

# LTE PHY description Link Level Simulator

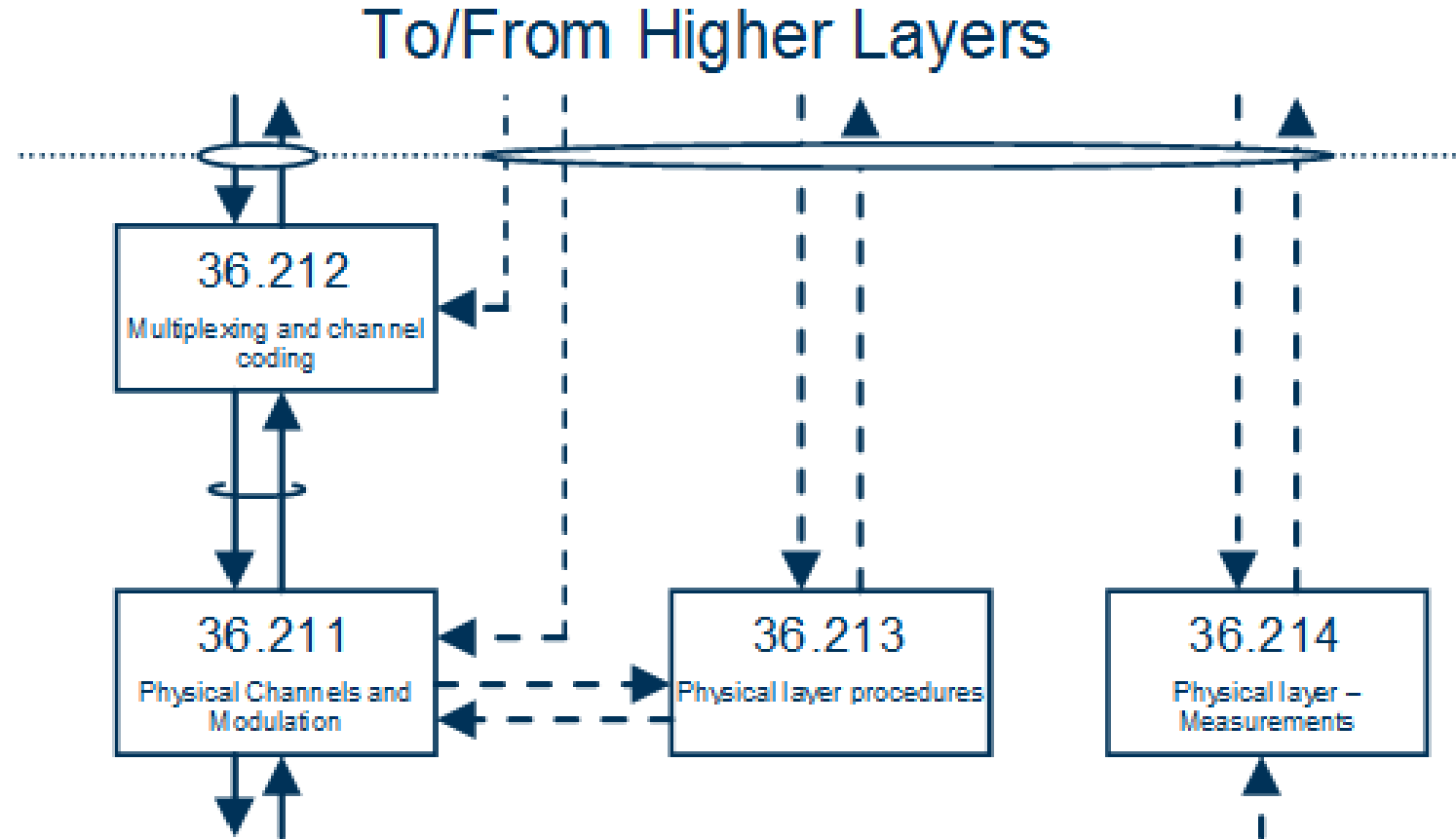
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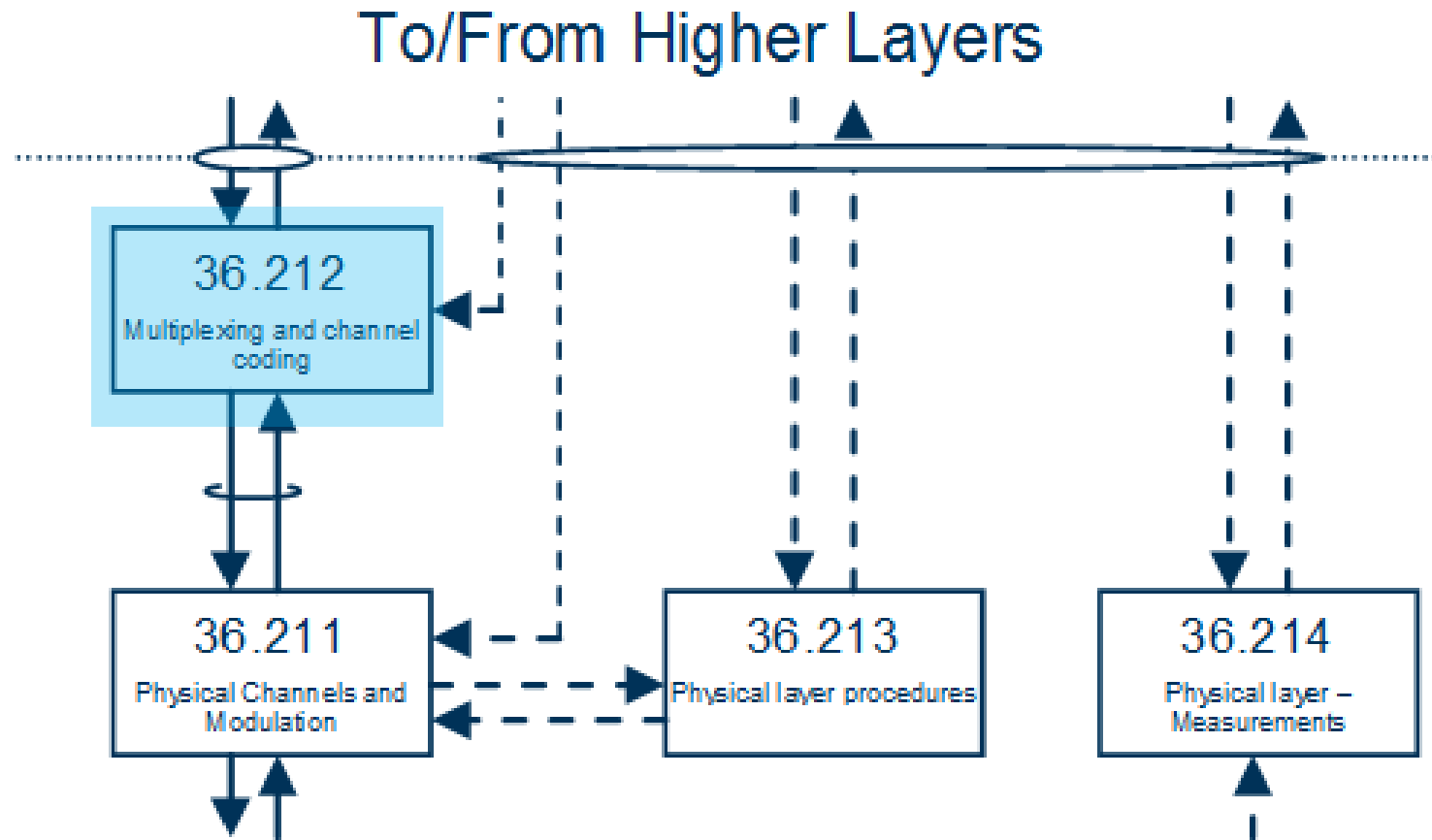
- Physical layer documents
- from TS 36.201



