Polarization Reversal Operator for Polarization-Encoded Quantum Communications

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Abstract—A quantum bit error rate (OBER) based upperlayer protocol agnostic algorithm for polarization random drift compensation is proposed. In contrast to the blind methods currently used, the proposed algorithm finds the polarization reversal operator, which results in much lower overhead. The algorithm is based on the mapping of the QBER on the Poincaré sphere. The algorithm reverts the polarization random drift performing two QBER estimations and applying three rotations, at most. The uncertainty on the two QBER estimations defines an area over the sphere surface that is related with upper-layer protocols QBER threshold. In a "worst case" scenario, where a polarization linewidth of 200 AHz was considered leading to an average transmission window of 0.8 ms, and assuming a 3% QBER threshold, the algorithm provides polarization basis alignment using less than 3% of overhead, assuming an ideal EPC and single-photon detector. However, assuming an off-shelf EPC and single-photon detector, with 20 microseconds actuation time and 25% detection efficiency, the overhead is still below 8%. For a transmission window of 8 ms these values drop to 0.35% and 3%, respectively. This value of transmission windows is still much higher than what is commonly seen in buried fibers, which means that in a normal scenario, and considering realistic components, the algorithm should be able to actuate using less than 3% of overhead.

Index Terms—Communications Society, IEEE, IEEEtran, journal, LATEX, paper, template.

I. Introduction

QANTUM communications (QC) have been implemented using polarization encoding [1] [2]. Nevertheless, standard single-mode optical fibers do not preserve the state of polarization (SOP) and therefore an active polarization basis alignment (PBA) scheme is required, which preserves the quantum information [3]. In order to allow the large deployment of polarization encoding QC systems, the PBA scheme must be efficient, simple, upper layer protocol agnostic, and able to operate in a large variety of environment conditions.

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In [3], the authors quantitatively analyze an interrupted and a real-time method to implement basis alignment. They assess both methods considering the polarization drift time and tracking speed, showing that interrupted PBA is only feasible in stable environments, like low mechanical-stressed buried fibers [3]. In real-time two different approaches have been presented: wavelength-division multiplexing polarization basis alignment (WDM-PBA) [4] [5] [6] [7], and time-division multiplexing polarization basis alignment (TDM-PBA) [5] [8] [9] [10]. In [6], a protocol agnostic scheme is proposed using WDM-PBA in aerial fibers. In [7] it was reported that polarization drift induces a QBER exceeding 2.5% in less than 7 ms in aerial-fibers, when no automatic compensation scheme is used. In [7], SOP tracking is performed using a hillclimbing algorithm in conjugation with a WDM polarization tracking scheme. In [10], it is shown that the achievable reach can be increased by using TDM-PBA based schemes. TDM-PBA may be implemented using classical [5] [8] or quantum reference signals [9]. In classical based TDM-PBA the copropagation can produce a strong degradation in the weak quantum signals [5]. In [9], a TDM quantum reference signal is transmitted along with the quantum data signal, also avoiding the need of using classical and quantum receivers. In [11], a protocol dependent real-time scheme, free of reference signals is presented, where the QKD unveiled bits are used to feed the algorithm to compensate random polarization drifts. That method has the advantage of not add additional overhead, but is not protocol agnostic which can limit its usage. In [12], an accurate QBER estimation method is proposed, and a QBER based PBA method is presented. That method is simple, upperlayer protocol agnostic and able to operate under different external conditions [12]. However, it uses a blind algorithm to align the polarization basis, which makes it quite inefficient, namely under large external condition perturbations [6] [7]. Furthermore, this method uses 12.5% of overhead on average in a laboratory environment, where the polarization remains stable for much longer than in aerial optical fiber installations. In [13], a theoretical polarization drift model, which is able to describe random polarization rotations for installed fibers under different external conditions based on a single parameter is presented. This parameter, named polarization linewidth, mimics how fast the SOP changes with time [13]. In aerial fibers, due to random polarization drifts, transmission windows as short as 1 ms have been reported in polarization encoding quantum communication systems [14].

In this paper, we develop a method to compensate the polarization random drift in optical fibers by mapping the

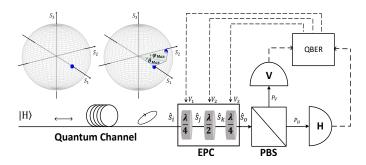


Fig. 1. Horizontal SOP evolution throughout an optical fiber, rotation stage, and detection probabilities at receiver (P_V and P_H). EPC: Electronic Polarization Controller. PBS: Polarization Beam Splitter. V: Single-photon detector in the PBS vertical port. H: Single-photon detector in the PBS horizontal port. V_1 , V_2 and V_3 : Voltages applied on the EPC to induce a certain rotation.

estimated QBER on the Poincaré sphere. In contrast to the current blind methods used to compensate polarization random drift, this method solves the problem of finding the correct polarization reversal operator thereby consuming less bandwidth. We show that polarization random drift can be reversed applying appropriate polarization rotations on the Poincaré sphere, in three iterations at most. This method is able to operate under different external perturbations and it is upper-layer protocol agnostic, does not need auxiliary classical signals, extra spectral bands, nor additional hardware, and provides polarization basis alignment in less than tens of microseconds, which makes it suitable even for aerial fibers applications.

