

The Spectral Representation for Real-Valued Stationary Processes

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Theorem 1. (Real Spectral Representation for Stationary Processes) *Let $\{\xi(t), t \in \mathbb{R}\}$ be a real-valued, zero-mean, second-order stationary process with covariance function $r(t) = \mathbb{E}[\xi(t) \xi(0)]$ and spectral distribution function $F(\omega)$. Then there exist real-valued random measures $\{U(\omega), \omega \geq 0\}$ and $\{V(\omega), \omega \geq 0\}$ with orthogonal increments such that:*

1. Process Representation:

$$\xi(t) = \int_0^\infty [\cos(\omega t) dU(\omega) + \sin(\omega t) dV(\omega)] \quad (1)$$

2. Covariance Representation:

$$r(t) = \int_0^\infty \cos(\omega t) dF(\omega) \quad (2)$$

3. Orthogonality Properties:

$$\mathbb{E}[U(\omega)] = \mathbb{E}[V(\omega)] = 0 \quad (3)$$

$$\mathbb{E}[dU(\omega_1) dU(\omega_2)] = \mathbb{E}[dV(\omega_1) dV(\omega_2)] = \delta(\omega_1 - \omega_2) dF(\omega_1) \quad (4)$$

$$\mathbb{E}[dU(\omega_1) dV(\omega_2)] = 0 \quad \text{for all } \omega_1, \omega_2 \geq 0 \quad (5)$$

Proof. 1. **Construction from Complex Representation:** From the complex spectral representation theorem, there holds

$$\xi(t) = \int_{-\infty}^\infty e^{i\omega t} d\zeta(\omega) \quad (6)$$

where $\zeta(\omega)$ is a complex-valued random measure with orthogonal increments and $\mathbb{E}[d\zeta(\omega_1) d\overline{\zeta(\omega_2)}] = \delta(\omega_1 - \omega_2) dF_{two}(\omega)$ for the two-sided spectral distribution function $F_{two}(\omega)$.

2. **Reality Condition:** As $\xi(t)$ is real-valued,

$$\xi(t) = \overline{\xi(t)} = \int_{-\infty}^\infty e^{-i\omega t} d\overline{\zeta(\omega)} \quad (7)$$

3. **Symmetry Property:** This reality condition requires the spectral random measure to satisfy

$$d\zeta(-\omega) = d\overline{\zeta(\omega)} \quad (8)$$

for all ω .

4. **Factorization into Real Random Measures:** For $\omega > 0$, define

$$dU(\omega) = d\zeta(\omega) + d\zeta(-\omega) = 2 \Re[d\zeta(\omega)] \quad (9)$$

$$dV(\omega) = i[d\zeta(\omega) - d\zeta(-\omega)] = -2 \Im[d\zeta(\omega)] \quad (10)$$

where \Re and \Im denote the real and imaginary parts.

5. **Derivation of Real Spectral Representation:**

$$\begin{aligned} \xi(t) &= \int_0^\infty e^{i\omega t} d\zeta(\omega) + \int_0^\infty e^{-i\omega t} d\zeta(-\omega) \\ &= \int_0^\infty e^{i\omega t} d\zeta(\omega) + \int_0^\infty e^{-i\omega t} d\overline{\zeta(\omega)} \\ &= \int_0^\infty [e^{i\omega t} + e^{-i\omega t}] \Re[d\zeta(\omega)] + \int_0^\infty i[e^{i\omega t} - e^{-i\omega t}] \Im[d\zeta(\omega)] \\ &= \int_0^\infty 2 \cos(\omega t) \Re[d\zeta(\omega)] - \int_0^\infty 2 \sin(\omega t) \Im[d\zeta(\omega)] \\ &= \int_0^\infty \cos(\omega t) dU(\omega) + \int_0^\infty \sin(\omega t) dV(\omega) \end{aligned} \quad (11)$$

6. **Orthogonality Verification:** For the real-valued process with one-sided representation, the spectral distribution function $F(\omega)$ is related to the two-sided function by $dF(\omega) = 2 dF_{two}(\omega)$ for $\omega > 0$. Since the real and imaginary parts of $d\zeta(\omega)$ are orthogonal with equal variances:

$$\mathbb{E}[[\Re[d\zeta(\omega)]]^2] = \mathbb{E}[[\Im[d\zeta(\omega)]]^2] = \frac{1}{2} dF_{two}(\omega) = \frac{1}{4} dF(\omega) \quad (12)$$

Therefore,

$$\mathbb{E}[dU(\omega)^2] = \mathbb{E}[dV(\omega)^2] = 4 \cdot \frac{1}{4} dF(\omega) = dF(\omega) \quad (13)$$