# TS 36.212 – Multiplexing and Channel Coding



# Multiplexing and Channel Coding



# Transport block processing (downlink)

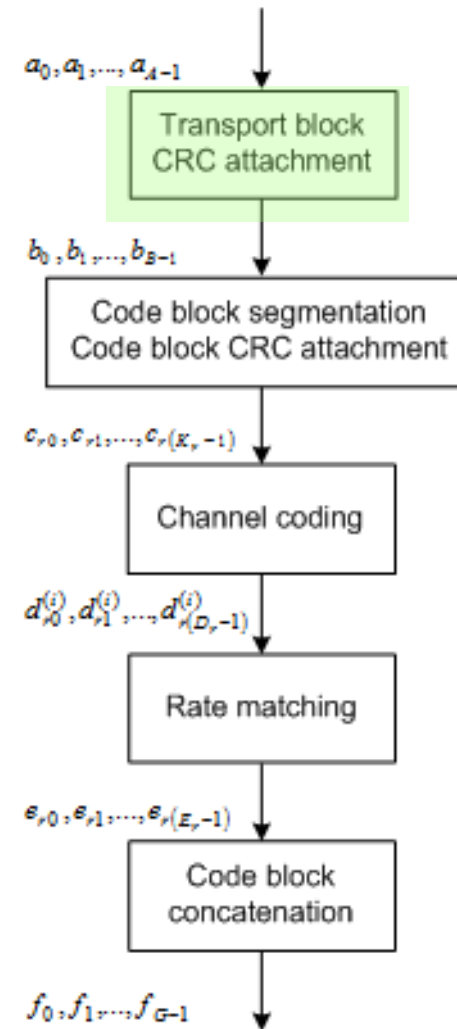


Figure 5.3.2-1: Transport block processing for DL-SCH, PCH and MCH.

- Input bits  $a_0, a_1, \dots, a_{A-1}$ , where  $A$  is the size of input sequence
- Parity bits  $p_0, p_1, \dots, p_{L-1}$ , where  $L$  is the number of parity bits
- The parity bits are generated by one of the following cyclic generator polynomials:
  - $g_{\text{CRC24A}}(D) = [D^{24} + D^{23} + D^{18} + D^{17} + D^{14} + D^{11} + D^{10} + D^7 + D^6 + D^5 + D^4 + D^3 + D + 1]$  and;
  - $g_{\text{CRC24B}}(D) = [D^{24} + D^{23} + D^6 + D^5 + D + 1]$  for a CRC length  $L = 24$  and;
  - $g_{\text{CRC16}}(D) = [D^{16} + D^{12} + D^5 + 1]$  for a CRC length  $L = 16$ .
  - $g_{\text{CRC8}}(D) = [D^8 + D^7 + D^4 + D^3 + D + 1]$  for a CRC length of  $L = 8$ .
- The use of one polynomial or another depends on the type of channel. In this simulator we will simulate only PDSCH.

