Linear Dynamic Systems Notes

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19 August 2011

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Following notes based on the EE263 Course Reader. Orgmode → LaTeX 2_ε

Matlab files Linear Algebra
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Official Lecture Notes

1.1 Interpretation for $y = Ax$

- A a transformation matrix
- y an observed measurement, x an unknown
- x is input, y is output; for rows i and columns j in A, $\mathbf{i} \to \mathbf{y}$, $\mathbf{j} \to \mathbf{x}$
  - indexed as **output, input**
- Lower diagonal Matrix should make you expect or have a vague thought about causality.
  - $a_{ij} = 0$ for $i < j \Rightarrow y_i$ only depends on $x_1, ..., x_i$
  - A is diagonal; output only depends on input
- Sparcity pattern (block of zeroes) in a matrix should have you wonder why... usually not coincidental
- A_{35} positive $\Rightarrow x_5$ increased corresponds to an increase in 3rd output, y_3

1.2 Example in notes

1.2.1 Linear elastic structure

- a_{11} probably positive
  - x_1 input gives positive push to y_1 output

1.2.2 Force/Torque on rigid body

- Net torque/force on body is linearly related to force inputs

1.2.3 Thermal System

- despite normally needing a Poisson equation, the steady state heat of the different locations can be represented as a linear system $Ax = y$

2 Lecture 3

2.1 Review of Linearization (affine approximation)

1. If $f : \mathbf{R}^n \to \mathbf{R}^m$ is differentiable at $x_0 \in \mathbf{R}^n$, then x near $x_0 \Rightarrow f(x)$ very near $f(x_0) + Df(x_0)(x - x_0)$ where
$$Df(x_0)_{ij} = \left.\frac{\partial f_i}{\partial x_j}\right|_{x_0}$$
is the derivative (Jacobian) matrix.
2. with $y = f(x), y_0 = f(x_0)$, define 'input deviation' $\delta x := x - x_0$, 'output deviation' $\delta y := y - y_0$
3. then we have $\delta y \approx Df(x_0)\delta x$
   i.e., we get a linear function for looking at how the output changes with small changes in the input
4. When deviations are small, they are approximately related by a linear function

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2.1.1 Good intuition range-finding problem

2.2 Intepretations of $Ax = y$

- Scaled combination of columns in A

  $A = [a_1 a_2 ... a_n] \rightarrow y = x_1 a_1 + x_2 a_2 + ... + x_n a_n$

- Looking at the rows: the multiplication is the inner product of rows and vector x
- I/O ordering is backwards in control; in A_{ij}, j refers to input and i refers to output
  - indexing is likewise backwards in block diagram indexing

$AB = I$; $\tilde{a}_i^T \cdot b_j = 0$ if $i \neq j$, where $\tilde{a}_i$ is the ith row in A, and b_j is jth column in B

2.2.1 Computing $C = AB$

How to compute this faster than using formula $c_{ij} = \sum_{k=1} A_{ik} B_{kj}$: have computer break it up into submatrices and do the multiplication on them

2.2.2 Zero nullspace (in notes)

- if 0 is only element in $\mathcal{N}(A)$
  - A is one-to-one
    - linear transformation doesn't lose information
  - columns of A are independent, and basis for their span
  - A has a left inverse $\Rightarrow$ you can use this to undo the transformation and find the input x, given the output y
  - $|A^T A| \neq 0$

2.2.3 Interpretations of nullspace

supposing $z \in \mathcal{N}(A)$

- Measurements
  - z is undetectable from sensors
  - x and $x + z$ are indistinguishable from sensors: $Ax = A(x + z)$
  - the nullspace characterizes ambiguity in x from measurement $y = Ax$
    - large nullspace is bad
- $y = Ax$ is output from input x
  - z is input with no result
  - x and $x + z$ have same result
  - the nullspace characterizes freedom of input choice for given result
    - large nullspace is good: more room for optimization because of more input possibilities

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2.2.4 Range

Range of $A \in \mathbb{R}^{m \times n}$ defined as $\mathcal{R}(A) = \{Ax | x \in \mathbb{R}^n\} \subseteq \mathbb{R}^m$

- The set of vectors that can be 'hit' by linear mapping $y = Ax$
- span of columns of A
- set of vectors y for which $Ax = y$ has a solution
- possible sensor measurement
  - in design, you'll want to throw an exception if a measurement is outside the range; the sensor is bad
  - test for a bad 13th sensor: remove 13th row in A; if the reduced y is in the range of the reduced matrix, the 13th sensor might not have been bad
- Onto matrices
  A is 'onto' if $\mathcal{R}(A) = \mathbb{R}^m$
  - you can solve $Ax = y$ for any y
  - columns of A span $\mathbb{R}^m$
  - A has a right inverse B s.t. $AB = I$
    - can do $ABy = A(By) = y$: you want an x that gives you y? Here it is.
    - Design procedure
  - rows of A are independent
    - a.k.a., $\mathcal{N}(A^T) = \{0\}$
  - $|AA^T| \neq 0$
- Interpretations of range
  - supposing $v \in \mathcal{R}(A)$
    - v reachable
    - else, not reachable
- Inverse
  Note: square matrices are impractical for engineering. They don't let you take advantoge of redundant sensors/controllers, or let you build a robust system to take care of broken sensors
  - $A \in \mathbb{R}^{n \times n}$ is invertible or nonsingular if det $A \neq 0$
    - columns of A are basis for $\mathbb{R}^n$
    - rows of A are basis for $\mathbb{R}^n$
    - $y = Ax$ has a unique solution x for every $y \in \mathbb{R}^n$
    - A has left and right inverse $A^{-1} \in \mathbb{R}^{n \times n}$, s.t. $AA^{-1} = A^{-1}A = I$
    - $\mathcal{N}(A) = \{0\}$
    - $\mathcal{R}(A) = \mathbb{R}^n$
    - det $A^T A = |AA^T| \neq 0$
  - Dual basis intepretation of inverse
    a_i are columns of A, and $\tilde{b}_i^T$ are rows of $B = A^{-1}$
    - $y = Ax$, column by column, looks like $y = x_1 a_1 + ... + x_n a_n$
      - multiply both sides of $y = Ax$ by $A^{-1} = B$ gives $x = By$
      - so $x_i = \tilde{b}_i^T y$

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$$\begin{bmatrix} \vdots & \vdots & \vdots & \vdots \\ a_1 & a_2 & ... & a_n \\ \vdots & \vdots & \vdots & \vdots \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} = \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_n \end{bmatrix}$$

$$x = A^{-1}y$$

$$\begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} = \begin{bmatrix} \cdots & \tilde{b}_1^T & \cdots \\ \cdots & \tilde{b}_2^T & \cdots \\ \cdots & \vdots & \cdots \\ \cdots & \tilde{b}_n^T & \cdots \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_n \end{bmatrix}$$

$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_n \end{bmatrix} = \begin{bmatrix} x_1 a_1 + ... + x_n a_n \end{bmatrix} = \begin{bmatrix} (\tilde{b}_1^T y) a_1 + ... + (\tilde{b}_n^T y) a_n \end{bmatrix}$$

Beautiful thing:

$$y = \sum_{i=1}^{n} (\tilde{b}_i^T y) a_i$$

2.2.5 Rank of matrix

Rank of $A \in \mathbb{R}^{m \times n}$ as $\mathbf{rank}(A) = \mathbf{dim}\mathcal{R}(A)$

- $\mathbf{rank}(A) = \mathbf{rank}(A^T)$
- $\mathbf{rank}(A)$ is maximum number of independent columns or rows of A: $\mathbf{rank}(A) \leq \mathbf{min}(m, n)$
- $\mathbf{rank}(A) + \mathbf{dim}\mathcal{N}(A) = n$
- Conservation of degrees of freedom (dimension)
  - $\mathbf{rank}(A)$ is dimension of set 'hit' by mapping $y = Ax$
  - $\mathbf{dim}\mathcal{N}(A)$ is dimension of set of x 'crushed' to zero by $y = Ax$
  - Example
    - $A \in \mathbb{R}^{20 \times 10} \mathbf{rank}(A) = 8$
      - you can do 8 dimensions worth of stuff
      - 10 knobs, 2 redundant knobs, which is $\mathbf{dim}\mathcal{N}(A) = 2$
- Coding interpretation of rank
  - rank of product: $\mathbf{rank}(BC) \leq \mathbf{min}\{\mathbf{rank}(B), \mathbf{rank}(C)\}$
  - supposedly really cool stuff based on this
  - low rank matrices let you do fast computations

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2.2.6 Various wrap-up items

- RMS

$$\mathbf{rms}(x) = \left(\frac{1}{n}\sum_{i=1}^{n}\right)^{1/2} = \frac{\|x\|}{\sqrt{n}}$$

- Inner product
  $\langle x, y\rangle := x_1y_1 + x_2y_2 + \cdots + x_ny_n = x^Ty$
  - intepretation of inner product signs:
  - $x^Ty > 0$: acute; roughly point in same direction
  - $x^Ty > 0$: obtuse; roughly point in opposite direction
- Orthonormal set of vectors
  - set of k vectors $u_1, u_2, ..., u_k \in \mathbb{R}^n$ orthonormal; $U = [u_1 \cdots u_k]$
  - $U^TU = I_k \leftrightarrow$ set of column vectors of U are orthonormal
  - **warning**: $UU^T \neq I_k$ if $k < n$
    - say U is 10×3, U^T is 3×10, rank of U is $3 \Rightarrow$ rank of UU^T is at most 3
    - but UU^T will be a 10×10 matrix, so it can't be the identity matrix

3 Lecture 5

A good source for more on orthogonality at University of Minnesota

3.1 Geometric properties of orthonormal vectors

- columns of U are ON $\Rightarrow$ mapping under U preserves distances
  - $w = Uz \Rightarrow \|w\| = \|z\|$
- Also preserves inner product
- Also preserves angles
- Something like a rigid transformation

3.2 Orthonormal basis for $\mathbb{R}^n$

- if there are n orthonormal vectors (remember, with dimension n), it forms an orthonormal basis for $\mathbb{R}^n$
- $U^{-1} = U^T$
  - $\boxed{U^TU = I \Leftrightarrow U\text{'s column vectors form an orthonormal basis for } \mathbb{R}^n}$
  - $\sum_{i=1}^{n} u_iu_i^T = I \in \mathbb{R}^{n\times n}$ (known as a dyad, or outer product; inner products reverses the two and gives a scalar, outer gives a matrix)
  - outer products take 2 vectors, possibly of different sizes, and multiplies every combination of elements one with another

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3.3 Expansion in orthonormal basis

- U orthogonal $\Rightarrow x = UU^T$
- $x = \sum_{i=1}^{n} \left(u_i^T x\right) u_i$
  - because $U^T U = I$, the thing in sum is really $u_i u_i^T x$
  - $u_i^T x$ is really a scalar, so this can be moved to the front of u_i, giving our result
  - This says x is a linear combination of u_i's

3.4 Gram-Schmidt procedure

- $a_1, ..., a_k \in \mathbb{R}^n$ are LI; G-S finds ON vectors $q_1, ..., q_k$ s.t.

