

Universidad de Costa Rica

Facultad de Ingeniería

Escuela de Ingeniería Eléctrica

Laboratorio de microcontroladores

Laboratorio 1:

Introducción a microcontroladores y manejo de GPIOS

Daniela Oporta Abarca B55160

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1. Introducción

El presente informe corresponde a la documentación correspondiente a la realización de una tómbola simplificada de bingo donde por medio de la aplicación de GPIOs, el Microcontrolador PIC12f675 y distintos elementos electrónicos pasivos se establece un esquemático de circuito realizado por medio del software Simulide en conjunto con Lenguaje de alto nivel C para generar números de 2 dígitos aleatorios que serán respresentados por medio del método BCD de 7 segmentos en dos pantallas de la misma configuración para así generar los números del juego bingo. El código y esquemático poseen un botón y únicamente 16 bolas con números por lo que el juego se reinicia luego de reproducir el valor 99 en las pantallas de display led. A continuación se presentará la información técnica relevante respecto a hojas de datos del fabricante de los elementos, lógica de hardware y flujo en C para la solución de problemas por medio de la generación de firmware para Microcontroladores.

2. Nota teórica

Se utiliza el microcontrolador PIC12f675 el cual posee las siguientes características:

- 8 pines de input/output
- Un conversor A/D de 10 bits
- 64 bytes de memoria SRAM
- 128 bytes de memoria EEPROM
- 1 comparador
- Timer 8/16-bit 1/1
- Oscilador en corriente directa de 20MHz
- Rango de corriente
- Comparador análogo
- Oscilador interno
- Una corriente máxima de entrada en el pin V_{dd} de 250mA
- Una corriente de salida máxima en el pin V_{ss} de 300mA
- Una tensión de operación entre un rango de 2,2 V hasta 5,5 V

La topología del microcontrolador corresponde a la siguiente:

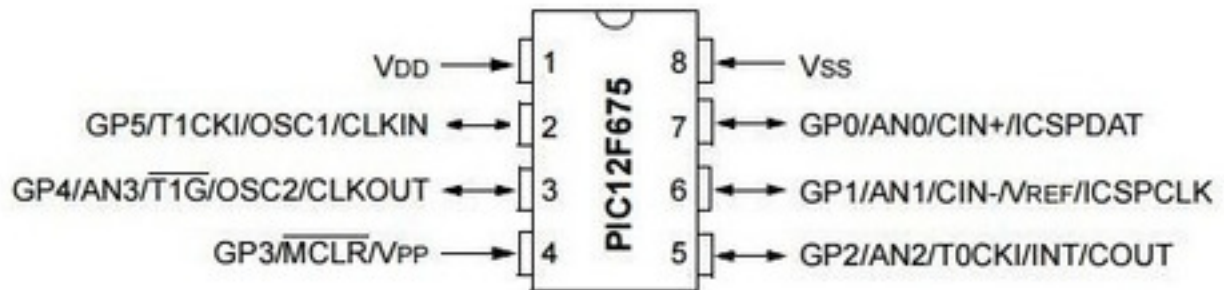


Figura 1: Microcontrolador PIC12f675

En la figura 5 se observan los 8 pines presentes en el PIC12f675 en donde se utilizan los registros de operación digital mencionados a continuación:

- Registro Trisio, registro para el modo de operación individual de cada uno de los pines, con valores en alto iguales a 1 o en bajo iguales a 0, en estos los 1 corresponden a entradas y los 0 a salidas.
- Registro GPIO de lectura de estado de pin y escritura al latch de salida. En este los 5 bits menos significativos 5:0 si están en 1 sucede porque el puerto de pin es mayor al umbral de tensión V_{IH} y si tienen por otro lado un valor 0 el puerto de pin es menor al valor de tensión umbral V_{IH} .
- Registros ANSEL o CMCON de inicialización para configuración del canal analógico y la entrada digital. El registro CMCOM registro debe ser modificado al utilizar los pines GP0, GP1 y GP2 como entradas. El ANSEL se configura para conversión y selección analógica y además debe ser modificado al utilizar cualquier pin como entrada.
- Registro CONFIG, de configuración de pines.

Se presentan a continuación los registros manipulados:

Dirección	Nombre Registro	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
05h	GPIO	-	-	GP5	GP4	GP3	GP2	GP1	GP0
85h	TRISIO	-	-	TRISIO05	TRISIO04	TRISIO03	TRISIO02	TRISIO01	TRISIO00

Figura 2: Tabla con registros utilizados, Creación propia.

Además se presentan las especificaciones eléctricas en la siguiente figura correspondiente a la hoja de datos del fabricante con las especificaciones del microcontrolador:

12.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings†

Ambient temperature under bias.....	-40 to +125°C
Storage temperature	-65°C to +150°C
Voltage on VDD with respect to VSS	-0.3 to +6.5V
Voltage on MCLR with respect to Vss	-0.3 to +13.5V
Voltage on all other pins with respect to Vss	-0.3V to (VDD + 0.3V)
Total power dissipation ⁽¹⁾	800 mW
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, I _{IK} (V _I < 0 or V _I > VDD).....	± 20 mA
Output clamp current, I _{OK} (V _O < 0 or V _O > VDD).....	± 20 mA
Maximum output current sunk by any I/O pin.....	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by all GPIO	125 mA
Maximum current sourced all GPIO	125 mA

Figura 3: MIC

Se hace uso además de resistencias y capacitores para configurar la lectura digital por medio de un switch donde se genere un cambio de estado al presionar un botón.

3. Desarrollo

Para el MIC utilizado en el desarrollo del esquemático se tomaron en cuenta situaciones específicas del modelo PIC12f675 para evitar sobre calentamiento y sobre cargas en los elementos del circuito.

Por medio del método BCD 7 segmentos se estableció el flujo del laboratorio, aplicando el uso del método se utilizaron elementos del software SimulIDE para generar la visibilización mediante 4 puertos de salida únicamente. El método de display de 7 segmentos corresponde a un sistema de decodificación BCD hacia 7 LEDS de forma alargada que conforman el número "8".^{en} un display, dicho número debido a que por medio del encendido y apagado de los LEDS es posible representar todos los números naturales por estos 7 segmentos. Existe el método de cátodo común en el cual los cátodos de todos los LEDS alargados se encuentran conectados y se deben conectar a tierra en el circuito para visualizar los números por la entrada de

voltaje que se le aplique, para el laboratorio se utilizó el método de cátodo común. Debido a las escasas salidas disponibles en el PIC12f765 se utiliza el método de decodificación del software llamado BcdTo7S-1185 donde por medio del mismo es posible convertir números binarios conformados por 4 bits a señales digitales de 7 segmentos que controlan el display de LED para la muestra del número.

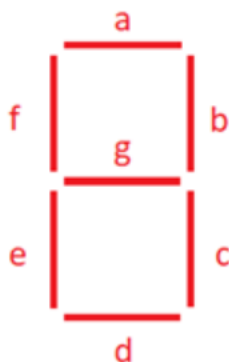


Figura 4: Display 7 segmentos. Tomado de [2].

Como se puede observar el "8" cuenta con los segmentos: a, b, c, d, e, f y g los cuales son los que se deben de activar y desactivar de una forma estandarizada respecto al tráfico de los bits para lograr el display. Para el caso de este laboratorio se utilizaron 2 pantallas de este tipo. La tabla de verdad con la que se establecieron los 10 posibles números naturales corresponde a la siguiente:

Numero Decimal	A	B	C	D	a	b	c	d	e	f	g
0	0	0	0	0	1	1	1	1	1	1	0
1	0	0	0	1	0	1	1	0	0	0	0
2	0	0	1	0	1	1	0	1	1	0	1
3	0	0	1	1	1	1	1	1	0	0	1
4	0	1	0	0	0	1	1	0	0	1	1
5	0	1	0	1	1	0	1	1	0	1	1
6	0	1	1	0	1	0	1	1	1	1	1
7	70	1	1	1	1	1	1	0	0	0	0
8	1	0	0	0	1	1	1	1	1	1	1
9	1	0	0	1	1	1	1	0	0	1	1

De la cual representamos números de dos dígitos mediante 8 bits de entrada que generan las 2 combinaciones de salidas para los 7 segmentos de los 2 displays 7S BCD. En la realización del código se debió tomar en cuenta cómo en este modelo específico se tiene el registro de configuración fuera del espacio de memoria del programa y pertenece al de configuración de memoria accedido durante la programación.

Para la configuración del MIC se inicializaron los registros importantes como entradas o salidas, por convención de programación se establecieron los pines GP0, GP1, GP2 y GP4 todos como salidas al designarlos como 0x00.

Luego, se realizó la configuración de cada pin de acuerdo con la capacidad eléctrica de cada uno para garantizar un comportamiento adecuado en el circuito, esto de acuerdo con la imagen 3. A la topología del circuito además se le agregó un pull down resistor esto para generar una lectura digital de tipo pull down en donde el estado sea bajo a menos que se cambie intencionalmente a alto por medio de una entrada lógica positiva.

El flujo de tareas para la creación del dado se creó con la inicialización de periféricos y el microcontrolador para luego dictaminar las instrucciones que utilizaría el MCU en forma cíclica, se establecieron ciclos anidados para abastecer necesidades de alternancia por la poca memoria disponible en el MIC.

Para la implementación se utilizó el algoritmo por software con contadores para la generación de un número aleatorio donde por medio de un contador for desde 0 se crea un ciclo que recorre números de 2 dígitos hasta 100, esto debido a que nuestro diseño al contar con un botón para aprobar el paso de la tensión de la fuente se logra crear una condición de parada en un valor dentro del rango del for no esperado ni con posibilidad de ser previsualizado por el usuario dando funcionalidad de "random". La naturaleza de ciclos anidados se da debido a que el ciclo for se encuentra dentro de un ciclo while (1) que es infinitamente recorrido mientras el circuito se encuentre en modo encendido, donde el for inicia dentro de este while(1) y dentro de un while con condición de que se tiene un contador menor a 17 el cual cada vez que se presiona el botón se le agrega un 1 al contador para cumplir con el aspecto de la tómbola de tener únicamente 16 bolas. A continuación se demuestra lo explicado:

```
void ejecucion()
{
    unsigned int time = 200;
    //int random = 0;
    int contador=0;

    while ( 1 )
    {
        while(contador!=17)
        {
            //opciones de numeros infinitas entre el 0 y el 9
            for (int random = 0; random < 100; random++){
                GP4 = 0;
                if(GP5 == 1)
                {
                    int unidad = random%10;
                    int dec = (int)random/10;

                    contador+=1;
                }
            }
        }
    }
}
```

Figura 5: Funciones anidadas. Creación propia.

Se implementó además una función de tiempo para programar un retraso de modo que la

visibilidad de las luces LED sean percibidas por el ojo humano. Para el main se estableció el GP5 como la única entrada del MCU respecto al trisio, se declaró también una variable de tiempo para visibilidad de las luces. El registro CONFIG no se declara. Se declararon los auxiliares contadores i y j para la función de delay de tiempo además.

```
TRISIO = 0b00100000 ;  
GPIO = 0x00 ;  
ANSEL = 0 ;  
int tiempo = 200 ;  
//CONFIG = 0 ;  
int i ;  
int j ;
```

En el main se establece una llamada a la función ejecución donde se tiene el cuerpo del programa manejado por medio de casos en el switch donde de acuerdo a la tabla BCD 7S si se quiere por ejemplo transmitir un número 9 porque es el valor random dado por el programa se tendrán las siguientes salidas:

GP0=1,GP1=0,GP2=0 y GP4=1 los cuales encienden los segmentos a,b,c,d,f y g que corresponden al número 9 en display.

Para la conexión del decodificador se establecieron las entradas en función de las salidas del PIC y sus salidas corresponden a los segmentos, luego se colocó una resistencia para evitar sobrecalentamiento de acuerdo a los valores recomendados para el dispositivo entre el decodificador y el display. Se conectaron 2 displays con distinta conexión que generan números aleatorios no repetidos y a las 16 pulsiones de botón demuestran un 99, esto por medio del uso de delay para generar la visualización desde el ojo humano de que se están generando al mismo tiempo los números, ya que no es posible crear paralelismo por carencia de memoria.

A continuación se presenta el esquemático del circuito.

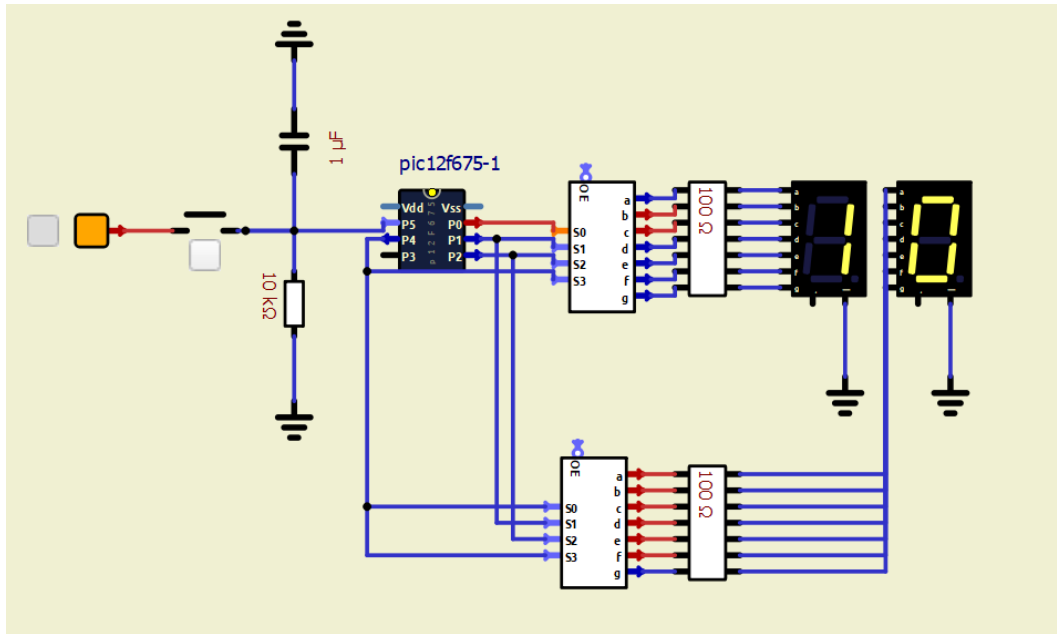


Figura 6: Resultados Cara 1

3.1. Precios de componentes

Además se presenta la lista de precios de los componentes utilizados:

Componente	Valor (\$)
Capacitor	0.38
Resistencia	0.5
Pulsador	0.4
Microcontrolador	2
Decodificador	10
Display BCD	6

3.2. Análisis de resultados

Luego de configurar el software y hardware se realizaron pruebas donde al encender el circuito se reflejaban de forma correcta los números sin ser repetidos en los display CBD 7s, para efectos prácticos se realizaron inicialmente pruebas donde se reestableciera el circuito para 4 pulsiones de botón al detallar que todo estaba corriendo correctamente se aumentaron las mismas a 16 donde se generó un reset forzoso por medio del llamado a la función main(), creando una corrida del código desde el inicio. A continuación se presentan pruebas de los valores random obtenidos:

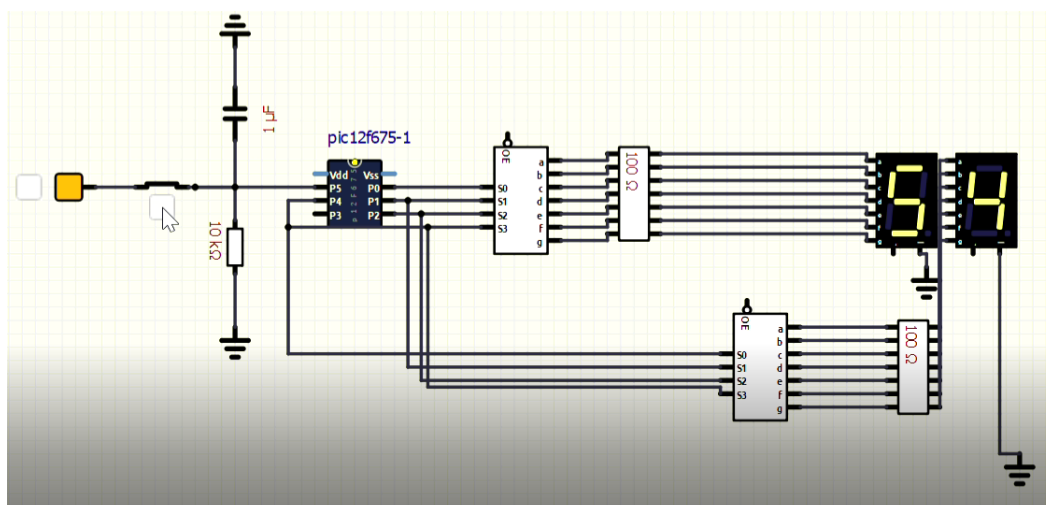


Figura 7: Resultados prueba 1. Creación propia.

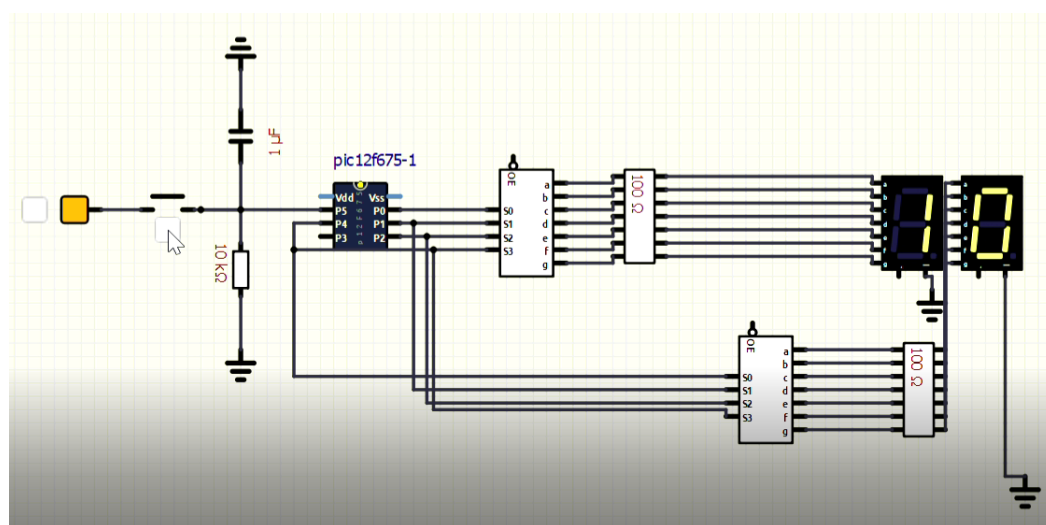


Figura 8: Resultados prueba 1. Creación propia.

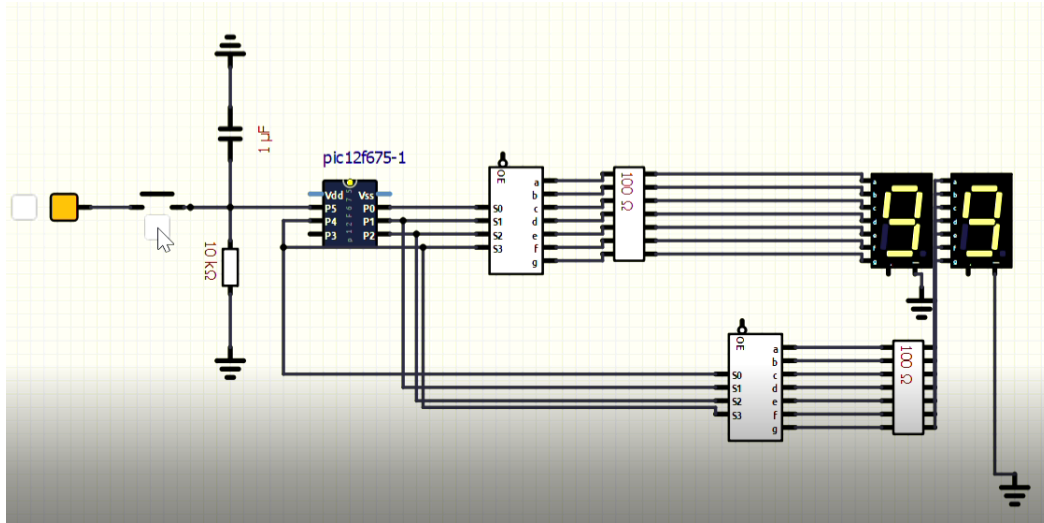


Figura 9: Resultados prueba 17. Creación propia.

Se denota como al encender el circuito en el simulador y presionar el switch se reflejan las luces con los resultados esperados, además se nota como el simulador puede presentar problemas si se tiene el botón presionado más tiempo del necesario ya que crea un efecto de cable en el que el delay toma más tiempo en ser reflejado alargando el tiempo de "apagado" hacia el ojo humano.

Como observación adicional, al tener un microcontrolador con escasez en memoria al compilar no existían errores en la lógica sin embargo para varias variaciones del código planteadas el sistema indicaba errores de falta de memoria para poder llegar a cabo las acciones, para el caso final del código generado no existieron más sin embargo sí warnings respecto al alocamiento de memoria.

```
warning: Relocation of section "UDL_display_0" failed, relocating to a shared memory location.
warning: Relocation of section "UDL_idata_0" failed, relocating to a shared memory location.
warning: Relocation of section "UDL_divsint_0" failed, relocating to a shared memory location.
warning: Relocation of section "UDL_divuint_0" failed, relocating to a shared memory location.
warning: Relocation of section "IDD_display_0" failed, relocating to a shared memory location.
warning: Relocation of section "IDD_idata_0" failed, relocating to a shared memory location.
```

Figura 10: Observaciones en el código. Creación propia.

4. Conclusiones

La utilización del software aproxima a los estudiantes en la práctica a la utilización de microcontroladores simulando por medio de una implementación híbrida de código y hardware así como la búsqueda de formas innovadoras de abordar problemas no conocidos anteriormente en el ámbito ingenieril.

El uso de periféricos de tamaño limitado de entradas y salidas proporcionaron a los estudiantes oportunidad de implementar conocimientos correspondientes a circuitos digitales para por medio de compuertas lógicas llegar a lo solicitado tomando de 4 salidas 7 diferentes

opciones a ser reflectadas por medio de los LEDS. Esto gracias a los decodificadores del software, además se aprendió una forma nueva de visualizar leds para representar símbolos por medio de su combinación dependiente de salidas controladas por el MIC.

Se implementó lenguaje C para cargar un archivo .hex de modo que se controló un MIC por código fuente de alto nivel como firmware.

Se tomaron en cuenta además los datos de importancia de la hoja del fabricante para el MIC específico utilizado en este laboratorio (PIC12F675) para así operar el circuito respetando umbrales y parámetros para un correcto funcionamiento aplicando conocimientos de laboratorio eléctrico.

4.1. Recomendaciones

Se recomienda para la creación del esquemático planificar cómo generar salidas por medio de la tabla combinacional, además de crear diagramas de flujo para la creación de cualquier código a programar especialmente al utilizar recursividad repetitivamente. Además, se recomienda trabajar a base de la hoja del fabricante como buena práctica ya que el uso de simulador nos exime de utilizar hardware físico sin embargo se implementa con el objetivo de crear una correcta implementación evitando a futuro dañar equipo físico así como en las mismas se tiene la información importante de registros y manejo de los microcontroladores.

5. Bibliografía

MSc. Marco Villalta Fallas, GPIOs y el PIC12F629. Escuela de Ingeniería Eléctrica, 2022.

“Decodificador BCD a 7 segmentos”, Ciencia y Tecnología, 12-oct-2020. [En línea]. Disponible en: <https://cienciayt.com/electronica/sistemas-digitales/decodificador-bcd-a-7-segmentos/>. [Consultado: 02-sep-2022].

6. Git-Avances.

Link del repositorio Git: <https://github.com/danielaoporta/src>

7. Apéndices



PIC12F629/675

Data Sheet

8-Pin FLASH-Based 8-Bit
CMOS Microcontrollers

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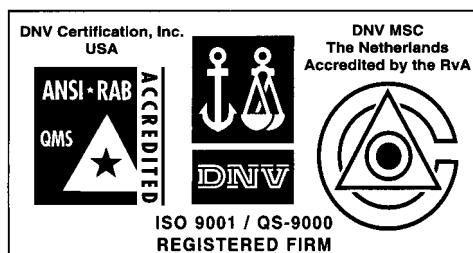
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8-Pin FLASH-Based 8-Bit CMOS Microcontroller

High Performance RISC CPU:

- Only 35 instructions to learn
 - All single cycle instructions except branches
- Operating speed:
 - DC - 20 MHz oscillator/clock input
 - DC - 200 ns instruction cycle
- Interrupt capability
- 8-level deep hardware stack
- Direct, Indirect, and Relative Addressing modes

Special Microcontroller Features:

- Internal and external oscillator options
 - Precision Internal 4 MHz oscillator factory calibrated to $\pm 1\%$
 - External Oscillator support for crystals and resonators
 - 5 μ s wake-up from SLEEP, 3.0V, typical
- Power saving SLEEP mode
- Wide operating voltage range - 2.0V to 5.5V
- Industrial and Extended temperature range
- Low power Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Brown-out Detect (BOD)
- Watchdog Timer (WDT) with independent oscillator for reliable operation
- Multiplexed $\overline{\text{MCLR}}$ /Input-pin
- Interrupt-on-pin change
- Individual programmable weak pull-ups
- Programmable code protection
- High Endurance FLASH/EEPROM Cell
 - 100,000 write FLASH endurance
 - 1,000,000 write EEPROM endurance
 - FLASH/Data EEPROM Retention: > 40 years

Low Power Features:

- Standby Current:
 - 1 nA @ 2.0V, typical
- Operating Current:
 - 8.5 μ A @ 32 kHz, 2.0V, typical
 - 100 μ A @ 1 MHz, 2.0V, typical
- Watchdog Timer Current
 - 300 nA @ 2.0V, typical
- Timer1 oscillator current:
 - 4 μ A @ 32 kHz, 2.0V, typical

Peripheral Features:

- 6 I/O pins with individual direction control
- High current sink/source for direct LED drive
- Analog comparator module with:
 - One analog comparator
 - Programmable on-chip comparator voltage reference (CVREF) module
 - Programmable input multiplexing from device inputs
 - Comparator output is externally accessible
- Analog-to-Digital Converter module (PIC12F675):
 - 10-bit resolution
 - Programmable 4-channel input
 - Voltage reference input
- Timer0: 8-bit timer/counter with 8-bit programmable prescaler
- Enhanced Timer1:
 - 16-bit timer/counter with prescaler
 - External Gate Input mode
 - Option to use OSC1 and OSC2 in LP mode as Timer1 oscillator, if INTOSC mode selected
- In-Circuit Serial Programming™ (ICSP™) via two pins

Device	Program Memory	Data Memory		I/O	10-bit A/D (ch)	Comparators	Timers 8/16-bit
	FLASH (words)	SRAM (bytes)	EEPROM (bytes)				
PIC12F629	1024	64	128	6	—	1	1/1
PIC12F675	1024	64	128	6	4	1	1/1

* 8-bit, 8-pin devices protected by Microchip's Low Pin Count Patent: U.S. Patent No. 5,847,450. Additional U.S. and foreign patents and applications may be issued or pending.

PIC12F629/675

Pin Diagrams

8-pin PDIP, SOIC, DFN-S

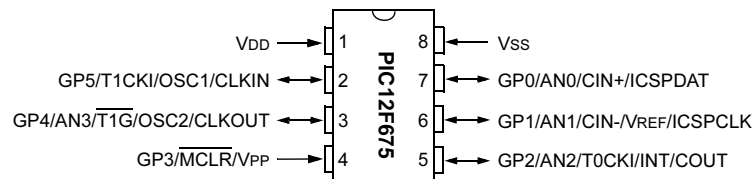
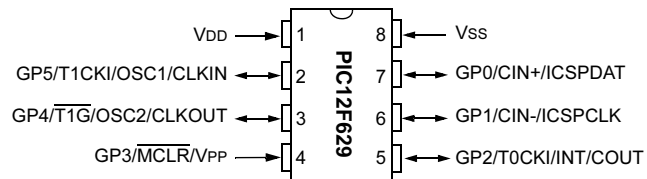


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PIC12F629/675

NOTES:

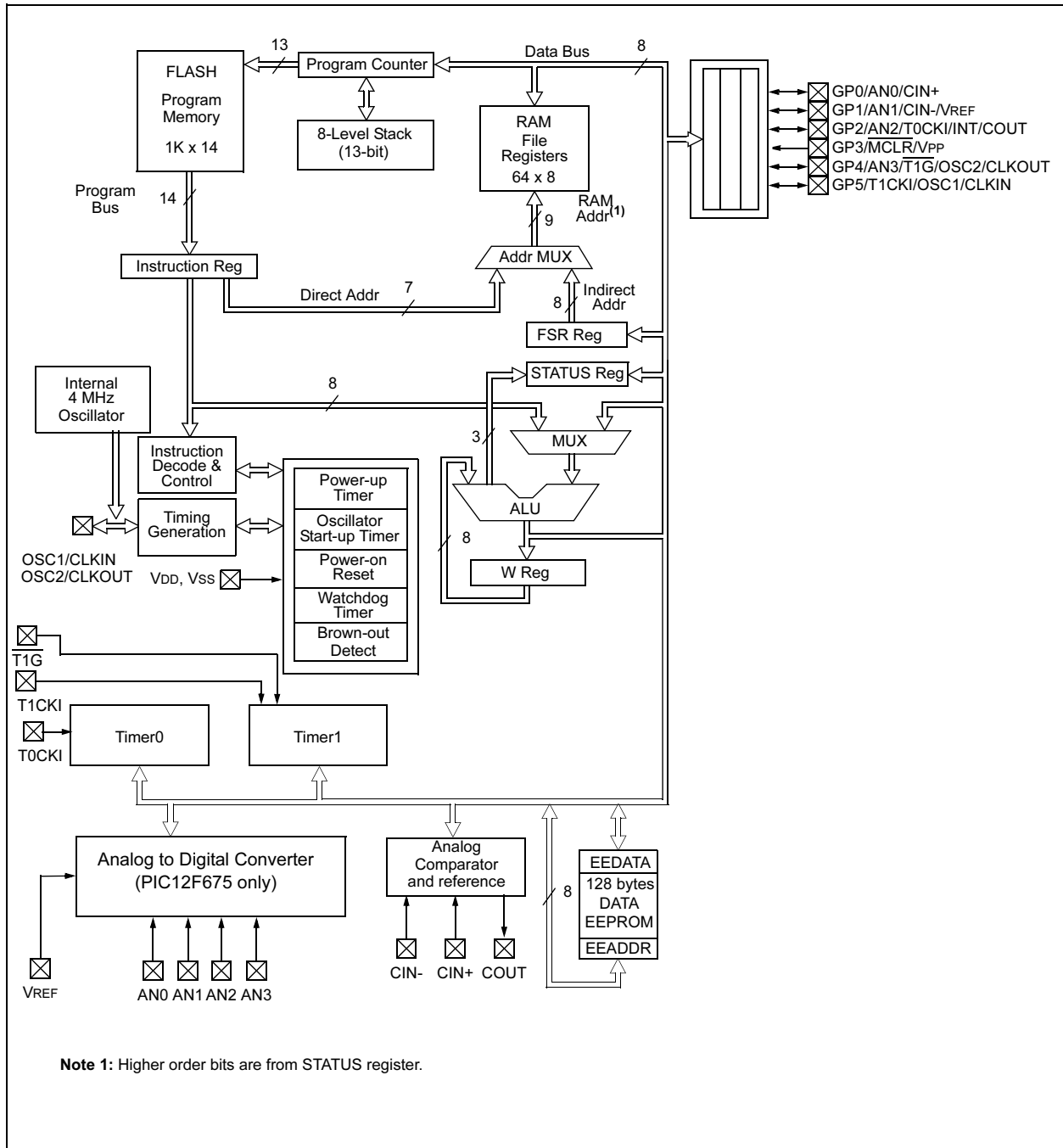
1.0 DEVICE OVERVIEW

This document contains device specific information for the PIC12F629/675. Additional information may be found in the PICmicro™ Mid-Range Reference Manual (DS33023), which may be obtained from your local Microchip Sales Representative or downloaded from the Microchip web site. The Reference Manual should be considered a complementary document to this Data

Sheet, and is highly recommended reading for a better understanding of the device architecture and operation of the peripheral modules.

The PIC12F629 and PIC12F675 devices are covered by this Data Sheet. They are identical, except the PIC12F675 has a 10-bit A/D converter. They come in 8-pin PDIP, SOIC, and MLF-S packages. Figure 1-1 shows a block diagram of the PIC12F629/675 devices. Table 1-1 shows the Pinout Description.

FIGURE 1-1: PIC12F629/675 BLOCK DIAGRAM



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TABLE 1-1: PIC12F629/675 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
GP0/AN0/CIN+/ICSPDAT	GP0	TTL	CMOS	Bi-directional I/O w/ programmable pull-up and interrupt-on-change
	AN0	AN		A/D Channel 0 input
	CIN+	AN		Comparator input
	ICSPDAT	TTL	CMOS	Serial programming I/O
GP1/AN1/CIN-/VREF/ICSPCLK	GP1	TTL	CMOS	Bi-directional I/O w/ programmable pull-up and interrupt-on-change
	AN1	AN		A/D Channel 1 input
	CIN-	AN		Comparator input
	VREF	AN		External voltage reference
	ICSPCLK	ST		Serial programming clock
GP2/AN2/T0CKI/INT/COUT	GP2	ST	CMOS	Bi-directional I/O w/ programmable pull-up and interrupt-on-change
	AN2	AN		A/D Channel 2 input
	T0CKI	ST		TMR0 clock input
	INT	ST		External interrupt
	COUT		CMOS	Comparator output
GP3/MCLR/VPP	GP3	TTL		Input port w/ interrupt-on-change
	MCLR	ST		Master Clear
	VPP	HV		Programming voltage
GP4/AN3/T1G/OSC2/CLKOUT	GP4	TTL	CMOS	Bi-directional I/O w/ programmable pull-up and interrupt-on-change
	AN3	AN		A/D Channel 3 input
	T1G	ST		TMR1 gate
	OSC2		XTAL	Crystal/resonator
	CLKOUT		CMOS	Fosc/4 output
GP5/T1CKI/OSC1/CLKIN	GP5	TTL	CMOS	Bi-directional I/O w/ programmable pull-up and interrupt-on-change
	T1CKI	ST		TMR1 clock
	OSC1	XTAL		Crystal/resonator
	CLKIN	ST		External clock input/RC oscillator connection
Vss	Vss	Power		Ground reference
VDD	VDD	Power		Positive supply

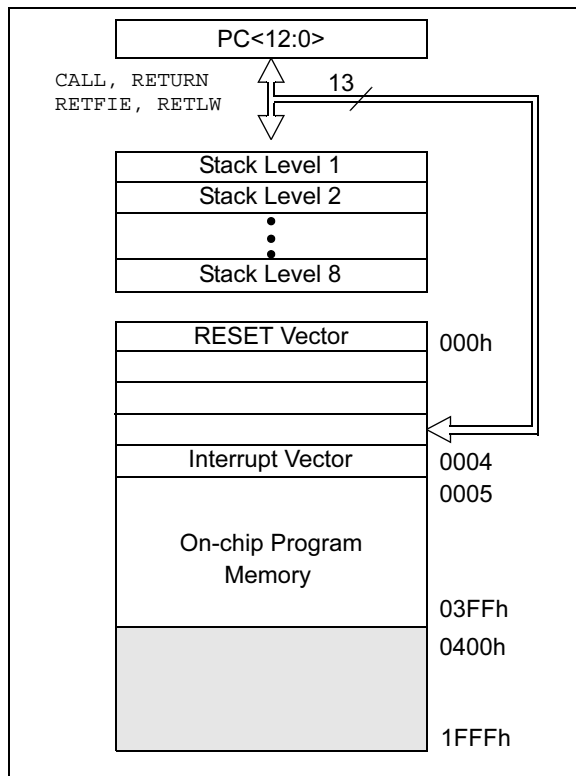
Legend: Shade = PIC12F675 only
TTL = TTL input buffer, ST = Schmitt Trigger input buffer

2.0 MEMORY ORGANIZATION

2.1 Program Memory Organization

The PIC12F629/675 devices have a 13-bit program counter capable of addressing an 8K x 14 program memory space. Only the first 1K x 14 (0000h - 03FFh) for the PIC12F629/675 devices is physically implemented. Accessing a location above these boundaries will cause a wrap around within the first 1K x 14 space. The RESET vector is at 0000h and the interrupt vector is at 0004h (see Figure 2-1).

FIGURE 2-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC12F629/675



2.2 Data Memory Organization

The data memory (see Figure 2-2) is partitioned into two banks, which contain the General Purpose registers and the Special Function registers. The Special Function registers are located in the first 32 locations of each bank. Register locations 20h-5Fh are General Purpose registers, implemented as static RAM and are mapped across both banks. All other RAM is unimplemented and returns '0' when read. RP0 (STATUS<5>) is the bank select bit.

- RP0 = 0 Bank 0 is selected
- RP0 = 1 Bank 1 is selected

Note: The IRP and RP1 bits STATUS<7:6> are reserved and should always be maintained as '0's.

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file is organized as 64 x 8 in the PIC12F629/675 devices. Each register is accessed, either directly or indirectly, through the File Select Register FSR (see Section 2.4).

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2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function registers are registers used by the CPU and peripheral functions for controlling the desired operation of the device (see Table 2-1). These registers are static RAM.

The special registers can be classified into two sets: core and peripheral. The Special Function registers associated with the “core” are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

FIGURE 2-2: DATA MEMORY MAP OF THE PIC12F629/675

File Address		File Address	
Indirect addr. ⁽¹⁾	00h	Indirect addr. ⁽¹⁾	80h
TMR0	01h	OPTION_REG	81h
PCL	02h	PCL	82h
STATUS	03h	STATUS	83h
FSR	04h	FSR	84h
GPIO	05h	TRISIO	85h
	06h		86h
	07h		87h
	08h		88h
	09h		89h
PCLATH	0Ah	PCLATH	8Ah
INTCON	0Bh	INTCON	8Bh
PIR1	0Ch	PIE1	8Ch
	0Dh		8Dh
TMR1L	0Eh	PCON	8Eh
TMR1H	0Fh		8Fh
T1CON	10h	OSCCAL	90h
	11h		91h
	12h		92h
	13h		93h
	14h		94h
	15h	WPU	95h
	16h	IOC	96h
	17h		97h
	18h		98h
CMCON	19h	VRCON	99h
	1Ah	EEDATA	9Ah
	1Bh	EEADR	9Bh
	1Ch	EECON1	9Ch
	1Dh	EECON2 ⁽¹⁾	9Dh
ADRESH ⁽²⁾	1Eh	ADRESL ⁽²⁾	9Eh
ADCON0 ⁽²⁾	1Fh	ANSEL ⁽²⁾	9Fh
	20h		A0h
General Purpose Registers 64 Bytes		accesses 20h-5Fh	
	5Fh		DFh
	60h		E0h
	7Fh		FFh
Bank 0		Bank 1	


 Unimplemented data memory locations, read as '0'.
 1: Not a physical register.
 2: PIC12F675 only.

TABLE 2-1: SPECIAL FUNCTION REGISTERS SUMMARY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Page
Bank 0											
00h	INDF ⁽¹⁾	Addressing this Location uses Contents of FSR to Address Data Memory								0000 0000	18,59
01h	TMR0	Timer0 Module's Register								xxxx xxxx	27
02h	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	17
03h	STATUS	IRP ⁽²⁾	RP1 ⁽²⁾	RP0	\overline{TO}	\overline{PD}	Z	DC	C	0001 1xxx	11
04h	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	18
05h	GPIO	—	—	GPIO5	GPIO4	GPIO3	GPIO2	GPIO1	GPIO0	--xx xxxx	19
06h	—	Unimplemented								—	—
07h	—	Unimplemented								—	—
08h	—	Unimplemented								—	—
09h	—	Unimplemented								—	—
0Ah	PCLATH	—	—	—	Write Buffer for Upper 5 bits of Program Counter				---	0 0000	17
0Bh	INTCON	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF	0000 0000	13
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	15
0Dh	—	Unimplemented								—	—
0Eh	TMR1L	Holding Register for the Least Significant Byte of the 16-bit Timer1								xxxx xxxx	30
0Fh	TMR1H	Holding Register for the Most Significant Byte of the 16-bit Timer1								xxxx xxxx	30
10h	T1CON	—	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	TMR1CS	TMR1ON	-000 0000	32
11h	—	Unimplemented								—	—
12h	—	Unimplemented								—	—
13h	—	Unimplemented								—	—
14h	—	Unimplemented								—	—
15h	—	Unimplemented								—	—
16h	—	Unimplemented								—	—
17h	—	Unimplemented								—	—
18h	—	Unimplemented								—	—
19h	CMCON	—	COUT	—	CINV	CIS	CM2	CM1	CM0	-0-0 0000	35
1Ah	—	Unimplemented								—	—
1Bh	—	Unimplemented								—	—
1Ch	—	Unimplemented								—	—
1Dh	—	Unimplemented								—	—
1Eh	ADRESH ⁽³⁾	Most Significant 8 bits of the Left Shifted A/D Result or 2 bits of the Right Shifted Result								xxxx xxxx	42
1Fh	ADCON0 ⁽³⁾	ADFM	VCFG	—	—	CHS1	CHS0	$\overline{GO/DONE}$	ADON	00-- 0000	43,59

Legend: — = unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented

Note 1: This is not a physical register.

2: These bits are reserved and should always be maintained as '0'.

3: PIC12F675 only.

