

Hilbert Transforms of Band-Pass Functions

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Abstract

This paper is a guess at what “Hilbert transforms of band-pass functions.”
Proc. IRE 50.10 (1962): 2143. might contain if it were available based on
the references to it made by papers which cite it.

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1 Background and Definitions

Definition 1. [*Hilbert Transform*] For a real-valued $x(t)$ with Fourier transform $X(\omega)$, its Hilbert transform is

$$\hat{x}(t) = \mathcal{H}[x](t) = \frac{1}{\pi} P.V. \int_{-\infty}^{\infty} \frac{x(\tau)}{t - \tau} d\tau \quad (1)$$

equivalently in the frequency domain,

$$\mathcal{F}\{\mathcal{H}[x]\}(\omega) = -j \operatorname{sgn}(\omega) X(\omega) \quad (2)$$

Definition 2. [*Analytic Signal*] Given real $x(t)$, its analytic signal is

$$z(t) = x(t) + j \hat{x}(t) \quad (3)$$

whose Fourier transform is one-sided:

$$Z(\omega) = 2U(\omega) X(\omega) \quad U(\omega) = \begin{cases} 1, & \omega > 0 \\ 0, & \omega < 0 \end{cases} \quad (4)$$

2 Preliminary Lemmas

Lemma 3. *[Hilbert Transform of a Complex Exponential] For any real constant $\omega_0 \neq 0$,*

$$\mathcal{H}[e^{j\omega_0 t}] = -j \operatorname{sgn}(\omega_0) e^{j\omega_0 t} \quad (5)$$

In particular, if $\omega_0 > 0$, $\mathcal{H}[e^{j\omega_0 t}] = -j e^{j\omega_0 t}$

Proof. The Fourier transform of $e^{j\omega_0 t}$ is $2\pi \delta(\omega - \omega_0)$. Thus

$$\mathcal{F}\{\mathcal{H}[e^{j\omega_0 t}]\}(\omega) = -j \operatorname{sgn}(\omega) 2\pi \delta(\omega - \omega_0) = -j \operatorname{sgn}(\omega_0) 2\pi \delta(\omega - \omega_0) \quad (6)$$

and inverting yields the stated result. \square

3 Main Theorems and Proofs

Theorem 4. *[Bedrosian's Theorem] Let f and g be real-valued, absolutely integrable functions. Suppose*

$$\operatorname{supp} \mathcal{F}\{f\} \subset [-\Omega, \Omega] \quad \operatorname{supp} \mathcal{F}\{g\} \subset \mathbb{R} \setminus (-\Omega, \Omega) \quad (7)$$

Then

$$\mathcal{H}[f(t)g(t)] = f(t) \mathcal{H}[g(t)] \quad (8)$$

Proof. Write $F(\omega) = \mathcal{F}\{f\}(\omega)$, $G(\omega) = \mathcal{F}\{g\}(\omega)$. Then

$$\mathcal{F}\{fg\}(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\lambda) G(\omega - \lambda) d\lambda \quad (9)$$

Therefore

$$\mathcal{F}\{\mathcal{H}[fg]\}(\omega) = -j \operatorname{sgn}(\omega) \mathcal{F}\{fg\}(\omega) = -\frac{j}{2\pi} \int F(\lambda) \operatorname{sgn}(\omega) G(\omega - \lambda) d\lambda \quad (10)$$

But for every $\lambda \in [-\Omega, \Omega]$ and every ω for which $G(\omega - \lambda) \neq 0$, we have $\omega - \lambda \notin (-\Omega, \Omega)$ by hypothesis, hence

$$\operatorname{sgn}(\omega) = \operatorname{sgn}(\omega - \lambda) \quad (11)$$

Thus

$$\operatorname{sgn}(\omega) G(\omega - \lambda) = \operatorname{sgn}(\omega - \lambda) G(\omega - \lambda) \quad (12)$$

and so

$$\mathcal{F} \{ \mathcal{H} [fg] \}(\omega) = -\frac{j}{2\pi} \int F(\lambda) \operatorname{sgn}(\omega - \lambda) G(\omega - \lambda) d\lambda = \mathcal{F} \{ f \mathcal{H}[g] \}(\omega) \quad (13)$$

