



# Components in wireless technologies (5XTC0)

## Lab 1: Computer-aided circuit simulation tool QUCS

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# Agenda

- Why QUCS?
- Download & installation of QUCS
- Instructions
- Exercises using QUCS

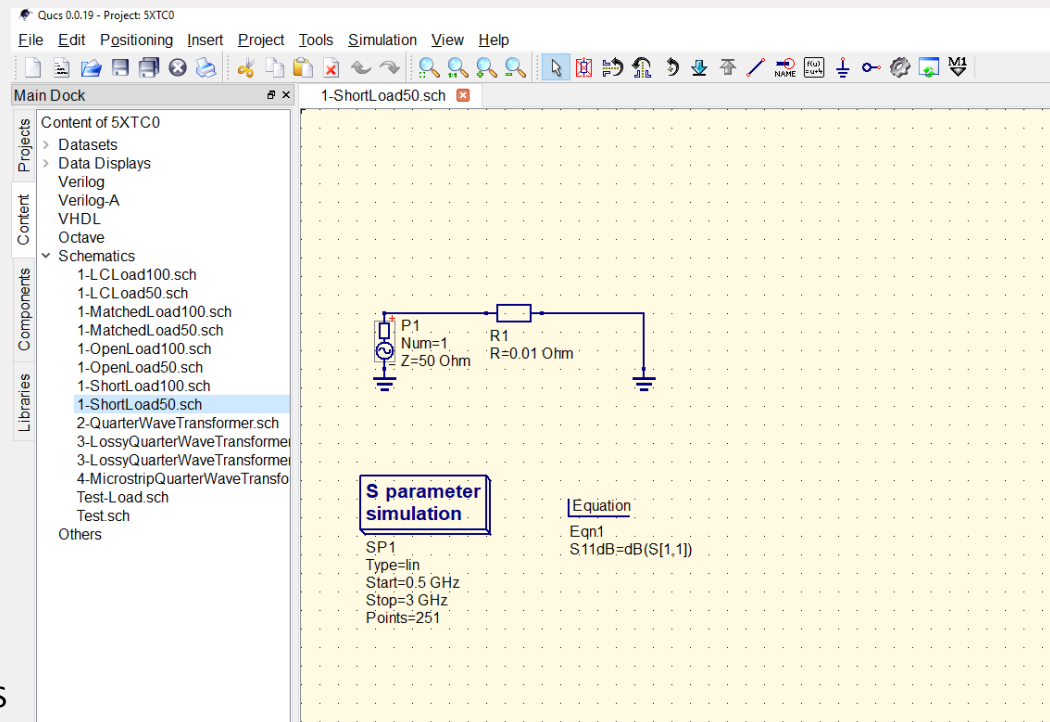


Figure 1: QUCS

# Why QUCS?

QUCS (Quite Universal Circuit Simulator) is open-source and free of use circuit simulator. It has a similar interface and functionality as the commercial software ADS (Advanced Design System) from Keysight.

Ideal for learning how to use a (RF) circuit simulator.

It has some bugs/quirks.

Example: press ENTER after entering a value instead of clicking away.

There are a couple forks of the project (QUCS-S, QucsStudio etc). Feel free to use another version. We use QUCS v0.0.19 in this course for explanations.