This paper contains five sections. In section II, the algorithm is detailed. In section III, the QBER estimation impact on polarization compensation algorithm efficiency is discussed. In section IV the algorithm behaviour is assessed. A case study is considered assuming a continuous polarization random drift following [13]. Finally, in section IV the main conclusions of this work are summarized.

II. ALGORITHM DESCRIPTION

An optical field SOP can be represented in the 3D-Stokes space [13]. We start by showing that it is also possible to map the QBER on the Poincaré sphere, and then find the appropriate polarization reversal operator. We assume a receiver with two single-photon detectors, see Fig.1. Without any loss of generality, an error at the receiver occurs when a horizontally polarized photon at the quantum communication channel input follows the vertical path of the polarization beam splitter (PBS) at the receiver, inducing a click on the detector V, see Fig.1. Note that any initial polarization state can be reduced to the previous case by a solid rotation of the Poincaré sphere, in the same way that any SOP can be used as a reference for the null QBER.

When a photon reaches the PBS it has the probability P_H to follow the horizontal path [10],

$$P_H = \frac{1}{2}(1 + \cos\theta\cos\varphi),\tag{1}$$

and has the probability P_V to follow the vertical path,

$$P_V = 1 - P_H, \tag{2}$$

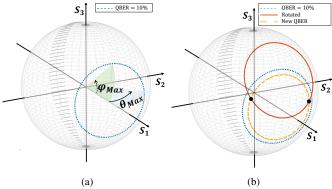


Fig. 2. (a) Circle of a sphere with all possible states on Poincaré sphere that correspond to the QBER = 10%. (b) Circle of a sphere that corresponds to the QBER = 10% rotated considering $\theta_{\rm max}$ and $\varphi_{\rm max}$, and circle of a sphere with all possible states on Poincaré sphere that corresponds to the QBER after the previous rotation. The two • represent the two intersection points that correspond to the two possible SOP location.

where angles $\theta/2$ and $\varphi/2$ correspond to the orientation and ellipticity angles of the arriving photon SOP on Poincaré sphere representation, respectively [15]. Considering a horizontal state at the fiber input, we can write

QBER
$$(\theta, \varphi) = 1 - \frac{1}{2}(1 + \cos\theta\cos\varphi).$$
 (3)

Therefore, a QBER specifies a set of possible orientation and ellipticity angles. This set of values define a circle of a sphere on the Poincaré sphere, which corresponds to a QBER with reference to a given initial state of polarization. In the present case, apart from an horizontal polarized photon at the input of the quantum communication channel, we are also assuming fully polarized light. Therefore the normalized Stokes parameter s_1 can be written as

$$s_1 = \cos(\theta)\cos(\varphi),$$
 (4)

where $\theta \in [0, 2\pi]$, $\varphi \in [-\pi/2, \pi/2]$. The QBER can also be written in terms of s_1 as

QBER
$$(s_1) = \frac{1}{2}(1 - s_1).$$
 (5)

Thus, the circle of a sphere resulting from a QBER value defines a set of possible SOP locations, which are at the same distance from the reference point,

$$d(QBER) = 2 \arcsin(\sqrt{QBER}).$$
 (6)

As we can see in Fig.2a, a single value of QBER has more than one possible polarization reversal operator associated with it, even though a single received SOP leads to a single QBER value. Lets assume that a particular polarization rotation leads to a QBER of 10%, see Fig. 2a. Looking into Fig.2a we can see that the polarization reversal operator still remains unknown, although it is restrict to rotations that lead the SOP from $(s_1, s_2, s_3)^T = (1, 0, 0)^T$, i.e. a horizontal initial state of polarization, to a point on the circle of the sphere that represents the 10% QBER. A subsequent deterministic rotation in conjunction with a new QBER calculation allows to reduce the number of possible polarization reversal operators to only

$$R_{T}(\theta, \varphi) = R_{1}(\varphi)R_{2}(\varphi)R_{3}(\theta), \tag{7}$$

where, R_1 , R_2 , and R_3 are the rotation matrices around the axis S_1 , S_2 , and S_3 , respectively,

$$\mathbf{R}_{\mathbf{1}}(\varphi) = \begin{bmatrix} 1 & 0 & 0 \\ 0 & \cos \varphi & -\sin \varphi \\ 0 & \sin \varphi & \cos \varphi \end{bmatrix}, \tag{8}$$

$$R_{2}(\varphi) = \begin{bmatrix} \cos \varphi & 0 & \sin \varphi \\ 0 & 1 & 0 \\ -\sin \varphi & 0 & \cos \varphi \end{bmatrix}, \tag{9}$$

and

$$\mathbf{R_3}(\theta) = \begin{bmatrix} \cos \theta & -\sin \theta & 0\\ \sin \theta & \cos \theta & 0\\ 0 & 0 & 1 \end{bmatrix}. \tag{10}$$