$$\mathbf{span}(a_1, ..., a_r) = \mathbf{span}(q_1, ..., q_r)$$

  for $r \leq k$
- so $q_1, ..., q_r$ is an ON basis for $\text{span}(a_1, ..., a_r)$
- Basic method: orthogonalize each vector wrt the previous ones, then normalize result
  1. $\tilde{q}_1 = a_1$
  2. normalize: $q_1 = \tilde{q}_1 / \|\tilde{q}_1\|$
  3. remove q_1 component from a_2: $\tilde{q}_2 = a_2 - (q_1^T a_2) q_1$
  4. normalize q_2
  5. remove q_1, q_2 components: $\tilde{q}_3 = a_3 - (q_1^T a_3) q_1 - (q_2^T a_3) q_2$
  6. normalize q_3
- $a_i = (q_1^T a_i) q_1 + (q_2^T a_i) q_2 + \cdots + (q_{i-1}^T a_i) q_{i-1} + \|\tilde{q}_i\| q_i$
  - $= r_{1i} q_1 + r_{2i} q_2 + \cdots + r_{ii} q_i$ ($r_{ii} \geq 0$ is the length of $\tilde{q}_i$)

3.5 QR decomposition

This can be written as $A = QR$, where $A \in \mathbb{R}^{n \times k}, Q \in \mathbb{R}^{n \times k}, R \in \mathbb{R}^{k \times k}$

$$\begin{bmatrix} a_1 & a_2 & \cdots & a_k \end{bmatrix} = \begin{bmatrix} q_1 & q_2 & \cdots & q_k \end{bmatrix} \begin{bmatrix} r_{11} & r_{12} & \cdots & r_{1k} \\ 0 & r_{22} & \cdots & r_{2k} \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & r_{kk} \end{bmatrix}$$

- R triangular because computation of a_i only involves up to q_i
  - a sort of causality, since you can calculate q_7 without seeing q_8
- Columns of Q are ON basis for $\mathcal{R}(A)$

3.6 General Gram Schmidt procedure ('rank revealing QR algorithm')

- Basically the same, but if one of the $\tilde{q}_i$'s is zero (meaning a_i is dependent on previous a vectors), then just go to the next column
- referring to notes, upper staircase notation shows which vectors are dependent on previous ones (columns without the x's)
  - entries with x are 'corner' entries

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3.7 Applications

- check if $b \in \mathbf{span}(a_1, a2, ..., a_k)$
- Factorize matrix A

3.8 Least Squares Approximation

- Overdetermined linear equation (tall, skinny, more equations than unknowns, dimensionally redundant system of equations)

4 Lecture 6

On skinny, full rank matrices

4.1 Overdetermined equations

- Skinny, more equations than unknowns
- Given $y = Ax, A \in \mathbb{R}^{m \times n}$, a randomly-chosen y in $\mathbb{R}^m$ has 0 probability of being in the range of A
- To *approximately* solve for y, minimize norm of error (residual) $r = Ax - y$
- find $x = x_{ls}$ (least squares approx.) that minimizes $\|r\|$

4.2 Least Squares 'Solution'

- square $\|r\|$, get expansion, set gradient wrt x equal to zero
- $\boxed{x_{ls} = (A^T A)^{-1} A^T y} = B_{ls} y$ (linear operation)
- $A^T A$ should be invertible, square, full rank
- $(A^T A)^{-1} A^T$ is a generalized inverse (is only inverse for square matrices, though)
  - Also known as the $A^\dagger$, 'pseudo-inverse'
  - Which is a left inverse of A

4.3 Projection on $\mathcal{R}(A)$

Ax_{ls} is the point closest to y (i.e., projection of y onto $\mathcal{R}(A)$)

- $Ax_{ls} = \mathbf{proj}_{\mathcal{R}(A)}(y) = \left(A(A^T A)^{-1} A^T\right) y$

4.4 Orthogonality principle

The optimal residual is orthogonal to $C(A)$

- $r = Ax_{ls} - y = (A(A^T A)^{-1} A^T - I)y$ orthogonal to $C(A)$
- $\langle r, Az \rangle = y^T (A(A^T A)^{-1} A^T - I)^T Az = 0$ for all $z \in \mathbb{R}^n$

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4.5 Least-squares via QR factorization

A is still skinny, full rank

- Factor as $A = QR$; $Q^TQ = I_n, R \in \mathbb{R}^{n\times n}$ upper triangular, invertible
- pseudo-inverse: $(A^TA)^{-1}A^T = R^{-1}Q^T \Rightarrow \boxed{R^{-1}Q^Ty = x_{ls}}$
- Pretty straight-forward
- Matlab for least squares approximation

```
xl = inv(A' * A)*A'y; # So common that has shorthand in MATLAB
xl = A\y;                 # Works for non-skinny matrices, may do unexpected things
```

4.6 Full QR factorization

- $A = \begin{bmatrix} Q_1 & Q_2 \end{bmatrix} \begin{bmatrix} R_1 \\ 0 \end{bmatrix}$
  - New Q is square, orthogonal matrix; R_1 is square, upper triangular, invertible
- Remember, multiplying by orthogonal matrix doesn't changet the norm:
  - $\|Ax - y\|^2 = \|R_1x - Q_1^Ty\|^2 + \|Q_2^Ty\|^2$
  - Find least squares approximation with $x_{ls} = R_1^{-1}Q_1^Ty$ (zeroes first term)

4.7 Applications for least squares approximations

- if there is some noise v in $y = Ax + v$
  - you can't reconstruct x, but you can get close with the approximation
- Estimation: choose some $\hat{x}$ that minimizes $\|A\hat{x} - y\|$, which is the deviation between the think we observed, and what we would have observed in the absence of noise

4.8 BLUE: Best linear unbiased estimator

- A still full rank and skinny; have a 'linear estimator' $\hat{x} = By$ (B is fat)
  - $\hat{x} = B(Ax + v)$
- Called unbiased if there is no estimation error when there's no noise; the estimator works perfectly in the absence of noise
  - if $v = 0$ and $BA = I$; B is left inverse/perfect reconstructor
- Estimation error uf unbiased linear estimator is $x - \hat{x} = sBv$, so we want B to be small and $BA = I$; small means error isn't sensitive to the noise
- The pseudo-inverse is the smallest left inverse of A:
  - $A^\dagger = (A^TA)^{-1}A^T$
  - $\sum_{i,j} B_{ij}^2 \geq \sum_{i,j} A_{ij}^{\dagger 2}$

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4.9 Range-finding example

- Find ranges to 4 beacons from an unknown position x
- $y = -\begin{bmatrix} k_1^T \\ k_2^T \\ k_3^T \\ k_4^T \end{bmatrix} x + v$
- actual position $x = (5.59, 10.58)$; measurement $y = (-11.95, -2.84, -9.81, 2.81)$
  - these numbers aren't consistent in $Ax = y$, since there's also the error; there is no such x value that can give this y value
- There are 2 redundant sensors (2 more y values than x values); one method for estimating $\hat{x}$ is 'just enough' method: you only need 2 y values; take inverse of top half of A and pad the rest of the matrix with 0's
- use $\hat{x} = B_{justenough}y = \begin{bmatrix} \begin{bmatrix} k_1^T \\ k_2^T \end{bmatrix}^{-1} 0 & 0 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} 0 & -1.0 & 0 & 0 \\ -1.12 & .5 & 0 & 0 \end{bmatrix} y = \begin{bmatrix} 2.84 \\ 11.9 \end{bmatrix}$
- Least Squares method: $\hat{x}A^\dagger y =$ this has a much smaller norm of error
- Just enough estimator doesn't seem to have good performance... unless last two measurements were really off, since JEM only takes 2 measurements into account

4.10 Quantizer example

Super-impressive least squares estimate; more precise than A-D converter

4.11 Least Squares data-fitting

- use functions $f_1, f_2, ..., f_n : S \to \mathbb{R}$ are called regressors or basis functions
- applications
  - interpolation- , extrapolation, smoothing of data

Applications	
interpolation	don't have sensors in specific location, but want the temperature
extrapolation	get good basis functions for better interpolation
data smoothing	de-noise measurements
simple, approximate data model	Get a million samples, use the data-fitting to get a simple approximate function

4.12 Least-squares polynomial fitting

- Vandermonde matrix?

5 Lecture 7

5.1 Least-squares polynomial fitting, cont'd

- have data samples $(t_i, y_i), i = 1, ..., m$
- fit coefficients a_i of polynomial $p(t) = a_0 + a_1 t + \cdots + a_{n-1}t^{t-1}$ so that when evaluated at t_i it will give you the associated y value
- basis functions are $f_j(t) = t^{j-1}, j = 1, ..., n$

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- use Vandermonde matrix A ('polynomial evaluator matrix'):

$$A = \begin{bmatrix} 1 & t_1 & t_1^2 & ... & t_1^{n-1} \\ 1 & t_2 & t_2^2 & ... & t_2^{n-1} \\ & \vdots & & & \vdots \\ 1 & t_m & t_m^2 & ... & t_m^{n-1} \end{bmatrix}$$

- side note: use this when you want to fit throughout an interval, use a Taylor series fit if you want it close to a point

5.2 Growing sets of regressors

- Given ordered set of vectors; find best fit with first vector, then best fit with first and second, then best fit with first three...
- These vectors called *regressors*, or columns
- Say you have some *master list* A with n columns, and $A^{(p)}$ will be the matrix with the first p columns of it
  - we want to minimize different sets of $\|A^{(p)}x - y\|$
  - i.e., project y onto a growing span $\{a_1, a_2, ..., a_p\}$
- Solution for each $p \leq n$ given by $x_{ls}^{(p)} = (A_p^T A_p)^{-1} A_p^T y = R_p^{-1} Q_p^T y$
  - In MATLAB, `A(:,1:p)\y`, though technically it's faster to do a sort of `for` loop
- Residual, $\| \sum_{i=1}^{p} x_i a_i - y\|$ reduces as p (number of columns) increases
  - though it may be same as residual with previous value of p if the optimal $x_1 = 0$, when $y \perp a_1$
  - if the residual drops 15% from that of previous value of p, you say that a_1 explains 15% of y

5.3 Least-squares system identification (important topic)

- measure input, output $u(t), y(t)$ for $t = 0, ..., N$ of unknown system, and try to get a model of system
- example: moving average (MA) model with n delays (try to approximate what are the weights h_i for each delay)
  - see equation/matrix in notes, though there are different ways to write it
  - get best answer with LSA

5.4 Model order selection

- how large should n be?
- the larger, the smaller prediction error on *data used to form model*
- but at a certain point, predictive ability of model on other I/O data from same system worsens
- probably best to choose the 'knee' on the graph on notes slide for prediction of new data

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5.4.1 Cross-validation

- check with new data, only if you're getting small residuals on data you've already seen
- when n gets too large (greater than $n = 10$ on graph), the error with 'validation data' actually gets larger
- this example is ideal, since $n = 10$ is the obvious order for the model
- **Application note**: in medical, many industries, there's a firm wall between validation data and model-developing data, so someone *else* tests your model
- in this example, it is known as *overfit* when the validation data error gets larger for n too large