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TABLE 2-1: SPECIAL FUNCTION REGISTERS SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Page
Bank 1											
80h	INDF ⁽¹⁾	Addressing this Location uses Contents of FSR to Address Data Memory								0000 0000	18,59
81h	OPTION_REG	GPPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	12,28
82h	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	17
83h	STATUS	IRP ⁽²⁾	RP1 ⁽²⁾	RP0	T0	PD	Z	DC	C	0001 1xxx	11
84h	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	18
85h	TRISIO	—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0	--11 1111	19
86h	—	Unimplemented								—	—
87h	—	Unimplemented								—	—
88h	—	Unimplemented								—	—
89h	—	Unimplemented								—	—
8Ah	PCLATH	—	—	—	Write Buffer for Upper 5 bits of Program Counter					---0 0000	17
8Bh	INTCON	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF	0000 0000	13
8Ch	PIE1	EEIE	ADIE	—	—	CMIE	—	—	TMR1IE	00-- 0--0	14
8Dh	—	Unimplemented								—	—
8Eh	PCON	—	—	—	—	—	—	POR	BOD	---- --0x	16
8Fh	—	Unimplemented								—	—
90h	OSCCAL	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	—	—	1000 00--	16
91h	—	Unimplemented								—	—
92h	—	Unimplemented								—	—
93h	—	Unimplemented								—	—
94h	—	Unimplemented								—	—
95h	WPU	—	—	WPU5	WPU4	—	WPU2	WPU1	WPU0	--11 -111	20
96h	IOC	—	—	IOC5	IOC4	IOC3	IOC2	IOC1	IOC0	--00 0000	21
97h	—	Unimplemented								—	—
98h	—	Unimplemented								—	—
99h	VRCON	VREN	—	VRR	—	VR3	VR2	VR1	VR0	0-0- 0000	40
9Ah	EEDATA	Data EEPROM Data Register								0000 0000	47
9Bh	EEADR	—	Data EEPROM Address Register							-000 0000	47
9Ch	EECON1	—	—	—	—	WRERR	WREN	WR	RD	---- x000	48
9Dh	EECON2 ⁽¹⁾	EEPROM Control Register 2								---- ----	48
9Eh	ADRESL ⁽³⁾	Least Significant 2 bits of the Left Shifted A/D Result of 8 bits or the Right Shifted Result								xxxx xxxx	42
9Fh	ANSEL ⁽³⁾	—	ADCS2	ADCS1	ADCS0	ANS3	ANS2	ANS1	ANS0	-000 1111	44,59

Legend: — = unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition,
shaded = unimplemented

- Note 1:** This is not a physical register.
Note 2: These bits are reserved and should always be maintained as '0'.
Note 3: PIC12F675 only.

2.2.2.1 STATUS Register

The STATUS register, shown in Register 2-1, contains:

- the arithmetic status of the ALU
- the RESET status
- the bank select bits for data memory (SRAM)

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper three bits and set the Z bit. This leaves the STATUS register as 000u u1uu (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any STATUS bits. For other instructions not affecting any STATUS bits, see the "Instruction Set Summary".

Note 1: Bits IRP and RP1 (STATUS<7:6>) are not used by the PIC12F629/675 and should be maintained as clear. Use of these bits is not recommended, since this may affect upward compatibility with future products.

2: The C and DC bits operate as a Borrow and Digit Borrow out bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.

REGISTER 2-1: STATUS — STATUS REGISTER (ADDRESS: 03h OR 83h)

Reserved	Reserved	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	TO	PD	Z	DC	C
bit 7							bit 0

- bit 7 **IRP:** This bit is reserved and should be maintained as '0'
- bit 6 **RP1:** This bit is reserved and should be maintained as '0'
- bit 5 **RP0:** Register Bank Select bit (used for direct addressing)
0 = Bank 0 (00h - 7Fh)
1 = Bank 1 (80h - FFh)
- bit 4 **TO:** Time-out bit
1 = After power-up, CLRWDI instruction, or SLEEP instruction
0 = A WDT time-out occurred
- bit 3 **PD:** Power-down bit
1 = After power-up or by the CLRWDI instruction
0 = By execution of the SLEEP instruction
- bit 2 **Z:** Zero bit
1 = The result of an arithmetic or logic operation is zero
0 = The result of an arithmetic or logic operation is not zero
- bit 1 **DC:** Digit carry/borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions)
For borrow, the polarity is reversed.
1 = A carry-out from the 4th low order bit of the result occurred
0 = No carry-out from the 4th low order bit of the result
- bit 0 **C:** Carry/borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions)
1 = A carry-out from the Most Significant bit of the result occurred
0 = No carry-out from the Most Significant bit of the result occurred

Note: For borrow the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either the high or low order bit of the source register

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

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2.2.2.2 OPTION Register

The OPTION register is a readable and writable register, which contains various control bits to configure:

- TMR0/WDT prescaler
- External GP2/INT interrupt
- TMR0
- Weak pull-ups on GPIO

Note: To achieve a 1:1 prescaler assignment for TMR0, assign the prescaler to the WDT by setting PSA bit to '1' (OPTION<3>). See Section 4.4.

REGISTER 2-2: OPTION_REG — OPTION REGISTER (ADDRESS: 81h)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
GPPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

- bit 7 **GPPU:** GPIO Pull-up Enable bit
1 = GPIO pull-ups are disabled
0 = GPIO pull-ups are enabled by individual port latch values
- bit 6 **INTEDG:** Interrupt Edge Select bit
1 = Interrupt on rising edge of GP2/INT pin
0 = Interrupt on falling edge of GP2/INT pin
- bit 5 **T0CS:** TMR0 Clock Source Select bit
1 = Transition on GP2/T0CKI pin
0 = Internal instruction cycle clock (CLKOUT)
- bit 4 **T0SE:** TMR0 Source Edge Select bit
1 = Increment on high-to-low transition on GP2/T0CKI pin
0 = Increment on low-to-high transition on GP2/T0CKI pin
- bit 3 **PSA:** Prescaler Assignment bit
1 = Prescaler is assigned to the WDT
0 = Prescaler is assigned to the TIMER0 module
- bit 2-0 **PS2:PS0:** Prescaler Rate Select bits

Bit Value	TMR0 Rate	WDT Rate
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

2.2.2.3 INTCON Register

The INTCON register is a readable and writable register, which contains the various enable and flag bits for TMR0 register overflow, GPIO port change and external GP2/INT pin interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-3: INTCON — INTERRUPT CONTROL REGISTER (ADDRESS: 0Bh OR 8Bh)

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF
bit 7							bit 0	

bit 7

GIE: Global Interrupt Enable bit
1 = Enables all unmasked interrupts
0 = Disables all interrupts

bit 6

PEIE: Peripheral Interrupt Enable bit
1 = Enables all unmasked peripheral interrupts
0 = Disables all peripheral interrupts

bit 5

T0IE: TMR0 Overflow Interrupt Enable bit
1 = Enables the TMR0 interrupt
0 = Disables the TMR0 interrupt

bit 4

INTE: GP2/INT External Interrupt Enable bit
1 = Enables the GP2/INT external interrupt
0 = Disables the GP2/INT external interrupt

bit 3

GPIE: Port Change Interrupt Enable bit⁽¹⁾
1 = Enables the GPIO port change interrupt
0 = Disables the GPIO port change interrupt

bit 2

T0IF: TMR0 Overflow Interrupt Flag bit⁽²⁾
1 = TMR0 register has overflowed (must be cleared in software)
0 = TMR0 register did not overflow

bit 1

INTF: GP2/INT External Interrupt Flag bit
1 = The GP2/INT external interrupt occurred (must be cleared in software)
0 = The GP2/INT external interrupt did not occur

bit 0

GPIF: Port Change Interrupt Flag bit
1 = When at least one of the GP5:GP0 pins changed state (must be cleared in software)
0 = None of the GP5:GP0 pins have changed state

Note 1: IOC register must also be enabled to enable an interrupt-on-change.

2: T0IF bit is set when TIMER0 rolls over. TIMER0 is unchanged on RESET and should be initialized before clearing T0IF bit.

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

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2.2.2.4 PIE1 Register

The PIE1 register contains the interrupt enable bits, as shown in Register 2-4.

Note: Bit PEIE (INTCON<6>) must be set to enable any peripheral interrupt.

REGISTER 2-4: PIE1 — PERIPHERAL INTERRUPT ENABLE REGISTER 1 (ADDRESS: 8Ch)

R/W-0	R/W-0	U-0	U-0	R/W-0	U-0	U-0	R/W-0
EEIE	ADIE	—	—	CMIE	—	—	TMR1IE
bit 7				bit 0			

- bit 7 **EEIE:** EE Write Complete Interrupt Enable bit
1 = Enables the EE write complete interrupt
0 = Disables the EE write complete interrupt
- bit 6 **ADIE:** A/D Converter Interrupt Enable bit (PIC12F675 only)
1 = Enables the A/D converter interrupt
0 = Disables the A/D converter interrupt
- bit 5-4 **Unimplemented:** Read as '0'
- bit 3 **CMIE:** Comparator Interrupt Enable bit
1 = Enables the comparator interrupt
0 = Disables the comparator interrupt
- bit 2-1 **Unimplemented:** Read as '0'
- bit 0 **TMR1IE:** TMR1 Overflow Interrupt Enable bit
1 = Enables the TMR1 overflow interrupt
0 = Disables the TMR1 overflow interrupt

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

2.2.2.5 PIR1 Register

The PIR1 register contains the interrupt flag bits, as shown in Register 2-5.

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-5: PIR1 — PERIPHERAL INTERRUPT REGISTER 1 (ADDRESS: 0Ch)

R/W-0	R/W-0	U-0	U-0	R/W-0	U-0	U-0	R/W-0
EEIF	ADIF	—	—	CMIF	—	—	TMR1IF
bit 7				bit 0			

- bit 7 **EEIF:** EEPROM Write Operation Interrupt Flag bit
 1 = The write operation completed (must be cleared in software)
 0 = The write operation has not completed or has not been started
- bit 6 **ADIF:** A/D Converter Interrupt Flag bit (PIC12F675 only)
 1 = The A/D conversion is complete (must be cleared in software)
 0 = The A/D conversion is not complete
- bit 5-4 **Unimplemented:** Read as '0'
- bit 3 **CMIF:** Comparator Interrupt Flag bit
 1 = Comparator input has changed (must be cleared in software)
 0 = Comparator input has not changed
- bit 2-1 **Unimplemented:** Read as '0'
- bit 0 **TMR1IF:** TMR1 Overflow Interrupt Flag bit
 1 = TMR1 register overflowed (must be cleared in software)
 0 = TMR1 register did not overflow

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

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2.2.2.6 PCON Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-on Reset (POR)
- Brown-out Detect (BOD)
- Watchdog Timer Reset (WDT)
- External MCLR Reset

The PCON Register bits are shown in Register 2-6.

REGISTER 2-6: PCON — POWER CONTROL REGISTER (ADDRESS: 8Eh)

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-x
—	—	—	—	—	—	POR	BOD
bit 7						bit 0	

bit 7-2 **Unimplemented:** Read as '0'

bit 1 **POR:** Power-on Reset STATUS bit

1 = No Power-on Reset occurred

0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

bit 0 **BOD:** Brown-out Detect STATUS bit

1 = No Brown-out Detect occurred

0 = A Brown-out Detect occurred (must be set in software after a Brown-out Detect occurs)

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

2.2.2.7 OSCCAL Register

The Oscillator Calibration register (OSCCAL) is used to calibrate the internal 4 MHz oscillator. It contains 6 bits to adjust the frequency up or down to achieve 4 MHz.

The OSCCAL register bits are shown in Register 2-7.

REGISTER 2-7: OSCCAL — OSCILLATOR CALIBRATION REGISTER (ADDRESS: 90h)

R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	U-0
CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	—	—
bit 7						bit 0	

bit 7-2 **CAL5:CAL0:** 6-bit Signed Oscillator Calibration bits

111111 = Maximum frequency

100000 = Center frequency

000000 = Minimum frequency

bit 1-0 **Unimplemented:** Read as '0'

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

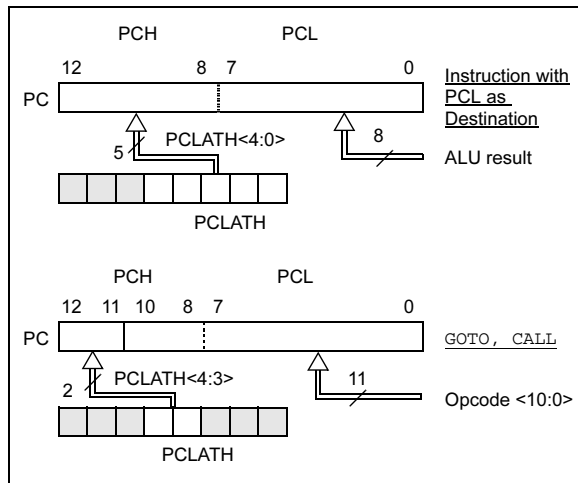
'0' = Bit is cleared

x = Bit is unknown

2.3 PCL and PCLATH

The program counter (PC) is 13-bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any RESET, the PC is cleared. Figure 2-3 shows the two situations for the loading of the PC. The upper example in Figure 2-3 shows how the PC is loaded on a write to PCL (PCLATH<4:0> → PCH). The lower example in Figure 2-3 shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> → PCH).

FIGURE 2-3: LOADING OF PC IN DIFFERENT SITUATIONS



2.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When performing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to the Application Note "Implementing a Table Read" (AN556).

2.3.2 STACK

The PIC12F629/675 family has an 8-level deep x 13-bit wide hardware stack (see Figure 2-1). The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed, or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1: There are no STATUS bits to indicate stack overflow or stack underflow conditions.

2: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions, or the vectoring to an interrupt address.

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2.4 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

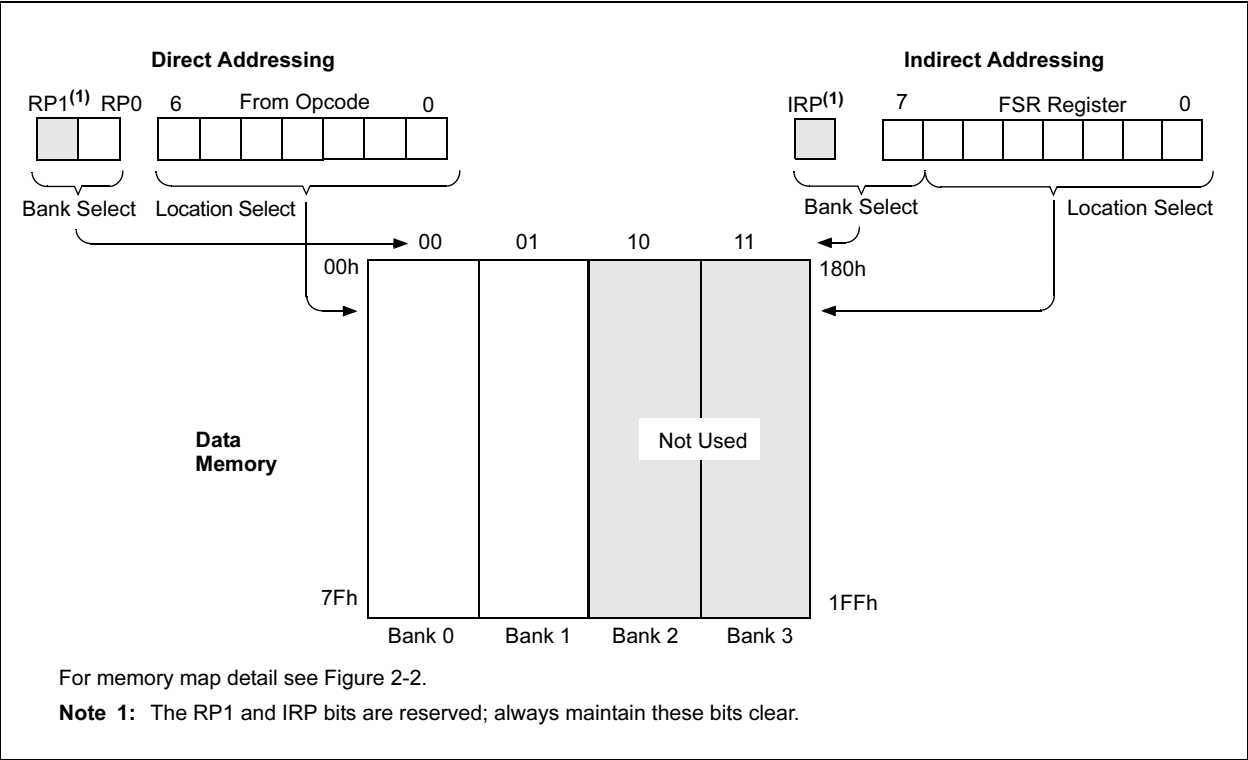
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no operation (although STATUS bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit (STATUS<7>), as shown in Figure 2-4.

A simple program to clear RAM location 20h-2Fh using indirect addressing is shown in Example 2-1.

EXAMPLE 2-1: INDIRECT ADDRESSING

```
movlw 0x20 ;initialize pointer
movwf FSR ;to RAM
NEXT   clrf INDF ;clear INDF register
       incf FSR ;inc pointer
       btfss FSR,4 ;all done?
       goto NEXT ;no clear next
CONTINUE ;yes continue
```

FIGURE 2-4: DIRECT/INDIRECT ADDRESSING PIC12F629/675



3.0 GPIO PORT

There are as many as six general purpose I/O pins available. Depending on which peripherals are enabled, some or all of the pins may not be available as general purpose I/O. In general, when a peripheral is enabled, the associated pin may not be used as a general purpose I/O pin.

Note: Additional information on I/O ports may be found in the PICmicro™ Mid-Range Reference Manual, (DS33023)

3.1 GPIO and the TRISIO Registers

GPIO is an 6-bit wide, bi-directional port. The corresponding data direction register is TRISIO. Setting a TRISIO bit (= 1) will make the corresponding GPIO pin an input (i.e., put the corresponding output driver in a Hi-impedance mode). Clearing a TRISIO bit (= 0) will make the corresponding GPIO pin an output (i.e., put the contents of the output latch on the selected pin). The exception is GP3, which is input only and its TRISIO bit will always read as '1'. Example 3-1 shows how to initialize GPIO.

Reading the GPIO register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified, and then written to the port data latch. GP3 reads '0' when MCLREN = 1.

The TRISIO register controls the direction of the GP pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISIO

register are maintained set when using them as analog inputs. I/O pins configured as analog inputs always read '0'.

Note: The ANSEL (9Fh) and CMCON (19h) registers (9Fh) must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'. The ANSEL register is defined for the PIC12F675.

EXAMPLE 3-1: INITIALIZING GPIO

```
bcf    STATUS,RP0    ;Bank 0
clrf   GPIO          ;Init GPIO
movlw  07h           ;Set GP<2:0> to
movwf  CMCON         ;digital IO
bsf    STATUS,RP0    ;Bank 1
clrf   ANSEL         ;Digital I/O
movlw  0Ch           ;Set GP<3:2> as inputs
movwf  TRISIO        ;and set GP<5:4,1:0>
                        ;as outputs
```

3.2 Additional Pin Functions

Every GPIO pin on the PIC12F629/675 has an interrupt-on-change option and every GPIO pin, except GP3, has a weak pull-up option. The next two sections describe these functions.

3.2.1 WEAK PULL-UP

Each of the GPIO pins, except GP3, has an individually configurable weak internal pull-up. Control bits WPUx enable or disable each pull-up. Refer to Register 3-3. Each weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset by the GPPU bit (OPTION<7>).

REGISTER 3-1: GPIO — GPIO REGISTER (ADDRESS: 05h)

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	GPIO5	GPIO4	GPIO3	GPIO2	GPIO1	GPIO0
bit 7							bit 0

bit 7-6: **Unimplemented:** Read as '0'
 bit 5-0: **GPIO<5:0>:** General Purpose I/O pin.
 1 = Port pin is >V_{IH}
 0 = Port pin is <V_{IL}

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

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REGISTER 3-2: TRISIO — GPIO TRISTATE REGISTER (ADDRESS: 85h)

U-0	U-0	R/W-x	R/W-x	R-1	R/W-x	R/W-x	R/W-x
—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0
bit 7				bit 0			

bit 7-6: **Unimplemented:** Read as '0'

bit 5-0: **TRISIO<5:0>**: General Purpose I/O Tri-State Control bit

1 = GPIO pin configured as an input (tri-stated)

0 = GPIO pin configured as an output.

Note: TRISIO<3> always reads 1.

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

REGISTER 3-3: WPU — WEAK PULL-UP REGISTER (ADDRESS: 95h)

U-0	U-0	R/W-1	R/W-1	U-0	R/W-1	R/W-1	R/W-1
—	—	WPU5	WPU4	—	WPU2	WPU1	WPU0
bit 7				bit 0			

bit 7-6: **Unimplemented:** Read as '0'

bit 5-4: **WPU<5:4>**: Weak Pull-up Register bit

1 = Pull-up enabled

0 = Pull-up disabled

bit 3: **Unimplemented:** Read as '0'

bit 2-0: **WPU<2:0>**: Weak Pull-up Register bit

1 = Pull-up enabled

0 = Pull-up disabled

Note 1: Global $\overline{\text{GPPU}}$ must be enabled for individual pull-ups to be enabled.

Note 2: The weak pull-up device is automatically disabled if the pin is in Output mode (TRISIO = 0).

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

3.2.2 INTERRUPT-ON-CHANGE

Each of the GPIO pins is individually configurable as an interrupt-on-change pin. Control bits IOC enable or disable the interrupt function for each pin. Refer to Register 3-4. The interrupt-on-change is disabled on a Power-on Reset.

For enabled interrupt-on-change pins, the values are compared with the old value latched on the last read of GPIO. The 'mismatch' outputs of the last read are OR'd together to set, the GP Port Change Interrupt flag bit (GPIF) in the INTCON register.

This interrupt can wake the device from SLEEP. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- Any read or write of GPIO. This will end the mismatch condition.
- Clear the flag bit GPIF.

A mismatch condition will continue to set flag bit GPIF. Reading GPIO will end the mismatch condition and allow flag bit GPIF to be cleared.

Note: If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the GPIF interrupt flag may not get set.

REGISTER 3-4: IOC — INTERRUPT-ON-CHANGE GPIO REGISTER (ADDRESS: 96h)

U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	—	IOC5	IOC4	IOC3	IOC2	IOC1	IOC0
bit 7		bit 0					

bit 7-6 **Unimplemented:** Read as '0'

bit 5-0 **IOC<5:0>:** Interrupt-on-Change GPIO Control bit
 1 = Interrupt-on-change enabled
 0 = Interrupt-on-change disabled

Note 1: Global interrupt enable (GIE) must be enabled for individual interrupts to be recognized.

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

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3.3 Pin Descriptions and Diagrams

Each GPIO pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the comparator or the A/D, refer to the appropriate section in this Data Sheet.

3.3.1 GP0/AN0/CIN+

Figure 3-1 shows the diagram for this pin. The GP0 pin is configurable to function as one of the following:

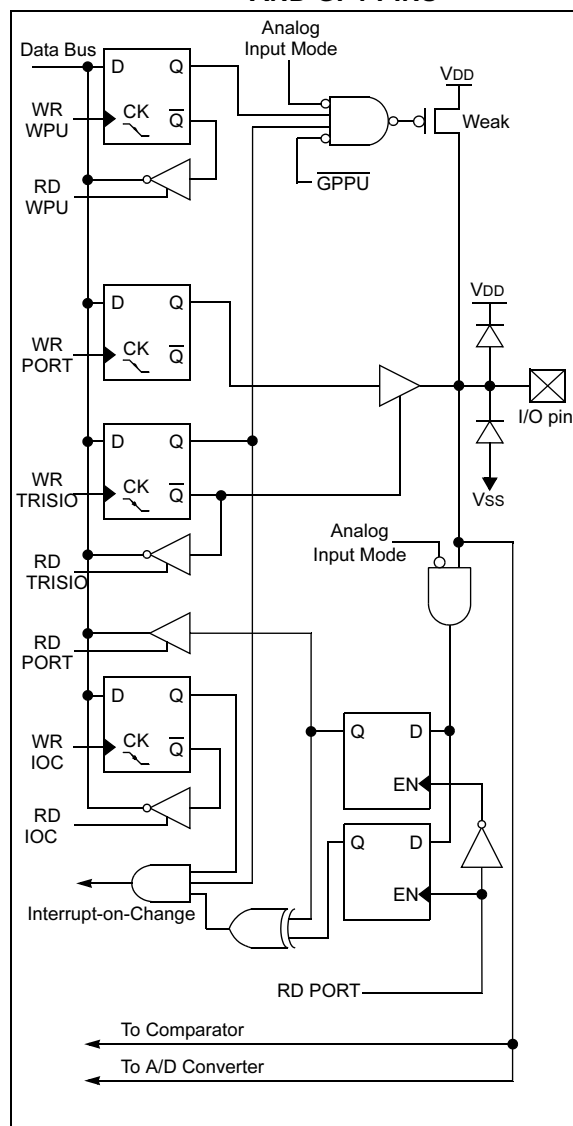
- a general purpose I/O
- an analog input for the A/D (PIC12F675 only)
- an analog input to the comparator

3.3.2 GP1/AN1/CIN-/VREF

Figure 3-1 shows the diagram for this pin. The GP1 pin is configurable to function as one of the following:

- as a general purpose I/O
- an analog input for the A/D (PIC12F675 only)
- an analog input to the comparator
- a voltage reference input for the A/D (PIC12F675 only)

FIGURE 3-1: BLOCK DIAGRAM OF GP0 AND GP1 PINS

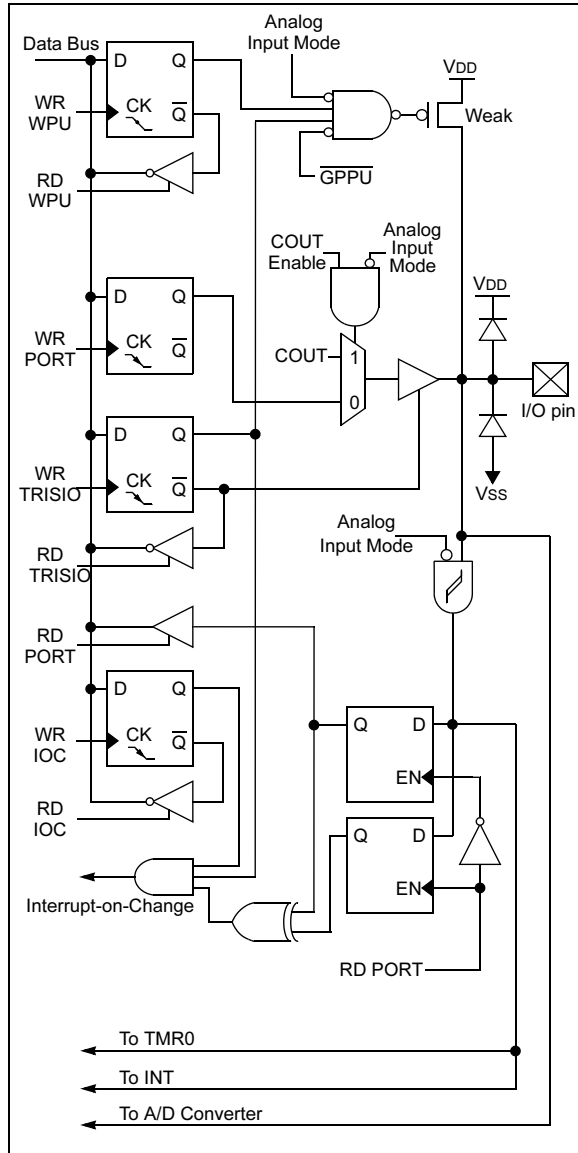


3.3.3 GP2/AN2/T0CKI/INT/COUT

Figure 3-2 shows the diagram for this pin. The GP2 pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the A/D (PIC12F675 only)
- the clock input for TMR0
- an external edge triggered interrupt
- a digital output from the comparator

FIGURE 3-2: BLOCK DIAGRAM OF GP2



3.3.4 GP3/MCLR/VPP

Figure 3-3 shows the diagram for this pin. The GP3 pin is configurable to function as one of the following:

- a general purpose input
- as Master Clear Reset

FIGURE 3-3: BLOCK DIAGRAM OF GP3

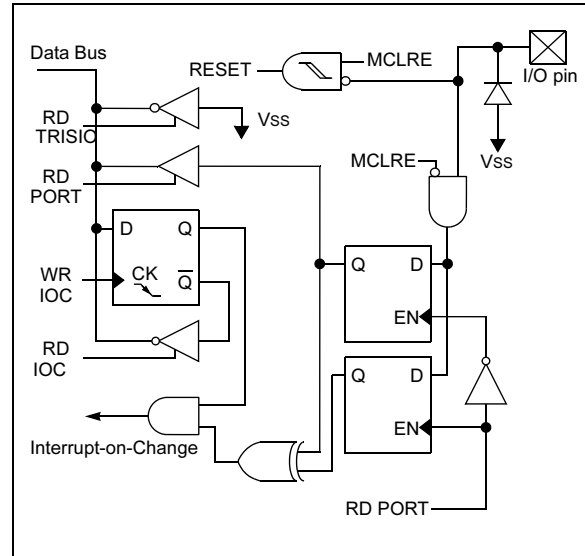


TABLE 3-1: SUMMARY OF REGISTERS ASSOCIATED WITH GPIO

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOD	Value on all other RESETS
05h	GPIO	—	—	GP5	GP4	GP3	GP2	GP1	GP0	--xx xxxx	--uu uuuu
0Bh/8Bh	INTCON	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF	0000 0000	0000 000u
19h	CMCON	—	COUT	—	CINV	CIS	CM2	CM1	CM0	-0-0 0000	-0-0 0000
81h	OPTION_REG	GPPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
85h	TRISIO	—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0	--11 1111	--11 1111
95h	WPU	—	—	WPU5	WPU4	—	WPU2	WPU1	WPU0	--11 -111	--11 -111
96h	IOC	—	—	IOC5	IOC4	IOC3	IOC2	IOC1	IOC0	--00 0000	--00 0000
9Fh	ANSEL	—	ADCS2	ADCS1	ADCS0	ANS3	ANS2	ANS1	ANS0	-000 1111	-000 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by GPIO.

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NOTES:

4.0 TIMER0 MODULE

The Timer0 module timer/counter has the following features:

- 8-bit timer/counter
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select
- Interrupt on overflow from FFh to 00h
- Edge select for external clock

Figure 4-1 is a block diagram of the Timer0 module and the prescaler shared with the WDT.

Note: Additional information on the Timer0 module is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

4.1 Timer0 Operation

Timer mode is selected by clearing the T0CS bit (OPTION_REG<5>). In Timer mode, the Timer0 module will increment every instruction cycle (without prescaler). If TMR0 is written, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

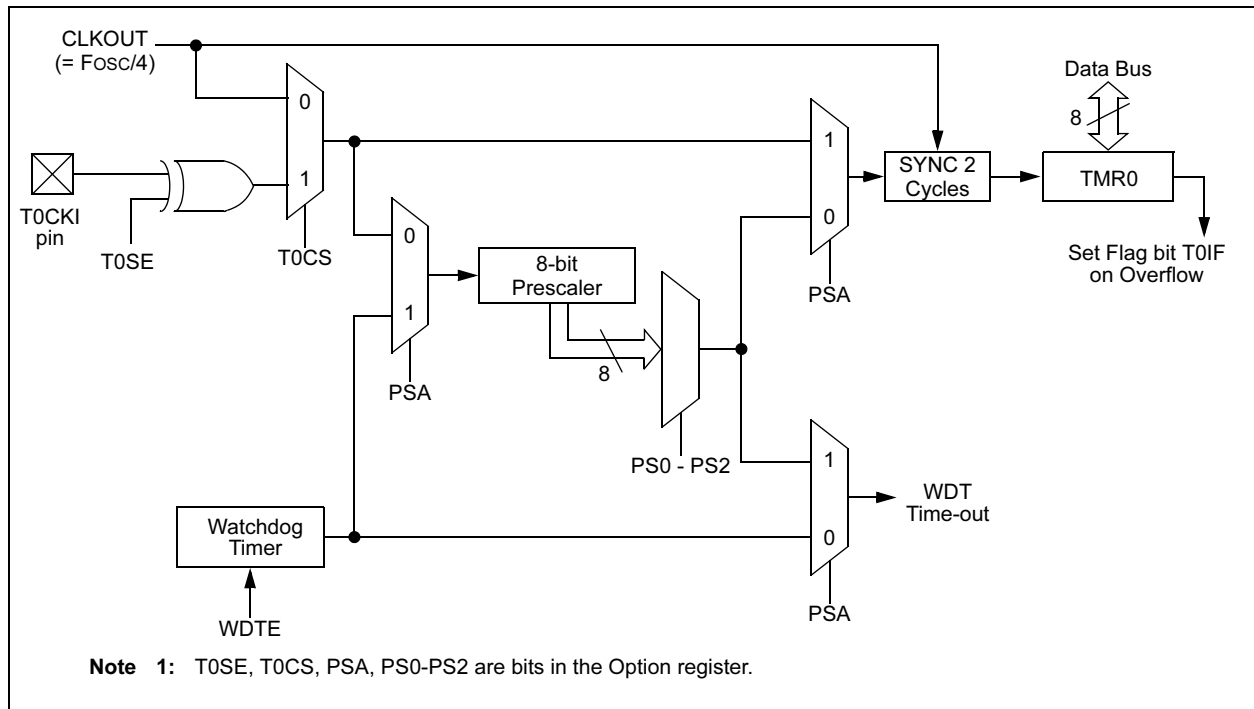
Counter mode is selected by setting the T0CS bit (OPTION_REG<5>). In this mode, the Timer0 module will increment either on every rising or falling edge of pin GP2/T0CKI. The incrementing edge is determined by the source edge (T0SE) control bit (OPTION_REG<4>). Clearing the T0SE bit selects the rising edge.

Note: Counter mode has specific external clock requirements. Additional information on these requirements is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

4.2 Timer0 Interrupt

A Timer0 interrupt is generated when the TMR0 register timer/counter overflows from FFh to 00h. This overflow sets the T0IF bit. The interrupt can be masked by clearing the T0IE bit (INTCON<5>). The T0IF bit (INTCON<2>) must be cleared in software by the Timer0 module Interrupt Service Routine before re-enabling this interrupt. The Timer0 interrupt cannot wake the processor from SLEEP since the timer is shut-off during SLEEP.

FIGURE 4-1: BLOCK DIAGRAM OF THE TIMER0/WDT PRESCALER



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4.3 Using Timer0 with an External Clock

When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI, with the internal phase clocks, is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, it is necessary for T0CKI to be high for at least 2Tosc (and

a small RC delay of 20 ns) and low for at least 2Tosc (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

Note: The ANSEL (9Fh) and CMCON (19h) registers must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'. The ANSEL register is defined for the PIC12F675.

REGISTER 4-1: OPTION_REG — OPTION REGISTER (ADDRESS: 81h)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
GPPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0
bit 7							bit 0

- bit 7 **GPPU:** GPIO Pull-up Enable bit
 1 = GPIO pull-ups are disabled
 0 = GPIO pull-ups are enabled by individual port latch values
- bit 6 **INTEDG:** Interrupt Edge Select bit
 1 = Interrupt on rising edge of GP2/INT pin
 0 = Interrupt on falling edge of GP2/INT pin
- bit 5 **T0CS:** TMR0 Clock Source Select bit
 1 = Transition on GP2/T0CKI pin
 0 = Internal instruction cycle clock (CLKOUT)
- bit 4 **T0SE:** TMR0 Source Edge Select bit
 1 = Increment on high-to-low transition on GP2/T0CKI pin
 0 = Increment on low-to-high transition on GP2/T0CKI pin
- bit 3 **PSA:** Prescaler Assignment bit
 1 = Prescaler is assigned to the WDT
 0 = Prescaler is assigned to the TIMER0 module
- bit 2-0 **PS2:PS0:** Prescaler Rate Select bits

Bit Value	TMR0 Rate	WDT Rate
000	1 : 2	1 : 1
001	1 : 4	1 : 2
010	1 : 8	1 : 4
011	1 : 16	1 : 8
100	1 : 32	1 : 16
101	1 : 64	1 : 32
110	1 : 128	1 : 64
111	1 : 256	1 : 128

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

4.4 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module, or as a postscaler for the Watchdog Timer. For simplicity, this counter will be referred to as “prescaler” throughout this Data Sheet. The prescaler assignment is controlled in software by the control bit PSA (OPTION_REG<3>). Clearing the PSA bit will assign the prescaler to Timer0. Prescale values are selectable via the PS2:PS0 bits (OPTION_REG<2:0>).

The prescaler is not readable or writable. When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., CLRF 1, MOVWF 1, BSF 1, x...etc.) will clear the prescaler. When assigned to WDT, a CLRWD instruction will clear the prescaler along with the Watchdog Timer.

4.4.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control (i.e., it can be changed “on the fly” during program execution). To avoid an unintended device RESET, the following instruction sequence (Example 4-1) must be executed when changing the prescaler assignment from Timer0 to WDT.

EXAMPLE 4-1: CHANGING PRESCALER (TIMER0→WDT)

```

bcf     STATUS,RP0    ;Bank 0
clrwdt                ;Clear WDT
clrf     TMR0          ;Clear TMR0 and
                        ; prescaler
bsf     STATUS,RP0    ;Bank 1

movlw    b'00101111'  ;Required if desired
movwf    OPTION_REG    ; PS2:PS0 is
clrwdt                ; 000 or 001
                        ;
movlw    b'00101xxx'  ;Set postscaler to
movwf    OPTION_REG    ; desired WDT rate
bcf     STATUS,RP0    ;Bank 0
    
```

To change prescaler from the WDT to the TMR0 module, use the sequence shown in Example 4-2. This precaution must be taken even if the WDT is disabled.

EXAMPLE 4-2: CHANGING PRESCALER (WDT→TIMER0)

```

clrwdt                ;Clear WDT and
                        ; postscaler
bsf     STATUS,RP0    ;Bank 1

movlw    b'xxxx0xxx'  ;Select TMR0,
                        ; prescale, and
                        ; clock source
movwf    OPTION_REG    ;
bcf     STATUS,RP0    ;Bank 0
    
```

TABLE 4-1: REGISTERS ASSOCIATED WITH TIMER0

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS
01h	TMR0	Timer0 Module Register								xxxx xxxx	uuuu uuuu
0Bh/8Bh	INTCON	GIE	PEIE	TOIE	INTE	GPIE	TOIF	INTF	GPIF	0000 0000	0000 000u
81h	OPTION_REG	GPPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
85h	TRISIO	—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0	--11 1111	--11 1111

Legend: — = Unimplemented locations, read as '0', u = unchanged, x = unknown.
Shaded cells are not used by the Timer0 module.

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5.0 TIMER1 MODULE WITH GATE CONTROL

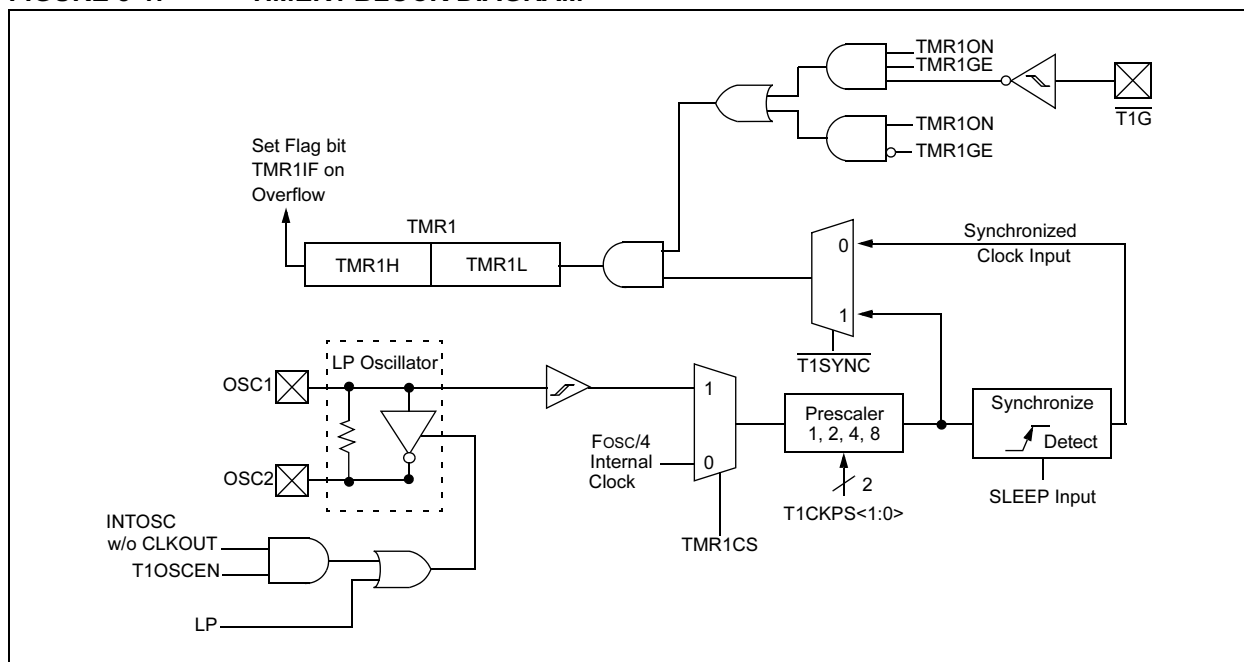
The PIC12F629/675 devices have a 16-bit timer. Figure 5-1 shows the basic block diagram of the Timer1 module. Timer1 has the following features:

- 16-bit timer/counter (TMR1H:TMR1L)
- Readable and writable
- Internal or external clock selection
- Synchronous or asynchronous operation
- Interrupt on overflow from FFFFh to 0000h
- Wake-up upon overflow (Asynchronous mode)
- Optional external enable input ($\overline{T1G}$)
- Optional LP oscillator

The Timer1 Control register (T1CON), shown in Register 5-1, is used to enable/disable Timer1 and select the various features of the Timer1 module.

Note: Additional information on timer modules is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

FIGURE 5-1: TIMER1 BLOCK DIAGRAM



5.1 Timer1 Modes of Operation

Timer1 can operate in one of three modes:

- 16-bit timer with prescaler
- 16-bit synchronous counter
- 16-bit asynchronous counter

In Timer mode, Timer1 is incremented on every instruction cycle. In Counter mode, Timer1 is incremented on the rising edge of the external clock input T1CKI. In addition, the Counter mode clock can be synchronized to the microcontroller system clock or run asynchronously.

In Counter and Timer modules, the counter/timer clock can be gated by the $\overline{T1G}$ input.

If an external clock oscillator is needed (and the microcontroller is using the INTOSC w/o CLKOUT), Timer1 can use the LP oscillator as a clock source.

Note: In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge.

5.2 Timer1 Interrupt

The Timer1 register pair (TMR1H:TMR1L) increments to FFFFh and rolls over to 0000h. When Timer1 rolls over, the Timer1 interrupt flag bit (PIR1<0>) is set. To enable the interrupt on rollover, you must set these bits:

- Timer1 interrupt Enable bit (PIE1<0>)
- PEIE bit (INTCON<6>)
- GIE bit (INTCON<7>).

