

Problem 1 [100 points] Feedback Linearization

Consider the dynamics of one-link robotic arm, given by

$$\begin{aligned} J_1 \ddot{q}_1 + F_1 \dot{q}_1 + \frac{K}{N} \left(q_2 - \frac{q_1}{N} \right) &= \tau, \\ J_2 \ddot{q}_2 + F_2 \dot{q}_2 + K \left(q_2 - \frac{q_1}{N} \right) &= -mgd \cos q_2, \end{aligned}$$

where q_1, q_2 denote angular positions (in rad), and τ denotes the actuation torque. The parameters $\{J_i, F_i\}_{i=1}^2$ respectively denote the moments of inertia and rotational damping coefficients of the actuator shaft (for $i = 1$) and the link (for $i = 2$), respectively. The parameters K, N, m, d denote the torsional spring constant, the transmission gear ratio, mass of the link, and distance of the center of gravity of the link from a fixed frame of reference, respectively. The **state vector** is $x := (q_1, q_2, \dot{q}_1, \dot{q}_2)^\top \in [0, 2\pi)^2 \times \mathbb{R}^2$ and the **control input** is $u := \tau \in \mathbb{R}$.

(a) [(5+5) + 20 = 30 points] Deciding feedback linearizability

(i) Write the dynamics of the robotic arm in **standard control-affine form** $\dot{x} = f(x) + g(x)u$, i.e., **explicitly write** the drift vector field $f(x)$ and the input vector field $g(x)$, both of size 4×1 , in terms of the state components x_1, x_2, x_3, x_4 and the parameters.

1. We can do some algebra on each equation to isolate the \ddot{q}_1 and \ddot{q}_2 terms on the RHS, and given $\tau = u$, we can express the dynamics as:

$$\dot{x} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} \dot{q}_1 \\ \dot{q}_2 \\ \ddot{q}_1 \\ \ddot{q}_2 \end{bmatrix} = \begin{bmatrix} x_3 \\ x_4 \\ \frac{-F_1 \dot{q}_1 - \frac{K}{N} \left(q_2 - \frac{q_1}{N} \right)}{J_1} \\ \frac{-F_2 \dot{q}_2 - K \left(q_2 - \frac{q_1}{N} \right) - dgm \cos q_2}{J_2} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{J_1} \\ 0 \end{bmatrix} u \quad (1)$$

$$= \begin{bmatrix} x_3 \\ x_4 \\ -\frac{F_1 x_3}{J_1} - \frac{K x_2}{J_1 N} + \frac{K x_1}{J_1 N^2} \\ -\frac{F_2 x_4}{J_2} - \frac{K d m \cos(x_2)}{J_2} - \frac{K x_2}{J_2} + \frac{K x_1}{J_2 N} \end{bmatrix} (f(x)) + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{J_1} \\ 0 \end{bmatrix} (g(x)) u \quad (2)$$

(ii) Determine the region in the state space $[0, 2\pi)^2 \times \mathbb{R}^2$, over which this system is full state static feedback linearizable. (**Hint:** use Lec. 16 p. 11, also Step 1 in p. 12.)

1. From Lec 16 pg 11, condition (i) we check is that the M matrix has full rank. We see that:

$$M = \begin{bmatrix} g(x^0) & adj_{fg}(x^0) & adj_f^2 g(x^0) & adj_f^3 g(x^0) \end{bmatrix} \quad (3)$$

$$= \begin{bmatrix} 0 & -\frac{1}{J_1} & -\frac{F_1}{J_1^2} & -\frac{F_1^2}{J_1^3} - \frac{K}{J_1^2 N^2} \\ 0 & 0 & 0 & -\frac{K}{J_1 J_2 N} \\ \frac{1}{J_1} & \frac{F_1}{J_1^2} & \frac{F_1^2}{J_1^3} + \frac{K}{J_1^2 N^2} & \frac{F_1^3}{J_1^4} + \frac{2F_1 K}{J_1^3 N^2} \\ 0 & 0 & \frac{K}{J_1 J_2 N} & \frac{K(F_1 J_2 + F_2 J_1)}{J_1^2 J_2^2 N} \end{bmatrix} \quad (4)$$

2. $rank(M) = 4 = n$, it is **full rank** over $[0, 2\pi)^2 \times \mathbb{R}^2$ since **no entries are a function of any element of the state**, they are all constant parameters.