Once a rotation has been performed, as shown in Fig.2b, another calculation of QBER is done. Lets assume the performed rotation was done using the orientation angle $\theta_{\text{max}}/4$ and ellipticity angle $\varphi_{max}/4$, where θ_{max} and φ_{max} are the maximum angles defined by the circle of the sphere corresponding to the first QBER value calculated, see Fig. 2a. A new value for OBER allows to draw another circle of a sphere on Poincaré sphere, which intercepts the previous rotated circle in two points, which are shown in Fig. 2b with circle marks. Therefore, the initial infinite number of possible polarization reversal operators is reduced to only two possibilities, which correspond to the reversal operator of the two intersection points. In order to obtain an analytical expression for the two intersection points, we can consider the parametric equations of a 3D circle, see (11). Note that m takes the value 1 for the initial QBER rotated circle, and 2 for the circle of a sphere after the QBER re-calculation,

$$\begin{cases} x^{(m)} = x_c^{(m)} + r_m \cos(\phi) x_{m1} + r_m \sin(\phi) x_{m2} \\ y^{(m)} = y_c^{(m)} + r_m \cos(\phi) y_{m1} + r_m \sin(\phi) y_{m2} \\ z^{(m)} = z_c^{(m)} + r_m \cos(\phi) z_{m1} + r_m \sin(\phi) z_{m2} \end{cases} , \tag{11}$$

where ϕ is a real value between 0 and 2π .

In (11), $(x_c^{(m)}, y_c^{(m)}, z_c^{(m)})$ are the center coordinates, and r_m is the radius of each circle m. Note that after measuring the QBER, a circle of sphere is defined. From (3), $(x_c^{(m)}, y_c^{(m)}, z_c^{(m)})$, r_m , and the plane containing the circle defined by the orthogonal vector $\vec{n} = \vec{v}_{m1} \times \vec{v}_{m2}$, where $\vec{v}_{m1} = (x_{m1}, y_{m1}, z_{m1})$ and $\vec{v}_{m2} = (x_{m2}, y_{m2}, z_{m2})$, can be readily obtained.

The two measured QBER values define two circles of a sphere that intersect in two points, which can be obtained from

$$\begin{cases} x^{(1)} = x^{(2)} \\ y^{(1)} = y^{(2)} \\ z^{(1)} = z^{(2)} \end{cases}$$
 (12)

and represented in the 3D-Stokes space by,

$$s_1^{(n)} = \cos \theta^{(n)} \cos \varphi^{(n)}$$

 $s_2^{(n)} = \sin \theta^{(n)} \cos \varphi^{(n)}$
 $s_3^{(n)} = \sin \varphi^{(n)},$ (13)

where $n \in \{1,2\}$. Subsequently, the algorithm chooses a value of n to perform a new rotation. Lets assume that we pick n = 1. After applying a rotation with angles $(\theta^{(1)}, \varphi^{(1)})$, the QBER is recalculated, see (3). If the QBER approaches zero, the polarization random drift has been compensated. Otherwise, the polarization random drift compensation can now be uniquely calculated by the following polarization reversal operator,

$$R_{\rm T}(\theta^{(2)}, \varphi^{(2)}) R_{\rm T}^{-1}(\theta^{(1)}, \varphi^{(1)}).$$
 (14)

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In any of the two scenarios, the algorithm needs only three QBER calculations and three rotations at most to revert the polarization random drift due to birefringence effects along the optical fiber link.

To summarize, the algorithm comprises the following stages:

- i Map the circle corresponding to the QBER calculated on Poincaré sphere according with (3), see Fig.2a.
- ii Perform a rotation with a given θ and φ , see Fig.2b.
- iii Calculate a new QBER, see Fig.2b. If the QBER decreases bellow a user defined threshold the algorithm stops to actuate in the current mode. Otherwise, it continues as following.
- iv Find the two intersection points, see (12) and (13).
- v Choose one out of the two intersection points to perform a new rotation, so that $R_T(\theta^{(i)}, \varphi^{(i)})\hat{s}^{(n)} = (1, 0, 0)^T$, see (7).
- vi Re-calculate the QBER. If the QBER approaches zero, the polarization random drift has been reverted. Otherwise, perform another rotation following (14).
- vii The random polarization drift is reverted.