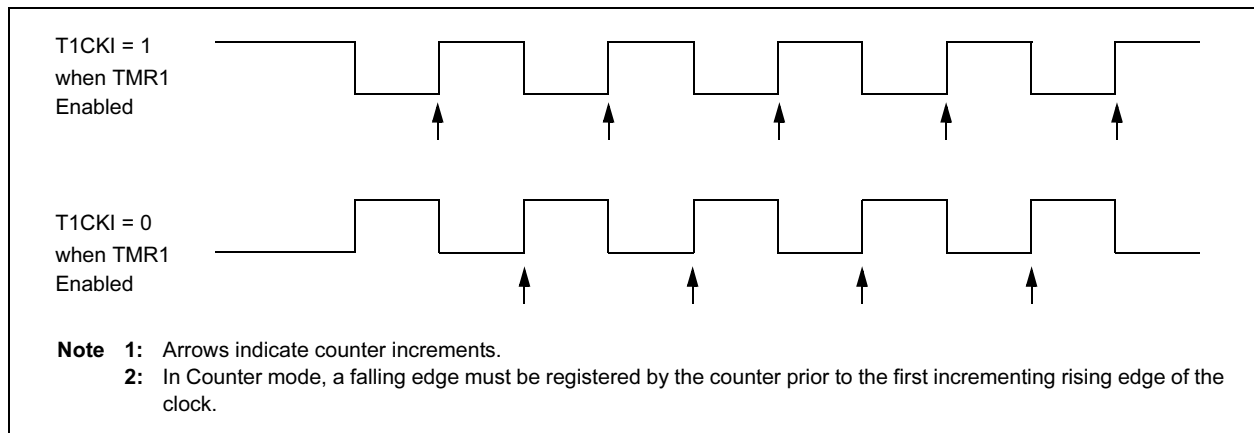
The interrupt is cleared by clearing the TMR1IF in the Interrupt Service Routine.

Note: The TMR1H:TMR1L register pair and the TMR1IF bit should be cleared before enabling interrupts.

5.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4, or 8 divisions of the clock input. The T1CKPS bits (T1CON<5:4>) control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

FIGURE 5-2: TIMER1 INCREMENTING EDGE



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REGISTER 5-1: T1CON — TIMER1 CONTROL REGISTER (ADDRESS: 10h)

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYN \overline{C}	TMR1CS	TMR1ON
bit 7							bit 0

- bit 7 **Unimplemented:** Read as '0'
- bit 6 **TMR1GE:** Timer1 Gate Enable bit
If TMR1ON = 0:
 This bit is ignored
If TMR1ON = 1:
 1 = Timer1 is on if T1G pin is low
 0 = Timer1 is on
- bit 5-4 **T1CKPS1:T1CKPS0:** Timer1 Input Clock Prescale Select bits
 11 = 1:8 Prescale Value
 10 = 1:4 Prescale Value
 01 = 1:2 Prescale Value
 00 = 1:1 Prescale Value
- bit 3 **T1OSCEN:** LP Oscillator Enable Control bit
If INTOSC without CLKOUT oscillator is active:
 1 = LP oscillator is enabled for Timer1 clock
 0 = LP oscillator is off
Else:
 This bit is ignored
- bit 2 **T1SYN \overline{C} :** Timer1 External Clock Input Synchronization Control bit
TMR1CS = 1:
 1 = Do not synchronize external clock input
 0 = Synchronize external clock input
TMR1CS = 0:
 This bit is ignored. Timer1 uses the internal clock.
- bit 1 **TMR1CS:** Timer1 Clock Source Select bit
 1 = External clock from T1OSO/T1CKI pin (on the rising edge)
 0 = Internal clock (Fosc/4)
- bit 0 **TMR1ON:** Timer1 On bit
 1 = Enables Timer1
 0 = Stops Timer1

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

5.4 Timer1 Operation in Asynchronous Counter Mode

If control bit $\overline{T1SYNC}$ (T1CON<2>) is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during SLEEP and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (Section 5.4.1).

Note: The ANSEL (9Fh) and CMCON (19h) registers must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'. The ANSEL register is defined for the PIC12F675.

5.4.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L, while the timer is running from an external asynchronous clock, will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the timer register.

Reading the 16-bit value requires some care. Examples 12-2 and 12-3 in the PICmicro™ Mid-Range MCU Family Reference Manual (DS33023) show how to read and write Timer1 when it is running in Asynchronous mode.

5.5 Timer1 Oscillator

A crystal oscillator circuit is built-in between pins OSC1 (input) and OSC2 (amplifier output). It is enabled by setting control bit T1OSCEN (T1CON<3>). The oscillator is a low power oscillator rated up to 37 kHz. It will continue to run during SLEEP. It is primarily intended for a 32 kHz crystal. Table 9-2 shows the capacitor selection for the Timer1 oscillator.

The Timer1 oscillator is shared with the system LP oscillator. Thus, Timer1 can use this mode only when the system clock is derived from the internal oscillator. As with the system LP oscillator, the user must provide a software time delay to ensure proper oscillator start-up.

While enabled, TRISIO4 and TRISIO5 are set. GP4 and GP5 read '0' and TRISIO4 and TRISIO5 are read '1'.

Note: The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to enabling Timer1.

5.6 Timer1 Operation During SLEEP

Timer1 can only operate during SLEEP when setup in Asynchronous Counter mode. In this mode, an external crystal or clock source can be used to increment the counter. To setup the timer to wake the device:

- Timer1 must be on (T1CON<0>)
- TMR1IE bit (PIE1<0>) must be set
- PEIE bit (INTCON<6>) must be set

The device will wake-up on an overflow. If the GIE bit (INTCON<7>) is set, the device will wake-up and jump to the Interrupt Service Routine on an overflow.

TABLE 5-1: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS
0Bh/8Bh	INTCON	GIE	PEIE	TOIE	INTE	GPIE	TOIF	INTF	GPIF	0000 0000	0000 000u
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	00-- 0--0
0Eh	TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
0Fh	TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx xxxx	uuuu uuuu
10h	T1CON	—	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	TMR1CS	TMR1ON	-000 0000	-uuu uuuu
8Ch	PIE1	EEIE	ADIE	—	—	CMIE	—	—	TMR1IE	00-- 0--0	00-- 0--0

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer1 module.

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NOTES:

6.0 COMPARATOR MODULE

The PIC12F629/675 devices have one analog comparator. The inputs to the comparator are multiplexed with the GP0 and GP1 pins. There is an on-chip Comparator Voltage Reference that can also

be applied to an input of the comparator. In addition, GP2 can be configured as the comparator output. The Comparator Control Register (CMCON), shown in Register 6-1, contains the bits to control the comparator.

REGISTER 6-1: CMCON — COMPARATOR CONTROL REGISTER (ADDRESS: 19h)

U-0	R-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	COUT	—	CINV	CIS	CM2	CM1	CM0
bit 7			bit 0				

6.1 Comparator Operation

A single comparator is shown in Figure 6-1, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 6-1 represent the uncertainty due to input offsets and response time.

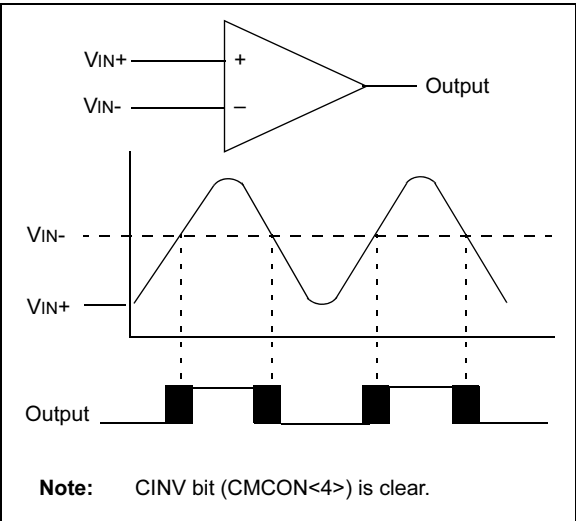
Note: To use CIN+ and CIN- pins as analog inputs, the appropriate bits must be programmed in the CMCON (19h) register.

The polarity of the comparator output can be inverted by setting the CINV bit (CMCON<4>). Clearing CINV results in a non-inverted output. A complete table showing the output state versus input conditions and the polarity bit is shown in Table 6-1.

TABLE 6-1: OUTPUT STATE VS. INPUT CONDITIONS

Input Conditions	CINV	COUT
VIN- > VIN+	0	0
VIN- < VIN+	0	1
VIN- > VIN+	1	1
VIN- < VIN+	1	0

FIGURE 6-1: SINGLE COMPARATOR



6.2 Comparator Configuration

There are eight modes of operation for the comparator. The CMCON register, shown in Register 6-1, is used to select the mode. Figure 6-2 shows the eight possible modes. The TRISIO register controls the data direction of the comparator pins for each mode. If the

Comparator mode is changed, the comparator output level may not be valid for a specified period of time. Refer to the specifications in Section 12.0.

Note: Comparator interrupts should be disabled during a Comparator mode change. Otherwise, a false interrupt may occur.

FIGURE 6-2: COMPARATOR I/O OPERATING MODES

<p>Comparator Reset (POR Default Value - low power) CM2:CM0 = 000</p> <p>GP1/CIN- A GP0/CIN+ A GP2/COU- D</p> <p>Off (Read as '0')</p>	<p>Comparator Off (Lowest power) CM2:CM0 = 111</p> <p>GP1/CIN- D GP0/CIN+ D GP2/COU- D</p> <p>Off (Read as '0')</p>
<p>Comparator without Output CM2:CM0 = 010</p> <p>GP1/CIN- A GP0/CIN+ A GP2/COU- D</p> <p>COUT</p>	<p>Comparator w/o Output and with Internal Reference CM2:CM0 = 100</p> <p>GP1/CIN- A GP0/CIN+ D GP2/COU- D</p> <p>From CVREF Module</p> <p>COUT</p>
<p>Comparator with Output and Internal Reference CM2:CM0 = 011</p> <p>GP1/CIN- A GP0/CIN+ D GP2/COU- D</p> <p>From CVREF Module</p> <p>COUT</p>	<p>Multiplexed Input with Internal Reference and Output CM2:CM0 = 101</p> <p>GP1/CIN- A GP0/CIN+ A GP2/COU- D</p> <p>CIS = 0 CIS = 1</p> <p>From CVREF Module</p> <p>COUT</p>
<p>Comparator with Output CM2:CM0 = 001</p> <p>GP1/CIN- A GP0/CIN+ A GP2/COU- D</p> <p>COUT</p>	<p>Multiplexed Input with Internal Reference CM2:CM0 = 110</p> <p>GP1/CIN- A GP0/CIN+ A GP2/COU- D</p> <p>CIS = 0 CIS = 1</p> <p>From CVREF Module</p> <p>COUT</p>
<p>A = Analog Input, ports always reads '0' D = Digital Input CIS = Comparator Input Switch (CMCON<3>)</p>	

A simplified circuit for an analog input is shown in Figure 6-3. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and VSS. The analog input, therefore, must be between VSS and VDD. If the input voltage deviates from this

range by more than 0.6V in either direction, one of the diodes is forward biased and a latchup may occur. A maximum source impedance of 10 k Ω is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

Legend:

CPIN	= Input Capacitance
VT	= Threshold Voltage
I _{LEAKAGE}	= Leakage Current at the pin due to Various Junctions
RIC	= Interconnect Resistance
Rs	= Source Impedance
VA	= Analog Voltage

The comparator output, COUT, is read through the CMCON register. This bit is read only. The comparator output may also be directly output to the GP2 pin in three of the eight possible modes, as shown in Figure 6-2. When in one of these modes, the output on GP2 is asynchronous to the internal clock. Figure 6-4 shows the comparator output block diagram.

The TRISIO<2> bit functions as an output enable/disable for the GP2 pin while the comparator is in an Output mode.

2: Analog levels on any pin that is defined as a digital input, may cause the input buffer to consume more current than is specified.

The logic diagram illustrates the internal structure of the GP0/GP1 pin driver. It features two D-type flip-flops (labeled Q, D, EN) and a 3-to-1 multiplexer. The top flip-flop's D input is connected to the output of an OR gate, which has inputs from 'To Data Bus' (via an inverter) and 'RD CMCON'. Its Q output is connected to 'To GP2/T0CKI pin'. The bottom flip-flop's D input is connected to the output of an AND gate, which has inputs from 'Set CMIF bit' and 'RD CMCON'. Its Q output is connected to the 'EN' input of the top flip-flop. The outputs of both flip-flops are connected to an AND gate, which also receives 'CINV' as an input. The output of this AND gate is connected to the '+' input of a comparator. The '-' input of the comparator is connected to 'CVREF'. The output of the comparator is connected to the 'CM2:CM0' input of a 3-to-1 multiplexer. The multiplexer's other two inputs are connected to 'GP0/CIN+' and 'GP1/CIN-'. The output of the multiplexer is connected to the 'To GP2/T0CKI pin' line. A 'RESET' signal is connected to the 'EN' input of the bottom flip-flop.

6.5 Comparator Reference

The comparator module also allows the selection of an internally generated voltage reference for one of the comparator inputs. The internal reference signal is used for four of the eight Comparator modes. The VRCON register, Register 6-2, controls the voltage reference module shown in Figure 6-5.

6.5.1 CONFIGURING THE VOLTAGE REFERENCE

The voltage reference can output 32 distinct voltage levels, 16 in a high range and 16 in a low range.

The following equations determine the output voltages:

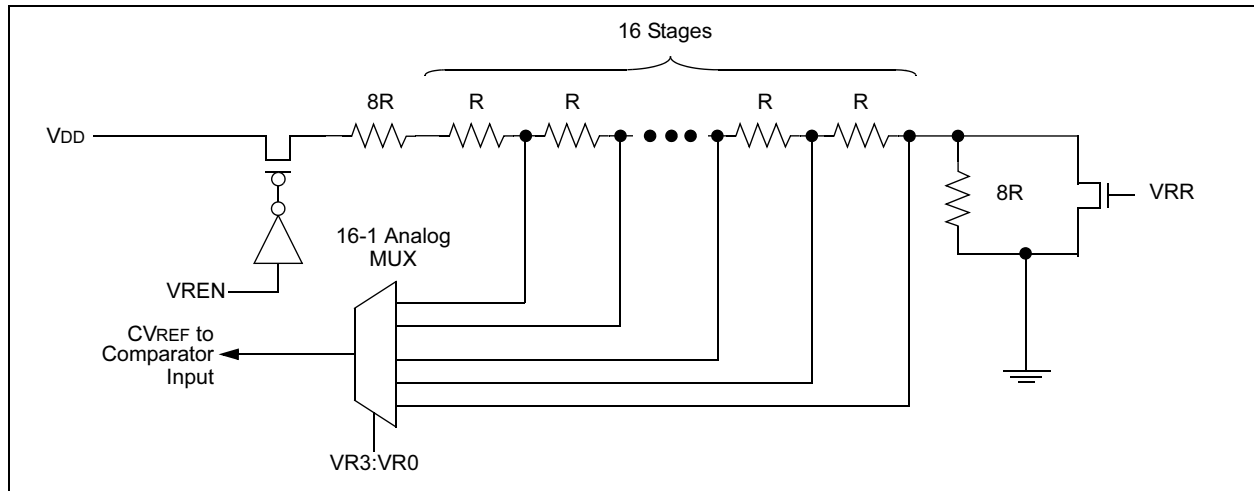
$$\text{VRR} = 1 \text{ (low range): } \text{CVREF} = (\text{VR3:VR0} / 24) \times \text{VDD}$$

$$\text{VRR} = 0 \text{ (high range): } \text{CVREF} = (\text{VDD} / 4) + (\text{VR3:VR0} \times \text{VDD} / 32)$$

6.5.2 VOLTAGE REFERENCE ACCURACY/ERROR

The full range of VSS to VDD cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 6-5) keep CVREF from approaching VSS or VDD. The Voltage Reference is VDD derived and therefore, the CVREF output changes with fluctuations in VDD. The tested absolute accuracy of the Comparator Voltage Reference can be found in Section 12.0.

FIGURE 6-5: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM



6.6 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output is ensured to have a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (Table 12-7).

6.7 Operation During SLEEP

Both the comparator and voltage reference, if enabled before entering SLEEP mode, remain active during SLEEP. This results in higher SLEEP currents than shown in the power-down specifications. The additional current consumed by the comparator and the voltage reference is shown separately in the specifications. To minimize power consumption while in SLEEP mode, turn off the comparator, CM2:CM0 = 111, and voltage reference, VRCON<7> = 0.

While the comparator is enabled during SLEEP, an interrupt will wake-up the device. If the device wakes up from SLEEP, the contents of the CMCON and VRCON registers are not affected.

6.8 Effects of a RESET

A device RESET forces the CMCON and VRCON registers to their RESET states. This forces the comparator module to be in the Comparator Reset mode, CM2:CM0 = 000 and the voltage reference to its off state. Thus, all potential inputs are analog inputs with the comparator and voltage reference disabled to consume the smallest current possible.

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REGISTER 6-2: VRCON — VOLTAGE REFERENCE CONTROL REGISTER (ADDRESS: 99h)

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
VREN	—	VRR	—	VR3	VR2	VR1	VR0
bit 7				bit 0			

- bit 7 **VREN:** CVREF Enable bit
 1 = CVREF circuit powered on
 0 = CVREF circuit powered down, no IDD drain
- bit 6 **Unimplemented:** Read as '0'
- bit 5 **VRR:** CVREF Range Selection bit
 1 = Low range
 0 = High range
- bit 4 **Unimplemented:** Read as '0'
- bit 3-0 **VR3:VR0:** CVREF value selection $0 \leq VR[3:0] \leq 15$
 When VRR = 1: $CVREF = (VR3:VR0 / 24) * V_{DD}$
 When VRR = 0: $CVREF = V_{DD}/4 + (VR3:VR0 / 32) * V_{DD}$

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

6.9 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of the comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<6>, to determine the actual change that has occurred. The CMIF bit, PIR1<3>, is the comparator interrupt flag. This bit must be reset in software by clearing it to '0'. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

The CMIE bit (PIE1<3>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit must also be set. If any of these bits are cleared, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- Any read or write of CMCON. This will end the mismatch condition.
- Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition, and allow flag bit CMIF to be cleared.

Note: If a change in the CMCON register (COUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CMIF (PIR1<3>) interrupt flag may not get set.

TABLE 6-2: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS
0Bh/8Bh	INTCON	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF	0000 0000	0000 000u
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	00-- 0--0
19h	CMCON	—	COUT	—	CINV	CIS	CM2	CM1	CM0	-0-0 0000	-0-0 0000
8Ch	PIE1	EEIE	ADIE	—	—	CMIE	—	—	TMR1IE	00-- 0--0	00-- 0--0
85h	TRISIO	—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0	--11 1111	--11 1111
99h	VRCON	VREN	—	VRR	—	VR3	VR2	VR1	VR0	0-0- 0000	0-0- 0000

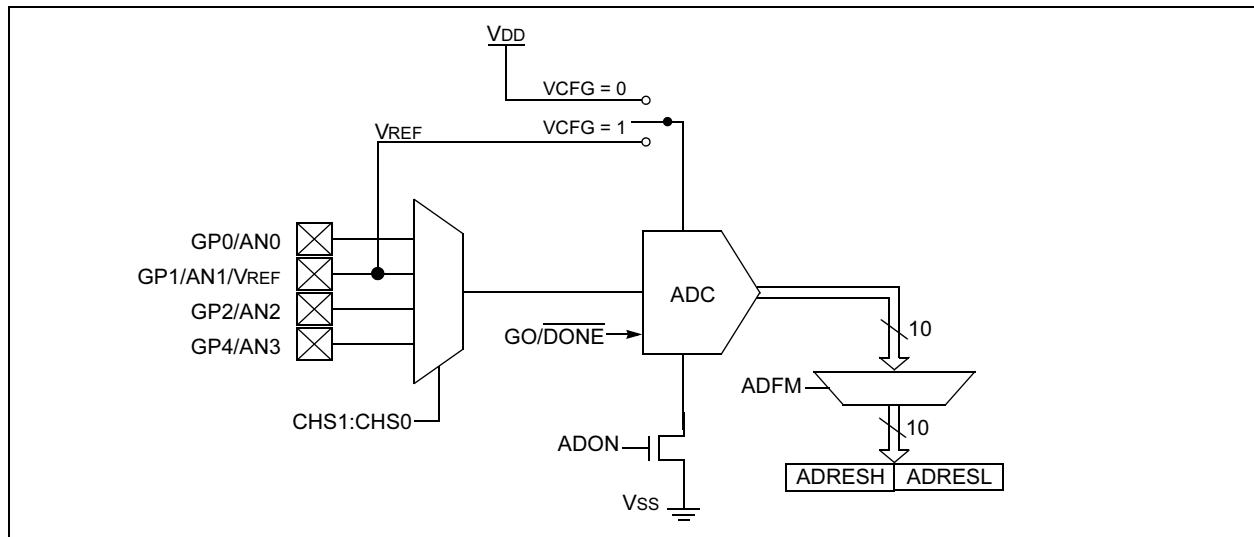
Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the comparator module.

7.0 ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE (PIC12F675 ONLY)

The analog-to-digital converter (A/D) allows conversion of an analog input signal to a 10-bit binary representation of that signal. The PIC12F675 has four analog inputs, multiplexed into one sample and hold circuit.

The output of the sample and hold is connected to the input of the converter. The converter generates a binary result via successive approximation and stores the result in a 10-bit register. The voltage reference used in the conversion is software selectable to either VDD or a voltage applied by the VREF pin. Figure 7-1 shows the block diagram of the A/D on the PIC12F675.

FIGURE 7-1: A/D BLOCK DIAGRAM



7.1 A/D Configuration and Operation

There are two registers available to control the functionality of the A/D module:

1. ADCON0 (Register 7-1)
2. ANSEL (Register 7-2)

7.1.1 ANALOG PORT PINS

The ANS3:ANS0 bits (ANSEL<3:0>) and the TRISIO bits control the operation of the A/D port pins. Set the corresponding TRISIO bits to set the pin output driver to its high impedance state. Likewise, set the corresponding ANS bit to disable the digital input buffer.

Note: Analog voltages on any pin that is defined as a digital input may cause the input buffer to conduct excess current.

7.1.2 CHANNEL SELECTION

There are four analog channels on the PIC12F675, AN0 through AN3. The CHS1:CHS0 bits (ADCON0<3:2>) control which channel is connected to the sample and hold circuit.

7.1.3 VOLTAGE REFERENCE

There are two options for the voltage reference to the A/D converter: either VDD is used, or an analog voltage applied to VREF is used. The VCFG bit (ADCON0<6>)

controls the voltage reference selection. If VCFG is set, then the voltage on the VREF pin is the reference; otherwise, VDD is the reference.

7.1.4 CONVERSION CLOCK

The A/D conversion cycle requires 11 TAD. The source of the conversion clock is software selectable via the ADCS bits (ANSEL<6:4>). There are seven possible clock options:

- Fosc/2
- Fosc/4
- Fosc/8
- Fosc/16
- Fosc/32
- Fosc/64
- FRC (dedicated internal RC oscillator)

For correct conversion, the A/D conversion clock (1/TAD) must be selected to ensure a minimum TAD of 1.6 μ s. Table 7-1 shows a few TAD calculations for selected frequencies.

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TABLE 7-1: TAD vs. DEVICE OPERATING FREQUENCIES

A/D Clock Source (TAD)		Device Frequency			
Operation	ADCS2:ADCS0	20 MHz	5 MHz	4 MHz	1.25 MHz
2 TOSC	000	100 ns ⁽²⁾	400 ns ⁽²⁾	500 ns ⁽²⁾	1.6 µs
4 TOSC	100	200 ns ⁽²⁾	800 ns ⁽²⁾	1.0 µs ⁽²⁾	3.2 µs
8 TOSC	001	400 ns ⁽²⁾	1.6 µs	2.0 µs	6.4 µs
16 TOSC	101	800 ns ⁽²⁾	3.2 µs	4.0 µs	12.8 µs ⁽³⁾
32 TOSC	010	1.6 µs	6.4 µs	8.0 µs ⁽³⁾	25.6 µs ⁽³⁾
64 TOSC	110	3.2 µs	12.8 µs ⁽³⁾	16.0 µs ⁽³⁾	51.2 µs ⁽³⁾
A/D RC	x11	2 - 6 µs ^(1,4)	2 - 6 µs ^(1,4)	2 - 6 µs ^(1,4)	2 - 6 µs ^(1,4)

Legend: Shaded cells are outside of recommended range.

Note 1: The A/D RC source has a typical TAD time of 4 µs for VDD > 3.0V.

2: These values violate the minimum required TAD time.

3: For faster conversion times, the selection of another clock source is recommended.

4: When the device frequency is greater than 1 MHz, the A/D RC clock source is only recommended if the conversion will be performed during SLEEP.

7.1.5 STARTING A CONVERSION

The A/D conversion is initiated by setting the GO/DONE bit (ADCON0<1>). When the conversion is complete, the A/D module:

- Clears the GO/DONE bit
- Sets the ADIF flag (PIR1<6>)
- Generates an interrupt (if enabled).

If the conversion must be aborted, the GO/DONE bit can be cleared in software. The ADRESH:ADRESL registers will not be updated with the partially complete A/D conversion sample. Instead, the ADRESH:ADRESL registers will retain the value of the

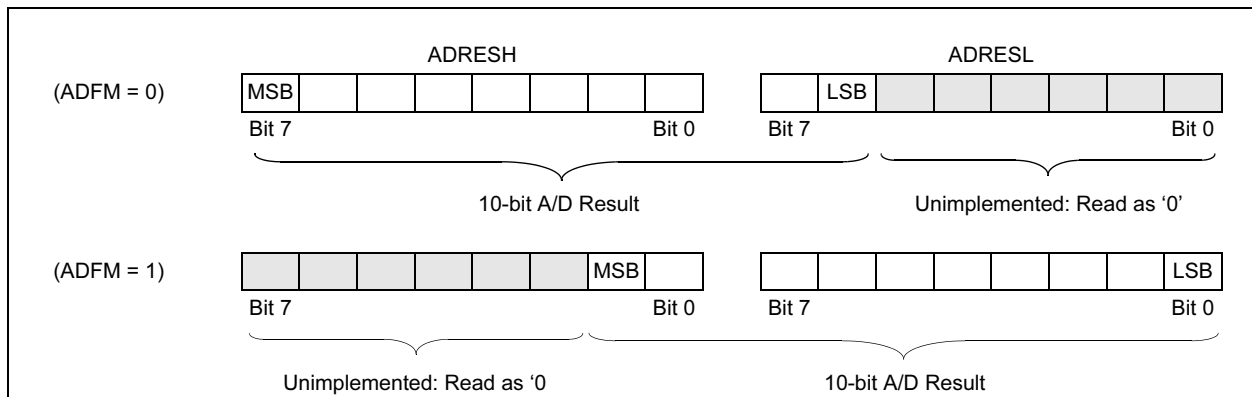
previous conversion. After an aborted conversion, a 2 TAD delay is required before another acquisition can be initiated. Following the delay, an input acquisition is automatically started on the selected channel.

Note: The GO/DONE bit should not be set in the same instruction that turns on the A/D.

7.1.6 CONVERSION OUTPUT

The A/D conversion can be supplied in two formats: left or right shifted. The ADFM bit (ADCON0<7>) controls the output format. Figure 7-2 shows the output formats.

FIGURE 7-2: 10-BIT A/D RESULT FORMAT



REGISTER 7-1: ADCON0 — A/D CONTROL REGISTER (ADDRESS: 1Fh)

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	VCFG	—	—	CHS1	CHS0	GO/DONE	ADON

bit 7

bit 0

- bit 7 **ADFM:** A/D Result Formed Select bit
 1 = Right justified
 0 = Left justified
- bit 6 **VCFG:** Voltage Reference bit
 1 = VREF pin
 0 = VDD
- bit 5-4 **Unimplemented:** Read as zero
- bit 3-2 **CHS1:CHS0:** Analog Channel Select bits
 00 = Channel 00 (AN0)
 01 = Channel 01 (AN1)
 10 = Channel 02 (AN2)
 11 = Channel 03 (AN3)
- bit 1 **GO/DONE:** A/D Conversion Status bit
 1 = A/D conversion cycle in progress. Setting this bit starts an A/D conversion cycle.
 This bit is automatically cleared by hardware when the A/D conversion has completed.
 0 = A/D conversion completed/not in progress
- bit 0 **ADON:** A/D Conversion STATUS bit
 1 = A/D converter module is operating
 0 = A/D converter is shut-off and consumes no operating current

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared x = Bit is unknown

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REGISTER 7-2: ANSEL — ANALOG SELECT REGISTER (ADDRESS: 9Fh)

U-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1	R/W-1
—	ADCS2	ADCS1	ADCS0	ANS3	ANS2	ANS1	ANS0
bit 7							bit 0

bit 7 **Unimplemented:** Read as '0'.

bit 6-4 **ADCS<2:0>:** A/D Conversion Clock Select bits

000 = Fosc/2

001 = Fosc/8

010 = Fosc/32

×11 = FRC (clock derived from a dedicated internal oscillator = 500 kHz max)

100 = Fosc/4

101 = Fosc/16

110 = Fosc/64

bit 3-0 **ANS3:ANS0:** Analog Select bits

(Between analog or digital function on pins AN<3:0>, respectively.)

1 = Analog input; pin is assigned as analog input⁽¹⁾

0 = Digital I/O; pin is assigned to port or special function

Note 1: Setting a pin to an analog input automatically disables the digital input circuitry, weak pull-ups, and interrupt-on-change. The corresponding TRISIO bit must be set to Input mode in order to allow external control of the voltage on the pin.

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

7.2 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 7-3. The source impedance (R_s) and the internal sampling switch (R_{ss}) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (R_{ss}) impedance varies over the device voltage (V_{DD}), see Figure 7-3. **The maximum recommended impedance for analog sources is 10 k Ω .** As the impedance is decreased, the acquisition time may be decreased.

After the analog input channel is selected (changed), this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, Equation 7-1 may be used. This equation assumes that 1/2 LSB error is used (1024 steps for the A/D). The 1/2 LSB error is the maximum error allowed for the A/D to meet its specified resolution.

To calculate the minimum acquisition time, T_{ACQ} , see the PICmicro™ Mid-Range Reference Manual (DS33023).

EQUATION 7-1: ACQUISITION TIME

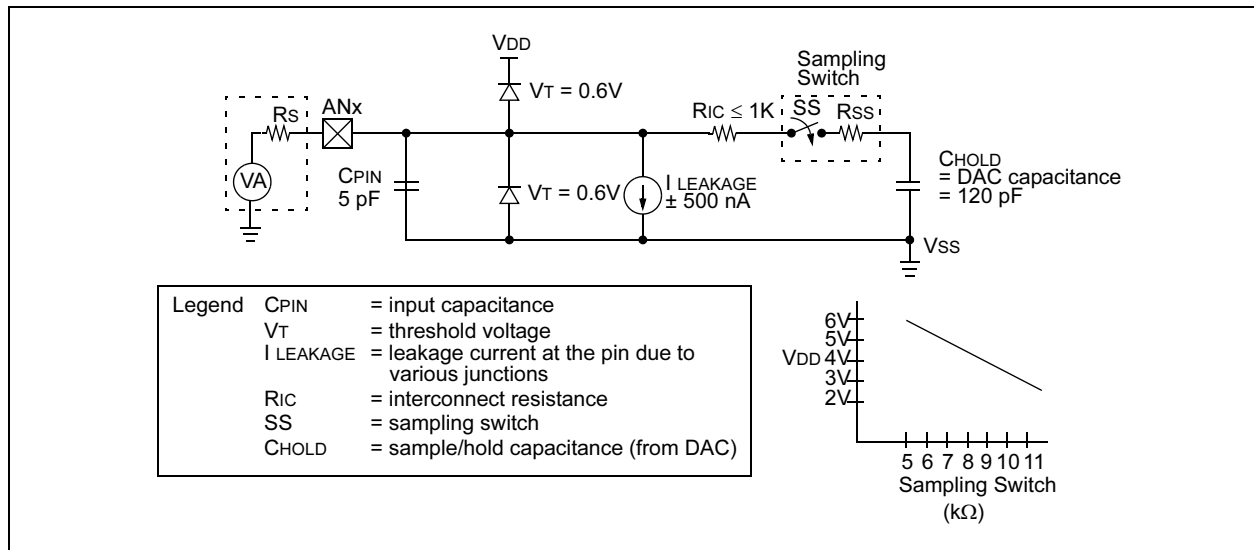
$$\begin{aligned}
 T_{ACQ} &= \text{Amplifier Settling Time} + \\
 &\quad \text{Hold Capacitor Charging Time} + \\
 &\quad \text{Temperature Coefficient} \\
 &= T_{AMP} + T_C + T_{COFF} \\
 &= 2\mu s + T_C + [(\text{Temperature} - 25^\circ\text{C})(0.05\mu s/^\circ\text{C})] \\
 T_C &= \text{CHOLD} (R_{IC} + R_{SS} + R_s) \ln(1/2047) \\
 &= -120\text{pF} (1\text{k}\Omega + 7\text{k}\Omega + 10\text{k}\Omega) \ln(0.0004885) \\
 &= 16.47\mu s \\
 T_{ACQ} &= 2\mu s + 16.47\mu s + [(50^\circ\text{C} - 25^\circ\text{C})(0.05\mu s/^\circ\text{C})] \\
 &= 19.72\mu s
 \end{aligned}$$

Note 1: The reference voltage (V_{REF}) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 k Ω . This is required to meet the pin leakage specification.

FIGURE 7-3: ANALOG INPUT MODEL



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7.3 A/D Operation During SLEEP

The A/D converter module can operate during SLEEP. This requires the A/D clock source to be set to the internal RC oscillator. When the RC clock source is selected, the A/D waits one instruction before starting the conversion. This allows the SLEEP instruction to be executed, thus eliminating much of the switching noise from the conversion. When the conversion is complete, the GO/DONE bit is cleared, and the result is loaded into the ADRESH:ADRESL registers. If the A/D interrupt is enabled, the device awakens from SLEEP. If the A/D interrupt is not enabled, the A/D module is turned off, although the ADON bit remains set.

When the A/D clock source is something other than RC, a SLEEP instruction causes the present conversion to be aborted, and the A/D module is turned off. The ADON bit remains set.

7.4 Effects of RESET

A device RESET forces all registers to their RESET state. Thus the A/D module is turned off and any pending conversion is aborted. The ADRESH:ADRESL registers are unchanged.

TABLE 7-2: SUMMARY OF A/D REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOD	Value on all other RESETS
05h	GPIO	—	—	GPIO5	GPIO4	GPIO3	GPIO2	GPIO1	GPIO0	--xx xxxx	--uu uuuu
0Bh, 8Bh	INTCON	GIE	PEIE	T0IE	INTE	GPIE	T0IF	INTF	GPIF	0000 0000	0000 000u
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	00-- 0--0
1Eh	ADRESH	Most Significant 8 bits of the Left Shifted A/D result or 2 bits of the Right Shifted Result								xxxx xxxx	uuuu uuuu
1Fh	ADCON0	ADFM	VCFG	—	—	CHS1	CHS0	GO	ADON	00-- 0000	00-- 0000
85h	TRISIO	—	—	TRISIO5	TRISIO4	TRISIO3	TRISIO2	TRISIO1	TRISIO0	--11 1111	--11 1111
8Ch	PIE1	EEIE	ADIE	—	—	CMIE	—	—	TMR1IE	00-- 0--0	00-- 0--0
9Eh	ADRESL	Least Significant 2 bits of the Left Shifted A/D Result or 8 bits of the Right Shifted Result								xxxx xxxx	uuuu uuuu
9Fh	ANSEL	—	ADCS2	ADCS1	ADCS0	ANS3	ANS2	ANS1	ANS0	-000 1111	-000 1111

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for A/D converter module.

8.0 DATA EEPROM MEMORY

The EEPROM data memory is readable and writable during normal operation (full VDD range). This memory is not directly mapped in the register file space. Instead, it is indirectly addressed through the Special Function Registers. There are four SFRs used to read and write this memory:

- EECON1
- EECON2 (not a physically implemented register)
- EEDATA
- EEADR

EEDATA holds the 8-bit data for read/write, and EEADR holds the address of the EEPROM location being accessed. PIC12F629/675 devices have 128 bytes of data EEPROM with an address range from 0h to 7Fh.

The EEPROM data memory allows byte read and write. A byte write automatically erases the location and writes the new data (erase before write). The EEPROM data memory is rated for high erase/write cycles. The write time is controlled by an on-chip timer. The write time will vary with voltage and temperature as well as from chip to chip. Please refer to AC Specifications for exact limits.

When the data memory is code protected, the CPU may continue to read and write the data EEPROM memory. The device programmer can no longer access this memory.

Additional information on the Data EEPROM is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

REGISTER 8-1: EEDAT — EEPROM DATA REGISTER (ADDRESS: 9Ah)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0
bit 7				bit 0			

bit 7-0 **EEDATn:** Byte value to write to or read from Data EEPROM

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

REGISTER 8-2: EEADR — EEPROM ADDRESS REGISTER (ADDRESS: 9Bh)

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	EADR6	EADR5	EADR4	EADR3	EADR2	EADR1	EADR0
bit 7				bit 0			

bit 7 **Unimplemented:** Should be set to '0'

bit 6-0 **EEADR:** Specifies one of 128 locations for EEPROM Read/Write Operation

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 - n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

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8.1 EEADR

The EEADR register can address up to a maximum of 128 bytes of data EEPROM. Only seven of the eight bits in the register (EEADR<6:0>) are required. The MSb (bit 7) is ignored.

The upper bit should always be '0' to remain upward compatible with devices that have more data EEPROM memory.

8.2 EECON1 AND EECON2 REGISTERS

EECON1 is the control register with four low order bits physically implemented. The upper four bits are non-implemented and read as '0's.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared in hardware at completion

of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a MCLR Reset, or a WDT Time-out Reset during normal operation. In these situations, following RESET, the user can check the WRERR bit, clear it, and rewrite the location. The data and address will be cleared, therefore, the EEDATA and EEADR registers will need to be re-initialized.

Interrupt flag bit EEIF in the PIR1 register is set when write is complete. This bit must be cleared in software.

EECON2 is not a physical register. Reading EECON2 will read all '0's. The EECON2 register is used exclusively in the Data EEPROM write sequence.

REGISTER 8-3: EECON1 — EEPROM CONTROL REGISTER (ADDRESS: 9Ch)

U-0	U-0	U-0	U-0	R/W-x	R/W-0	R/S-0	R/S-0
—	—	—	—	WRERR	WREN	WR	RD
bit 7				bit 0			

bit 7-4 **Unimplemented:** Read as '0'

bit 3 **WRERR:** EEPROM Error Flag bit

- 1 = A write operation is prematurely terminated (any MCLR Reset, any WDT Reset during normal operation or BOD detect)
- 0 = The write operation completed

bit 2 **WREN:** EEPROM Write Enable bit

- 1 = Allows write cycles
- 0 = Inhibits write to the data EEPROM

bit 1 **WR:** Write Control bit

- 1 = Initiates a write cycle (The bit is cleared by hardware once write is complete. The WR bit can only be set, not cleared, in software.)
- 0 = Write cycle to the data EEPROM is complete

bit 0 **RD:** Read Control bit

- 1 = Initiates an EEPROM read (Read takes one cycle. RD is cleared in hardware. The RD bit can only be set, not cleared, in software.)
- 0 = Does not initiate an EEPROM read

Legend:

S = Bit can only be set

R = Readable bit

- n = Value at POR

W = Writable bit

'1' = Bit is set

U = Unimplemented bit, read as '0'

'0' = Bit is cleared

x = Bit is unknown

8.3 READING THE EEPROM DATA MEMORY

To read a data memory location, the user must write the address to the EEADR register and then set control bit RD (EECON1<0>), as shown in Example 8-1. The data is available, in the very next cycle, in the EEDATA register. Therefore, it can be read in the next instruction. EEDATA holds this value until another read, or until it is written to by the user (during a write operation).

EXAMPLE 8-1: DATA EEPROM READ

```
bsf    STATUS,RP0    ;Bank 1
movlw  CONFIG_ADDR   ;
movwf  EEADR          ;Address to read
bsf    EECON1,RD      ;EE Read
movf   EEDATA,W       ;Move data to W
```

8.4 WRITING TO THE EEPROM DATA MEMORY

To write an EEPROM data location, the user must first write the address to the EEADR register and the data to the EEDATA register. Then the user must follow a specific sequence to initiate the write for each byte, as shown in Example 8-2.

EXAMPLE 8-2: DATA EEPROM WRITE

Required Sequence	bsf	STATUS,RP0	;Bank 1
	bsf	EECON1,WREN	;Enable write
	bcf	INTCON,GIE	;Disable INTs
	movlw	55h	;Unlock write
	movwf	EECON2	;
	movlw	AAh	;
	movwf	EECON2	;
	bsf	EECON1,WR	;Start the write
	bsf	INTCON,GIE	;Enable INTs

The write will not initiate if the above sequence is not exactly followed (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. We strongly recommend that interrupts be disabled during this code segment. A cycle count is executed during the required sequence. Any number that is not equal to the required cycles to execute the required sequence will prevent the data from being written into the EEPROM.

Additionally, the WREN bit in EECON1 must be set to enable write. This mechanism prevents accidental writes to data EEPROM due to errant (unexpected) code execution (i.e., lost programs). The user should keep the WREN bit clear at all times, except when updating EEPROM. The WREN bit is not cleared by hardware.

After a write sequence has been initiated, clearing the WREN bit will not affect this write cycle. The WR bit will be inhibited from being set unless the WREN bit is set.

At the completion of the write cycle, the WR bit is cleared in hardware and the EE Write Complete Interrupt Flag bit (EEIF) is set. The user can either enable this interrupt or poll this bit. The EEIF bit (PIR<7>) register must be cleared by software.

8.5 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the Data EEPROM should be verified (see Example 8-3) to the desired value to be written.

EXAMPLE 8-3: WRITE VERIFY

```
bcf    STATUS,RP0    ;Bank 0
:      ;Any code
bsf    STATUS,RP0    ;Bank 1 READ
movf   EEDATA,W       ;EEDATA not changed
                        ;from previous write
bsf    EECON1,RD      ;YES, Read the
                        ;value written
xorwf  EEDATA,W
btfss  STATUS,Z        ;Is data the same
goto   WRITE_ERR      ;No, handle error
:      ;Yes, continue
```

8.5.1 USING THE DATA EEPROM

The Data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that are updated often). Frequently changing values will typically be updated more often than specifications D120 or D120A. If this is not the case, an array refresh must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in FLASH program memory.

8.6 PROTECTION AGAINST SPURIOUS WRITE

There are conditions when the device may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been built in. On power-up, WREN is cleared. Also, the Power-up Timer (72 ms duration) prevents EEPROM write.

The write initiate sequence and the WREN bit together help prevent an accidental write during:

- brown-out
- power glitch
- software malfunction

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8.7 DATA EEPROM OPERATION DURING CODE PROTECT

Data memory can be code protected by programming the CPD bit to '0'.

When the data memory is code protected, the CPU is able to read and write data to the Data EEPROM. It is recommended to code protect the program memory when code protecting data memory. This prevents anyone from programming zeroes over the existing code (which will execute as NOPs) to reach an added routine, programmed in unused program memory, which outputs the contents of data memory. Programming unused locations to '0' will also help prevent data memory code protection from becoming breached.