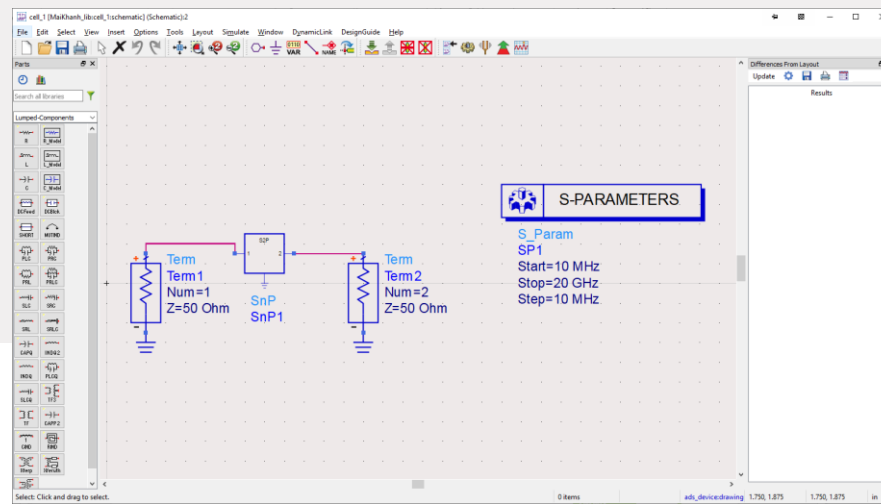
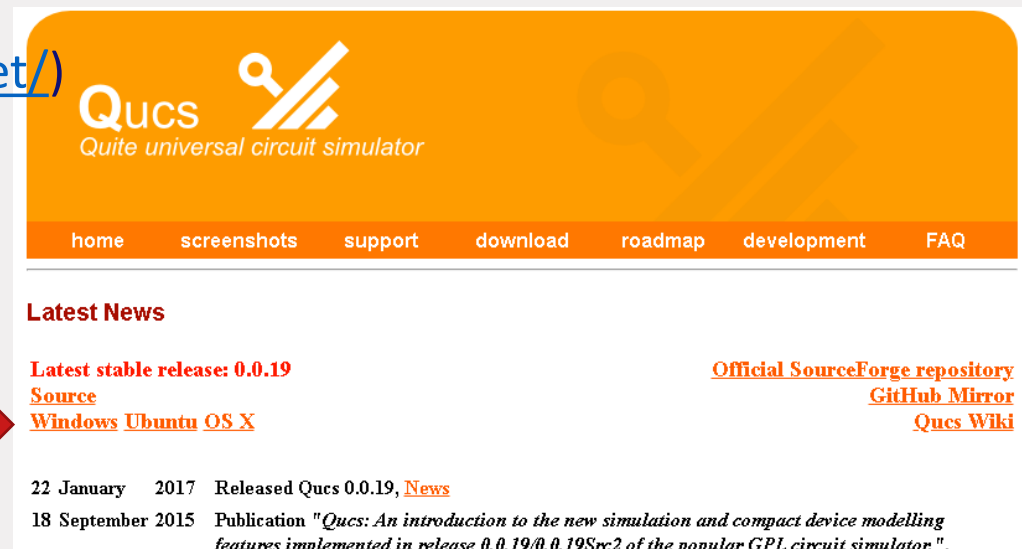


Figure 2: ADS Schematic view

# Downloading QUCS v0.0.19

- Windows:
  - from TU/e ([LINK](#))
  - from SourceForge (<http://qucs.sourceforge.net/>)
- Ubuntu or Mac OS X
  - from SourceForge (<http://qucs.sourceforge.net/>)



# Installing / Starting QUCS

- Windows
  - Unzip and start QUCS via "qucs.bat" in the root folder of QUCS. When starting Qucs.exe in the binary folder, the simulator can crash for an unclear reason(!)
- Ubuntu
  - Install .deb package and start shortcut
- Mac
  - Install .pkg package and start shortcut

# Questions?

If you have issues downloading or installing QUCS, please contact

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040 247 3417  
Flux 7.089



# Lab 1: Computer-aided circuit simulation tool QUCS

## Exercises



# Instructions

You can work in pairs if you like or work solo.

Use the template answer document to write the results and copy-paste the schematics and graphs in. Explain then the results in your own words.

When you are done, save the document with name:  
5XTC0 - LAB1 - STUDENTNUMBER & NAME.docx

Then upload the document on CANVAS.

Time during this lab should be sufficient. Submission deadline is set to 23:59 tomorrow (14-02-2025).



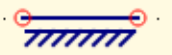
# Symbols that you should use in these exercises


Components - simulations - S-parameter simulation

**S parameter  
simulation**

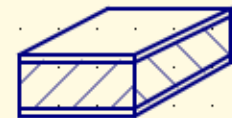
SP1  
Type=lin  
Start=1 GHz  
Stop=10 GHz  
Points=19

Components - transmission lines -  
Transmission Line      Microstrip Line

  
Line1  
 $Z=50 \text{ Ohm}$   
 $L=1 \text{ mm}$

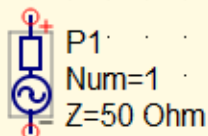
  
MS1  
Subst=Subst1  
 $W=1 \text{ mm}$   
 $L=10 \text{ mm}$

Substrate



Subst1  
 $\epsilon_r=9.8$   
 $h=1 \text{ mm}$   
 $t=35 \text{ um}$   
 $\tan\delta=2e-4$   
 $\rho=0.022e-6$   
 $D=0.15e-6$