7. **Covariance Function:** The covariance is given by

$$\begin{aligned} r(t) &= \mathbb{E}[\xi(t) \xi(0)] \\ &= \mathbb{E}\left[\left(\int_0^\infty \cos(\omega t) dU(\omega) + \sin(\omega t) dV(\omega)\right) \int_0^\infty dU(\omega')\right] \\ &= \int_0^\infty \cos(\omega t) \mathbb{E}[dU(\omega)^2] \end{aligned} \quad (14)$$

where cross-terms vanish by orthogonality and the sine term vanishes since $\sin(\omega \cdot$

$0) = 0$. Using $\mathbb{E} [dU(\omega)^2] = dF(\omega)$:

$$r(t) = \int_0^\infty \cos(\omega t) dF(\omega) \quad (15) \quad \square$$

Corollary 2. (Physical Interpretation) *In the real spectral representation:*

1. $\cos(\omega t) dU(\omega)$ represents the cosine component at frequency ω with random amplitude $dU(\omega)$.
2. $\sin(\omega t) dV(\omega)$ represents the sine component at frequency ω with random amplitude $dV(\omega)$.
3. $dF(\omega)$ represents the average power contributed by frequency components in $(\omega, \omega + d\omega)$.
4. The random measures $U(\omega)$ and $V(\omega)$ are uncorrelated and have equal variance increments.

Theorem 3. (U and V Random Measures) *For a real-valued stationary process $\xi(t)$ with mean-square continuous sample paths and spectral representation*

$$\xi(t) = \int_0^\infty [\cos(\omega t) dU(\omega) + \sin(\omega t) dV(\omega)] \quad (16)$$

the random measures $U(\omega)$ and $V(\omega)$ are given explicitly by:

1. ***U-process formula:***

$$U(\omega) = \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{\sin(\omega t)}{t} \xi(t) dt \quad (17)$$

2. ***V-process formula:***

$$V(\omega) = - \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{1 - \cos(\omega t)}{t} \xi(t) dt \quad (18)$$

3. ***Incremental form:***

$$U(\omega_2) - U(\omega_1) = \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{\sin(\omega_2 t) - \sin(\omega_1 t)}{t} \xi(t) dt \quad (19)$$

$$V(\omega_2) - V(\omega_1) = - \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{\cos(\omega_1 t) - \cos(\omega_2 t)}{t} \xi(t) dt \quad (20)$$

Proof. The formulas follow from the Fourier inversion theorem applied to the complex spectral measure. Starting from the complex inversion formula:

$$\zeta(\omega) = \lim_{T \rightarrow \infty} \frac{1}{2\pi} \int_{-T}^T \frac{1 - e^{-i\omega t}}{-it} \xi(t) dt \quad (21)$$

Using the definitions $dU(\omega) = 2 \Re [d\zeta(\omega)]$ and $dV(\omega) = -2 \Im [d\zeta(\omega)]$, and the identity $1 - e^{-i\omega t} = (1 - \cos(\omega t)) + i \sin(\omega t)$, we obtain:

$$U(\omega) = 2 \Re \left[\lim_{T \rightarrow \infty} \frac{1}{2\pi} \int_{-T}^T \frac{1 - e^{-i\omega t}}{-it} \xi(t) dt \right] = \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{\sin(\omega t)}{t} \xi(t) dt \quad (22)$$

$$V(\omega) = -2 \Im \left[\lim_{T \rightarrow \infty} \frac{1}{2\pi} \int_{-T}^T \frac{1 - e^{-i\omega t}}{-it} \xi(t) dt \right] = - \lim_{T \rightarrow \infty} \frac{1}{\pi} \int_{-T}^T \frac{1 - \cos(\omega t)}{t} \xi(t) dt \quad (23)$$

□

Remark 4. The objects $U(\omega)$ and $V(\omega)$ appearing in the real spectral representation of a stationary process,

$$\xi(t) = \int_0^\infty \cos(\omega t) dU(\omega) + \int_0^\infty \sin(\omega t) dV(\omega) \quad (24)$$

are *random measures* (or random set functions) on the frequency axis $[0, \infty)$. Their main property is that their increments over disjoint frequency intervals are orthogonal, i.e., uncorrelated (and independent if Gaussian). The notation $U(\omega)$ denotes the cumulative random measure up to frequency ω :

$$U(\omega) = U([0, \omega]) \quad V(\omega) = V([0, \omega]) \quad (25)$$

For a stationary process with mean-square continuous sample paths, each sample path uniquely determines the corresponding random measures through the inversion formulas given above.