3. We also need to check condition (ii) from the same reference, we see that:

$$\begin{bmatrix} g(x^0) & adj_{fg}(x^0) \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad (5)$$

$$\begin{bmatrix} g(x^0) & adj_f^2 g(x^0) \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad (6)$$

$$\begin{bmatrix} adj_{fg}(x^0) & adj_f^2 g(x^0) \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad (7)$$

4. For each of the lie brackets in 3., we see that the lie bracket does **not** add rank, and each 2 pair of $g(x^0), adj_{fg}(x^0), adj_f^2 g(x^0)$ are **not** full rank (they are rank(2)). Combining each pair of $g(x^0), adj_{fg}(x^0), adj_f^2 g(x^0)$ with the pair's lie bracket would still be **rank(2) < rank(3)**.

5. Then by definition the span is closed under Lie bracket, **is involutive** over $[0, 2\pi)^2 \times \mathbb{R}^2$

6. Then from Lec 16 pg 11, we've checked for the 2 constructive conditions, and from 16 pg 12, Step 1, we've verified the **system is feedback linearizable**

(b) [10 + 10 = 20 points] Finding $\lambda(x)$ and r

(i) **Show that** the scalar field $\lambda(x)$ in the Theorem given in Lec. 16 p. 10, for this system, can be taken as $\lambda(x) = x_2$. (**Hint:** use Lec. 16 p. 13, Step 2.)

1. We can see that for $\lambda(x) = x_2$, $\frac{\partial \lambda}{\partial x} = \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix}$

2. We also can see that since $g(x^0)[2] = adj_{fg}(x^0)[2] = adj_f^2 g(x^0)[2] = 0$:

$$\mathcal{L}_g \lambda(x) = \begin{bmatrix} 0 & 0 & \frac{1}{J_1} & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} = 0 \quad (8)$$

$$\mathcal{L}_{adj_{fg}(x^0)} \lambda(x) = \begin{bmatrix} -\frac{1}{J_1} & 0 & \frac{F_1}{J_1^2} & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} = 0 \quad (9)$$

$$\mathcal{L}_{adj_f^2 g(x^0)} \lambda(x) = \begin{bmatrix} -\frac{F_1}{J_1^2} & 0 & \frac{F_1^2}{J_1^3} + \frac{K}{J_1^2 N^2} & \frac{K}{J_1 J_2 N} \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} = 0 \quad (10)$$

3. We can also see that since $adj_f^3 g(x^0)[2] \neq 0$:

$$\mathcal{L}_{adj_f^3 g(x^0)} \lambda(x) = \begin{bmatrix} -\frac{F_1^3}{J_1^4} - \frac{K}{J_1^2 N^2} & -\frac{K}{J_1 J_2 N} & \frac{F_1^3}{J_1^4} + \frac{2F_1 K}{J_1^3 N^2} & \frac{K(F_1 J_2 + F_2 J_1)}{J_1^2 J_2^2 N} \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} = -\frac{K}{J_1 J_2 N} \neq 0 \quad (11)$$

4. Then from Lec 16 pg 13, Step 2, we see that $\lambda(x) = x_2$ **satisfies the desired system of PDEs**.

(ii) **By directly computing the relative degree r** (a positive integer), **prove that** the state equation derived in part (a)(i), augmented with the output equation $y = \lambda(x) = x_2$, indeed has relative degree 4, that is, satisfies the $r = n$ condition in the region determined in part (a)(ii). (**Hint:** use Lec. 15 p. 12 for computing the relative degree)

1. We can show first that for our system, $y = h(x) = \lambda(x) = x_2$, and:

$$\mathcal{L}_g \mathcal{L}_f^0 h = \mathcal{L}_g h = \mathcal{L}_g \lambda(x) = \begin{bmatrix} 0 & 0 & \frac{1}{J_1} & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \\ 0 \end{bmatrix} = 0 \quad (12)$$

$$\mathcal{L}_g \mathcal{L}_f^1 h = \mathcal{L}_g \left\langle \frac{\partial h}{\partial x}, f \right\rangle = \mathcal{L}_g \langle [0 \ 1 \ 0 \ 0], f \rangle = \mathcal{L}_g x_4 = \begin{bmatrix} 0 & 0 & \frac{1}{J_1} & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} = 0 \quad (13)$$

