

Limits on which specs can actually be achieved

- plant dynamics
- control effort by actuators

## Calculating $e_{ss}$

### Final value theorem (FVT)

If a signal  $Y(s)$  has all of its poles in the OLHP, except possibly a single pole at  $s=0$ , then

$$y_{ss} = \lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} sY(s).$$

Proof.  $L(\dot{y}(t)) = \int_0^{\infty} \dot{y}(t) e^{-st} dt = sY(s) - y(0)$

$$\Rightarrow \lim_{s \rightarrow 0} \int_0^{\infty} \dot{y}(t) e^{-st} dt = \lim_{s \rightarrow 0} sY(s) - y(0)$$

$$\int_0^{\infty} \dot{y}(t) \lim_{s \rightarrow 0} e^{-st} dt = \int_0^{\infty} \dot{y}(t) dt = \left[ y(t) \right]_0^{\infty} = \lim_{t \rightarrow \infty} y(t) - y(0) = y_{ss} - y(0)$$

only well-defined if all poles of  $Y(s)$  are in the OLHP, except possibly one at  $s=0$

$$\Rightarrow y_{ss} = \lim_{s \rightarrow 0} sY(s).$$

$$e(t) = r(t) - y(t)$$

Corollary. If a system  $H(s)$  is stable and its input is  $u(t) = 1(t)$

then its steady-state output is:  $y_{ss} = H(0)$ .  $H(0)$  is sometimes called the DC gain of the system.

Proof.  $Y(s) = H(s)U(s) = H(s)\frac{1}{s}$  has all of its poles in the OLHP (since  $H(s)$  is stable) except for exactly one pole at  $s=0$ .  
 $\Rightarrow$  by FVT,  $y_{ss} = \lim_{s \rightarrow 0} sY(s) = \lim_{s \rightarrow 0} sH(s)\frac{1}{s} = \lim_{s \rightarrow 0} H(s) = H(0)$

$$\text{Ex. } H(s) = \frac{K}{sT+1} \Rightarrow y_{ss} = \frac{K}{0+1} = K$$

$$\text{Ex. } H(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \Rightarrow y_{ss} = 1$$

$$\text{Ex. } H(s) = 18 \frac{(s+1)(s+2)}{(s+3)(s+4)} \Rightarrow y_{ss} = 18 \frac{(1)(2)}{(3)(4)} = 3$$

$$\text{Ex. } H(s) = \int \frac{s+1}{s-2} \Rightarrow y_{ss} \text{ is undefined (or } \infty) \text{ unstable!}$$

For a stable system  $H(s)$ ,  $e_{ss} = 1 - y_{ss} = 1 - H(0)$ .

## dynamic responses of LTI systems

- no calculators
- no cheat sheets
- no notes