5.5 Growing sets of measurements

- similar to GSo Regressors, except you add new rows, not columns
- this would happen if we're estimating a parameter x (which is constant)
- Solution: $x_{ls} = \left(\sum_{i=1}^{m} a_i a_i^T\right)^{-1} \sum_{i=1}^{m} y_i a_i$
- new way to think of least squares equation

5.6 Recursive ways to do least squares

- don't have to re-add for each new measurement
  - i.e., memory is bounded
  - use equation from notes; solution is $x_{ls}(m) = P(m)^{-1}q(m)$

5.7 Fast update algorithm for recursive LS

- Was a big deal back in the day; somewhat still

5.8 Multi-objective least squares

- Sometimes you have 2+ objectives to minimize
  - say $J_1 = \|Ax - y\|^2$ (what we've done so far)
  - and $J_2 = \|Fx - g\|^2$
  - these are usually competing (minimize one at cost of other)
- Variable in question is $x \in \mathbb{R}^n$
- Plot in notes shows plot of $(J_1(x_i), J_2(x_i))$
- Some points are unambiguously worse than others, but there is some ambiguity when $J_1(x_1) < J_1(x_2)$, while $J_2(x_1) > J_2(x_2)$
- Fix this ambiguity with 'weighted-sum objective'
- $J_1 + \mu J_2 = \|Ax - y\|^2 + \mu\|Fx - g\|^2$
  - Say, there's a trade-off between smoothness (no noise) and better fit; μ can have different dimensions if J_2 does
- Use slope of μ in graph ('indifference curve', in economics) [slide 7-6]

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6 Lecture 8

Multi-objective least-squares

6.1 Plot of achievable objective pairs

- if it approximates an L shape (has a 'knee'), the knee is usually the obvious optimal location, so least-squares isn't as helpful
  - optimal point isn't very sensitive to μ
- Other extreme: trade-off curve looks linear (negative slope), where it's zero-sum
  - optimal point very sensitive to μ
  - slope commonly called *exchange rate curve*
- In this class, they must be convex curves (cup up/outward)
- To find Pareto optimal points, minimize $J_1 + \mu J_2 = \alpha$
  - on plot, can have level curves with slope μ
  - Find point on Pareto Optimal Curve that has slope μ

6.2 Minimizing weighted-sum objective

- note: norm-squared of a stacked vector is norm-square of the top+norm-square of bottom

$$J_1 + \mu J_2 = \|Ax - y\|^2 + \mu\|Fx - g\|^2 = \left\| \begin{bmatrix} A \\ \sqrt{\mu}F \end{bmatrix} x - \begin{bmatrix} y \\ \sqrt{\mu}g \end{bmatrix} \right\|^2$$
$$= \left\| \tilde{A}x - \tilde{y} \right\|$$

where

$$\tilde{A} = \begin{bmatrix} A \\ \sqrt{\mu}F \end{bmatrix}, \tilde{y} = \begin{bmatrix} y \\ \sqrt{\mu}g \end{bmatrix}$$

If $\tilde{A}$ is full rank,

$$x = \left(\tilde{A}^T\tilde{A}\right)^{-1}\tilde{A}^T\tilde{y} \quad (1)$$
$$= (A^TA + \mu F^TF)^{-1}(A^Ty + \mu F^Tg) \quad (2)$$

Note: to plot the tradeoff curve, calculate the minimizer x_μ, and plot the resulting pairs (J_1, J_2) In MATLAB, `[A; sqrt(mu) * F]\[y;sqrt(mu) * g]`

6.3 Example: frictionless table

- y is final position at $t = 10$; $y = a^Tx$, $a \in \mathbb{R}^{10}$
- $J_1 = (y-1)^2$, (final position difference from $y = 1$ squared)
- $J_2 = \|x\|^2$ sum of force squares
- Q: Why do we often care about sum of squares? A: **It's easy to analyze** (not necessarily because it corresponds to energy)
  - max $|x_i|$ corresponds to maximum thrust
  - $\sum |x_i|$ corresponds to fuel use
- Optimal tradeoff curve is quadratic

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6.4 Regularized least-squares

- famous example of multi-objective least squares
  - second J term is simply $J_2 = \|x\|$, though first is the same: $J_1 = \|Ax - y\|^2$
- Tychonov regularization works for *any* A
  - *regularized* least-squares solution: $\boxed{x_\mu = (A^TA + \mu I)^{-1}A^Ty}$, for $F = I, g = 0$

Show $(A^TA+\mu I)$ is invertible, no matter what size/values of A (assuming $\mu > 0$): If this is *not* invertible (singular), it means some nonzero vector z gets mapped to zero ($z \in \mathcal{N}(A)$)

$$(A^TA + \mu I)z = 0, z \neq 0 \tag{3}$$
$$z^T(A^TA + \mu I)z = 0 \text{ since } z^T\vec{0} = 0 \tag{4}$$
$$z^TA^TAz + \mu z^Tz = 0 \tag{5}$$
$$\|Az\|^2 + \mu\|z\|^2 = 0 \tag{6}$$
$$z = \vec{0} \tag{7}$$

So, z can only be zero, meaning $\mathcal{N}(A) = \{0\} \Rightarrow (A^TA - \mu I)$ is invertible. This is also why μ must be positive. Or, you know it's invertible, since it is full rank (and skinny) when you stack μI below it (see definition of $\tilde{A}$).

- Application of Regularized least-squares
  - estimation/inversion
  - $Ax - y$ is sensor residual
  - prior information that x is really small
  - or, model only accurate for small x
  - Tychonov solution trades off sensor fit and size of x
- Image processing example
  - Laplacian regularization
    - image reconstruction problem
  - x is vectorized version of image
  - $\|Ax - y\|^2$ is difference from real image
  - Want new objective to minimize roughness
    - vector Dx (from new matrix D) which has difference between neighboring pixels as elements
      - D_vx measures vertical difference
      - D_hx measures horizontal difference
      - Nullspace is vector where there is no variation between pixels
  - minimize $\|Ax - y\|^2 + \mu\|[D_hx\ D_vx]^T\|^2$
    - if μ is turned way up, it'll be all smoothed out
    - if you care about total size of image, you can add another parameter λ: $\|Ax-y\|^2+\mu\|[D_hx\ D_vx]^T\|^2+\lambda\|x\|^2$

6.5 Nonlinear least squares (NLLS) problem

- find $x \in \mathbb{R}^n$ that minimizes $\|r(x)\|^2 = \sum_{i=1}^{m} r_i(x)^2$
- $r(x)$ is vector of residuals; $r(x) = Ax - y \Rightarrow$ problem reduces to linear least squares problem
- in general, can't **really** solve a NLLS problem, but can find good heuristics to get a locally optimal solution

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6.6 Gauss-Newton method for NLLS

- Start guess for x
- Loop
  - linearize r near current guess
  - new guess is linear LS solution, using linearized r
  - if convergence, stop
- Linearize?
  - Jacobian: $(Dr)_{ij} = \partial r_i / \partial x_j$
  - Linearization: $r(x) \approx r(x^{(k)}) + Dr(x^{(k)})(x - x^{(k)})$
  - Set this linearized approximation equal to $r(x) \approx A^{(k)}x - b^{(k)}$
    - $A^{(k)} = Dr(x^{(k)})$
    - $b^{(k)} = Dr(x^{(k)})x^{(k)} - r(x^{(k)})$
  - See rest in notes
  - At k th iteration, approximate NLLS problem by linear LS problem:
    - $\|r(x)\|^2 \approx \left\|A^{(k)}x - b^{(k)}\right\|^2$
      - if you wanna make this really cool add a $\mu\|x - x^{(k)}\|^2$ term on RHS
      - called a 'trust region term';
      - first (original) part says to minimize sum of squares for *model*
      - trust region term says 'but don't go far from where you are now'
- Could also linearize without calculus; works really well
  - See 'particle filter'

6.7 G-N example

- Nice graph and residual plot
- As practical matter, good to run simulation several times (with different initial guesses)
- 'exuastive simulation'

6.8 Underdetermined linear equations

- $A \in \mathbb{R}^{m \times n}, m < n$ (A is fat)
- more variables than equations
- x is underspecified
- For this sectian **assume A is full rank**
- Set of all solutions has form $\{x|Ax = y\} = \{x_p + z|z \in \mathcal{N}(A)\}$
- solution has dim $\mathcal{N}(A) = n - m$ 'degrees of freedom'
  - many DOF: good for design (flexibility), bad for estimation (stuff you don't/can't know with available measurements)

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6.9 Least norm solution

- $\boxed{x_{ls} = A^T(AA^T)^{-1}y}$
  - similar to our familiar skinny A version: $x_{ls} = (A^TA)^{-1}A^Ty$
  - mnemonic: $(\cdot)^{-1}$ thing must be square
    - if A skinny, both \$A A ^{T} and \$ A^TA could be square (syntactically)
    - semantically, you need the up and down patterns that will form the smallest square, i.e., full rank matrix

7 Lecture 9 pt 2

Thank you fucking Suncheon.

7.1 General norm minimization with equality constraints

- Problem: $\boxed{\text{minimize } \|Ax-b\| \text{ subject to } Cx = d, \text{ with variable } x}$
- Least squares/least norm are special cases
  - Least norm: set $A = I, b = 0$, then you just have norm of x subject to some linear equations
- Same as: minimize $(1/2)\|Ax-b\|^2$ subject to $Cx = d$
- Lagrangian is... long ugly thing... look at notes
  - a bit easier to look at block matrix format

$$\begin{bmatrix} A^TA & C^T \\ C & 0 \end{bmatrix} \begin{bmatrix} x \\ \lambda \end{bmatrix} = \begin{bmatrix} A^Tb \\ d \end{bmatrix}$$

- recover least squares (maybe) by eliminating C from matrix (not setting to zero, but only having 1 row/column in first matrix)

7.2 Autonomous linear dynamical systems

"What the class is nominally about"

- In continuous time, autonomous LDS has form $\dot{x} = Ax$
- Solution: $x(t) = e^{ta}x(0)$
- $x(t) \in \mathbb{R}^n$ is called the state
  - n is state dimension
- A basically maps where you are (x) to where you're going ($\dot{x}$)
  - has units of s^{-1}, frequency
- Example illustration: vector fields

7.3 Block diagrams

- use integrators to express $\dot{x} = Ax$ instead of differentiators
  - block called 'bank of integrators'
  - historically used because of analog, mechanical computers
- notches to express n signals

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7.4 Linear circuit example

8 Lecture 10

Examples of autonomous linear dynamical systems, $\dot{x} = Ax$

8.1 Example: Series reaction $A \to B \to C$

$$\dot{x} = \begin{bmatrix} -k_1 & 0 & 0 \\ k_1 & -k_2 & 0 \\ 0 & k_2 & 0 \end{bmatrix} x$$

- For second row, first term on rhs of $\dot{x}_2 = k_1 x_1 - k_2 x_1$ is *buildup*
- Note: Column sums are 0 implies conservation of mass/materials;

8.2 Discrete time Markov chain

- $x(t+1) = Ax(t)$
- $x(t) = A^t x(0)$
- Given current state, the matrix of *transition probabilities* P will tell you probabilities of the next state, given the current state