- Input bits  $a_0, a_1, \dots, a_{A-1}$ , where A is the size of input sequence
- Parity bits  $p_0, p_1, \dots, p_{L-1}$ , where L is the number of parity bits
- The encoding is performed in a systematic form (the input data is embedded in the output data):

$$a_0 D^{A+23} + a_1 D^{A+22} + \dots + a_{A-1} D^{24} + p_0 D^{23} + p_1 D^{22} + \dots + p_{22} D^1 + p_{23}$$

- The bits after CRC attachment are denoted by:  $b_0, b_1, \dots, b_{B-1}$ , where  $B = A + L$ . The relationship between  $a_k$  and  $b_k$  is:
  - $b_k = a_k$ , for  $k = 0, 1, 2, \dots, A - 1$
  - $b_k = p_k$ , for  $k = A, A + 1, \dots, A + L - 1$

# Transport block processing (downlink)

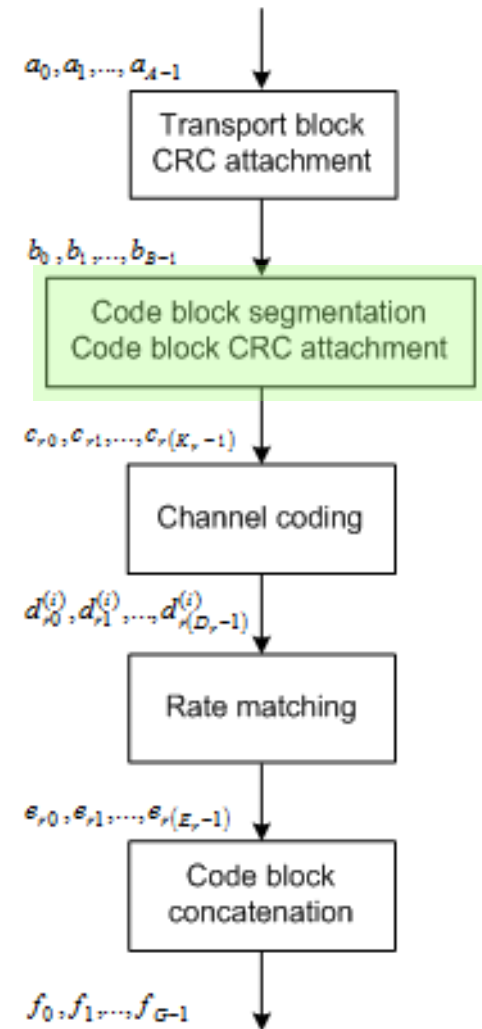


Figure 5.3.2-1: Transport block processing for DL-SCH, PCH and MCH.



# Code block segmentation and code block CRC attachment

- The maximum code block size is:  $Z = 6144$
- If  $B \geq Z$ , segmentation is performed and an additional CRC sequence of  $L = 24$  is attached to each code block. Generator polynomial  $g_{CRC24B}(D)$  is used in this case.

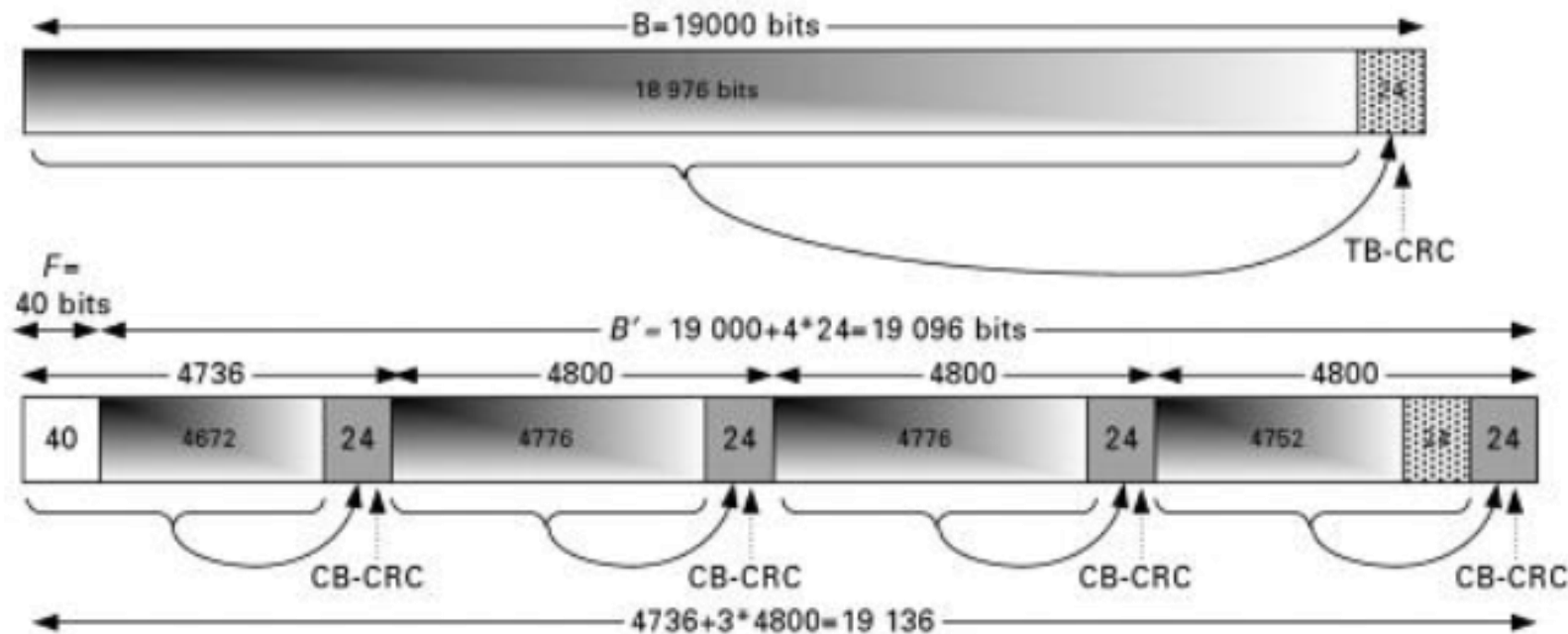


Figure 11.11. An example of codeblock segmentation.