TABLE 8-1: REGISTERS/BITS ASSOCIATED WITH DATA EEPROM

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS	
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	00-- 0--0	
9Ah	EEDATA	EEPROM Data Register								0000 0000	0000 0000	
9Bh	EEADR	—	EEPROM Address Register								-000 0000	-000 0000
9Ch	EECON1	—	—	—	—	WRERR	WREN	WR	RD	---- x000	---- q000	
9Dh	EECON2 ⁽¹⁾	EEPROM Control Register 2								---- ----	---- ----	

Legend: x = unknown, u = unchanged, - = unimplemented read as '0', q = value depends upon condition.

Shaded cells are not used by Data EEPROM module.

Note 1: EECON2 is not a physical register.

9.0 SPECIAL FEATURES OF THE CPU

Certain special circuits that deal with the needs of real time applications are what sets a microcontroller apart from other processors. The PIC12F629/675 family has a host of such features intended to:

- maximize system reliability
- minimize cost through elimination of external components
- provide power saving operating modes and offer code protection.

These features are:

- Oscillator selection
- RESET
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Detect (BOD)
- Interrupts
- Watchdog Timer (WDT)
- SLEEP
- Code protection
- ID Locations
- In-Circuit Serial Programming

The PIC12F629/675 has a Watchdog Timer that is controlled by configuration bits. It runs off its own RC oscillator for added reliability. There are two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in RESET until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay of 72 ms (nominal) on power-up only, designed to keep the part in RESET while the power supply stabilizes. There is also circuitry to reset the device if a brown-out occurs, which can provide at least a 72 ms RESET. With these three functions on-chip, most applications need no external RESET circuitry.

The SLEEP mode is designed to offer a very low current Power-down mode. The user can wake-up from SLEEP through:

- External RESET
- Watchdog Timer wake-up
- An interrupt

Several oscillator options are also made available to allow the part to fit the application. The INTOSC option saves system cost while the LP crystal option saves power. A set of configuration bits are used to select various options (see Register 9-1).

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9.1 Configuration Bits

The configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1') to select various device configurations, as shown in Register 9-1. These bits are mapped in program memory location 2007h.

Note: Address 2007h is beyond the user program memory space. It belongs to the special configuration memory space (2000h - 3FFFh), which can be accessed only during programming. See PIC12F629/675 Programming Specification for more information.

REGISTER 9-1: CONFIG — CONFIGURATION WORD (ADDRESS: 2007h)

R/P-1	R/P-1	U-0	U-0	U-0	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
BG1	BG0	—	—	—	$\overline{\text{CPD}}$	$\overline{\text{CP}}$	BODEN	MCLRE	$\overline{\text{PWRT}}\overline{\text{E}}$	WDTE	F0SC2	F0SC1	F0SC0	
bit 13														bit 0

bit 13-12 **BG1:BG0:** Bandgap Calibration bits for BOD and POR voltage⁽¹⁾

00 = Lowest bandgap voltage

11 = Highest bandgap voltage

bit 11-9 **Unimplemented:** Read as '0'

bit 8 **CPD:** Data Code Protection bit⁽²⁾

1 = Data memory code protection is disabled

0 = Data memory code protection is enabled

bit 7 **$\overline{\text{CP}}$:** Code Protection bit⁽³⁾

1 = Program Memory code protection is disabled

0 = Program Memory code protection is enabled

bit 6 **BODEN:** Brown-out Detect Enable bit⁽⁴⁾

1 = BOD enabled

0 = BOD disabled

bit 5 **MCLRE:** GP3/ $\overline{\text{MCLR}}$ pin function select⁽⁵⁾

1 = GP3/ $\overline{\text{MCLR}}$ pin function is $\overline{\text{MCLR}}$

0 = GP3/ $\overline{\text{MCLR}}$ pin function is digital I/O, $\overline{\text{MCLR}}$ internally tied to VDD

bit 4 **$\overline{\text{PWRT}}\overline{\text{E}}$:** Power-up Timer Enable bit

1 = PWRT disabled

0 = PWRT enabled

bit 3 **WDTE:** Watchdog Timer Enable bit

1 = WDT enabled

0 = WDT disabled

bit 2-0 **F0SC2:F0SC0:** Oscillator Selection bits

111 = RC oscillator: CLKOUT function on GP4/OSC2/CLKOUT pin, RC on GP5/OSC1/CLKIN

110 = RC oscillator: I/O function on GP4/OSC2/CLKOUT pin, RC on GP5/OSC1/CLKIN

101 = INTOSC oscillator: CLKOUT function on GP4/OSC2/CLKOUT pin, I/O function on GP5/OSC1/CLKIN

100 = INTOSC oscillator: I/O function on GP4/OSC2/CLKOUT pin, I/O function on GP5/OSC1/CLKIN

011 = EC: I/O function on GP4/OSC2/CLKOUT pin, CLKIN on GP5/OSC1/CLKIN

010 = HS oscillator: High speed crystal/resonator on GP4/OSC2/CLKOUT and GP5/OSC1/CLKIN

001 = XT oscillator: Crystal/resonator on GP4/OSC2/CLKOUT and GP5/OSC1/CLKIN

000 = LP oscillator: Low power crystal on GP4/OSC2/CLKOUT and GP5/OSC1/CLKIN

Note 1: The Bandgap Calibration bits are factory programmed and must be read and saved prior to erasing the device as specified in the PIC12F629/675 Programming Specification. These bits are reflected in an export of the configuration word. Microchip Development Tools maintain all calibration bits to factory settings.

2: The entire data EEPROM will be erased when the code protection is turned off.

3: The entire program memory will be erased, including OSCCAL value, when the code protection is turned off.

4: Enabling Brown-out Detect does not automatically enable Power-up Timer.

5: When MCLR is asserted in INTOSC or RC mode, the internal clock oscillator is disabled.

Legend:

P = Programmed using ICSP

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

1 = bit is set

0 = bit is cleared

x = bit is unknown

9.2 Oscillator Configurations

9.2.1 OSCILLATOR TYPES

The PIC12F629/675 can be operated in eight different oscillator option modes. The user can program three configuration bits (FOSC2 through FOSC0) to select one of these eight modes:

- LP Low Power Crystal
- XT Crystal/Resonator
- HS High Speed Crystal/Resonator
- RC External Resistor/Capacitor (2 modes)
- INTOSC Internal Oscillator (2 modes)
- EC External Clock In

Note: Additional information on oscillator configurations is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

9.2.2 CRYSTAL OSCILLATOR / CERAMIC RESONATORS

In XT, LP or HS modes a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation (see Figure 9-1). The PIC12F629/675 oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may yield a frequency outside of the crystal manufacturers specifications. When in XT, LP or HS modes, the device can have an external clock source to drive the OSC1 pin (see Figure 9-2).

FIGURE 9-1: CRYSTAL OPERATION (OR CERAMIC RESONATOR) HS, XT OR LP OSC CONFIGURATION

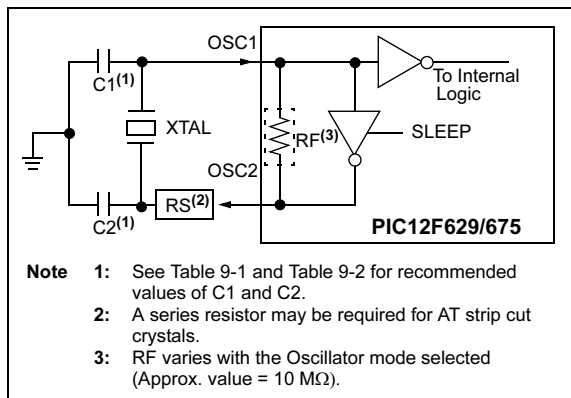


FIGURE 9-2: EXTERNAL CLOCK INPUT OPERATION (HS, XT, EC, OR LP OSC CONFIGURATION)

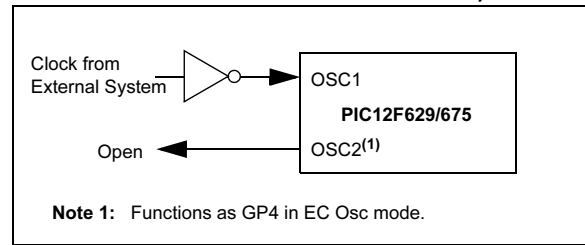


TABLE 9-1: CAPACITOR SELECTION FOR CERAMIC RESONATORS

Ranges Characterized:			
Mode	Freq	OSC1(C1)	OSC2(C2)
XT	455 kHz	68 - 100 pF	68 - 100 pF
	2.0 MHz	15 - 68 pF	15 - 68 pF
	4.0 MHz	15 - 68 pF	15 - 68 pF
HS	8.0 MHz	10 - 68 pF	10 - 68 pF
	16.0 MHz	10 - 22 pF	10 - 22 pF

Note 1: Higher capacitance increases the stability of the oscillator but also increases the start-up time. These values are for design guidance only. Since each resonator has its own characteristics, the user should consult the resonator manufacturer for appropriate values of external components.

TABLE 9-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Mode	Freq	OSC1(C1)	OSC2(C2)
LP	32 kHz	68 - 100 pF	68 - 100 pF
XT	100 kHz	68 - 150 pF	150 - 200 pF
	2 MHz	15 - 30 pF	15 - 30 pF
	4 MHz	15 - 30 pF	15 - 30 pF
HS	8 MHz	15 - 30 pF	15 - 30 pF
	10 MHz	15 - 30 pF	15 - 30 pF
	20 MHz	15 - 30 pF	15 - 30 pF

Note 1: Higher capacitance increases the stability of the oscillator but also increases the start-up time. These values are for design guidance only. Rs may be required in HS mode as well as XT mode to avoid overdriving crystals with low drive level specification. Since each crystal has its own characteristics, the user should consult the crystal manufacturer for appropriate values of external components.

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9.2.3 EXTERNAL CLOCK IN

For applications where a clock is already available elsewhere, users may directly drive the PIC12F629/675 provided that this external clock source meets the AC/DC timing requirements listed in Section 12.0. Figure 9-2 shows how an external clock circuit should be configured.

9.2.4 RC OSCILLATOR

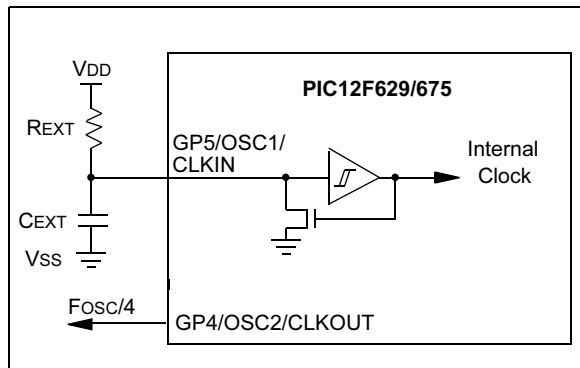
For applications where precise timing is not a requirement, the RC oscillator option is available. The operation and functionality of the RC oscillator is dependent upon a number of variables. The RC oscillator frequency is a function of:

- Supply voltage
- Resistor (REXT) and capacitor (CEXT) values
- Operating temperature.

The oscillator frequency will vary from unit to unit due to normal process parameter variation. The difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low CEXT values. The user also needs to account for the tolerance of the external R and C components. Figure 9-3 shows how the R/C combination is connected.

Two options are available for this Oscillator mode which allow GP4 to be used as a general purpose I/O or to output Fosc/4.

FIGURE 9-3: RC OSCILLATOR MODE



9.2.5 INTERNAL 4 MHz OSCILLATOR

When calibrated, the internal oscillator provides a fixed 4 MHz (nominal) system clock. See Electrical Specifications, Section 12.0, for information on variation over voltage and temperature.

Two options are available for this Oscillator mode which allow GP4 to be used as a general purpose I/O or to output Fosc/4.

9.2.5.1 Calibrating the Internal Oscillator

A calibration instruction is programmed into the last location of program memory. This instruction is a RETLW XX, where the literal is the calibration value. The literal is placed in the OSCCAL register to set the calibration of the internal oscillator. Example 9-1 demonstrates how to calibrate the internal oscillator. For best operation, decouple (with capacitance) VDD and VSS as close to the device as possible.

Note: Erasing the device will also erase the pre-programmed internal calibration value for the internal oscillator. The calibration value must be saved prior to erasing part as specified in the PIC12F629/675 Programming specification. Microchip Development Tools maintain all calibration bits to factory settings.

EXAMPLE 9-1: CALIBRATING THE INTERNAL OSCILLATOR

```
bsf    STATUS, RP0    ;Bank 1
call   3FFh           ;Get the cal value
movwf  OSCCAL         ;Calibrate
bcf    STATUS, RP0    ;Bank 0
```

9.2.6 CLKOUT

The PIC12F629/675 devices can be configured to provide a clock out signal in the INTOSC and RC oscillator modes. When configured, the oscillator frequency divided by four (Fosc/4) is output on the GP4/OSC2/CLKOUT pin. Fosc/4 can be used for test purposes or to synchronize other logic.

9.3 RESET

The PIC12F629/675 differentiates between various kinds of RESET:

- Power-on Reset (POR)
- WDT Reset during normal operation
- WDT Reset during SLEEP
- $\overline{\text{MCLR}}$ Reset during normal operation
- $\overline{\text{MCLR}}$ Reset during SLEEP
- Brown-out Detect (BOD)

Some registers are not affected in any RESET condition; their status is unknown on POR and unchanged in any other RESET. Most other registers are reset to a "RESET state" on:

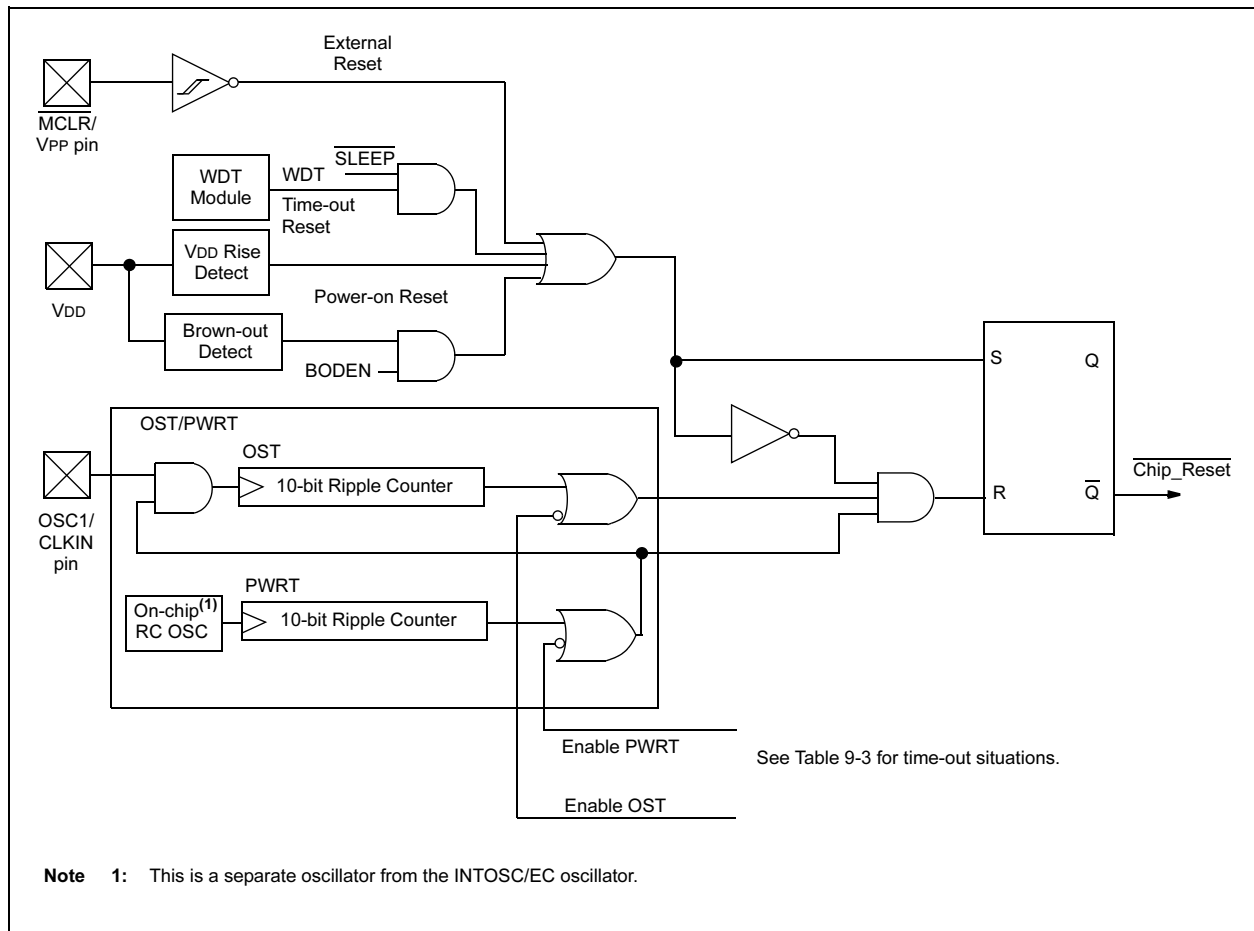
- Power-on Reset
- $\overline{\text{MCLR}}$ Reset
- WDT Reset
- WDT Reset during SLEEP
- Brown-out Detect (BOD) Reset

They are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits are set or cleared differently in different RESET situations as indicated in Table 9-4. These bits are used in software to determine the nature of the RESET. See Table 9-7 for a full description of RESET states of all registers.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 9-4.

The $\overline{\text{MCLR}}$ Reset path has a noise filter to detect and ignore small pulses. See Table 12-4 in Electrical Specifications Section for pulse width specification.

FIGURE 9-4: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



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9.3.1 $\overline{\text{MCLR}}$

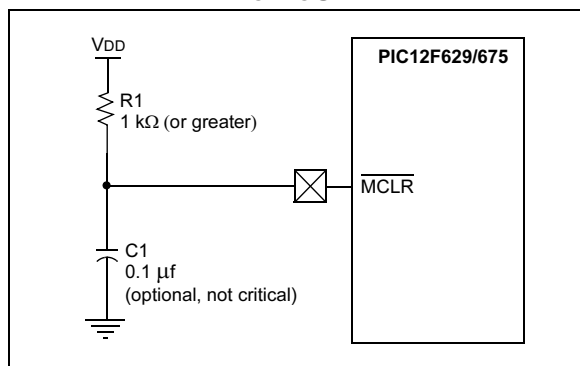
PIC12F629/675 devices have a noise filter in the $\overline{\text{MCLR}}$ Reset path. The filter will detect and ignore small pulses.

It should be noted that a WDT Reset does not drive $\overline{\text{MCLR}}$ pin low.

The behavior of the ESD protection on the $\overline{\text{MCLR}}$ pin has been altered from previous devices of this family. Voltages applied to the pin that exceed its specification can result in both $\overline{\text{MCLR}}$ Resets and excessive current beyond the device specification during the ESD event. For this reason, Microchip recommends that the $\overline{\text{MCLR}}$ pin no longer be tied directly to VDD. The use of an RC network, as shown in Figure 9-5, is suggested.

An internal $\overline{\text{MCLR}}$ option is enabled by setting the $\overline{\text{MCLRE}}$ bit in the configuration word. When enabled, $\overline{\text{MCLR}}$ is internally tied to VDD. No internal pull-up option is available for the $\overline{\text{MCLR}}$ pin.

FIGURE 9-5: RECOMMENDED $\overline{\text{MCLR}}$ CIRCUIT



9.3.2 POWER-ON RESET (POR)

The on-chip POR circuit holds the chip in RESET until VDD has reached a high enough level for proper operation. To take advantage of the POR, simply tie the $\overline{\text{MCLR}}$ pin through a resistor to VDD. This will eliminate external RC components usually needed to create Power-on Reset. A maximum rise time for VDD is required. See Electrical Specifications for details (see Section 12.0). If the BOD is enabled, the maximum rise time specification does not apply. The BOD circuitry will keep the device in RESET until VDD reaches VBOD (see Section 9.3.5).

Note: The POR circuit does not produce an internal RESET when VDD declines.

When the device starts normal operation (exits the RESET condition), device operating parameters (i.e., voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in RESET until the operating conditions are met.

For additional information, refer to Application Note AN607 "Power-up Trouble Shooting".

9.3.3 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 72 ms (nominal) time-out on power-up only, from POR or Brown-out Detect. The Power-up Timer operates on an internal RC oscillator. The chip is kept in RESET as long as PWRT is active. The PWRT delay allows the VDD to rise to an acceptable level. A configuration bit, $\overline{\text{PWRTÉ}}$ can disable (if set) or enable (if cleared or programmed) the Power-up Timer. The Power-up Timer should always be enabled when Brown-out Detect is enabled.

The Power-up Time delay will vary from chip to chip and due to:

- VDD variation
- Temperature variation
- Process variation.

See DC parameters for details (Section 12.0).

9.3.4 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-up Timer (OST) provides a 1024 oscillator cycle (from OSC1 input) delay after the PWRT delay is over. This ensures that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP and HS modes and only on Power-on Reset or wake-up from SLEEP.

9.3.5 BROWN-OUT DETECT (BOD)

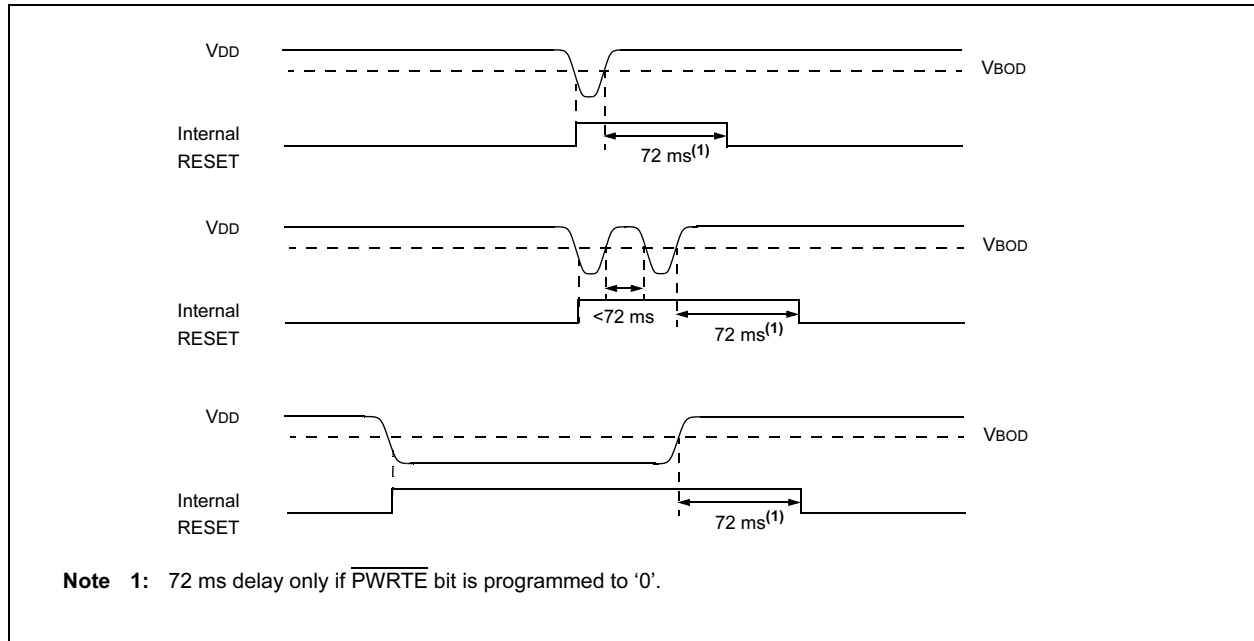
The PIC12F629/675 members have on-chip Brown-out Detect circuitry. A configuration bit, BODEN, can disable (if clear/programmed) or enable (if set) the Brown-out Detect circuitry. If VDD falls below VBOD for greater than parameter (TBOD) in Table 12-4 (see Section 12.0), the Brown-out situation will reset the device. This will occur regardless of VDD slew-rate. A RESET is not guaranteed to occur if VDD falls below VBOD for less than parameter (TBOD).

On any RESET (Power-on, Brown-out, Watchdog, etc.), the chip will remain in RESET until VDD rises above BVDD (see Figure 9-6). The Power-up Timer will now be invoked, if enabled, and will keep the chip in RESET an additional 72 ms.

Note: A Brown-out Detect does not enable the Power-up Timer if the $\overline{\text{PWRTE}}$ bit in the configuration word is set.

If VDD drops below BVDD while the Power-up Timer is running, the chip will go back into a Brown-out Detect and the Power-up Timer will be re-initialized. Once VDD rises above BVDD, the Power-up Timer will execute a 72 ms RESET.

FIGURE 9-6: BROWN-OUT SITUATIONS



9.3.6 TIME-OUT SEQUENCE

On power-up, the time-out sequence is as follows: first, PWRT time-out is invoked after POR has expired. Then, OST is activated. The total time-out will vary based on oscillator configuration and $\overline{\text{PWRTE}}$ bit status. For example, in EC mode with $\overline{\text{PWRTE}}$ bit erased (PWRT disabled), there will be no time-out at all. Figure 9-7, Figure 9-8 and Figure 9-9 depict time-out sequences.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, the time-outs will expire. Then bringing $\overline{\text{MCLR}}$ high will begin execution immediately (see Figure 9-8). This is useful for testing purposes or to synchronize more than one PIC12F629/675 device operating in parallel.

Table 9-6 shows the RESET conditions for some special registers, while Table 9-7 shows the RESET conditions for all the registers.

9.3.7 POWER CONTROL (PCON) STATUS REGISTER

The power CONTROL/STATUS register, PCON (address 8Eh) has two bits.

Bit0 is $\overline{\text{BOD}}$ (Brown-out). $\overline{\text{BOD}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent RESETS to see if $\overline{\text{BOD}} = 0$, indicating that a brown-out has occurred. The $\overline{\text{BOD}}$ STATUS bit is a don't care and is not necessarily predictable if the brown-out circuit is disabled (by setting BODEN bit = 0 in the Configuration word).

Bit1 is $\overline{\text{POR}}$ (Power-on Reset). It is a '0' on Power-on Reset and unaffected otherwise. The user must write a '1' to this bit following a Power-on Reset. On a subsequent RESET, if POR is '0', it will indicate that a Power-on Reset must have occurred (i.e., VDD may have gone too low).

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TABLE 9-3: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up		Brown-out Detect		Wake-up from SLEEP
	$\overline{\text{PWRTE}} = 0$	$\overline{\text{PWRTE}} = 1$	$\overline{\text{PWRTE}} = 0$	$\overline{\text{PWRTE}} = 1$	
XT, HS, LP	TPWRT + 1024•Tosc	1024•Tosc	TPWRT + 1024•Tosc	1024•Tosc	1024•Tosc
RC, EC, INTOSC	TPWRT	—	TPWRT	—	—

TABLE 9-4: STATUS/PCON BITS AND THEIR SIGNIFICANCE

$\overline{\text{POR}}$	$\overline{\text{BOD}}$	$\overline{\text{TO}}$	$\overline{\text{PD}}$	
0	u	1	1	Power-on Reset
1	0	1	1	Brown-out Detect
u	u	0	u	WDT Reset
u	u	0	0	WDT Wake-up
u	u	u	u	$\overline{\text{MCLR}}$ Reset during normal operation
u	u	1	0	$\overline{\text{MCLR}}$ Reset during SLEEP

Legend: u = unchanged, x = unknown

TABLE 9-5: SUMMARY OF REGISTERS ASSOCIATED WITH BROWN-OUT

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS ⁽¹⁾
03h	STATUS	IRP	RP1	RPO	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	0001 1xxx	000q quuu
8Eh	PCON	—	—	—	—	—	—	$\overline{\text{POR}}$	$\overline{\text{BOD}}$	---- --0x	---- --uq

Legend: u = unchanged, x = unknown, – = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: Other (non Power-up) Resets include $\overline{\text{MCLR}}$ Reset, Brown-out Detect and Watchdog Timer Reset during normal operation.

TABLE 9-6: INITIALIZATION CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	000h	0001 1xxx	---- --0x
$\overline{\text{MCLR}}$ Reset during normal operation	000h	000u uuuu	---- --uu
$\overline{\text{MCLR}}$ Reset during SLEEP	000h	0001 0uuu	---- --uu
WDT Reset	000h	0000 uuuu	---- --uu
WDT Wake-up	PC + 1	uuu0 0uuu	---- --uu
Brown-out Detect	000h	0001 1uuu	---- --10
Interrupt Wake-up from SLEEP	PC + 1 ⁽¹⁾	uuu1 0uuu	---- --uu

Legend: u = unchanged, x = unknown, – = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and global enable bit GIE is set, the PC is loaded with the interrupt vector (0004h) after execution of PC+1.

TABLE 9-7: INITIALIZATION CONDITION FOR REGISTERS

Register	Address	Power-on Reset	<ul style="list-style-type: none"> MCLR Reset during normal operation MCLR Reset during SLEEP WDT Reset Brown-out Detect⁽¹⁾ 	<ul style="list-style-type: none"> Wake-up from SLEEP through interrupt Wake-up from SLEEP through WDT time-out
W	—	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h/80h	—	—	—
TMR0	01h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h/82h	0000 0000	0000 0000	PC + 1 ⁽³⁾
STATUS	03h/83h	0001 1xxx	000q quuu ⁽⁴⁾	uuuq quuu ⁽⁴⁾
FSR	04h/84h	xxxx xxxx	uuuu uuuu	uuuu uuuu
GPIO	05h	--xx xxxx	--uu uuuu	--uu uuuu
PCLATH	0Ah/8Ah	---0 0000	---0 0000	---u uuuu
INTCON	0Bh/8Bh	0000 0000	0000 000u	uuuu uuq ⁽²⁾
PIR1	0Ch	00-- 0--0	00-- 0--0	qq-- q--q ^(2,5)
T1CON	10h	-000 0000	-uuu uuuu	-uuu uuuu
CMCON	19h	-0-0 0000	-0-0 0000	-u-u uuuu
ADRESH	1Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADCON0	1Fh	00-- 0000	00-- 0000	uu-- uuuu
OPTION_REG	81h	1111 1111	1111 1111	uuuu uuuu
TRISIO	85h	--11 1111	--11 1111	--uu uuuu
PIE1	8Ch	00-- 0--0	00-- 0--0	uu-- u--u
PCON	8Eh	---- --0x	---- --uu ^(1,6)	---- --uu
OSCCAL	90h	1000 00--	1000 00--	uuuu uu--
WPU	95h	--11 -111	--11 -111	uuuu uuuu
IOC	96h	--00 0000	--00 0000	--uu uuuu
VRCON	99h	0-0- 0000	0-0- 0000	u-u- uuuu
EEDATA	9Ah	0000 0000	0000 0000	uuuu uuuu
EEADR	9Bh	-000 0000	-000 0000	-uuu uuuu
EECON1	9Ch	---- x000	---- q000	---- uuuu
EECON2	9Dh	---- ----	---- ----	---- ----
ADRESL	9Eh	xxxx xxxx	uuuu uuuu	uuuu uuuu
ANSEL	9Fh	-000 1111	-000 1111	-uuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: If V_{DD} goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON and/or PIR1 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 9-6 for RESET value for specific condition.

5: If wake-up was due to data EEPROM write completing, Bit 7 = 1; A/D conversion completing, Bit 6 = 1; Comparator input changing, bit 3 = 1; or Timer1 rolling over, bit 0 = 1. All other interrupts generating a wake-up will cause these bits to = u.

6: If RESET was due to brown-out, then bit 0 = 0. All other RESETS will cause bit 0 = u.

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FIGURE 9-7: TIME-OUT SEQUENCE ON POWER-UP ($\overline{\text{MCLR}}$ NOT TIED TO V_{DD}): CASE 1

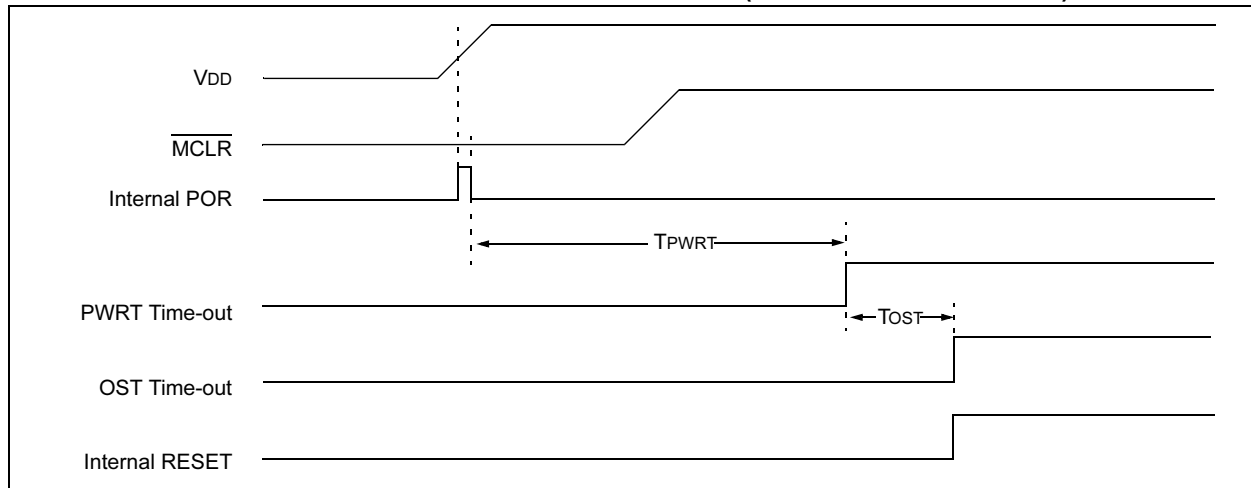


FIGURE 9-8: TIME-OUT SEQUENCE ON POWER-UP ($\overline{\text{MCLR}}$ NOT TIED TO V_{DD}): CASE 2

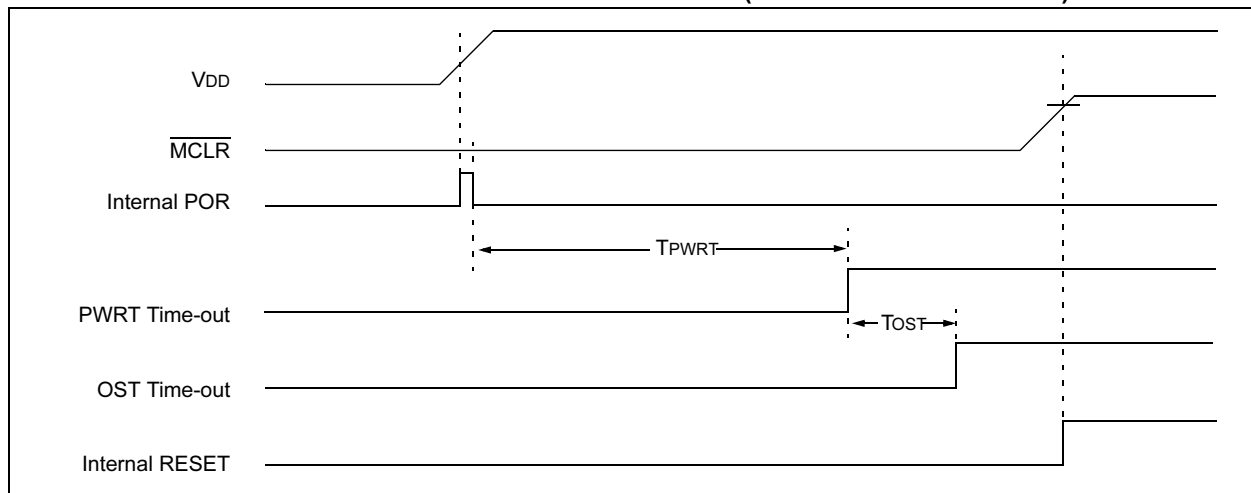
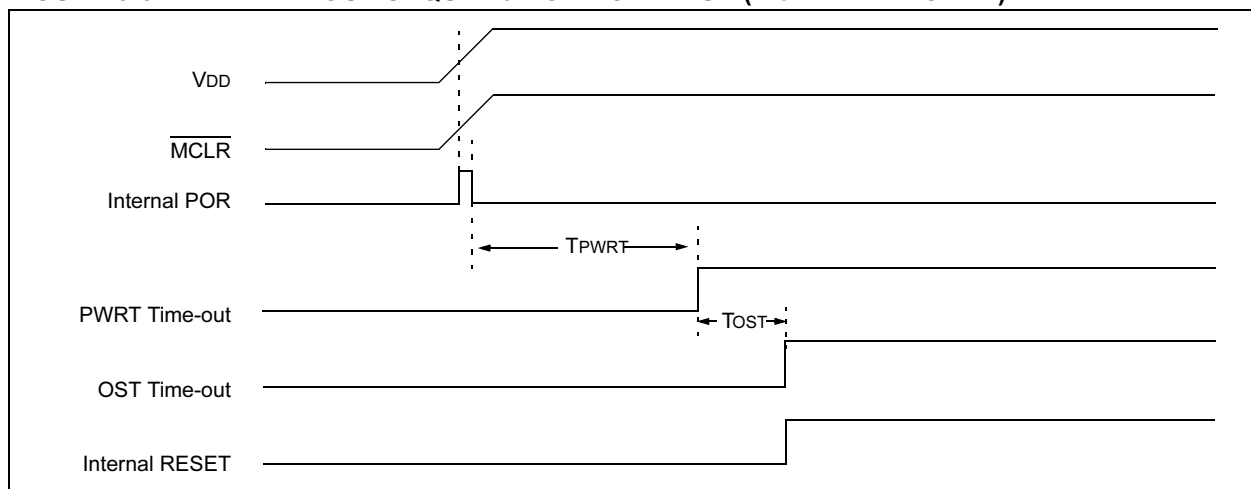


FIGURE 9-9: TIME-OUT SEQUENCE ON POWER-UP ($\overline{\text{MCLR}}$ TIED TO V_{DD})



9.4 Interrupts

The PIC12F629/675 has 7 sources of interrupt:

- External Interrupt GP2/INT
- TMR0 Overflow Interrupt
- GPIO Change Interrupts
- Comparator Interrupt
- A/D Interrupt (PIC12F675 only)
- TMR1 Overflow Interrupt
- EEPROM Data Write Interrupt

The Interrupt Control register (INTCON) and Peripheral Interrupt register (PIR) record individual interrupt requests in flag bits. The INTCON register also has individual and global interrupt enable bits.

A global interrupt enable bit, GIE (INTCON<7>) enables (if set) all unmasked interrupts, or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in INTCON register and PIE register. GIE is cleared on RESET.

The return from interrupt instruction, RETFIE, exits interrupt routine, as well as sets the GIE bit, which re-enables unmasked interrupts.

The following interrupt flags are contained in the INTCON register:

- INT pin interrupt
- GP port change interrupt
- TMR0 overflow interrupt

The peripheral interrupt flags are contained in the special register PIR1. The corresponding interrupt enable bit is contained in Special Register PIE1.

The following interrupt flags are contained in the PIR register:

- EEPROM data write interrupt
- A/D interrupt
- Comparator interrupt
- Timer1 overflow interrupt

When an interrupt is serviced:

- The GIE is cleared to disable any further interrupt
- The return address is pushed onto the stack
- The PC is loaded with 0004h

Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid GP2/INT recursive interrupts.

For external interrupt events, such as the INT pin, or GP port change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends upon when the interrupt event occurs (see Figure 9-11). The latency is the same for one or two-cycle instructions. Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The

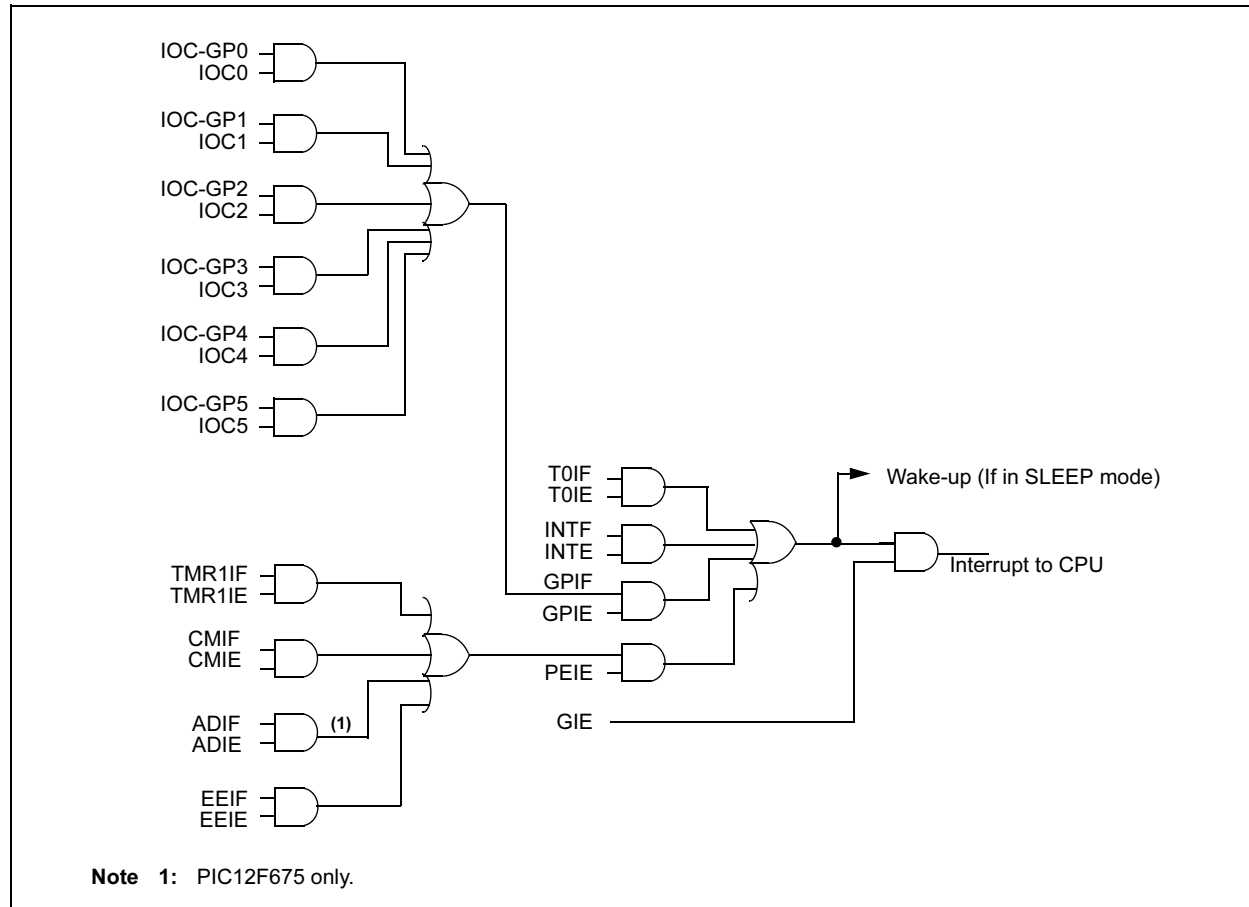
interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

Note 1: Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.

