

PCA9622

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

Rev. 5 — 2 June 2014

Product data sheet

1. General description

The PCA9622 is an I²C-bus controlled 16-bit LED driver optimized for voltage switch dimming and blinking 100 mA Red/Green/Blue/Amber (RGBA) LEDs. Each LED output has its own 8-bit resolution (256 steps) fixed frequency individual PWM controller that operates at 97 kHz with a duty cycle that is adjustable from 0 % to 99.6 % to allow the LED to be set to a specific brightness value. An additional 8-bit resolution (256 steps) group PWM controller has both a fixed frequency of 190 Hz and an adjustable frequency between 24 Hz to once every 10.73 seconds with a duty cycle that is adjustable from 0 % to 99.6 % that is used to either dim or blink all LEDs with the same value.

Each LED output can be off, on (no PWM control), set at its individual PWM controller value or at both individual and group PWM controller values. The PCA9622 operates with a supply voltage range of 2.3 V to 5.5 V and the 100 mA open-drain outputs allow voltages up to 40 V.

The PCA9622 is one of the first LED controller devices in a new Fast-mode Plus (Fm+) family. Fm+ devices offer higher frequency (up to 1 MHz) and more densely populated bus operation (up to 4000 pF).

The active LOW Output Enable input pin (\overline{OE}) blinks all the LED outputs and can be used to externally PWM the outputs, which is useful when multiple devices must be dimmed or blinked together without using software control.

Software programmable LED Group and three Sub Call I²C-bus addresses allow all or defined groups of PCA9622 devices to respond to a common I²C-bus address, allowing for example, all red LEDs to be turned on or off at the same time or marquee chasing effect, thus minimizing I²C-bus commands. Seven hardware address pins allow up to 126 devices on the same bus.

The Software Reset (SWRST) Call allows the master to perform a reset of the PCA9622 through the I²C-bus, identical to the Power-On Reset (POR) that initializes the registers to their default state causing the outputs to be set HIGH (LED off). This allows an easy and quick way to reconfigure all device registers to the same condition.

The PCA9622, PCA9625 and PCA9635 software is identical and if the PCA9622 on-chip 100 mA NAND FETs do not provide enough current or voltage to drive the LEDs, then the PCA9635 with larger current or higher voltage external drivers can be used.



16-bit Fm+ I²C-bus 100 mA 40 V LED driver

2. Features and benefits

- 16 LED drivers. Each output programmable at:
 - Off
 - On
 - Programmable LED brightness
 - Programmable group dimming/blinking mixed with individual LED brightness
- 1 MHz Fast-mode Plus compatible I²C-bus interface with 30 mA high drive capability on SDA output for driving high capacitive buses
- 256-step (8-bit) linear programmable brightness per LED output varying from fully off (default) to maximum brightness using a 97 kHz PWM signal
- 256-step group brightness control allows general dimming (using a 190 Hz PWM signal) from fully off to maximum brightness (default)
- 256-step group blinking with frequency programmable from 24 Hz to 10.73 s and duty cycle from 0 % to 99.6 %
- Sixteen open-drain outputs can sink between 0 mA to 100 mA and are tolerant to a maximum off state voltage of 40 V. No input function.
- Output state change programmable on the Acknowledge or the STOP Command to update outputs byte-by-byte or all at the same time (default to 'Change on STOP').
- Active LOW Output Enable (OE) input pin allows for hardware blinking and dimming of the LEDs
- 7 hardware address pins allow 126 PCA9622 devices to be connected to the same I²C-bus and to be individually programmed
- 4 software programmable I²C-bus addresses (one LED Group Call address and three LED Sub Call addresses) allow groups of devices to be addressed at the same time in any combination (for example, one register used for 'All Call' so that all the PCA9622s on the I²C-bus can be addressed at the same time and the second register used for three different addresses so that ¹/₃ of all devices on the bus can be addressed at the same time in a group). Software enable and disable for I²C-bus address.
- Software Reset feature (SWRST Call) allows the device to be reset through the l²C-bus
- 25 MHz internal oscillator requires no external components
- Internal power-on reset
- Noise filter on SDA/SCL inputs
- No glitch on power-up
- Supports hot insertion
- Low standby current
- Operating power supply voltage (V_{DD}) range of 2.3 V to 5.5 V
- 5.5 V tolerant inputs on non-LED pins
- -40 °C to +85 °C operation
- ESD protection exceeds 2000 V HBM per JESD22-A114 and 1000 V CDM per JESD22-C101
- Latch-up testing is done to JEDEC Standard JESD78 which exceeds 100 mA
- Packages offered: TSSOP32

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

3. Applications

- RGB or RGBA LED drivers
- LED status information
- LED displays
- LCD backlights
- Keypad backlights for cellular phones or handheld devices

4. Ordering information

Table 1. Ordering information

| Type number | Topside | Package | | |
|-------------|-----------|---------|---|----------|
| marking | | Name | Description | Version |
| PCA9622DR | PCA9622DR | TSSOP32 | plastic thin shrink small outline package; 32 leads; body width 6.1 mm; lead pitch 0.65 mm | SOT487-1 |

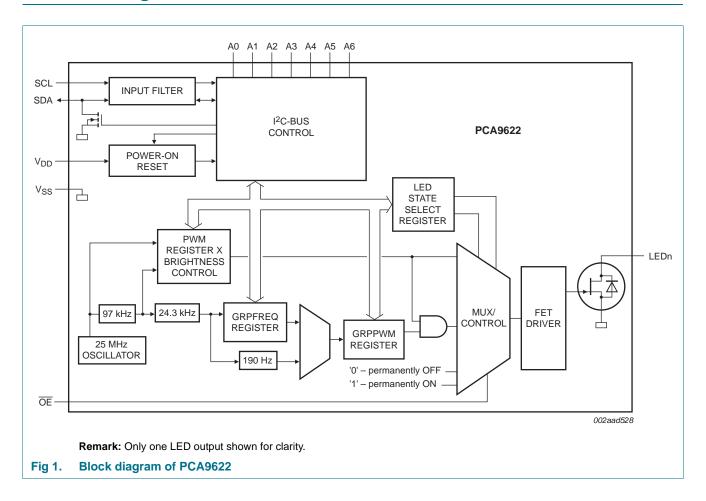
4.1 Ordering options

Table 2. Ordering options

| 71 | Orderable part number | Package | 9 | Minimum order quantity | Temperature |
|-----------|-----------------------|---------|-----------------------------------|------------------------|---|
| PCA9622DR | PCA9622DR,118 | TSSOP32 | Reel 13" Q1/T1 *Standard mark SMD | 2000 | $T_{amb} = -40 ^{\circ}\text{C} \text{ to } +85 ^{\circ}\text{C}$ |

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

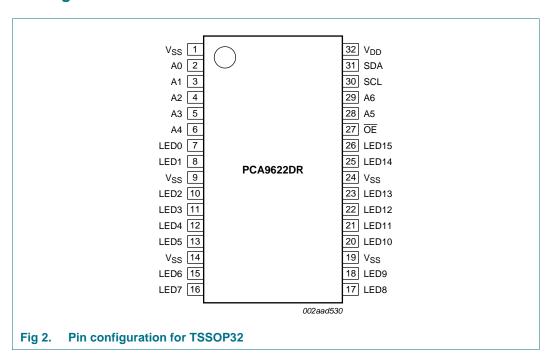
5. Block diagram



16-bit Fm+ I²C-bus 100 mA 40 V LED driver

6. Pinning information

6.1 Pinning



6.2 Pin description

Table 3. Pin description

| Symbol | Pin | Туре | Description |
|----------|-----|--------------|-----------------|
| V_{SS} | 1 | power supply | supply ground |
| A0 | 2 | I | address input 0 |
| A1 | 3 | I | address input 1 |
| A2 | 4 | I | address input 2 |
| A3 | 5 | I | address input 3 |
| A4 | 6 | I | address input 4 |
| LED0 | 7 | 0 | LED driver 0 |
| LED1 | 8 | 0 | LED driver 1 |
| V_{SS} | 9 | power supply | supply ground |
| LED2 | 10 | 0 | LED driver 2 |
| LED3 | 11 | 0 | LED driver 3 |
| LED4 | 12 | 0 | LED driver 4 |
| LED5 | 13 | 0 | LED driver 5 |
| V_{SS} | 14 | power supply | supply ground |
| LED6 | 15 | 0 | LED driver 6 |
| LED7 | 16 | 0 | LED driver 7 |
| LED8 | 17 | 0 | LED driver 8 |
| LED9 | 18 | 0 | LED driver 9 |

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

| Symbol | Pin | Туре | Description |
|-----------------|-----|--------------|--------------------------|
| V_{SS} | 19 | power supply | supply ground |
| LED10 | 20 | 0 | LED driver 10 |
| LED11 | 21 | 0 | LED driver 11 |
| LED12 | 22 | 0 | LED driver 12 |
| LED13 | 23 | 0 | LED driver 12 |
| V _{SS} | 24 | power supply | supply ground |
| LED14 | 25 | 0 | LED driver 14 |
| LED15 | 26 | 0 | LED driver 15 |
| OE | 27 | I | active LOW output enable |
| A5 | 28 | I | address input 5 |
| A6 | 29 | I | address input 6 |
| SCL | 30 | I | serial clock line |
| SDA | 31 | I/O | serial data line |
| | | | |

supply voltage

Table 3. Pin description ... continued

7. Functional description

 V_{DD}

Refer to Figure 1 "Block diagram of PCA9622".

power supply

7.1 Device addresses

32

Following a START condition, the bus master must output the address of the slave it is accessing.

There are a maximum of 128 possible programmable addresses using the 7 hardware address pins. Two of these addresses, Software Reset and LED All Call, cannot be used because their default power-up state is ON, leaving a maximum of 126 addresses. Using other reserved addresses, as well as any other Sub Call address, reduces the total number of possible addresses even further.

7.1.1 Regular I²C-bus slave address

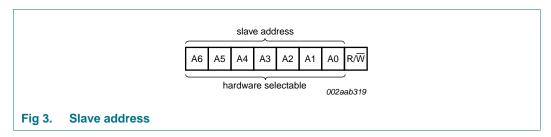
The I²C-bus slave address of the PCA9622 is shown in <u>Figure 3</u>. To conserve power, no internal pull-up resistors are incorporated on the hardware selectable address pins and they must be pulled HIGH or LOW.

Remark: Using reserved I²C-bus addresses will interfere with other devices, but only if the devices are on the bus and/or the bus will be open to other I²C-bus systems at some later date. In a closed system where the designer controls the address assignment these addresses can be used since the PCA9622 treats them like any other address. The LED All Call, Software Rest and PCA9564 or PCA9665 slave address (if on the bus) can never be used for individual device addresses.

- PCA9622 LED All Call address (1110 000) and Software Reset (0000 0110) which are active on start-up
- PCA9564 (0000 000) or PCA9665 (1110 000) slave address which is active on start-up

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

- 'reserved for future use' I2C-bus addresses (0000 011, 1111 1XX)
- slave devices that use the 10-bit addressing scheme (1111 0XX)
- slave devices that are designed to respond to the General Call address (0000 000)
- High-speed mode (Hs-mode) master code (0000 1XX)



The last bit of the address byte defines the operation to be performed. When set to logic 1 a read is selected, while a logic 0 selects a write operation.

7.1.2 LED All Call I²C-bus address

- Default power-up value (ALLCALLADR register): E0h or 1110 000
- Programmable through I²C-bus (volatile programming)
- At power-up, LED All Call I²C-bus address is enabled. PCA9622 sends an ACK when E0h (R/W = 0) or E1h (R/W = 1) is sent by the master.

See Section 7.3.8 "ALLCALLADR, LED All Call I2C-bus address" for more detail.

Remark: The default LED All Call I²C-bus address (E0h or 1110 000) must not be used as a regular I²C-bus slave address since this address is enabled at power-up. All the PCA9622s on the I²C-bus acknowledge the address if sent by the I²C-bus master.