Note that the voltages applied on the EPC, V_1 , V_2 , V_3 , can be written in terms of angles θ and φ . These voltages induce a certain phase shift on the wave-plates, which implies a rotation of the SOP based on a set of rotation angles, χ_1 , χ_2 , χ_3 . These angles can be written in terms of the orientation and ellipticity angles, θ and φ . Looking into Fig. 1, a random SOP \hat{s}_i inputs the EPC facing the first quarter-wave-plate(QWP) that outputs, in turn, the state of polarization \hat{s}_i ,

$$\hat{s}_i = R(\chi_1) M_{\lambda/4} R(-\chi_1) \hat{s}_i, \tag{15}$$

where $M_{\lambda/4}$ is the QWP matrix [16] and $R = R_3(2\chi_1)$ is the rotation matrix of the wave-plate . The angle χ_1 is given by [16],

$$\chi_1 = \frac{1}{2} \arctan \left(\frac{\sin \theta \cos \varphi}{\cos \theta \sin \varphi} \right). \tag{16}$$

The second wave-plate is a half-wave-plate (HWP), and transforms the linear SOP \hat{s}_j into another linear SOP [16], which in practice means a rotation by θ around S_3 when $\varphi = 0$,

$$\hat{s}_k = R(\chi_2) M_{\lambda/2} R(-\chi_2) \hat{s}_i, \tag{17}$$

where, $M_{\lambda/2}$ is the HWP matrix, and $\hat{s}_j = (s_{1j}, s_{2j}, 0)^T$ defined by (10) in [16]. In this way,

$$\chi_2 = \frac{1}{4} \arctan\left(\frac{s_{2j}}{s_{1j}}\right),\tag{18}$$

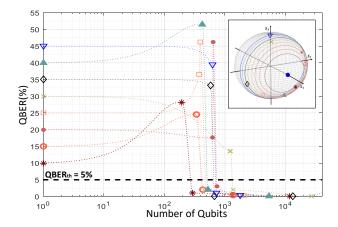


Fig. 3. QBER evolution during random polarization random drift compensation algorithm running. Markers represent QBER measurements. The SOP locations are represented on the inset sphere on the right.

where, s_{1j} , s_{2j} are defined by (19) and (20), respectively.

$$s_{1j} = s_{1i} \cos^2(2\chi_1) + s_{2i} \cos(2\chi_1) \sin(2\chi_1)$$

$$+ s_{3i} \sin(2\chi_1),$$
(19)

$$s_{2j} = s_{1i} \cos 2\chi_1 \sin(2\chi_1) + s_{2i} \sin^2(2\chi_1)$$

$$-s_{3i} \cos(2\chi_1),$$
(20)

In addition, $s_{1i} = \sin \theta \cos \phi$ and $s_{2i} = \cos \theta \sin \varphi$, see (13). Finally, the EPC output SOP is defined as,

$$\hat{s}_o = R(\chi_3) M_{\lambda/4} R(-\chi_3) \hat{s}_k, \tag{21}$$

where,

$$\chi_3 = \frac{1}{2} \arctan\left(\frac{s_{2k}}{s_{1k}}\right). \tag{22}$$

Fig. 3 shows numerical results for different initial QBER values between 10% and 45%. Note that in a practical scenario QBER is estimated taken into account a certain number of received qubits, N_r using

$$\widehat{\text{QBER}} = \frac{e_r}{N_r},\tag{23}$$

where e_r is the number of errors in N_r qubits [12]. This estimation is performed with a certain confidence interval which depends on the number of qubits that we use to perform it. In this section, we are assuming that N_r is large enough to provide an accurate estimation of the QBER. In the next section, we are going to consider the impact of N_r during the algorithm's running. We also assume a threshold QBER_{th} = 5%. Above the threshold, the quantum communication system cannot operate. The goal of the presented algorithm is to force the QBER to be below the threshold. Fig. 3 shows eight different cases corresponding to different initial conditions, where it is shown that regardless the respective initial QBER, the final QBER is always below the threshold. The inset in Fig. 3 shows the location of the different SOPs on Poincaré sphere, where each one is on the circle of a sphere corresponding to the QBER measured. In Fig. 3 every case starts from an initial QBER estimation, i.e the first marker. The algorithm starts from this initial QBER estimation and performs the first

rotation. The second marker is the QBER estimation after the first rotation. Here, the algorithm chooses one of the two intersection points, see (12). After that, we wait for a 5 errors event, or for 100 qubits received to estimate the new QBER. Note that more than 5 errors implies an estimated QBER larger than the threshold, in this case where a 5% threshold was assumed. When it wrongly chooses the intersection point, the next marker is a QBER above the QBER_{th}. On the other hand, when it rightly chooses the intersection point, the next marker is a OBER below threshold, and the following. A final marker with a high accuracy on QBER estimation is also included in the figure, to show the proper operation of the algorithm. In Fig.3, we have shown that the algorithm find the appropriate polarization reversal operator and compensates any polarization random drift leading the initial QBER to a value below the threshold after two or three rotations, at most.