8.3 Numerical integration of continuous system

- for a small time step h, find about where you'll be in h seconds b
- $x(t+h) \approx x(t) + h\dot{x}(t) = (I + hA)x(t)$
- problem: when you do it for a long time, error can build up pretty high

8.4 Higher order linear dynamical systems ($\dot{x} = Ax$)

$x^{(k)} = A_{k-1}x^{(k-1)} + \cdots + A_1 x^{(1)} + A_0 x, x(t) \in \mathbb{R}^n$

- define new variable

$$z = \begin{bmatrix} x \\ x^{(1)} \\ \vdots \\ x^{(k-1)} \end{bmatrix} \in \mathbb{R}^{nk}, \dot{z} = \begin{bmatrix} x^{(1)} \\ \vdots \\ x^{(k)} \end{bmatrix} = \begin{bmatrix} 0 & I & 0 & \cdots & 0 \\ 0 & 0 & I & \cdots & 0 \\ \vdots & & & & \vdots \\ 0 & 0 & 0 & \cdots & I \\ A_0 & A_1 & A_2 & \cdots & A_{k-1} \end{bmatrix} z$$

- 'upshift x, and zero-pad'
- z is the state, not x
- in notes, black diagram with chain of integrators

8.5 Example: Mechanical systems

- Ex: K_{12} is 'cross-stiffness', how much stiffness you'd feel at node 1 from node 2

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8.6 Linearization near equilibrium point

Equilibrium point corresponds to constant solution ($f(x_e) = 0, x(t) = x_e$)

- if you start at an equilibrium point, you'll stay there
- if you start *near* equilibrium point
  - veer off (unstable)
  - go towards equilibrium (stable)
  - something in between
- but, you never stay at an unstable equilibrium position, since equation is really $\dot{x} = f(x) + w(t)$, where $w(t)$ is noise
- Near equilibrium point, $\dot{\delta x}(t) \approx Df(x_e)\delta x(t)$, where D is the Jacobian
  - similar to euler forward equation
- Don't fully trust approximations on approximations (but hope they work)

8.7 Example: pendulum linearization

- $ml^2\ddot{\theta} = -lmg\sin\theta$
- rewrite as 1st order DE with state $x = [\theta\ \dot{\theta}]^T = [x_1\ x_2]^T$:

$$\dot{x} = \begin{bmatrix} x_2 \\ -(g/l)\sin x_1 \end{bmatrix}$$

- $\exists$ equilibrium point at $x = 0$ (and π), so we linearize system near $x_e = 0$, using a Jacobian matrix:

$$\dot{\delta x} = \begin{bmatrix} \frac{\partial x_2}{\partial x_1} & \frac{\partial x_2}{\partial x_2} \\ \frac{\partial}{\partial x_1}\left(-(g/l)\sin x_1\right)|_{x_1=0} & \frac{\partial}{\partial x_2}(-(g/l)\sin x_1) \end{bmatrix}\delta x = \begin{bmatrix} 0 & 1 \\ -g/l & 0 \end{bmatrix}\delta x$$

9 Lecture 11

Solution via Laplace transform and matrix exponential Remember, we've already overloaded $\dot{x} = ax$. Now, we'll overload exponentials to apply to matrices $x(t) = e^{ta}x(0)$.

9.1 Laplace transform

- $z : \mathbb{R}_+ \to \mathbb{R}^{p\times q}$ (function that maps non-negative real scalars to matrices)
- Laplace transform: $Z = \mathcal{L}(z)$, defined by $Z(s) = \int_0^\infty e^{-st}z(t)dt$
- Region of convergence of Z is mostly for confusing students
- Derivative property: $\mathcal{L}(\dot{z}) = sZ(s) - z(0)$

So, we can use the Laplace transform to solve $\dot{x} = Ax$. Take Laplace: $sX(s) - x(0) = AX(s)$, rewrite as $(sI - A)X(s) = x(0)$, so $X(s) = (sI - A)^{-1}x(0)$. Then take the inverse transform: $\boxed{x(t) = \mathcal{L}^{-1}\left((sI - A)^{-1}\right)x(0)}$

- takes advantage if linearity of the Laplace transform
- $(sI - A)^{-1}$ is called the *resolvent* of A
  - but not defined for eigenvalues of A; s, ST $\det(sI - A)$=0
- $\boxed{\Phi = \mathcal{L}^{-1}((sI - A)^{-1})}$ is called the *state-transition matrix*, which maps the initial state to state at time t: $\boxed{x(t) = \Phi(t)x(0)}$

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9.2 Example: Harmonic oscillator

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} x$$

- To solve for s, get the resolvent, then apply the Laplacion to it *elementwise*, getting

$$x(t) = \begin{bmatrix} \cos t & \sin t \\ -\sin t & \cos t \end{bmatrix} x(0)$$

Which is a circular rotation matrix. The solutions to $\dot{x} = ax$ is $x(t) = e^{ta}x(0)$

- a positive: exponential growth
- a negative: exponential decay
- $a = 0$: constant

9.3 Example: Double Integrator

- Note, with scalars, x in $\dot{x} = ax$ grows exponentially in time, and cannot grow linearly, as with matrices (can have a t element in matrix)
- What is first column of $\Phi(t)$ say? It tells what the state trajectory is if the initial condition was e_1 (second column tells what it is if $x(0) = e_2$)
- First row says the linear combination that x_1 is at time t given $x(0)$

9.4 Characteristic polynomial

$\mathcal{X}(s) = \mathbf{det}(sI - A)$; called a *monic* polynomial

- roots of $\mathcal{X}$ are eigenvalues of A, and $\mathcal{X}$ has real coefficients, so e-values are real or occur in conjugate pairs

9.5 Get eigenvalues of A and poles of resolvent

Use Cramer's rule to get i, j entry:

$$(-1)^{i+j} \frac{\det \Delta_{ij}}{\det(sI - A)},$$

where Δ_{ij} is $sI - A$ with j th row and i th column deleted. Poles of entries of resolvent **must** be eigenvalues of A.

9.6 Matrix exponential

How to overload exponentials for matrices; start with $(I - C)^{-1} = I + C + C^2 + \ldots$ Series converges if |eigenvalues of C |<1. Do series expansion of resolvent, then take the Laplacian of the series, which looks like the form for the expansion of e^{ta} (though square matrices replace scalars). So we end by learning that the state transition matrix, $\Phi(t)$ is the matrix exponential e^{tA}.

- Many scalar exponential properties don't extend to matrix exponential; with scalars, this is wrong: $e^{A+B} = e^A e^B$ (unless A and B commute: $AB = BA$)
- But this is ok: $e^{-A} = (e^A)^{-1}$
- So, how do you find the matrix exponential:

Find e^A,

$$A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$$

Found $e^{tA} = \mathcal{L}^{-1}(sI - A)^{-1}$ in earlier example, so just plug in $t = 1$.

- Matlab: `expm(A)`, not elementwise `exp(A)`

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9.7 Time transfer property

Summary: for $\dot{x} = Ax$, $x(t) = \Phi(t)x(0) = \boxed{e^{tA}x(0)}$. $\boxed{\text{The matrix } e^{tA} \text{ propagates initial condition into state at time } t.}$
Also propagates backward in time if $t < 0$.

If given $x(12)$, find $x(0)$ via $e^{-12A}x(12)$.

- Can use first order forward Euler approximate state update for small t
- Discretized autonomous LDS: $z(k+1) = e^{hA}z(k)$ (not an approximation for these equations)

9.8 Application: sampling a continuous time system

10 Lecture 12

Piecewise constant system: A is constant for certain intervals of time.

- Qualitative behavior of $x(t)$
  - Eigenvalues determine (possible) behavior of x
  - Can plot eigenvalues on complex axes; like pole plot
  - Can put x in summation form with polynomial coefficient and exponential terms

10.1 Stability

- $\dot{x} = Ax$ is stable if $e^{tA} \to 0$ as $t \to \infty$
  - means that state $x(t)$ converges to 0 as $t \to \infty$, no matter $x(0)$
  - all trajectories of $\dot{x} = Ax$ converge to 0 as $t \to \infty$
  - $\dot{x} = Ax$ is stable iff all eigenvalues of A have negative real part

10.2 Eigenvectors and diagonalization

- $\lambda \in \mathbb{C}$ is an eigenvalue of $A \in \mathbb{C}^{n \times n}$ if (characteristic polynomial)

$$\mathcal{X}(\lambda) = \det(\lambda I - A) = 0$$

- i.e., $(\lambda I - A)$ is singular, not invertible, $\mathcal{N}$ not equal to the 0 set

Equivalent to:

- $\exists$ nonzero $v \in \mathbb{C}^n$ s.t. $(\lambda I - A)v = 0$: $\boxed{Av = \lambda v}$ (v is the eigenvector)
  - columns are dependent
- $\exists$ nonzero $w \in \mathbb{C}^n$ s.t. $w^T(\lambda I - A) = 0$: $\boxed{w^T A = \lambda w^T}$ (w is the *left eigenvector*)
  - rows are dependent
- real A can still have complex e-pairs
- A, λ real $\Rightarrow \lambda$ is associated with a real v
- conjugate (negate imaginary term of complex number[s])
- hermitian conjugate (and transpose)

10.3 Scaling intepretation

Av is simply scaled version of v (λ times); all components get magnified by the same amount

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10.4 Dynamic intepretation

For $Av = \lambda v$, if $\dot{x} = Ax, x(0) = v \Rightarrow \boxed{x(t) = e^{\lambda t}v} = e^{tA}v$.

- $A^2v = \lambda^2 v$
- So you just need a scalar in front of the v to calculate $x(t)$!
- An eigenvector is an initial condition $x(0)$ for which the entire trajectory is really simple.
- solution $x(t) = e^{\lambda t}v$ is a mode of $\dot{x} = Ax$ (associated with eigenvalue λ)

10.5 Invariant set

a set $S \subseteq \mathbb{R}^n$ is *invariant* under $\dot{x} = Ax$ if whenever $x(t) \in S$, then $x(\tau) \in S$ for all $\tau \geq t$ (you stay stuck within the set)

- vector field intepretation: trajectories only cut *into* S

If a single point is an invariant set, it must be in the nullspace; $S = \{x_0\} \Leftrightarrow x_0 \in \mathcal{N}(A)$, so $Ax_0 = 0 = \dot{x}$.

- line $\{tv | t \in \mathbb{R}\}$ is invariant for eigenvector v

10.6 Complex eigenvectors

- for $a \in \mathbb{C}$, complex trajectory $ae^{\lambda t}v$ satisfies $\dot{x} = Ax$, as well as *real* part

$$x(t) = \text{Re}(ae^{\lambda t}v)$$
$$= e^{\sigma t}\begin{bmatrix} v_{re} & v_{im}\end{bmatrix}\begin{bmatrix}\cos\omega t & \sin\omega t \\ -\sin\omega t & \cos\omega t\end{bmatrix}\begin{bmatrix}\alpha \\ -\beta\end{bmatrix}$$

where

$$v = v_{re} + jv_{im}, \lambda = \sigma + j\omega, a = \alpha + j\beta$$

- σ gives logarithmic growth/decay factor
- ω gives angular velocity of rotation in plane
- trajectory stays in *invariant plane* span $\{v_{re}, v_{im}\}$

10.7 Dynamic interpretation: left eigenvectors

10.8 Summary:

- *right eigenvectors* are initial conditions from which resulting motion is simple (i.e., remains on line or in plane)
- *left eigenvectors* give linear functions of state that are simple, for any initial condition

10.9 Example- companion matrix

- Easy to get the characteristic polynomial
- General truth: with these matrices you can't generally tell the system behavior by just looking at it
- If you push a signal through an integrator, it gets less wiggly
- By multiplying by the left eigenvector, you've filtered out the sinusoid?

