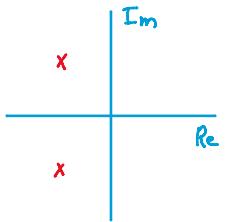


### 3 - Frequency response - Type 3 building blocks, nonminimum-phase systems & sketching procedure

#### Frequency response analysis

- 3<sup>rd</sup> class / Class 3: Second order complex conjugate pair of poles and zeros
- Frequency response of RHP zero
- Drawing rules bode plot



#### Class 3

$$\left[ \left( \frac{jw}{\omega_n} \right)^2 + 2 \left[ \frac{jw}{\omega_n} + 1 \right] \right]^{\pm 1}$$

Damping coefficient

Break point  $\frac{1}{\omega_n}$

natural frequency

#### Behaviour:

- $w/\omega_n \gg 1$ ,  $\pm 40\text{dB}/\text{decade}$
- $w/\omega_n \approx 1$ , resonant peak  $M_r$
- Phase lead/lag  $180^\circ$ ,  $90^\circ$  at  $\omega_n$

$$\frac{1}{\left( \frac{jw}{\omega_n} \right)^2 + 2 \left[ \frac{jw}{\omega_n} + 1 \right]} \quad (\text{Complex conjugate pole pair})$$

$$\text{Re: } \frac{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)}{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)^2 + \left( 2 \frac{\omega}{\omega_n} \right)^2}$$

$$\text{Im: } \frac{-2 \frac{\omega}{\omega_n}}{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)^2 + \left( 2 \frac{\omega}{\omega_n} \right)^2}$$

#### Three ranges:

- 1)  $\omega \ll \omega_n$
- 2)  $\omega \gg \omega_n$
- 3)  $\omega = \omega_n$

#### $\omega \ll \omega_n$

~~$\frac{\omega}{\omega_n}$~~  is very small

$$\text{Re: } \frac{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)}{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)^2 + \left( 2 \frac{\omega}{\omega_n} \right)^2} \approx \frac{1}{1+0}$$

$$\text{Im: } \frac{-2 \frac{\omega}{\omega_n}}{\left( 1 - \left( \frac{\omega}{\omega_n} \right)^2 \right)^2 + \left( 2 \frac{\omega}{\omega_n} \right)^2} \approx \frac{0}{1+0}$$

Gain:  $\sqrt{1^2 + 0^2} = 1 = 0\text{dB}$

phase:  $\arctan(0/1) = 0^\circ$

$\omega \gg \omega_n$

$$1 - \left(\frac{\omega}{\omega_n}\right)^2 \approx -\left(\frac{\omega}{\omega_n}\right)^2$$

$$\text{Re} : \frac{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)}{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta\left(\frac{\omega}{\omega_n}\right)\right)^2} \approx \frac{-\left(\frac{\omega}{\omega_n}\right)^2}{\left(-\left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta\left(\frac{\omega}{\omega_n}\right)\right)^2} \approx -\left(\frac{\omega}{\omega_n}\right)^2 = -\left(\frac{\omega}{\omega_n}\right)^{-2}$$

$$\text{Im} : \frac{-2\zeta\left(\frac{\omega}{\omega_n}\right)}{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta\left(\frac{\omega}{\omega_n}\right)\right)^2} \approx \frac{-2\zeta}{\left(\frac{\omega}{\omega_n}\right)^3}$$

↑ small compared to

$$\text{Gain} : \left[ \left(\frac{1}{\omega/\omega_n}\right)^2 + \left(\frac{-2\zeta}{\omega/\omega_n}\right)^2 \right]^{1/2} \approx \left(\frac{1}{\omega/\omega_n}\right)^{1/2} = \left(\frac{\omega}{\omega_n}\right)^{-1/2}$$

↑ small compared to

$$\text{In decibels} : 20 \log_{10} \left(\frac{\omega}{\omega_n}\right)^{-1/2} = \pm 20 \log_{10} \left(\frac{\omega}{\omega_n}\right)$$

$$\text{Phase} : \arctan \left( \frac{-2\zeta\left(\frac{\omega}{\omega_n}\right)^{-3}}{\left(\frac{\omega}{\omega_n}\right)^{-2}} \right) = \arctan \left( -2\zeta\left(\frac{\omega}{\omega_n}\right)^{-1} \right)$$

$= 0^\circ$  sign information ignored,  $\pm 180^\circ$

$\omega = \omega_n$

$$\frac{\omega_0}{\omega_n} = 1$$

$$\text{Re} : \frac{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)}{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta\left(\frac{\omega}{\omega_n}\right)\right)^2} = 0$$

$$\text{Im} : \frac{-2\zeta\left(\frac{\omega}{\omega_n}\right)}{\left(1 - \left(\frac{\omega}{\omega_n}\right)^2\right)^2 + \left(2\zeta\left(\frac{\omega}{\omega_n}\right)\right)^2} = \frac{-1}{2\zeta}$$