III. IMPACT OF QBER ESTIMATION ACCURACY

In order to assess the impact of the QBER estimation accuracy in the algorithm performance, we are going to use a new coordinate γ , such that

$$\cos(\gamma) = \cos\theta\cos\varphi,\tag{24}$$

where γ is the angle between the axis S_1 , and the vector defined by the SOP. In this way, the QBER in (3) can be written in terms of γ as

QBER
$$(\gamma) = \frac{1}{2}(1 - \cos \gamma)$$
. (25)

The number of qubits required for each algorithm iteration is the sum of the qubits used for each QBER estimation, occurring in stages (i), (iii) and (vi), see section II. The last QBER estimation, at stage (vi), does not lead to any rotation, and therefore does not require high accuracy. In this way, we can assume that

$$n_1, n_2 \gg n_3, \tag{26}$$

where n_1 , n_2 and n_3 are the number of qubits used in QBER estimations at (i), (iii), and (vi), respectively. Therefore, the total number of qubits required to compensate the polarization random drift can be written as, see (15) from [12],

$$n_b \simeq n_1(\Delta QBER_1, QBER_1, \alpha)$$

+ $n_2(\Delta QBER_2, QBER_2, \alpha),$ (27)

where $\Delta QBER_1$ and $\Delta QBER_2$ are the uncertainty associated with the QBER₁ and QBER₂ estimations at stage (i) and (iii) of the algorithm, respectively, and $1 - \alpha$ is the confidence interval. Note that the QBER estimation uncertainty, at stages (i) and (iii), can be written as

$$\Delta QBER_i = QBER(\gamma_i + \delta \gamma_i) - QBER(\gamma_i - \delta \gamma_i), \qquad (28)$$

where $\delta \gamma_i$ is the maximum deviation on γ_i , see Fig. 4. At stage (iv) of the algorithm, i.e. after two QBER estimations, an area can be defined due to the QBER estimation uncertainties, see inset on Fig. 4.

From (25) and (28), the uncertainty $\triangle QBER_i$ at stages (i) and (iii) can be related to the corresponding γ_i , as well as to $\delta \gamma_i$, using

$$\Delta QBER_i \approx \delta \gamma_i \sin \gamma_i. \tag{29}$$

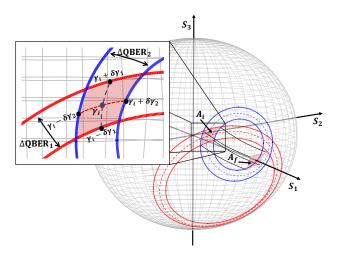


Fig. 4. Representation of the area defined by the uncertainties of the first and second QBER estimations on the Poincaré sphere surface, A_i . This area is preserved after the final rotation, i.e. $A_i = A_f$.

The induced rotation into the SOP associated with the QBER estimations, at stage (v), is going to place the uncertainty area, A_i , around the target SOP, preserving its shape, A_f , as shown in Fig. 4. Note that the final QBER, QBER_f, is null at $\gamma_f = 0$, therefore from (25) we obtain

$$\Delta QBER_f = QBER(\delta \gamma_f) \approx \frac{\delta \gamma_f^2}{4}.$$
 (30)

The final QBER estimation uncertainty depends on the first and second QBER estimations, and as a consequence, the ΔQBER_f , see (30), depends on the $\delta \gamma_1$ and $\delta \gamma_2$. Note that the uncertainties ΔQBER_1 and ΔQBER_2 define an area that remains constant after the final rotation, see Fig. 4. In the worst case scenario, $\delta \gamma_f$ will be the sum of both uncertainties $\delta \gamma_1$ and $\delta \gamma_2$,

$$\delta \gamma_f \le \delta \gamma_1 + \delta \gamma_2. \tag{31}$$

For a given confidence interval $1 - \alpha$, the algorithm satisfies

$$P(QBER_f \ge \Delta QBER_f) \le 2\alpha.$$
 (32)

Following this discussion, we can calculate the number of qubits required for QBER estimation at stages (i) and (iii) of the algorithm, so that the polarization control random drift algoritm assures a QBER_f below a certain QBER threshold, QBER_{th}, with a certain probability. Note that in a small rotation regime in stage (ii), we can also assume that $\delta \gamma_1 \approx \delta \gamma_2 \approx \delta \gamma$, and $\gamma_1 \approx \gamma_2 \approx \gamma$, which implies Δ QBER₁ $\approx \Delta$ QBER₂. Therefore $n_1 \approx n_2 \approx n$, and so that the total number of required qubits will be $n_b = 2n$.

