

- State-space to matrix fraction

Given $\dot{x} = Ax + Bu$

$$y = Cx + Du \quad \nearrow D=0$$

$$G(s) = C(sI - A)^{-1}B + \cancel{D} = D_L^{-1}(s) N_L(s) = N_R(s) D_R^{-1}(s)$$

D, N_L, D_R &
polynomial
matrices

where

$$D_L = I - C(sI - A + JC)^{-1}J \quad \}$$

$$N_L = C(sI - A + JC)^{-1}B \quad \}$$

where

$$A - JC \text{ stable}$$

$$D_R = I - k(sI - A + BK)^{-1}B \quad \}$$

$$N_R = C(sI - A + BK)^{-1}B \quad \}$$

where

$$A - BK \text{ stable}$$

Further, using K, J above, define

$$Y_L = k(sI - A + BK)^{-1}J$$

$$X_L = I + C(sI - A + BK)^{-1}J$$

$$Y_R = k(sI - A + JC)^{-1}J$$

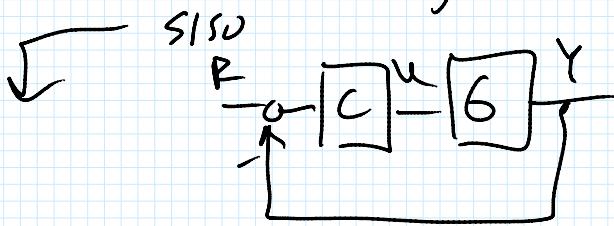
$$X_R = I + k(sI - A + JC)^{-1}B$$

Then $\begin{bmatrix} Y_R & X_R \\ -N_R & D_R \end{bmatrix} \begin{bmatrix} D_L & -X_L \\ N_L & Y_L \end{bmatrix} = \begin{bmatrix} I & 0 \\ 0 & I \end{bmatrix}$

Diagonal terms

$$\begin{aligned} Y_L D_L + X_R N_L &= I \\ D_R Y_R + N_R X_L &= I \end{aligned} \quad \left. \begin{array}{l} \text{Bezout} \\ \text{equations} \end{array} \right.$$

MIMO Block Diagrams



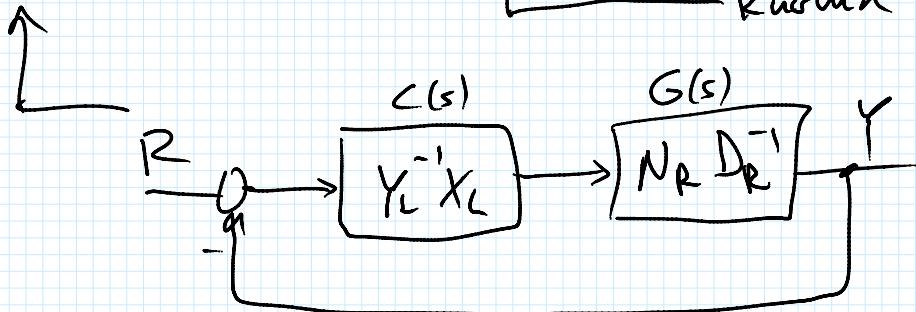
$$G = \frac{N(s)}{D(s)}$$

$$C = \frac{X(s)}{Y(s)}$$

$$\frac{E(s)}{C(s)} = K_P + \frac{K_1}{s} = \frac{K_P s + K_1}{s}$$

$$\frac{Y}{R} = \frac{C \cdot G}{1 + C \cdot G} = \frac{N(s) X(s)}{D(s) Y(s) + N(s) X(s)}$$

↑ ↑ ↑
To find K₁ K_P



from earlier

$$Y = G \cdot C (I + GC)^{-1} R$$

$$= G (I + CG)^{-1} C R$$

$$Y = N_R (Y_L D_R + X_L N_R)^{-1} X_L R$$

compare to SISO

$$y = n(dy + nx)^{-1} x \cdot R$$

call $Y_L D_R + X_L N_R = D(s)$

let $\Delta_d(s) = \text{desired } \Delta(s)$

\Rightarrow Given $D_r, N_r, \Delta_d(s)$

Find X_L, Y_L

so that $\Delta(s) = \Delta_d(s)$

Solvable if (A, B, C) is cont/des

i.e., $G(s)$ is minimal

In MIMO case $G(s)$ minimal means no common factors between $D_r(s)$ and $N_r(s)$