$$\mathcal{L}_g \mathcal{L}_f^2 h = \mathcal{L}_g \left\langle \frac{\partial}{\partial x} \mathcal{L}_f^1 h, f \right\rangle = \mathcal{L}_g \left\langle \frac{\partial}{\partial x} x_4, f \right\rangle = \mathcal{L}_g \left\langle \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix}, f \right\rangle \quad (14)$$

$$= \mathcal{L}_g f[3] = \left\langle \frac{\partial}{\partial x} f[3], g \right\rangle = \left\langle \begin{bmatrix} \frac{K}{J_2 N} & \frac{-K + dgm \sin(x_2)}{J_2} & 0 & -\frac{F_2}{J_2} \end{bmatrix}, \begin{bmatrix} 0 \\ 0 \\ \frac{1}{J_1} \\ 0 \end{bmatrix} \right\rangle = 0 \quad (15)$$

$$\mathcal{L}_g \mathcal{L}_f^3 h = \mathcal{L}_g \left\langle \frac{\partial}{\partial x} \mathcal{L}_f^2 h, f \right\rangle = \mathcal{L}_g \left\langle \frac{\partial}{\partial x} f[3], f \right\rangle \quad (16)$$

$$= \left\langle \frac{\partial}{\partial x} \mathcal{L}_f^3 h, g \right\rangle = \frac{K}{J_1 J_2 N} \neq 0 \quad (17)$$

2. From 1., we see that $\mathcal{L}_g \mathcal{L}_f^0 h = \mathcal{L}_g \mathcal{L}_f^1 h = \mathcal{L}_g \mathcal{L}_f^2 h = 0$, and $\mathcal{L}_g \mathcal{L}_f^3 h \neq 0$.
3. This means t hat it will at $r = 3 + 1, y = \lambda(x) = x_2, y' = \mathcal{L}_f^1 \lambda(x) + \mathcal{L}_g \mathcal{L}_f^3 \lambda(x)u$ has a **nonzero u** coefficient.

4. From 2. and Lec 15 pg 12, we prove that this SISO system has relative degree $3 + 1 = 4 = n$

(c) [(5+5+5+5) + 5 + 25 = 50 points] Normal form and closed loop simulation

(i) Use the Steps 3 and 4 in Lec. 16 p. 14-15 to **derive the feedback linearizing tuple** $(\tau(\cdot), \alpha(\cdot), \beta(\cdot))$. **Also clearly write down** the control system in the feedback linearized coordinates with the new 4×1 state vector z and the new control $v \in \mathbb{R}$, where $z := \tau(x)$ and $u = \alpha(x) + \beta(x)v$.

1. We can derive $\alpha(\cdot)$:

$$\alpha(x) = \frac{-\left\langle \frac{\partial}{\partial x} \mathcal{L}_f^3 h, f \right\rangle}{\left\langle \frac{\partial}{\partial x} \mathcal{L}_f^3 h, g \right\rangle} \quad (18)$$

$$= F_1 x_3 + \frac{F_2^3 J_1 N x_4}{J_2^2 K} + \frac{F_2^2 G J_1 N d m \cos(x_2)}{J_2^2 K} + \frac{F_2^2 J_1 N x_2}{J_2^2} - \frac{F_2^2 J_1 x_1}{J_2^2} \quad (19)$$

$$+ \frac{2F_2 G J_1 N d m x_4 \sin(x_2)}{J_2 K} - \frac{2F_2 J_1 N x_4}{J_2} + \frac{F_2 J_1 x_3}{J_2} \quad (20)$$

$$+ \frac{G^2 J_1 N d^2 m^2 \sin(x_2) \cos(x_2)}{J_2 K} - \frac{G J_1 N d m x_4^2(t) \cos(x_2)}{J_2} \quad (21)$$

$$+ \frac{G J_1 N d m x_2 \sin(x_2)}{J_2} - \frac{G J_1 N d m \cos(x_2)}{J_2} - \frac{G J_1 d m x_1 \sin(x_2)}{J_2} \quad (22)$$

$$- \frac{J_1 K N x_2}{J_2} + \frac{J_1 K x_1}{J_2} + \frac{K x_2}{N} - \frac{K x_1}{N^2} \quad (23)$$

2. We can derive $\beta(\cdot)$:

$$\beta(x) = \frac{1}{\mathcal{L}_g \mathcal{L}_f^3 \lambda(x)} = \frac{J_1 J_2 N}{K} \quad (24)$$