7.1.3 LED Sub Call I²C-bus addresses

- 3 different I2C-bus addresses can be used
- Default power-up values:
 - SUBADR1 register: E2h or 1110 001
 - SUBADR2 register: E4h or 1110 010
 - SUBADR3 register: E8h or 1110 100
- Programmable through I²C-bus (volatile programming)
- At power-up, Sub Call I²C-bus addresses are disabled. PCA9622 does not send an ACK when E2h (R/W = 0) or E3h (R/W = 1), E4h (R/W = 0) or E5h (R/W = 1), or E8h (R/W = 0) or E9h (R/W = 1) is sent by the master.

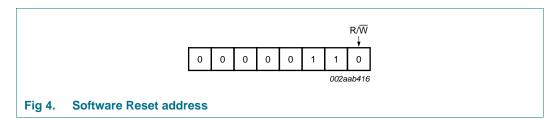
See Section 7.3.7 "SUBADR1 to SUBADR3, I2C-bus subaddress 1 to 3" for more detail.

Remark: The default LED Sub Call I²C-bus addresses may be used as regular I²C-bus slave addresses as long as they are disabled.

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.1.4 Software Reset I²C-bus address

The address shown in Figure 4 is used when a reset of the PCA9622 must be performed by the master. The Software Reset address (SWRST Call) must be used with $R/\overline{W} = logic 0$. If $R/\overline{W} = logic 1$, the PCA9622 does not acknowledge the SWRST. See Section 7.6 "Software reset" for more detail.

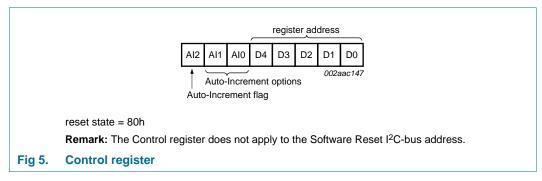


Remark: The Software Reset I²C-bus address is a reserved address and cannot be used as a regular I²C-bus slave address or as an LED All Call or LED Sub Call address.

7.2 Control register

Following the successful acknowledgement of the slave address, LED All Call address or LED Sub Call address, the bus master sends a byte to the PCA9622, which is stored in the Control register.

The lowest 5 bits are used as a pointer to determine which register is accessed (D[4:0]). The highest 3 bits are used as Auto-Increment flag and Auto-Increment options (AI[2:0]).



When the Auto-Increment flag is set (AI2 = logic 1), the five low-order bits of the Control register are automatically incremented after a read or write. This allows the user to program the registers sequentially. Four different types of Auto-Increment are possible, depending on AI1 and AI0 values.

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

Table 4. Auto-Increment options

| AI2 | Al1 | AI0 | Function |
|-----|-----|-----|---|
| 0 | 0 | 0 | no Auto-Increment |
| 1 | 0 | 0 | Auto-Increment for all registers. D[4:0] roll over to '0 0000' after the last register (1 1011) is accessed. |
| 1 | 0 | 1 | Auto-Increment for individual brightness registers only. D[4:0] roll over to '0 0010' after the last register (1 0001) is accessed. |
| 1 | 1 | 0 | Auto-Increment for global control registers only. D[4:0] roll over to '1 0010' after the last register (1 0011) is accessed. |
| 1 | 1 | 1 | Auto-Increment for individual and global control registers only. D[4:0] roll over to '0 0010' after the last register (1 0011) is accessed. |

Remark: Other combinations not shown in $\underline{\text{Table 4}}$ (AI[2:0] = 001, 010, and 011) are reserved and must not be used for proper device operation.

AI[2:0] = 000 is used when the same register must be accessed several times during a single I^2C -bus communication, for example, changes the brightness of a single LED. Data is overwritten each time the register is accessed during a write operation.

AI[2:0] = 100 is used when all the registers must be sequentially accessed, for example, power-up programming.

AI[2:0] = 101 is used when the 16 LED drivers must be individually programmed with different values during the same I²C-bus communication, for example, changing color setting to another color setting.

AI[2:0] = 110 is used when the LED drivers must be globally programmed with different settings during the same I²C-bus communication, for example, global brightness or blinking change.

AI[2:0] = 111 is used when individual and global changes must be performed during the same I²C-bus communication, for example, changing a color and global brightness at the same time.

Only the 5 least significant bits D[4:0] are affected by the Al[2:0] bits.

When the Control register is written, the register entry point determined by D[4:0] is the first register that is addressed (read or write operation), and can be anywhere between 0 0000 and 1 1011 (as defined in Table 5). When Al[2] = 1, the Auto-Increment flag is set and the rollover value at which the register increment stops and goes to the next one is determined by Al[2:0]. See Table 4 for rollover values. For example, if the Control register = 1111 0100 (F4h), then the register addressing sequence is (in hexadecimal): $14 \rightarrow ... \rightarrow 1B \rightarrow 00 \rightarrow ... \rightarrow 13 \rightarrow 02 \rightarrow ... \rightarrow 13 \rightarrow 02 \rightarrow ... \rightarrow 13 \rightarrow 02 \rightarrow ...$ as long as the master keeps sending or reading data.

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

7.3 Register definitions

Table 5. Register summary[1]

| Register number | D4 | D3 | D2 | D1 | D0 | Name | Туре | Function |
|-----------------|----|----|----|----|----|------------|------------|---|
| 00h | 0 | 0 | 0 | 0 | 0 | MODE1 | read/write | Mode register 1 |
| 01h | 0 | 0 | 0 | 0 | 1 | MODE2 | read/write | Mode register 2 |
| 02h | 0 | 0 | 0 | 1 | 0 | PWM0 | read/write | brightness control LED0 |
| 03h | 0 | 0 | 0 | 1 | 1 | PWM1 | read/write | brightness control LED1 |
| 04h | 0 | 0 | 1 | 0 | 0 | PWM2 | read/write | brightness control LED2 |
| 05h | 0 | 0 | 1 | 0 | 1 | PWM3 | read/write | brightness control LED3 |
| 06h | 0 | 0 | 1 | 1 | 0 | PWM4 | read/write | brightness control LED4 |
| 07h | 0 | 0 | 1 | 1 | 1 | PWM5 | read/write | brightness control LED5 |
| 08h | 0 | 1 | 0 | 0 | 0 | PWM6 | read/write | brightness control LED6 |
| 09h | 0 | 1 | 0 | 0 | 1 | PWM7 | read/write | brightness control LED7 |
| 0Ah | 0 | 1 | 0 | 1 | 0 | PWM8 | read/write | brightness control LED8 |
| 0Bh | 0 | 1 | 0 | 1 | 1 | PWM9 | read/write | brightness control LED9 |
| 0Ch | 0 | 1 | 1 | 0 | 0 | PWM10 | read/write | brightness control LED10 |
| 0Dh | 0 | 1 | 1 | 0 | 1 | PWM11 | read/write | brightness control LED11 |
| 0Eh | 0 | 1 | 1 | 1 | 0 | PWM12 | read/write | brightness control LED12 |
| 0Fh | 0 | 1 | 1 | 1 | 1 | PWM13 | read/write | brightness control LED13 |
| 10h | 1 | 0 | 0 | 0 | 0 | PWM14 | read/write | brightness control LED14 |
| 11h | 1 | 0 | 0 | 0 | 1 | PWM15 | read/write | brightness control LED15 |
| 12h | 1 | 0 | 0 | 1 | 0 | GRPPWM | read/write | group duty cycle control |
| 13h | 1 | 0 | 0 | 1 | 1 | GRPFREQ | read/write | group frequency |
| 14h | 1 | 0 | 1 | 0 | 0 | LEDOUT0 | read/write | LED output state 0 |
| 15h | 1 | 0 | 1 | 0 | 1 | LEDOUT1 | read/write | LED output state 1 |
| 16h | 1 | 0 | 1 | 1 | 0 | LEDOUT2 | read/write | LED output state 2 |
| 17h | 1 | 0 | 1 | 1 | 1 | LEDOUT3 | read/write | LED output state 3 |
| 18h | 1 | 1 | 0 | 0 | 0 | SUBADR1 | read/write | I ² C-bus subaddress 1 |
| 19h | 1 | 1 | 0 | 0 | 1 | SUBADR2 | read/write | I ² C-bus subaddress 2 |
| 1Ah | 1 | 1 | 0 | 1 | 0 | SUBADR3 | read/write | I ² C-bus subaddress 3 |
| 1Bh | 1 | 1 | 0 | 1 | 1 | ALLCALLADR | read/write | LED All Call I ² C-bus address |

^[1] Only $D[4:0] = 0\,0000$ to 1 1011 are allowed and are acknowledged. $D[4:0] = 1\,1100$ to 1 1111 are reserved and are not acknowledged.

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.3.1 Mode register 1, MODE1

Table 6. MODE1 - Mode register 1 (address 00h) bit description Legend: * default value.

| Bit | Symbol | Access | Value | Description | | |
|-----|----------|-----------|-------|--|--|-----------------------------------|
| 7 | Al2 | read only | | read only 0 Register Auto-Increr | | Register Auto-Increment disabled. |
| | | | 1* | Register Auto-Increment enabled. | | |
| 6 | Al1 | read only | 0* | Auto-Increment bit 1 = 0. | | |
| | | | 1 | Auto-Increment bit 1 = 1. | | |
| 5 | AI0 | read only | 0* | Auto-Increment bit 0 = 0. | | |
| | | | 1 | Auto-Increment bit 0 = 1. | | |
| 4 | SLEEP[1] | R/W | 0 | Normal mode ^[2] . | | |
| | | | 1* | Low-power mode. Oscillator off[3]. | | |
| 3 | SUB1 | R/W | 0* | PCA9622 does not respond to I ² C-bus subaddress 1. | | |
| | | | 1 | PCA9622 responds to I ² C-bus subaddress 1. | | |
| 2 | SUB2 | R/W | 0* | PCA9622 does not respond to I ² C-bus subaddress 2. | | |
| | | | 1 | PCA9622 responds to I ² C-bus subaddress 2. | | |
| 1 | SUB3 | R/W | 0* | PCA9622 does not respond to I ² C-bus subaddress 3. | | |
| | | | 1 | PCA9622 responds to I ² C-bus subaddress 3. | | |
| 0 | ALLCALL | R/W | 0 | PCA9622 does not respond to LED All Call I ² C-bus address. | | |
| | | | 1* | PCA9622 responds to LED All Call I ² C-bus address. | | |

^[1] Bit 4 must be programmed with logic 0 for proper device operation.

7.3.2 Mode register 2, MODE2

Table 7. MODE2 - Mode register 2 (address 01h) bit description Legend: * default value.

| Bit | Symbol | Access | Value | Description | | |
|-----|--------|-----------|-------|---|--|--|
| 7 | - | read only | 0* | reserved | | |
| 6 | - | read only | 0* | reserved | | |
| 5 | DMBLNK | R/W | 0* | group control = dimming | | |
| | | | 1 | group control = blinking | | |
| 4 | INVRT | R/W | 0* | reserved; write must always be a logic 0 | | |
| 3 | OCH | R/W | 0* | outputs change on STOP command ^[1] | | |
| | | | 1 | outputs change on ACK | | |
| 2 | - | R/W | 1* | reserved; write must always be a logic 1[2] | | |
| 1 | - | R/W | 0* | reserved; write must always be a logic 0[2] | | |
| 0 | - | R/W | 1* | reserved; write must always be a logic 12 | | |

^[1] Change of the outputs at the STOP command allows synchronizing outputs of more than one PCA9622. Applicable to registers from 02h (PWM0) to 17h (LEDOUT) only.

^[2] It takes $500 \mu s$ max. for the oscillator to be up and running once SLEEP bit has been set to logic 0. Timings on LEDn outputs are not guaranteed if PWMx, GRPPWM or GRPFREQ registers are accessed within the $500 \mu s$ window.

^[3] No blinking or dimming is possible when the oscillator is off.

^[2] Remark: If you change these bits from their default values, the device will not perform as expected.

