The Discrete Fourier Transform*

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1 Motivation

We want to numerically approximate coefficients in a Fourier series. The first step is to see how the trapezoidal rule applies when numerically computing the integral $(2\pi)^{-1} \int_0^{2\pi} F(t) dt$, where F(t) is a continuous, 2π -periodic function. Applying the trapezoidal rule with the stepsize taken to be $h = 2\pi/n$ for some integer $n \ge 1$ results in

$$(2\pi)^{-1} \int_0^{2\pi} F(t)dt \approx \frac{1}{n} \sum_{j=0}^{n-1} Y_j,$$

where $Y_j := F(hj) = F(2\pi j/n)$, $j = 1 \dots n-1$. We remark that we made use of $Y_n = F(2\pi) = F(0) = Y_0$ in employing the trapezoidal rule to arrive at the right hand side of the equation above. Recall that the coefficients in a Fourier series expansion for a continuous, 2π -periodic function f(t) have the form

$$c_k = \frac{1}{2\pi} \int_0^{2\pi} f(t) \exp(-ikt) dt.$$

We can apply the version of the trapezoidal rule derived above to approximately calculate the c_k 's, since $f(t) \exp(-ikt)$ is 2π -periodic. Doing so yields

$$c_k \approx \frac{1}{n} \sum_{j=0}^{n-1} f(2\pi j/n) \exp(-2\pi i j k/n) = \frac{1}{n} \sum_{j=0}^{n-1} y_j \overline{w}^{jk},$$

where $y_j = f(2\pi j/n)$ and $w = \exp(2\pi i/n)$. If we replace k by k + n, the right hand side of the last equation is unchanged, for $\overline{w}^n = \exp(-2\pi i) = 1$.

^{*}These notes are based on [1, Chapter 3].

Consequently, only the approximations to c_k for $k = 0 \dots n-1$ need be calculated. Given these approximations, however, one may recover y_j , $j = 0 \dots n-1$. To see this, let

$$\hat{y}_k = \sum_{j=0}^{n-1} y_j \overline{w}^{jk},$$

so that $c_k \approx \hat{y}_k/n$. Multiply both sides by $w^{k\ell}$ and sum over k:

$$\sum_{k=0}^{n-1} \hat{y}_k w^{k\ell} = \sum_{j=0}^{n-1} y_j \sum_{k=0}^{n-1} w^{(\ell-j)k}.$$

The sum over k on the right can be evaluated via the algebraic identity

$$\sum_{k=0}^{n-1} z^k = \begin{cases} \frac{z^n - 1}{z - 1} & \text{if } z \neq 1\\ n & \text{if } z = 1. \end{cases}$$

Recalling that $w^n = 1$, setting $z = w^{j-\ell}$ above, and noting that $w^{j-\ell} \neq 1$ unless $j = \ell$, one gets

$$\sum_{k=0}^{n-1} w^{(\ell-j)k} = \begin{cases} 0 & \text{if } j \neq \ell \\ n & \text{if } j = \ell. \end{cases}$$

Consequently, we find that

$$\frac{1}{n} \sum_{k=0}^{n-1} \hat{y}_k w^{k\ell} = y_{\ell} \,.$$

Thus the y's can be calculated if we know the c's or \hat{y} 's.

2 Definition

Let S_n be the set of periodic sequences of complex numbers with period n. The set S_n forms a complex vector space under the operations of entry-by-entry addition and entry-by-entry multiplication by a scalar. Let $y = \{y_j\}_{j=-\infty}^{\infty} \in S_n$, so that $y_{j+n} = y_j$ for all j. We can associate to each y a new sequence \hat{y} via

$$\hat{y}_k = \sum_{j=0}^{n-1} y_j \overline{w}^{jk}.$$

This is the same formula that we used to find \hat{y}_k in §1; the only differences are that the y_j 's are do not necessarily come from a continuous function, and that the index k above is not resricted to $\{0, \ldots, n-1\}$. The sequence \hat{y} is periodic with period n. To see this, note that

$$\hat{y}_{k+n} = \sum_{j=0}^{n-1} y_j \overline{w}^{j(k+n)} = \sum_{j=0}^{n-1} y_j \overline{w}^{jk} \overline{w}^{nj}$$

$$= \sum_{j=0}^{n-1} y_j \overline{w}^{jk} \quad [\overline{w}^n = e^{-(2\pi i/n)n} = 1]$$

$$= \hat{y}_k$$

Put another way, $\hat{y} \in \mathcal{S}_n$. The mapping $y \in \mathcal{S}_n \mapsto \hat{y} \in \mathcal{S}_n$ defines the discrete Fourier transform. We will write $\hat{y} = \mathcal{F}[y]$. In addition, the formula derived in §1 giving y_j 's in terms of \hat{y}_k 's certainly applies here as well. Thus, after changing the "dummy" indices, one gets this formula for y_j 's in terms of \hat{y}_k 's:

$$y_j = \frac{1}{n} \sum_{k=0}^{n-1} \hat{y}_k w^{jk} .$$

This is the inversion formula for the DFT. We denote the inverse correspondence $\hat{y} \in \mathcal{S}_n \mapsto y \in \mathcal{S}_n$ by $y = \mathcal{F}^{-1}[\hat{y}]$.

