

Sensing analysis based on plasmon induced transparency in nanocavity-coupled waveguide

Shiping Zhan, Hongjian Li,* Zhihui He, Boxun Li, Zhiquan Chen, and Hui Xu

College of Physics and Electronics, Central South University, Changsha 410083, China

* lihj398@126.com

Abstract: We report the sensing characteristic based on plasmon induced transparency in nanocavity-coupled metal-dielectric-metal waveguide analytically and numerically. A simple model for the sensing nature is first presented by the coupled mode theory. We show that the coupling strength and the resonance detuning play important roles in optimizing the sensing performance and the detection limit of sensor, and an interesting double-peak sensing is also obtained in such plasmonic sensor. In addition, the specific refractive index width of the dielectric environment is discovered in slow-light sensing and the relevant sensitivity can be enhanced. The proposed model and findings provide guidance for fundamental research of the integrated plasmonic nanosensor applications and designs.

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1. Introduction

The extraordinary property of surface plasmon polaritons (SPPs), manipulating light on nanoscale structure beyond the diffraction limit [1], provides opportunities for many fascinating researches like biomedicine [2], optical filter [3], solar cell [4] and plasmonic sensor [5]. As a promising application, the plasmonic sensors based on varied systems have attracted people's attention in the recent years [6–9]. Previous works showed the plasmon induced transparency (PIT) based sensor can be fabricated in metamaterial [7], and the PIT sensor performs better than the non-PIT one due to the steeper variation in the optical spectrum [10]. In addition, the PIT system possesses slow-light that can enhance the light-matter interaction, which can be used in sensors and slow-light devices [11, 12].

Lately, the nanoresonator-coupled metal-dielectric-metal (NCMDM) waveguide structure, with advantages of easy fabrication, light manipulation and convenient integration, has also been found available for the sensor research [13–15]. Lu et al reported a plasmonic sensor in a waveguide system and got a figure of merit (FOM) of ~500 [13], and Qi et al also observed a sensing phenomenon in the MDM waveguide structure and achieved FOMs over 650 [14]. Nevertheless, these researches mainly focused on the sensors' performance numerically, the further sensing optimization based on analytical model is seldom discussed. In addition, many reports revealed that the PIT effect can also be realized in NCMDM waveguide systems [15–18]. From the above discussion, we think that the PIT waveguide system is an ideal candidate for the integrated plasmonic sensor. Very recently, Huang et al reported a PIT-based sensor in a MDM waveguide system and found the sensing performance can be enhanced [19]. However, more efforts still need to be made in the analytical and numerical research on the PIT-based sensor and its tuning mechanism in NCMDM waveguide system.

In this paper, a simple PIT-based nanosensor is proposed and an analytical expression of figure of merit is derived for the first time to our knowledge to characterize the sensing property. The coupling strength and the resonance detuning between the cavities are useful for the sensing tuning and optimization. And an interesting double-peak sensing is obtained by choosing the detuning in a certain range. Finally, the slow-light sensing of the PIT-based sensor is discussed and the relevant sensitivity can also be regulated.

2. Structure and analytical expression

The plasmonic nanosensor and its equivalent model are shown in Fig. 1. Two cavities with length L_1 (cavity 1), L_2 (cavity 2) and width w are placed above the bus waveguide. The coupling distances are h and s , respectively. When the incident pulse comes into the waveguide, SPP wave generates in both metal-dielectric interfaces. According to the evanescent coupling, two excitation pathways interfere destructively and result in an EIT-like spectrum. Cavity 1 and 2 can also be analogically regarded as the subradiant and the superradiant mode, respectively [20]. The permittivity of the dielectric environment during the sensing process is ϵ_d . And the metal is silver, with permittivity ϵ_m defined by the Drude model [21]: $\epsilon_m(\omega) = \epsilon_\infty - \omega_p^2 / (\omega^2 + i\omega\gamma_p)$, where ω stands for the angle frequency of the incident wave, $\epsilon_\infty = 3.7$, $\omega_p = 1.38 \times 10^{16}$ rad/s is the bulk plasmon frequency, and $\gamma_p = 2.73 \times 10^{13}$ rad/s stands for the damping rate.