2: When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The interrupts which were ignored are still pending to be serviced when the GIE bit is set again.

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FIGURE 9-10: INTERRUPT LOGIC



9.4.1 GP2/INT INTERRUPT

External interrupt on GP2/INT pin is edge-triggered; either rising if INTEDG bit (OPTION<6>) is set, of falling, if INTEDG bit is clear. When a valid edge appears on the GP2/INT pin, the INTF bit (INTCON<1>) is set. This interrupt can be disabled by clearing the INTE control bit (INTCON<4>). The INTF bit must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The GP2/INT interrupt can wake-up the processor from SLEEP if the INTE bit was set prior to going into SLEEP. The status of the GIE bit decides whether or not the processor branches to the interrupt vector following wake-up. See Section 9.7 for details on SLEEP and Figure 9-13 for timing of wake-up from SLEEP through GP2/INT interrupt.

Note: The ANSEL (9Fh) and CMCON (19h) registers must be initialized to configure an analog channel as a digital input. Pins configured as analog inputs will read '0'. The ANSEL register is defined for the PIC12F675.

9.4.2 TMR0 INTERRUPT

An overflow (FFh → 00h) in the TMR0 register will set the T0IF (INTCON<2>) bit. The interrupt can be enabled/disabled by setting/clearing T0IE (INTCON<5>) bit. For operation of the Timer0 module, see Section 4.0.

9.4.3 GPIO INTERRUPT

An input change on GPIO change sets the GPIF (INTCON<0>) bit. The interrupt can be enabled/disabled by setting/clearing the GPIE (INTCON<3>) bit. Plus individual pins can be configured through the IOC register.

Note: If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the GPIF interrupt flag may not get set.

9.4.4 COMPARATOR INTERRUPT

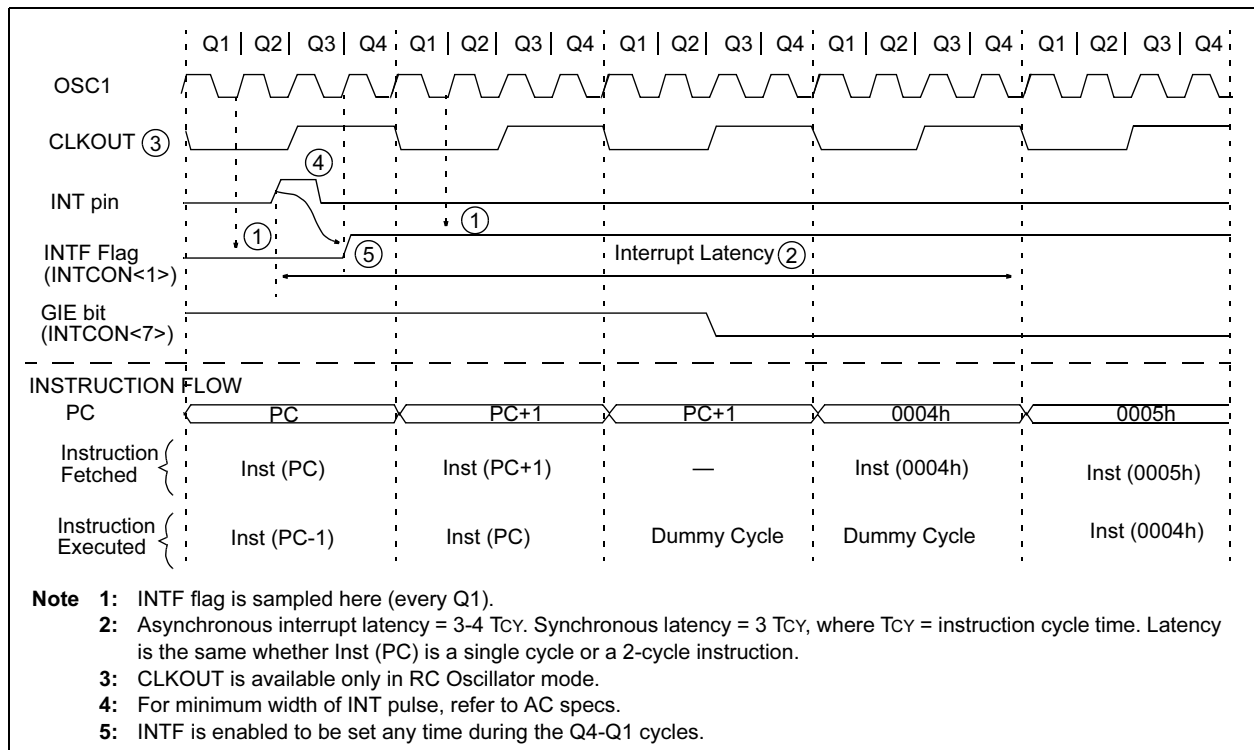
See Section 6.9 for description of comparator interrupt.

9.4.5 A/D CONVERTER INTERRUPT

After a conversion is complete, the ADIF flag (PIR<6>) is set. The interrupt can be enabled/disabled by setting or clearing ADIE (PIE<6>).

See Section 7.0 for operation of the A/D converter interrupt.

FIGURE 9-11: INT PIN INTERRUPT TIMING



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TABLE 9-8: SUMMARY OF INTERRUPT REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS
0Bh, 8Bh	INTCON	GIE	PEIE	TOIE	INTE	GPIE	TOIF	INTF	GPIF	0000 0000	0000 000u
0Ch	PIR1	EEIF	ADIF	—	—	CMIF	—	—	TMR1IF	00-- 0--0	00-- 0--0
8Ch	PIE1	EEIE	ADIE	—	—	CMIE	—	—	TMR1IE	00-- 0--0	00-- 0--0

Legend: x = unknown, u = unchanged, - = unimplemented read as '0', q = value depends upon condition.

Shaded cells are not used by the Interrupt module.

9.5 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt, e.g., W register and STATUS register. This must be implemented in software.

Example 9-2 stores and restores the STATUS and W registers. The user register, W_TEMP, must be defined in both banks and must be defined at the same offset from the bank base address (i.e., W_TEMP is defined at 0x20 in Bank 0 and it must also be defined at 0xA0 in Bank 1). The user register, STATUS_TEMP, must be defined in Bank 0. The Example 9-2:

- Stores the W register
- Stores the STATUS register in Bank 0
- Executes the ISR code
- Restores the STATUS (and bank select bit register)
- Restores the W register

EXAMPLE 9-2: SAVING THE STATUS AND W REGISTERS IN RAM

```

MOVWF  W_TEMP      ;copy W to temp register,
                    ;could be in either bank
SWAPF  STATUS,W     ;swap status to be saved into W
BCF    STATUS,RP0    ;change to bank 0 regardless of
                    ;current bank
MOVWF  STATUS_TEMP  ;save status to bank 0 register
:
: (ISR)
:
SWAPF  STATUS_TEMP,W;swap STATUS_TEMP register into
                    ;W, sets bank to original state
MOVWF  STATUS       ;move W into STATUS register
SWAPF  W_TEMP,F     ;swap W_TEMP
SWAPF  W_TEMP,W     ;swap W_TEMP into W

```

9.6 Watchdog Timer (WDT)

The Watchdog Timer is a free running, on-chip RC oscillator, which requires no external components. This RC oscillator is separate from the external RC oscillator of the CLKIN pin and INTOSC. That means that the WDT will run, even if the clock on the OSC1 and OSC2 pins of the device has been stopped (for example, by execution of a SLEEP instruction). During normal operation, a WDT time-out generates a device RESET. If the device is in SLEEP mode, a WDT time-out causes the device to wake-up and continue with normal operation. The WDT can be permanently disabled by programming the configuration bit WDTE as clear (Section 9.1).

9.6.1 WDT PERIOD

The WDT has a nominal time-out period of 18 ms, (with no prescaler). The time-out periods vary with temperature, VDD and process variations from part to part (see DC specs). If longer time-out periods are desired, a prescaler with a division ratio of up to 1:128 can be assigned to the WDT under software control by writing to the OPTION register. Thus, time-out periods up to 2.3 seconds can be realized.

The CLRWDT and SLEEP instructions clear the WDT and the prescaler, if assigned to the WDT, and prevent it from timing out and generating a device RESET.

The \overline{TO} bit in the STATUS register will be cleared upon a Watchdog Timer time-out.

9.6.2 WDT PROGRAMMING CONSIDERATIONS

It should also be taken in account that under worst case conditions (i.e., VDD = Min., Temperature = Max., Max. WDT prescaler) it may take several seconds before a WDT time-out occurs.

FIGURE 9-12: WATCHDOG TIMER BLOCK DIAGRAM

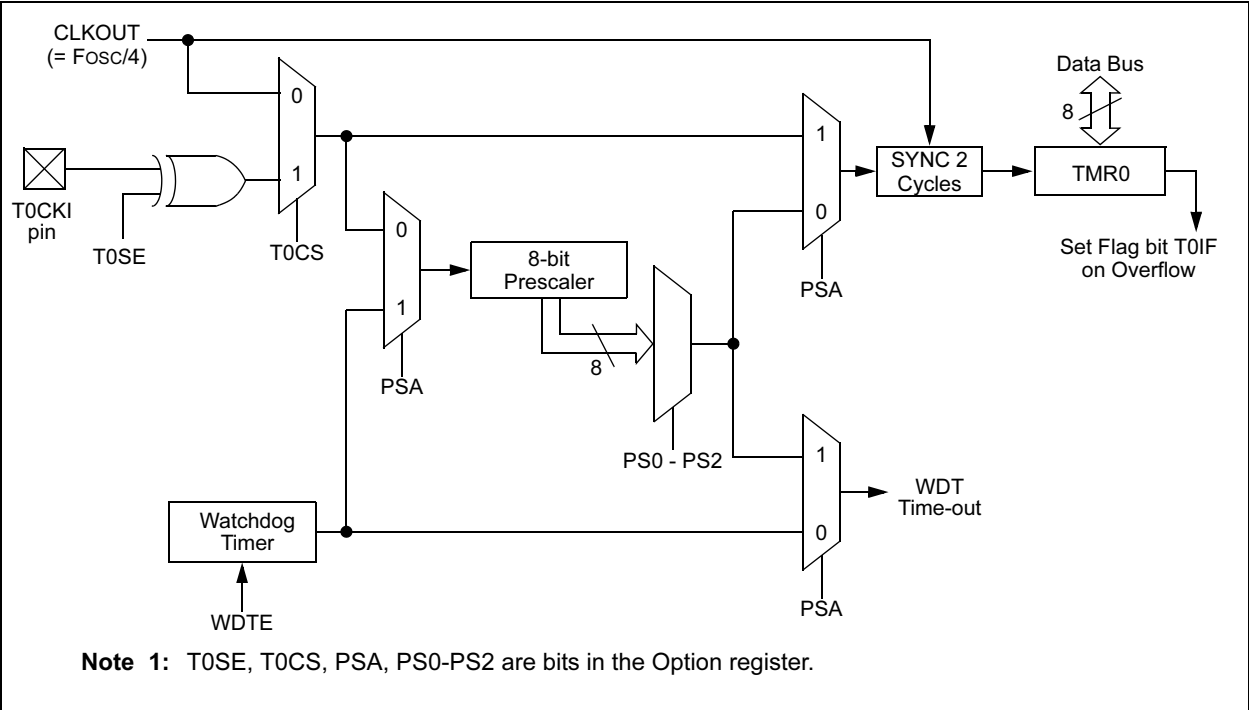


TABLE 9-9: SUMMARY OF WATCHDOG TIMER REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOD	Value on all other RESETS
81h	OPTION_REG	$\overline{\text{GPPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
2007h	Config. bits	$\overline{\text{CP}}$	BODEN	MCLRE	$\overline{\text{PWRT}}\overline{\text{E}}$	WDTE	F0SC2	F0SC1	F0SC0	uuuu uuuu	uuuu uuuu

Legend: u = Unchanged, shaded cells are not used by the Watchdog Timer.

9.7 Power-Down Mode (SLEEP)

The Power-down mode is entered by executing a SLEEP instruction.

If the Watchdog Timer is enabled:

- WDT will be cleared but keeps running
- $\overline{\text{PD}}$ bit in the STATUS register is cleared
- $\overline{\text{TO}}$ bit is set
- Oscillator driver is turned off
- I/O ports maintain the status they had before SLEEP was executed (driving high, low, or hi-impedance).

For lowest current consumption in this mode, all I/O pins should be either at VDD, or VSS, with no external circuitry drawing current from the I/O pin and the comparators and CVREF should be disabled. I/O pins that are hi-impedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The T0CKI input should also be at VDD or VSS for lowest current consumption. The contribution from on-chip pull-ups on GPIO should be considered.

The MCLR pin must be at a logic high level (VIHMC).

Note: It should be noted that a RESET generated by a WDT time-out does not drive MCLR pin low.

The first event will cause a device RESET. The two latter events are considered a continuation of program execution. The $\overline{\text{TO}}$ and $\overline{\text{PD}}$ bits in the STATUS register can be used to determine the cause of device RESET. The $\overline{\text{PD}}$ bit, which is set on power-up, is cleared when SLEEP is invoked. $\overline{\text{TO}}$ bit is cleared if WDT Wake-up occurred.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is pre-fetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up is regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction, then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have an NOP after the SLEEP instruction.

Note: If the global interrupts are disabled (GIE is cleared), but any interrupt source has both its interrupt enable bit and the corresponding interrupt flag bits set, the device will immediately wake-up from SLEEP. The SLEEP instruction is completely executed.

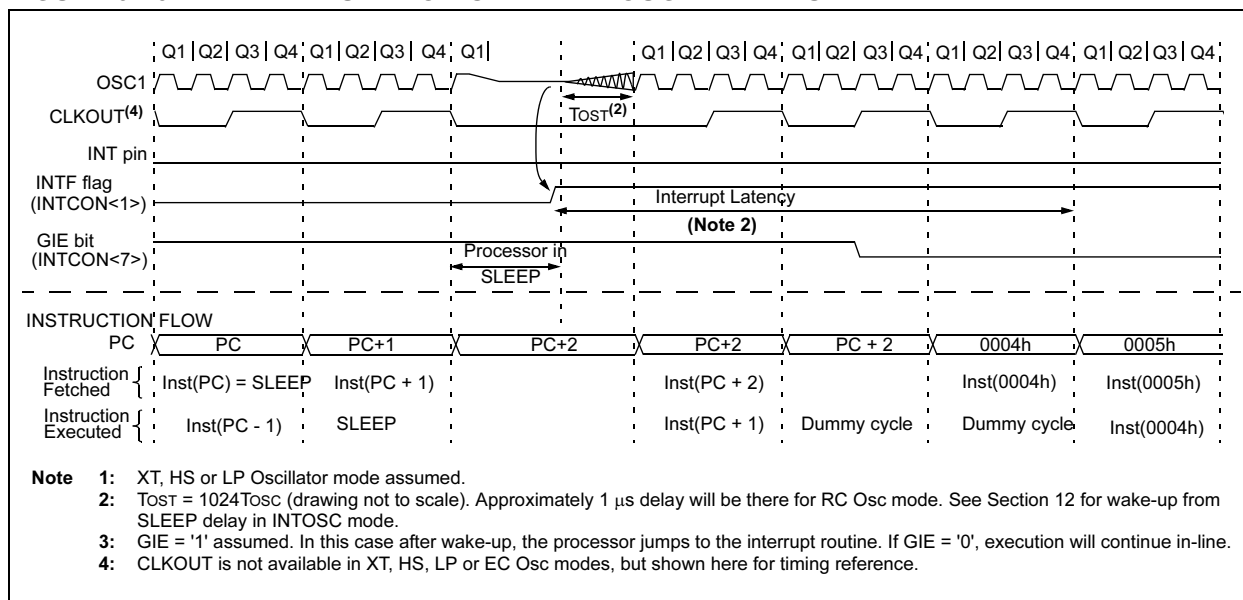
The WDT is cleared when the device wakes up from SLEEP, regardless of the source of wake-up.

9.7.1 WAKE-UP FROM SLEEP

The device can wake-up from SLEEP through one of the following events:

1. External RESET input on MCLR pin
2. Watchdog Timer Wake-up (if WDT was enabled)
3. Interrupt from GP2/INT pin, GPIO change, or a peripheral interrupt.

FIGURE 9-13: WAKE-UP FROM SLEEP THROUGH INTERRUPT



9.8 Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

Note: The entire data EEPROM and FLASH program memory will be erased when the code protection is turned off. The INTOSC calibration data is also erased. See PIC12F629/675 Programming Specification for more information.

9.9 ID Locations

Four memory locations (2000h-2003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution but are readable and writable during Program/Verify. Only the Least Significant 7 bits of the ID locations are used.

9.10 In-Circuit Serial Programming

The PIC12F629/675 microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for:

- power
- ground
- programming voltage

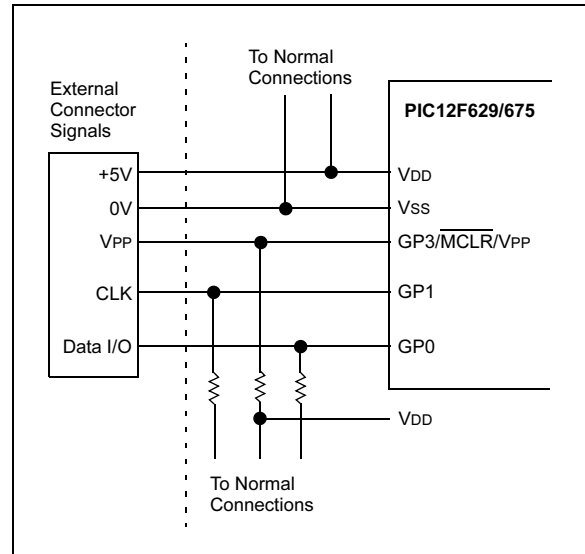
This allows customers to manufacture boards with unprogrammed devices, and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

The device is placed into a Program/Verify mode by holding the GP0 and GP1 pins low, while raising the MCLR (VPP) pin from VIL to VIH (see Programming Specification). GP0 becomes the programming data and GP1 becomes the programming clock. Both GP0 and GP1 are Schmitt Trigger inputs in this mode.

After RESET, to place the device into Programming/Verify mode, the program counter (PC) is at location 00h. A 6-bit command is then supplied to the device. Depending on the command, 14-bits of program data are then supplied to or from the device, depending on whether the command was a load or a read. For complete details of serial programming, please refer to the Programming Specifications.

A typical In-Circuit Serial Programming connection is shown in Figure 9-14.

FIGURE 9-14: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING CONNECTION



9.11 In-Circuit Debugger

Since in-circuit debugging requires the loss of clock, data and MCLR pins, MPLAB® ICD 2 development with an 8-pin device is not practical. A special 14-pin PIC12F675-ICD device is used with MPLAB ICD 2 to provide separate clock, data and MCLR pins and frees all normally available pins to the user.

This special ICD device is mounted on the top of the header and its signals are routed to the MPLAB ICD 2 connector. On the bottom of the header is an 8-pin socket that plugs into the user's target via the 8-pin stand-off connector.

When the $\overline{\text{ICD}}$ pin on the PIC12F675-ICD device is held low, the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB ICD 2. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 9-10 shows which features are consumed by the background debugger:

TABLE 9-10: DEBUGGER RESOURCES

I/O pins	ICDCLK, ICDDATA
Stack	1 level
Program Memory	Address 0h must be NOP 300h - 3FEh

For more information, see 8-Pin MPLAB ICD 2 Header Information Sheet (DS51292) available on Microchip's website (www.microchip.com).

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NOTES:

10.0 INSTRUCTION SET SUMMARY

The PIC12F629/675 instruction set is highly orthogonal and is comprised of three basic categories:

- **Byte-oriented** operations
- **Bit-oriented** operations
- **Literal and control** operations

Each PIC12F629/675 instruction is a 14-bit word divided into an **opcode**, which specifies the instruction type, and one or more **operands**, which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 10-1, while the various opcode fields are summarized in Table 10-1.

Table 10-2 lists the instructions recognized by the MPASM™ assembler. A complete description of each instruction is also available in the PICmicro™ Mid-Range Reference Manual (DS33023).

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator, which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an 8-bit or 11-bit constant, or literal value.

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a normal instruction execution time of 1 μs. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles, with the second cycle executed as a NOP.

Note: To maintain upward compatibility with future products, do not use the **OPTION** and **TRISIO** instructions.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

10.1 READ-MODIFY-WRITE OPERATIONS

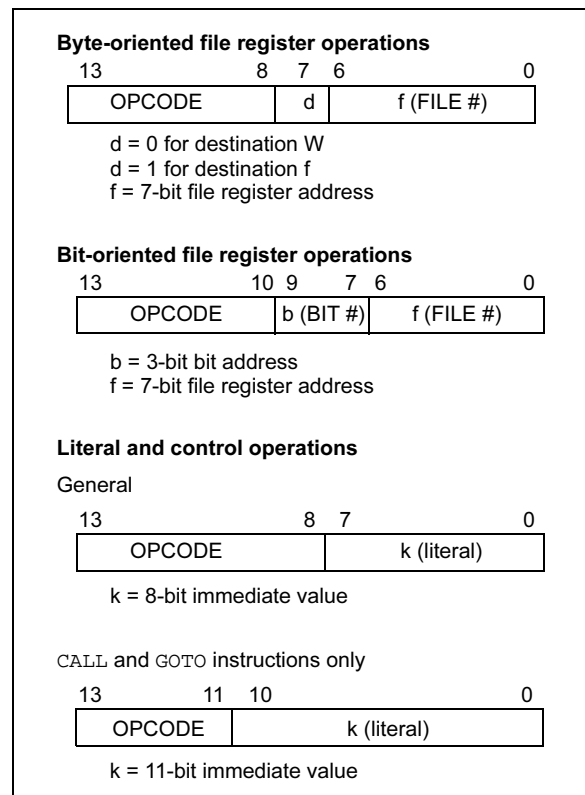
Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction, or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

For example, a **CLRF GPIO** instruction will read GPIO, clear all the data bits, then write the result back to GPIO. This example would have the unintended result that the condition that sets the GPIF flag would be cleared.

TABLE 10-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.
PC	Program Counter
TO	Time-out bit
PD	Power-down bit

FIGURE 10-1: GENERAL FORMAT FOR INSTRUCTIONS



PIC12F629/675

TABLE 10-2: PIC12F629/675 INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	14-Bit Opcode				Status Affected	Notes
				MSb		LSb			
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	1,2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1,2
CLRF	f	Clear f	1	00	0001	1fff	ffff	Z	2
CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1,2
DECf	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1,2
DECFSZ	f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1,2,3
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1,2
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1,2,3
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1,2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1,2
MOVWF	f	Move W to f	1	00	0000	1fff	ffff		
NOP	-	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	1,2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	1,2
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	1,2
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1,2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1,2
BIT-ORIENTED FILE REGISTER OPERATIONS									
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		1,2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		1,2
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
LITERAL AND CONTROL OPERATIONS									
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	-	Clear Watchdog Timer	1	00	0000	0110	0100	$\overline{TO}, \overline{PD}$	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	-	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	-	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	-	Go into Standby mode	1	00	0000	0110	0011	$\overline{TO}, \overline{PD}$	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk	kkkk	C,DC,Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

- Note 1:** When an I/O register is modified as a function of itself (e.g., `MOVF GPIO, 1`), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.
- Note 2:** If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.
- Note 3:** If Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

Note: Additional information on the mid-range instruction set is available in the PICmicro™ Mid-Range MCU Family Reference Manual (DS33023).

10.2 Instruction Descriptions

ADDLW Add Literal and W

Syntax: `[label] ADDLW k`

Operands: $0 \leq k \leq 255$

Operation: $(W) + k \rightarrow (W)$

Status Affected: C, DC, Z

Description: The contents of the W register are added to the eight-bit literal 'k' and the result is placed in the W register.

ADDWF Add W and f

Syntax: `[label] ADDWF f,d`

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(W) + (f) \rightarrow (\text{destination})$

Status Affected: C, DC, Z

Description: Add the contents of the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

ANDLW AND Literal with W

Syntax: `[label] ANDLW k`

Operands: $0 \leq k \leq 255$

Operation: $(W) .\text{AND.} (k) \rightarrow (W)$

Status Affected: Z

Description: The contents of W register are AND'd with the eight-bit literal 'k'. The result is placed in the W register.

ANDWF AND W with f

Syntax: `[label] ANDWF f,d`

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(W) .\text{AND.} (f) \rightarrow (\text{destination})$

Status Affected: Z

Description: AND the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

BCF Bit Clear f

Syntax: `[label] BCF f,b`

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: $0 \rightarrow (f)$

Status Affected: None

Description: Bit 'b' in register 'f' is cleared.

BSF Bit Set f

Syntax: `[label] BSF f,b`

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: $1 \rightarrow (f)$

Status Affected: None

Description: Bit 'b' in register 'f' is set.

BTFSS Bit Test f, Skip if Set

Syntax: `[label] BTFSS f,b`

Operands: $0 \leq f \leq 127$
 $0 \leq b < 7$

Operation: skip if $(f) = 1$

Status Affected: None

Description: If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2TCY instruction.

BTFSC Bit Test, Skip if Clear

Syntax: `[label] BTFSC f,b`

Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$

Operation: skip if $(f) = 0$

Status Affected: None

Description: If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2TCY instruction.

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CALL Call Subroutine

Syntax: `[label] CALL k`
Operands: $0 \leq k \leq 2047$
Operation: $(PC)+1 \rightarrow TOS$,
 $k \rightarrow PC<10:0>$,
 $(PCLATH<4:3>) \rightarrow PC<12:11>$
Status Affected: None
Description: Call Subroutine. First, return address $(PC+1)$ is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits $<10:0>$. The upper bits of the PC are loaded from PCLATH. `CALL` is a two-cycle instruction.

CLRWDTClear Watchdog Timer

Syntax: `[label] CLRWDTClear Watchdog Timer`
Operands: None
Operation: $00h \rightarrow WDT$
 $0 \rightarrow WDT \text{ prescaler}$,
 $1 \rightarrow \overline{TO}$
 $1 \rightarrow \overline{PD}$
Status Affected: \overline{TO} , \overline{PD}
Description: `CLRWDTClear Watchdog Timer` instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. STATUS bits \overline{TO} and \overline{PD} are set.

CLRF Clear f

Syntax: `[label] CLRF f`
Operands: $0 \leq f \leq 127$
Operation: $00h \rightarrow (f)$
 $1 \rightarrow Z$
Status Affected: Z
Description: The contents of register 'f' are cleared and the Z bit is set.

COMF Complement f

Syntax: `[label] COMF f,d`
Operands: $0 \leq f \leq 127$
 $d \in [0,1]$
Operation: $(\bar{f}) \rightarrow (\text{destination})$
Status Affected: Z
Description: The contents of register 'f' are complemented. If 'd' is 0, the result is stored in W. If 'd' is 1, the result is stored back in register 'f'.

CLRWClear W

Syntax: `[label] CLRW`
Operands: None
Operation: $00h \rightarrow (W)$
 $1 \rightarrow Z$
Status Affected: Z
Description: W register is cleared. Zero bit (Z) is set.

DECF Decrement f

Syntax: `[label] DECF f,d`
Operands: $0 \leq f \leq 127$
 $d \in [0,1]$
Operation: $(f) - 1 \rightarrow (\text{destination})$
Status Affected: Z
Description: Decrement register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

DECFSZ Decrement f, Skip if 0

Syntax: [*label*] DECFSZ f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) - 1 \rightarrow (\text{destination});$
skip if result = 0

Status Affected: None

Description: The contents of register 'f' are decremented. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.
If the result is 1, the next instruction is executed. If the result is 0, then a NOP is executed instead, making it a 2TCY instruction.

INCFSZ Increment f, Skip if 0

Syntax: [*label*] INCFSZ f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) + 1 \rightarrow (\text{destination});$
skip if result = 0

Status Affected: None

Description: The contents of register 'f' are incremented. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.
If the result is 1, the next instruction is executed. If the result is 0, a NOP is executed instead, making it a 2TCY instruction.

GOTO Unconditional Branch

Syntax: [*label*] GOTO k

Operands: $0 \leq k \leq 2047$

Operation: $k \rightarrow PC<10:0>$
 $PCLATH<4:3> \rightarrow PC<12:11>$

Status Affected: None

Description: GOTO is an unconditional branch. The eleven-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction.

IORLW Inclusive OR Literal with W

Syntax: [*label*] IORLW k

Operands: $0 \leq k \leq 255$

Operation: $(W) .OR. k \rightarrow (W)$

Status Affected: Z

Description: The contents of the W register are OR'ed with the eight-bit literal 'k'. The result is placed in the W register.

INCF Increment f

Syntax: [*label*] INCF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) + 1 \rightarrow (\text{destination})$

Status Affected: Z

Description: The contents of register 'f' are incremented. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.

IORWF Inclusive OR W with f

Syntax: [*label*] IORWF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(W) .OR. (f) \rightarrow (\text{destination})$

Status Affected: Z

Description: Inclusive OR the W register with register 'f'. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.

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MOVF **Move f**

Syntax: [*label*] MOVF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: (f) → (destination)

Status Affected: Z

Description: The contents of register f are moved to a destination dependant upon the status of d. If d = 0, destination is W register. If d = 1, the destination is file register f itself. d = 1 is useful to test a file register, since status flag Z is affected.

NOP **No Operation**

Syntax: [*label*] NOP

Operands: None

Operation: No operation

Status Affected: None

Description: No operation.

MOVLW **Move Literal to W**

Syntax: [*label*] MOVLW k

Operands: $0 \leq k \leq 255$

Operation: $k \rightarrow (W)$

Status Affected: None

Description: The eight-bit literal 'k' is loaded into W register. The don't cares will assemble as 0's.

RETIE **Return from Interrupt**

Syntax: [*label*] RETFIE

Operands: None

Operation: TOS → PC,
 1 → GIE

Status Affected: None

MOVWF **Move W to f**

Syntax: [*label*] MOVWF f

Operands: $0 \leq f \leq 127$

Operation: (W) → (f)

Status Affected: None

Description: Move data from W register to register 'f'.

RETLW **Return with Literal in W**

Syntax: [*label*] RETLW k

Operands: $0 \leq k \leq 255$

Operation: $k \rightarrow (W)$;
 TOS → PC

Status Affected: None

Description: The W register is loaded with the eight-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a two-cycle instruction.

RLF Rotate Left f through Carry

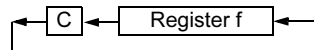
Syntax: [*label*] RLF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: See description below

Status Affected: C

Description: The contents of register 'f' are rotated one bit to the left through the Carry Flag. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is stored back in register 'f'.



SLEEP

Syntax: [*label*] SLEEP

Operands: None

Operation: 00h → WDT,
 0 → WDT prescaler,
 1 → \overline{TO} ,
 0 → \overline{PD}

Status Affected: \overline{TO} , \overline{PD}

Description: The power-down STATUS bit, \overline{PD} is cleared. Time-out STATUS bit, \overline{TO} is set. Watchdog Timer and its prescaler are cleared. The processor is put into SLEEP mode with the oscillator stopped.

RETURN Return from Subroutine

Syntax: [*label*] RETURN

Operands: None

Operation: TOS → PC

Status Affected: None

Description: Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.

SUBLW Subtract W from Literal

Syntax: [*label*] SUBLW k

Operands: $0 \leq k \leq 255$

Operation: $k - (W) \rightarrow (W)$

Status Affected: C, DC, Z

Description: The W register is subtracted (2's complement method) from the eight-bit literal 'k'. The result is placed in the W register.

RRF Rotate Right f through Carry

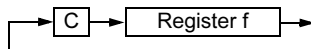
Syntax: [*label*] RRF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: See description below

Status Affected: C

Description: The contents of register 'f' are rotated one bit to the right through the Carry Flag. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.



SUBWF Subtract W from f

Syntax: [*label*] SUBWF f,d

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) - (W) \rightarrow (\text{destination})$

Status Affected: C, DC, Z

Description: Subtract (2's complement method) W register from register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

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SWAPF **Swap Nibbles in f**

Syntax: `[label] SWAPF f,d`

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f<3:0>) \rightarrow (\text{destination}<7:4>)$,
 $(f<7:4>) \rightarrow (\text{destination}<3:0>)$

Status Affected: None

Description: The upper and lower nibbles of register 'f' are exchanged. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed in register 'f'.

XORWF **Exclusive OR W with f**

Syntax: `[label] XORWF f,d`

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(W) .XOR. (f) \rightarrow (\text{destination})$

Status Affected: Z

Description: Exclusive OR the contents of the W register with register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

XORLW **Exclusive OR Literal with W**

Syntax: `[label] XORLW k`

Operands: $0 \leq k \leq 255$

Operation: $(W) .XOR. k \rightarrow (W)$

Status Affected: Z

Description: The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.

11.0 DEVELOPMENT SUPPORT

The PICmicro® microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB® IDE Software
- Assemblers/Compilers/Linkers
 - MPASM™ Assembler
 - MPLAB C17 and MPLAB C18 C Compilers
 - MPLINK™ Object Linker/
MPLIB™ Object Librarian
 - MPLAB C30 C Compiler
 - MPLAB ASM30 Assembler/Linker/Library
- Simulators
 - MPLAB SIM Software Simulator
 - MPLAB dsPIC30 Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - MPLAB ICE 4000 In-Circuit Emulator
- In-Circuit Debugger
 - MPLAB ICD 2
- Device Programmers
 - PRO MATE® II Universal Device Programmer
 - PICSTART® Plus Development Programmer
- Low Cost Demonstration Boards
 - PICDEM™ 1 Demonstration Board
 - PICDEM.net™ Demonstration Board
 - PICDEM 2 Plus Demonstration Board
 - PICDEM 3 Demonstration Board
 - PICDEM 17 Demonstration Board
 - PICDEM 18R Demonstration Board
 - PICDEM LIN Demonstration Board
 - PICDEM USB Demonstration Board
- Evaluation Kits
 - KEELOQ®
 - PICDEM MSC
 - microID®
 - CAN
 - PowerSmart®
 - Analog

11.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows® based application that contains:

- An interface to debugging tools
 - simulator
 - programmer (sold separately)
 - emulator (sold separately)
 - in-circuit debugger (sold separately)
- A full-featured editor with color coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High level source code debugging
- Mouse over variable inspection
- Extensive on-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
 - source files (assembly or C)
 - absolute listing file (mixed assembly and C)
 - machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost effective simulators, through low cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increasing flexibility and power.

11.2 MPASM Assembler

The MPASM assembler is a full-featured, universal macro assembler for all PICmicro MCUs.

The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contains source lines and generated machine code and COFF files for debugging.

The MPASM assembler features include:

- Integration into MPLAB IDE projects
- User defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

11.3 MPLAB C17 and MPLAB C18 C Compilers

The MPLAB C17 and MPLAB C18 Code Development Systems are complete ANSI C compilers for Microchip's PIC17CXXX and PIC18CXXX family of microcontrollers. These compilers provide powerful integration capabilities, superior code optimization and ease of use not found with other compilers.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

11.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK object linker combines relocatable objects created by the MPASM assembler and the MPLAB C17 and MPLAB C18 C compilers. It can link relocatable objects from pre-compiled libraries, using directives from a linker script.

The MPLIB object librarian manages the creation and modification of library files of pre-compiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/librarian features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

11.5 MPLAB C30 C Compiler

The MPLAB C30 C compiler is a full-featured, ANSI compliant, optimizing compiler that translates standard ANSI C programs into dsPIC30F assembly language source. The compiler also supports many command-line options and language extensions to take full advantage of the dsPIC30F device hardware capabilities, and afford fine control of the compiler code generator.

MPLAB C30 is distributed with a complete ANSI C standard library. All library functions have been validated and conform to the ANSI C library standard. The library includes functions for string manipulation, dynamic memory allocation, data conversion, time-keeping, and math functions (trigonometric, exponential and hyperbolic). The compiler provides symbolic information for high level source debugging with the MPLAB IDE.

11.6 MPLAB ASM30 Assembler, Linker, and Librarian

MPLAB ASM30 assembler produces relocatable machine code from symbolic assembly language for dsPIC30F devices. MPLAB C30 compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire dsPIC30F instruction set
- Support for fixed-point and floating-point data
- Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

11.7 MPLAB SIM Software Simulator

The MPLAB SIM software simulator allows code development in a PC hosted environment by simulating the PICmicro series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any pin. The execution can be performed in Single-Step, Execute Until Break, or Trace mode.

The MPLAB SIM simulator fully supports symbolic debugging using the MPLAB C17 and MPLAB C18 C Compilers, as well as the MPASM assembler. The software simulator offers the flexibility to develop and debug code outside of the laboratory environment, making it an excellent, economical software development tool.

11.8 MPLAB SIM30 Software Simulator

The MPLAB SIM30 software simulator allows code development in a PC hosted environment by simulating the dsPIC30F series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any of the pins.

The MPLAB SIM30 simulator fully supports symbolic debugging using the MPLAB C30 C Compiler and MPLAB ASM30 assembler. The simulator runs in either a Command Line mode for automated tasks, or from MPLAB IDE. This high speed simulator is designed to debug, analyze and optimize time intensive DSP routines.

11.9 MPLAB ICE 2000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 2000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PICmicro microcontrollers. Software control of the MPLAB ICE 2000 in-circuit emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PICmicro microcontrollers.

The MPLAB ICE 2000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft® Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

11.10 MPLAB ICE 4000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 4000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for high-end PICmicro microcontrollers. Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 4000 is a premium emulator system, providing the features of MPLAB ICE 2000, but with increased emulation memory and high speed performance for dsPIC30F and PIC18XXXX devices. Its advanced emulator features include complex triggering and timing, up to 2 Mb of emulation memory, and the ability to view variables in real-time.

The MPLAB ICE 4000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

11.11 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low cost, run-time development tool, connecting to the host PC via an RS-232 or high speed USB interface. This tool is based on the FLASH PICmicro MCUs and can be used to develop for these and other PICmicro microcontrollers. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the FLASH devices. This feature, along with Microchip's In-Circuit Serial Programming™ (ICSP™) protocol, offers cost effective in-circuit FLASH debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single-stepping and watching variables, CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real-time. MPLAB ICD 2 also serves as a development programmer for selected PICmicro devices.

11.12 PRO MATE II Universal Device Programmer

The PRO MATE II is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features an LCD display for instructions and error messages and a modular detachable socket assembly to support various package types. In Stand-alone mode, the PRO MATE II device programmer can read, verify, and program PICmicro devices without a PC connection. It can also set code protection in this mode.

11.13 PICSTART Plus Development Programmer

The PICSTART Plus development programmer is an easy-to-use, low cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus development programmer supports most PICmicro devices up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus development programmer is CE compliant.

11.14 PICDEM 1 PICmicro Demonstration Board

The PICDEM 1 demonstration board demonstrates the capabilities of the PIC16C5X (PIC16C54 to PIC16C58A), PIC16C61, PIC16C62X, PIC16C71, PIC16C8X, PIC17C42, PIC17C43 and PIC17C44. All necessary hardware and software is included to run basic demo programs. The sample microcontrollers provided with the PICDEM 1 demonstration board can be programmed with a PRO MATE II device programmer, or a PICSTART Plus development programmer. The PICDEM 1 demonstration board can be connected to the MPLAB ICE in-circuit emulator for testing. A prototype area extends the circuitry for additional application components. Features include an RS-232 interface, a potentiometer for simulated analog input, push button switches and eight LEDs.

11.15 PICDEM.net Internet/Ethernet Demonstration Board

The PICDEM.net demonstration board is an Internet/Ethernet demonstration board using the PIC18F452 microcontroller and TCP/IP firmware. The board supports any 40-pin DIP device that conforms to the standard pinout used by the PIC16F877 or PIC18C452. This kit features a user friendly TCP/IP stack, web server with HTML, a 24L256 Serial EEPROM for Xmodem download to web pages into Serial EEPROM, ICSP/MPLAB ICD 2 interface connector, an Ethernet interface, RS-232 interface, and a 16 x 2 LCD display. Also included is the book and CD-ROM *"TCP/IP Lean, Web Servers for Embedded Systems,"* by Jeremy Bentham

11.16 PICDEM 2 Plus Demonstration Board

The PICDEM 2 Plus demonstration board supports many 18-, 28-, and 40-pin microcontrollers, including PIC16F87X and PIC18FXX2 devices. All the necessary hardware and software is included to run the demonstration programs. The sample microcontrollers provided with the PICDEM 2 demonstration board can be programmed with a PRO MATE II device programmer, PICSTART Plus development programmer, or MPLAB ICD 2 with a Universal Programmer Adapter. The MPLAB ICD 2 and MPLAB ICE in-circuit emulators may also be used with the PICDEM 2 demonstration board to test firmware. A prototype area extends the circuitry for additional application components. Some of the features include an RS-232 interface, a 2 x 16 LCD display, a piezo speaker, an on-board temperature sensor, four LEDs, and sample PIC18F452 and PIC16F877 FLASH microcontrollers.

11.17 PICDEM 3 PIC16C92X Demonstration Board

The PICDEM 3 demonstration board supports the PIC16C923 and PIC16C924 in the PLCC package. All the necessary hardware and software is included to run the demonstration programs.

11.18 PICDEM 17 Demonstration Board

The PICDEM 17 demonstration board is an evaluation board that demonstrates the capabilities of several Microchip microcontrollers, including PIC17C752, PIC17C756A, PIC17C762 and PIC17C766. A programmed sample is included. The PRO MATE II device programmer, or the PICSTART Plus development programmer, can be used to reprogram the device for user tailored application development. The PICDEM 17 demonstration board supports program download and execution from external on-board FLASH memory. A generous prototype area is available for user hardware expansion.

11.19 PICDEM 18R PIC18C601/801 Demonstration Board

The PICDEM 18R demonstration board serves to assist development of the PIC18C601/801 family of Microchip microcontrollers. It provides hardware implementation of both 8-bit Multiplexed/De-multiplexed and 16-bit Memory modes. The board includes 2 Mb external FLASH memory and 128 Kb SRAM memory, as well as serial EEPROM, allowing access to the wide range of memory types supported by the PIC18C601/801.

11.20 PICDEM LIN PIC16C43X Demonstration Board

The powerful LIN hardware and software kit includes a series of boards and three PICmicro microcontrollers. The small footprint PIC16C432 and PIC16C433 are used as slaves in the LIN communication and feature on-board LIN transceivers. A PIC16F874 FLASH microcontroller serves as the master. All three microcontrollers are programmed with firmware to provide LIN bus communication.

11.21 PICKit™ 1 FLASH Starter Kit

A complete "development system in a box", the PICKit FLASH Starter Kit includes a convenient multi-section board for programming, evaluation, and development of 8/14-pin FLASH PIC® microcontrollers. Powered via USB, the board operates under a simple Windows GUI. The PICKit 1 Starter Kit includes the user's guide (on CD ROM), PICKit 1 tutorial software and code for various applications. Also included are MPLAB® IDE (Integrated Development Environment) software, software and hardware "Tips 'n Tricks for 8-pin FLASH PIC® Microcontrollers" Handbook and a USB Interface Cable. Supports all current 8/14-pin FLASH PIC microcontrollers, as well as many future planned devices.