- The bits output from code block segmentation are denoted by:  
 $c_{r0}, c_{r1}, \dots, c_r(k_r) - 1$
- $r$  is the code block number
- $K_r$  is the number of bits for the code block  $r$ .

- Segmentation is described in clause 5.1.2 of TS 36.212
- A very nice description is shown [on this link](#).

# Transport block processing (downlink)

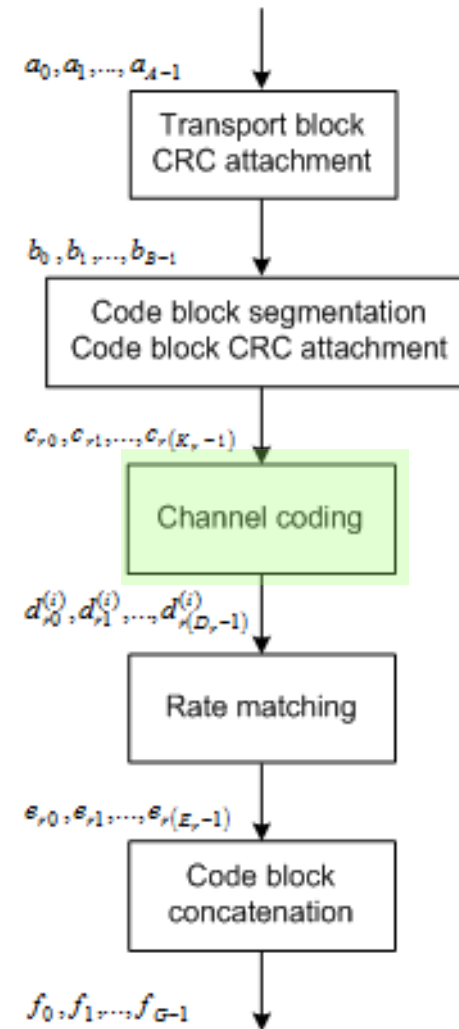


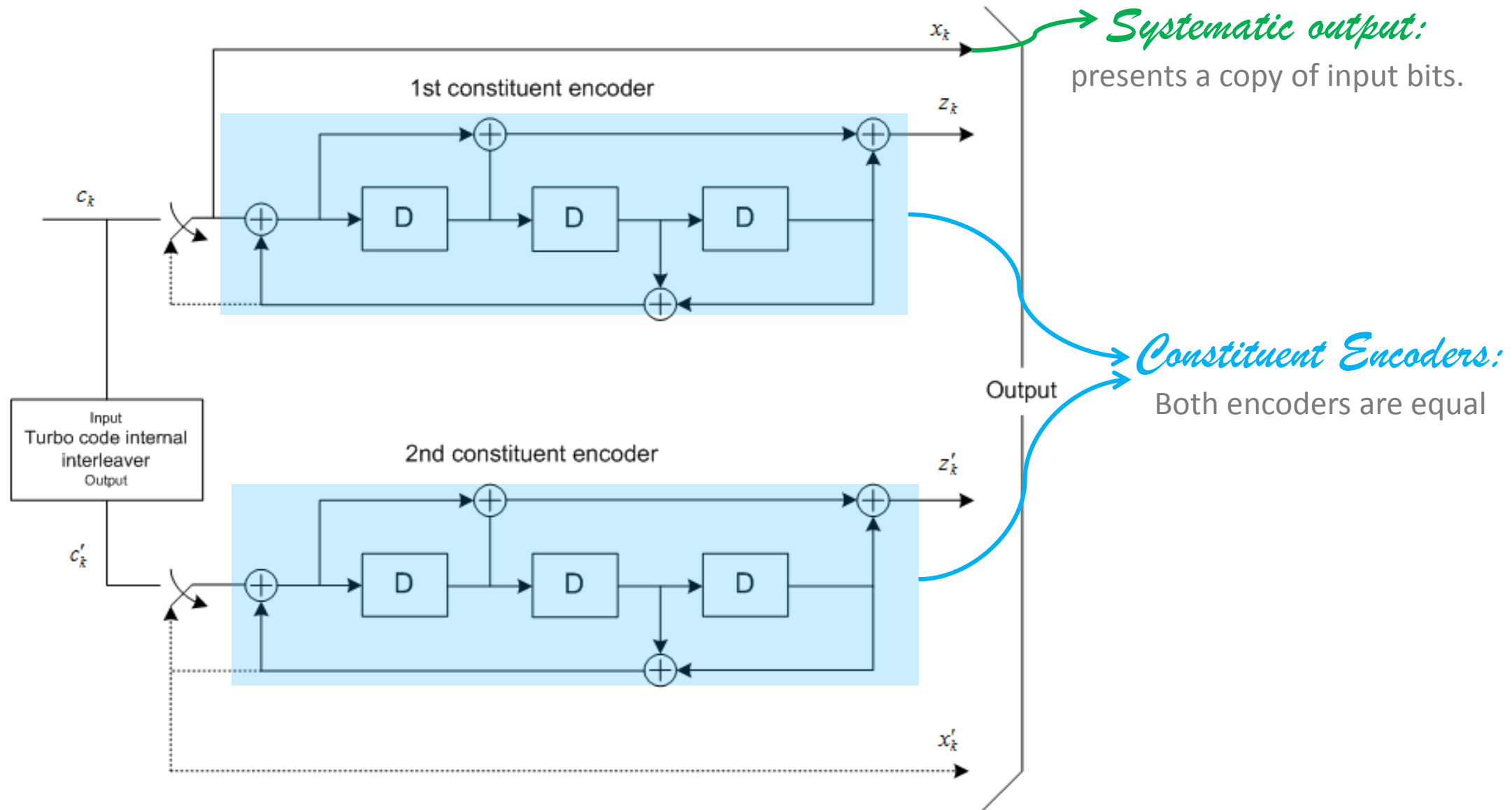
Figure 5.3.2-1: Transport block processing for DL-SCH, PCH and MCH.