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.3.3 PWM0 to PWM15, individual brightness control

Table 8. PWM0 to PWM15 - PWM registers 0 to 15 (address 02h to 11h) bit description Legend: * default value.

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|----------|-----|------------|--------|------------|-----------------------------|
| 02h | PWM0 | 7:0 | IDC0[7:0] | R/W | 0000 0000* | PWM0 Individual Duty Cycle |
| 03h | PWM1 | 7:0 | IDC1[7:0] | R/W | 0000 0000* | PWM1 Individual Duty Cycle |
| 04h | PWM2 | 7:0 | IDC2[7:0] | R/W | 0000 0000* | PWM2 Individual Duty Cycle |
| 05h | PWM3 | 7:0 | IDC3[7:0] | R/W | 0000 0000* | PWM3 Individual Duty Cycle |
| 06h | PWM4 | 7:0 | IDC4[7:0] | R/W | 0000 0000* | PWM4 Individual Duty Cycle |
| 07h | PWM5 | 7:0 | IDC5[7:0] | R/W | 0000 0000* | PWM5 Individual Duty Cycle |
| 08h | PWM6 | 7:0 | IDC6[7:0] | R/W | 0000 0000* | PWM6 Individual Duty Cycle |
| 09h | PWM7 | 7:0 | IDC7[7:0] | R/W | 0000 0000* | PWM7 Individual Duty Cycle |
| 0Ah | PWM8 | 7:0 | IDC8[7:0] | R/W | 0000 0000* | PWM8 Individual Duty Cycle |
| 0Bh | PWM9 | 7:0 | IDC9[7:0] | R/W | 0000 0000* | PWM9 Individual Duty Cycle |
| 0Ch | PWM10 | 7:0 | IDC10[7:0] | R/W | 0000 0000* | PWM10 Individual Duty Cycle |
| 0Dh | PWM11 | 7:0 | IDC11[7:0] | R/W | 0000 0000* | PWM11 Individual Duty Cycle |
| 0Eh | PWM12 | 7:0 | IDC12[7:0] | R/W | 0000 0000* | PWM12 Individual Duty Cycle |
| 0Fh | PWM13 | 7:0 | IDC13[7:0] | R/W | 0000 0000* | PWM13 Individual Duty Cycle |
| 10h | PWM14 | 7:0 | IDC14[7:0] | R/W | 0000 0000* | PWM14 Individual Duty Cycle |
| 11h | PWM15 | 7:0 | IDC15[7:0] | R/W | 0000 0000* | PWM15 Individual Duty Cycle |

A 97 kHz fixed frequency signal is used for each output. Duty cycle is controlled through 256 linear steps from 00h (0 % duty cycle = LED output off) to FFh (99.6 % duty cycle = LED output at maximum brightness). Applicable to LED outputs programmed with LDRx = 10 or 11 (LEDOUT0 to LEDOUT3 registers).

$$duty\ cycle = \frac{IDCx[7:0]}{256} \tag{1}$$

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.3.4 GRPPWM, group duty cycle control

Table 9. GRPPWM - Group brightness control register (address 12h) bit description Legend: * default value

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|----------|-----|----------|--------|-----------|-----------------|
| 12h | GRPPWM | 7:0 | GDC[7:0] | R/W | 1111 1111 | GRPPWM register |

When DMBLNK bit (MODE2 register) is programmed with logic 0, a 190 Hz fixed frequency signal is superimposed with the 97 kHz individual brightness control signal. GRPPWM is then used as a global brightness control allowing the LED outputs to be dimmed with the same value. The value in GRPFREQ is then a 'Don't care'.

General brightness for the 16 outputs is controlled through 256 linear steps from 00h (0 % duty cycle = LED output off) to FFh (99.6 % duty cycle = maximum brightness). Applicable to LED outputs programmed with LDRx = 11 (LEDOUT0 to LEDOUT3 registers).

When DMBLNK bit is programmed with logic 1, GRPPWM and GRPFREQ registers define a global blinking pattern, where GRPFREQ contains the blinking period (from 24 Hz to 10.73 s) and GRPPWM the duty cycle (ON/OFF ratio in %).

$$duty\ cycle = \frac{GDC[7:0]}{256} \tag{2}$$

7.3.5 GRPFREQ, group frequency

Table 10. GRPFREQ - Group Frequency register (address 13h) bit description Legend: * default value.

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|----------|-----|-----------|--------|------------|------------------|
| 13h | GRPFREQ | 7:0 | GFRQ[7:0] | R/W | 0000 0000* | GRPFREQ register |

GRPFREQ is used to program the global blinking period when DMBLNK bit (MODE2 register) is equal to 1. Value in this register is a 'Don't care' when DMBLNK = 0. Applicable to LED outputs programmed with LDRx = 11 (LEDOUT0 to LEDOUT3 registers).

Blinking period is controlled through 256 linear steps from 00h (41 ms, frequency 24 Hz) to FFh (10.73 s).

global blinking period =
$$\frac{GFRQ[7:0] + 1}{24}(s)$$
 (3)

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.3.6 LEDOUT0 to LEDOUT3, LED driver output state

Table 11. LEDOUT0 to LEDOUT3 - LED driver output state register (address 14h to 17h) bit description

Legend: * default value.

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|----------|-----|--------|--------|-------|----------------------------|
| 14h | LEDOUT0 | 7:6 | LDR3 | R/W | 00* | LED3 output state control |
| | | 5:4 | LDR2 | R/W | 00* | LED2 output state control |
| | | 3:2 | LDR1 | R/W | 00* | LED1 output state control |
| | | 1:0 | LDR0 | R/W | 00* | LED0 output state control |
| 15h | LEDOUT1 | 7:6 | LDR7 | R/W | 00* | LED7 output state control |
| | | 5:4 | LDR6 | R/W | 00* | LED6 output state control |
| | | 3:2 | LDR5 | R/W | 00* | LED5 output state control |
| | | 1:0 | LDR4 | R/W | 00* | LED4 output state control |
| 16h | LEDOUT2 | 7:6 | LDR11 | R/W | 00* | LED11 output state control |
| | | 5:4 | LDR10 | R/W | 00* | LED10 output state control |
| | | 3:2 | LDR9 | R/W | 00* | LED9 output state control |
| | | 1:0 | LDR8 | R/W | 00* | LED8 output state control |
| 17h | LEDOUT3 | 7:6 | LDR15 | R/W | 00* | LED15 output state control |
| | | 5:4 | LDR14 | R/W | 00* | LED14 output state control |
| | | 3:2 | LDR13 | R/W | 00* | LED13 output state control |
| | | 1:0 | LDR12 | R/W | 00* | LED12 output state control |

LDRx = 00 — LED driver x is off (default power-up state).

LDRx = 01 — LED driver x is fully on (individual brightness and group dimming/blinking not controlled).

LDRx = 10 — LED driver x individual brightness can be controlled through its PWMx register.

LDRx = 11 — LED driver x individual brightness and group dimming/blinking can be controlled through its PWMx register and the GRPPWM registers.

7.3.7 SUBADR1 to SUBADR3, I²C-bus subaddress 1 to 3

Table 12. SUBADR1 to SUBADR3 - I²C-bus subaddress registers 0 to 3 (address 18h to 1Ah) bit description

Legend: * default value.

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|----------|-----|---------|--------|-----------|-----------------------------------|
| 18h | SUBADR1 | 7:1 | A1[7:1] | R/W | 1110 001* | I ² C-bus subaddress 1 |
| | | 0 | A1[0] | R only | 0* | reserved |
| 19h | SUBADR2 | 7:1 | A2[7:1] | R/W | 1110 010* | I ² C-bus subaddress 2 |
| | | 0 | A2[0] | R only | 0* | reserved |
| 1Ah | SUBADR3 | 7:1 | A3[7:1] | R/W | 1110 100* | I ² C-bus subaddress 3 |
| | | 0 | A3[0] | R only | 0* | reserved |

Subaddresses are programmable through the I²C-bus. Default power-up values are E2h, E4h, E8h, and the device(s) will not acknowledge these addresses right after power-up (the corresponding SUBx bit in MODE1 register is equal to 0).

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

Once subaddresses have been programmed to their right values, SUBx bits must be set to logic 1 in order to have the device acknowledging these addresses (MODE1 register).

Only the 7 MSBs representing the I²C-bus subaddress are valid. The LSB in SUBADRx register is a read-only bit (0).

When SUBx is set to logic 1, the corresponding I²C-bus subaddress can be used during either an I²C-bus read or write sequence.

7.3.8 ALLCALLADR, LED All Call I²C-bus address

Table 13. ALLCALLADR - LED All Call I²C-bus address register (address 1Bh) bit description

Legend: * default value.

| Address | Register | Bit | Symbol | Access | Value | Description |
|---------|------------|-----|---------|--------|-------|---|
| 1Bh | ALLCALLADR | 7:1 | AC[7:1] | R/W | | ALLCALL I ² C-bus address register |
| | | 0 | AC[0] | R only | 0* | reserved |

The LED All Call I²C-bus address allows all the PCA9622s on the bus to be programmed at the same time (ALLCALL bit in register MODE1 must be equal to 1 (power-up default state)). This address is programmable through the I²C-bus and can be used during either an I²C-bus read or write sequence. The register address can also be programmed as a Sub Call.

Only the 7 MSBs representing the All Call I²C-bus address are valid. The LSB in ALLCALLADR register is a read-only bit (0).

If ALLCALL bit = 0, the device does not acknowledge the address programmed in register ALLCALLADR.

7.4 Active LOW output enable input

The active LOW output enable (\overline{OE}) pin, allows enabling or disabling all the LED outputs at the same time.

- When a LOW level is applied to \overline{OE} pin, all the LED outputs are enabled.
- When a HIGH level is applied to \overline{OE} pin, all the LED outputs are high-impedance.

The OE pin can be used as a synchronization signal to switch on/off several PCA9622 devices at the same time. This requires an external clock reference that provides blinking period and the duty cycle.

The OE pin can also be used as an external dimming control signal. The frequency of the external clock must be high enough not to be seen by the human eye, and the duty cycle value determines the brightness of the LEDs.

Remark: Do not use \overline{OE} as an external blinking control signal when internal global blinking is selected (DMBLNK = 1, MODE2 register) since it results in an undefined blinking pattern. Do not use \overline{OE} as an external dimming control signal when internal global dimming is selected (DMBLNK = 0, MODE2 register) since it results in an undefined dimming pattern.

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

Remark: During power-down, slow decay of voltage supplies may keep LEDs illuminated. Consider disabling LED outputs using HIGH level applied to \overline{OE} pin.

7.5 Power-on reset

When power is applied to V_{DD} , an internal power-on reset holds the PCA9622 in a reset condition until V_{DD} has reached V_{POR} . At this point, the reset condition is released and the PCA9622 registers and I^2C -bus state machine are initialized to their default states (all zeroes) causing all the channels to be deselected. Thereafter, V_{DD} must be lowered below 0.2 V to reset the device.

7.6 Software reset

The Software Reset Call (SWRST Call) allows all the devices in the I^2C -bus to be reset to the power-up state value through a specific formatted I^2C -bus command. To be performed correctly, it implies that the I^2C -bus is functional and that there is no device hanging the bus.

The SWRST Call function is defined as the following:

- 1. A START command is sent by the I²C-bus master.
- 2. The reserved SWRST I²C-bus address '0000 011' with the R/\overline{W} bit set to '0' (write) is sent by the I²C-bus master.
- 3. The PCA9622 device(s) acknowledge(s) after seeing the SWRST Call address '0000 0110' (06h) only. If the R/W bit is set to '1' (read), no acknowledge is returned to the I²C-bus master.
- 4. Once the SWRST Call address has been sent and acknowledged, the master sends 2 bytes with 2 specific values (SWRST data byte 1 and byte 2):
 - a. Byte 1 = A5h: the PCA9622 acknowledges this value only. If byte 1 is not equal to A5h, the PCA9622 does not acknowledge it.
 - b. Byte 2 = 5Ah: the PCA9622 acknowledges this value only. If byte 2 is not equal to 5Ah, then the PCA9622 does not acknowledge it.