11.22 PICDEM USB PIC16C7X5 Demonstration Board

The PICDEM USB Demonstration Board shows off the capabilities of the PIC16C745 and PIC16C765 USB microcontrollers. This board provides the basis for future USB products.

11.23 Evaluation and Programming Tools

In addition to the PICDEM series of circuits, Microchip has a line of evaluation kits and demonstration software for these products.

- KEELOQ evaluation and programming tools for Microchip's HCS Secure Data Products
- CAN developers kit for automotive network applications
- Analog design boards and filter design software
- PowerSmart battery charging evaluation/calibration kits
- IrDA® development kit
- microID development and rLab™ development software
- SEEVAL® designer kit for memory evaluation and endurance calculations
- PICDEM MSC demo boards for Switching mode power supply, high power IR driver, delta sigma ADC, and flow rate sensor

Check the Microchip web page and the latest Product Line Card for the complete list of demonstration and evaluation kits.

PIC12F629/675

NOTES:

12.0 ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings†

Ambient temperature under bias	-40 to +125°C
Storage temperature	-65°C to +150°C
Voltage on VDD with respect to VSS	-0.3 to +6.5V
Voltage on $\overline{\text{MCLR}}$ with respect to VSS	-0.3 to +13.5V
Voltage on all other pins with respect to VSS	-0.3V to (VDD + 0.3V)
Total power dissipation ⁽¹⁾	800 mW
Maximum current out of VSS pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, I _{IK} (V _I < 0 or V _I > VDD)	± 20 mA
Output clamp current, I _{OK} (V _O < 0 or V _O > VDD)	± 20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by all GPIO	125 mA
Maximum current sourced all GPIO	125 mA

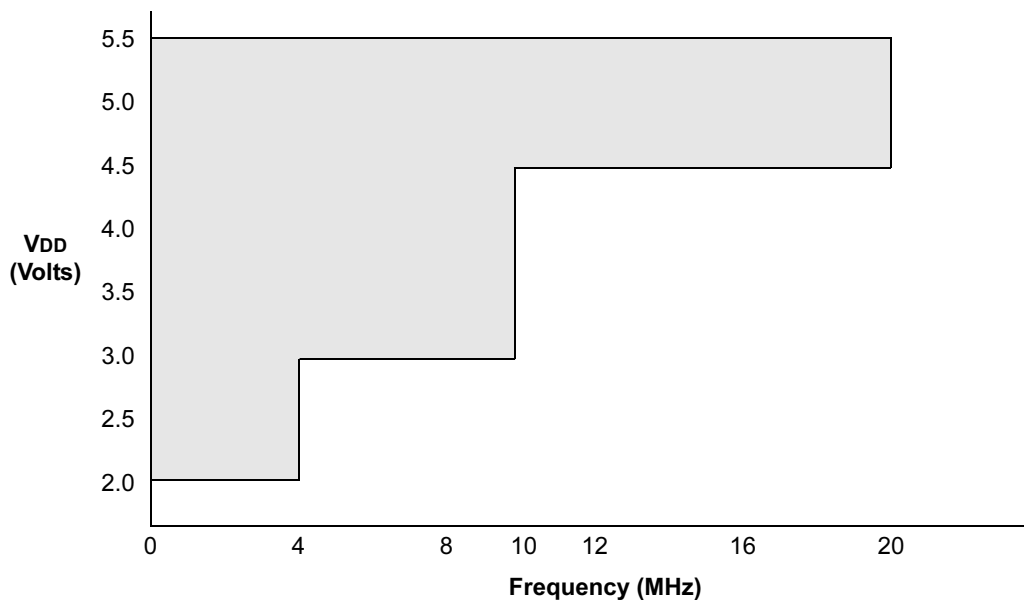
Note 1: Power dissipation is calculated as follows: $P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (V_{OI} \times I_{OL})$.

† **NOTICE:** Stresses above those listed under 'Absolute Maximum Ratings' may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Note: Voltage spikes below VSS at the $\overline{\text{MCLR}}$ pin, inducing currents greater than 80 mA, may cause latchup. Thus, a series resistor of 50-100 Ω should be used when applying a "low" level to the $\overline{\text{MCLR}}$ pin, rather than pulling this pin directly to VSS.

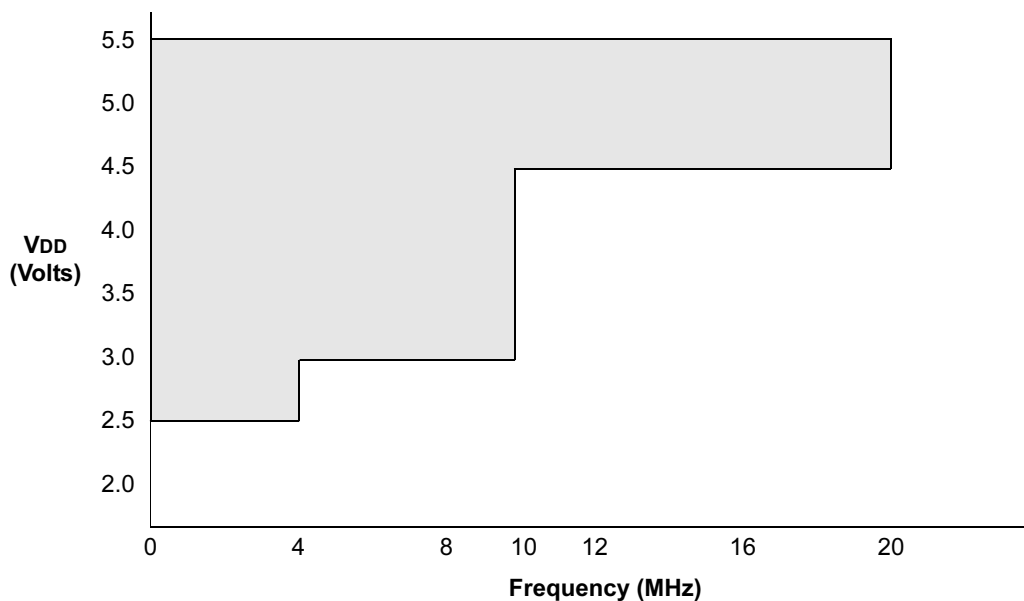
PIC12F629/675

FIGURE 12-1: PIC12F629/675 WITH A/D DISABLED VOLTAGE-FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$



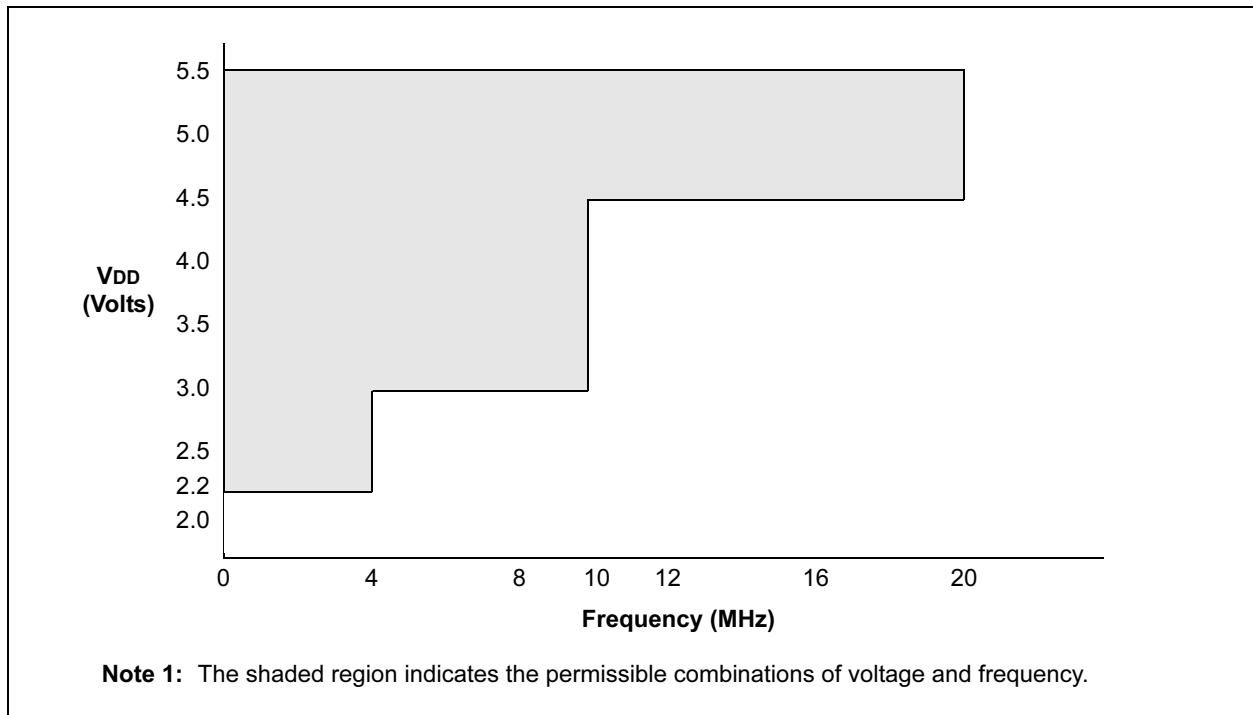
Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

FIGURE 12-2: PIC12F675 WITH A/D ENABLED VOLTAGE-FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

**FIGURE 12-3: PIC12F675 WITH A/D ENABLED VOLTAGE-FREQUENCY GRAPH,
 $0^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$**



PIC12F629/675

12.1 DC Characteristics: PIC12F629/675-I (Industrial), PIC12F629/675-E (Extended)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated)				
			Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended				
Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001 D001A D001B D001C D001D	VDD	Supply Voltage	2.0 2.2 2.5 3.0 4.5	— — — — —	5.5 5.5 5.5 5.5 5.5	V V V V V	FOSC ≤ 4 MHz: PIC12F629/675 with A/D off PIC12F675 with A/D on, 0°C to $+125^{\circ}\text{C}$ PIC12F675 with A/D on, -40°C to $+125^{\circ}\text{C}$ $4\text{ MHz} < \text{FOSC} \leq 10\text{ MHz}$
D002	VDR	RAM Data Retention Voltage⁽¹⁾	1.5*	—	—	V	Device in SLEEP mode
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	VSS	—	V	See section on Power-on Reset for details
D004	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05*	—	—	V/ms	See section on Power-on Reset for details
D005	VBOD		—	2.1	—	V	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

12.2 DC Characteristics: PIC12F629/675-I (Industrial)

		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial					
Param No.	Device Characteristics	Min	Typ†	Max	Units	Conditions	
						V _{DD}	Note
D010	Supply Current (I _{DD})	—	9	16	μA	2.0	Fosc = 32 kHz
		—	18	28	μA	3.0	LP Oscillator Mode
		—	35	54	μA	5.0	
D011		—	110	150	μA	2.0	Fosc = 1 MHz
		—	190	280	μA	3.0	XT Oscillator Mode
		—	330	450	μA	5.0	
D012		—	220	280	μA	2.0	Fosc = 4 MHz
		—	370	650	μA	3.0	XT Oscillator Mode
		—	0.6	1.4	mA	5.0	
D013		—	70	110	μA	2.0	Fosc = 1 MHz
		—	140	250	μA	3.0	EC Oscillator Mode
		—	260	390	μA	5.0	
D014		—	180	250	μA	2.0	Fosc = 4 MHz
		—	320	470	μA	3.0	EC Oscillator Mode
		—	580	850	μA	5.0	
D015		—	340	450	μA	2.0	Fosc = 4 MHz
		—	500	700	μA	3.0	INTOSC Mode
		—	0.8	1.1	mA	5.0	
D016		—	180	250	μA	2.0	Fosc = 4 MHz
		—	320	450	μA	3.0	EXTRC Mode
		—	580	800	μA	5.0	
D017		—	2.1	2.95	mA	4.5	Fosc = 20 MHz
		—	2.4	3.0	mA	5.0	HS Oscillator Mode

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all I_{DD} measurements in Active Operation mode are: OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to V_{DD}; MCLR = V_{DD}; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

PIC12F629/675

12.3 DC Characteristics: PIC12F629/675-I (Industrial)

		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial					
Param No.	Device Characteristics	Min	Typ†	Max	Units	Conditions	
						V _{DD}	Note
D020	Power-down Base Current (IPD)	—	0.99	700	nA	2.0	WDT, BOD, Comparators, V _{REF} , and T1OSC disabled
		—	1.2	770	nA	3.0	
		—	2.9	995	nA	5.0	
D021		—	0.3	1.5	μA	2.0	WDT Current ⁽¹⁾
		—	1.8	3.5	μA	3.0	
		—	8.4	17	μA	5.0	
D022		—	58	70	μA	3.0	BOD Current ⁽¹⁾
		—	109	130	μA	5.0	
D023		—	3.3	6.5	μA	2.0	Comparator Current ⁽¹⁾
		—	6.1	8.5	μA	3.0	
		—	11.5	16	μA	5.0	
D024		—	58	70	μA	2.0	CV _{REF} Current ⁽¹⁾
		—	85	100	μA	3.0	
		—	138	160	μA	5.0	
D025		—	4.0	6.5	μA	2.0	T1 Osc Current ⁽¹⁾
		—	4.6	7.0	μA	3.0	
		—	6.0	10.5	μA	5.0	
D026		—	1.2	775	nA	3.0	A/D Current ⁽¹⁾
		—	0.0022	1.0	μA	5.0	

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base I_{DD} or I_{PD} and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base I_{DD} or I_{PD} current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to V_{DD}.

12.4 DC Characteristics: PIC12F629/675-E (Extended)

		Standard Operating Conditions (unless otherwise stated)					
		Operating temperature -40°C ≤ TA ≤ +125°C for extended					
Param No.	Device Characteristics	Min	Typ†	Max	Units	Conditions	
						VDD	Note
D010E	Supply Current (IDD)	—	9	16	μA	2.0	Fosc = 32 kHz
		—	18	28	μA	3.0	LP Oscillator Mode
		—	35	54	μA	5.0	
D011E		—	110	150	μA	2.0	Fosc = 1 MHz
		—	190	280	μA	3.0	XT Oscillator Mode
		—	330	450	μA	5.0	
D012E		—	220	280	μA	2.0	Fosc = 4 MHz
		—	370	650	μA	3.0	XT Oscillator Mode
		—	0.6	1.4	mA	5.0	
D013E		—	70	110	μA	2.0	Fosc = 1 MHz
		—	140	250	μA	3.0	EC Oscillator Mode
		—	260	390	μA	5.0	
D014E		—	180	250	μA	2.0	Fosc = 4 MHz
		—	320	470	μA	3.0	EC Oscillator Mode
		—	580	850	μA	5.0	
D015E		—	340	450	μA	2.0	Fosc = 4 MHz
		—	500	780	μA	3.0	INTOSC Mode
		—	0.8	1.1	mA	5.0	
D016E		—	180	250	μA	2.0	Fosc = 4 MHz
		—	320	450	μA	3.0	EXTRC Mode
		—	580	800	μA	5.0	
D017E		—	2.1	2.95	mA	4.5	Fosc = 20 MHz
		—	2.4	3.0	mA	5.0	HS Oscillator Mode

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all I_{DD} measurements in Active Operation mode are: OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to V_{DD}; MCLR = V_{DD}; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

PIC12F629/675

12.5 DC Characteristics: PIC12F629/675-E (Extended)

		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended					
Param No.	Device Characteristics	Min	Typ†	Max	Units	Conditions	
						VDD	Note
D020E	Power-down Base Current (IPD)	—	0.00099	3.5	μA	2.0	WDT, BOD, Comparators, VREF, and T1OSC disabled
		—	0.0012	4.0	μA	3.0	
		—	0.0029	8.0	μA	5.0	
D021E		—	0.3	6.0	μA	2.0	WDT Current ⁽¹⁾
		—	1.8	9.0	μA	3.0	
		—	8.4	20	μA	5.0	
D022E		—	58	70	μA	3.0	BOD Current ⁽¹⁾
		—	109	130	μA	5.0	
D023E		—	3.3	10	μA	2.0	Comparator Current ⁽¹⁾
		—	6.1	13	μA	3.0	
		—	11.5	24	μA	5.0	
D024E		—	58	70	μA	2.0	CVREF Current ⁽¹⁾
		—	85	100	μA	3.0	
		—	138	165	μA	5.0	
D025E		—	4.0	10	μA	2.0	T1 Osc Current ⁽¹⁾
		—	4.6	12	μA	3.0	
		—	6.0	20	μA	5.0	
D026E		—	0.0012	6.0	μA	3.0	A/D Current ⁽¹⁾
		—	0.0022	8.5	μA	5.0	

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base IDD or IPD current from this limit. Max values should be used when calculating total current consumption.

2: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD.

12.6 DC Characteristics: PIC12F629/675-I (Industrial), PIC12F629/675-E (Extended)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D030 D030A D031 D032 D033 D033A	V _{IL}	Input Low Voltage I/O ports with TTL buffer with Schmitt Trigger buffer MCLR, OSC1 (RC mode) OSC1 (XT and LP modes) OSC1 (HS mode)	V _{SS} V _{SS} V _{SS} V _{SS} V _{SS} V _{SS}	— — — — — —	0.8 0.15 V _{DD} 0.2 V _{DD} 0.2 V _{DD} 0.3 0.3 V _{DD}	V V V V V V	4.5V ≤ V _{DD} ≤ 5.5V Otherwise Entire range (Note 1) (Note 1)
D040 D040A D041 D042 D043 D043A D043B	V _{IH}	Input High Voltage I/O ports with TTL buffer with Schmitt Trigger buffer MCLR OSC1 (XT and LP modes) OSC1 (HS mode) OSC1 (RC mode)	2.0 (0.25 V _{DD} +0.8) 0.8 V _{DD} 0.8 V _{DD} 1.6 0.7 V _{DD} 0.9 V _{DD}	— — — — — — —	V _{DD} V _{DD} V _{DD} V _{DD} V _{DD} V _{DD} V _{DD}	V V V V V V V	4.5V ≤ V _{DD} ≤ 5.5V otherwise entire range (Note 1) (Note 1)
D070	I _{PUR}	GPIO Weak Pull-up Current	50*	250	400*	μA	V _{DD} = 5.0V, V _{PIN} = V _{SS}
D060 D060A D060B D061 D063	I _{IL}	Input Leakage Current⁽³⁾ I/O ports Analog inputs V _{REF} MCLR ⁽²⁾ OSC1	— — — — —	± 0.1 ± 0.1 ± 0.1 ± 0.1 ± 0.1	± 1 ± 1 ± 1 ± 5 ± 5	μA μA μA μA μA	V _{SS} ≤ V _{PIN} ≤ V _{DD} , Pin at hi-impedance V _{SS} ≤ V _{PIN} ≤ V _{DD} V _{SS} ≤ V _{PIN} ≤ V _{DD} V _{SS} ≤ V _{PIN} ≤ V _{DD} V _{SS} ≤ V _{PIN} ≤ V _{DD} , XT, HS and LP osc configuration
D080 D083	V _{OL}	Output Low Voltage I/O ports OSC2/CLKOUT (RC mode)	— —	— —	0.6 0.6	V V	I _{OL} = 8.5 mA, V _{DD} = 4.5V (Ind.) I _{OL} = 1.6 mA, V _{DD} = 4.5V (Ind.) I _{OL} = 1.2 mA, V _{DD} = 4.5V (Ext.)
D090 D092	V _{OH}	Output High Voltage I/O ports OSC2/CLKOUT (RC mode)	V _{DD} - 0.7 V _{DD} - 0.7	— —	— —	V V	I _{OH} = -3.0 mA, V _{DD} = 4.5V (Ind.) I _{OH} = -1.3 mA, V _{DD} = 4.5V (Ind.) I _{OH} = -1.0 mA, V _{DD} = 4.5V (Ext.)

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

- Note 1:** In RC oscillator configuration, the OSC1/CLKIN pin is a Schmitt Trigger input. It is not recommended to use an external clock in RC mode.
- 2:** The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.
- 3:** Negative current is defined as current sourced by the pin.

PIC12F629/675

12.7 DC Characteristics: PIC12F629/675-I (Industrial), PIC12F629/675-E (Extended) (Cont.)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended				
Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D100	Cosc2	Capacitive Loading Specs on Output Pins OSC2 pin	—	—	15*	pF	In XT, HS and LP modes when external clock is used to drive OSC1
D101	CIO	All I/O pins	—	—	50*	pF	
D120	Ed	Data EEPROM Memory Byte Endurance	100K	1M	—	E/W	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ $+85^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ Using EECON to read/write V_{MIN} = Minimum operating voltage Provided no other specifications are violated $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$
D120A	Ed	Byte Endurance	10K	100K	—	E/W	
D121	VDRW	VDD for Read/Write	V_{MIN}	—	5.5	V	
D122	TDEW	Erase/Write cycle time	—	5	6	ms	
D123	TRETD	Characteristic Retention	40	—	—	Year	
D124	TREF	Number of Total Erase/Write Cycles before Refresh ⁽¹⁾	1M	10M	—	E/W	
D130	EP	Program FLASH Memory Cell Endurance	10K	100K	—	E/W	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ $+85^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ V_{MIN} = Minimum operating voltage Provided no other specifications are violated
D130A	ED	Cell Endurance	1K	10K	—	E/W	
D131	VPR	VDD for Read	V_{MIN}	—	5.5	V	
D132	VPEW	VDD for Erase/Write	4.5	—	5.5	V	
D133	TPEW	Erase/Write cycle time	—	2	2.5	ms	
D134	TRETD	Characteristic Retention	40	—	—	Year	

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: See Section 8.5.1 for additional information.

12.8 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created with one of the following formats:

- 1. TppS2ppS
- 2. TppS

T			
F	Frequency	T	Time

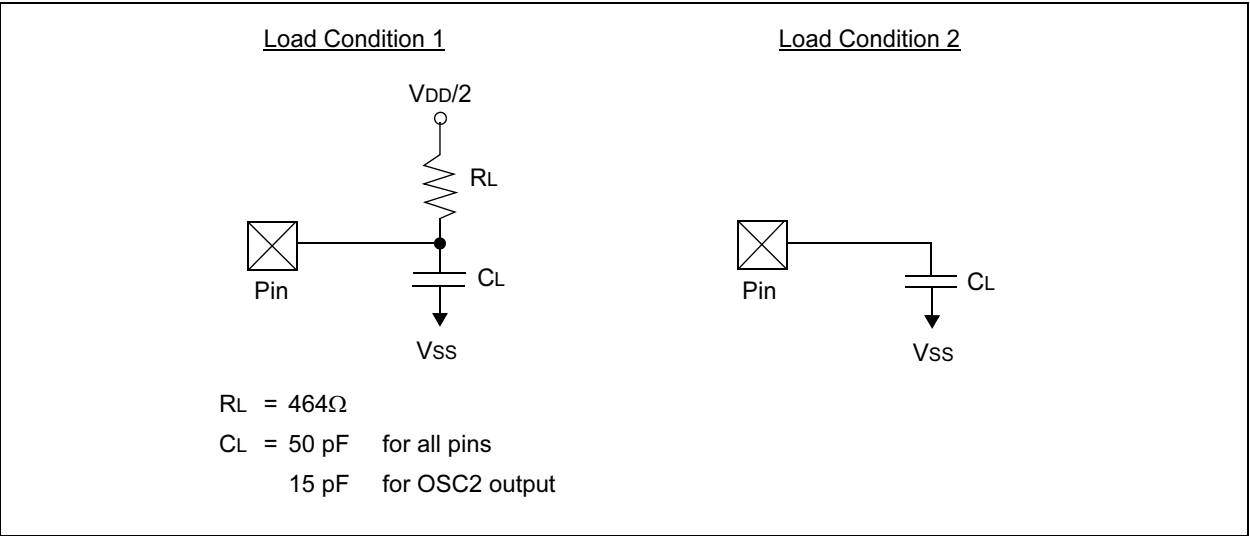
Lowercase letters (pp) and their meanings:

pp			
cc	CCP1	osc	OSC1
ck	CLKOUT	rd	\overline{RD}
cs	\overline{CS}	rw	\overline{RD} or \overline{WR}
di	SDI	sc	SCK
do	SDO	ss	\overline{SS}
dt	Data in	t0	T0CKI
io	I/O port	t1	T1CKI
mc	\overline{MCLR}	wr	\overline{WR}

Uppercase letters and their meanings:

S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (Hi-impedance)	V	Valid
L	Low	Z	Hi-impedance

FIGURE 12-4: LOAD CONDITIONS



PIC12F629/675

12.9 AC CHARACTERISTICS: PIC12F629/675 (INDUSTRIAL, EXTENDED)

FIGURE 12-5: EXTERNAL CLOCK TIMING

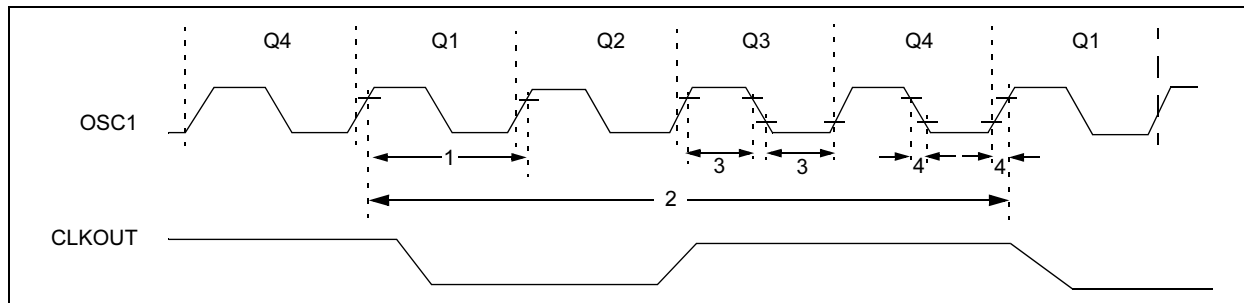


TABLE 12-1: EXTERNAL CLOCK TIMING REQUIREMENTS

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
	Fosc	External CLKIN Frequency ⁽¹⁾	DC	—	37	kHz	LP Osc mode
			DC	—	4	MHz	XT mode
			DC	—	20	MHz	HS mode
			DC	—	20	MHz	EC mode
		Oscillator Frequency ⁽¹⁾	5	—	37	kHz	LP Osc mode
			—	4	—	MHz	INTOSC mode
			DC	—	4	MHz	RC Osc mode
			0.1	—	4	MHz	XT Osc mode
			1	—	20	MHz	HS Osc mode
1	Tosc	External CLKIN Period ⁽¹⁾	27	—	∞	μs	LP Osc mode
			50	—	∞	ns	HS Osc mode
			50	—	∞	ns	EC Osc mode
			250	—	∞	ns	XT Osc mode
		Oscillator Period ⁽¹⁾	27	—	200	μs	LP Osc mode
			—	250	—	ns	INTOSC mode
			250	—	—	ns	RC Osc mode
			250	—	10,000	ns	XT Osc mode
			50	—	1,000	ns	HS Osc mode
2	Tcy	Instruction Cycle Time ⁽¹⁾	200	Tcy	DC	ns	Tcy = 4/Fosc
3	TosL, TosH	External CLKIN (OSC1) High	2*	—	—	μs	LP oscillator, TOSC L/H duty cycle
		External CLKIN Low	20*	—	—	ns	HS oscillator, TOSC L/H duty cycle
			100 *	—	—	ns	XT oscillator, TOSC L/H duty cycle
4	TosR, TosF	External CLKIN Rise	—	—	50*	ns	LP oscillator
		External CLKIN Fall	—	—	25*	ns	XT oscillator
			—	—	15*	ns	HS oscillator

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (Tcy) equals four times the input oscillator time-base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at 'min' values with an external clock applied to OSC1 pin. When an external clock input is used, the 'max' cycle time limit is 'DC' (no clock) for all devices.

TABLE 12-2: PRECISION INTERNAL OSCILLATOR PARAMETERS

Param No.	Sym	Characteristic	Freq Tolerance	Min	Typ†	Max	Units	Conditions
F10	FOSC	Internal Calibrated INTOSC Frequency	±1	3.96	4.00	4.04	MHz	VDD = 3.5V, 25°C
			±2	3.92	4.00	4.08	MHz	2.5V ≤ VDD ≤ 5.5V 0°C ≤ TA ≤ +85°C
			±5	3.80	4.00	4.20	MHz	2.0V ≤ VDD ≤ 5.5V -40°C ≤ TA ≤ +85°C (IND) -40°C ≤ TA ≤ +125°C (EXT)
F14	TOSC ST	Oscillator Wake-up from SLEEP start-up time*	—	—	6	8	μs	VDD = 2.0V, -40°C to +85°C
			—	—	4	6	μs	VDD = 3.0V, -40°C to +85°C
			—	—	3	5	μs	VDD = 5.0V, -40°C to +85°C

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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FIGURE 12-6: CLKOUT AND I/O TIMING

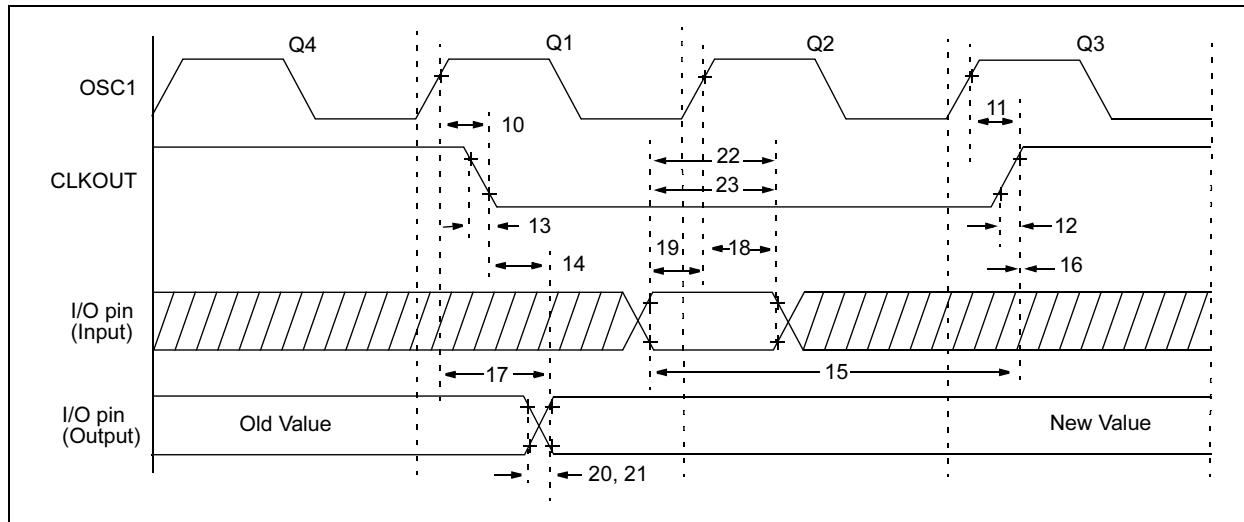


TABLE 12-3: CLKOUT AND I/O TIMING REQUIREMENTS

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
10	TosH2ckL	OSC1↑ to CLK-OUT↓	—	75	200	ns	(Note 1)
11	TosH2ckH	OSC1↑ to CLK-OUT↑	—	75	200	ns	(Note 1)
12	TckR	CLKOUT rise time	—	35	100	ns	(Note 1)
13	TckF	CLKOUT fall time	—	35	100	ns	(Note 1)
14	TckL2ioV	CLKOUT↓ to Port out valid	—	—	20	ns	(Note 1)
15	TioV2ckH	Port in valid before CLKOUT↑	Tosc + 200 ns	—	—	ns	(Note 1)
16	TckH2ioI	Port in hold after CLKOUT↑	0	—	—	ns	(Note 1)
17	TosH2ioV	OSC1↑ (Q1 cycle) to Port out valid	—	50	150 *	ns	
			—	—	300	ns	
18	TosH2ioI	OSC1↑ (Q2 cycle) to Port input invalid (I/O in hold time)	100	—	—	ns	
19	TioV2osH	Port input valid to OSC1↑ (I/O in setup time)	0	—	—	ns	
20	TioR	Port output rise time	—	10	40	ns	
21	TioF	Port output fall time	—	10	40	ns	
22	Tinp	INT pin high or low time	25	—	—	ns	
23	Trbp	GPIO change INT high or low time	Tcy	—	—	ns	

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated.

Note 1: Measurements are taken in RC mode where CLKOUT output is 4xTosc.

FIGURE 12-7: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

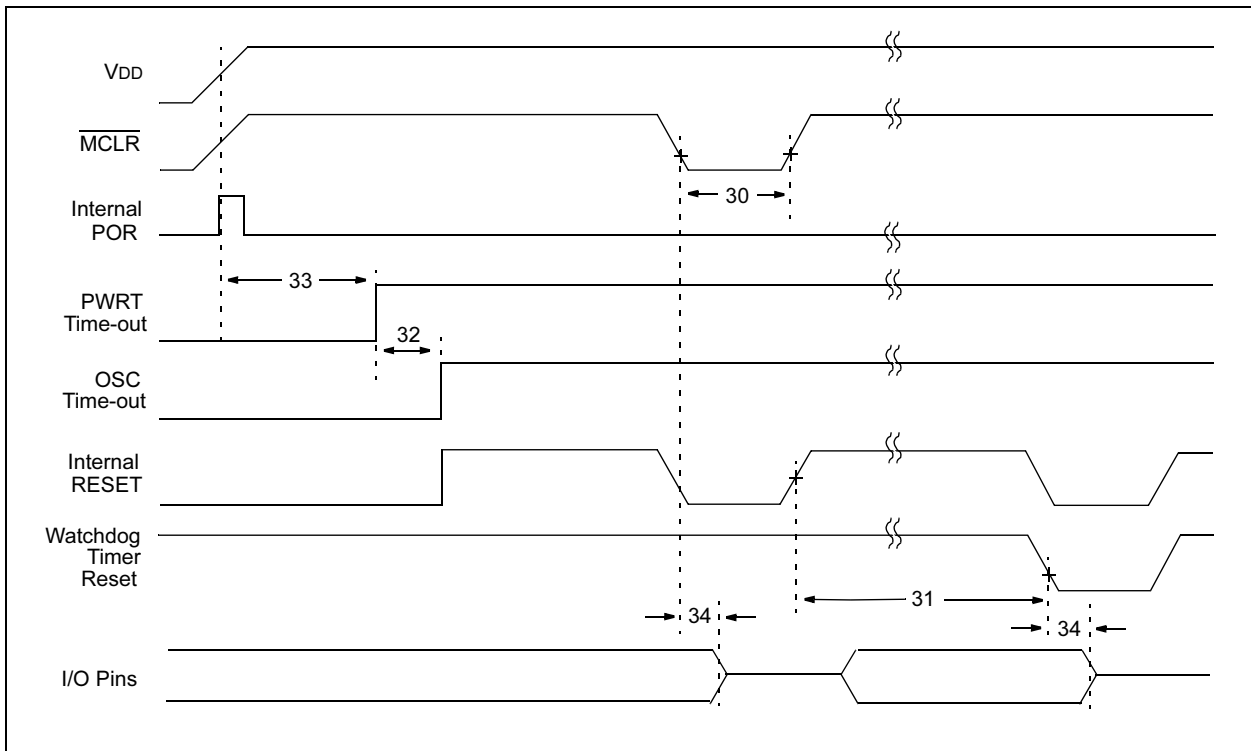
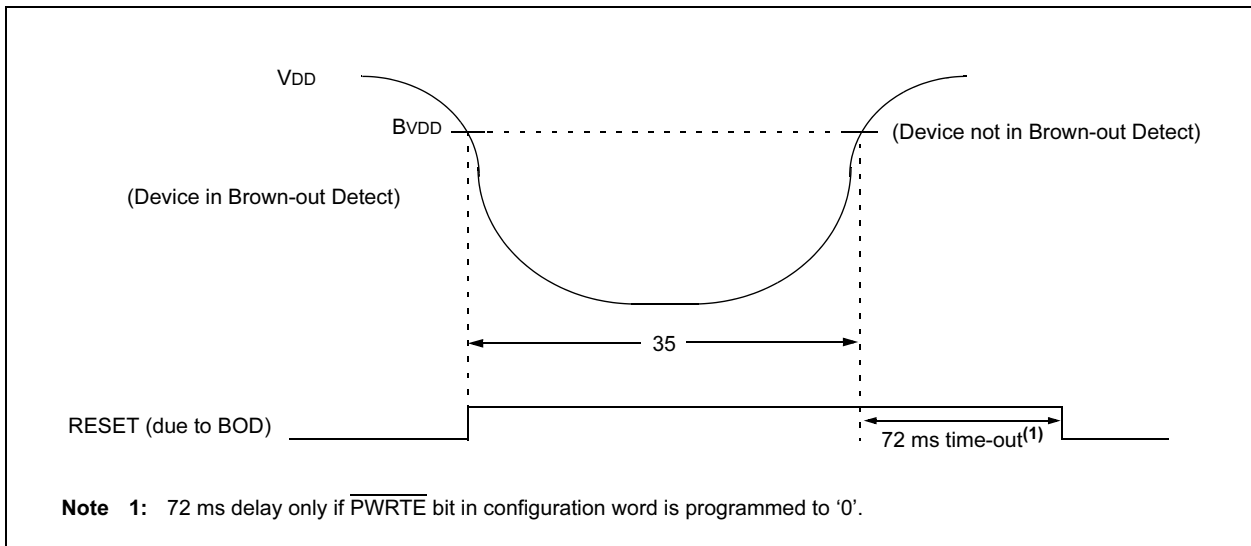


FIGURE 12-8: BROWN-OUT DETECT TIMING AND CHARACTERISTICS



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TABLE 12-4: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER, AND BROWN-OUT DETECT REQUIREMENTS

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
30	TMCL	MCLR Pulse Width (low)	2 TBD	— TBD	— TBD	μs ms	VDD = 5V, -40°C to +85°C Extended temperature
31	TWDT	Watchdog Timer Time-out Period (No Prescaler)	10 10	17 17	25 30	ms ms	VDD = 5V, -40°C to +85°C Extended temperature
32	TOST	Oscillation Start-up Timer Period	—	1024Tosc	—	—	Tosc = OSC1 period
33*	TPWRT	Power-up Timer Period	28* TBD	72 TBD	132* TBD	ms ms	VDD = 5V, -40°C to +85°C Extended Temperature
34	TIOZ	I/O Hi-impedance from MCLR Low or Watchdog Timer Reset	—	—	2.0	μs	
	BVDD	Brown-out Detect Voltage	2.025	—	2.175	V	
		Brown-out Hysteresis	TBD	—	—	—	
35	TBOD	Brown-out Detect Pulse Width	100*	—	—	μs	VDD ≤ BVDD (D005)

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 12-9: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

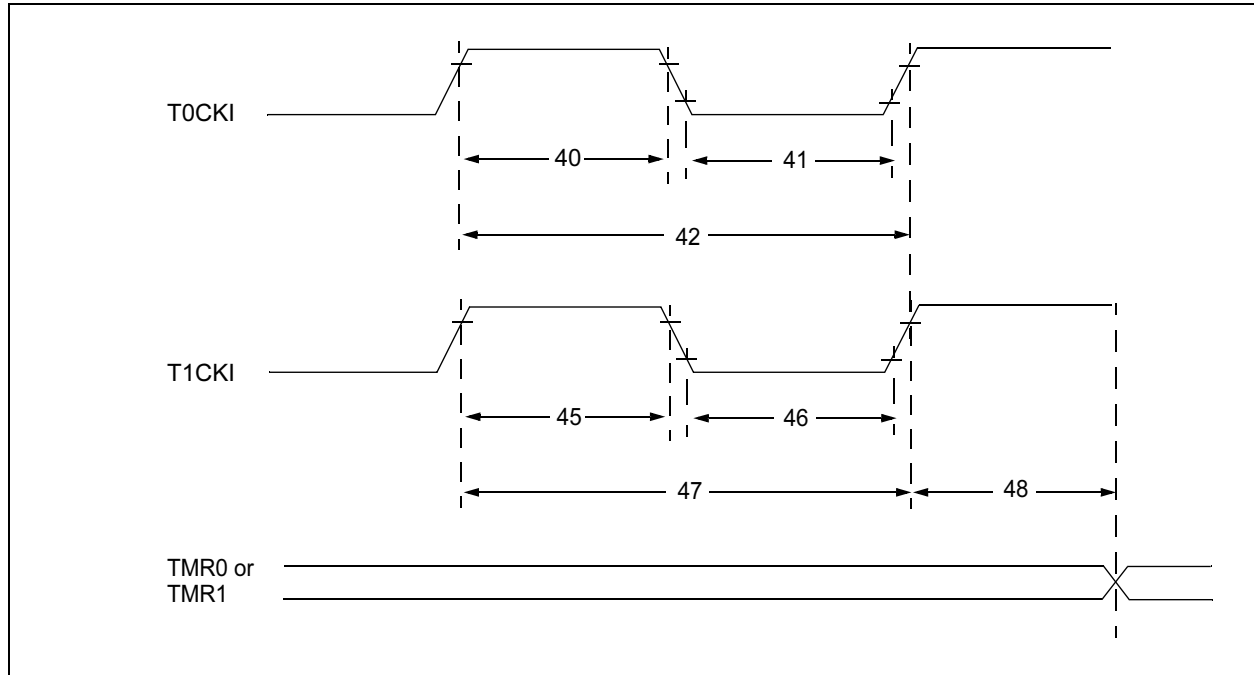


TABLE 12-5: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
40*	Tt0H	T0CKI High Pulse Width	No Prescaler	$0.5 T_{CY} + 20$	—	—	ns	
			With Prescaler	10	—	—	ns	
41*	Tt0L	T0CKI Low Pulse Width	No Prescaler	$0.5 T_{CY} + 20$	—	—	ns	
			With Prescaler	10	—	—	ns	
42*	Tt0P	T0CKI Period		Greater of: 20 or $\frac{T_{CY} + 40}{N}$	—	—	ns	N = prescale value (2, 4, ..., 256)
45*	Tt1H	T1CKI High Time	Synchronous, No Prescaler	$0.5 T_{CY} + 20$	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
46*	Tt1L	T1CKI Low Time	Synchronous, No Prescaler	$0.5 T_{CY} + 20$	—	—	ns	
			Synchronous, with Prescaler	15	—	—	ns	
			Asynchronous	30	—	—	ns	
47*	Tt1P	T1CKI Input Period	Synchronous	Greater of: 30 or $\frac{T_{CY} + 40}{N}$	—	—	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous	60	—	—	ns	
	Ft1	Timer1 oscillator input frequency range (oscillator enabled by setting bit T1OSCEN)		DC	—	200*	kHz	
48	TCKEZtmr1	Delay from external clock edge to timer increment		$2 T_{osc}^*$	—	7 T_{osc}^*	—	

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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TABLE 12-6: COMPARATOR SPECIFICATIONS

Comparator Specifications		Standard Operating Conditions -40°C to +125°C (unless otherwise stated)				
Sym	Characteristics	Min	Typ	Max	Units	Comments
VOS	Input Offset Voltage	—	± 5.0	± 10	mV	
VCM	Input Common Mode Voltage	0	—	VDD - 1.5	V	
CMRR	Common Mode Rejection Ratio	+55*	—	—	db	
TRT	Response Time ⁽¹⁾	—	150	400*	ns	
TMC2coV	Comparator Mode Change to Output Valid	—	—	10*	µs	

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (VDD - 1.5)/2 while the other input transitions from VSS to VDD - 1.5V.