Both \mathcal{F} and \mathcal{F}^{-1} are linear transformations from \mathcal{S}_n to itself. Here are some additional properties that you can verify as exercises.

- 1. Shifts. If z is the periodic sequence formed from $y \in \mathcal{S}_n$ via $z_j = y_{j+1}$, then $\mathcal{F}[z]_k = w^k \mathcal{F}[y]_k$.
- 2. Convolutions. If $y \in \mathcal{S}_n$ and $z \in \mathcal{S}_n$, then the sequence defined by $[y*z]_j := \sum_{m=0}^{n-1} y_m z_{j-m}$ is also in \mathcal{S}_n . The sequence y*z is called the convolution of y and z.
- 3. The Convolution Theorem: $\mathcal{F}[y*z]_k = \mathcal{F}[y]_k \mathcal{F}[z]_k$.

3 An application

Consider the differential equation

$$u'' + au' + bu = f(t),$$

where f is a continuous, 2π -periodic function of t. There is a well-known analytical method for finding the unique periodic solution to this equation

(cf. Boyce & DiPrima, fifth edition, §3.7.2—forced vibrations), provided f is known for all t. On the other hand, if we only know f at the points $t_j = jh$, where again $h = 2\pi/n$ for some integer $n \ge 1$, this method is no longer applicable.

Instead of directly trying to work with the differential equation itself, we will work with a discretized version of it. There a many ways of discretizing; the one that we will use here amounts to making these replacements:

$$u'(t) \longrightarrow \frac{u(t) - u(t-h)}{h},$$

$$u''(t) \longrightarrow \frac{u(t+h) + u(t-h) - 2u(t)}{h^2}.$$

Replacing u' and u'' in the differential equation and setting $t = 2\pi j/n$, we get the following difference equation for the sequence $u_j = u(2\pi j/n)$:

$$u_{j+1} + \alpha u_j + \beta u_{j-1} = h^2 f_j$$
,

where $f_j = f(2\pi j/n)$, $\alpha = bh^2 + ah - 2$, and $\beta = 1 - ah$.

Let $u \in \mathcal{S}_n$ be a solution to the difference equation derived above, and let $\hat{u} = \mathcal{F}[u]$. In addition, let $\hat{f} = \mathcal{F}[f]$. From the inversion formula for the DFT, we have

$$u_j = \frac{1}{n} \sum_{k=0}^{n-1} \hat{u}_k w^{jk}$$
 and $f_j = \frac{1}{n} \sum_{k=0}^{n-1} \hat{f}_k w^{jk}$.

Inserting these in the difference equation then yields, after multiplying by n,

$$\sum_{k=0}^{n-1} \hat{u}_k w^{k(j+1)} + \alpha \sum_{k=0}^{n-1} \hat{u}_k w^{jk} + \beta \sum_{k=0}^{n-1} \hat{u}_k w^{k(j-1)} = \sum_{k=0}^{n-1} h^2 \hat{f}_k w^{jk}.$$

Combining terms and doing an algebraic manipulation then results in this:

$$\sum_{k=0}^{n-1} (w^k + \alpha + \beta \overline{w}^k) \hat{u}_k w^{jk} = \sum_{k=0}^{n-1} h^2 \hat{f}_k w^{jk}.$$

Taking the inverse DFT of both sides and dividing by $w^j + \alpha + \beta \overline{w}^j$, which we assume is never 0, we find that

$$\hat{u}_k = h^2 (w^k + \alpha + \beta \overline{w}^k)^{-1} \hat{f}_k.$$

Thus we have found the DFT of u. Inverting this then recovers u itself. In the next section we will discuss methods for fast computation of the DFT and its inverse.