# Channel Coding (general information)



- Channel coding can be divided in two methods:
  - Backward error correction (BEC): provides only error detection
  - Forward error correction (FEC): provides error detection and correction - ↓ retransmissions.
- Codes can be:
  - Systematic: the input bits appear on the output
  - Nonsystematic: the input bits do not appear on the output
- Important feature:
  - Code weight: number of “non-zero” symbols in the code word – high weight code words can be distinguished easier.

# Turbo Coding (5.1.3.2)



# Channel Coding (5.1.3)

- Bits on the encoder input:  $c_0, c_1, \dots, c_{K-1}$
- Bits at the encoder output:  $d_0^i, d_1^i, \dots, d_{D-1}^i$ , where  $D$  is the number of encoded bits per output stream and  $i$  is the encoder output stream index.
- Two coding schemes are applied on LTE
  - Tail biting convolutional encoding
  - Turbo Coding

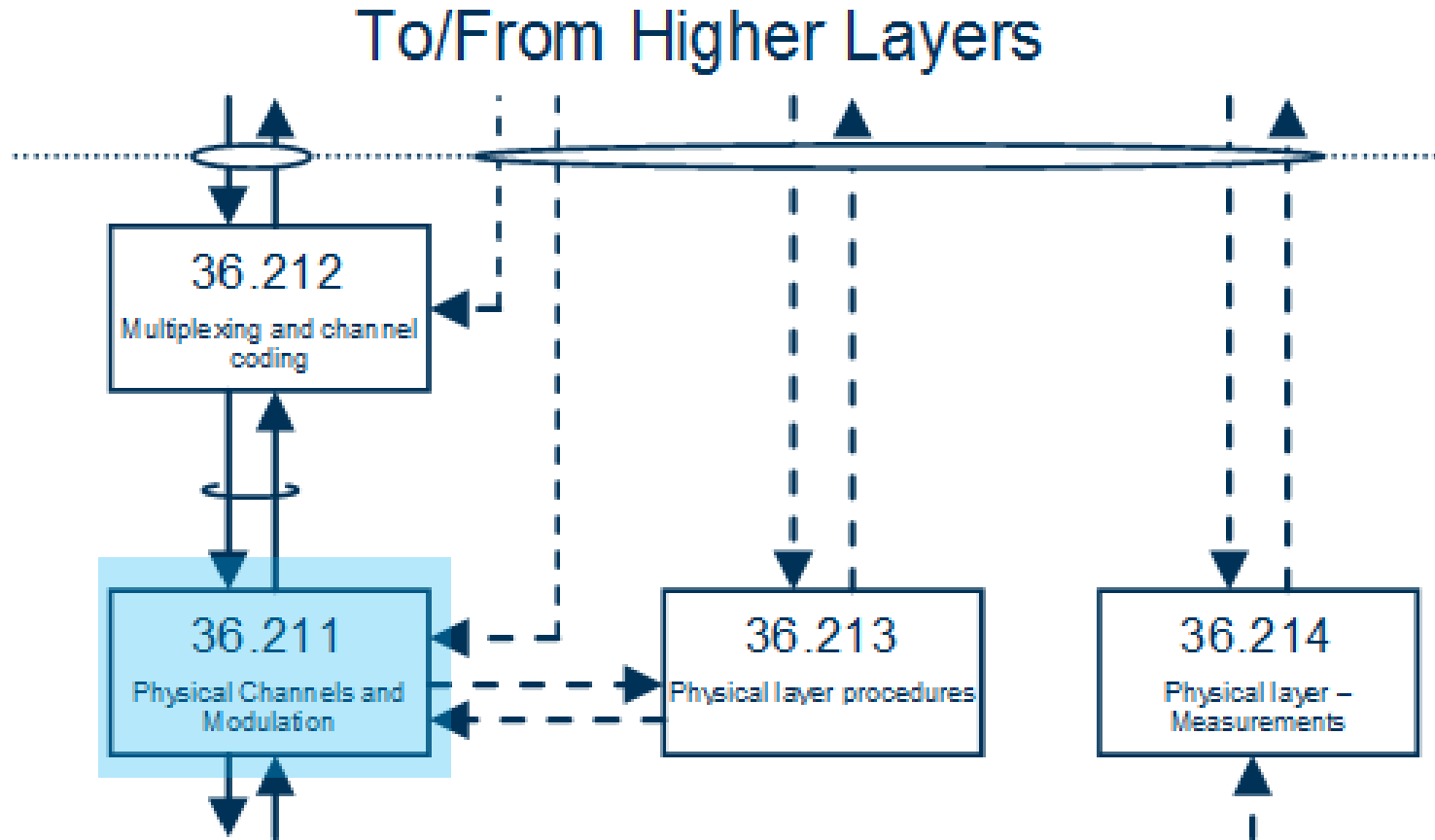
**On Link Simulations, we are interested on the channel that carries data**