Fig. 5 shows the number of qubits used to perform each QBER estimation, n, calculated given a certain QBER_{th}, using (30) and (31) to calculate $\delta \gamma$, and using (25) to obtain γ in order to obtain the initial QBER uncertainty. Using the initial QBER uncertainty, the initial QBER, and for a given confidence level using (27), we can obtain the total number of required qubits, n_b , and subsequently n, the number of qubits to estimate QBER in stage (i) and (iii). Fig. 5 shows that the number of qubits required, n, is almost independent of

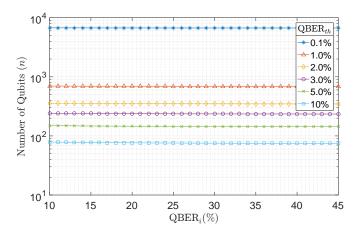


Fig. 5. Number of qubits, n, required for each QBER estimation, at stages (i) and (iii) of the algorithm, aiming to reach a final QBER below the threshold. A confidence interval of $1 - \alpha = 99\%$ was used, and n was calculated for different QBER_{th} considering an initial QBER from 10% to 45%.

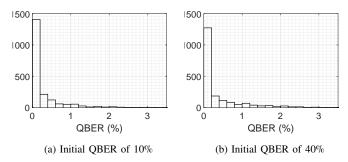


Fig. 6. Histogram of the final QBER for two different values for the initial QBER. A QBER_{th} of 3% was defined as well as a confidence level of 99% for each QBER estimation, at stages (i) and (iii) of the algorithm.

the initial QBER, QBER_i. On the contrary, the final QBER threshold has a high impact on the number of qubits required. As one can see in Fig. 5 the number of qubits is inversely proportional to the QBER threshold, i.e for smaller QBER_{th} a larger number of qubits is required. Table I summarizes the number of qubits required for each QBER estimation, at stages (i) and (iii), for a given QBER threshold.

TABLE I Number of qubits required for estimation of QBER in terms of uncertainty QBER $_{\mathrm{TH}}$ for a confidence level of 99%.

QBER _{th}	n(QBER _{th})
0.1 %	6743
1.0 %	699
2.0 %	357
3.0 %	243
5.0 %	150
10.0 %	79

In order to assess the algorithm, we perform a simulation for a QBER threshold of 3%, considering two initial values for QBER, 10% and 40%. The SOP at the receiver input is randomly chosen between all possible SOP on the circle

of a sphere corresponding with the desirable initial QBER. Following Table I, and considering the 3% threshold, we use 243 qubits to estimate each QBER, at stages (i) and (iii) of the algorithm. We run 1000 simulations for each initial QBER, and the results are shown in Fig. 6. Moreover, the reached QBER estimation was performed with a high accuracy, 3500 qubits were used. Note that this estimation is not part of the algorithm, and it was only performed here to assess the algorithm performance. The obtained results shows that for an initial QBER of 10% and 40%, the reached QBER is above the QBER_{th} only in 1.3% and 1.8% of the cases, respectively, which is in good agreement with (32), considering a confidence level of 99%, i.e. $\alpha = 1\%$.

IV. ALGORITHM OVERHEAD

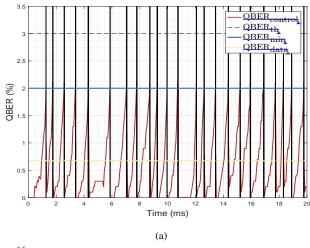
We applied the proposed algorithm to a realistic quantum optical communication system using polarization encoding. We assumed that the system operates at 100 Mqubit/s, which is a reasonable value considering current single photon detectors technology [17].

A. With Perfect Devices

In a first instance, we only assess the algorithm actuation impact on a quantum communication system considering perfect devices, i.e. an EPC that actuates instantaneously, ideal single-photon sources, zero signal attenuation and single-photon detectors with unitary efficiency. We also assumed that the number of photons in the control qubits was adjusted to make the non-click probability and the double click probability negligible. Our goal was to maintain the QBER threshold and protocol overhead bellow 3%, These values should allow current quantum communication protocols to operate smoothly [1811] 9].

der heavy external conditions, for instance in aerial optical fiber installations, polarization can remain stable in a window as short as 1 ms [14]. To consider this "worst case" scenario, we modeled the polarization drift following [13], with a polarization linewidth of 20μ Hz, see Fig. 7a. In Fig. 7b we consider a better scenario where polarization remains stable for around 10 ms, where a polarization linewidth of 200 nHz was used.