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11 Lecture 13

11.1 Example: Markov chain

Probability vector $p \in \mathbb{R}^n$ that you're in each of n states: $p(t+1) = Pp(t)$. This probability evolves in time by being multiplied by state transition matrix P.

- $p_i(t) = \mathbf{Prob}(z(t) = i) \Rightarrow \sum_{i=1}^{n} p_i(t) = 1$
- sum of each column is 1
  - called stochastic
- i.e., $[1\ 1\ \cdots 1]$ is a left eigenvector of P with $\lambda = 1$
- so $\det(I - P)$=0, so there's also a nonzero right eigenvector s.t. $Pv = v$
  - v can always be chosen to have non-negative elements, and can be normalized
- **Interpretation**: v is an equilibrium distribution; you don't change your *probability* distribution in time; always in v
  - if v unique, it's called the steady-state distribution of the Markov chain

11.2 Diagonalization

- $v_1, ..., v_n$ is LI set of eigenvectors of $A \in \mathbb{R}^{n\times n}$: $Av_i = \lambda_i v_i$
- Concatenate in matrix language:

$$A \begin{bmatrix} v_1 & \cdots & v_n \end{bmatrix} = \begin{bmatrix} v_1 & \cdots & v_n \end{bmatrix} \begin{bmatrix} \lambda_1 & & & \\ & \lambda_2 & & \\ & & \ddots & \\ & & & \lambda_n \end{bmatrix}$$

or, $AT = T\Lambda$, or $T^{-1}AT = \Lambda$

- note, T is invertible, since its columns are linearly independent
- This is why, while $Av = \lambda v$ is more commonly used for a scalar eigenvalue, $\boxed{Av = v\lambda}$ is more general, as it can represent a vector of eigenvalues λ.
- so, A is diagonalizable if
  - $\exists\ T$ s.t. $T^{-1}AT = \Lambda$ is diagonal
  - A has a set of linearly independent eigenvectors
    - if A not diagonalizable, it is called defective

11.3 Not all matrices diagonalizable

i.e.,

$$A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$$

11.4 Distinct eigenvalues

fact: distinct eigenvalues in $A \Rightarrow A$ diagonalizable

- converse not true, i.e., $I \in \mathbb{R}^{7\times 7}$

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11.5 Diagonalization and left eigenvectors

rewrite $T^{-1}AT = \Lambda$ as $T^{-1}A = \Lambda T^{-1}$:

$$\begin{bmatrix} w_1^T \\ \vdots \\ w_n^T \end{bmatrix} A = \Lambda \begin{bmatrix} w_1^T \\ \vdots \\ w_n^T \end{bmatrix}$$

- remember that Λ is diagonal matrix, and multiplying by a diagonal matrix on the left is equivalent to scaling rows of the matrix
  - on the right scales the columns

Remeber left/right multiplication results (whether it scales columns or rows) with 2×2 matrix multiplication:

$$\begin{bmatrix} 2 & 0 \\ 0 & 3 \end{bmatrix} \begin{bmatrix} x_1 & x_2 \\ y_1 & y_2 \end{bmatrix} = \begin{bmatrix} 2x_1 & 2x_2 \\ 3y_1 & 3y_2 \end{bmatrix}$$

I.e., right multiplication of diagonal matrix scales the rows.

- Take LI set of eigenvectors as columns, invert that matrix, then the rows are **left** eigenvectors
- An eigenvector is still an eigenvector after being scaled; so any can be normalized

11.6 Modal form

Take a LI set of eigenvectors from A, shove them together as columns of new matrix $T =$ "A is diagonalizable by T"

- can define new coordinates by $x = T\tilde{x}$:
- $\tilde{x}$ is coordinates of x in the T expansion; modal (or eigenvector) expansion
  - $\tilde{x}$ is x in terms of the eigenvectors

$$T\dot{\tilde{x}} = AT\tilde{x} \Leftrightarrow \dot{\tilde{x}} = T^{-1}AT\tilde{x} \Leftrightarrow \dot{\tilde{x}} = \Lambda\tilde{x}$$

- in new coordinate system, system is diagonal (decoupled)
- normally, with $\dot{x} = Ax$, there's a ton of cross-gains from input x_i to output y_j, where all the outputs depend on all the inputs (assuming A has only non-zero entries)
  - $\boxed{\text{diagonalized system decouples it}}$; trajectory consists of n independent modes:

$$\tilde{x}_i(t) = e^{\lambda_i t}\tilde{x}_i(0)$$

11.7 Real modal form

when eigenvalues ($\Rightarrow T$) are complex

- notes show block diagram of complex mode (note if real parts σ are removed, you get harmonic oscillator)

11.8 Diagonalization simplification

Simplifies calculation of:

- resolvent
- powers (A^k)
- exponential ($e^A = T\mathbf{diag}(e^{\lambda_1}, \ldots, e^{\lambda_n})T^{-1}$)
- So, diagonalization is largely a conceptual tool, and sometimes gives great computational advantage

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11.9 Simplify for analytical functions of a matrix

11.10 Solution via diagonalization

$\dot{x} = Ax$ solution is $x(t) = e^{tA}x(0)$

- with diagonalization, solution given as

$$x(t) = \sum_{i=1}^{n} e^{\lambda_i t}(w_i^T x(0))v_i$$

11.11 Interpretation

- (left eigenvectors) decompose initial state $x(0)$ into modal components $w_i^T x(0)$
- $e^{\lambda_i t}$ term propagates i th mode forward t seconds
- reconstruct state as linear combination of (right eigenvectors)

11.12 Application

Finding $x(0)$ that gives stable solution.

11.13 Stability of discrete-time systems

- powers of complex numbers s^k go to zero if $|s| < 1$
  - imaginary part tells how much of a rotation at each step you get
- $\boxed{x(t+1) = Ax(t) \text{ is stable iff all eigenvalues of } A \text{ have magnitude less than one}}$
- spectral radius of $A : \rho(A) = \mathbf{max}|\lambda_i|$
  - so it is a stable system iff $\rho(A) < 1$
  - ρ gives rough growth or decay

11.14 Jordan Canonical form

- *Any* matrix $A \in \mathbb{R}^{n\times n}$ can be expressed in Jordan-canonical form (via 'similarity transformation,' for some invertible matrix T^{-1})

$$T^{-1}AT = J = \begin{bmatrix} J_1 & & \\ & \ddots & \\ & & J_q \end{bmatrix}$$

where

$$J_i = \begin{bmatrix} \lambda_i & 1 & & \\ & \lambda_i & \ddots & \\ & & \ddots & 1 \\ & & & \lambda_i \end{bmatrix} \in \mathbb{C}^{n_i \times n_i}$$

- J is 'upper bidiagonal'
- Jordan form is unique (up to permutations of blocks- blocks might be in different places in the diagonal)
- *Almost* strictly a conceptual tool; almost never used for numerical computations
- Jordan forms are inutil if the matrix is already diagonalizable

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- When you get into Jordan form, you can use a chain of integrators to represent it in block diagram form
- Jordan blocks refer to dynamics blocks that cannot be decoupled
- Jordan blocks yield:
  - repeated poles in resolvent
  - terms of form $t^p e^{t\lambda}$ in e^{tA}: solution to $\dot{x} = Ax$ is in form of polynomials, with $t^n e^{\lambda t}$ terms

11.15 Jordan block example

$$\dot{x} = \begin{bmatrix} 0 & 1 & & & \\ & 0 & 1 & & \\ & & 0 & 1 & \\ & & & 0 & 1 \\ & & & & \end{bmatrix} x$$

- Eigenvalues are all zero
- in the solutions you should expect $t^p e^0$ ('constant') terms: solution of form $t + t^2 + t^3$

12 Lecture 14

- Jordan canonical form: 'closest you can get to diagonalizable form, if matrix isn't diagonalizable'
- via inverse Laplace transform:

$$e^{tJ_\lambda} = e^{t\lambda}(I + tF_1 + \cdots + (t^{k-1}/(k-1)!)F_{k-1})$$

$$= e^{t\lambda} \begin{bmatrix} 1 & t & \cdots & t^{k-1}/(k-1)! \\ & 1 & \cdots & t^{k-2}/(k-2)! \\ & & \ddots & \vdots \\ & & & 1 \end{bmatrix}$$

Jordan-blocks should be associated in your mind with polynomials times $e^{\lambda t}$

- All FIR filters have Jordan blocks

12.1 J-C application: Caley-Hamilton theorem

- if proving things about matrices, first show it for diagonalizable matrices, then with Jordan-Canonical form
- for any $n \times n$ matrix A, the powers of $A : I, A^2, A^3, ..., A^n$ span $\mathbb{R}^{n \times n}$
  - $\mathcal{X}(A) = 0$ (characteristic polynomial; by the time you get to n, you're getting dependent matrices)
- wrong proof for evaluating characteristic polynomial $\mathcal{X}(s) = \mathbf{det}(sI - A)$ at A:
  - $\mathcal{X}(A) = \mathbf{det}(AI - A) = \mathbf{det}(0) = 0$
  - wrong, since definition of $\mathcal{X}$ involves sI term, which is diagonal matrix (so A can't be plugged into it)
- correct example for

$$A = \begin{bmatrix} 1 & 2 \\ 3 & 4 \end{bmatrix}$$

$$\begin{aligned} \mathcal{X} &= A^2 - 5A - 2I && (8) \\ &= \begin{bmatrix} 7 & 10 \\ 15 & 22 \end{bmatrix} - 5\begin{bmatrix} 1 & 2 \\ 3 & 4 \end{bmatrix} - 2I && (9) \\ &= 0 && (10) \end{aligned}$$

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- Caley-Hamilton theorem: the square of a (square) matrix is a linear combination of the Identity matrix and itself
  - the inverse is also a linear combination of I and powers of A:
  - rewrite C-H theorem: $\mathcal{X}(A) = A^n + a_{n-1}A^{n-1} + \cdots + a_0 I = 0$ as
    - $I = A\left(-(a_1/a_0)I - (a_2/a_0)A - \cdots - (1/a_0)A^{n-1}\right)$

$$A^{-1} = -(a_1/a_0)I - (a_2/a_0)A - \cdots - (1/a_0)A^{n-1}$$

- Corollary: for every $p \in \mathbb{Z}_+$,

$$A^p \in \mathbf{span}\{I, A, A^2 ..., A^{n-1}\}$$

12.2 Solving huge matrix equations

- usually have a method for fast matrix vector multiplication: Az
- if you can do a quick Az calculation, you can also do a quick $A^2z, ...A^{n-1}z$ calculation
  - using C-H thm, you can put scalars in front of them to get A^{-1}