4 The Fast Fourier Transform

Let us consider the DFT of a periodic sequence y with period n=2N. The \hat{y}_k 's are calculated via

$$\hat{y}_k = \sum_{j=0}^{2N-1} y_j \overline{w}^{jk}.$$

Splitting the sum above into a sum over even and odd integers yields

$$\hat{y}_k = \sum_{j=0}^{N-1} y_{2j} \overline{w}^{2jk} + \sum_{j=0}^{N-1} y_{2j+1} \overline{w}^{(2j+1)k}$$

$$= \sum_{j=0}^{N-1} y_{2j} \overline{W}^{jk} + \overline{w}^k \left(\sum_{j=0}^{N-1} y_{2j+1} \overline{W}^{jk} \right),$$

where $W:=\exp(2\pi i/N)=w^2$. We may rewrite this in terms of DFT's with $n\to N$:

$$\hat{y}_k = \mathcal{F}[\{y_0, y_2, \cdots, y_{2N-2}\}]_k + \overline{w}^k \mathcal{F}[\{y_1, y_3, \cdots, y_{2N-1}\}]_k$$

A further savings is possible. In the last equation, let $k \to k + N$ and use these facts: (1) $\mathcal{F}[y^{\text{even}}]$ and $\mathcal{F}[y^{\text{odd}}]$ both have period N. (2) $\overline{w}^{k+N} = \overline{w}^k \exp(-\pi i) = -\overline{w}^k$. The result is that for $0 \le k \le N - 1$ we have

$$\begin{cases} \hat{y}_k = \mathcal{F}[\{y_0, y_2, \cdots, y_{2N-2}\}]_k + \overline{w}^k \mathcal{F}[\{y_1, y_3, \cdots, y_{2N-1}\}]_k \\ \hat{y}_{k+N} = \mathcal{F}[\{y_0, y_2, \cdots, y_{2N-2}\}]_k - \overline{w}^k \mathcal{F}[\{y_1, y_3, \cdots, y_{2N-1}\}]_k. \end{cases}$$

Similar formulas can be derived for the inverse DFT; they are:

$$\begin{cases} y_k = \frac{1}{2} \left\{ \mathcal{F}^{-1} [\{\hat{y}_0, \hat{y}_2, \cdots, \hat{y}_{2N-2}\}]_k + w^k \mathcal{F}^{-1} [\{\hat{y}_1, \hat{y}_3, \cdots, \hat{y}_{2N-1}\}]_k \right\} \\ y_{k+N} = \frac{1}{2} \left\{ \mathcal{F}^{-1} [\{\hat{y}_0, \hat{y}_2, \cdots, \hat{y}_{2N-2}\}]_k - w^k \mathcal{F}^{-1} [\{\hat{y}_1, \hat{y}_3, \cdots, \hat{y}_{2N-1}\}]_k \right\}. \end{cases}$$

(The factor of $\frac{1}{2}$ appears because the inversion formula has a "1/n" in it.)

What is the computational "cost" of using the formulas above versus ordinary matrix methods, where there are $4n^2$ real multiplications used in the computation? Set $n=2^L$ and let K_L be the number of real multiplications required to compute $\mathcal{F}[y]$ by the method above. From the formulas derived above, one sees that to compute $\mathcal{F}[y]$, one needs to compute $\mathcal{F}[y^{\text{even}}]$ and $\mathcal{F}[y^{\text{odd}}]$. This takes $2K_{L-1}$ real multiplications. In addition, one must multiply \overline{w}^k and $\mathcal{F}[y^{\text{odd}}]_k$, for $k=0,\ldots,2^{L-1}-1$, which requires $4\times 2^{L-1}$ real multiplications. The result is that K_L is related to K_{L-1} via

$$K_L = 2K_{L-1} + 2^{L+1}$$

When L=0, $n=2^0=1$ and no multiplications are required; thus, $K_0=0$. Inserting L=1 in the last equation, we find that $K_1=1\times 2^2$. Similarly, setting L=2 then yields $K_2=2\times 2^3$. Similarly, one finds that $K_3=3\times 2^4$, $K_4=4\times 2^5$, and so on. The general formula is $K_L=L\times 2^{L+1}=2L\times 2^L$. Again setting $n=2^L$ and noting that $L=\log_2 n$, we see that the number of real multiplications required is $2n\log_2 n$.

To get an idea of how much faster than matrix multiplication this method is, suppose that we want to take the DFT of data with $n=2^{12}=4,096$ points. The conventional method requires $2^{26}\approx 7\times 10^7$ real multiplications. Using the FFT method to get the DFT requires $2\times 2^{12}\times 12\approx 10^5$ real multiplications, making the FFT roughly 700 times as fast.

We remark that similar algorithms can be obtained for $n = N_1 N_2 \cdots N_m$, although the fastest one is obtained in the case discussed above. For a discussion of this and related topics, one should consult the references below.

Previous: Pointwise convergence of Fourier series

Next: Splines and finite element spaces

References

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