

$$\forall t \in \text{Reals}, \quad S \circ D_\tau = D_\tau \circ S.$$

SoD means add system then add on delay, whereas DoS is delay then add system

- $S \circ D_\tau$ - delay input by τ , then apply S
- $D_\tau \circ S$ - apply S to the input, then delay by τ

Interchanging the two systems must yield the same output, i.e. a delayed input $D_\tau(x)$ produces a delayed output $D_\tau(y)$.

Example 1

constant



$$y(t) = Cx(t)$$

Let $\hat{x}(t) = D_\tau(x(t))$. Then

delay on
system

$$\begin{aligned}\hat{y}(t) &= C\hat{x}(t) \\ &= Cx(t - \tau) \\ &= y(t - \tau)\end{aligned}$$



Example 2



$$y(t) = Cx^3(t)$$

$$\begin{aligned}\hat{y}(t) &= C(\hat{x}(t))^3 \\ &= C(x(t - \tau))^3 \\ &= y(t - \tau)\end{aligned}$$

Example 3

reverses signal in time.

$$y(t) = x(-t)$$

$$\begin{aligned}\hat{y}(t) &= \hat{x}(-t) \\ &= x(-t - \tau)\end{aligned}$$

But $y(t - \tau) = x(-t + \tau)$. **not time invariant**

Exercise 1

Is the following time-invariant?

$$y(t) = 10x(t) + 5x(t - 2)$$

$$\hat{x}(t) = x(t - T)$$
$$\hat{y}(t) = 10x(t - T) + 5x(t - 2 - T)$$

*time invariant

Exercise 2

time invariant :

Is the following time-invariant?

$$y(t) = tx(t)$$

if coefficient depends on
the time, it is not time
invariant

Discrete-time time-invariance

TI = TIME INVARIANT

$$D_M(x(n)) = x(n - M) \quad \text{Int, \# of samples}$$

$$S(D_M(x)) = D_M(S(x))$$

Examples:

$$y(n) = x(n) + 0.1(x(n))^2$$

$$y(n) = \cos(2\pi n)x(n)$$

$$y(n) = \cos(2\pi n/9)x(n)$$

Int
doesn't affect n

doesn't depend on 'n'

$x(n) \leftarrow TI$

Int NOT TI

Review - Complex Numbers

- Define i to be the square root of -1 , in other words, $(i)(i) = -1$.
- A *complex number* is the sum of a real number and an imaginary number, written $z = x + iy$, where the *real part* of z ($\text{Re}\{z\}$) is x and the *imaginary part* of z ($\text{Im}\{z\}$) is y .

Arithmetic

- Addition/subtraction. Let $z_1 = x_1 + iy_1$, $z_2 = x_2 + iy_2$.

$$z_1 + z_2 = (x_1 + x_2) + i(y_1 + y_2)$$

$$z_1 - z_2 = (x_1 - x_2) + i(y_1 - y_2)$$

- Multiplication.

$$z_1 z_2 = (x_1 + iy_1)(x_2 + iy_2) = (x_1 x_2 - y_1 y_2) + i(x_1 y_2 + x_2 y_1)$$

Complex conjugate

$$z = x + iy$$

$$z^* = x - iy$$

$$z + z^* = 2\operatorname{Re}\{z\}$$

$$z - z^* = 2i\operatorname{Im}\{z\}$$

$$zz^* = x^2 + y^2$$

$$|z| = \sqrt{zz^*}$$

Complex exponentials

The Taylor series expansion for an exponential function is

$$e^z = \sum_{k=0}^{\infty} \frac{z^k}{k!} = 1 + z + \frac{z^2}{2!} + \frac{z^3}{3!} + \dots$$

Let z be an imaginary number, $z = i\theta$.

$$\begin{aligned} e^{i\theta} &= 1 + i\theta + \frac{(i\theta)^2}{2!} + \frac{(i\theta)^3}{3!} + \dots \\ &= \left[1 - \frac{\theta^2}{2} + \frac{\theta^4}{4!} - \dots \right] + i \left[\theta - \frac{\theta^3}{3!} + \frac{\theta^5}{5!} - \dots \right] \\ &= \cos(\theta) + i \sin(\theta) \end{aligned}$$

Polar representation

$$z = x + iy = re^{i\theta}$$

where

$$r = |z| = \sqrt{x^2 + y^2}$$

and

$$\theta = \arg(z) = \tan^{-1}(y/x).$$

Complex systems

Let $x(t)$ have range equal to the complex numbers, and a be a complex constant. Define

$$\begin{aligned}(ax)(t) &= a(x(t)) \\ (x_1 + x_2)(t) &= x_1(t) + x_2(t)\end{aligned}$$

Let S map complex-valued functions to complex-valued functions. These are called *complex systems*.

Linearity

- Homogeneity: $S(ax) = aS(x)$
- Additivity: $S(x_1 + x_2) = S(x_1) + S(x_2)$
- Linearity: Homogeneity and Additivity.

Example 1.

$$y(t) = 10x(t) + 5x(t - 2).$$