12.3 LTI system inputs & outputs

LTI systems with input and outputs have the form $\dot{x} = Ax + Bu, y = Cx + Du$; Ax is the *drift* term, Bu is the *input* term. For

12.4 Interpretations

- state derivative $\dot{x}$ is the sum of the autonomous term Ax and one term per input, $b_i u_i$
  - if columns of B are independent, each input u_i gives another degree of freedom for $\dot{x}$
  - column i of B represents how much input i affects $\dot{x}$
  - row i of B represents how much input i affects the i th component of $\dot{x}$
- Good block diagram in notes, pg. 13-5

12.5 Transfer matrix

- Solution to $\dot{x} = Ax + Bu$ via Laplace transform is $\boxed{x(t) = e^{tA}x(0) + \int_0^t e^{(t-\tau)A}Bu(\tau)d\tau}$
  - $e^{tA}x(0)$ is the (unforced or) autonomous response
  - $e^{tA}B$ is the 'input-to-state impulse matrix'
    - integral term is convolution
  - $(sI - A)^{-1}B$ is input-to-state *transfer matrix*; the resolvent $\times$ B
- Readout equation, $y = Cx + Du$ solution (via Laplacian)
  - $\boxed{y(t) = Ce^{tA}x(0) + \int_0^t Ce^{(t-\tau)A}Bu(\tau)d\tau + Du(t)}$
  - transfer function: $H(s) = C(sI - A)^{-1}B + D$
  - impulse matrix/impulse response: $h(t) = Ce^{tA}B + D\delta(t)$
  - zero initial condition gives $Y(s) = H(s)U(s), y = h * u$
    - H_{ij} is transfer function from input u_j to output y_i

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- Mnemonic for convolution intuition $y(t) = \int_0^t h(t-\tau)u(\tau)d\tau$:
  - output is 'linear combination' or 'mixture' of input in the past $u(\tau)$, weighted by how many seconds ago something happened
  - τ is 'seconds ago'
  - example: if $h(7)$ is really big $\rightarrow$ output depends a *lot* on the input 7 seconds ago
    - remember, the h term has been delayed and flipped: $h(-(\tau - t))$
  - if $h(t)$ decays, the system has *fading memory*

12.6 Impulse matrix

- $h(t) = Ce^{tA}B + D\delta(t), x(0) = 0, y = h * u$

$$y_i(t) = \sum_{j=1}^{m} \int_0^t h_{ij}(t-\tau)u_j(\tau)d\tau$$

- for rain/river metaphor, where rain is input and river level is output, this matrix is summed over the input, with different regions of rainfall u_j and different locations of the river where rainfall is measured, $y_i(t)$

12.7 Step response

$$s(t) = \int_0^t h(\tau)d\tau$$

12.8 Spring dashpot example

12.9 RC circuit example

13 Lecture 15

13.1 DC or static gain matrix

- transfer matrix at $s = 0$ is $H(0) = -CA^{-1}B + D \in \mathbb{R}^{m\times p}$
- DC transfer matrix describes system under static conditions: x, u, y constant
  - $\Rightarrow y = H(0)u$
  - if system is stable, $H(0) = \int_0^\infty h(t)dt = \lim_{t\to\infty} s(t)$

13.2 Discretization with piecewise constant inputs

13.3 Causality

13.4 Idea of state

- $x(t)$ is state of system at time t:
  - future output depends only on *current state* and *future output*
    - sufficient statistic of what happened in the past to be able to predict the future
  - bridge between past inputs and future outputs
    - 'past and future are conditionally independent, given the state' (-machine learning)

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13.5 Change of coordinates

- start with LDS $\dot{x} = Ax + Bu, y = Cx + Du$
- change coordinates in $\mathbb{R}^n$ to $\tilde{x}$ with $x = T\tilde{x}$

$$\dot{\tilde{x}} = T^{-1}\dot{x} = T^{-1}(Ax + Bu) = T^{-1}AT\tilde{x} + T^{-1}Bu$$

- LDS can be expressed as

$$\dot{\tilde{x}} = \tilde{A}\tilde{x} + \tilde{B}u, y = \tilde{C}\tilde{x} + \tilde{D}u$$

where

$$\tilde{A} = T^{-1}AT, \tilde{B} = T^{-1}B, \tilde{C} = CT, \tilde{D} = D$$

- You might want to do this to, i.e., get an A with a bunch of zeros, since the system might be much easier to implement

13.6 Standard forms for LDS

Can change coordinates to put A in various forms (diagonal, real modal, Jordan ...)

13.7 Discrete-time systems

(Another block diagram in notes)

13.8 Z-transform

13.9 Discrete-time transfer function

- re-derivation of stuff we know, using the frequency domain

14 Last Section, Singular Value Decomposition

14.1 Eigenvalues of symmetric matrices

Given $A \in \mathbb{R}^{n\times n} : A = A^T$

- eigenvalues always real
- $Av = \lambda v, v \neq 0, v \in \mathbb{C}^n$; remember conjugate transpose, $\bar{v}^T = v^H = v^{*T}$, or, in MATLAB,

```
v'
```

and remember, $\bar{a}a = |a|^2$

$$\Rightarrow v^H Av = v^H(Av) = \lambda v^H v = \lambda \sum_{i=1}^{n} |v_1|^2$$

$$... = \bar{\lambda} \sum_{i=1}^{n} |v_1|^2$$

so $\lambda = \bar{\lambda} \Rightarrow \lambda \in \mathbb{R}^n$

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14.2 Eigenvectors of symmetric matrices

fact: There is a set of orthonormal eigenvectors of A, $q_1, ..., q_n : Aq_i = \lambda_i q_i, q_i^T q_j = \delta_{ij}$

- In matrix form, $\exists$ orthogonal $Q : Q^{-1}AQ = Q^T AQ = \Lambda$
- or, rewrite *dyadic expansion of A*:

$$A = Q\Lambda Q^T = \sum_{i=1}^{n} \lambda (q_i q_i^T)$$

- *engineering etiquette aside*: "The eigenvectors of a symmetric matrix are orthonormal" makes no sense; any matrix has zillions of eigenvectors
  - but rather, "you can choose the eigenvectors of a symmetric matrix to be orthonormal"

14.3 Interpretations

Given $A = Q\Lambda Q^T$. Remember, $Q^{-1} = Q^T$. (Notes have block diagram.) This means, to multiply a vector x by A, first multiply it by Q^T, then Λ, then Q

- first operation result $Q^T x = Q^{-1}x$ 'resolves x in the q_i coordinates'
- next: simple to multiply by diagonal matrix Λ (simply scaling the matrix)
- last: multiplying by Q 'reconstitutes the output'
- multiplying by Q^T, Λ, Q represents a 'coding,' scaling, reconstruction operation
- Application: in JPEG, do DCT transformation, quantize in middle, then decode the image in last step
- Geometrically:
  - rotate by Q^T
  - dilation by A
  - rotate back by Q

Decomposition (review)

$$A = \sum_{i=1}^{n} \lambda_i q_i q_i^T$$

expresses A as linear combination of 1-dimensional projections.

If a matrix is real and symmetric, (and the eigenvalues of a matrix are distinct,) then any set of associated eigenvectors is orthogonal (and can be normalized).

$$Av_i = \lambda_i v_i, \|v_i\| = 1$$

Then,

$$v_i^T(Av_j) = \lambda_j v_i^T v_j = (Av_i)^T v_j = \lambda_i v_i^T v_j$$

so $(\lambda_i - \lambda_j)v_i^T v_j = 0$ for $i \neq j, \lambda_i \neq \lambda_j$ $\boxed{\Rightarrow v_i^T v_j = 0}$ So distinct eigenvectors of a symmetric real matrix are 'orthogonal'

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14.4 Example: RC circuit

15 Lecture 16

15.1 Quadratic forms

Previously, we did mappings of form $y = Ax$, or $\dot{x} = Ax$. We're no longer doing problems with a linear structure.

- New form is *quadratic form*

$$f(x) = x^T Ax = \sum_{i,j=1}^{n} A_{ij} x_i x_j$$

- $(x^T Ax)^T = x^T A^T x$
- $x^T Ax = x^T((A + A^T)/2)x$ is another form of quadratic form; the middle part is the average of itself (member of A) and its associated trasposed self
- weird etiquette: A should be symmetric if you're dealing with quadratic forms
  - often may want to symmetrize it to make it safe
  - i.e., this quadratic form

$$x^T \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x = x_1 x_2$$

$$x^T \begin{bmatrix} 0 & 3/2 \\ -1/2 & 0 \end{bmatrix} x = x_1 x_2$$

But the canonical form would have $1/2$ in the 2nd and 3rd entry (row-wise).

- If A is diagonal in LDS, $y = Ax$, it means there is no cross-coupling. If A is diagonal in a quadratic form, it gives you a weighted sum of squares.
- **uniqueness:** if $x^T Ax = x^T Bx$ for all $x \in \mathbb{R}^n$ and $A = A^T, B = B^T$, then $A = B$

15.2 Examples of quadratic forms

- $\|Bx\|^2 = x^T(B^T B)x; B^T B is already symmetric$
- $\sum_{i=1}^{n-1}(x_{i+1} - x_i)^2$. This measures 'wiggliness' of signal; sum of squares of difference from one to the next signal
  - to get in quadratic form, multiply out: $x_{i+1}^2 + x_i^2 - 2x_i x_{i+1}$:

$$\begin{bmatrix} x_1 & x_2 & x_3 \end{bmatrix} \begin{bmatrix} 1 & -1 & 0 \\ -1 & 2 & -1 \\ 0 & -1 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

- Zeros in corners of matrix, since there is no product between x_1, x_3
- $\|Fx\|^2 - \|Gx\|^2$
  - $x^T(F^T F - G^T G)x$
- quadratic surface: level set: $\{x | f(x) = a\}$
- quadratic region: $\{x | f(x) \leq a\}$
  - unit ball: $\{x | x^T x \leq 1\}$
  - unit sphere: $\{x | x^T x = 1\}$

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15.3 Inequalities for quadratic forms

If $A = A^T$, it is diagonalizable: $A = Q\Lambda Q^T$ with sorted eigenvalues in Λ

$$\begin{aligned} x^T Ax &= x^T Q\Lambda Q^T x && (11)\\ &= (Q^T x)^T \Lambda (Q^T x) && (12)\\ &= \sum_{i=1}^{n} \lambda_i (q_i^T x)^2 && (13)\\ &\leq \lambda_1 \sum_{i=1}^{n} (q_i^T x)^2 && (14)\\ &= \lambda_1 \|x\|^2 && (15)\\ & && (16) \end{aligned}$$