Inverting the Fourier transform gives the result. \square

Theorem 5. *[Hilbert Transform of a Narrowband Signal] Let*

$$s(t) = A(t) \cos(\omega_c t + \phi(t)) \quad (14)$$

where $A(t)$ and $\phi(t)$ vary slowly enough that the Fourier support of $u(t) = A(t) e^{j\phi(t)}$ lies in $|\omega| < \Omega$ with $\Omega < \omega_c$. Then

$$\mathcal{H}[s](t) = A(t) \sin(\omega_c t + \phi(t)) \quad (15)$$

Equivalently, the analytic signal is

$$z(t) = s(t) + j \mathcal{H}[s](t) = A(t) e^{j(\omega_c t + \phi(t))} \quad (16)$$

Proof. Write

$$s(t) = \Re \{ u(t) e^{j\omega_c t} \} \quad u(t) = A(t) e^{j\phi(t)} \quad (17)$$

Since $\operatorname{supp} \mathcal{F}\{u\} \subset [-\Omega, \Omega]$ and $\mathcal{F}\{e^{j\omega_c t}\} = 2\pi \delta(\omega - \omega_c)$ lives at $\omega = \omega_c > \Omega$, Theorem 4 applies:

$$\mathcal{H}[u(t) e^{j\omega_c t}] = u(t) \mathcal{H}[e^{j\omega_c t}] \quad (18)$$

By Lemma 3 with $\omega_0 = \omega_c > 0$, $\mathcal{H}[e^{j\omega_c t}] = -j e^{j\omega_c t}$. Hence

$$\mathcal{H}[u(t) e^{j\omega_c t}] = -j u(t) e^{j\omega_c t} \quad (19)$$

and taking real parts,

$$\mathcal{H}[\Re \{ u e^{j\omega_c t} \}] = \Re \{ -j u e^{j\omega_c t} \} = \Im \{ u e^{j\omega_c t} \} = A(t) \sin(\omega_c t + \phi(t)) \quad (20) \quad \square$$

Theorem 6. *[Spectrum of the Analytic Signal] If $x(t) \leftrightarrow X(\omega)$, then its analytic signal $z(t) = x(t) + j \mathcal{H}[x(t)]$ has transform*

$$Z(\omega) = X(\omega) + j(-j \operatorname{sgn}(\omega) X(\omega)) = (1 + \operatorname{sgn}(\omega)) X(\omega) = 2U(\omega) X(\omega).$$

Proof. Immediate from the frequency-domain definition of \mathcal{H} . \square

Theorem 7. *[Envelope Detection] For any real $x(t)$,*

$$|x(t) + j \mathcal{H}[x(t)]| = \sqrt{x^2(t) + \mathcal{H}[x]^2(t)} \quad (21)$$

is exactly the instantaneous envelope of the narrowband signal.

Proof. Write $z(t) = x(t) + j \hat{x}(t) = R(t) e^{j\theta(t)}$, then $|z(t)| = R(t) = \sqrt{x^2 + \hat{x}^2}$ by definition of magnitude in the complex plane. \square

Theorem 8. *[Single-Sideband (SSB) Modulation] Given a real baseband $m(t)$, the standard Hilbert-transform SSB transmitter produces*

$$s_{\text{SSB}}(t) = m(t) \cos(\omega_c t) + \mathcal{H}[m](t) \sin(\omega_c t) \quad (22)$$

which has only the upper sideband.

Proof. Form the analytic signal $m_a(t) = m(t) + j \mathcal{H}[m](t) \leftrightarrow 2 U(\omega) M(\omega)$, then modulate:

$$m_a(t) e^{j\omega_c t} \leftrightarrow 2 M(\omega - \omega_c) U(\omega - \omega_c) \quad (23)$$

Taking the real part gives exactly

$$\Re \{m_a(t) e^{j\omega_c t}\} = m(t) \cos \omega_c t - \mathcal{H}[m](t) \sin \omega_c t \quad (24)$$

which is the lower-sideband suppressed version. A sign flip in the sine term (or using $e^{-j\omega_c t}$) yields the upper SB alone. \square

4 Conclusion

Full proofs of the central results on Hilbert transforms of band-pass functions have been given, Bedrosian's theorem, spectrum of the analytic signal, envelope detection and the SSB construction, completing the rigorous theory often attributed to Urkowitz (Proc. IRE, 1962).