Based on coupled mode theory analysis [22], the stable state of cavity 2 can be described by a harmonic oscillator model as follows

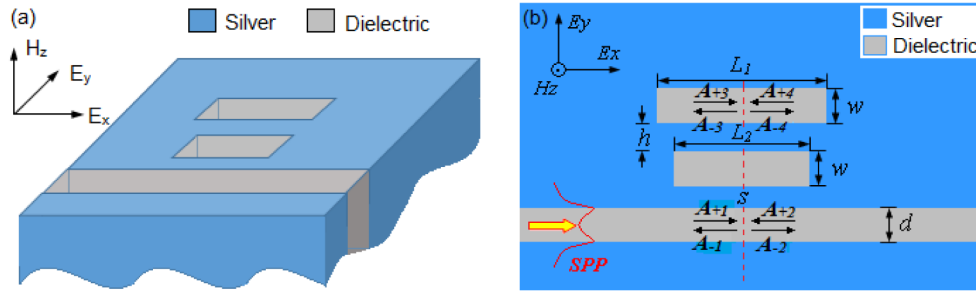


Fig. 1. The (a)schematic and (b)top view of the 2D plasmonic sensor in NCMDM waveguide system.

$$i\omega a = (i\omega_0 - 1/\tau_i - 1/\tau_w - 1/\tau_c)a + A_{+1}\sqrt{1/\tau_w} + A_{+3}\sqrt{1/\tau_c} + A_{+4}\sqrt{1/\tau_c}. \quad (1)$$

Here a represents the mode amplitude of cavity 2, ω (λ) is the angular frequency (wavelength) of the input optical pulse, ω_0 is the resonant frequency, $1/\tau_i = \omega_0/(2Q_i)$ is the internal loss, $1/\tau_w = \omega_0/(2Q_w)$ and $1/\tau_c = \omega_0/(2Q_c)$ are the coupling loss to the waveguide and cavity 1, respectively. Q_i , Q_w , Q_c are the related quality factors. The energy amplitudes in the bus waveguide and cavity 1 are depicted by $A_{\pm j}$ ($j = 1, 2, 3, 4$). And they also satisfy the following relations

$$A_{-2} = A_{+1} - \sqrt{1/\tau_w}a, A_{-3} = A_{+4} - \sqrt{1/\tau_c}a, A_{-4} = A_{+3} - \sqrt{1/\tau_c}a, \quad (2)$$

$$A_{+3} = A_{-3}C\exp(-i\varphi), A_{+4} = A_{-4}C\exp(-i\varphi), \quad (3)$$

where $\varphi = 2\pi\text{Re}(n_{\text{eff}})L_1/\lambda + \psi$ represents the phase shift for a half roundtrip in cavity 1, C is the related attenuation coefficient, and ψ is the additional phase shift by end-reflection in cavity 1. The effective refractive index n_{eff} in a MDM waveguide with width w can be obtained in [16]. Thus the transmission coefficient at frequency ω of the entire system with dielectric refractive index n , where $n = (\epsilon_d)^{1/2}$, is derived as

$$T(\omega, n) = |A_{-2}/A_{+1}|^2 = |1 - \omega_0/(2KQ_w + \omega_0)|^2, \quad (4a)$$

where

$$K(\omega, n) = i(\omega - \omega_0) + \omega_0/2Q_i + \omega_0(1 - C\exp(i\varphi))/[2Q_c(1 + C\exp(i\varphi))]. \quad (4b)$$

For numerical understanding of Eq. (4), the FDTD simulation with grid size $\Delta x = \Delta y = 2.5\text{nm}$ is provided for structure with $w = d = 50\text{nm}$, $s = 20\text{nm}$, $h = 40\text{nm}$, $L_1 = 415\text{nm}$ and $L_2 = 400\text{nm}$ [23]. In Fig. 2(a), a typical EIT-like feature (blue cycles) is observed, where a transparency peak at 674nm is located between two dips at 650nm and 691.6nm [10]. That is consistent with the analytical one (red solid lines) based on Eq. (4).

