An oscilatory process (Priestley 1965) X_t can be represented as:

$$X_t = \int_{-\infty}^{\infty} e^{i\omega t} A_t(\omega) \ dZ(\omega) \tag{1}$$

where $A_t(\omega)$ is the time-varying gain function and $dZ(\omega)$ represents a process with orthogonal increments.

Theorem 1. [Fourier Domain Relationship] The relationship between the gain function $A_t(\omega)$ and the time-varying filter $h_t(u)$ is given by:

$$A_t(\omega) e^{i\omega t} = \int_{-\infty}^{\infty} h_t(u) e^{i\omega u} du$$
 (2)

Remark 2. Note that in some literature, $h_t(u)$ may be denoted as $a(t, \tau)$, where t is the time parameter and τ or u represents the lag parameter.

Theorem 3. [Explicit Definition of $h_t(u)$] The time-varying filter $h_t(u)$ is explicitly defined as:

$$h_t(u) = \frac{1}{2\pi} \int_{-\infty}^{\infty} A_t(\omega) e^{i\omega(t-u)} d\omega$$
 (3)

Proof. We start with the Fourier domain relationship:

$$A_t(\omega) e^{i\omega t} = \int_{-\infty}^{\infty} h_t(u) e^{i\omega u} du$$
 (4)

To isolate $h_t(u)$, we apply the inverse Fourier transform by multiplying both sides by $e^{-i\omega v}$ and integrating with respect to ω :

$$\int_{-\infty}^{\infty} A_t(\omega) e^{i\omega t} e^{-i\omega v} d\omega = \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} h_t(u) e^{i\omega u} du \right) e^{-i\omega v} d\omega$$
 (5)

$$\int_{-\infty}^{\infty} A_t(\omega) e^{i\omega(t-v)} d\omega = \int_{-\infty}^{\infty} h_t(u) \left(\int_{-\infty}^{\infty} e^{i\omega(u-v)} d\omega \right) du$$
 (6)

The inner integral on the right-hand side is:

$$\int_{-\infty}^{\infty} e^{i\omega(u-v)} d\omega = 2\pi \delta(u-v)$$
(7)

where $\delta(\cdot)$ is the Dirac delta function.

Therefore:

$$\int_{-\infty}^{\infty} A_t(\omega) e^{i\omega(t-v)} d\omega = \int_{-\infty}^{\infty} h_t(u) \cdot 2\pi \delta(u-v) du$$
(8)

$$=2\pi h_t(v) \tag{9}$$

Solving for $h_t(v)$ and replacing v with u:

$$h_t(u) = \frac{1}{2\pi} \int_{-\infty}^{\infty} A_t(\omega) e^{i\omega(t-u)} d\omega$$
 (10)

This gives the explicit definition of $h_t(u)$.

Theorem 4. [Alternative Process Representation] The process X_t can also be represented as:

$$X_t = \int_{-\infty}^{\infty} h_t(u) X_S(t - u) du$$
(11)

where $X_S(t)$ is a stationary process with power spectral density $S_{XX}(\omega)$.

Proof. Starting from the original spectral representation:

$$X_t = \int_{-\infty}^{\infty} e^{i\omega t} A_t(\omega) \ dZ(\omega)$$
 (12)

We substitute the Fourier relationship:

$$X_t = \int_{-\infty}^{\infty} \frac{1}{e^{i\omega t}} \left(\int_{-\infty}^{\infty} h_t(u) e^{i\omega u} du \right) e^{i\omega t} dZ(\omega)$$
 (13)

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h_t(u) e^{i\omega u} du dZ(\omega)$$
(14)

Interchanging the order of integration:

$$X_t = \int_{-\infty}^{\infty} h_t(u) \left(\int_{-\infty}^{\infty} e^{i\omega u} \ dZ(\omega) \right) du \tag{15}$$

$$= \int_{-\infty}^{\infty} h_t(u) \left(\int_{-\infty}^{\infty} e^{i\omega(t - [t - u])} dZ(\omega) \right) du$$
 (16)

Define the stationary process:

$$X_S(t-u) = \int_{-\infty}^{\infty} e^{i\omega(t-u)} dZ(\omega)$$
 (17)

Therefore:

$$X_t = \int_{-\infty}^{\infty} h_t(u) X_S(t - u) du$$
(18)

This demonstrates that X_t can be represented as the output of a time-varying filter $h_t(u)$ applied to a stationary process $X_s(t)$.

1 Conclusion

The explicit definition of the time-varying filter $h_t(u)$ in terms of the gain function $A_t(\omega)$ is:

$$h_t(u) = \frac{1}{2\pi} \int_{-\infty}^{\infty} A_t(\omega) e^{i\omega(t-u)} d\omega$$
 (19)

This relationship provides a critical link between the frequency-domain representation using the gain function and the time-domain representation using the time-varying filter. The presence of the factor $e^{i\omega(t-u)}$ in the integrand is essential and distinguishes this from a simple inverse Fourier transform.