TABLE 12-7: COMPARATOR VOLTAGE REFERENCE SPECIFICATIONS

Voltage Reference Specifications		Standard Operating Conditions -40°C to +125°C (unless otherwise stated)				
Sym	Characteristics	Min	Typ	Max	Units	Comments
	Resolution	—	VDD/24*	—	LSb	Low Range (VRR = 1)
		—	VDD/32	—	LSb	High Range (VRR = 0)
	Absolute Accuracy	—	—	± 1/2	LSb	Low Range (VRR = 1)
		—	—	± 1/2*	LSb	High Range (VRR = 0)
	Unit Resistor Value (R)	—	2k*	—	Ω	
	Settling Time ⁽¹⁾	—	—	10*	µs	

* These parameters are characterized but not tested.

Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from 0000 to 1111.

TABLE 12-8: PIC12F675 A/D CONVERTER CHARACTERISTICS:

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
A01	NR	Resolution	—	—	10 bits	bit	
A02	EABS	Total Absolute Error*	—	—	±1	LSb	VREF = 5.0V
A03	EIL	Integral Error	—	—	±1	LSb	VREF = 5.0V
A04	EDL	Differential Error	—	—	±1	LSb	No missing codes to 10 bits VREF = 5.0V
A05	EFS	Full Scale Range	2.2*	—	5.5*	V	
A06	EOFF	Offset Error	—	—	±1	LSb	VREF = 5.0V
A07	EGN	Gain Error	—	—	±1	LSb	VREF = 5.0V
A10	—	Monotonicity	—	guaranteed ⁽³⁾	—	—	VSS ≤ VAIN ≤ VREF+
A20 A20A	VREF	Reference Voltage	2.0 2.5	—	— VDD + 0.3	V	Absolute minimum to ensure 10-bit accuracy
A21	VREF	Reference V High (VDD or VREF)	VSS	—	VDD	V	
A25	VAIN	Analog Input Voltage	VSS	—	VREF	V	
A30	ZAIN	Recommended Impedance of Analog Voltage Source	—	—	10	kΩ	
A50	IREF	VREF Input Current ⁽²⁾	10 —	— —	1000 10	μA μA	During VAIN acquisition. Based on differential of VHOLD to VAIN. During A/D conversion cycle.

* These parameters are characterized but not tested.

† Data in 'Typ' column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from External VREF or VDD pin, whichever is selected as reference input.

3: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

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FIGURE 12-10: PIC12F675 A/D CONVERSION TIMING (NORMAL MODE)

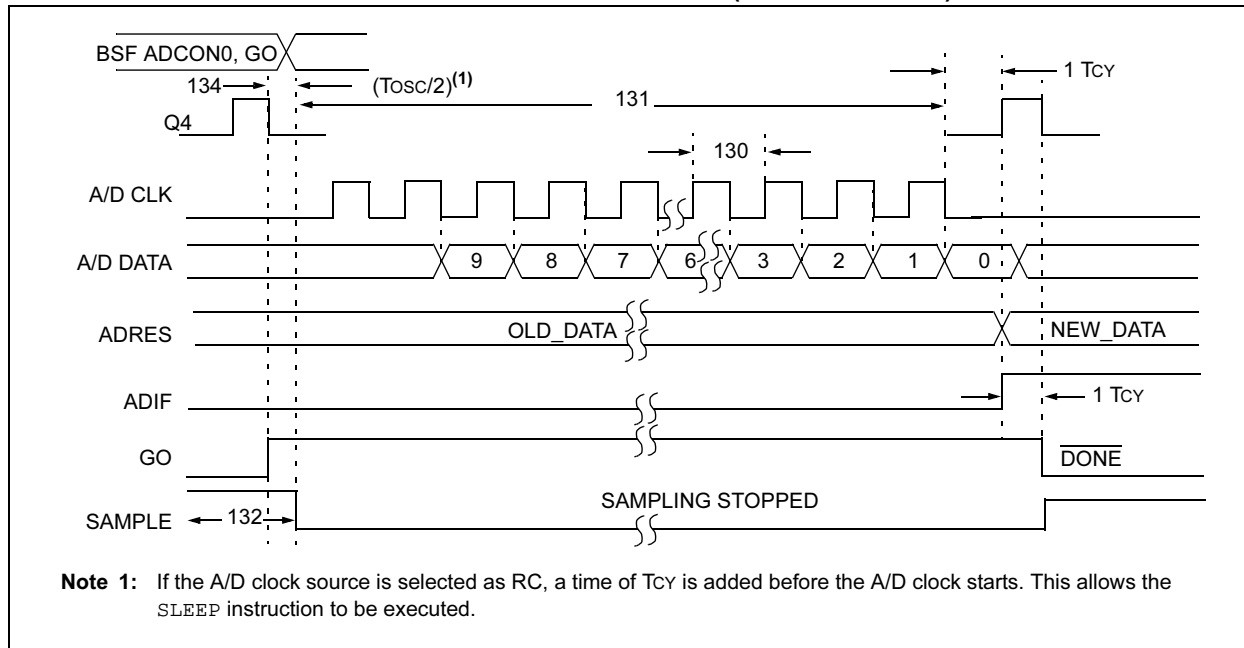


TABLE 12-9: PIC12F675 A/D CONVERSION REQUIREMENTS

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
130	TAD	A/D Clock Period	1.6	—	—	μs	TOSC based, $V_{REF} \geq 3.0\text{V}$
130	TAD	A/D Internal RC Oscillator Period	3.0*	—	—	μs	TOSC based, V_{REF} full range
			3.0*	6.0	9.0*	μs	ADCS<1:0> = 11 (RC mode)
			2.0*	4.0	6.0*	μs	At $V_{DD} = 2.5\text{V}$
						μs	At $V_{DD} = 5.0\text{V}$
131	TCNV	Conversion Time (not including Acquisition Time) ⁽¹⁾	—	11	—	TAD	Set GO bit to new data in A/D result register
132	TACQ	Acquisition Time	(Note 2)	11.5	—	μs	The minimum time is the amplifier settling time. This may be used if the “new” input voltage has not changed by more than 1 LSb (i.e., 4.1 mV @ 4.096V) from the last sampled voltage (as stored on CHOLD).
			5*	—	—	μs	
134	TGO	Q4 to A/D Clock Start	—	Tosc/2	—	—	If the A/D clock source is selected as RC, a time of T_{CY} is added before the A/D clock starts. This allows the SLEEP instruction to be executed.

* These parameters are characterized but not tested.

† Data in ‘Typ’ column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRES register may be read on the following T_{CY} cycle.

Note 2: See Section 7.1 for minimum conditions.

FIGURE 12-11: PIC12F675 A/D CONVERSION TIMING (SLEEP MODE)

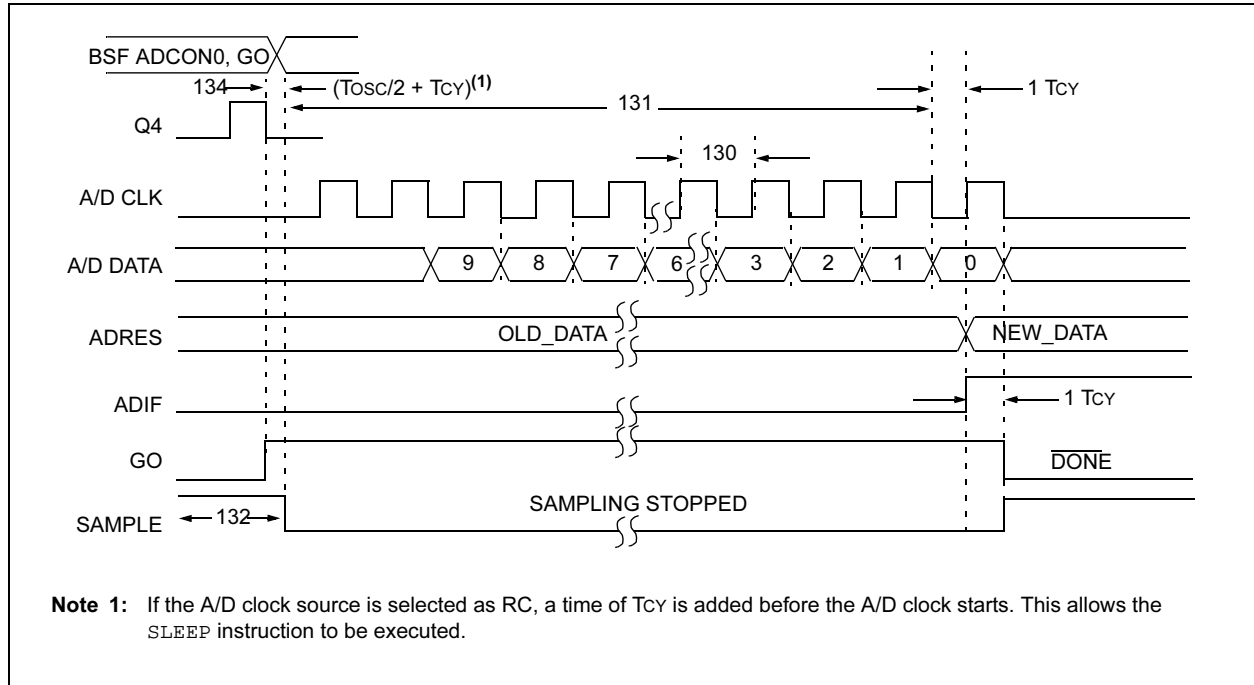


TABLE 12-10: PIC12F675 A/D CONVERSION REQUIREMENTS (SLEEP MODE)

Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
130	TAD	A/D Clock Period	1.6	—	—	μs	$V_{REF} \geq 3.0\text{V}$
130	TAD	A/D Internal RC Oscillator Period	3.0*	—	—	μs	V_{REF} full range
			3.0*	6.0	9.0*	μs	$ADCS\langle 1:0 \rangle = 11$ (RC mode)
			2.0*	4.0	6.0*	μs	At $V_{DD} = 2.5\text{V}$
							At $V_{DD} = 5.0\text{V}$
131	TCNV	Conversion Time (not including Acquisition Time) ⁽¹⁾	—	11	—	TAD	
132	TACQ	Acquisition Time	(Note 2)	11.5	—	μs	The minimum time is the amplifier settling time. This may be used if the “new” input voltage has not changed by more than 1 LSb (i.e., 4.1 mV @ 4.096V) from the last sampled voltage (as stored on CHOLD).
			5*	—	—	μs	
134	TGO	Q4 to A/D Clock Start	—	$T_{OSC}/2 + T_{CY}$	—	—	If the A/D clock source is selected as RC, a time of T_{CY} is added before the A/D clock starts. This allows the <code>SLEEP</code> instruction to be executed.

* These parameters are characterized but not tested.

† Data in ‘Typ’ column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: ADRES register may be read on the following T_{CY} cycle.

Note 2: See Section 7.1 for minimum conditions.

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NOTES:

13.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

The graphs and tables provided in this section are for **design guidance** and are **not tested**.

In some graphs or tables, the data presented are **outside specified operating range** (i.e., outside specified V_{DD} range). This is for **information only** and devices are ensured to operate properly only within the specified range.

The data presented in this section is a **statistical summary** of data collected on units from different lots over a period of time and matrix samples. 'Typical' represents the mean of the distribution at 25°C. 'Max' or 'min' represents (mean + 3σ) or (mean - 3σ) respectively, where σ is standard deviation, over the whole temperature range.

FIGURE 13-1: TYPICAL I_{PD} vs. V_{DD} OVER TEMP (-40°C TO +25°C)

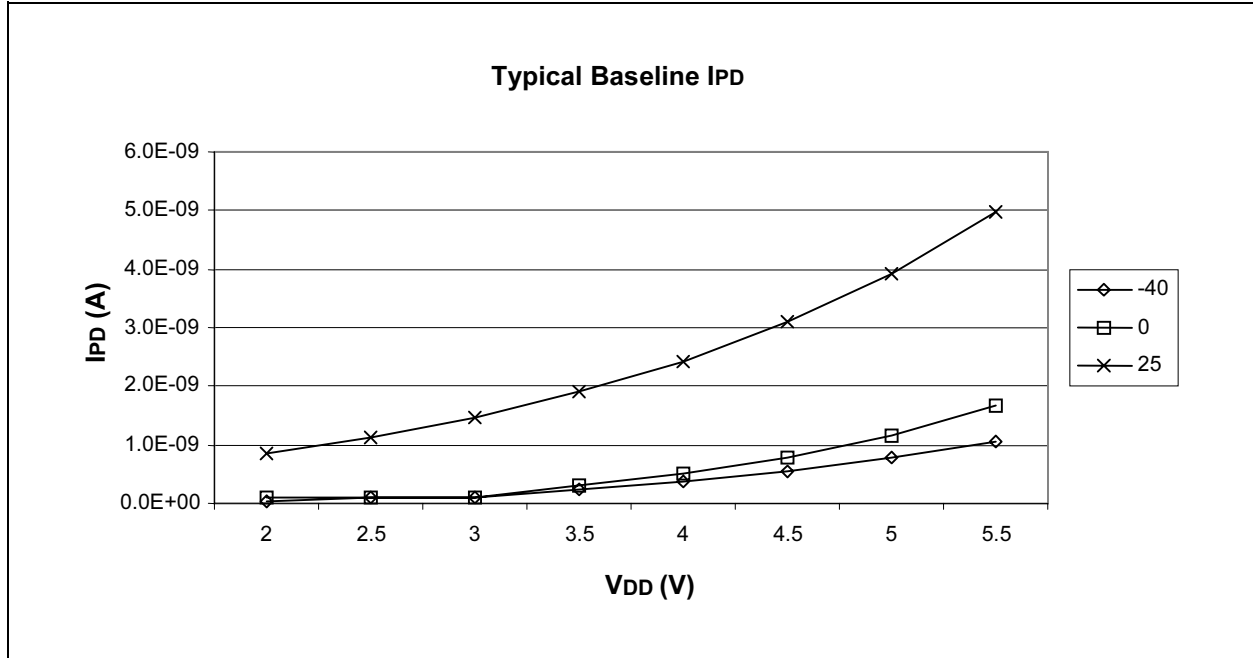
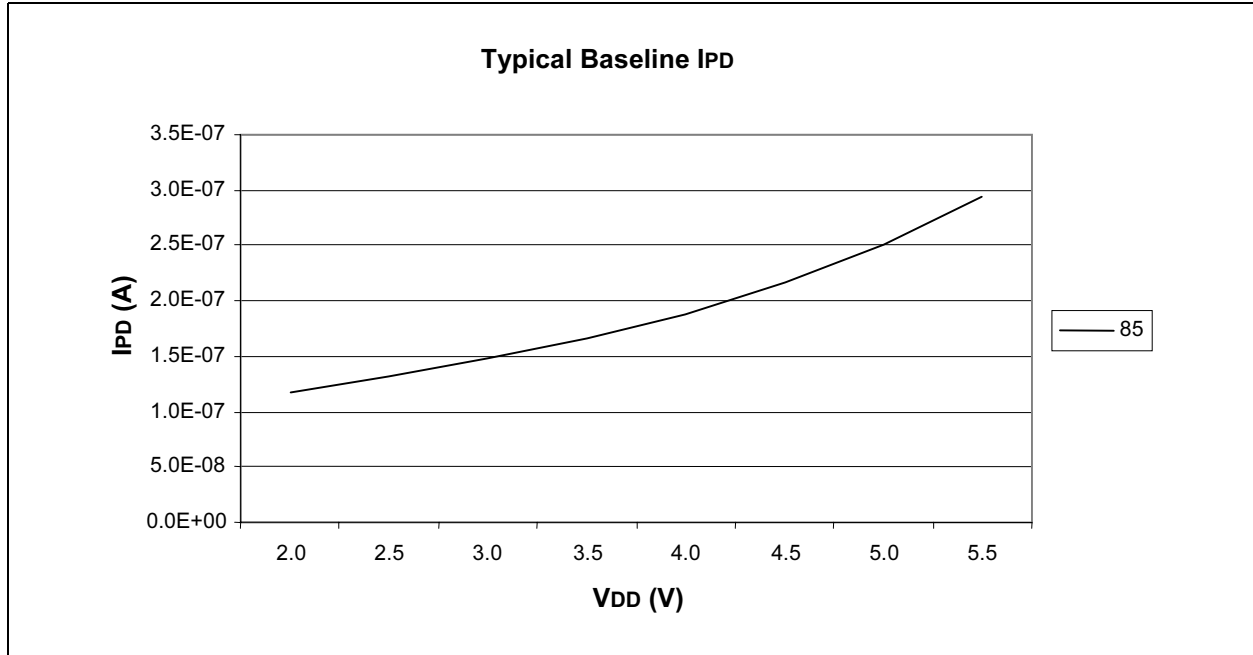


FIGURE 13-2: TYPICAL I_{PD} vs. V_{DD} OVER TEMP (+85°C)



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FIGURE 13-3: TYPICAL IPD vs. VDD OVER TEMP (+125°C)

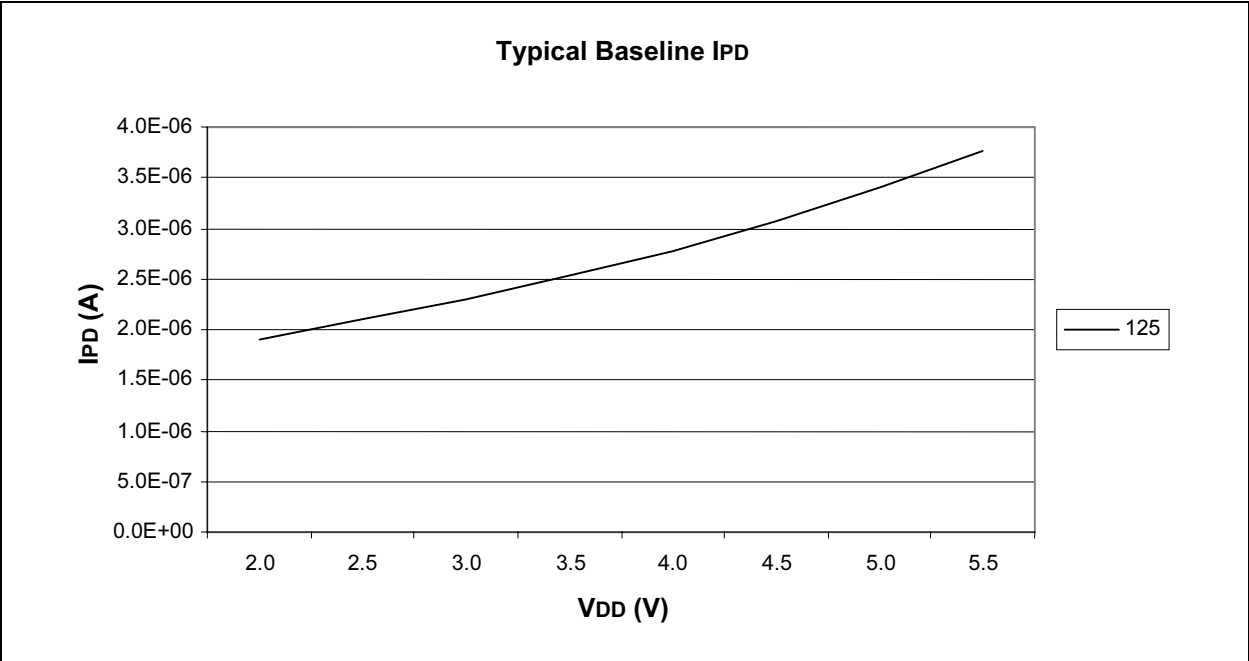


FIGURE 13-4: MAXIMUM IPD vs. VDD OVER TEMP (-40°C TO +25°C)

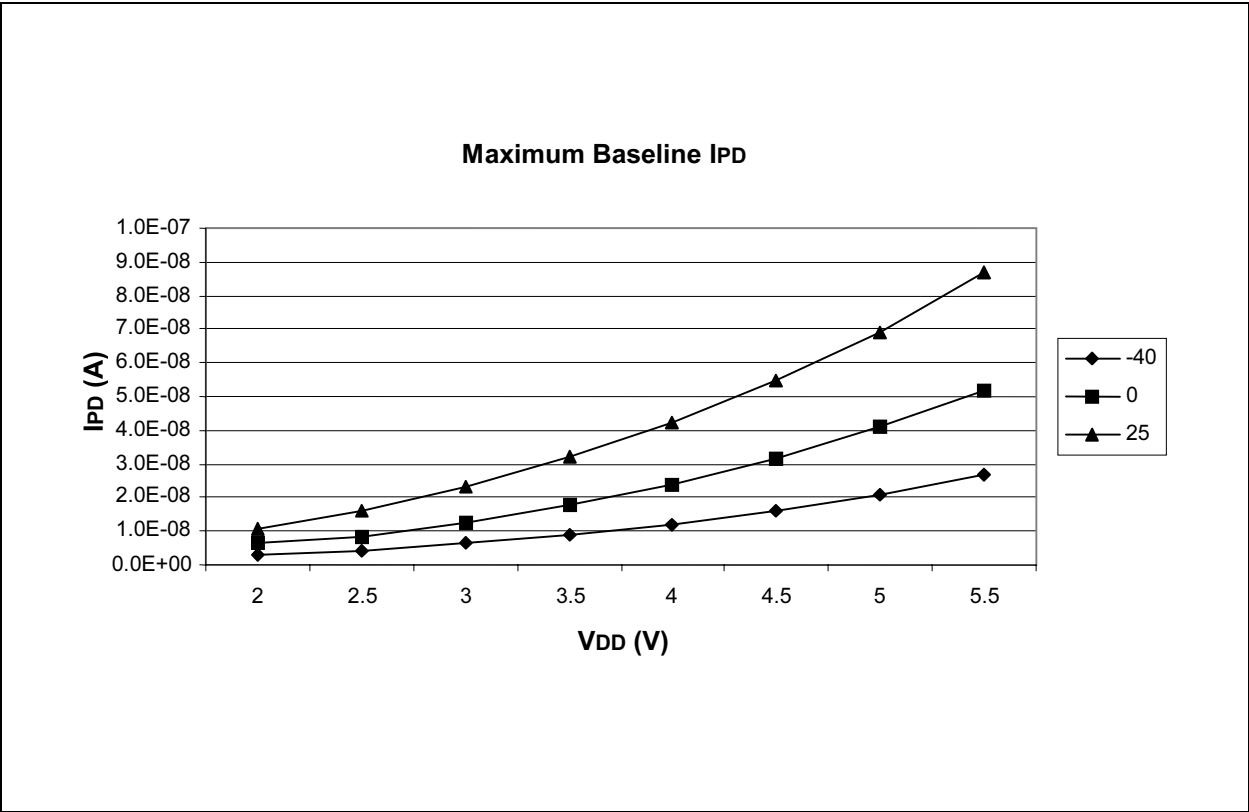


FIGURE 13-5: MAXIMUM IPD vs. VDD OVER TEMP (+85°C)

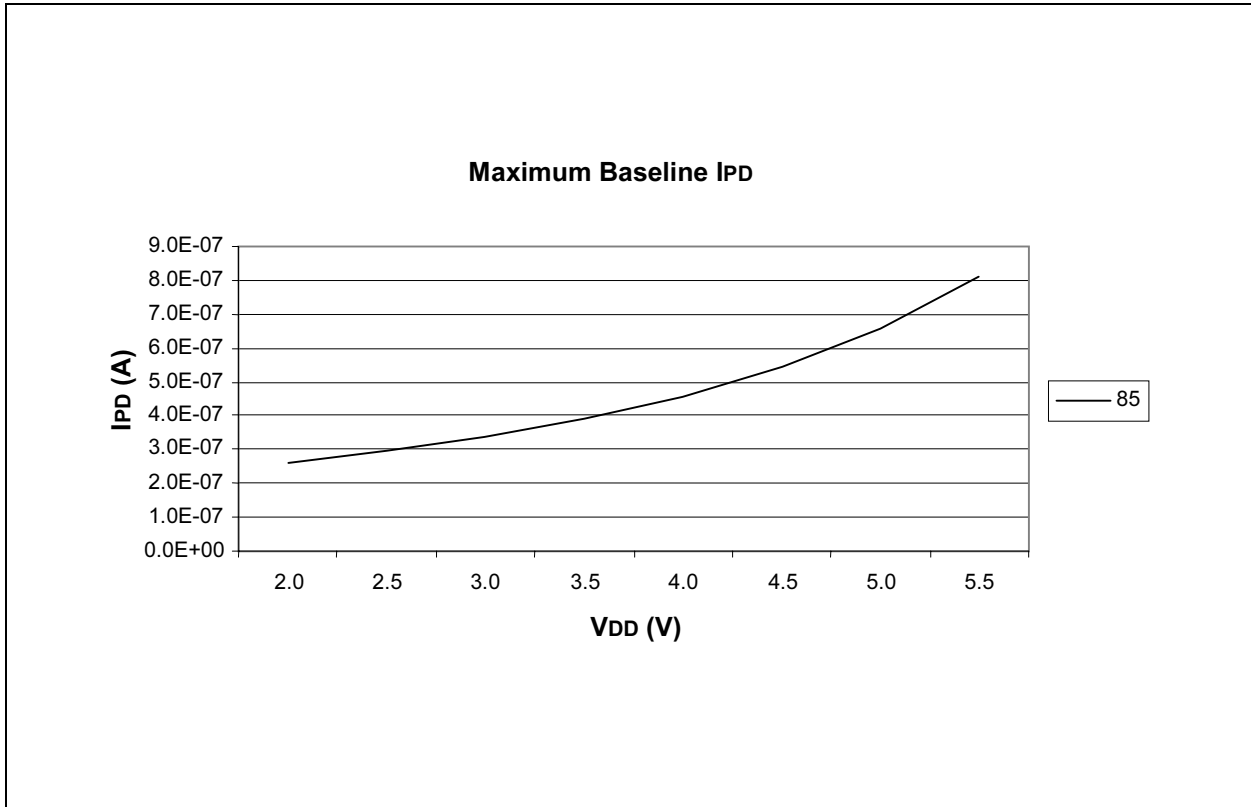
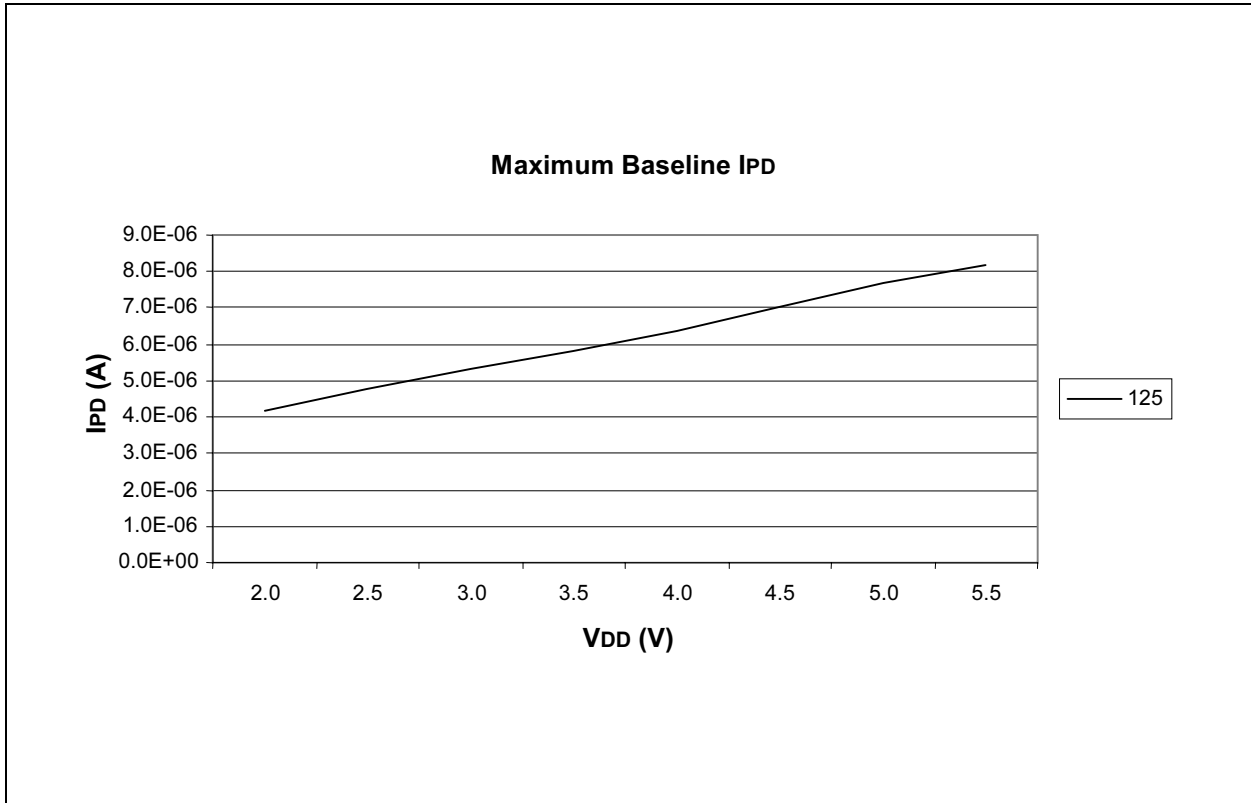


FIGURE 13-6: MAXIMUM IPD vs. VDD OVER TEMP (+125°C)



PIC12F629/675

FIGURE 13-7: TYPICAL I_{PD} WITH BOD ENABLED vs. V_{DD} OVER TEMP (-40°C TO +125°C)

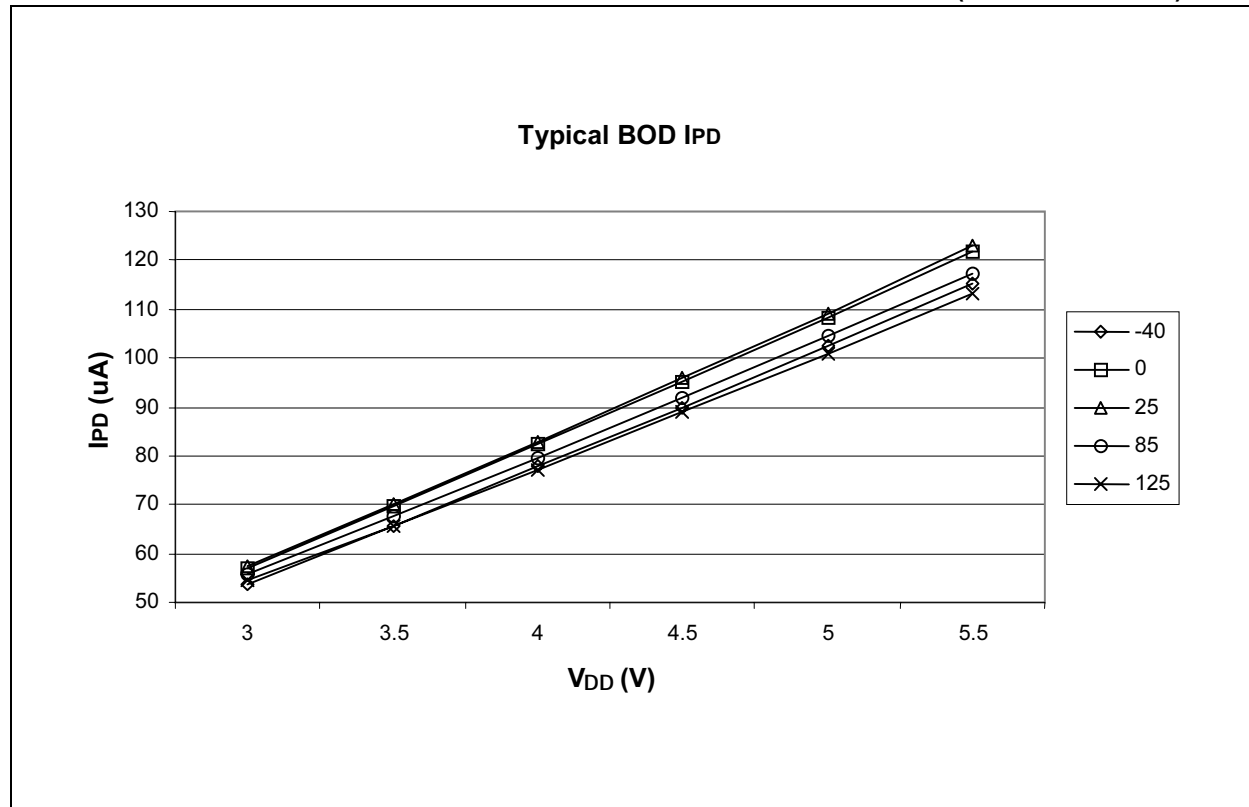


FIGURE 13-8: TYPICAL I_{PD} WITH CMP ENABLED vs. V_{DD} OVER TEMP (-40°C TO +125°C)

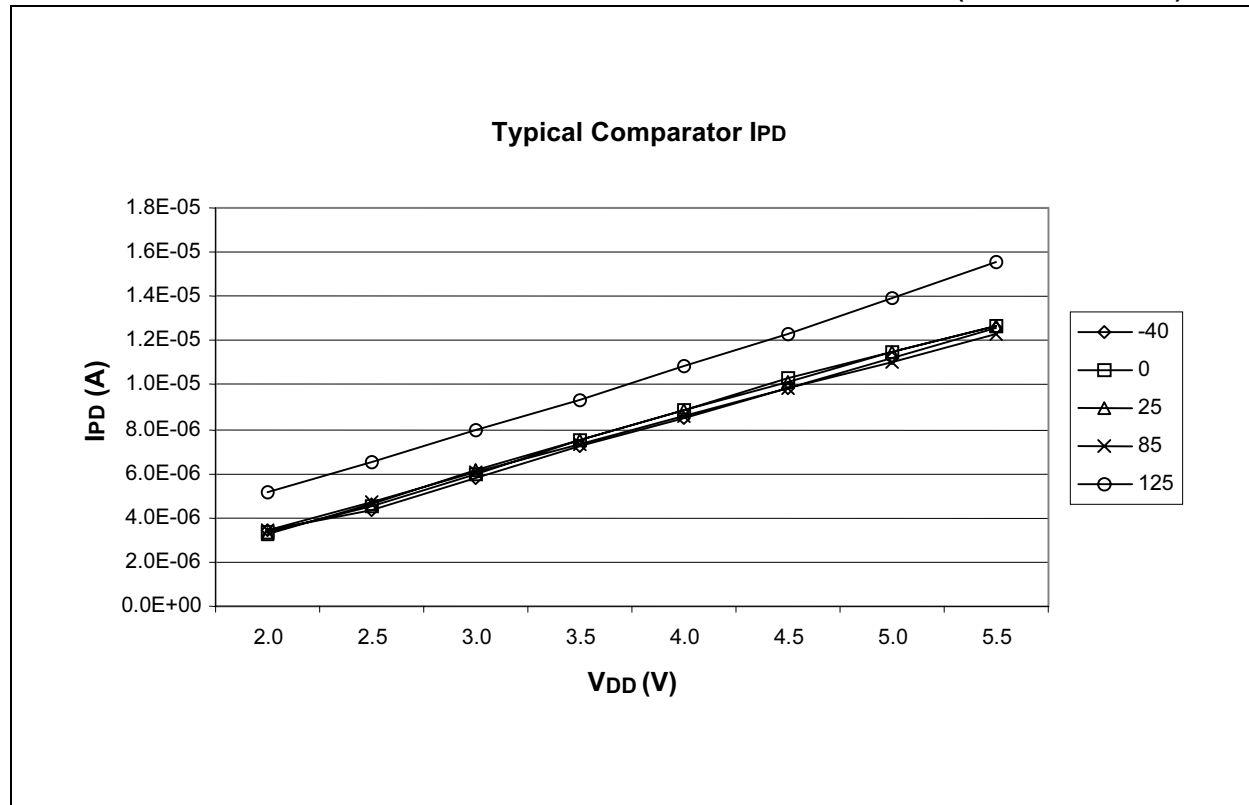


FIGURE 13-9: TYPICAL I_{PD} WITH A/D ENABLED vs. V_{DD} OVER TEMP (-40°C TO +25°C)

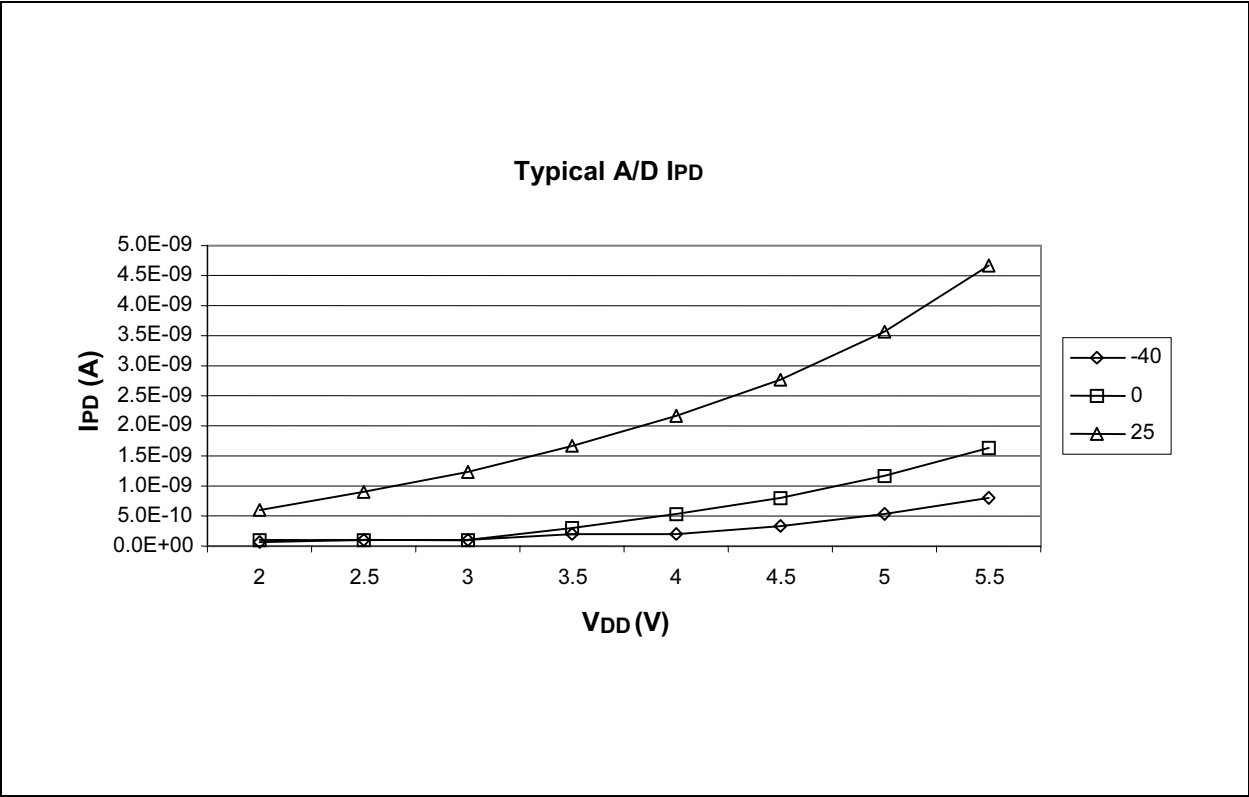
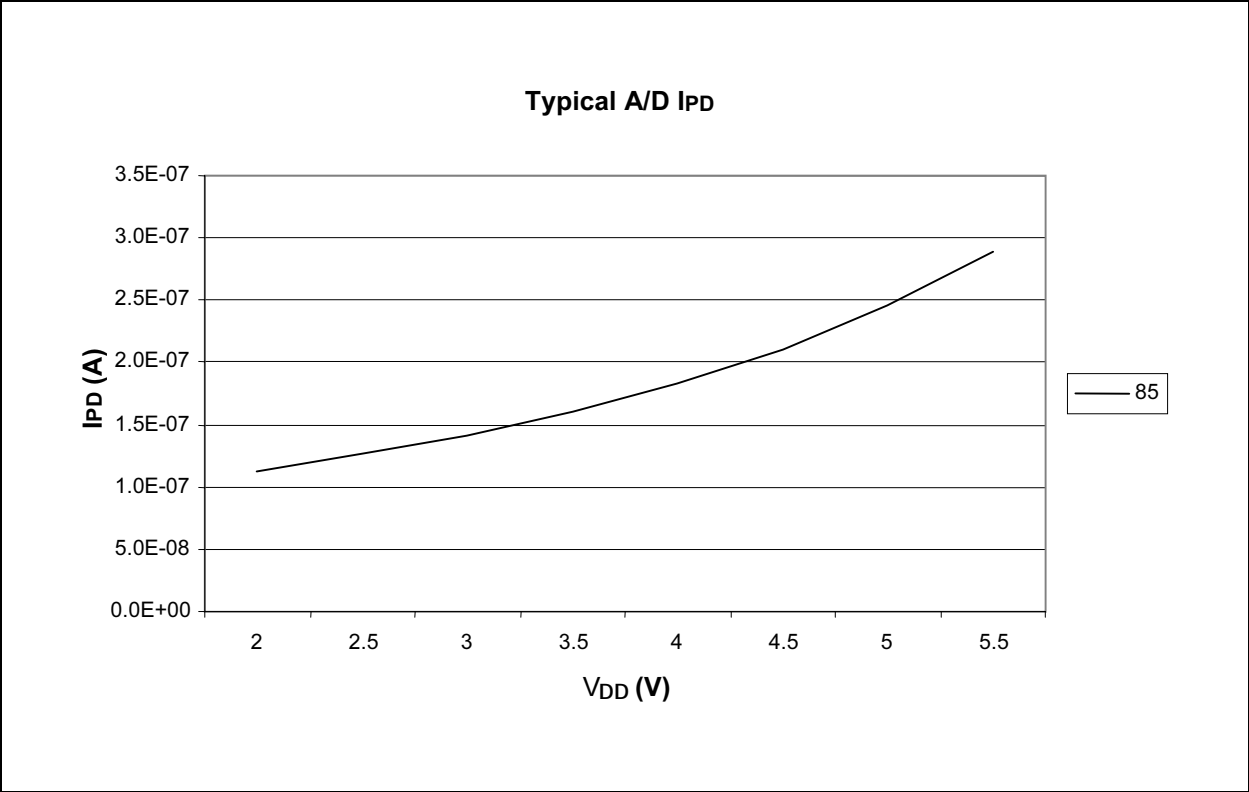