Thus, the transmission at frequency ω for the system with dielectric material refractive index $n + \Delta n$ can be derived as

$$T(\omega, n + \Delta n) = \left| 1 - \frac{\omega_1 / 2Q_w'}{i(\omega - \omega_1) + \omega_1 / 2Q_i' + \omega_1 / 2Q_w' + (\omega_1 / 2Q_c)(1 - Ce^{i\varphi'}) / (1 + Ce^{i\varphi'})} \right|^2, \quad (5)$$

where $\omega_1 = \omega_0 / (n + \Delta n)$, Q_i' and Q_w' are the resonance frequency and quality factors when the environment refractive index increases from n to $n + \Delta n$, while φ' is the half roundtrip phase shift in cavity 1. It's worthy to notice that those values are obtained due to the approximation that the line shape of the optical response spectrum nearly maintains the same, only leading to a red-shift [13, 14], and it's also been proved by the later analysis.

Based on the definition of figure of merit [8] and the consideration that detection used in sensor, such as biosensor, usually by measuring the light intensity variation for one particular wavelength, here we give a precise analytical expression for FOM at frequency ω as

$$FOM(\omega) = \frac{\Delta T}{T \Delta n} = \frac{T(\omega, n + \Delta n) - T(\omega, n)}{T(\omega, n) \Delta n}, \quad (6)$$

where $T(\omega, n)$ and $T(\omega, n + \Delta n)$ have been obtained in Eq. (4) and (5), ΔT is the intensity variation at frequency ω caused by environment refractive index change Δn . The simulated sensing response for the PIT-based sensor in Fig. 2(a) is shown in Fig. 2(b). The spectrum

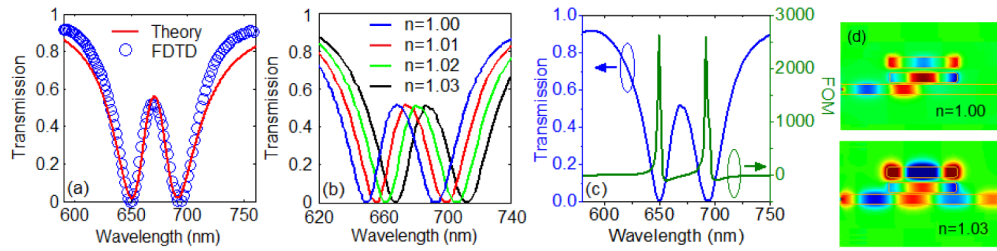


Fig. 2. (a) Simulated transmission (blue cycles) and theoretical fitting (red solid lines) as $\omega_0 = 2.81 \times 10^{15} \text{rad/s}$, $Q_c = 100$, $Q_w = 10$, $Q_i = 339$, $C = 0.95$, $\varphi = 4.81 \text{rad}$. (b) Transmission for with varied n . (c) Transmission and FOM as functions of wavelength. (d) Magnetic field distributions of incident pulse at 691.6nm for $n = 1.00$ and $n = 1.03$.

shifts to longer wavelength and almost keeps the line-shape. A FOM of ~ 2600 at 691.6nm is obtained, which is much higher than that in [13, 14], here Δn is chosen to be 0.01 . Figure 2(c) shows the transmission and FOM as functions of wavelength. The maximum FOM nearly appears at the transmission dips, which can be explained by the lowest initial intensity T here. The magnetic field distribution of the incident pulse at 691.6nm with $n = 1.00$ and 1.03 are displayed in Fig. 2(d). As the surrounding refractive index increases, the plasmonic sensor also shows a switching property.

