

# Multi Component Carrier, Sub-Band DPD and GNURadio Implementation

Chance Tarver<sup>1</sup>, Mahmoud Abdelaziz<sup>2</sup>, Lauri Kuitila<sup>2</sup>, and Joseph Cavallaro<sup>1</sup>  
Rice University<sup>1</sup> and Tampere University of Technology<sup>2</sup>

## Motivation

- **Spectrum Scarcity → Frequency Agile Standards**
  - Non-contiguous Transmission
  - Carrier Aggregation (CA) in LTE-Advanced
  - Cognitive Radio
  - 5G Cellular
- **Non-contiguous carriers intermodulate**
  - Caused by nonlinearities in power amplifiers (PAs)
  - Undesired spurious emissions (spurs)
  - Could interfere with nearby channels
  - Self-interference to own receiver when using FDD
- **Current 4G chipsets support up to 4 carriers**
  - Snapdragon 835
  - 4x20 MHz carrier aggregation downlink, 2x20 MHz uplink
- *Need efficient way to linearize for this scenario*

## Power Spectral Density

## Related Works

- **Reduce Power**
  - Operate in a more linear PA region
  - Less range and less power efficient
- **Full-Band Digital Predistortion (DPD)**
  - Computationally expensive
  - Does not scale for noncontiguous carriers
  - Requires large sampling rate as carrier spacing grows
- **Sub-Band DPD**
  - Previously explored by the authors with the WARP SDR RF Board
  - Observes and applies DPD to individual spurs
  - Can reduce the necessary sampling rate and complexity
  - Has only been considered for 2 carriers

## Main Idea

- **Learn DPD coefficient,  $\alpha$ , for each spur**
  - Iteratively learn coefficients as necessary using adaptive, LMS algorithm.
  - Apply them as in Equation 5 to reduce spurious emissions.

## Mathematical Model

- **PA Inputs and Outputs:**
  - Parallel Hamerstein baseband PA model
  - Third order, memoryless

$$\textbf{Input: } x(n) = \sum_{i=1}^N x_i(n) e^{2\pi j n \frac{f_i}{f_s}}, \quad (1)$$

$$\textbf{Output: } y(n) = \beta_1 x(n) + \beta_3 |x(n)|^2 x(n). \quad (2)$$

- **IM3 Spurious Signals and Their Locations**

$$y_{IM3:i,j,k}(n) = x_i^*(n) x_j(n) x_k(n), \quad (3)$$

$$f_{IM3:i,j,k} = -f_i + f_j + f_k. \quad (4)$$

- **DPD Processing**

$$\begin{aligned} \tilde{x}(n) = & \sum_{i=1}^N x_i(n) e^{2\pi j n \frac{f_i}{f_s} n} \\ & + \sum_{i=1}^N \sum_{j=1}^N \sum_{k=1}^N \alpha_{i,j,k} x_i^*(n) x_j(n) x_k(n) e^{2\pi j n \frac{-f_i + f_j + f_k}{f_s}}. \end{aligned} \quad (5)$$

## MATLAB Simulation

- **LTE-Advanced CA Scenario**
  - Three, 20MHz CCs, scenario CA 41C-41A
  - Bandpass filter around passband on the RF Frontend
  - Only one spurious region is in violation of the emission limits
- **Result**
  - Rapid convergence of coefficients
  - Suppression below the emission limit in the spurious region

## DPD Coefficient Learning

- **LMS Adaptive Training**

$$\alpha_{i,j,k}(n+1) = \alpha_{i,j,k}(n) - \mu \frac{x_i^*(n) x_j(n) x_k(n) e_{i,j,k}^*(n)}{||x_i^*(n) x_j(n) x_k(n)||} \quad (6)$$

- Decorrelates the error signal,  $e$ , with basis functions from Equation 3 to learn the DPD coefficients,  $\alpha$ .
- Rate is controlled by LMS parameter  $\mu$ .
- Learning is resilient to temperature changes, PA Gain changes, and changes in signal characteristics.

## GNURadio Simulator

## Future Work

- **Main carrier linearization**
- **Hardware testing on the USRP with external PAs**

Can change carrier magnitude and placement, PA model, and DPD application dynamically.