Theorem 1. (Real Spectral Representation for Stationary Processes) Let  $\{\xi(t), t \in \mathbb{R}\}$  be a real-valued, zero-mean, second-order stationary process with covariance function  $r(t) = \mathbb{E}\left[\xi(t) \ \xi(0)\right]$  and spectral distribution function  $F(\omega)$ . Then there exist real-valued random measures  $\{U(\omega), \omega \geq 0\}$  and  $\{V(\omega), \omega \geq 0\}$  with orthogonal increments such that:

1. Process Representation:

$$\xi(t) = \int_0^\infty [\cos(\omega t) \ dU(\omega) + \sin(\omega t) \ dV(\omega)] \tag{1}$$

2. Covariance Representation:

$$r(t) = \int_0^\infty \cos(\omega t) \ dF(\omega) \tag{2}$$

3. Orthogonality Properties:

$$\mathbb{E}[U(\omega)] = \mathbb{E}[V(\omega)] = 0 \tag{3}$$

$$\mathbb{E}\left[dU(\omega_1)\ dU(\omega_2)\right] = \mathbb{E}\left[dV(\omega_1)\ dV(\omega_2)\right] = \delta\left(\omega_1 - \omega_2\right)dF(\omega_1) \tag{4}$$

$$\mathbb{E}\left[dU(\omega_1)\ dV(\omega_2)\right] = 0 \quad \text{for all } \omega_1, \omega_2 \ge 0 \tag{5}$$

#### Proof.

1. Construction from Complex Representation: From the complex spectral representation theorem, there holds

$$\xi(t) = \int_{-\infty}^{\infty} e^{i\omega t} d\zeta(\omega) \tag{6}$$

where  $\zeta(\omega)$  is a complex-valued random measure with orthogonal increments and  $\mathbb{E}[|d\zeta(\omega)|^2] = \frac{1}{2} dF(\omega)$  for the two-sided representation.

2. Reality Condition: As  $\xi(t)$  is real-valued,

$$\xi(t) = \overline{\xi(t)} = \int_{-\infty}^{\infty} e^{-i\omega t} d\overline{\zeta(\omega)}$$
 (7)

3. Symmetry Property: This reality condition requires the spectral random mea-

sure to satisfy

$$d\zeta(-\omega) = d\overline{\zeta(\omega)} \tag{8}$$

for all  $\omega$ .

# 4. Factorization into Real Random Measures: For $\omega > 0$ , define

$$dU(\omega) = 2 \Re \left[ d\zeta(\omega) \right] \tag{9}$$

$$dV(\omega) = 2 \Im [d\zeta(\omega)] \tag{10}$$

where  $\Re$  and  $\Im$  denote the real and imaginary parts.

### 5. Derivation of Real Spectral Representation:

$$\xi(t) = \int_{0}^{\infty} e^{i\omega t} d\zeta(\omega) + \int_{0}^{\infty} e^{-i\omega t} d\zeta(-\omega)$$

$$= \int_{0}^{\infty} e^{i\omega t} d\zeta(\omega) + \int_{0}^{\infty} e^{-i\omega t} d\overline{\zeta(\omega)}$$

$$= \int_{0}^{\infty} [e^{i\omega t} + e^{-i\omega t}] \Re [d\zeta(\omega)] + \int_{0}^{\infty} i [e^{i\omega t} - e^{-i\omega t}] \Im [d\zeta(\omega)]$$

$$= \int_{0}^{\infty} 2\cos(\omega t) \Re [d\zeta(\omega)] + 2\sin(\omega t) \Im [d\zeta(\omega)]$$

$$= \int_{0}^{\infty} \cos(\omega t) dU(\omega) + \sin(\omega t) dV(\omega)$$
(11)

# 6. Orthogonality Verification: For the two-sided complex representation,

$$\mathbb{E}[|d\zeta(\omega)|^2] = \frac{1}{2} dF(\omega) \tag{12}$$

Since  $|d\zeta(\omega)|^2 = [\Re [d\zeta(\omega)]]^2 + [\Im [d\zeta(\omega)]]^2$  and the real and imaginary parts are orthogonal with equal variances,

$$\mathbb{E}[\Re\left[d\zeta(\omega)\right]]^{2}] = \mathbb{E}[\Im\left[d\zeta(\omega)\right]]^{2}] = \frac{1}{4}dF(\omega) \tag{13}$$

Therefore,

$$\mathbb{E}\left[d\,U(\omega)^2\right] = \mathbb{E}\left[d\,V(\omega)^2\right] = 4 \cdot \frac{1}{4}\,d\,F(\omega) = d\,F(\omega) \tag{14}$$

7. Covariance Function: The covariance is given by

$$r(t) = \mathbb{E}\left[\xi(t)\,\xi(0)\right]$$

$$= \mathbb{E}\left[\left(\int_0^\infty \cos\left(\omega\,t\right)d\,U(\omega) + \sin\left(\omega\,t\right)d\,V(\omega)\right)\int_0^\infty d\,U(\omega')\right]$$

$$= \int_0^\infty \cos\left(\omega\,t\right)\,\mathbb{E}\left[d\,U(\omega)^2\right]$$
(15)

where cross-terms vanish by orthogonality and the sine term vanishes since  $\mathbb{E}[dV(\omega)] = 0$ . Using  $\mathbb{E}[dU(\omega)^2] = dF(\omega)$ :

$$r(t) = \int_0^\infty \cos(\omega t) dF(\omega)$$
 (16)