3. Numerical results

To verify the availability of the analytical sensing expression in Eq. (6), we first focus on the coupling strength between the two cavities, which is represented by Q_c . Figure 3(a) is the theoretical evolution of FOM with Q_c , here we take $\Delta n = 0.01$ and the structure parameters used in Fig. 2(a). It can be observed that the FOM maximum for the left dip (dip 1) increases

gradually while decreases for the right one (dip 2) when Q_c gets enlarged, and a highest value reaches almost 820. This is because the lower coupling strength makes the closer transmission dips, thus the ΔT s caused by Δn for the two dips vary differently, which result in this phenomenon. Numerical analysis obtained by FDTD method is provided to support this explanation. The simulated intensity variations at dips caused by $\Delta n = 0.01$ for different coupling distance h are shown in Fig. 3(b). When h changes from 32 to 64nm, indicating the enlarged Q_c , the ΔT increases for dip 1 (black squares) while decreases for the second (red circles). Since the dip transmissions are quite small, we make them constant values approximately, which leads to the proportional relationship between FOM and ΔT . Thus, as Q_c increases, the opposite variation of ΔT leads to the different trend of FOM, which agrees with the theoretical results in Fig. 3(a). For further understanding the impact of the coupling strength, we discuss the lossless situation of subradiant cavity for $C = 1$, shown in Fig. 3(c). Different from Fig. 3(a), the FOM in lossless case almost maintains the same for the first dip but shows a nonmonotonic change for the second one when Q_c increases. We find lowering the attenuation in the subradiant resonator can also adjust the sensing property. Figure 3(d)-(f) are the transmission spectra for varied coupling strength. It is found that the transmission change ΔT for the right dip firstly increases from 0.165 to 0.374 but then drops to 0.206, thus the related FOM behaves a nonmonotonic variation according to the former discussion. In addition, we know for realistic sensor device, there should be a detection limit [24]. That means when the sensing is under certain condition, the plasmonic sensor can show good performance. But if the sensing work is beyond that condition, the sensor will lose efficacy. Such as when T is quite small, the sensor may not be able to detect small refractive index change, which leads to a detection limit. Here we use the definition of detection limit in Ref [24], as: $\Delta n_{lim} = (\Delta n / \Delta T) \Delta T_{lim}$, where Δn_{lim} represents the smallest refractive index change of the tested material or analyte, ΔT and Δn have been defined above, while ΔT_{lim} is the minimum transmission change that can be detected, which is determined by the optical detector. We also define a modified figure of merit as $FOM_m = \Delta T / \Delta n$ accordingly. Can the structure parameter also affect the detection limit of the PIT-based sensor? Now we discuss this issue. For sensing analysis of dip 1, the detection limit Δn_{lim} and FOM_m as functions of the coupling distance h are shown in Fig. 3(g). A tradeoff is found between the Δn_{lim} and the FOM_m due to the fact that the detection limit Δn_{lim} is inversely proportional to $\Delta T / \Delta n$. The Δn_{lim} decreases while the FOM_m increases as the coupling gets weak. A highest Δn_{lim} of $\sim 0.137 \Delta T_{lim}$ is obtained under strong coupling with $h = 32$ nm, while the FOM_m reaches its maximum of ~ 20 under weak coupling with $h = 64$ nm. Since the ΔT_{lim} is usually a constant that determined by the optical detector, thus the coefficient in front of ΔT_{lim} can properly describe the detection limit of a plasmonic sensor. However, for dip 2, the highest Δn_{lim} of $\sim 0.313 \Delta T_{lim}$ is obtained under weak coupling with $h = 64$ nm, while the FOM_m reaches its maximum of ~ 14 under strong coupling with $h = 32$ nm, shown in Fig. 3(h), which is opposite to the results of dip 1. This can also be simply explained by the relationship between the intensity variations ΔT and the coupling distance h in Fig. 3(b). These interesting findings can be helpful in optimizing the sensing property and the detection limit of PIT-based sensor by tuning the coupling strength between the superradiant and the subradiant cavities.