TrCH	Coding scheme	Coding rate
UL-SCH	Turbo coding	1/3
DL-SCH		
PCH		
MCH		
BCH	Tail biting convolutional coding	1/3

# TS 36.211 – Physical Channels and modulation

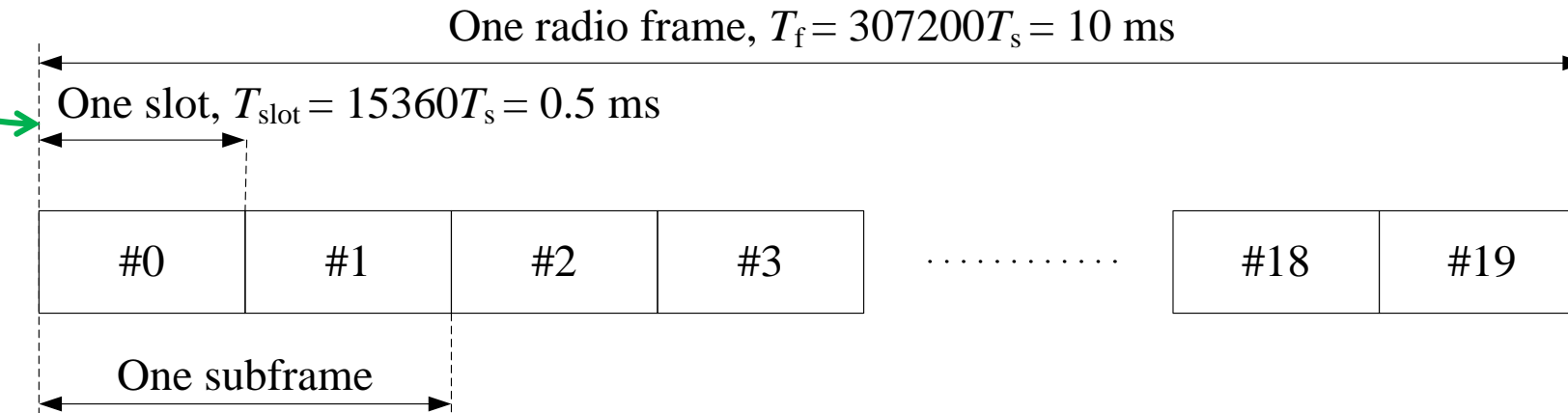


# Physical Channels and Modulation



# Frame structure – Type 1 (FDD)

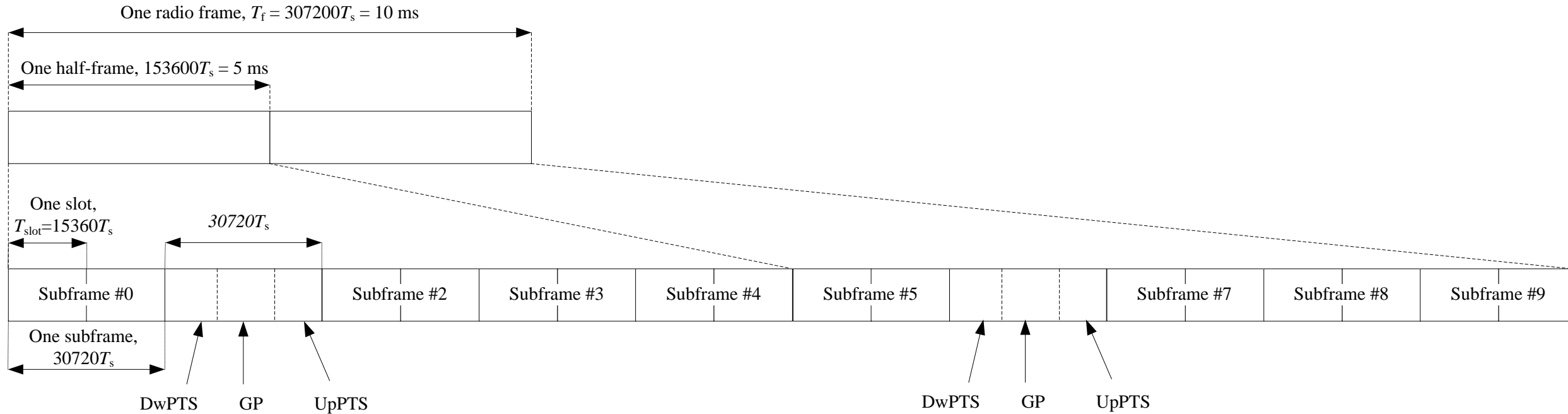
This was agreed to be on the first version of implemented PHY



- Simpler
- More suitable for Link level simulations



# Frame structure – Type 2 (TDD)



- There are special subframes used for switching between downlink and uplink transmissions