- So, how big can the quadratic form be?
  - $x^T Ax \leq \lambda_1 x^T x$
- Another important inequality

$$\lambda_n x^T x \leq x^T Ax \leq \lambda_1 x^T x$$

- λ_1 sometimes called $\lambda_{\max}$, λ_n called $\lambda_{\min}$
- for $\boxed{A = A^T} \in \mathbb{R}^{n\times n}$, A is *positive semidefinite* if $x^T Ax \geq 0$ for all x
  - notation: $A \geq 0$
  - iff $\lambda_{min}(A) \geq 0$: all eigenvalues nonnegative
  - A is *positive definite* if $x^T Ax > 0$ for all $x \neq 0$
- $A \geq 0 \Rightarrow A_{ij} \geq 0$
- if A not positive or negative [semi]definite, it is *indefinite*
- matrices **can be** *incomparable*:
  - $A \not\geq B$ **and** $B \not\geq A$, where it will depend on the x (depend on the direction you're looking at it from)
- $A \leq B$ statement requires that both be square matrices, and probably symmetric (depending on social norms)
  - if not symmetric, it means $\frac{A+A^T}{2} \leq \frac{B+B^T}{2}$

15.4 Ellipsoids

- if $A^T > 0$, the set $\mathcal{E} = \{x | x^T Ax \leq 1\}$ is an *ellipsoid* in $\mathbb{R}^n$, centered at 0
- semi-axes given by $s_i = \lambda_i^{-1/2} q_i$
  - eigenvectors determine directions of semiaxes
  - eigenvalues determine lengths of semiaxes
  - $\sqrt{\lambda_{max}/\lambda_{min}}$ gives maximum eccentricity
- eigenvalue intuition in $\mathbb{R}^3$
  - large pancake shape $\Rightarrow$ one large eigenvalue, two slightly smaller eigenvalues
  - long cigar shape $\Rightarrow$ two large eigenvalues, one small eigenvalue

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15.5 Gain of matrix in a direction

For $A \in \mathbb{R}^{m \times n}$ (not necessarily symmetric anymore), for $x \in \mathbb{R}^n$, $\|Ax\|/\|x\|$ gives *amplification factor* or *gain* of A in direction x, and will vary with different x directions

- Questions we will ask
  - maximum gain of A
    - $\boxed{\max_{x \neq 0} \frac{\|Ax\|}{\|x\|}}$
    - *matrix norm* or *spectral norm*, and denoted $\|A\|$ (overloading)
    - $\max_{x \neq 0} \frac{\|Ax\|}{\|x\|} = \max_{x \neq 0} \frac{x^T A^T A x}{\|x\|^2} = \lambda_{\max}(A^T A)$
    - so, $\|A\| = \sqrt{\lambda_{\max}(A^T A)}$
  - minimum gain of A
    - $\min_{x \neq 0} \|Ax\|/\|x\| = \sqrt{\lambda_{\min}(A^T A)}$
  - how does gain of A vary with direction
- Will give us a *quantitative way* to talk about nullspace

15.6 Example

15.7 Properties of matrix norm

- triangle inequality
- scaling
- definiteness: $\|A\| = 0 \Leftrightarrow A = 0$
- norm of a product inequality

16 Lecture 17

16.1 SVD

Singular value decomposition, for *any* matrix A.

$$A = U \Sigma V^T$$

where

- $A \in \mathbb{R}^{m \times n}, \mathbf{Rank}(A) = r$
- $A \in \mathbb{R}^{m \times r}, U^T U = I$
- $V \in \mathbb{R}^{n \times r}, V^T V = I$
- $\Sigma = \mathbf{diag}(\sigma_1, ..., \sigma_r)$, where $\sigma_1 \geq ... \geq \sigma_r > 0$

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16.2 Dyadic expansion: (U and V have orthonormal columns)

$$A = U\Sigma V^T = \sum_{i=1}^{r} \sigma_i u_i v_i^T$$

- σ_i are the nonzero *singular values* of A
- v_i are the *right* or *input singular vectors* of A
- u_i are the *left* or *output singular vectors* of A

$$A^T A = (U\Sigma V^T)^T (U\Sigma V^T) = V\Sigma^2 V^T$$

- v_i are eigenvectors of $A^T A$ corresponding to nonzero eigenvalues
- $\sigma_i = \sqrt{\lambda_i(A^T A)}$ (and $\lambda_i(A^T A) = 0$ for $i > r$)
- $\|A\| = \sigma_1$

$$AA^T = (U\Sigma V^T)(U\Sigma V^T)^T = U\Sigma^2 U^T$$

- u_i are eigenvectors of AA^T corresponding to nonzero eigenvalues
- $\sigma_i = \sqrt{\lambda_i(AA^T)}$ (and $\lambda_i(AA^T) = 0$ for $i > r$)
- $u_1, ..., u_r$ are orthonormal basis for range(A)
- $v_1, ..., v_r$ are orthonormal basis for $\mathcal{N}(A)^\perp$

16.3 Interpretations

$A = U\Sigma V^T$

linear mapping $y = Ax$ can be decomposed as

- compute coefficients of x along input directions $v_1, ..., v_r$
- scale coefficients by σ_i
- reconstitute along output directions $u_1, ..., u_r$
- u_1 is highest gain output direction
- v_1 is most sensitive (highest gain) input direction
- $Av_1 = \sigma_1 u_1$

SVD gives clearer picture of gain as function of input/output directions.

16.4 General pseudo-inverse

- $A \neq 0 \Rightarrow A = U\Sigma V^T$

General pseudo-inverse: $\boxed{A^\dagger = V\Sigma^{-1}U^T}$ is the pseudo-inverse of A.

If A is skinny and full rank,

$$A^\dagger = (A^T A)^{-1} A^T$$

If A is fat and full rank,

$$A^\dagger = A^T (AA^T)^{-1}$$

In the general case, (if A is not full-rank)

$$X_{ls} = \{z : \|Az - y\| = \min_w \|Aw - y\|\}$$

is the set of least squares solutions.

$x_{pinv} = A^\dagger y \in X_{ls}$ has minimum norm on X_{ls}, so this is the minimum-norm, least squares solution

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16.5 Pseudo-inverse via regularization

Minimizer of $\|Ax-y\|^2+\mu\|x\|^2$ is given as

$$x_\mu = (A^TA+\mu I)^{-1}A^Ty$$

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17.1 Sensitivity of linear equations to data error

To perturbations $\delta x, \delta y$. Looking at *relative error* gives us

$$\frac{\|\delta x\|}{\|x\|} \leq \|A\|\|A^{-1}\|\frac{\|\delta y\|}{\|y\|}$$

$$\kappa(A) = \text{ cond}(A) = \|A\|\|A^{-1}\| = \sigma_{\max}(A)/\sigma_{\min}(A)$$

So, the relative error in solution $x \leq$ condition number $\cdot$ relative error in data y. In terms of # of bits of guaranteed accuracy, '# of bits accuracy in solution $\approx$ # bits accuracy in data $-\log_2\kappa$.'

17.2 Low rank applications

If you want to decompose A into $A=BC$, you can remember QR factorization (does skinny-fat factorization), but we'd also want an A that is approximated by such a factorization. Suppose $A\in\mathbb{R}^{m\times n}, \mathbf{Rank}(A)=r$ with SVD $A=U\Sigma V^T=\sum_{i=1}^r\sigma_iu_iv_i^T$, and we want $\hat{A}:\mathbf{Rank}(\hat{A})\leq p<r:\|A-\hat{A}\|$ is minimized.

The optimal rank p approximator of A is

$$\hat{A}=\sum_{i=1}^p\sigma_iu_iv_i^T$$

- $\|A-\hat{A}\| = \|\sum_{i=p+1}^r\sigma_iu_iv_i^T\| = \sigma_{p+1}$
- SVD dyads $u_iv_i^T$ are ranked in order of 'importance'; take p to get rank p approximant
- keep in mind Frobenius norm $\|A-\hat{A}\|_F = \|A(i)-\hat{A}(i)\| = \sqrt{\sum_{i,j}(A_{ij}-\hat{A}_{ij})^2}$
- If you have a set of 100 $\mathbb{R}^{10}$ prices that has dimension=3, it means there are 3 underlying factors; this would give you 'stunning' predictive power
- You would *almost* (i.e., practically) have a dimension 3 matrix if the SVD Σ had 3 singular values that were significantly larger than the rest

17.3 State transfer

17.4 Reachability

Consider state transfer from $x(0)=0$ to $x(t)$

- $x(t)$ is reachable in t seconds
- define $\mathcal{R}_t\subseteq\mathbb{R}^n$ as the set of points reachable in t seconds for CT system $\dot{x}=Ax+Bu$,

$$\mathcal{R}_t=\left\{\int_0^t e^{(t-\tau)A}Bu(\tau)d\tau\Big|u:[0,t]\to\mathbb{R}^m\right\}$$

and for DT system $x(t+1)=Ax(t)+Bu(t)$,

$$\mathcal{R}_t=\left\{\sum_{\tau=0}^{t-1}A^{t-1-\tau}Bu(\tau)\Big|u(t)\in\mathbb{R}^m\right\}$$

If a state is reachable in 1 s, it's reachable in 2 s (just leave it at zero for the first second) if you're starting from zero initial state (in both CT and DT systems).

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17.5 Reachability for discrete-time LDS

DT system $x(t+1) = Ax(t) + Bu(t), x(t) \in \mathbb{R}^n$

$$x(t) = \mathcal{C}_t \begin{bmatrix} u(t-1) \\ \vdots \\ u(0) \end{bmatrix}$$

where $\mathcal{C}_t = [B\ Ab\ \cdots\ A^{t-1}B]$. So reachable set at t is $\mathcal{R}_t = \text{range}(\mathcal{C}_t)$

If we want to see if a state is reachable in 1 step, see if it's in the range of B.

If we want to see if it's reachable in 2 steps, see if it's in the range of B, AB. Caley-Hamilton theorem lets us express each A^k for $x \geq n$ as linear combination of $A^0, ..., A^{n-1}$.

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But, you can't increase your reach after $t \geq n$ seconds (or epochs). I.e.,

$$\mathcal{R}_t = \begin{cases} \textbf{range}(\mathcal{C}_t) & t < n \\ \textbf{range}(\mathcal{C}) & t \geq n \end{cases}$$

where $\mathcal{C} = \mathcal{C}_n$ is the *controllability matrix*.

