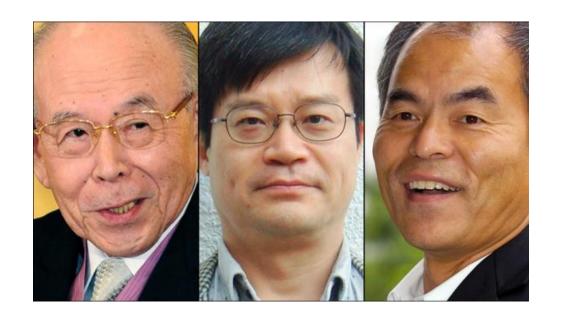
# BJT Amplifiers (Analog Circuits)

EE 101

S. Lodha

Reference material: L. Bobrow's Book

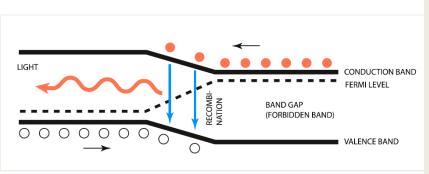
# Nobel Prize in Physics 2014

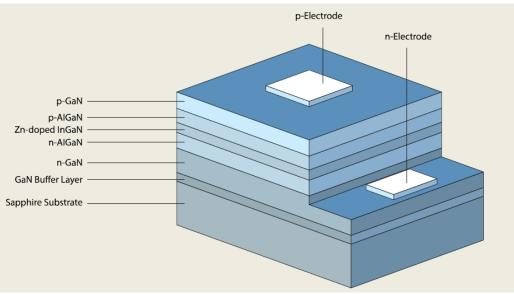


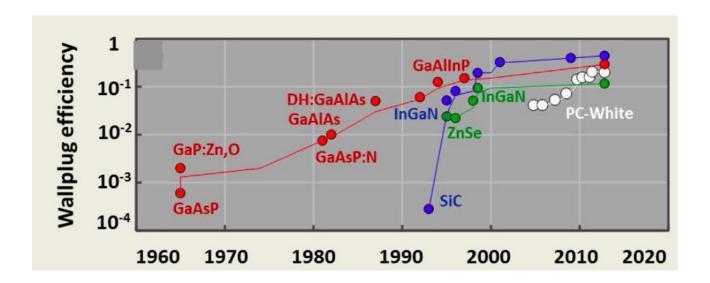
Isamu Akasaki, Hiroshi Amano and Shuji Nakamura

EFFICIENT BLUE LIGHT-EMITTING DIODES LEADING TO BRIGHT AND ENERGY-SAVING WHITE LIGHT SOURCES

### Blue LED



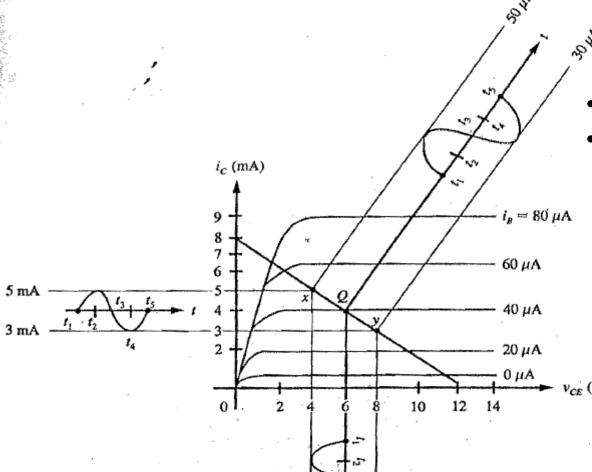




## **Impact**

- Transformed entire lighting industry
  - 20-30% of electricity is used for lighting
  - Standard bulb → 16 lm/W (4% efficiency)
  - LED  $\rightarrow$  300 lm/W (50% efficiency)
    - 100,000 hour lifetime
- Other applications
  - Displays in electronics
  - Sensors etc.