	Zeros	Poles
$\zeta < 1/2$	-	+
$\zeta > 1/2$	+	-

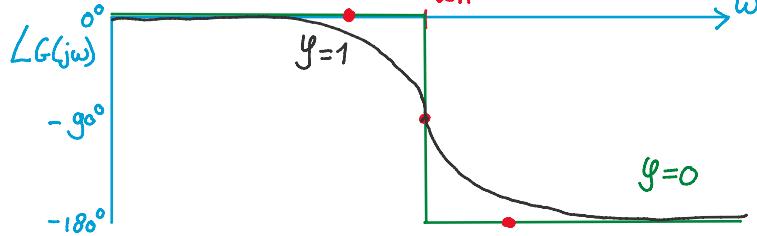
+ : amplification  
- : attenuation

$$\text{Phase} : \arctan \left( \frac{\text{Im}}{\text{Re}} \right) = \infty$$

Very small

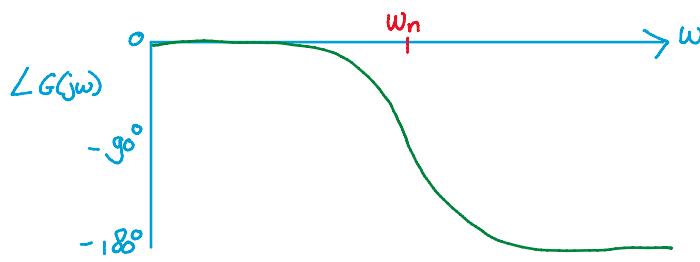
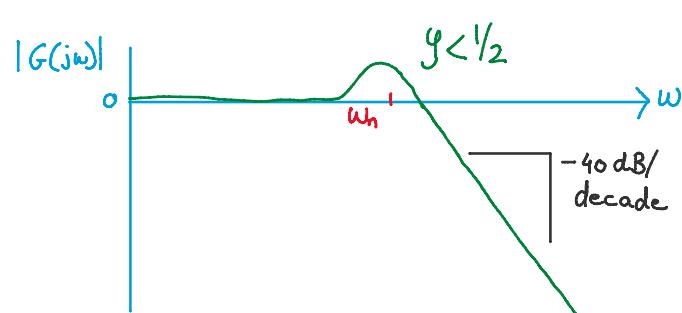
$$\text{Gain} : \sqrt{\left(\frac{-1}{2\zeta}\right)^2} = \frac{1}{2\zeta}, \text{ In decibels } \pm 20 \log_{10} (2\zeta)$$

$$\begin{cases} \zeta = 1/2 & \log_{10}(2\zeta) = 0 \\ \zeta < 1/2 & \log_{10}(2\zeta) \text{ negative} \\ \zeta > 1/2 & \log_{10}(2\zeta) \text{ positive} \end{cases}$$



### Class 3 bode plot

Complex conjugate pole pair



### RHP zeros

$$G_1 = 10 \frac{s+1}{s+10}$$

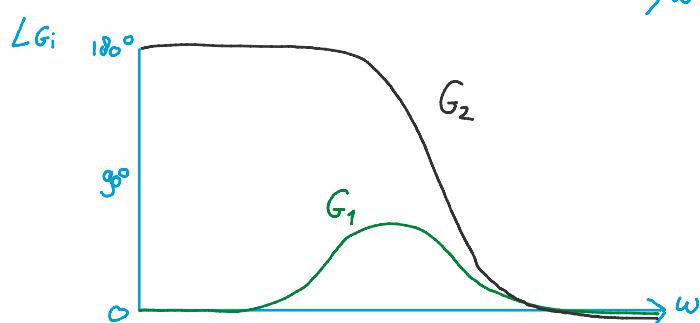
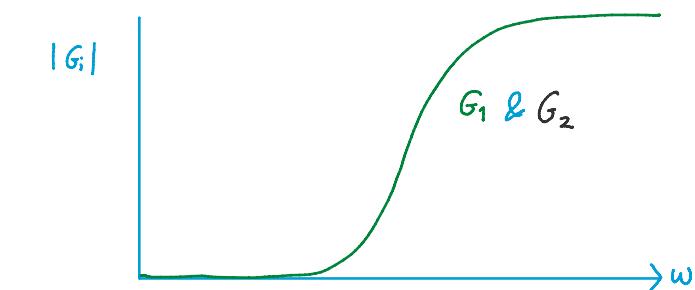
$$G_2 = 10 \frac{s-1}{s+10}$$

### RHP zero

Gain:  $20 \text{ dB/decade}$

Phase:  $-90^\circ$

$$|G_1(j\omega)| = |G_2(j\omega)|$$



## Design procedure bode plots

- 1.) Write TF in bode form, identify building blocks and their break point frequencies
- 2.) Determine  $n$  in  $K_0(j\omega_b)^n$ , plot low Frequency asymptote through  $K_0$   
Bode gain-phase: gain  $n \times 20 \text{ dB/decade}$  phase  $n \times 90^\circ$
- 3.) Extend asymptote until first break point. Change slope based on pole/zero at break points  
Continue in ascending order
- 4.) At the break points, magnitude deviates  $3\text{dB}$  (1st order) or  $\vartheta$ -dependent (2nd order)
- 5.) Plot low Frequency asymptote  $n \times 90^\circ$
- 6.) Phase change of  $\pm 90^\circ$  or  $\pm 180^\circ$  at break points
- 7.) Draw asymptotes for phase and intersect them
- 8.) Account for  $\pi^\circ$  phase difference at asymptote intersection

## Summary

- Frequency response of 2nd order system as function of  $\vartheta$   
$$\left[ \left( \frac{j\omega}{\omega_n} \right)^2 + 2\zeta \left( \frac{j\omega}{\omega_n} \right) + 1 \right]^{\frac{1}{2}}$$
- Zero in LHP VS RHP
- Overall design procedure  
Transfer function  $\rightarrow$  Bode plot  $\rightarrow$  System properties