

# 1 The Discrete Fourier Transform

The Fourier transform establishes a bijective mapping between temporal and frequency domain representations of signals. For discrete-time signals of finite duration, the Discrete Fourier Transform (DFT) provides this mapping:

**Definition 1. (Discrete Fourier Transform (DFT))** *Given a discrete-time signal  $x(k)$  where  $k = 0, 1, \dots, N - 1$ , the DFT yields the same number of complex coefficients  $X(k)$ :*

$$X(k) = \sum_{n=0}^{N-1} x(n) \exp\left(-i \frac{2\pi k n}{N}\right) \quad (1)$$

*The inverse DFT reconstructs the original signal:*

$$x(k) = \frac{1}{N} \sum_{n=0}^{N-1} X(n) \exp\left(i \frac{2\pi k n}{N}\right) \quad (2)$$

The DFT presupposes periodic extension of the finite-length sequence  $x(n)$ , a fact that underlies many spectral estimation artifacts.

## 1.1 The Fast Fourier Transform

The Fast Fourier Transform denotes a family of algorithms that compute the DFT with  $O(N \log N)$  complexity rather than the  $O(N^2)$  complexity of direct matrix multiplication. Crucially, the FFT produces numerically identical results to the DFT; scaling relationships derived for the DFT apply without modification to FFT implementations.

## 1.2 Magnitude and Phase

The complex output  $X(k)$  encodes both magnitude and phase information:

$$X(k) = |X(k)| \exp(i \arg(X(k))) \quad (3)$$

where  $|X(k)|$  represents the magnitude and  $\arg(X(k))$  the phase of the  $k$ -th frequency component.

**Definition 2. (Argument (Phase))** For a non-zero complex number  $z = a + ib$  where  $a, b \in \mathbb{R}$ , the principal argument (phase) with range  $(-\pi, \pi]$  is:

$$\arg(z) = \begin{cases} 2 \arctan\left(\frac{b}{\sqrt{a^2 + b^2} + a}\right), & \text{if } b \neq 0 \text{ or } a > 0, \\ \pi, & \text{if } b = 0 \text{ and } a < 0. \end{cases} \quad (4)$$

where  $\arg(0)$  is undefined. This follows from the half-angle identity

$$\lambda = 2 \arctan\left(\frac{\sin(\lambda)}{1 + \cos(\lambda)}\right), \quad \lambda \in (-\pi, \pi) \quad (5)$$

where

$$a(r, \lambda) = r \cos(\lambda) \quad (6)$$

and

$$b(r, \lambda) = r \sin(\lambda) \quad (7)$$

with

$$r = \sqrt{a^2 + b^2} \quad (8)$$

The conjugate relationship

$$\arg(\bar{z}) = \begin{cases} -\arg(z), & \text{if } z \notin (-\infty, 0], \\ \pi, & \text{if } z \in (-\infty, 0) \end{cases} \quad (9)$$

and the identity

$$b + i a = i \bar{z} \quad (10)$$

yield:

$$\arg(i \bar{z}) = \begin{cases} 2 \arctan\left(\frac{a}{\sqrt{a^2 + b^2} + b}\right), & \text{if } a \neq 0 \text{ or } b > 0, \\ \pi, & \text{if } a = 0 \text{ and } b < 0. \end{cases} \quad (11)$$

For real-valued input signals  $x(n) \in \mathbb{R}$ , the DFT exhibits conjugate symmetry:

$$X(k) = X^*(N - k) \quad (12)$$

Consequently, the magnitude spectrum is symmetric:

$$|X(k)| = |X(N - k)| \quad (13)$$

and the phase spectrum is antisymmetric modulo  $2\pi$ :

$$\arg(X(k)) \equiv -\arg(X(N-k)) \pmod{2\pi} \quad (14)$$

with exact equality  $\arg(X(k)) = -\arg(X(N-k))$  holding when  $X(k)$  is not a negative real number. Thus in the real-valued (symmetric) case the coefficients contain redundant information.