If more than 2 bytes of data are sent, the PCA9622 does not acknowledge any more.

5. Once the right 2 bytes (SWRST data byte 1 and byte 2 only) have been sent and correctly acknowledged, the master sends a STOP command to end the SWRST Call: the PCA9622 then resets to the default value (power-up value) and is ready to be addressed again within the specified bus free time (t_{BUF}).

The I²C-bus master must interpret a non-acknowledge from the PCA9622 (at any time) as a 'SWRST Call Abort'. The PCA9622 does not initiate a reset of its registers. This happens only when the format of the SWRST Call sequence is not correct.

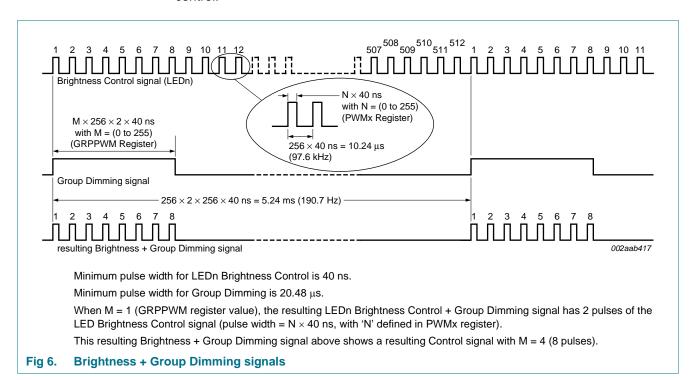
16-bit Fm+ I2C-bus 100 mA 40 V LED driver

7.7 Individual brightness control with group dimming/blinking

A 97 kHz fixed frequency signal with programmable duty cycle (8 bits, 256 steps) is used to control individually the brightness for each LED.

On top of this signal, one of the following signals can be superimposed (this signal can be applied to the 4 LED outputs):

- A lower 190 Hz fixed frequency signal with programmable duty cycle (8 bits, 256 steps) is used to provide a global brightness control.
- A programmable frequency signal from 24 Hz to ¹/_{10.73} Hz (8 bits, 256 steps) with programmable duty cycle (8 bits, 256 steps) is used to provide a global blinking control.



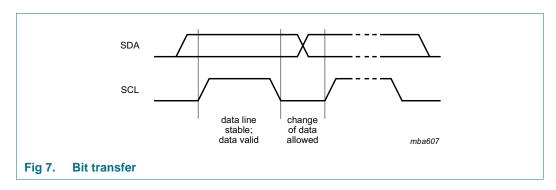
16-bit Fm+ I²C-bus 100 mA 40 V LED driver

8. Characteristics of the I²C-bus

The I²C-bus is for 2-way, 2-line communication between different ICs or modules. The two lines are a serial data line (SDA) and a serial clock line (SCL). Both lines must be connected to a positive supply via a pull-up resistor when connected to the output stages of a device. Data transfer may be initiated only when the bus is not busy.

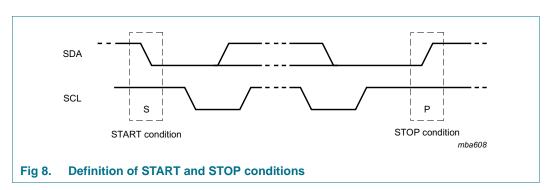
8.1 Bit transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable during the HIGH period of the clock pulse as changes in the data line at this time are interpreted as control signals (see Figure 7).



8.1.1 START and STOP conditions

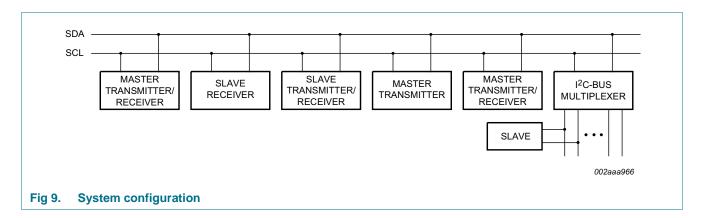
Both data and clock lines remain HIGH when the bus is not busy. A HIGH-to-LOW transition of the data line while the clock is HIGH is defined as the START condition (S). A LOW-to-HIGH transition of the data line while the clock is HIGH is defined as the STOP condition (P) (see <u>Figure 8</u>).



8.2 System configuration

A device generating a message is a 'transmitter'; a device receiving is the 'receiver'. The device that controls the message is the 'master' and the devices which are controlled by the master are the 'slaves' (see Figure 9).

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

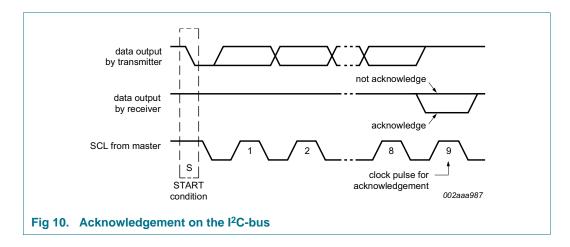


8.3 Acknowledge

The number of data bytes transferred between the START and the STOP conditions from transmitter to receiver is not limited. Each byte of 8 bits is followed by one acknowledge bit. The acknowledge bit is a HIGH level put on the bus by the transmitter, whereas the master generates an extra acknowledge related clock pulse.

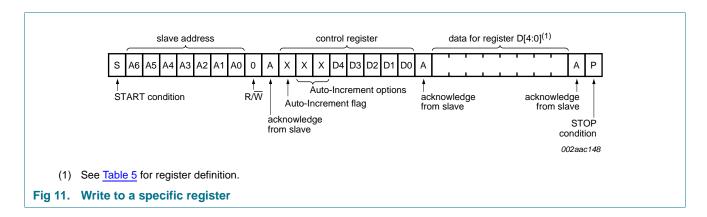
A slave receiver which is addressed must generate an acknowledge after the reception of each byte. Also a master must generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter. The device that acknowledges has to pull down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse; set-up time and hold time must be taken into account.

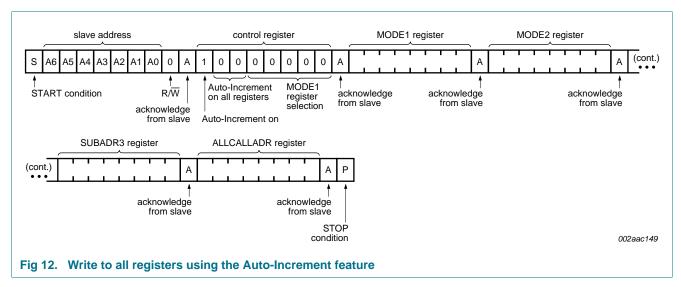
A master receiver must signal an end of data to the transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this event, the transmitter must leave the data line HIGH to enable the master to generate a STOP condition.

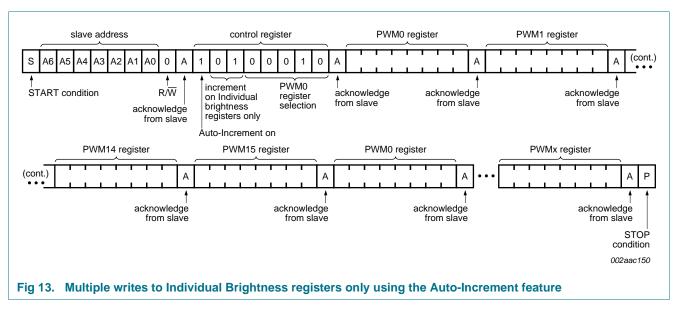


16-bit Fm+ I2C-bus 100 mA 40 V LED driver

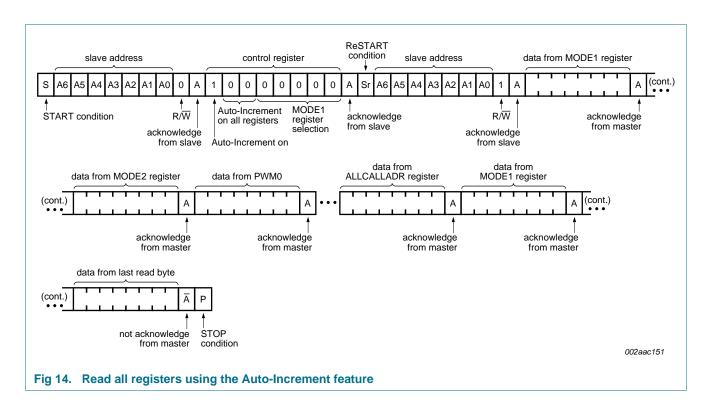
9. Bus transactions

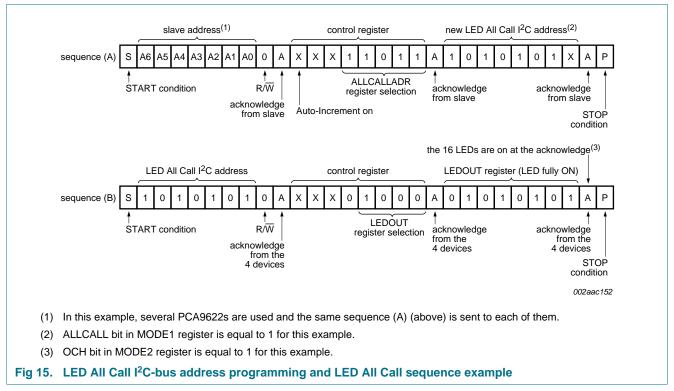






16-bit Fm+ I2C-bus 100 mA 40 V LED driver

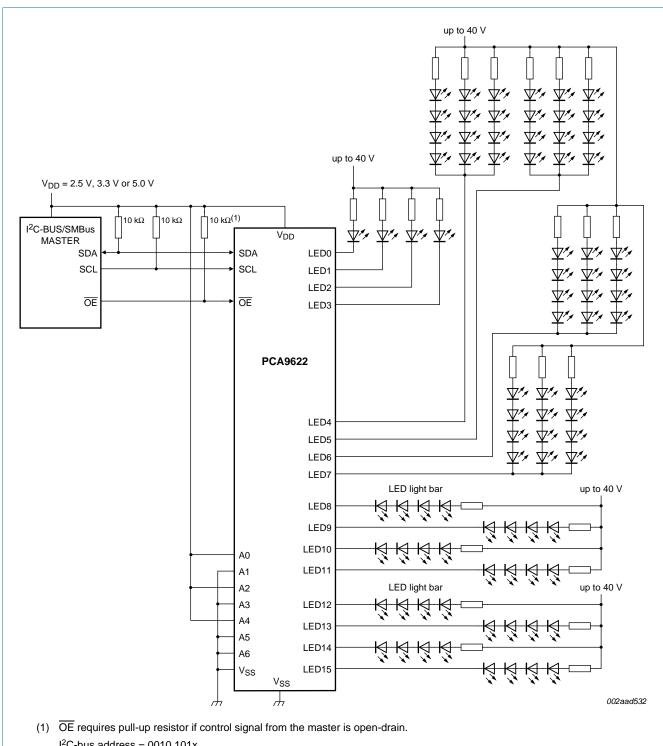




PCA9622 NXP Semiconductors

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

10. Application design-in information



 I^2 C-bus address = 0010 101x.

 $\textbf{Remark:} \ \, \text{During power-down}, \ \, \text{slow decay of voltage supplies may keep LEDs illuminated.} \ \, \text{Consider disabling LED outputs using HIGH level applied to } \, \overline{\text{OE}} \, \text{pin.}$

Fig 16. Typical application

Product data sheet

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16-bit Fm+ I2C-bus 100 mA 40 V LED driver

10.1 Junction temperature calculation

A device junction temperature can be calculated when the ambient temperature or the case temperature is known.