The polarization control system comprises two operation modes: a monitoring mode and an actuation mode. In the monitoring mode the QBER is a user defined sliding window. We assumed a 1 ms sliding window for the QBER estimation in the situation corresponding to high polarization linewidth, see Fig. 7a, and a 10 ms in a more stable situation corresponding to a low polarization linewidth, see Fig. 7b. The user also defines a QBER mine smaller than the QBER threshold. As long as the estimated QBER remains below this QBER_{min}, the algorithm runs only in the monitoring mode. We assumed a 2% value for the QBER_{min}. Considering the QBER_{th} of 3% and the uncertainty values associated with the QBER estimation presented in table I, we used 1 control qubit in each 100 qubits for QBER monitoring in the situation represented in Fig. 7a, which allows to estimate the QBER in the 1 ms sliding window with 1000



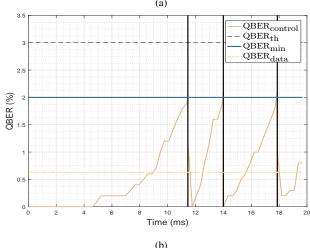


Fig. 7. (a) - QBER monitoring with actuation of the proposed algorithm for polarization random drift compensation in a worst situation, where polarization linewidth is high, i.e in the order of $20\mu Hz$.(b) - QBER monitoring with actuation of the proposed algorithm for polarization random drift compensation in a environment controlled situation, where polarization linewidth is low, i.e in the order of 200 nHz.

qubits. In Fig. 7b we used 1 control qubit in each 500 qubits for QBER monitoring, which allows to estimate the QBER in the 10 ms sliding window with 2000 qubits. Note that the number of photons in the control qubits was adjusted to make the non-click probability negligible. When the QBER rises above the QBER_{min}, the algorithm enters in the actuation mode. In this mode, the algorithm follows the steps presented in section III. In the actuation mode, all transmitted qubits are used for polarization control. After the algorithm actuation, the QBER estimation window is reset.

In order to assess the algorithm's performance, we measured the algorithm's overhead, actuation frequency, average QBER, and maximum QBER for both situations shown in Fig. 7a and Fig. 7b. The algorithm's overhead is defined as the ratio between the sum of qubits used for polarization monitoring and control, and all transmitted qubits. The algorithm's actuation frequency is defined as the number of times that the algorithm actuates per unit of time. The average and maximum QBER are calculated considering the data qubits. To demonstrate the algorithm execution on the defined scenario, we performed

several simulations during 20 ms time windows. In Fig. 7a and Fig. 7b, we present the results of a single simulation run for each case, all the other results are similar.

Fig. 7a and Fig. 7b show the evolution of OBER according with a high and low polarization linewidth, respectively, using the proposed algorithm to find the polarization reversal operator to compensate the drift. As it is shown, whenever the QBER rises above the QBER_{min}, 2.0%, the algorithm actuates, being able to reestablish the qubits data transmission in less than 30 mic onds in Fig. 7a, and in less than 10 microseconds in Fig. 7b. The actuation times are represented by vertical lines in the plots. The width of the lines correspond to the algorithm actuation time. In average, the algorithm actuates 1.15 times per millisecond, imposing a transmission window of 0.8 ms with an overhead of 2.54% in the case represented in Fig. 7a. For the case presented in Fig. 7b the algorithm actuates 0.15 times per millisecond, imposing a transmission window of 8 ms with an overhead of 0.31%. The average QBER during data gubits transmission remained below the 3% threshold in both situations, and a maximum QBER of 2.1% was obtained. Note that, even in the low polarization linewidth considered scenario the data transmission window is around 8 ms on average, which is well far from the transmission windows that we find in buried optical fibers, and we achieve an overhead well bellow the value presented in [12] of 12.5% obtained in a laboratory scenario where the transmission windows are in the order of seconds.

B. With Imperfect Devices

Now, we will assess the algorithm's performance considering the available imperfect devices and its technical limitations. First, we consider the power losses throughout the optical fiber channel. We assumed a typical value for attenuation of 0.2 dB/km, which is the loss through a silica optical fiber for 1550 nm [20]. Since we are working on a quantum regime, the effect of attenuation will highly impact the no-click probability. Furthermore, the no-click event is also caused by the detection efficiency of single-photon detectors. In order to overcome the issues related with no-click events, and reduce its impact on algorithm's performance, the number of photons in control pulses is optimized. By increasing the number of photons in each control pulse, where the photons are not perfectly aligned due polarization random drift, can lead to double-click events. These events are also caused by dark-counts, which depend in part on single-photon detectors efficiency. Both, no-click and double-click events will impact the overhead consumed by the algorithm, since the qubits measured in that situation are discarded and not taken into account for QBER estimation. Now, the need to established a balanced number of photons in control pulses has arisen, and they should be defined to be convenient for the best trade-off between both no-click and double-click events. In this way, we perform the same simulation considering different numbers of photons per control pulse, with a transmission rate of 100 Mqubits/s, an average number of photons per data pulse of 0.1 at the transmitter output, since it assures a pulse intensity that prevents the beam splitting attacks [21]. Two optical fiber