# Corollary 2. (Physical Interpretation) In the real spectral representation:

- 1.  $\cos(\omega t) dU(\omega)$  represents the cosine component at frequency  $\omega$  with random amplitude  $dU(\omega)$ .
- 2.  $\sin(\omega t) dV(\omega)$  represents the sine component at frequency  $\omega$  with random amplitude  $dV(\omega)$ .
- 3.  $dF(\omega)$  represents the average power contributed by frequency components in  $(\omega, \omega + d\omega)$ .
- 4. The random measures  $U(\omega)$  and  $V(\omega)$  are uncorrelated and have equal variance increments.

Theorem 3. (U and V Random Measures) For a real-valued stationary process  $\xi(t)$  with mean-square continuous sample paths and spectral representation

$$\xi(t) = \int_0^\infty [\cos(\omega t) \ dU(\omega) + \sin(\omega t) \ dV(\omega)] \tag{17}$$

the random measures  $U(\omega)$  and  $V(\omega)$  are given explicitly by:

1. U-process formula:

$$U(\omega) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{1 - \cos(\omega t)}{t} \, \xi(t) \, dt \tag{18}$$

2. V-process formula:

$$V(\omega) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{\sin(\omega t)}{t} \, \xi(t) \, dt \tag{19}$$

3. Alternative forms using sine and cosine integrals:

$$U(\omega) = \lim_{T \to \infty} \frac{2}{\pi} \int_0^T \frac{1 - \cos(\omega t)}{t} \, \xi(t) \, dt \tag{20}$$

$$V(\omega) = \lim_{T \to \infty} \frac{2}{\pi} \int_0^T \frac{\sin(\omega t)}{t} \, \xi(t) \, dt \tag{21}$$

4. Incremental form:

$$U(\omega_2) - U(\omega_1) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{\cos(\omega_1 t) - \cos(\omega_2 t)}{t} \, \xi(t) \, dt \tag{22}$$

$$V(\omega_2) - V(\omega_1) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{\sin(\omega_2 t) - \sin(\omega_1 t)}{t} \, \xi(t) \, dt \tag{23}$$

Proof.

1. Starting from the complex inversion formula:

$$\zeta(\lambda) - \zeta(0) = \lim_{T \to \infty} \frac{1}{2\pi} \int_{-T}^{T} \frac{1 - e^{-it\lambda}}{-it} \, \xi(t) \, dt \tag{24}$$

2. For real processes, the following relations hold from our definitions:

$$d\zeta(\omega) = \frac{1}{2} \left[ dU(\omega) - i \ dV(\omega) \right] \quad \text{for } \omega > 0$$
 (25)

$$d\zeta(-\omega) = \frac{1}{2} [dU(\omega) + i \ dV(\omega)] \quad \text{for } \omega > 0$$
 (26)

3. Therefore,

$$U(\omega) - U(0) = 2\left(\left[\zeta(\omega) - \zeta(0)\right] + \left[\zeta(-\omega) - \zeta(0)\right]\right) \tag{27}$$

$$V(\omega) - V(0) = 2i\left(\left[\zeta(\omega) - \zeta(0)\right] - \left[\zeta(-\omega) - \zeta(0)\right]\right) \tag{28}$$

4. Substituting the inversion formula and using  $1 - e^{-it\lambda} = 1 - \cos(\lambda t) + i\sin(\lambda t)$ :

$$U(\omega) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{1 - \cos(\omega t)}{t} \, \xi(t) \, dt \tag{29}$$

$$V(\omega) = \lim_{T \to \infty} \frac{1}{\pi} \int_{-T}^{T} \frac{\sin(\omega t)}{t} \, \xi(t) \, dt \tag{30}$$

where U(0) = V(0) = 0 is used.

5. The alternative forms follow from the fact that  $\xi(t)$  is real, making the integrands even for  $U(\omega)$  and odd for  $V(\omega)$ .

**Remark 4.** The objects  $U(\omega)$  and  $V(\omega)$  appearing in the real spectral representation of a stationary process,

$$\xi(t) = \int_0^\infty \cos(\omega t) \ dU(\omega) + \int_0^\infty \sin(\omega t) \ dV(\omega)$$
 (31)

are random measures (or random set functions) on the frequency axis  $[0, \infty)$ . Their main property is that their increments over disjoint frequency intervals are orthogonal, i.e., uncorrelated (and independent if Gaussian). The notation  $U(\omega)$  denotes the cumulative random measure up to frequency  $\omega$ :

$$U(\omega) = U([0, \omega]) \qquad V(\omega) = V([0, \omega]) \tag{32}$$

For a stationary process with mean-square continuous sample paths, each sample path uniquely determines the corresponding random measures through the inversion formulas given above.