18.1 General state transfer

$$x(t_f) = A^{t_f - t_i} x(t_i) + \mathcal{C}_{t_f - t_i} \begin{bmatrix} u(t_f - 1) \\ \vdots \\ u(t_i) \end{bmatrix}$$

- First term on rhs is what would happen without any input (the drift term)

so we can transfer $x(t_i)$ to $x(t_f) = x_{des}$ iff

$$x_{des} - A^{t_f - t_i} x(t_i) \in \mathbb{R}_{t_f - t_i}$$

- we're not hoping x_{des} is reachable, we want it minus the drift
  - So it is possible to be able to reach a state in 4 steps, but not in 5
- This reduces to the reachability problem
- if system is controllable, any state transfer can be achieved in $\leq n$ steps
- important special case: driving state to zero

18.2 Minimum time control problem

Given

$$x(t+1) = Ax(t) + Bu(t), \quad x(0) = x_0$$

how do you minimize t? Check if $A^t x_0 \in \mathcal{R}([B\ \cdots\ A^{t-1}B])$

18.3 Least-norm input for reachability

assume system is reachable, $\mathbf{Rank}(\mathcal{C}_t) = n$ to steer $x(0) = 0$ to $x(t) = x_{des}$, inputs $u(0), ..., u(t-1)$ to satisfy

$$x_{des} = \mathcal{C}_t \begin{bmatrix} u(t-1) \\ \vdots \\ u(0) \end{bmatrix}$$

among all u that steer $x(0) = 0$ to $x(t) = x_{des}$, the one that minimizes

$$\sum_{\tau=0}^{t-1} \|u(\tau)\|^2$$

The least-square solution for the input is given by

$$\begin{bmatrix} u_{ln}(t-1) \\ \vdots \\ u_{ln}(0) \end{bmatrix} = \mathcal{C}_t^T (\mathcal{C}_t \mathcal{C}_t^T)^{-1} x_{des}$$

18.4 Stability

18.5 Continuous time reachability

consider $\dot{x} = Ax + Bu$ with $x(t) \in \mathbb{R}^n$

reachable set at time t is

$$\mathcal{R}_t = \left\{ \int_0^t e^{(t-\tau)A} B u(\tau) d\tau \,|u : [0,t] \to \mathbb{R}^m \right\}$$

for $t > 0$, $\mathcal{R}_t = \mathcal{R} = \mathbf{range}(\mathcal{C})$ where

$$\mathcal{C} = [B \; AB \; \cdots \; A^{n-1}B]$$

is the controllability matrix of (A, B)

- According to the model, you can hit anything infinitely fast

18.6 Impulsive inputs

18.7 Example

18.8 Least-norm input for reachability

Assume that $\dot{x} = Ax + Bu$ is reachable. We want u that steers $x(0) = 0$ to $x(t) = x_{des}$ and minimizes

$$\int_0^t \|u(\tau)\|^2 d\tau$$

let's discretize system with interval $h = t/N$ (eventually let $N \to \infty$). So u is piecewise constant:

$$u(\tau) = u_d(k) \text{ for } xh \le \tau < (k+1)h, \quad k = 0, ..., N-1$$

... Get long expression for least norm solution for u

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19.1 Observability and state estimation

Consider DT system

$$x(t+1) = Ax(t) + Bu(t) + w(t), \quad y(t) = Cx(t) + Du(t) + v(t)$$

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- w is state disturbance/noise
- v is sensor noise/error
- state estimation problem: estimate $x(s)$ from

$$u(0), ..., u(t-1), \quad y(0), ..., y(t-1)$$

- $s = 0$: estimate initial state
- $s = t - 1$: estimate current state
- $s = t$: estimate/predict next state
- estimate $\tilde{x}(s)$, called *observer* or *state estimator*

19.2 Noiseless case

Find $x(0)$, with no state or measurement noise

$$x(t+1) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

then we have

$$\begin{bmatrix} y(0) \\ \vdots \\ y(t-1) \end{bmatrix} = \mathcal{O}_t x(0) + \mathcal{T}_t \begin{bmatrix} u(0) \\ \vdots \\ u(t-1) \end{bmatrix}$$

where

$$\mathcal{O}_t = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{t-1} \end{bmatrix} \quad \mathcal{T}_t = \begin{bmatrix} D & 0 & \cdots & \\ CB & D & 0 & \cdots \\ \vdots & & & \\ CA^{t-2}B & CA^{t-3}B & \cdots & CB & D \end{bmatrix}$$

- $\mathcal{O}_t$ maps initial state into resulting output over $[0, t-1]$
- $\mathcal{T}_t$ maps input to output over $[0, t-1]$

19.3 Observability matrix

19.4 Least-squares observers

19.5 Parting thoughts

19.5.1 Linear algebra

19.5.2 Levels of understanding

- High school: 17 variables, 17 equations $\rightarrow$ usually has unique solution
  - 80 vars, 60 eqns $\rightarrow$ probably 20 extra degrees of freedom
- Platonic view ('math')
  - singular, rank, range, nullspace, Jordan form, controllability
  - stuff is true or false
- Quantitative
  - based on least-squares, SVD
  - gives numerical measures for ideas like singularity, rank, etc
  - interpretation depends on (practical) context
  - very useful in practice

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20 Appendix

Some special things to remember.

20.1 Inverse, transpose properties

- $(AB)^{-1} = B^{-1}A^{-1}$
- $(A^{-1})^T = (A^T)^{-1} = A^{-T}$

20.2 Invertibility implications

For an n-by-n matrix A

Invertible	mnemonic
$\|A\| \neq 0$	$\|A\| = 0 \Rightarrow$ you can't compute the inverse - (remember base case 2×2 matrix inverse involves $1/\|A\|$ term)
non-singular	singular $\Rightarrow$ the matrix sends a nontrivial subspace to the singular subspace, $\{0\}$
A is full rank	linearly independent columns (invertibility $\Rightarrow$ 1-to-1/injective)
$\mathcal{N}(A) = \{0\}$	linearly independent columns
$\mathcal{R}(A) = \mathbb{R}^n$	linearly independent columns
$Ax = b$ has unique solution for every b	- no more than one solution (can't add members of $\mathcal{N}(A)$ for multiple b) - one solution, since $\mathcal{R}(A) = \mathbb{R}^n$; everything reachable/surjective - one solution found using the unique inverse of A
$\text{rref}(A) = I_n$	
A is a product of elementary matrices	

20.3 Tips

- to sum up all the elements in matrix A, multiply by one vectors: $\sum_{i,j}(A)_{i,j} = \mathbf{1}^T A \mathbf{1}$
- each row i in Z is a linear combination of rows $i, ..., n$ in $Y : Z = UY$, where U is upper triangular: $U_{ij} = 0$ for $i > j$
  - more in 2.24, hw2
- $E_{i,j} = e_i e_j^T \in \mathbb{R}^{n \times n}, i, j = 1, ..., n$ is $n \times n$ matrix with a 1 in i, j th entry, and zero elsewhere
  - has n^2 dimensions
- When you do a polynomial of a similarity matrix, you can pull out the outer matrices: $\mathcal{X}(T\Lambda T^{-1}) = T\mathcal{X}(\Lambda)T^{-1}$
- If $AB = 0$, and each is full rank, (neither is zero), then $\mathcal{R}(B) \subseteq \mathcal{N}(A)$
- $A^k = 0, k \geq 1 \Rightarrow$ eigenvalues of A are zero. If A is diagonalizable, then A^k can be expressed as $T\Lambda^k T^{-1}$, where Λ is diagonal, consisting of the eigenvalues multiplied by the k th power, since T is non-singular.
  - **Rank**$(A) = r \Rightarrow A \in \mathbb{R}^{n \times n}$ has exactly r non-zero eigenvalues
- Every eigenvalue of AB is an eigenvalue of BA, if $A, B \in \mathbb{R}^{n \times n}$: Given λ is eigenvalue of $AB : ABv = \lambda v, BA(Bv) = B(AB)v = \lambda Bv$. Thus Bv is an eigenvector of BA assocated with λ
- To check whether $\boxed{x \in \mathcal{R}(A)}$, check whether $\boxed{rank([A\ x]) = rank(A)}$
- Least squares approximation that minimizes x in $Ax = y, x_{ls} = (A^T A)^{-1} A^T y$ is given in MATLAB by `xls=A\y;`
  - only works if A is skinny or square

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- compute using QR (economy) factorization with

```
[Q,R]=qr(A,0); % compute economy QR decomposition
xls=R\(Q'*y);
```

- Least norm solution, $x_{ln} = A^T(AA^T)^{-1}y$, can be computed by `xln=A'*inv(A*A')*y;`
  - A cannot be full rank or skinny
  - via QR factorization:

```
[Q,R]=qr(A',0);
xln=Q*(R'\y);
```

    * for more, see here

20.4 Unrelated stuff

20.4.1 Series

$$S = \sum_{k=0}^{N} a^k = a^0 + a^1 + ... + a^N \tag{17}$$

$$aS = \sum_{k=0}^{N} a^k = a^1 + a^1 + ... + a^{N+1} \tag{18}$$

$$aS - S = -1 + a^{N+1} \tag{19}$$

$$S = \frac{a^{N+1} - 1}{a - 1} \tag{20}$$

20.4.2 Taylor/Mclauren

Intuition for approximating a function at a point (say at $x = 0$): assuming we know $f(0), f'(0), f''(0), f'''(0)$, put into polynomials

$$p_0(x) = f(0) \tag{21}$$

$$p_1(x) = f(0) + f'(0)x \tag{22}$$

$$p_2(x) = f(0) + f'(0)x + \frac{1}{2}f''(0)x^2 \tag{23}$$

$$p_3(x) = f(0) + f'(0)x + \frac{1}{2}f''(0)x^2 + \frac{1}{2}\frac{1}{3}f'''(0)x^3 \tag{24}$$

$$\tag{25}$$

Our polynomial $p_n(x)$ will both have the same value at $x = 0$, as well as the same n th derivatives when evaluated at 0. I.e.,

$$p_2'(x) = f'(0) + f''(0)x \tag{26}$$

$$p_2'(0) = f'(0) \tag{27}$$

$$\tag{28}$$

In general, the n th order McLauren approximation polynomial of $f(x)$ at point x_1 will be given as

$$f(x) \approx f(x_1) + f'(x_1)x + \frac{1}{2}f''(x_1)x^2 + \frac{1}{2 \cdot 3}f^{(3)}x^3 + ... + \frac{1}{n!}f^{(n)}x^n$$

In general, the n th order Taylor approximation polynomial of $f(x)$ at point c will be given as

$$f(x) \approx f(c) + f'(c)(x - c) + \frac{1}{2}f''(c)(x - c)^2 + ... + \frac{1}{n!}f^{(n)}(x - c)^n$$

So when you evaluate derivatives at $x = c$, the higher order terms will still drop to zero, leaving you with the exact derivative value.

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- Cosine McLauren (around 0)

$$\cos(x) = 1 - \frac{x^2}{2!} + \frac{x^4}{4!} - \frac{x^6}{6!} + ...$$

- Sine McLauren (around 0)

$$\sin(x) = x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + ...$$

- e McLauren (around 0)

$$e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + ...$$

21 Homework assignments

Homework 1	Lecture 4	2.1–2.4, 2.6, 2.9, 2.12, +
Homework 2	Lecture 6	3.2, 3.3, 3.10, 3.11, 3.16, 3.17, +
Homework 3	Lecture 8	2.17, 3.13, 4.1–4.3, 5.1, 6.9, +
Homework 4	Lecture 10	5.2, 6.2, 6.5, 6.12, 6.14, 6.26, 7.3, 8.2
Homework 5	Lecture 13	10.2, 10.3, 10.4, +
Homework 6	Lecture 14	9.9, 10.5, 10.6, 10.8, 10.14, 11.3, and 11.6a
Homework 7	Lecture 16	10.9, 10.11, 10.19, 11.13, 12.1, 13.1, +
Homework 8	Lecture 18	13.17, 14.2, 14.3, 14.4, 14.6, 14.8, 14.9, 14.11, 14.13, 14.21, 14.33, +
Homework 9	Lecture 20	14.16, 14.26, 15.2, 15.3, 15.6, 15.8, 15.10, and 15.11

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