3. We can derive $\tau(\cdot)$:

$$z = \tau(x) = \begin{pmatrix} \lambda(x) \\ \mathcal{L}_f \lambda(x) \\ \mathcal{L}_f^2 \lambda(x) \\ \mathcal{L}_f^3 \lambda(x) \end{pmatrix} \quad (25)$$

$$\begin{bmatrix} z_1 \\ z_2 \\ z_3 \\ z_4 \end{bmatrix} = \begin{bmatrix} x_2 \\ x_4 - \frac{F_2 x_4}{J_2} - \frac{G d m \cos(x_2)}{J_2} - \frac{K x_2}{J_2} + \frac{K x_1}{J_2 N} \\ F_2 \left(-\frac{F_2 x_4}{J_2} - \frac{G d m \cos(x_2)}{J_2} - \frac{K x_2}{J_2} + \frac{K x_1}{J_2 N} \right) \\ -\frac{F_2 \left(-\frac{F_2 x_4}{J_2} - \frac{G d m \cos(x_2)}{J_2} - \frac{K x_2}{J_2} + \frac{K x_1}{J_2 N} \right)}{J_2} + \left(\frac{G d m \sin(x_2)}{J_2} - \frac{K}{J_2} \right) x_4 + \frac{K x_3}{J_2 N} \end{bmatrix} \quad (26)$$

4. Finally, for the new control system in new coordinates:

$$\begin{pmatrix} \dot{z}_1 \\ \dot{z}_2 \\ \dot{z}_3 \\ \dot{z}_4 \end{pmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{pmatrix} z_1 \\ z_2 \\ z_3 \\ z_4 \end{pmatrix} + \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} v \quad (27)$$

(ii) Give physical interpretation of the components of the vector z (think of their units).

1. We see that for the elements of $z = \tau(x)$
2. z_1 is joint position
3. z_2 is velocity of that joint
4. z_3 is the acceleration of that joint, as we derived in (a)(i) for \ddot{q}_2
5. z_4 is the Lie derivative of $z_3 = \ddot{q}_2$ along f , intuitively it is how the acceleration term $z_3 = \ddot{q}_2$ changes as we move along the system dynamics vector field f .

(iii) **Use a simple pole placement controller** $v = k^\top z$ where $k := (k_1, k_2, k_3, k_4)^\top \in \mathbb{R}^4$ to stabilize the feedback linearized states to the origin. (in $\pi/6, \pi/3, 1, 2$). (**Hint:** recall that over the region in which the system is feedback linearizable, the map τ must be a diffeomorphism, and $x = \tau^{-1}(z)$)

1. We can define $x = \tau^{-1}(z)$:

$$\tau^{-1}(z) = \begin{pmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{pmatrix} = \begin{bmatrix} \frac{N(F_2 z_2 + J_2 z_3 + K z_1 + dgm \cos(z_1))}{K} \\ z_1 \\ \frac{N(F_2 z_3 + J_2 z_4 + K z_2 - dgm z_2 \sin(z_1))}{K} \\ z_2 \end{bmatrix} \quad (28)$$

2. We can use τ to translate the initial state in $x(t=0)$ to $z(t=0)$, and then use the z coordinate closed-loop dynamics ($v = k^\top z$) to integrate out $z(t)$:

$$\begin{pmatrix} \dot{z}_1 \\ \dot{z}_2 \\ \dot{z}_3 \\ \dot{z}_4 \end{pmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{pmatrix} z_1 \\ z_2 \\ z_3 \\ z_4 \end{pmatrix} + \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} (k_1 \ k_2 \ k_3 \ k_4) \begin{pmatrix} z_1 \\ z_2 \\ z_3 \\ z_4 \end{pmatrix} \quad (29)$$

$$= \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ k_1 & k_2 & k_3 & k_4 \end{bmatrix} \begin{pmatrix} z_1 \\ z_2 \\ z_3 \\ z_4 \end{pmatrix} \quad (30)$$

3. Using the matrix in 2. (the companion matrix), we can see that if we want the eigenvalues to be at -1, -2, -3, -4, then we set the k values = the negative characteristic polynomial coefficients: $k = [-24, -50, -35, -10]$, and we can plot a simulator output as below, one simulating by z , the other using u and original system dynamics (as expected they are equivalent). Derivation in **derivation_1c.py**, simulation in **1c.py**.