Components - sources - Power Source

  
P1  
Num=1  
 $Z=50 \text{ Ohm}$

Insert - Insert Equation

Equation

Eqn1  
 $y=1$

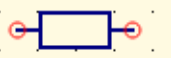
Components - lumped components -


Resistor


Capacitor

Inductor

Ground

  
R1  
 $R=50 \text{ Ohm}$

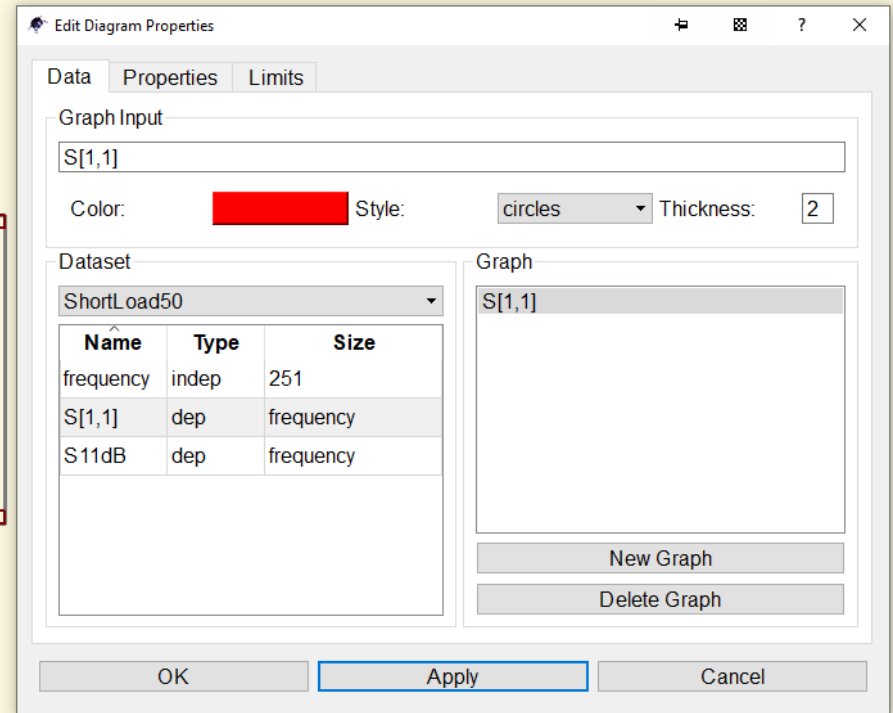
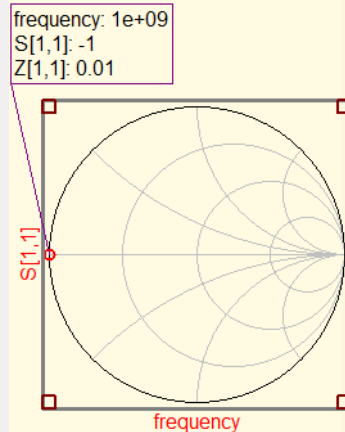
  
C1  
 $C=1 \text{ pF}$

  
L1  
 $L=1 \text{ nH}$

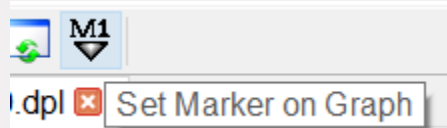


# Tips & Tricks

How to make a dot visible in Smith Chart:  
Use “Circles”  
and set  
“Thickness” to 2



Place the marker via



and click on the trace

Select the marker and press the arrow buttons on your keyboard to change the frequency of the marker

# Equations to use for transmission lines

Terminated transmission line equations (week 1):

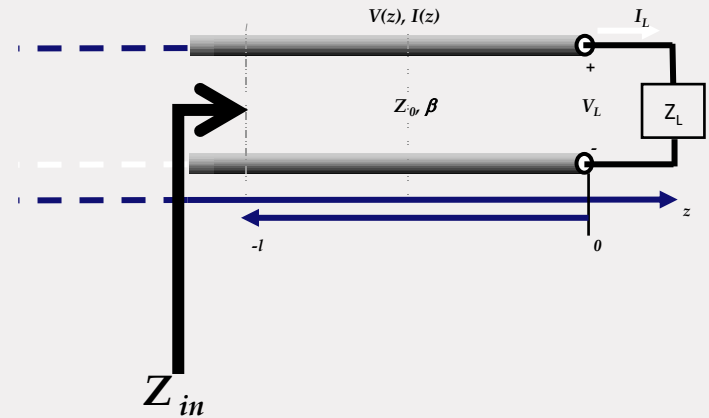
$$\Gamma(z = 0) = \frac{V_0^-}{V_0^+} = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$\Gamma(z = -l) = \frac{V_0^- e^{-j\beta l}}{V_0^+ e^{j\beta l}} = \Gamma e^{-2j\beta l},$$