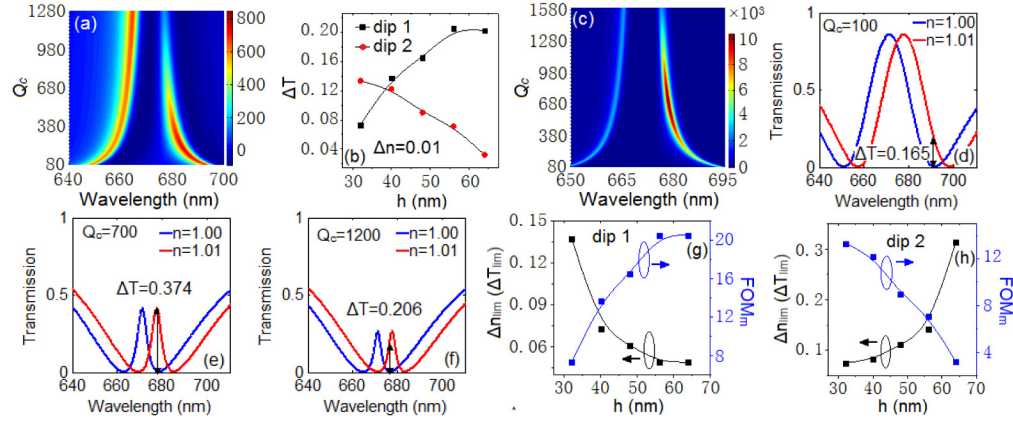


Fig. 3. The evolution of FOM with Q_c for $\Delta n = 0.01$ in (a) lossy and (c) lossless system. (b) Intensity variation ΔT at dips caused by $\Delta n = 0.01$ for different coupling distance h . (d)-(f) The transmission spectra for $n = 1.00$ and 1.01 with $Q_c = 100, 700$ and 1200 , respectively. (g) and (h) are Δn_{lim} and FOM_m as functions of coupling distance h .

Besides the coupling strength, the resonance detuning between the two cavities, tuned by length L_I , also can affect the sensing characteristic. Figure 4(a) is the FOM evolution with cavity length L_I . The white dash line represents $L_I = 415\text{nm}$ which guarantees the symmetric transmission spectrum. When L_I gets larger (smaller) than 415nm , only one obvious left (right) sensing peak appears. Particularly, when $L_I = 415\text{nm}$, two comparable sensing peaks with FOM of nearly 700 are found. We name it as a double-peak (DP) sensing phenomenon. The double-peak sensing may perform more functionally than the conventional single-peak sensing. This interesting phenomenon was also found in [14]. Figure 4(b) is the simulated intensity variation ΔT caused by $\Delta n = 0.01$ for different L_I . It can be seen that the difference between the two ΔT s gets enlarged when L_I becomes larger or smaller than 415nm . Thus there should be an operating width of L_I that ensures the DP sensing. However, the attenuation of subradiant cavity makes such width quite small, only about several nanometers. If the width can be broadened, it could be beneficial for the sensor fabrication and meanwhile lower the technique difficulty. Figure 4(c) shows the evolution of FOM with L_I in the ideal situation ($C = 1$) for different Q_c . For potential applications, high FOM is required, we choose the minimum of 500 in the color label for discussion. When $Q_c = 100$, the DP sensing range is observed from $L_I = 376$ to 473nm , resulting in an operating width of 97nm . Here some outer part of Fig. 4(c) is not given for clarity. Similarly, the operating DP width for $Q_c = 200$ and 400 are calculated as 66 and 43nm , respectively. It's found that the coupling strength can affect the width and a stronger coupling makes a wider operating range, however, a relatively lower FOM. Later on, we briefly discuss the influence of coefficient C , shown in Fig. 4(d). The lower attenuation results in the larger DP width. And the coupling strength can also be regarded as the further tuning factor. This can be helpful for the design and improvement of highly tunable sensor with multi-sensing application. In addition, similar with the discussion in Fig. 3(g)-3(h), the detection limit Δn_{lim} and the FOM_m of the plasmonic sensor can also be regulated by the resonance detuning between superradiant and the subradiant cavities, shown in Fig. 4(e)-4(f). For sensing analysis of dip 1, the Δn_{lim} increases while the FOM_m decreases as L_I gets weak. A highest Δn_{lim} of $\sim 0.2\Delta T_{lim}$ is obtained when $L_I = 435\text{nm}$, while the FOM_m reaches its maximum of ~ 24 for $L_I = 395\text{nm}$. For dip 2, however, the highest Δn_{lim} of $\sim 0.294\Delta T_{lim}$ is obtained for $L_I = 395\text{nm}$, while the modified figure of merit gets a peak value of ~ 40 . This can be simply explained by the Fano-shaped spectra that results from the resonance detuning between the superradiant and the subradiant modes [13]. These results can be useful in optimizing the sensing property and the detection limit by tuning the resonance detuning in PIT-based sensor.