## **BJT Amplifier: Collector Characteristics**



- Active region operation
- Q → quiescent operating point

$$i_C = I_{CO} + i_c = 4 + \sin \omega t$$
 mA

$$i_B = I_{BQ} + i_b = 40 + 10\sin\omega t \text{ } \mu A$$

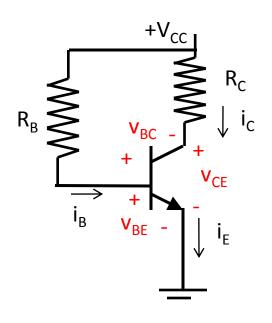
$$v_{CE}$$
 (V)  $v_{CE} = V_{CE} + v_{ce} = 6 - 2\sin \omega t$  V

AC current gain  $h_{fe}$ = $\beta_{ac}$ 

$$h_{fe} = \frac{i_c}{i_h} = \frac{\Delta i_C}{\Delta i_R} = \frac{i_{Cx} - i_{Cy}}{i_{Rx} - i_{Ry}} = \frac{(5 - 3) \times 10^{-3}}{(50 - 30) \times 10^{-6}} = 100$$

# DC biasing of BJT: Fixed Bias

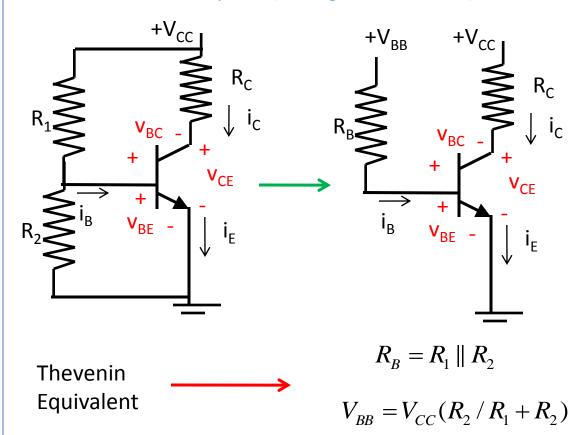
#### Example 1



$$i_{B} = \frac{V_{CC} - v_{BE}}{R_{B}} \qquad i_{C} = h_{FE}i_{B}$$

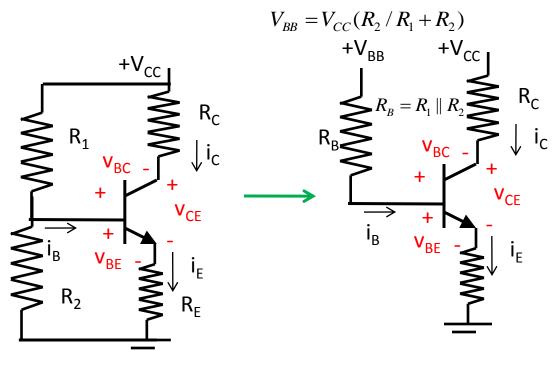
$$v_{CE} = -R_C h_{FE} i_B + V_{CC}$$

Example 2 (Voltage divider bias)



- In fixed bias, variations (temperature, transistor-to-transistor) in  $h_{\text{FE}}$  can affect Q point
  - h<sub>FF</sub> increases, i<sub>C</sub> increases

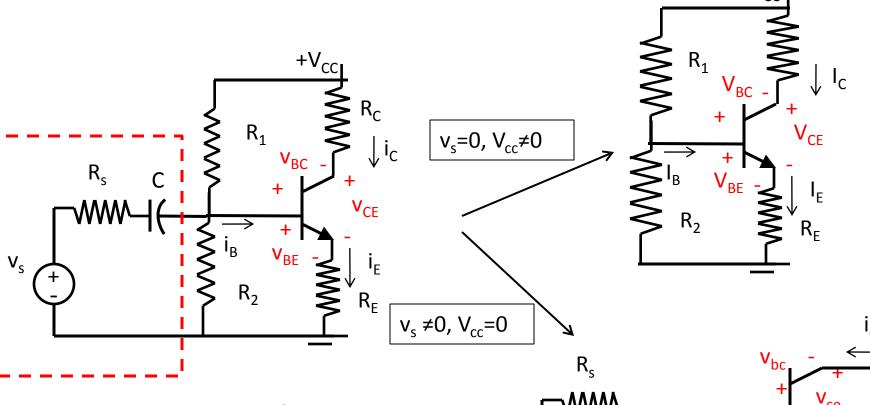
## DC biasing of BJT: Self-Bias



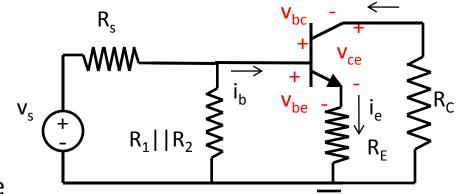
$$V_{BB} = R_B i_B + v_{BE} + R_E (i_B + i_C) \Rightarrow i_B = \frac{V_{BB} - v_{BE}}{R_B + (1 + h_{FE})R_E}$$