matrices

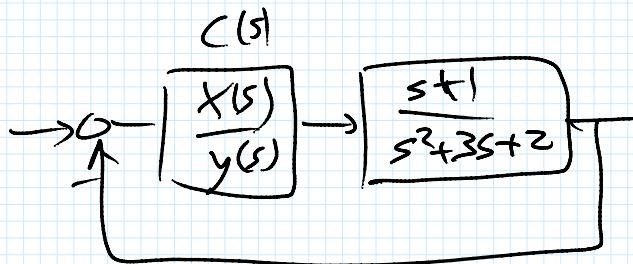
common factors means there is

a matrix $U(s)$ so that $D_r^{-1} = U^{-1} \bar{D}_r^{-1}$

$$D_r = \bar{D}_r(s) U(s) \quad G(s) = N_r D_r^{-1}$$

$$N_r = \bar{N}_r(s) U(s)$$

Ex $\Sigma(s)$



$$G = \frac{(s+1)}{(s+1)(s+2)} = \frac{1}{s+2}$$

$$G(s) = \frac{N(s)}{D(s)} \text{ where } N(s) = (s+1) \\ D(s) = s^3 + 3s^2 \\ = (s+1)(s+2)$$

common factor $(s+1)$

for this feedback system

characteristic
polynomial

$$s(s) = y(s) \cdot d(s) + x(s) n(s)$$

$$= y(s)(s+1)(s+2) + x(s)(s+1)$$

$$= (s+1) \{ y(s)(s+2) + x(s) \}$$

common factor equivalent to
uncontrollable eigenvalue
in A matrix

\Rightarrow have to be working with a
minimized, irreducible system to
design controllers

when N, D have no common
factors they are called co-prime

6.4 Poles & Zeros of MIMO Systems

for (A, B, C, D) it's easy: poles = eigenvalues

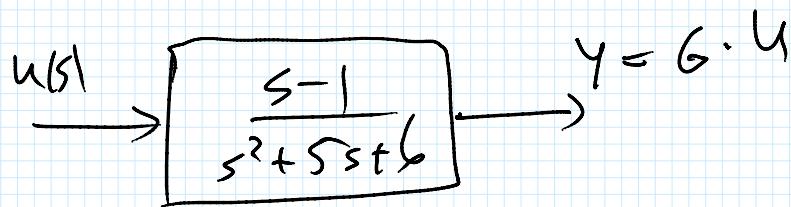
- SISO case

$$G(s) = \frac{n(s)}{d(s)}$$

← zeros = roots of $n(s)$
(zeros)

→ poles = roots of $d(s)$
(zeros)

- poles determine stability, modes of system
- zeros? define signals that can't pass through system



$$\begin{aligned} u(t) &= e^{st} + 2u \\ u(s) &= \frac{1}{s-1} \end{aligned} \Rightarrow y = \frac{s+1}{(s-1)} \cdot \frac{1}{s+1} = \frac{1}{s^2 + 5s + 6}$$

$$s \in \mathbb{C} \rightarrow 0$$

but $u(t) \rightarrow \infty$
input is blocked

- MIMO case

Given matrix $G(s)$, $s = c$ is a zero
of $G(s)$ if there exists constant vector

N, w so that $G(c)v=0$ and $wG(c)=0$
 i.e., $|G(s)|_{s=c}$ loses rank

$$\text{Ex. } G(s) = \begin{bmatrix} \frac{4}{(s+1)(s+4)} & \frac{-1}{s+1} \\ \frac{2}{s+1} & \frac{-1}{2(s+1)(s+2)} \end{bmatrix}$$

Not: $\underset{s=-3}{G(-3)} = \begin{bmatrix} 2 & 1/2 \\ -1 & -1/4 \end{bmatrix}$ has rank 1
 $\Rightarrow -3$ is a zero

can also do this with matrix fractions

$$G = N_R D_R^{-1} = D_L^{-1} N_L$$

zeros are values of s where N_R (and N_L) lose rank

• Poles of MIMO systems

with matrix fraction, poles are where D_R (and D_L) lose rank

in terms of $G(s)$, define poles as the roots of pd_e polynomial

roots of Pde polynomial

least common denominator of
all minors of $G(s)$

determinants of all $m \times m$
submatrices of a matrix