Since the sensor is based on a PIT system, thus besides of the conventional sensing above, a slow-light sensing [12, 25] should also be realized. Attributing to the steep phase dispersion in transparency window, the group velocity of light can be slowed down significantly and results in a time delay. The decrease of group velocity allows not only the light to ‘feel’ changes in the refractive index for a longer time, but also the enhancement of light intensity by the pulse compression [26], give rise to good sensing performance. The group index N_g can be calculated through $N_g = c/v_g = c/H \cdot \tau_g = c/H \cdot (d\theta(\omega)/d\omega)$ [27], where v_g is the group velocity, τ_g is delay time, and the phase shift $\theta(\omega)$ is the function of angular frequency ω , $H = 1000\text{nm}$ is the length of the plasmonic system. Figure 5 (a) is the group index for $Q_c = 600$ and $L_l = 415\text{nm}$ with $n = 1.000, 1.005$ and 1.012 . When $n = 1.005$, the pulse centered at 679nm has a group index of 2 as it passes through the system. This means the light speed is halved, and the group index (or intensity) of the pulse increases as n increases, but reaches a maximum of ~ 22 when $n = 1.012$. Here we consider $N_g \geq 2$ as slow-light region and define the sensing width with slow-light (SWS) as $\text{SWS} = 1.012 - 1.005 = 0.007$. The SWS reveals the available detection limit of a sensor for slow-light sensing. Figure 5(b) shows the SWS as a function of L_l for different Q_c . It is found that both the resonance detuning and the coupling strength can adjust the SWS of plasmonic sensor. From $L_l = 400$ to 415nm , the SWS decreases gradually and lower Q_c leads to

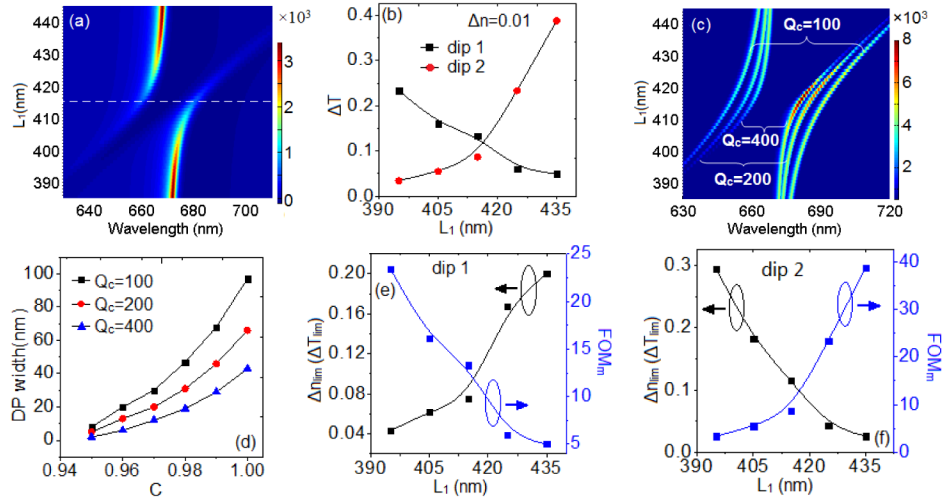


Fig. 4. The evolution of FOM (a) with L_l in the lossy case of $C = 0.95$, (b) The simulated transmission change ΔT at two dips as a function of L_l . (c) The evolution of FOM with L_l for different Q_c in the lossless case of $C = 1$, (d) The relationship between DP width and attenuation coefficient C for different Q_c . (e) and (f) are Δn_{lim} and FOM_m as functions of cavity length L_l .

