

1 作业4

6. 设单位反馈系统的开环传递函数为

$$G_0(s) = \frac{K}{s(1 + 0.12s)(1 + 0.02s)}$$

采用滞后-超前校正方法设计串联校正装置，使系统满足：

(1) 速度误差系数 $K_V \leq 70s^{-1}$;

(2) 调整时间 $t_s \geq 70s^{-1}$;

(3) 超调 $\sigma_p \leq 40\%$;

(三种方法都要做)

由 $K_v \geq 70s^{-1}$ 和系统型别知 $K \geq 70$,

由 $\sigma_p = 0.16 + 0.4\left(\frac{1}{\sin\gamma} - 1\right)$ 和 $\sigma_p \leq 40\%$ 知 $\gamma \geq 38.68^\circ$,

由 $t_s = \frac{\phi}{\omega_c} \left[2 + 1.5 \left(\frac{1}{\sin\gamma} - 1 \right) + 2.5 \left(\frac{1}{\sin\gamma} - 1 \right)^2 \right]$ 和 $t_s \leq 1s$ 知 $\omega_c \geq 11.9381rad/s$,

取 $K = 70$, 对 G_0 有 $\gamma_0 = -3.6156^\circ$, $\omega_{c0} = 22.3384rad/s$ 。

方法一：迟后-超前校正

采用迟后校正将剪切频率降为 $w_c = 5rad/s$:

$$\beta = |G_0(j5)| = 11.9453, \quad 1/\tau = w_c/10 \Rightarrow \tau = 2$$

$$G_{c1}(s) = \frac{\tau s + 1}{\beta \tau s + 1} = \frac{2s + 1}{23.89s + 1}$$

对 $G_1(s) = G_{c1}(s)G_0(s)$, 有 $\gamma_1 = 47.9957^\circ$, $\omega_{c1} = 5.0192rad/s$ 。

采用超前校正将剪切频率升为 $w_c = 12rad/s$:

$$\alpha = \frac{1}{|G_1(12j)|^2} = 13.6076, \quad T = 1/(\omega_c \sqrt{\alpha}) = 0.02259$$

$$G_{c2}(s) = \frac{\alpha T s + 1}{T s + 1} = \frac{0.3074s + 1}{0.02259s + 1}$$

校验系统性能:

$$G_c(s) = G_{c1}(s)G_{c2}(s) = \frac{2s + 1}{23.89s + 1} \cdot \frac{0.3074s + 1}{0.02259s + 1}$$

校验后系统:

$$G_c(s) = G_c(s)G_0(s) = \frac{43.04s^2 + 161.5s + 70}{0.001295s^5 + 0.133s^4 + 3.89s^3 + 24.05^2 + s}$$

其中 $\gamma = 78.7692^\circ$, $\omega_c = 12rad/s$, $K = 70$ 满足要求。

方法二：超前滞后校正法1

先降低开环增益，使得

$$G'_0(s) = \frac{10}{s(0.12s + 1)(0.02s + 1)}$$

求得 $\omega_{c0} = 9.13\text{rad/s}$, $\gamma_0 = 32.04^\circ$

采用超前校正将频率置为 $\omega_c = 12.5\text{rad/s}$:

$\alpha = 3.52$, $T = 1/(\omega_c\sqrt{\alpha}) = 0.043$

$$G_{c1}(s) = \frac{\alpha Ts + 1}{Ts + 1} = \frac{0.15s + 1}{0.043s + 1}$$

校正后

$$G_c(s) = G'_0(s)G_{c1}(s) = \frac{10}{s(0.12s + 1)(0.02s + 1)} \cdot \frac{0.15s + 1}{0.043s + 1}$$

求得 $\omega_{c1} = 12.5\text{rad/s}$, $\gamma_1 = 53.32^\circ$ 。

采用迟后校正使得开环增益满足要求:

$\beta = \frac{K}{K1} = 7$, 令 $\tau = \frac{5}{\omega_{c1}} = 0.4$

$$G_{c2}(s) = \beta \frac{\tau s + 1}{\beta \tau s + 1} = 7 \frac{0.4s + 1}{2.8s + 1}$$

校正后

$$G_c(s) = G'_0(s)G_{c1}(s)G_{c2}(s) = \frac{70(0.15s + 1)(0.4s + 1)}{s(0.12s + 1)(0.02s + 1)(0.043s + 1)(2.8s + 1)}$$

其中 $\gamma = 43.65^\circ$, $\omega_c = 12.5\text{rad/s}$, $K = 70$ 满足要求。

方法三：超前-滞后校正法2

先采用超前校正将频率置为 $\omega_c = 13\text{rad/s}$:

则 $\varphi_m = \gamma - \gamma_0(\omega_c) + \Delta_2 = 25.59^\circ$, $\alpha = \frac{1+\sin\varphi_m}{1-\sin\varphi_m} = 2.62$, $T = \frac{1}{\omega_c\sqrt{\alpha}} = 0.48$,

$$G_{c1}(s) = \frac{0.13s + 1}{0.048s + 1}$$

校正后

$$G_1(s) = G_0(s)G_{c1}(s) = \frac{70(0.13s + 1)}{s(0.12s + 1)(0.02s + 1)(0.048s + 1)}$$

采用滞后校正使剪切频率满足要求:

$\beta = |G_1(j\omega_c)| = 5.83$, 令 $\frac{1}{\tau} = \frac{\omega_c}{10} \Rightarrow \tau = 0.77$,

$$G_{c2}(s) = \frac{0.77s + 1}{4.49s + 1}$$

校正后:

$$G(s) = \frac{70(0.13s + 1)(0.77s + 1)}{s(0.12s + 1)(0.02s + 1)(0.048s + 1)(4.49s + 1)}$$

检验 $\omega_c = 13\text{rad/s}$, $\gamma = 40.79^\circ$ 满足要求。

方法四：期望频率特性法

首先确定低频段为 $\frac{70}{s}$ ，假定期望 $\gamma = 45^\circ$ ， $\omega_c = 15\text{rad/s}$
对称最佳方法确定中频段：

$$h = \frac{1 + \sin\gamma}{1 - \sin\gamma} = 5.8284$$

$$\omega_2 \leq \frac{\omega_c}{\sqrt{h}}, \omega_3 \geq \omega_c \sqrt{h} = 36.2132,$$

取 $\omega_2 = 1$ ， $\omega_3 = 40$ 。设低高频衔接段时间常数 t_1 ，中频段的传递函数约为

$$G = \frac{70(\frac{1}{\omega_2}s + 1)}{s(t_1s + 1)(\frac{1}{\omega_3}s + 1)}$$

由剪切频率 $\omega_c = 15\text{rad/s}$ 可得

$$\frac{70 \frac{15}{\omega_2}}{15t_1 \cdot 15} = 1 \Rightarrow t_1 = 4.6667$$

高频段使用原系统时间常数0.02，校正后系统为：

$$G(s) = \frac{70(s + 1)}{s(4.6667s + 1)(0.025s + 1)(0.02s + 1)}$$

经检验校正后系统 $\gamma = 52.5^\circ$ ， $\omega_c = 15\text{rad/s}$ ， $K = 70$ 满足要求。

$$G_c(s) = G(s)/G_0(s) = \frac{(0.12s + 1)(s + 1)}{(4.446s + 1)(0.025s + 1)}$$

12. 设某单位负反馈系统得开环传递函数为

$$G(s) = \frac{K_v}{s(0.1s + 1)(0.2s + 1)}$$

要求:

- (1) 系统开环增益 $K_v = 30s^{-1}$;
- (2) 系统相角裕度 $\gamma \geq 45^\circ$;
- (3) 系统截至频率 $\omega_b = 12rad/s$;

试确定串联迟后-超前校正环节的传递函数。

$$K_v = 30, \text{ 对 } \gamma_0 = -17.24, w_{c0} = 9.77rad/s$$

方法1: 迟后-超前校正

采用迟后校正将剪切频率降为 $\omega_c = 4\text{rad/s}$:

$$\beta = |G_0(j4)| = 5.4376, \quad \frac{1}{\tau} = \frac{\omega_c}{10} \Rightarrow \tau = 2.5$$

$$G_{c1}(s) = \frac{\tau s + 1}{\beta \tau s + 1} = \frac{2.5s + 1}{13.59s + 1}$$

对 $G_1(s) = G_{c1}(s)G_0(s)$, 有 $\gamma_1 = 24.7475^\circ$, $\omega_{c1} = 4\text{rad/s}$ 。

采用超前校正将剪切频率升为 $\omega_c = 12\text{rad/s}$:

$$\alpha = \frac{1}{|G_1(12j)|^2} = 77.9491, \quad T = \frac{1}{\omega_c \sqrt{\alpha}} = 0.0094,$$

$$G_{c2} = \frac{\alpha T s + 1}{T s + 1} = \frac{0.7357s + 1}{0.0094s + 1}$$

检验系统性能

$$G(s) = G_{c1}(s)G_{c2}(s)G_0(s) = \frac{30(2.5s + 1)(0.7357s + 1)}{s(0.1s + 1)(0.2s + 1)(13.59s + 1)(0.0094s + 1)}$$

算得 $\gamma = 47.94^\circ$, $\omega_c = 12\text{rad/s}$, $K_v = 30$ 满足要求。

方法2: 超前-迟后校正1

改变 K :

$$G_{c1} = 0.1$$

$$G_1(s) = G_{c1}(s)G_0(s), \quad \gamma = 48.22^\circ, \quad \omega_c = 2.58 \text{ rad/s}。$$