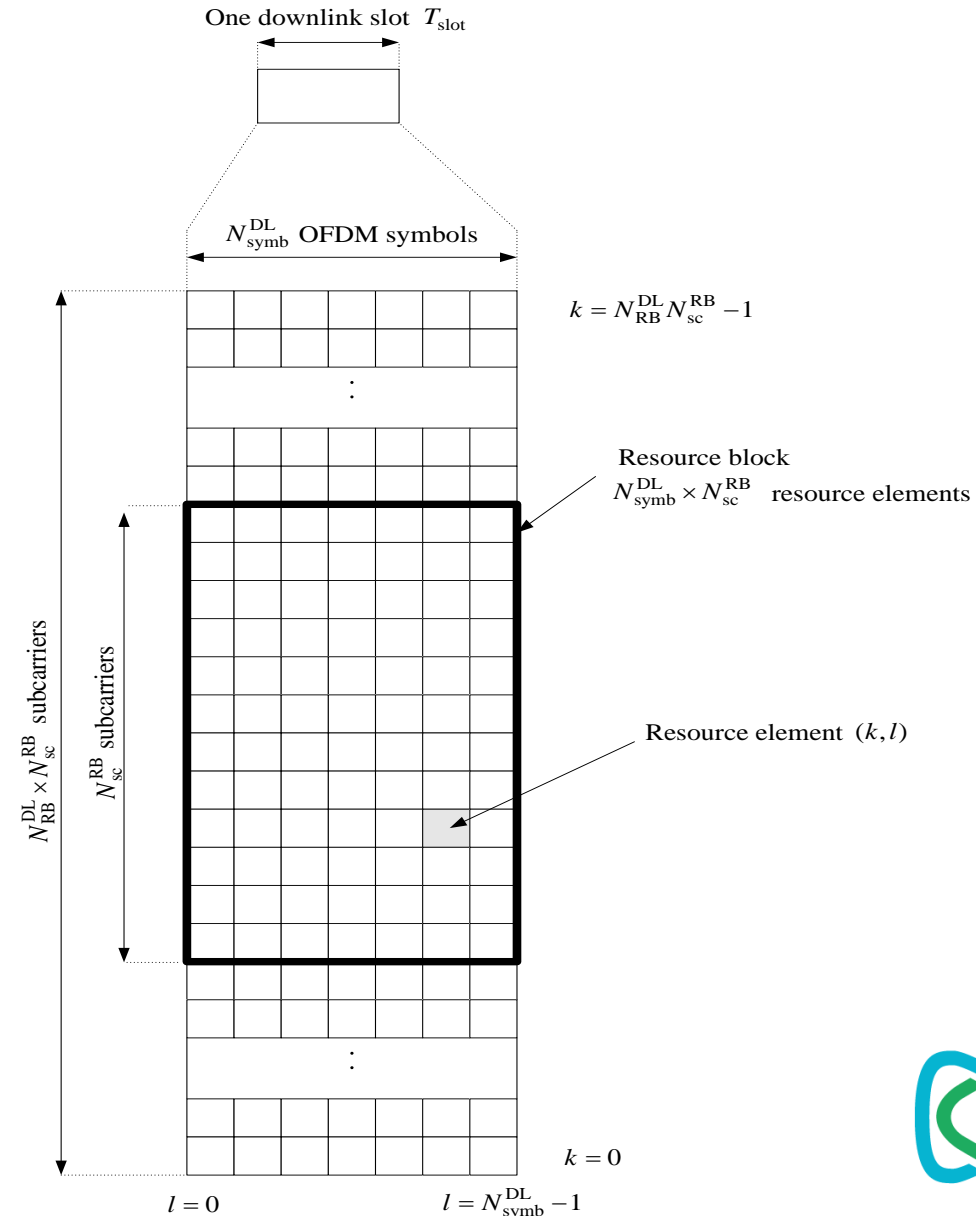
# Frame structure – Type 2 (TDD)

- Frame configuration doesn't allow DL only or UL only simulations.

Uplink-downlink configuration	Downlink-to-Uplink Switch-point periodicity	Subframe number									
		0	1	2	3	4	5	6	7	8	9
0	5 ms	D	S	U	U	U	D	S	U	U	U
1	5 ms	D	S	U	U	D	D	S	U	U	D
2	5 ms	D	S	U	D	D	D	S	U	D	D
3	10 ms	D	S	U	U	U	D	D	D	D	D
4	10 ms	D	S	U	U	D	D	D	D	D	D
5	10 ms	D	S	U	D	D	D	D	D	D	D
6	5 ms	D	S	U	U	U	D	S	U	U	D

# Downlink Description – Resource Elements

- Each element in the resource grid, identified by the index  $(k,l)$
- $k = 0, \dots, N_{RB}^{DL} N_{sc}^{RB} - 1$
- $l = 0, \dots, N_{symb}^{DL} - 1$



# Downlink Description – Resource Blocks

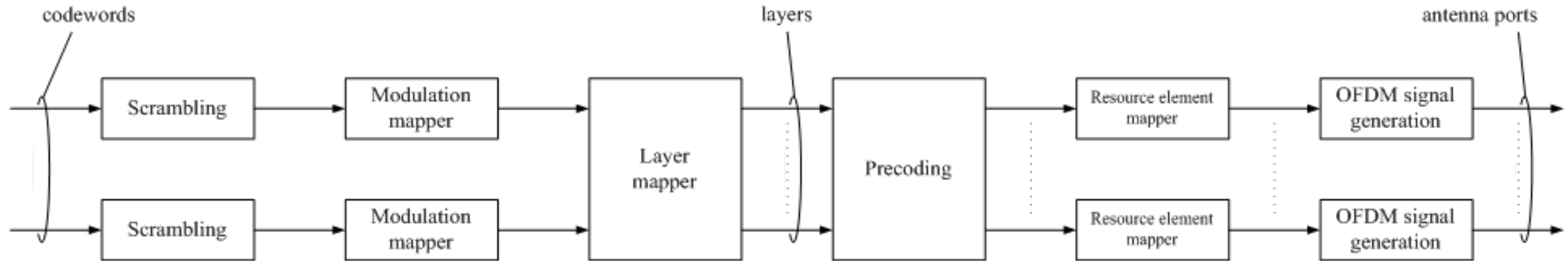
- Physical Resource block pair:
  - 2 physical RBs in one subframe having the same resource-block number