When the ambient temperature is known, the junction temperature is calculated using Equation 4 and the ambient temperature, junction to ambient thermal resistance and power dissipation.

$$T_i = T_{amb} + R_{th(i-a)} \times P_{tot} \tag{4}$$

where:

T_i = junction temperature

 T_{amb} = ambient temperature

R_{th(i-a)} = junction to ambient thermal resistance

 P_{tot} = (device) total power dissipation

When the case temperature is known, the junction temperature is calculated using <u>Equation 5</u> and the case temperature, junction to case thermal resistance and power dissipation.

$$T_{j} = T_{case} + R_{th(j-c)} \times P_{tot}$$
 (5)

where:

T_i = junction temperature

 T_{case} = case temperature

 $R_{th(i-c)}$ = junction to case thermal resistance

 $P_{tot} = (device) total power dissipation$

Here are two examples regarding how to calculate the junction temperature using junction to case and junction to ambient thermal resistance. In the first example (Section 10.1.1), given the operating condition and the junction to ambient thermal resistance, the junction temperature of PCA9622DR, in the TSSOP32 package, is calculated for a system operating condition in 50 °C¹ ambient temperature. In the second example (Section 10.1.2), based on a specific customer application requirement where only the case temperature is known, applying the junction to case thermal resistance equation, the junction temperature of the PCA9622DR, in the TSSOP32 package, is calculated.

PCA9622

 ^{50 °}C is a typical temperature inside an enclosed system. The designers should feel free, as needed, to perform their own calculation using the examples.

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

10.1.1 Example 1: T_j calculation of PCA9622DR, in TSSOP32 package, when T_{amb} is known

 $R_{th(i-a)} = 83 \, ^{\circ}C/W$

 $T_{amb} = 50 \, ^{\circ}C$

LED output low voltage (LED V_{OL}) = 0.5 V

LED output current per channel = 80 mA

Number of outputs = 16

 $I_{DD(max)} = 12 \text{ mA}$

 $V_{DD(max)} = 5.5 \text{ V}$

I²C-bus clock (SCL) maximum sink current = 25 mA

I²C-bus data (SDA) maximum sink current = 25 mA

- 1. Find Ptot (device total power dissipation):
 - output total power = $80 \text{ mA} \times 16 \times 0.5 \text{ V} = 640 \text{ mW}$
 - chip core power consumption = 12 mA \times 5.5 V = 66 mW
 - SCL power dissipation = 25 mA × 0.4 V = 10 mW
 - SDA power dissipation = 25 mA \times 0.4 V = 10 mW

$$P_{tot} = (640 + 66 + 10 + 10) \text{ mW} = 726 \text{ mW}$$

2. Find T_i (junction temperature):

$$T_i = (T_{amb} + R_{th(i-a)} \times P_{tot}) = (50 \text{ °C} + 83 \text{ °C/W} \times 726 \text{ mW}) = 110.26 \text{ °C}$$

10.1.2 Example 2: T_i calculation where only T_{case} is known

This example uses a customer's specific application of the PCA9622DR, 16-channel LED controller in the TSSOP32 package, where only the case temperature (T_{case}) is known.

$$T_i = T_{case} + R_{th(i-c)} \times P_{tot}$$
, where:

 $R_{th(j-c)} = 23 \, ^{\circ}C/W$

T_{case} (measured) = 94.6 °C

V_{OL} of LED ~ 0.5 V

 $I_{DD(max)} = 12 \text{ mA}$

 $V_{DD(max)} = 5.5 V$

LED output voltage LOW = 0.5 V

LED output current:

60 mA on 1 port = $(60 \text{ mA} \times 1)$

50 mA on 6 ports = $(50 \text{ mA} \times 6)$

40 mA on 2 ports = $(40 \text{ mA} \times 2)$

20 mA on 7 ports = $(20 \text{ mA} \times 7)$

I²C-bus maximum sink current on clock line = 25 mA

I²C-bus maximum sink current on data line = 25 mA

16-bit Fm+ I2C-bus 100 mA 40 V LED driver

- 1. Find P_{tot} (device total power dissipation)
 - output current (60 mA \times 1 port); output power (60 mA \times 1 \times 0.5 V) = 30 mW
 - output current (50 mA \times 6 ports); output power (50 mA \times 6 \times 0.5 V) = 150 mW
 - output current (40 mA \times 2 ports); output power (40 mA \times 2 \times 0.5 V) = 40 mW
 - output current (20 mA \times 7 ports); output power (20 mA \times 7 \times 0.5 V) = 70 mW

Output total power = 290 mW

- chip core power consumption = 12 mA \times 5.5 V = 66 mW
- SCL power dissipation = 25 mA × 0.4 V = 10 mW
- SDA power dissipation = 25 mA \times 0.4 V = 10 mW

P_{tot} (device total power dissipation) = 376 mW

2. Find T_i (junction temperature):

$$T_j = T_{case} + R_{th(j-a)} \times P_{tot} = 94.6 \text{ °C} + 23 \text{ °C/W} \times 376 \text{ mW} = 103.25 \text{ °C}$$

11. Limiting values

Table 14. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

| Symbol | Parameter | Conditions | Min | Max | Unit |
|-----------------------|--------------------------------|------------------------------|-----------------------|------|------|
| V_{DD} | supply voltage | | -0.5 | +6.0 | V |
| V _{I/O} | voltage on an input/output pin | | V _{SS} - 0.5 | 5.5 | V |
| V _{drv(LED)} | LED driver voltage | | V _{SS} - 0.5 | 40 | V |
| I _{O(LEDn)} | output current on pin LEDn | | - | 100 | mA |
| I _{OL(tot)} | total LOW-level output current | $V_{OL} = 0.5 \text{ V}$ [1] | 1600 | - | mA |
| I _{SS} | ground supply current | per V _{SS} pin | - | 800 | mA |
| P _{tot} | total power dissipation | T _{amb} = 25 °C | - | 1.8 | W |
| | | T _{amb} = 85 °C | - | 0.72 | W |
| P/ch | power dissipation per channel | T _{amb} = 25 °C | - | 100 | mW |
| | | T _{amb} = 85 °C | - | 45 | mW |
| Tj | junction temperature | [2] | - | 125 | °C |
| T _{stg} | storage temperature | | -65 | +150 | °C |
| T _{amb} | ambient temperature | operating | -40 | +85 | °C |

^[1] Each bit must be limited to a maximum of 100 mA and the total package limited to 1600 mA due to internal busing limits.

[2] Refer to Section 10.1 for junction temperature calculation.

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

Table 15. TSSOP32 power dissipation and output current capability

| Measurement | TSSOP32 | | | | |
|---|---|--|--|--|--|
| T _{amb} = 25 °C | | | | | |
| maximum power dissipation (chip + output drivers) | 1200 mW | | | | |
| maximum power dissipation (output drivers only) | 1110 mW | | | | |
| maximum drive current per channel | $<\frac{1110 \ mW}{16-bit \times 0.5 \ V} = 138.8 \ mA \ [1]$ | | | | |
| T _{amb} = 60 °C | | | | | |
| maximum power dissipation (chip + output drivers) | 723 mW | | | | |
| maximum power dissipation (output drivers only) | 637 mW | | | | |
| maximum drive current per channel | $<\frac{637 \ mW}{16-bit \times 0.5 \ V} = 79.6 \ mA$ | | | | |
| T _{amb} = 80 °C | | | | | |
| maximum power dissipation (chip + output drivers) | 542 mW | | | | |
| maximum power dissipation (output drivers only) | 456 mW | | | | |
| maximum drive current per channel | $<\frac{456 \ mW}{16-bit \times 0.5 \ V} = 57 \ mA$ | | | | |

^[1] This value signifies package's ability to handle more than 100 mA per output driver. The device's maximum current rating per output is 100 mA.

12. Thermal characteristics

Table 16. Thermal characteristics

| Symbol | Parameter | Conditions | Тур | Unit |
|----------------------|---|-------------|-----|------|
| R _{th(j-a)} | thermal resistance from junction to ambient | TSSOP32 [1] | 83 | °C/W |
| R _{th(j-c)} | thermal resistance from junction to case | TSSOP32 [1] | 23 | °C/W |

^[1] Calculated in accordance with JESD 51-7.

16-bit Fm+ I²C-bus 100 mA 40 V LED driver

13. Static characteristics

Table 17. Static characteristics

 V_{DD} = 2.3 V to 5.5 V; V_{SS} = 0 V; T_{amb} = -40 °C to +85 °C; unless otherwise specified.

