

1 The Discrete Fourier Transform

The Fourier transform establishes a bijective mapping between temporal and frequency domain representations of signals. For discrete-time signals of finite duration, the Discrete Fourier Transform (DFT) provides this mapping:

Definition 1. (Discrete Fourier Transform (DFT)) *Given a discrete-time signal $x(k)$ where $k = 0, 1, \dots, N - 1$, the DFT yields the same number of complex coefficients $X(k)$:*

$$X(k) = \sum_{n=0}^{N-1} x(n) \exp\left(-i \frac{2\pi k n}{N}\right) \quad (1)$$

The inverse DFT reconstructs the original signal:

$$x(k) = \frac{1}{N} \sum_{n=0}^{N-1} X(n) \exp\left(i \frac{2\pi k n}{N}\right) \quad (2)$$

The DFT presupposes periodic extension of the finite-length sequence $x(n)$, a fact that underlies many spectral estimation artifacts.

1.1 The Fast Fourier Transform

The Fast Fourier Transform denotes a family of algorithms that compute the DFT with $O(N \log N)$ complexity rather than the $O(N^2)$ complexity of direct matrix multiplication. Crucially, the FFT produces numerically identical results to the DFT; scaling relationships derived for the DFT apply without modification to FFT implementations.

1.2 Magnitude and Phase

The complex output $X(k)$ encodes both magnitude and phase information:

$$X(k) = |X(k)| \exp(i \arg(X(k))) \quad (3)$$

where $|X(k)|$ represents the magnitude and $\arg(X(k))$ the phase of the k -th frequency component.

Definition 2. (Argument (Phase)) For a non-zero complex number $z = a + i b$ where $a, b \in \mathbb{R}$, the principal argument (phase) with range $(-\pi, \pi]$ is:

$$\arg(z) = \begin{cases} 2 \arctan\left(\frac{b}{\sqrt{a^2 + b^2} + a}\right), & \text{if } b \neq 0 \text{ or } a > 0, \\ \pi, & \text{if } b = 0 \text{ and } a < 0. \end{cases} \quad (4)$$

where $\arg(0)$ is undefined. This follows from the half-angle identity

$$\lambda = 2 \arctan\left(\frac{\sin(\lambda)}{1 + \cos(\lambda)}\right), \quad \lambda \in (-\pi, \pi) \quad (5)$$

where

$$a(r, \lambda) = r \cos(\lambda) \quad (6)$$

and

$$b(r, \lambda) = r \sin(\lambda) \quad (7)$$

with

$$r = \sqrt{a^2 + b^2} \quad (8)$$

The conjugate relationship

$$\arg(\bar{z}) = \begin{cases} -\arg(z), & \text{if } z \notin (-\infty, 0], \\ \pi, & \text{if } z \in (-\infty, 0) \end{cases} \quad (9)$$

and the identity

$$b + i a = i \bar{z} \quad (10)$$

yield:

$$\arg(i \bar{z}) = \begin{cases} 2 \arctan\left(\frac{a}{\sqrt{a^2 + b^2} + b}\right), & \text{if } a \neq 0 \text{ or } b > 0, \\ \pi, & \text{if } a = 0 \text{ and } b < 0. \end{cases} \quad (11)$$

For real-valued input signals $x(n) \in \mathbb{R}$, the DFT exhibits conjugate symmetry:

$$X(k) = X^*(N - k) \quad (12)$$

Consequently, the magnitude spectrum is symmetric:

$$|X(k)| = |X(N - k)| \quad (13)$$

and the phase spectrum is antisymmetric modulo 2π :

$$\arg(X(k)) \equiv -\arg(X(N-k)) \pmod{2\pi} \quad (14)$$

with exact equality $\arg(X(k)) = -\arg(X(N-k))$ holding when $X(k)$ is not a negative real number. Thus in the real-valued (symmetric) case the coefficients contain redundant information.