- $$n_{PRB} = \left\lfloor \frac{k}{N_{SC}^{RB}} \right\rfloor$$

This was agreed to be on the first version of implemented PHY

Configuration		$N_{sc}^{RB}$	$N_{symbol}^{DL}$
Normal cyclic prefix	$\Delta f = 15 \text{ kHz}$	12	7
	$\Delta f = 15 \text{ kHz}$		6
Extended cyclic prefix	$\Delta f = 7.5 \text{ kHz}$	24	3

# Downlink Description – General structure



# Downlink Description – Scrambling

- For each codeword  $q$ , the block of bits  $b^q(0), \dots, b^q(M_{bit}^q - 1)$  shall be scrambled prior to modulation, according to (Section 6.3.1):

$$\tilde{b}^{(q)}(i) = (b^{(q)}(i) + c^{(q)}(i)) \bmod 2$$

$$c_{init} = \begin{cases} n_{RNTI} \cdot 2^{14} + q \cdot 2^{13} \cdot \left\lfloor \frac{N_C = 1600}{s} \right\rfloor \cdot 2^9 + N_{ID}^{cell} & \text{for PDSCH} \\ \left\lfloor n_s / 2 \right\rfloor \cdot 2^9 + N_{ID}^{MBSFN} & \text{for PMCH} \end{cases}$$

$$c(n) = (x_1(n + N_C) + x_2(n + N_C)) \bmod 2$$

$$x_1(n + 31) = (x_1(n + 3) + x_1(n)) \bmod 2$$

$$x_2(n + 31) = (x_2(n + 3) + x_2(n + 2) + x_2(n + 1) + x_2(n)) \bmod 2$$

$$N_C = 1600$$

- Where  $n_{RNTI}$  corresponds to the RNTI associated with the PDSCH described in clause 7.1 in TS 36.213

# Downlink Description – Modulation

- For each codeword  $q$ , the block of scrambled bits  $\tilde{b}^q(0), \dots, \tilde{b}^q(M_{bit}^q - 1)$  shall be modulated (TS 36.2011 – clause 7.1 )

16 QAM

BPSK

$b(i)$	$I$	$Q$
0	$1/\sqrt{2}$	$1/\sqrt{2}$
1	$-1/\sqrt{2}$	$-1/\sqrt{2}$

$b(i), b(i+1), b(i+2), b(i+3)$	$I$	$Q$
0000	$1/\sqrt{10}$	$1/\sqrt{10}$
0001	$1/\sqrt{10}$	$3/\sqrt{10}$
0010	$3/\sqrt{10}$	$1/\sqrt{10}$
0011	$3/\sqrt{10}$	$3/\sqrt{10}$
0100	$1/\sqrt{10}$	$-1/\sqrt{10}$
0101	$1/\sqrt{10}$	$-3/\sqrt{10}$
0110	$3/\sqrt{10}$	$-1/\sqrt{10}$
0111	$3/\sqrt{10}$	$-3/\sqrt{10}$
1000	$-1/\sqrt{10}$	$1/\sqrt{10}$
1001	$-1/\sqrt{10}$	$3/\sqrt{10}$
1010	$-3/\sqrt{10}$	$1/\sqrt{10}$
1011	$-3/\sqrt{10}$	$3/\sqrt{10}$
1100	$-1/\sqrt{10}$	$-1/\sqrt{10}$
1101	$-1/\sqrt{10}$	$-3/\sqrt{10}$
1110	$-3/\sqrt{10}$	$-1/\sqrt{10}$
1111	$-3/\sqrt{10}$	$-3/\sqrt{10}$

QPSK

$b(i), b(i+1)$	$I$	$Q$
00	$1/\sqrt{2}$	$1/\sqrt{2}$
01	$1/\sqrt{2}$	$-1/\sqrt{2}$
10	$-1/\sqrt{2}$	$1/\sqrt{2}$
11	$-1/\sqrt{2}$	$-1/\sqrt{2}$

- 64 QAM is described on Table 7.1.4-1

- 256 QAM is described on Table 7.1.4-2

# Downlink Description – Layer Mapper

- Complex-valued modulation symbols for each of the codewords should be mapped on 1 or several layers.
  - $v$  is the number of layers,  $M_{symbol}^{layer}$  is the number of modulation symbols per layer
  - $x(i) = [x^0(i) \dots x^{v-1}(i)]$ ,  $i = 0, 1, \dots, M_{symbol}^{layer} - 1$
- **Single Antenna port**

$$x^0(i) = d^0(i)$$



# Downlink Description – Layer Mapper

- $v$  is the number of layers,  $M_{symbol}^{layer}$  is the number of modulation symbols per layer
- $P$  is the number of antenna ports
- **Multiple Antenna Ports**
  - Spatial Multiplexing
    - ✓  $v \leq P$
    - ✓ Codeword to layer mapping defined on Table 6.3.3.2-1
  - Transmit Diversity
    - ✓  $v = P$
    - ✓ Codeword to layer mapping defined on Table 6.3.3.3-1

# Downlink Description – Precoding

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Falta fazer essa parte!!!!



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# THANK YOU

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