| VDD supply voltage 0.0 pin VDD; operating mode; no load; fSCL = 1 MHz 2.3 - 5.5 V IDD supply current on pin VDD; operating mode; no load; fSCL = 1 MHz - 0.2 4 mA VDD = 2.7 V - 0.2 4 mA VDD = 3.6 V - 8.5 12 mA ILath an pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; VDD pin VDD; no load; fSCL = 0 Hz; DDD; no load; no load; fSCL = 0 Hz; DDD; no load; no loa | Symbol | Parameter | Conditions | Min | Тур | Max | Unit |
|--|------------------|---------------------------|---|-------------|----------|---------------------|------|
| Supply current On pin V _{DD} ; operating mode; no load; f _{SCL} = 1 MHz | Supply | | | | <u> </u> | <u> </u> | |
| No load; f _{SCL} = 1 MHz V _{DD} = 2.7 V - 0.2 4 mA V _{DD} = 3.6 V - 0.2 6 mA V _{DD} = 5.5 V - 0.8.5 12 mA V _{DD} = 5.5 V - 0.8.5 12 mA V _{DD} = 2.7 V - 0.8.5 12 mA V _{DD} = 3.6 V - 0.8.5 12 mA V _{DD} = 2.7 V - 0.8.5 12 mA V _{DD} = 2.7 V - 0.8.5 12 mA V _{DD} = 2.5 V - 0.8.5 12 mA V _{DD} = 3.6 V - 0.8.5 12 mA V _{DD} = 3.6 V - 0.8.5 0.8 mA V _{DD} = 3.6 V - 0.8.2 7 μA V _{DD} = 3.6 V - 0.8.2 7 μA V _{DD} = 5.5 V - 0.8.2 7 μA V _{DD} = 5.5 V - 0.8.2 7 μA V _{DD} = 5.5 V - 0.8.2 7 μA V _{DD} = 5.5 V - 0.8.2 7 μA V _{DD} = 0.5 V - 0.8 V V _{DD} = 0.6 V V _{DD} V _{DD} | V_{DD} | supply voltage | | 2.3 | - | 5.5 | V |
| Variable | I _{DD} | supply current | • | | | | |
| Standby current Vo_D = 5.5 V - 8.5 12 mA | | | V _{DD} = 2.7 V | - | 0.2 | 4 | mA |
| Standby current Standby c | | | V _{DD} = 3.6 V | - | 2 | 6 | mA |
| | | | V _{DD} = 5.5 V | - | 8.5 | 12 | mA |
| $ V_{DD} = 3.6 \ V \\ V_{DD} = 5.5 \ V \\ V_{DD} = 5.0 \ V \\ V_{DD} = 5$ | I _{stb} | standby current | | | | | |
| V _{DD} = 5.5 V - 3.2 7 μA V _{POR} power-on reset voltage no load; V _I = V _{DD} or V _{SS} 11 - 1.70 2.0 V Input SCL; input/output SDA V _{IL} LOW-level input voltage -0.5 - +0.3V _{DD} V V _{IH} HIGH-level input voltage 0.7V _{DD} - 5.5 V IoL LOW-level output current V _{OL} = 0.4 V; V _{DD} = 2.3 V 20 - - mA I _L leakage current V _I = V _{DD} or V _{SS} -1 - +1 μA C ₁ input capacitance V _I = V _{DD} or V _{SS} -1 - +1 μA C ₁ input capacitance V _I = V _{SS} - 6 10 pF LED driver outputs V _{OL} = 0.5 V 100 - - 40 V I _O LED driver voltage 0 - 40 V I _O LED driver voltage 0 - - ±1 μA <td></td> <td></td> <td>V_{DD} = 2.7 V</td> <td>-</td> <td>1.3</td> <td>5</td> <td>μΑ</td> | | | V _{DD} = 2.7 V | - | 1.3 | 5 | μΑ |
| $\begin{array}{c ccccccccccccccccccccccccccccccccccc$ | | | V _{DD} = 3.6 V | - | 1.8 | 6 | μΑ |
| Input SCL; input/output SDA VIL LOW-level input voltage | | | V _{DD} = 5.5 V | - | 3.2 | 7 | μΑ |
| $\begin{array}{c} V_{LL} & LOW-level input voltage \\ V_{HH} & HIGH-level input voltage \\ V_{OL} & LOW-level output current \\ V_{OL} & 0.4 \ V; \ V_{DD} & 2.3 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 2.3 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.5 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 5.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 0.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 0.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 0.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 0.0 \ V \\ V_{OL} & 0.4 \ V; \ V_{DD} & 0.0 \ V \\ V_{OL} & 0.5 \ V \\ V_{OL} & 0.0 \ V_{OL} \\ V_{OL} & 0.0 \ $ | V _{POR} | power-on reset voltage | no load; $V_I = V_{DD}$ or V_{SS} | <u>l</u> - | 1.70 | 2.0 | V |
| $\begin{array}{c} V_{IH} & \text{HIGH-level input voltage} \\ V_{IOL} & \text{LOW-level output current} \\ V_{OL} = 0.4 \ V; \ V_{DD} = 2.3 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V \\ V_{OL} = 0.5 \ V_{OL} = 0.5 \ V \\ V_{OL} = 0.5 \ V_{OL} = 0.5 \ V \\ V_{OL} = 0.5 \ V_{OL} = 0.5 \ V_{OL} \\ V$ | Input SCL; | input/output SDA | | - | - | | _ |
| $\begin{array}{c} \text{MoL} \\ \text{NOL} \\ \text{LOW-level output current} \\ \text{VOL} = 0.4 \ \text{V}; \ \text{V}_{DD} = 2.3 \ \text{V} \\ \text{V}_{OL} = 0.4 \ \text{V}; \ \text{V}_{DD} = 5.0 \ \text{V} \\ \text{30} \\ \text{O} \\ \text$ | V _{IL} | LOW-level input voltage | | -0.5 | - | +0.3V _{DD} | V |
| $ \begin{array}{ c c c c c c c c } \hline V_{OL} = 0.4 \ V; \ V_{DD} = 5.0 \ V & 30 & - & - & mA \\ \hline V_{L} & leakage current & V_{I} = V_{DD} \ or \ V_{SS} & -1 & - & +1 & \muA \\ \hline C_{I} & input capacitance & V_{I} = V_{SS} & - & 6 & 10 & pF \\ \hline \textbf{LED driver outputs} \\ \hline V_{drv(LED)} & LED driver voltage & 0 & - & 40 & V \\ \hline I_{OL} & LOW-level output current & V_{OL} = 0.5 \ V & 22 & 100 & - & - & mA \\ \hline I_{LOH} & HIGH-level output leakage & V_{drv(LED)} = 5 \ V & - & \pm 1 & 15 & \muA \\ \hline C_{O} & output capacitance & V_{drv(LED)} = 40 \ V; \ V_{DD} = 2.3 \ V & - & 2 & 5 & \Omega \\ \hline \textbf{C}_{O} & output capacitance & V_{drv(LED)} = 40 \ V; \ V_{DD} = 2.3 \ V & - & 2.5 & 5 & pF \\ \hline \textbf{OE input} \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & 0.7 \ V_{DD} & - & 5.5 \ V \\ \hline Address inputs \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IL} & LOW-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline V_{IH} & HIGH-level input voltage & -0.5 & - & +0.3 \ V_{DD} \ V \\ \hline U_{IL} & input leakage current & -1 & - & +1 & +1 & \mu A \\ \hline \end{array}$ | V _{IH} | HIGH-level input voltage | | $0.7V_{DD}$ | - | 5.5 | V |
| $\begin{array}{c ccccccccccccccccccccccccccccccccccc$ | I _{OL} | LOW-level output current | $V_{OL} = 0.4 \text{ V}; V_{DD} = 2.3 \text{ V}$ | 20 | - | - | mA |
| $ \begin{array}{c ccccc} C_i & \text{input capacitance} & V_I = V_{SS} & - & 6 & 10 & pF \\ \hline \textbf{LED driver outputs} \\ \hline V_{drv(LED)} & \text{LED driver voltage} & 0 & - & 40 & V \\ \hline I_{OL} & \text{LOW-level output current} & V_{OL} = 0.5 \text{ V} & 2 & 100 & - & \pm 1 & \mu A \\ \hline I_{LOH} & \text{HIGH-level output leakage} & V_{drv(LED)} = 5 \text{ V} & - & \pm 1 & 15 & \mu A \\ \hline R_{on} & \text{ON-state resistance} & V_{drv(LED)} = 40 \text{ V}; V_{DD} = 2.3 \text{ V} & - & 2 & 5 & \Omega \\ \hline C_o & \text{output capacitance} & V_{drv(LED)} = 40 \text{ V}; V_{DD} = 2.3 \text{ V} & - & 2.5 & 5 & pF \\ \hline \hline \textbf{OE input} \\ \hline V_{IL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline V_{IH} & \text{HIGH-level input voltage} & 0.7 V_{DD} & - & 5.5 & V \\ \hline I_{LI} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{C}_i & \text{input capacitance} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & 5.5 & V \\ \hline \textbf{Multiput leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{C}_i & \text{input capacitance} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{LOW-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{HIGH-level input voltage} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{Input leakage current} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{Input leakage current} & -0.5 & - & +0.3 V_{DD} & V \\ \hline \textbf{VIL} & \text{Input leakage current} & -0.5 & - & +0.3 V_{DD} & - & 5.5 & V \\ \hline \textbf{IL} & \text{input leakage current} & -0.5 & - & +0.3 V_{DD} & - & 5.5 & V \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{input leakage current} & -1 & - & +1 & - & +1 & \mu A \\ \hline \textbf{IL} & \text{IL} & \text{Input leakage current} & -1 & - & +1 & - & +$ | | | $V_{OL} = 0.4 \text{ V}; V_{DD} = 5.0 \text{ V}$ | 30 | - | - | mA |
| LED driver outputs Vdrv(LED) LED driver voltage 0 - 40 V Vdrv(LED) LED driver voltage 0 - 40 V Vdrv(LED) LED driver voltage 0 - 40 V Vdrv(LED) ED driver voltage 0 - - mA Vdrv(LED) ED driver voltage Vdrv(LED) ED driver voltage - - ±1 15 μA Ron ON-state resistance Vdrv(LED) ED driver voltage - 2.5 5 pF OE input | IL | leakage current | $V_I = V_{DD}$ or V_{SS} | -1 | - | +1 | μΑ |
| $ V_{\text{drv(LED)}} \text{LED driver voltage} \qquad 0 \qquad - \qquad 40 \qquad V \\ I_{\text{OL}} \qquad \text{LOW-level output current} \qquad V_{\text{OL}} = 0.5 \text{ V} \qquad 2 \qquad 100 \qquad - \qquad mA \\ I_{\text{LOH}} \qquad \text{HIGH-level output leakage} \qquad V_{\text{drv(LED)}} = 5 \text{ V} \qquad - \qquad - \qquad \pm 1 \qquad \mu A \\ V_{\text{drv(LED)}} = 40 \text{ V} \qquad - \qquad \pm 1 \qquad 15 \qquad \mu A \\ V_{\text{oloc}} \qquad \text{ON-state resistance} \qquad V_{\text{drv(LED)}} = 40 \text{ V}; V_{\text{DD}} = 2.3 \text{ V} \qquad - \qquad 2 \qquad 5 \qquad \Omega \\ C_{\text{O}} \qquad \text{output capacitance} \qquad V_{\text{drv(LED)}} = 40 \text{ V}; V_{\text{DD}} = 2.3 \text{ V} \qquad - \qquad 2 \qquad 5 \qquad DF \\ \hline \textbf{OE input} \qquad \qquad - \qquad 0.5 \qquad 5 \qquad pF \\ \hline \textbf{OE input} \qquad \qquad V_{\text{IL}} \qquad \text{LOW-level input voltage} \qquad \qquad -0.5 \qquad - \qquad +0.3 V_{\text{DD}} \qquad V \\ V_{\text{H}} \qquad \text{HIGH-level input voltage} \qquad \qquad 0.7 V_{\text{DD}} \qquad - \qquad 5.5 \qquad V \\ I_{\text{LI}} \qquad \text{input leakage current} \qquad \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ C_{\text{i}} \qquad \text{input capacitance} \qquad \qquad -0.5 \qquad - \qquad +0.3 V_{\text{DD}} \qquad V \\ \hline \textbf{Address inputs} \qquad \qquad V_{\text{IL}} \qquad \text{LOW-level input voltage} \qquad \qquad -0.5 \qquad - \qquad +0.3 V_{\text{DD}} \qquad V \\ V_{\text{IH}} \qquad \text{HIGH-level input voltage} \qquad \qquad -0.5 \qquad - \qquad - \qquad 5.5 \qquad V \\ I_{\text{LI}} \qquad \text{input leakage current} \qquad \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address inputs} \qquad \qquad V_{\text{IL}} \qquad \text{LOW-level input voltage} \qquad \qquad -0.5 \qquad - \qquad +0.3 V_{\text{DD}} \qquad V \\ V_{\text{IH}} \qquad \text{HIGH-level input voltage} \qquad \qquad -0.5 \qquad - \qquad - \qquad 5.5 \qquad V \\ I_{\text{LI}} \qquad \text{input leakage current} \qquad \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address inputs} \qquad \qquad -1 \qquad - \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address input leakage current} \qquad \qquad -1 \qquad - \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address inputs} \qquad \qquad -1 \qquad - \qquad -1 \qquad - \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address input leakage current} \qquad \qquad -1 \qquad - \qquad -1 \qquad - \qquad -1 \qquad - \qquad +1 \qquad \mu A \\ \hline \textbf{Address input leakage current} \qquad \qquad -1 \qquad - \qquad -1 \qquad - \qquad -1 \qquad - \qquad -1 \qquad - \qquad -$ | Ci | input capacitance | $V_I = V_{SS}$ | - | 6 | 10 | pF |
| $\begin{array}{c} \text{IOL} \\ \text{IOL} \\ \text{IOH} \\ \\ \text{IICH} \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\$ | LED driver | outputs | | | | ' | |
| $\begin{array}{c} \text{I}_{\text{LOH}} & \text{HIGH-level output leakage} \\ \text{current} & \text{V}_{\text{drv(LED)}} = 5 \text{ V} \\ \text{V}_{\text{drv(LED)}} = 40 \text{ V} & - & \pm 1 & 15 & \mu \text{A} \\ \text{R}_{\text{on}} & \text{ON-state resistance} & \text{V}_{\text{drv(LED)}} = 40 \text{ V}; \text{V}_{\text{DD}} = 2.3 \text{ V} & - & 2 & 5 & \Omega \\ \text{C}_{\text{o}} & \text{output capacitance} & & - & 2.5 & 5 & \text{pF} \\ \hline \textbf{OE input} \\ \hline \textbf{V}_{\text{IL}} & \text{LOW-level input voltage} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \text{V}_{\text{IH}} & \text{HIGH-level input voltage} & & 0.7 \text{V}_{\text{DD}} & - & 5.5 & \text{V} \\ \text{I}_{\text{LI}} & \text{input leakage current} & & -1 & - & +1 & \mu \text{A} \\ \text{C}_{\text{i}} & \text{input capacitance} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{F} \\ \hline \textbf{Address inputs} \\ \hline \textbf{V}_{\text{IL}} & \text{LOW-level input voltage} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{V}_{\text{IH}} & \text{HIGH-level input voltage} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{V}_{\text{IL}} & \text{LOW-level input voltage} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{V}_{\text{IH}} & \text{HIGH-level input voltage} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{V}_{\text{IL}} & \text{input leakage current} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{I}_{\text{LI}} & \text{input leakage current} & & -0.5 & - & +0.3 \text{V}_{\text{DD}} & \text{V} \\ \hline \textbf{I}_{\text{LI}} & \text{input leakage current} & & -1 & - & +1 & \mu \text{A} \\ \hline \end{tabular}$ | $V_{drv(LED)}$ | LED driver voltage | | 0 | - | 40 | V |
| current $V_{drv(LED)} = 40 \text{ V}$ - ± 1 15 μA Ron ON-state resistance $V_{drv(LED)} = 40 \text{ V}$; $V_{DD} = 2.3 \text{ V}$ - 2 5 Ω Co output capacitance - 2.5 5 ρF OE input VIL LOW-level input voltage - 0.5 - $+0.3V_{DD}$ V VIH HIGH-level input voltage - 0.7 V_{DD} - 5.5 V ILI input leakage current - 1 - $+1$ μA Ci input capacitance - 15 $+0.3V_{DD}$ V Address inputs VIL LOW-level input voltage - 0.5 - $+0.3V_{DD}$ V HIGH-level input voltage - $+0.5$ - $+0$ | I _{OL} | LOW-level output current | V _{OL} = 0.5 V | 100 | - | - | mA |
| $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$ | I _{LOH} | HIGH-level output leakage | $V_{drv(LED)} = 5 V$ | - | - | ±1 | μΑ |
| $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$ | | current | $V_{drv(LED)} = 40 \text{ V}$ | - | ±1 | 15 | μΑ |
| OE input V _{IL} LOW-level input voltage -0.5 - +0.3V _{DD} V V _{IH} HIGH-level input voltage 0.7V _{DD} - 5.5 V I _{LI} input leakage current -1 - +1 μA C _i input capacitance - 15 40 pF Address inputs V _{IL} LOW-level input voltage -0.5 - +0.3V _{DD} V V _{IH} HIGH-level input voltage 0.7V _{DD} - 5.5 V I _{LI} input leakage current -1 - +1 μA | R _{on} | ON-state resistance | $V_{drv(LED)} = 40 \text{ V}; V_{DD} = 2.3 \text{ V}$ | - | 2 | 5 | Ω |
| $\begin{array}{c ccccccccccccccccccccccccccccccccccc$ | Co | output capacitance | | - | 2.5 | 5 | pF |
| $\begin{array}{c ccccccccccccccccccccccccccccccccccc$ | OE input | | | | | | |
| I_{LI} input leakage current -1 - +1 μA C_i input capacitance -1 5 40 pF Address inputs V_{IL} LOW-level input voltage -0.5 - $+0.3V_{DD}$ V V_{IH} HIGH-level input voltage $0.7V_{DD}$ - 5.5 V I_{LI} input leakage current -1 - $+1$ μA | V _{IL} | LOW-level input voltage | | -0.5 | - | +0.3V _{DD} | V |
| $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$ | V _{IH} | HIGH-level input voltage | | $0.7V_{DD}$ | - | 5.5 | V |
| Address inputs $V_{IL} \hspace{0.5cm} \text{LOW-level input voltage} \hspace{0.5cm} -0.5 \hspace{0.5cm} - \hspace{0.5cm} +0.3 V_{DD} \hspace{0.5cm} \text{V} \\ V_{IH} \hspace{0.5cm} \text{HIGH-level input voltage} \hspace{0.5cm} 0.7 V_{DD} \hspace{0.5cm} - \hspace{0.5cm} 5.5 \hspace{0.5cm} \text{V} \\ I_{LI} \hspace{0.5cm} \text{input leakage current} \hspace{0.5cm} -1 \hspace{0.5cm} -1 \hspace{0.5cm} +1 \hspace{0.5cm} \mu \text{A} \\ \end{array}$ | ILI | input leakage current | | -1 | - | +1 | μΑ |
| V_{IL} LOW-level input voltage -0.5 - $+0.3V_{DD}$ V V_{IH} HIGH-level input voltage $0.7V_{DD}$ - 5.5 V I_{LI} input leakage current -1 - $+1$ μA | Ci | input capacitance | | - | 15 | 40 | pF |
| V_{IH} HIGH-level input voltage 0.7 V_{DD} - 5.5 V | Address in | puts | | | | 1 | |
| I _{LI} input leakage current -1 - +1 μA | V _{IL} | LOW-level input voltage | | -0.5 | - | +0.3V _{DD} | V |
| I _{LI} input leakage current -1 - +1 μA | V _{IH} | HIGH-level input voltage | | $0.7V_{DD}$ | - | 5.5 | V |
| | ILI | input leakage current | | -1 | - | +1 | μΑ |
| | Ci | input capacitance | | - | 3.7 | 5 | pF |