- If  $h_{FE}$  increases  $\rightarrow$   $i_C$  tends to increase but  $i_B$  decreases and tends to decrease  $i_C$
- More stable Q point, also called self-biasing circuit

# How to add ac signal to DC bias?

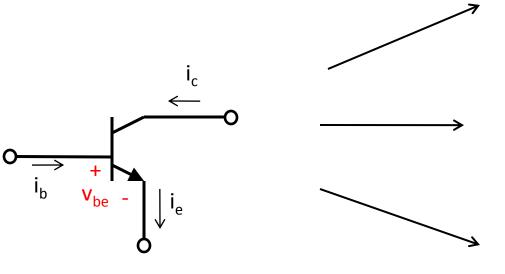


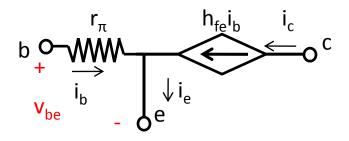
- Coupling capacitor C → open circuit for DC → no effect on dc Q point
- C large enough → short circuit for ac source v<sub>s</sub>
- With active (linear) region biasing → use principle of superposition to analyse

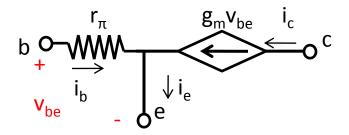


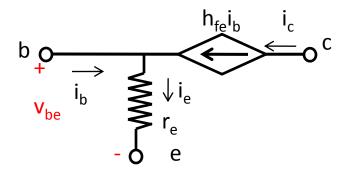
# Small signal ac models

- Ac signal is small such that
  - transistor operates in active mode
  - Non-linear effects/distortions can be ignored
    - o/p characteristics are linear and parallel

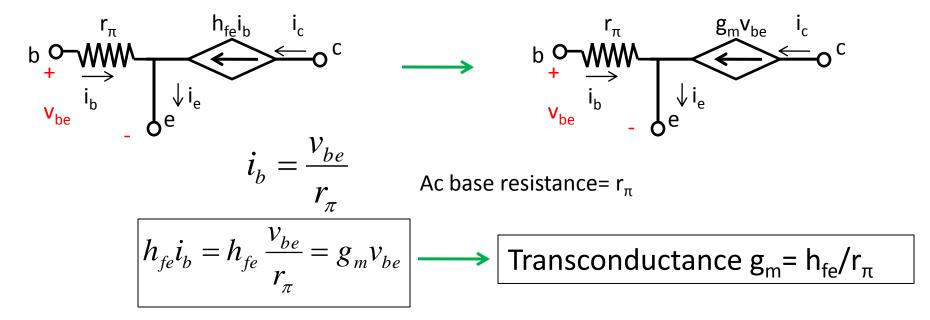








# Small signal model



$$\begin{array}{c|c} b & & & i_c \\ \downarrow & & \downarrow i_e \\ \hline v_{be} & & i_b \\ \hline \end{array}$$

$$r_{e} = \frac{v_{be}}{i_{b} + h_{fe}i_{b}} = \frac{v_{be}}{(1 + h_{fe})i_{b}} = \frac{r_{\pi}}{1 + h_{fe}} \approx \frac{r_{\pi}}{h_{fe}} = \frac{1}{g_{m}}$$

$$r_{\pi} = (1 + h_{fe})r_{e} \approx h_{fe}r_{e}$$

10

## $g_m, r_\pi, r_e$

For npn transistor in active mode

$$i_C = I_S(e^{v_{BE}/V_T} - 1) \approx I_S e^{v_{BE}/V_T}$$

$$di_C = d[I_S(e^{v_{BE}/V_T} - 1)] = \frac{1}{V_T}(I_S e^{v_{BE}/V_T}) dv_{BE} \approx \frac{i_C}{V_T} dv_{BE}$$

$$g_m = \frac{i_c}{v_{be}} = \frac{di_C}{dv_{BE}} = \frac{i_C}{V_T}$$

Biased at  $i_C = I_{CQ}$ ,

$$g_{m} = \frac{I_{CQ}}{V_{T}}$$

$$r_{e} = \frac{1}{g_{m}} = \frac{V_{T}}{I_{CQ}}$$

$$r_{\pi} = h_{fe}r_{e} = \frac{h_{fe}V_{T}}{I_{CQ}}$$

At room temperature,  $V_T=kT/q=0.026 V$