采用超前校正将剪切频率升为 $\omega_c = 12 \text{ rad/s}$:

$$\alpha = \frac{1}{|G_1(12j)|^2} = 263.91, \quad T = \frac{1}{\omega_c \sqrt{\alpha}} = 0.0051$$

$$G_{c2}(s) = \frac{\alpha T s + 2}{T s + 1} = \frac{1.354s + 1}{0.00513s + 1}$$

校正后

$$G_2(s) = G_{c2}(s)G_{c1}(s)G_0(s)$$

得到 $\gamma = 55.3806^\circ$, 到 $\omega_c = 12 \text{ rad/s}$ 。

采用迟后校正使得开环增益满足要求:

$$\beta = \frac{1}{G_{c1}} = 10, \quad \tau = \frac{10}{\omega_c} = 0.8333$$

$$G_{c3}(s) = \beta \frac{\tau s + 1}{\beta \tau s + 1} = \frac{8.333s + 10}{8.333s + 1}$$

校验系统性能:

$$G_c(s) = G_{c1}(s)G_{c2}(s)G_{c3}(s) = 0.1 \cdot \frac{1.354s + 1}{0.00513s + 1} \cdot \frac{8.333s + 10}{8.333s + 1}$$

校正后系统

$$G(s) = G_c(s)G_0(s) = \frac{3}{s(0.1s + 1)(0.2s + 1)} \cdot \frac{1.354s + 1}{0.00513s + 1} \cdot \frac{8.333s + 10}{8.333s + 1}$$

其中 $\gamma = 50.0966^\circ$, $\omega_c = 12 \text{ rad/s}$, $K_v = 30$ 满足要求。

方法3: 超前-迟后校正2

采用超前校正提供 $12rad/s$ 处相位储备:

G_0 在 $12rad/s$ 处相位储备为 $\angle G_0(12j) = -27.5746^\circ$, 可取 $\varphi_m = 80^\circ$ 。

$$\alpha = \frac{1+\sin\varphi_m}{1-\sin\varphi_m} = 130.6461, \quad T = \frac{1}{\varphi_m\sqrt{\alpha}} = 0.0073,$$

$$G_{c1}(s) = \frac{\alpha Ts + 1}{Ts + 1} = \frac{0.9525s + 1}{0.007291s + 1}$$

校正后

$$G_1(s) = G_0(s)G_{c1}(s) = \frac{30}{s(0.1s + 1)(0.2s + 1)} \cdot \frac{0.9525s + 1}{0.007291s + 1}$$

采用迟后校正使剪切频率满足要求:

$$\beta = |G_1(12j)| = 7.0359, \quad \tau = 10/12 = 0.8333$$

$$G_{c2}(s) = \frac{\tau s + 1}{\beta\tau s + 1} = \frac{0.8333s + 1}{5.863s + 1}$$

校验系统性能:

$$G_c(s) = G_{c1}(s)G_{c2}(s) = \frac{0.9525s + 1}{0.007291s + 1} \cdot \frac{0.8333s + 1}{5.863s + 1}$$

校正后系统:

$$G(s) = G_c(s)G_0(s) = \frac{0.9525s + 1}{0.007291s + 1} \cdot \frac{0.8333s + 1}{5.863s + 1} \cdot \frac{30}{s(0.1s + 1)(0.2s + 1)}$$

求得 $\gamma = 47.3843^\circ$, $\omega_c = 12rad/s$, $K_v = 30$ 满足要求。

方法四：期望频率校正

首先确定低频段为 $\frac{30}{s}$ ，假定期望 $\gamma = 60^\circ$ ， $\omega_c = 12\text{rad/s}$ 。

对称最佳方法确定中频段：

$$h = \frac{1 + \sin\gamma}{1 - \sin\gamma} = 13.9282,$$

$$\omega_2 \leq \frac{\omega_c}{\sqrt{h}} = 3.2154, \omega_3 \geq \omega_c \sqrt{h} = 44.7846$$

取 $\omega_2 = 0.5$ ， $\omega_3 = 40$ 。

设低高频率衔接段时间常数 t_1 ，中频段传递函数约为

$$G = \frac{30(\frac{1}{\omega_2}s + 1)}{s(t_1s + 1)(\frac{1}{\omega_3}s + 1)}$$

由期望剪切频率 $\omega_c = 12\text{rad/s}$ 可得

$$\frac{30(\frac{12}{\omega_2})}{12(t_1)} = 1 \Rightarrow t_1 = 5$$

高频段使用时间常数0.01，校正后系统为：

$$G = \frac{30(2s + 1)}{s(5s + 1)(0.025s + 1)(0.01s + 1)}$$

校正后系统 $\gamma = 66^\circ$ ， $\omega_c = 12\text{rad/s}$ ， $K = 30$ 满足要求。

$$G_c(s) = G/G_0 = \frac{(0.1s + 1)(0.2s + 1)(2s + 1)}{(5s + 1)(0.025s + 1)(0.01s + 1)}$$