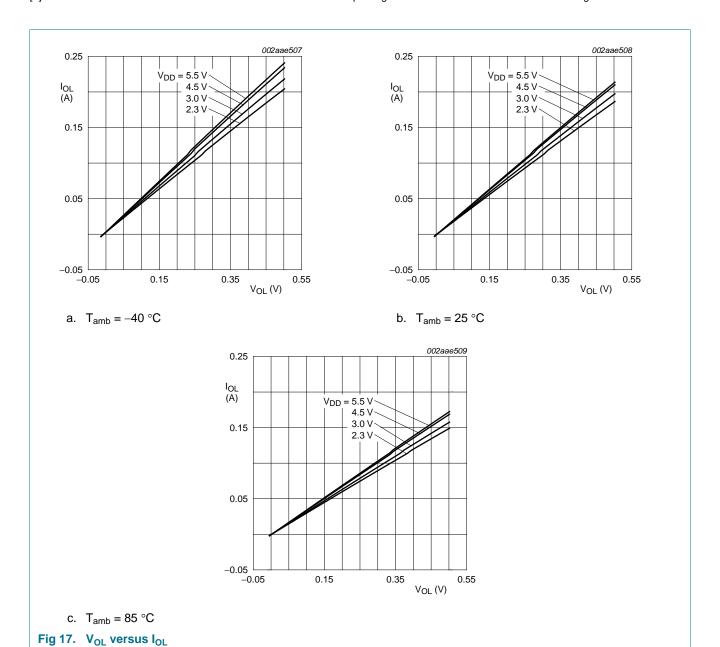
^[1] V_{DD} must be lowered to 0.2 V in order to reset part.

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16-bit Fm+ I²C-bus 100 mA 40 V LED driver

[2] Each bit must be limited to a maximum of 100 mA and the total package limited to 1600 mA due to internal busing limits.



16-bit Fm+ I²C-bus 100 mA 40 V LED driver

14. Dynamic characteristics

Table 18. Dynamic characteristics

| Symbol | Parameter | meter Conditions | | Standard-mode I ² C-bus | | Fast-mode I ² | C-bus | | mode ² C-bus | Unit |
|-----------------------|--|--|------------|------------------------------------|------|----------------------------|-------|------|----------------------------|------|
| | | | | Min | Max | Min | Max | Min | Max | |
| f _{SCL} | SCL clock frequency | | | 0 | 100 | 0 | 400 | 0 | 1000 | kHz |
| t _{BUF} | bus free time between a STOP and START condition | | | 4.7 | - | 1.3 | - | 0.5 | - | μS |
| t _{HD;STA} | hold time (repeated) START condition | | | 4.0 | - | 0.6 | - | 0.26 | - | μS |
| t _{SU;STA} | set-up time for a repeated START condition | | | 4.7 | - | 0.6 | - | 0.26 | - | μS |
| t _{SU;STO} | set-up time for STOP condition | | | 4.0 | - | 0.6 | - | 0.26 | - | μS |
| t _{HD;DAT} | data hold time | | | 0 | - | 0 | - | 0 | - | ns |
| t _{VD;ACK} | data valid acknowledge time | | <u>[1]</u> | 0.3 | 3.45 | 0.1 | 0.9 | 0.05 | 0.45 | μS |
| t _{VD;DAT} | data valid time | | [2] | 0.3 | 3.45 | 0.1 | 0.9 | 0.05 | 0.45 | μS |
| t _{SU;DAT} | data set-up time | | | 250 | - | 100 | - | 50 | - | ns |
| t_{LOW} | LOW period of the SCL clock | | | 4.7 | - | 1.3 | - | 0.5 | - | μS |
| t _{HIGH} | HIGH period of the SCL clock | | | 4.0 | - | 0.6 | - | 0.26 | - | μS |
| t _f | fall time of both SDA and SCL signals | | [3][4] | - | 300 | 20 + 0.1C _b [5] | 300 | - | 120 | ns |
| t _r | rise time of both SDA and SCL signals | | | - | 1000 | 20 + 0.1C _b [5] | 300 | - | 120 | ns |
| t _{SP} | pulse width of spikes that must be suppressed by the input filter | | <u>[6]</u> | - | 50 | - | 50 | - | 50 | ns |
| Output p | ropagation delay | | | | | | | | | |
| t _{PLH} | LOW to HIGH propagation delay | OE to LEDn; MODE2[1:0] = 01 | | - | - | - | - | - | 150 | ns |
| t _{PHL} | HIGH to LOW propagation delay | OE to LEDn; MODE2[1:0] = 01 | | - | - | - | - | - | 150 | ns |
| Output p | ort timing | | | | | | | | | |
| t _{d(SCL-Q)} | delay time from SCL to data output | SCL to LEDn; MODE2[3] = 1; outputs change on ACK | | - | - | - | - | - | 450 | ns |
| t _{d(SDA-Q)} | delay time from SDA to data output | SDA to LEDn; MODE2[3] = 0; outputs change on STOP condition | | - | - | - | - | - | 450 | ns |

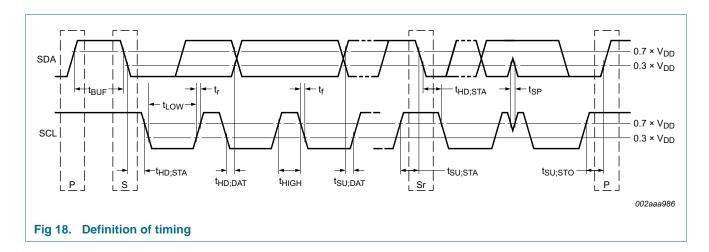
^[1] $t_{VD;ACK}$ = time for Acknowledgement signal from SCL LOW to SDA (out) LOW.

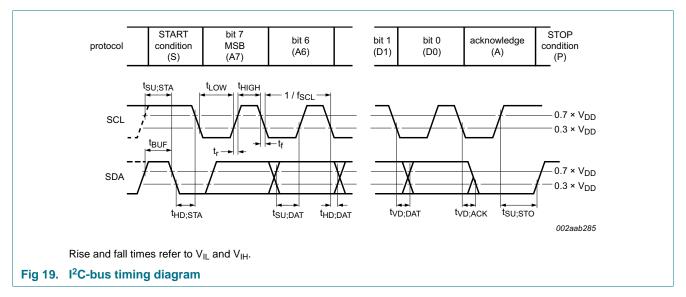
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16-bit Fm+ I²C-bus 100 mA 40 V LED driver

- [2] $t_{VD;DAT}$ = minimum time for SDA data out to be valid following SCL LOW.
- [3] A master device must internally provide a hold time of at least 300 ns for the SDA signal (refer to the V_{IL} of the SCL signal) in order to bridge the undefined region of SCL's falling edge.
- [4] The maximum t_f for the SDA and SCL bus lines is specified at 300 ns. The maximum fall time (t_f) for the SDA output stage is specified at 250 ns. This allows series protection resistors to be connected between the SDA and the SCL pins and the SDA/SCL bus lines without exceeding the maximum specified t_f.
- [5] $C_b = \text{total capacitance of one bus line in pF.}$
- [6] Input filters on the SDA and SCL inputs suppress noise spikes less than 50 ns.