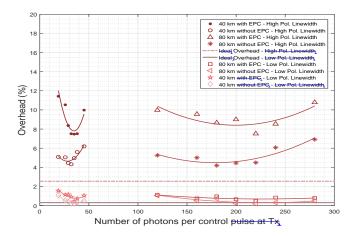


Fig. 8. Overhead measured for different average number of photons per control pulse considering two lengths for optical channel. For each fiber length the overhead were measured considering the EPC actuation time and considering it negligible. Two values for polarization linewidth were considered, a high value of 20 μ Hz, and a low value of 200 nHz. The overhead obtained considering perfect devices is also shown for the two polarization linewidths

channel lengths were defined to perform these simulations, 80 km and 40 km. Fig. 8 shows the overhead according to the average number of photons per pulse defined at the transmitter output. This number varies with optical channel length i order to assure seven different values for the average numb of photons per pulse that input the receiver. Note that, for 40 km optical channel length the number of photons needed at transmitter output is much lower that for 80 km optical channel length, since the signal suffers lower attenuation. The four curves placed above 4% overhead correspond to a high polarization linewidth, where the transmission window is the order of 1 ms, and it is shown that the overhead does not depend on the optical channel length, as well as the optimum number of photons per control pulse, which in both cases correspond to approximately 5 photons per pulse at the receiver input imposing an overhead bellow 5% considering the EPC actuation time negligible.

Due technological limitations, the EPC does not induce an instantaneous rotation, and it demands a certain time interval to stabilize the output SOP. In this way, the EPC actuation time should be considered in the system assessment. An EPC actuation time of 20μ s was assumed [6], which corresponds to increase the overhead by 2000 qubits in each performed rotation for a transmission rate of 100 MQubits/s. Fig. 8 also shows the overhead resulted from adding the EPC actuation time for both optical channel lengths. Even through the overhead increases due this technological limitation it remains below 9%, which nevertheless is lower than the value presented in [12] considering a shorter transmission window.

The overhead obtained for both ideal scenarios previously considered is also shown in Fig. 8 corresponding to 2.54% and 0.31% for high and low polarization linewidth, respectively. Moreover, overhead according to the number of photons per control pulse was also calculated for both optical channel lengths considering a low value for polarization linewidth,

i.e. 200 nHz that imposes a transmission window of 8 ms on average. As shown in Fig. 8, the overhead remains bellow 1.5% for every number of photons considered, even taken into account all equipment technological limitations. Note that, even the biggest considered transmission window is still applied in aerial optical fiber scenarios, which means that in buried optical fiber networks this transmission window is higher, and so that the overhead can decrease.

V. CONCLUSION

We presented an algorithm to automatically compensate the polarization random drift in polarization-encoding based quantum transmission systems. This algorithm is based on QBER estimation and on its representation on the Poincaré sphere, allowing to find the appropriate polarization reversal operator that enables to compensate polarization random drift avoiding bandwidth consuming blind techniques. From the estimated QBER, a circle on the Poincaré sphere is defined leading to a set of possible polarization states. By performing a deterministic rotation, the algorithm reduces this set of polarization states to only two possible SOP. From this two possible states of polarization the algorithm is able to compensate the polarization random drift in a very short time.

It was shown that the proposed algorithm is always able to force the QBER to a value bellow a user-defined threshold in three iterations, at most. In addition, the uncertainty in the final QBER was related with an area calculated in the Poincaré sphere surface based on two QBER estimation uncertainties. From this area, we obtain the number of qubits required in the QBER estimations to guarantee a final QBER bellow the threshold.

Moreover, the proposed algorithm was applied on a case study, first assuming an ideal scenario where the technical devices limitations are neglected. It was assumed two values for polarization linewidth, 20μ Hz as a high polarization linewidth, which imposes a transmission window of around 0.8 ms, and 200 nHz as a low value, which imposes a transmission window of 8 ms. In both situations, the algorithm was capable of maintain the OBER bellow the 3% threshold using only 2.54% of overhead, and 0.31% of overhead, respectively. Furthermore, when imperfect devices limitations are taken into account, namely the EPC actuation time, no ideal single-photon sources, attenuation, and single-photon detectors efficiency, the overhead increases but never crosses the 9%. This value was obtained for a worst case, and even though it is lower than the value of 12.5% presented in [12] acquired for a scenario with a much bigger transmission window. An optimum average number of photons per control pulse at receiver of 5 was found. Regarding the low polarization linewidth scenario, the overhead decreases to a value bellow 1.5%. This transmission window is well smaller than what is found in buried optical fibers. Therefore, the proposed algorithm should be able to find the polarization reversal operator to compensate polarization random drift in buried optical fiber networks using very little bandwidth.

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