FIGURE 13-10: TYPICAL I_{PD} WITH A/D ENABLED vs. V_{DD} OVER TEMP (+85°C)



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FIGURE 13-11: TYPICAL I_{PD} WITH A/D ENABLED vs. V_{DD} OVER TEMP (+125°C)

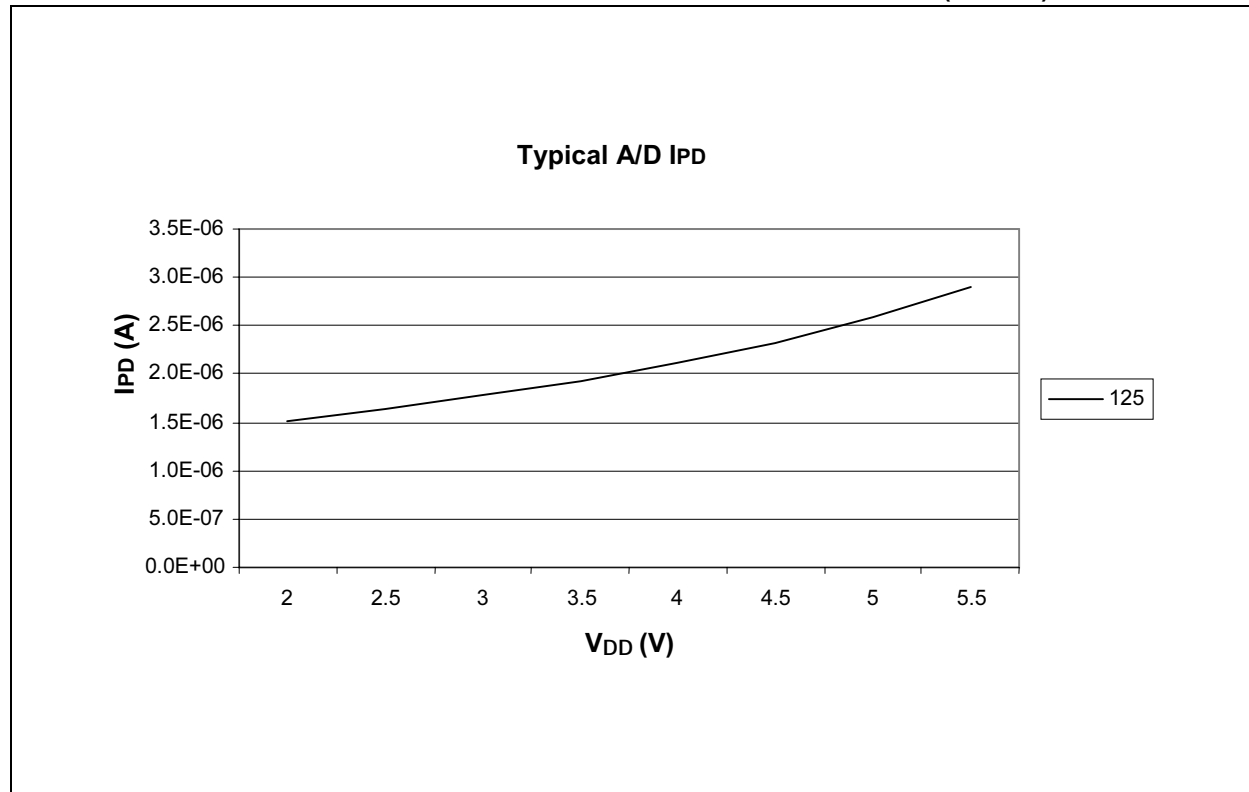


FIGURE 13-12: TYPICAL I_{PD} WITH T1 OSC ENABLED vs. V_{DD} OVER TEMP (-40°C TO +125°C), 32 KHZ, C1 AND C2=50 pF)

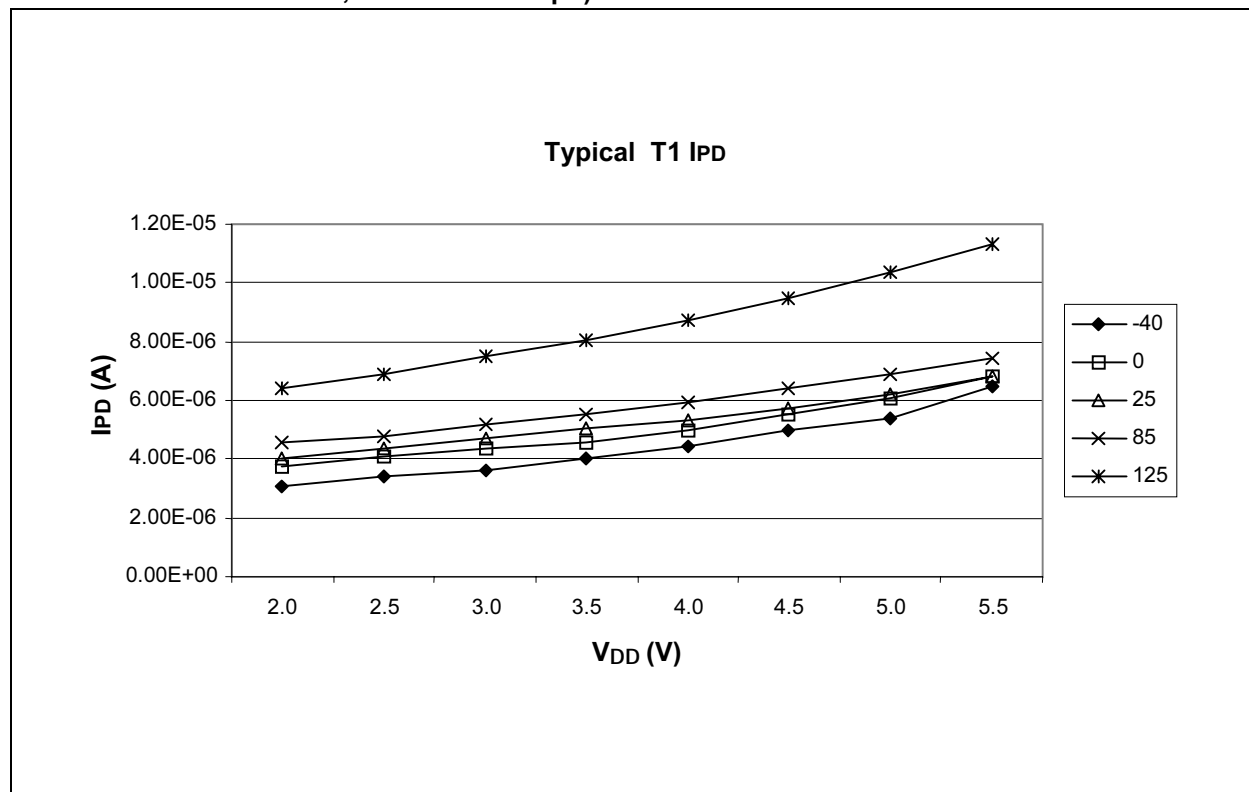


FIGURE 13-13: TYPICAL I_{PD} WITH CV_{REF} ENABLED vs. V_{DD} OVER TEMP (-40°C TO +125°C)

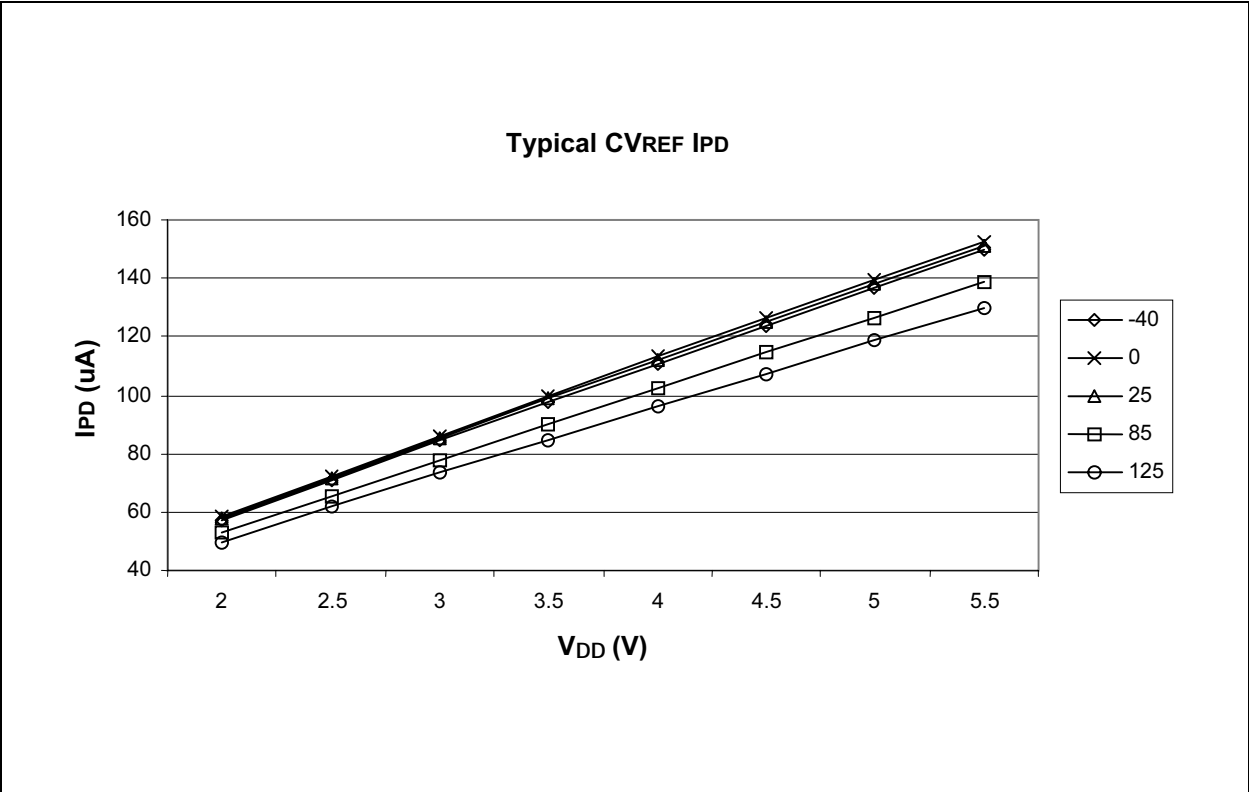
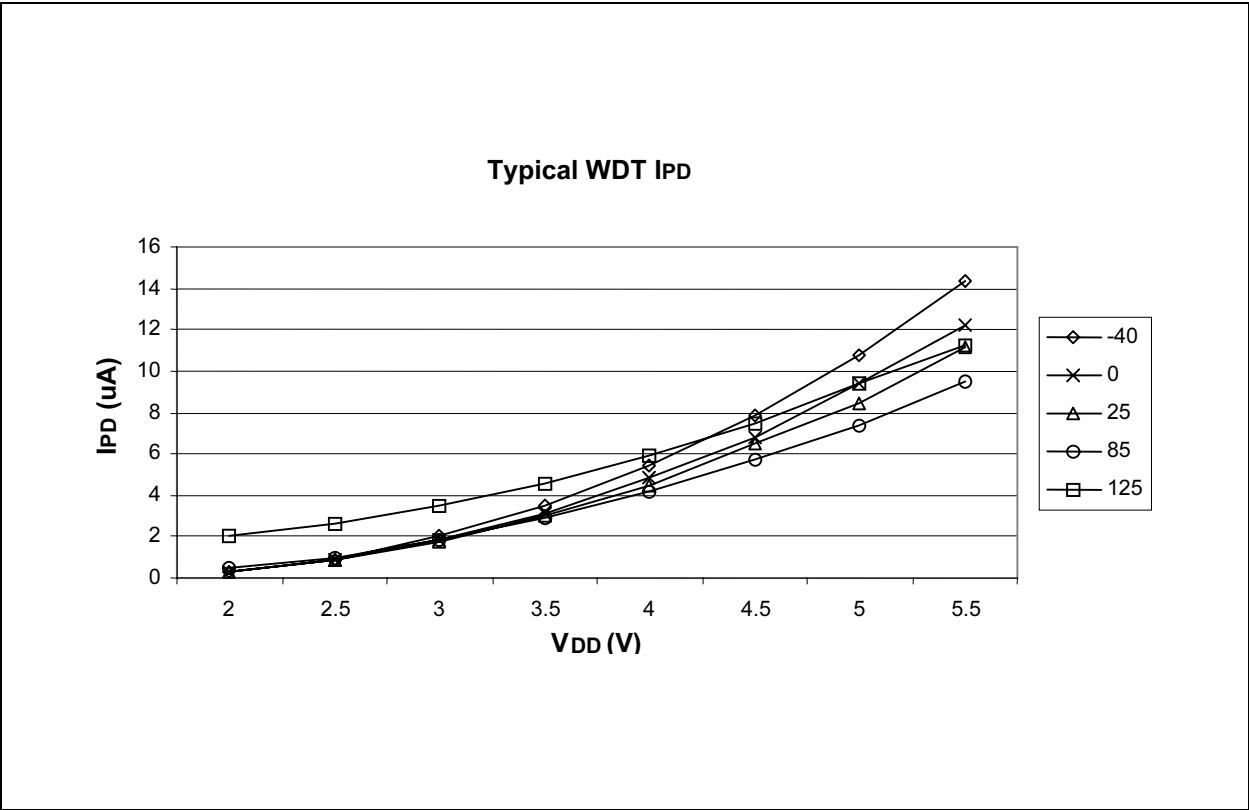


FIGURE 13-14: TYPICAL I_{PD} WITH WDT ENABLED vs. V_{DD} OVER TEMP (-40°C TO +125°C)



PIC12F629/675

FIGURE 13-15: MAXIMUM AND MINIMUM INTOSC FREQ vs. TEMPERATURE WITH 0.1μF AND 0.01μF DECOUPLING (VDD = 3.5V)

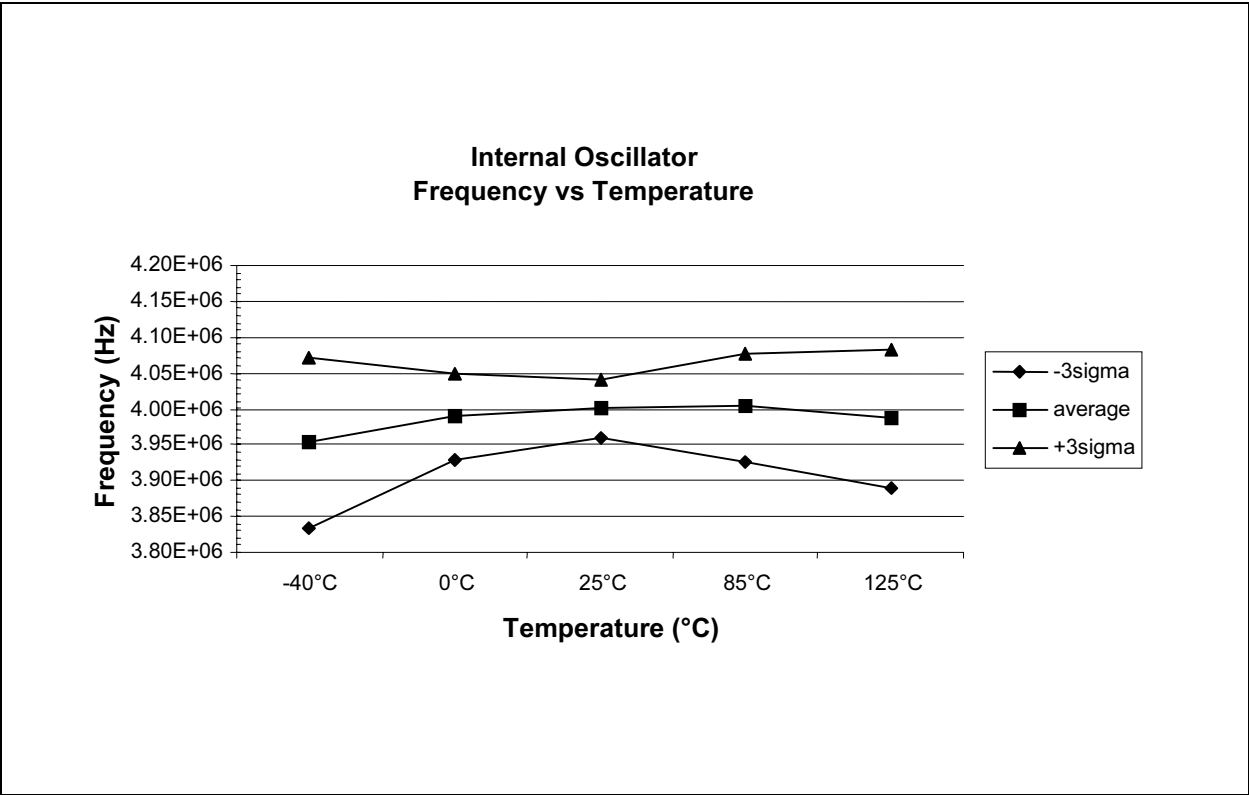


FIGURE 13-16: MAXIMUM AND MINIMUM INTOSC FREQ vs. VDD WITH 0.1μF AND 0.01μF DECOUPLING (+25°C)

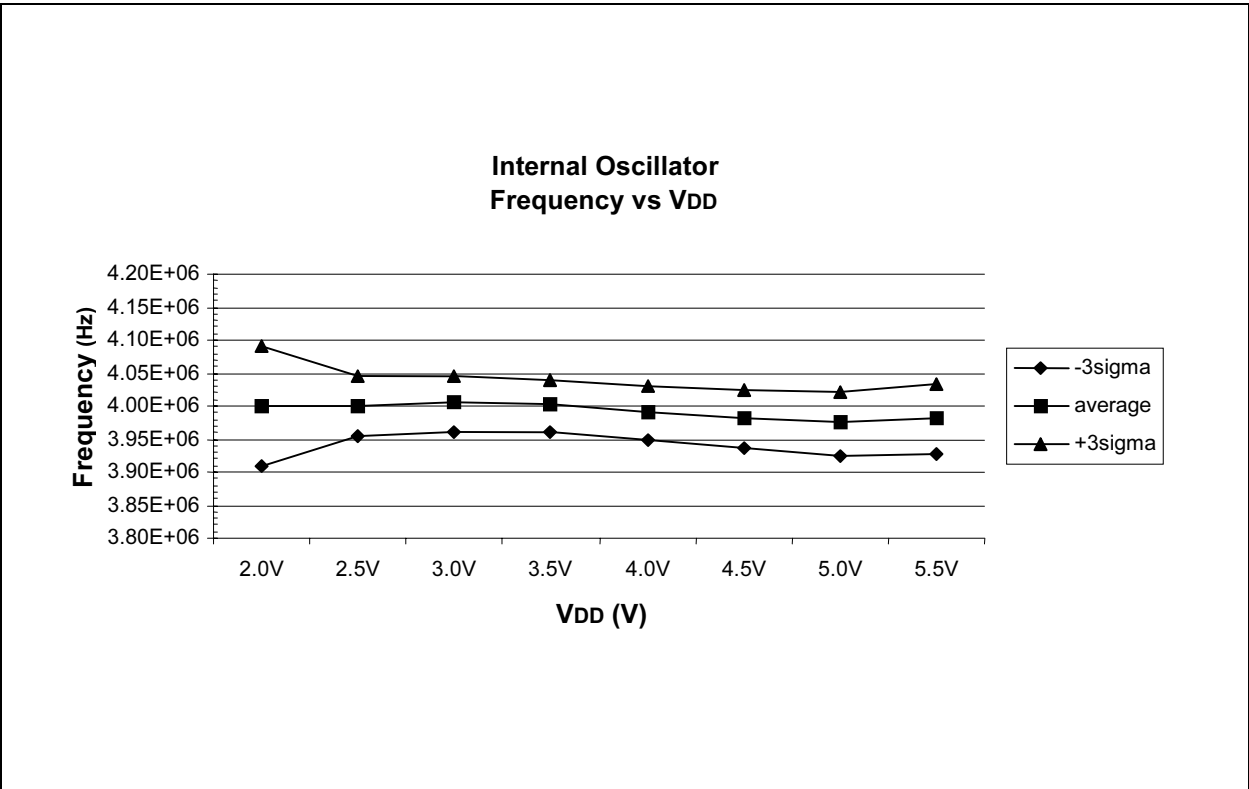
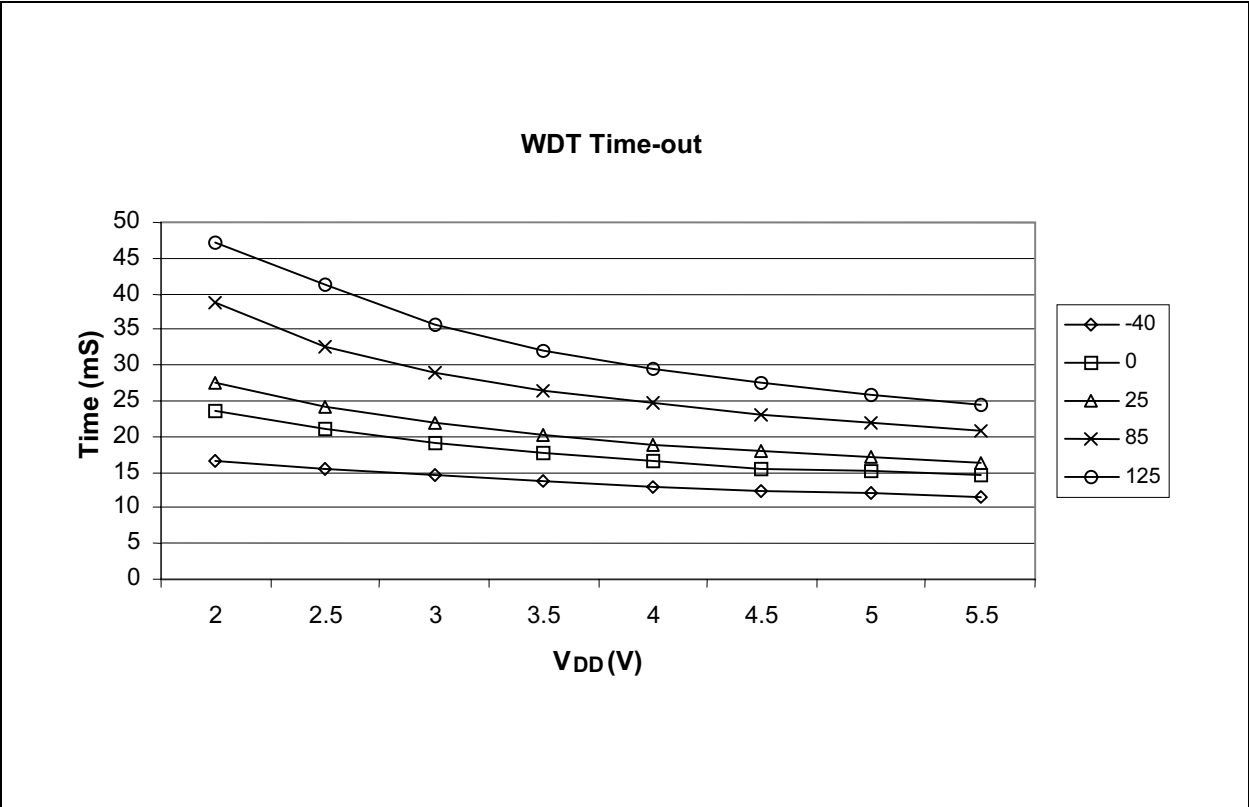


FIGURE 13-17: TYPICAL WDT PERIOD vs. VDD (-40°C TO +125°C)



PIC12F629/675

NOTES:

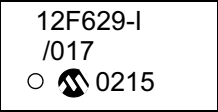
14.0 PACKAGING INFORMATION

14.1 Package Marking Information

8-Lead PDIP (Skinny DIP)



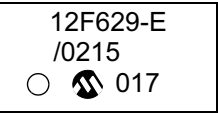
Example



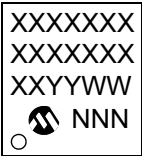
8-Lead SOIC



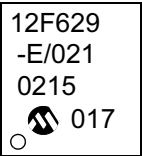
Example



8-Lead DFN-S



Example



Legend:	XX...X	Customer specific information*
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line thus limiting the number of available characters for customer specific information.		

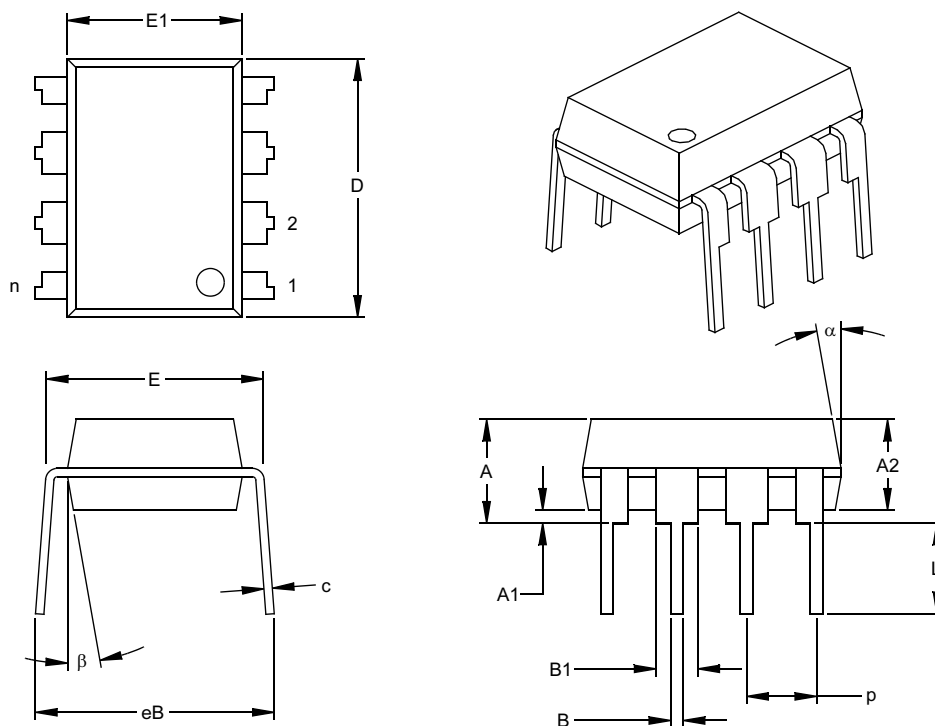
* Standard PICmicro device marking consists of Microchip part number, year code, week code, and traceability code. For PICmicro device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

PIC12F629/675

14.2 Package Details

The following sections give the technical details of the packages.

8-Lead Plastic Dual In-line (P) – 300 mil (PDIP)



Units		INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	p		.100			2.54	
Top to Seating Plane	A	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	c	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	B	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	§ eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

* Controlling Parameter

§ Significant Characteristic

Notes:

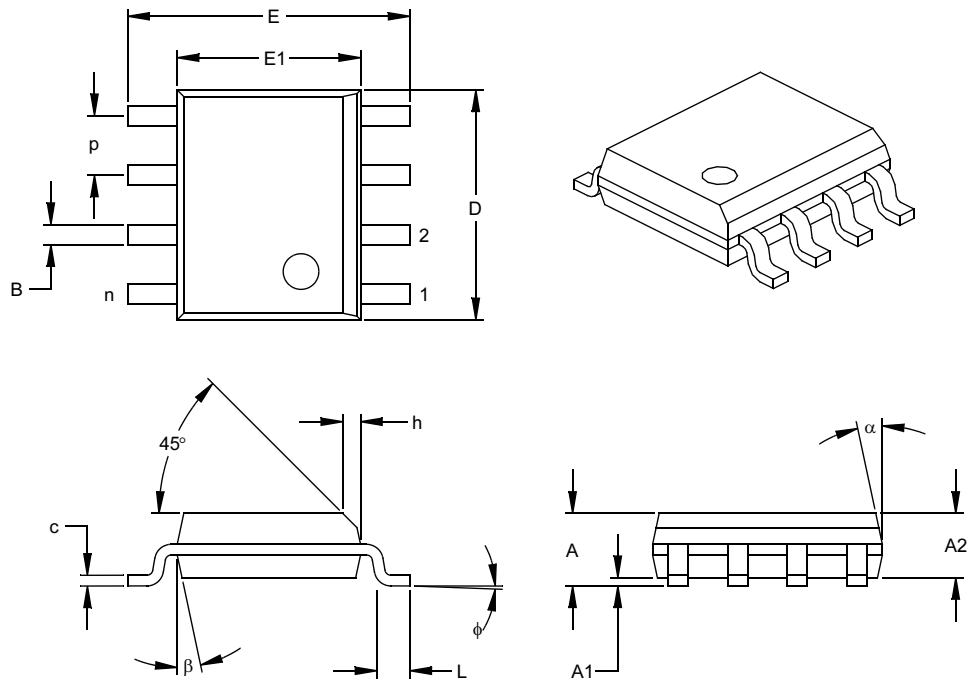
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

.010" (0.254mm) per side.

JEDEC Equivalent: MS-001

Drawing No. C04-018

8-Lead Plastic Small Outline (SN) – Narrow, 150 mil (SOIC)



Units		INCHES*			MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	p		.050			1.27	
Overall Height	A	.053	.061	.069	1.35	1.55	1.75
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25
Overall Width	E	.228	.237	.244	5.79	6.02	6.20
Molded Package Width	E1	.146	.154	.157	3.71	3.91	3.99
Overall Length	D	.189	.193	.197	4.80	4.90	5.00
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51
Foot Length	L	.019	.025	.030	0.48	0.62	0.76
Foot Angle	φ	0	4	8	0	4	8
Lead Thickness	c	.008	.009	.010	0.20	0.23	0.25
Lead Width	B	.013	.017	.020	0.33	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15

* Controlling Parameter
§ Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

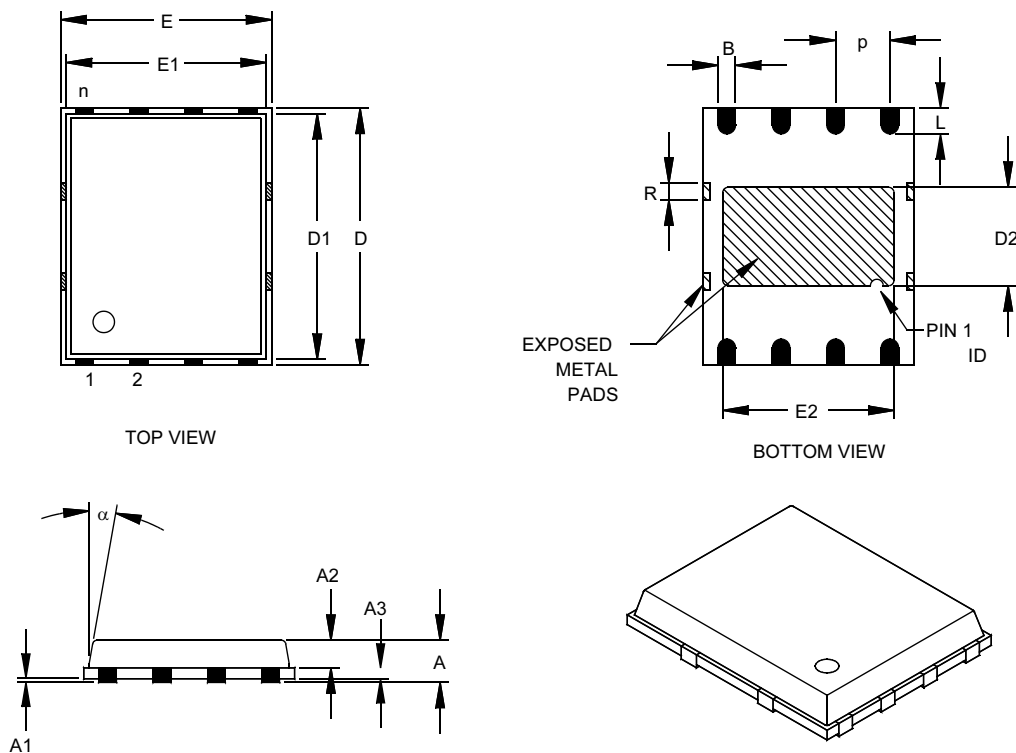
.010" (0.254mm) per side.

JEDEC Equivalent: MS-012

Drawing No. C04-057

PIC12F629/675

8-Lead Plastic Dual Flat No Lead Package (MF) 6x5 mm Body (DFN-S)



Units		INCHES			MILLIMETERS*		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	p		.050 BSC			1.27 BSC	
Overall Height	A		.033	.039		0.85	1.00
Molded Package Thickness	A2		.026	.031		0.65	0.80
Standoff	A1	.000	.0004	.002	0.00	0.01	0.05
Base Thickness	A3		.008 REF.			0.20 REF.	
Overall Length	E		.194 BSC			4.92 BSC	
Molded Package Length	E1		.184 BSC			4.67 BSC	
Exposed Pad Length	E2	.152	.158	.163	3.85	4.00	4.15
Overall Width	D		.236 BSC			5.99 BSC	
Molded Package Width	D1		.226 BSC			5.74 BSC	
Exposed Pad Width	D2	.085	.091	.097	2.16	2.31	2.46
Lead Width	B	.014	.016	.019	0.35	0.40	0.47
Lead Length	L	.020	.024	.030	0.50	0.60	0.75
Tie Bar Width	R		.014			.356	
Mold Draft Angle Top	α			12°			12°

*Controlling Parameter

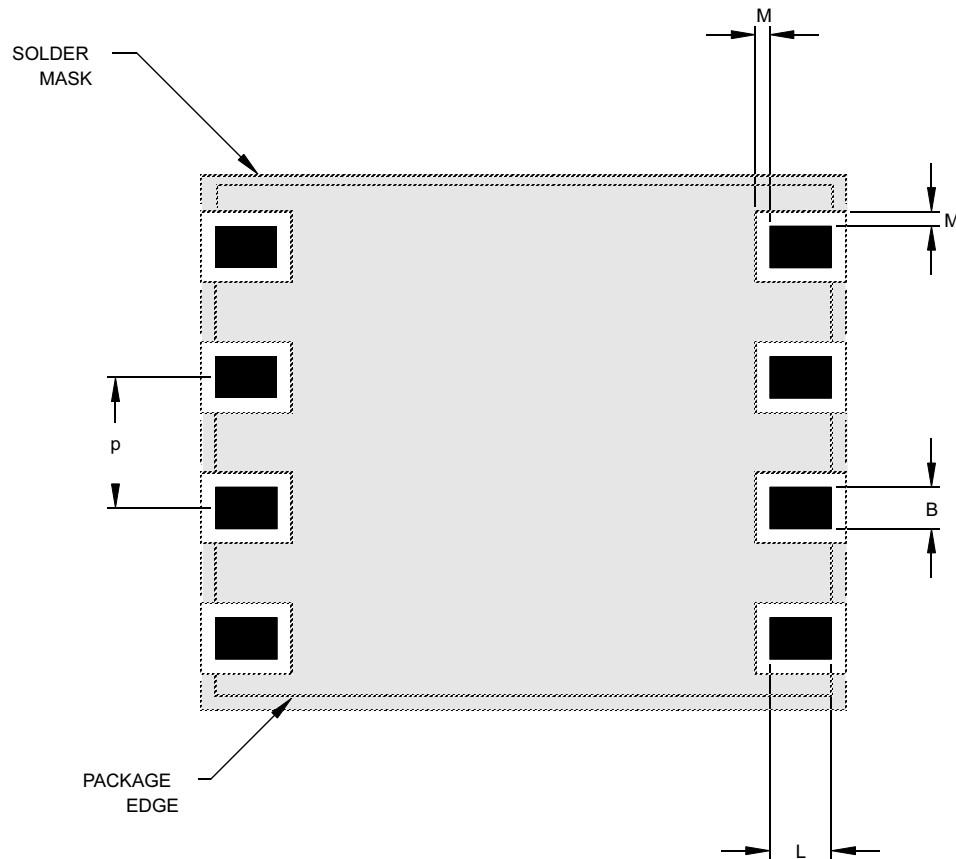
Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

JEDEC equivalent: pending

Drawing No. C04-113

8-Lead Plastic Dual Flat No Lead Package (MF) 6x5 mm Body (DFN-S) Land Pattern and Solder Mask



Units		INCHES			MILLIMETERS*		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Pitch	p	.050 BSC			1.27 BSC		
Pad Width	B	.014	.016	.019	0.35	0.40	0.47
Pad Length	L	.020	.024	.030	0.50	0.60	0.75
Pad to Solder Mask	M	.005		.006	0.13		0.15

*Controlling Parameter

Drawing No. C04-2113

PIC12F629/675

NOTES:

APPENDIX A: DATA SHEET REVISION HISTORY

Revision A

This is a new data sheet.

Revision B

Added characterization graphs.

Updated specifications.

Added notes to indicate Microchip programmers maintain all calibration bits to factory settings and the PIC12F675 ANSEL register must be initialized to configure pins as digital I/O.

Updated MLF-S package name to DFN-S.

APPENDIX B: DEVICE DIFFERENCES

The differences between the PIC12F629/675 devices listed in this data sheet are shown in Table B-1.

TABLE B-1: DEVICE DIFFERENCES

Feature	PIC12F629	PIC12F675
A/D	No	Yes

APPENDIX C: DEVICE MIGRATIONS

This section is intended to describe the functional and electrical specification differences when migrating between functionally similar devices (such as from a PIC16C74A to a PIC16C74B).

Not Applicable

APPENDIX D: MIGRATING FROM OTHER PICmicro® DEVICES

This discusses some of the issues in migrating from other PICmicro devices to the PIC12F6XX family of devices.

D.1 PIC12C67X to PIC12F6XX

TABLE 1: FEATURE COMPARISON

Feature	PIC12C67X	PIC12F6XX
Max Operating Speed	10 MHz	20 MHz
Max Program Memory	2048 bytes	1024 bytes
A/D Resolution	8-bit	10-bit
Data EEPROM	16 bytes	64 bytes
Oscillator Modes	5	8
Brown-out Detect	N	Y
Internal Pull-ups	GP0/1/3	GP0/1/2/4/5
Interrupt-on-change	GP0/1/3	GP0/1/2/3/4/5
Comparator	N	Y

Note: This device has been designed to perform to the parameters of its data sheet. It has been tested to an electrical specification designed to determine its conformance with these parameters. Due to process differences in the manufacture of this device, this device may have different performance characteristics than its earlier version. These differences may cause this device to perform differently in your application than the earlier version of this device.

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Device	PIC12F6XX: Standard VDD range PIC12F6XXT: (Tape and Reel)		
Temperature Range	I = -40°C to +85°C E = -40°C to +125°C		
Package	P = PDIP SN = SOIC (Gull wing, 150 mil body) MF = MLF-S		
Pattern	3-Digit Pattern Code for QTP (blank otherwise)		

Examples:
a) PIC12F629 – E/P 301 = Extended Temp., PDIP package, 20 MHz, QTP pattern #301
b) PIC12F675 – I/SO = Industrial Temp., SOIC package, 20 MHz

* JW Devices are UV erasable and can be programmed to any device configuration. JW Devices meet the electrical requirement of each oscillator type.

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Fuzhou 350001, China
Tel: 86-591-7503506 Fax: 86-591-7503521

China - Hong Kong SAR

Microchip Technology Hongkong Ltd.
Unit 901-6, Tower 2, Metroplaza
223 Hing Fong Road
Kwai Fong, N.T., Hong Kong
Tel: 852-2401-1200 Fax: 852-2401-3431

China - Shanghai

Microchip Technology Consulting (Shanghai)
Co., Ltd.
Room 701, Bldg. B
Far East International Plaza
No. 317 Xian Xia Road
Shanghai, 200051
Tel: 86-21-6275-5700 Fax: 86-21-6275-5060

China - Shenzhen

Microchip Technology Consulting (Shanghai)
Co., Ltd., Shenzhen Liaison Office
Rm. 1812, 18/F, Building A, United Plaza
No. 5022 Binhe Road, Futian District
Shenzhen 518033, China
Tel: 86-755-82901380 Fax: 86-755-82966626

China - Qingdao

Rm. B505A, Fullhope Plaza,
No. 12 Hong Kong Central Rd.
Qingdao 266071, China
Tel: 86-532-5027355 Fax: 86-532-5027205

India

Microchip Technology Inc.
India Liaison Office
Marketing Support Division
Divyasree Chambers
1 Floor, Wing A (A3/A4)
No. 11, O'Shaugnessey Road
Bangalore, 560 025, India
Tel: 91-80-2290061 Fax: 91-80-2290062

Japan

Microchip Technology Japan K.K.
Benex S-1 6F
3-18-20, Shinyokohama
Kohoku-Ku, Yokohama-shi
Kanagawa, 222-0033, Japan
Tel: 81-45-471-6166 Fax: 81-45-471-6122

Korea

Microchip Technology Korea
168-1, Youngbo Bldg. 3 Floor
Samsung-Dong, Kangnam-Ku
Seoul, Korea 135-882
Tel: 82-2-554-7200 Fax: 82-2-558-5934

Singapore

Microchip Technology Singapore Pte Ltd.
200 Middle Road
#07-02 Prime Centre
Singapore, 188980
Tel: 65-6334-8870 Fax: 65-6334-8850

Taiwan

Microchip Technology (Barbados) Inc.,
Taiwan Branch
11F-3, No. 207
Tung Hua North Road
Taipei, 105, Taiwan
Tel: 886-2-2717-7175 Fax: 886-2-2545-0139

EUROPE

Austria

Microchip Technology Austria GmbH
Durisolstrasse 2
A-4600 Wels
Austria
Tel: 43-7242-2244-399
Fax: 43-7242-2244-393

Denmark

Microchip Technology Nordic ApS
Regus Business Centre
Lautrup hof 1-3
Ballerup DK-2750 Denmark
Tel: 45 4420 9895 Fax: 45 4420 9910

France

Microchip Technology SARL
Parc d'Activite du Moulin de Massy
43 Rue du Saule Trapu
Batiment A - 1er Etage
91300 Massy, France
Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79

Germany

Microchip Technology GmbH
Steinheilstrasse 10
D-85737 Ismaning, Germany
Tel: 49-089-627-144-100
Fax: 49-089-627-144-44

Italy

Microchip Technology SRL
Via Quasimodo, 12
20025 Legnano (MI)
Milan, Italy
Tel: 39-0331-742611 Fax: 39-0331-466781

United Kingdom

Microchip Ltd.
505 Eskdale Road
Winnersh Triangle
Wokingham
Berkshire, England RG41 5TU
Tel: 44 118 921 5869 Fax: 44-118 921-5820

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