larger quantity for each L_l . The difference at $L_l = 395\text{nm}$ may result from the large resonance detuning. For sensing application, higher SWS guarantees wider refractive index range of the dielectric environment or the tested analyte. Based on the figure of merit analysis by M. Povinelli [28], we further give the sensitivity of system with slow-light at frequency ω as $\text{FOM}_{N_g}(\omega) = [N_g(\omega, n + \Delta n) - N_g(\omega, n)]n / (N_g(\omega, n)\Delta n)$, here Δn is chosen to be the SWS. Figure 5(c) depicts the relationship between the sensing width and the FOM_{N_g} for structures with varied L_l when $Q_c = 600$. A tradeoff exists apparently and higher sensitivity makes smaller sensing width. The balance between these two factors can be interesting and will be discussed

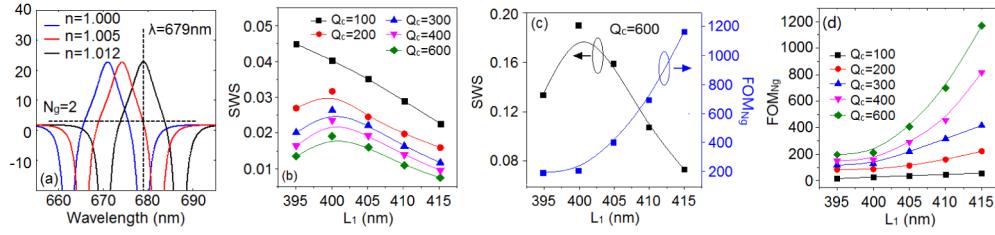


Fig. 5. (a) The group index for $Q_c = 600$, $L_1 = 415\text{nm}$ with different refractive index, (b) The SWS as a function of L_1 with different Q_c , (c) The relationship between the SWS and the FOM_{Ng} for varied L_1 when $Q_c = 600$. (d) FOM_{Ng} for varied L_1 with different Q_c .

in the future. The detuning of resonance can also affect the SWS and the FOM_{Ng} , lower detuning results in a higher sensitivity and a maximum of ~ 1200 is achieved when $L_1 = 415\text{nm}$ for the nearly zero detuning. The relation between L_1 and FOM_{Ng} under varied coupling strength is displayed in Fig. 5(d). For each constant coupling strength, the FOM_{Ng} increases as L_1 gets larger, but the slope strongly depends on the quality factor Q_c . This is because the SWS decreases while the maximum group index $N_g(\omega, n + \Delta n)$ increases when enlarging L_1 , and the slow-light sensitivity is proportional to the $N_g(\omega, n + \Delta n)$ but inversely proportional to the SWS according to the definition. Moreover, the overall sensitivities for higher Q_c are larger than those for smaller Q_c which may be attributed to the steeper dispersion in transparency windows. According to the discussion above, the slow-light sensing in the PIT-based sensor is found enhanced by weak coupling and small detuning between the superradiant cavity and the subradiant cavity. These findings may provide help for the theoretical and application research of plasmonic nanosensor with slow-light.

4. Conclusions

In summary, the sensing property of PIT-based sensor in nanocavity-coupled MDM waveguide is investigated, and a simple analytical expression is presented for discussion. It is found that the sensing performance and the detection limit of our proposed sensor can be adjusted by the coupling strength, resonance detuning and dissipation, the double-peak sensing phenomenon can be observed within certain detuning. Finally, the refractive index range for the slow-light sensing is discussed and the related sensitivity is realized adjustable. This work may provide guidance for fundamental applications and designs of plasmonic nanosensor.

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