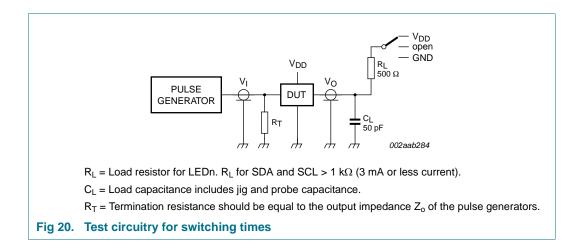




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15. Test information

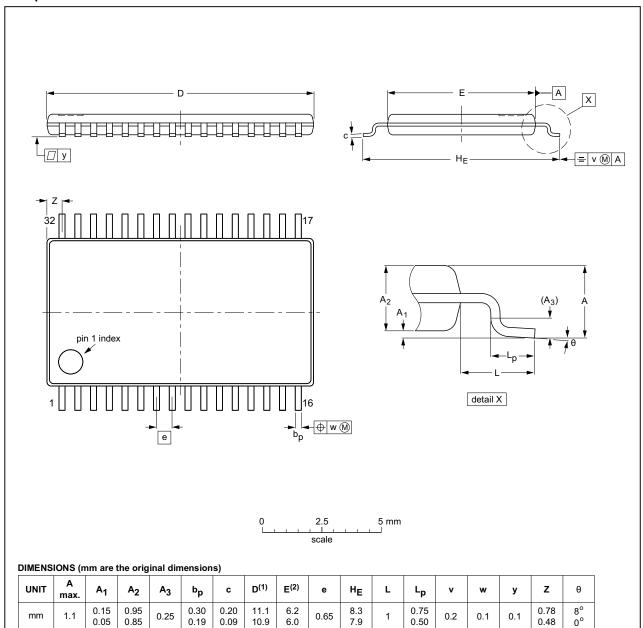


16-bit Fm+ I2C-bus 100 mA 40 V LED driver

16. Package outline

TSSOP32: plastic thin shrink small outline package; 32 leads; body width 6.1 mm; lead pitch 0.65 mm

SOT487-1



Notes

- 1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.
- 2. Plastic interlead protrusions of 0.25 mm maximum per side are not included.

| IEC JEHA JEHA | | EUROPEAN | REFERENCES | | | |
|-----------------|----------------------|------------|------------|-----|----------|--|
| | TION ISSUE DATE | PROJECTION | JEDEC | IEC | VERSION | |
| SOT487-1 MO-153 | 99-12-27 03-02-18 | | MO-153 | | SOT487-1 | |

Fig 21. Package outline SOT487-1 (TSSOP32)

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17. Handling information

All input and output pins are protected against ElectroStatic Discharge (ESD) under normal handling. When handling ensure that the appropriate precautions are taken as described in *JESD625-A* or equivalent standards.

18. Soldering of SMD packages

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note *AN10365* "Surface mount reflow soldering description".

18.1 Introduction to soldering

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

18.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- Through-hole components
- Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- Board specifications, including the board finish, solder masks and vias
- Package footprints, including solder thieves and orientation
- The moisture sensitivity level of the packages
- Package placement
- Inspection and repair
- Lead-free soldering versus SnPb soldering

18.3 Wave soldering

Key characteristics in wave soldering are:

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- Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- Solder bath specifications, including temperature and impurities

18.4 Reflow soldering

Key characteristics in reflow soldering are:

- Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see <u>Figure 22</u>) than a SnPb process, thus reducing the process window
- Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with Table 19 and 20

Table 19. SnPb eutectic process (from J-STD-020D)

| Package thickness (mm) | Package reflow temperature (°C) | | | |
|------------------------|---------------------------------|-------|--|--|
| | Volume (mm³) | | | |
| | < 350 | ≥ 350 | | |
| < 2.5 | 235 | 220 | | |
| ≥ 2.5 | 220 | 220 | | |

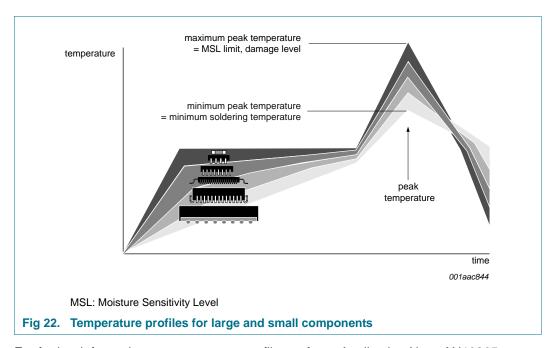
Table 20. Lead-free process (from J-STD-020D)

| Package thickness (mm) | Package reflow temperature (°C) | | | | | |
|------------------------|---------------------------------|-------------|--------|--|--|--|
| | Volume (mm³) | | | | | |
| | < 350 | 350 to 2000 | > 2000 | | | |
| < 1.6 | 260 | 260 | 260 | | | |
| 1.6 to 2.5 | 260 | 250 | 245 | | | |
| > 2.5 | 250 | 245 | 245 | | | |

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see Figure 22.

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For further information on temperature profiles, refer to Application Note *AN10365* "Surface mount reflow soldering description".

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19. Soldering: PCB footprints

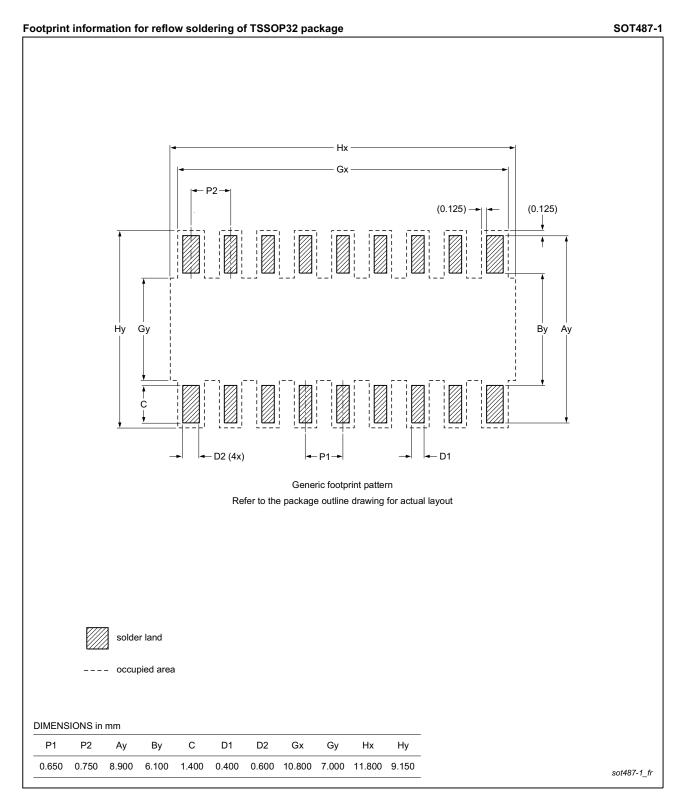


Fig 23. PCB footprint for SOT487-1 (TSSOP32); reflow soldering

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20. Abbreviations

Table 21. Abbreviations

| Acronym | Description |
|----------------------|--|
| CDM | Charged-Device Model |
| DUT | Device Under Test |
| ESD | ElectroStatic Discharge |
| FET | Field-Effect Transistor |
| HBM | Human Body Model |
| I ² C-bus | Inter-Integrated Circuit bus |
| LED | Light Emitting Diode |
| LCD | Liquid Crystal Display |
| LSB | Least Significant Bit |
| MSB | Most Significant Bit |
| NMOS | Negative-channel Metal-Oxide Semiconductor |
| PCB | Printed-Circuit Board |
| PMOS | Positive-channel Metal-Oxide Semiconductor |
| PWM | Pulse Width Modulation |
| RGB | Red/Green/Blue |
| RGBA | Red/Green/Blue/Amber |
| SMBus | System Management Bus |

21. Revision history

Table 22. Revision history

| Document ID | Release date | Data sheet status | Change notice | Supersedes | | | | |
|----------------|--|---|---------------|-------------|--|--|--|--|
| PCA9622 v.5 | 20140602 | Product data sheet | - | PCA9622 v.4 | | | | |
| Modifications: | Table 5 "RegiTable 6 "MOD cross-reference | n 4.1 "Ordering options" ster summary[1]": deleted old table note [2] E1 - Mode register 1 (address 00h) bit description": added (new) Table note [1] and its ce at SLEEP bit (bit 4) n 19 "Soldering: PCB footprints" | | | | | | |
| PCA9622 v.4 | 20120906 | Product data sheet | - | PCA9622 v.3 | | | | |
| PCA9622 v.3 | 20090831 | Product data sheet | - | PCA9622 v.2 | | | | |
| PCA9622 v.2 | 20090611 | Product data sheet | - | PCA9622 v.1 | | | | |
| PCA9622 v.1 | 20090327 | Product data sheet | - | - | | | | |

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22. Legal information

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| Document status[1][2] | Product status[3] | Definition |
|--------------------------------|-------------------|---|
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| Preliminary [short] data sheet | Qualification | This document contains data from the preliminary specification. |
| Product [short] data sheet | Production | This document contains the product specification. |

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
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16-bit Fm+ I2C-bus 100 mA 40 V LED driver

24. Contents

11

| 1 | General description | . 1 | 12 | Thermal characteristics | 26 |
|----------------|---|-----|------|---------------------------|----|
| 2 | Features and benefits | . 2 | 13 | Static characteristics | 27 |
| 3 | Applications | . 3 | 14 | Dynamic characteristics | 29 |
| 4 | Ordering information | . 3 | 15 | Test information | 31 |
| 4.1 | Ordering options | . 3 | 16 | Package outline | 32 |
| 5 | Block diagram | . 4 | 17 | Handling information | 33 |
| 6 | Pinning information | . 5 | 18 | Soldering of SMD packages | |
| 6.1 | Pinning | | 18.1 | Introduction to soldering | |
| 6.2 | Pin description | | 18.2 | Wave and reflow soldering | |
| 7 | Functional description | . 6 | 18.3 | Wave soldering | |
| 7.1 | Device addresses | | 18.4 | Reflow soldering | 34 |
| 7.1.1 | Regular I ² C-bus slave address | | 19 | Soldering: PCB footprints | 36 |
| 7.1.2 | LED All Call I ² C-bus address | . 7 | 20 | Abbreviations | 37 |
| 7.1.3 | LED Sub Call I ² C-bus addresses | | 21 | Revision history | 37 |
| 7.1.4 | Software Reset I ² C-bus address | | 22 | Legal information | |
| 7.2 | Control register | | 22.1 | Data sheet status | |
| 7.3 | Register definitions | | 22.2 | Definitions | |
| 7.3.1 7.3.2 | Mode register 1, MODE1 | | 22.3 | Disclaimers | 38 |
| 7.3.2 | PWM0 to PWM15, individual brightness | 11 | 22.4 | Trademarks | |
| 7.3.3 | control | 12 | 23 | Contact information | 39 |
| 7.3.4 | GRPPWM, group duty cycle control | | 24 | Contents | |
| 7.3.5 | GRPFREQ, group frequency | | | | |
| 7.3.6 | LEDOUT0 to LEDOUT3, LED driver output | | | | |
| | state | 14 | | | |
| 7.3.7 | SUBADR1 to SUBADR3, I ² C-bus subaddress | | | | |
| | | | | | |
| 7.3.8 | ALLCALLADR, LED All Call I ² C-bus address. | | | | |
| 7.4 | Active LOW output enable input | | | | |
| 7.5 | Power-on reset | | | | |
| 7.6 7.7 | Software reset | 10 | | | |
| 1.1 | dimming/blinking | 17 | | | |
| 8 | Characteristics of the I ² C-bus | | | | |
| 8.1 | Bit transfer | | | | |
| 8.1.1 | START and STOP conditions | - | | | |
| 8.2 | System configuration | | | | |
| 8.3 | Acknowledge | | | | |
| 9 | | 20 | | | |
| 10 | Application design-in information | | | | |
| 10.1 | Junction temperature calculation | | | | |
| 10.1.1 | Example 1: T _i calculation of PCA9622DR, in | | | | |
| | TSSOP32 package, when T _{amb} is known | 24 | | | |
| 10.1.2 | Example 2: T _i calculation where only T _{case} is | | | | |
| | known | 24 | | | |

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