$$\Gamma = \Gamma(z = 0).$$

$$Z_{in}(z = -l) = Z_0 \frac{Z_L + jZ_0 \tan \beta l}{Z_0 + jZ_L \tan \beta l}$$

$$\lambda = \frac{2\pi}{\beta}, f = \frac{c}{\lambda}$$



# Exercise 1 – Transmission lines

*Simulate in QUCS using S-parameters:*

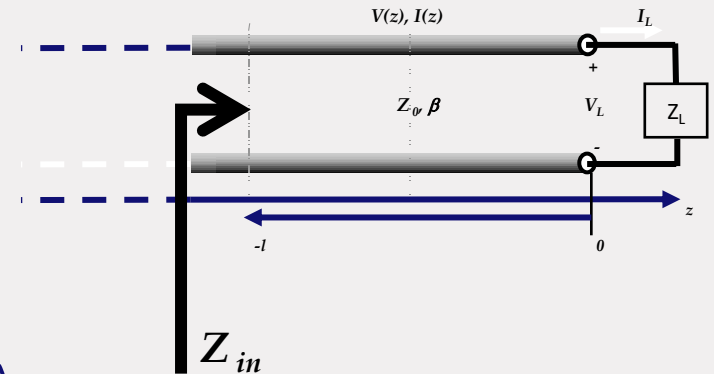
- Short-circuit load ( $Z_0 = 50 \text{ Ohm}$ )
- Short-circuit load ( $Z_0 = 100 \text{ Ohm}$ )
- Open-ended load ( $Z_0 = 50 \text{ Ohm}$ )
- Open-ended load ( $Z_0 = 100 \text{ Ohm}$ )
- Matched load ( $Z_0 = 50 \text{ Ohm}$ ,  $Z_L = 50 \text{ Ohm}$ )
- Matched load ( $Z_0 = 100 \text{ Ohm}$ ,  $Z_L = 100 \text{ Ohm}$ )
- LC lumped elements (resonator) and 50 Ohm resistor as load ( $Z_0 = 50 \text{ Ohm}$ ) (resonating frequency = 1 GHz)
- LC lumped elements (resonator) and 100 Ohm resistor as load ( $Z_0 = 100 \text{ Ohm}$ ) (resonating frequency = 1 GHz)

*Output data:*

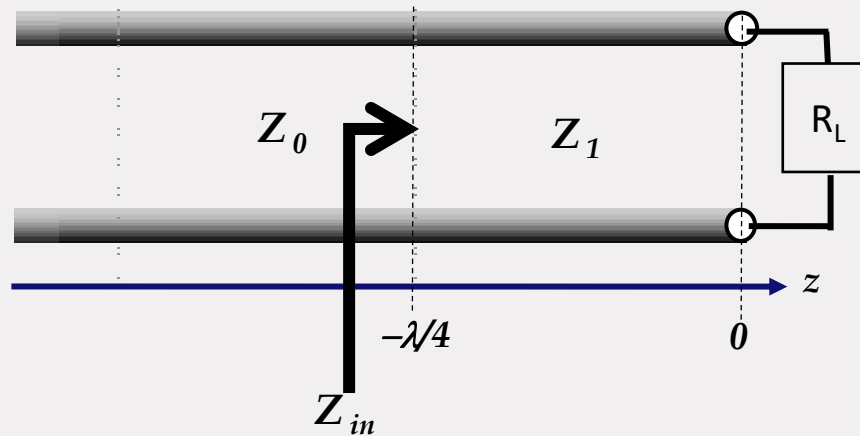
$S_{11}$  ( $=\Gamma$ ) versus frequency (0.5GHz – 3.0GHz; 251 points)

Linear, Logarithmic (dB) and Smith chart

Explain the results of these simulations



# Equations for quarter-wave transformer



Input impedance at  $z = -\lambda/4$ :

$$Z_{in} = Z_1 \frac{R_L + jZ_1 \tan \beta l}{Z_1 + jR_L \tan \beta l}$$

Now  $\beta l = (2\pi / \lambda)(\lambda / 4) = \pi / 2$

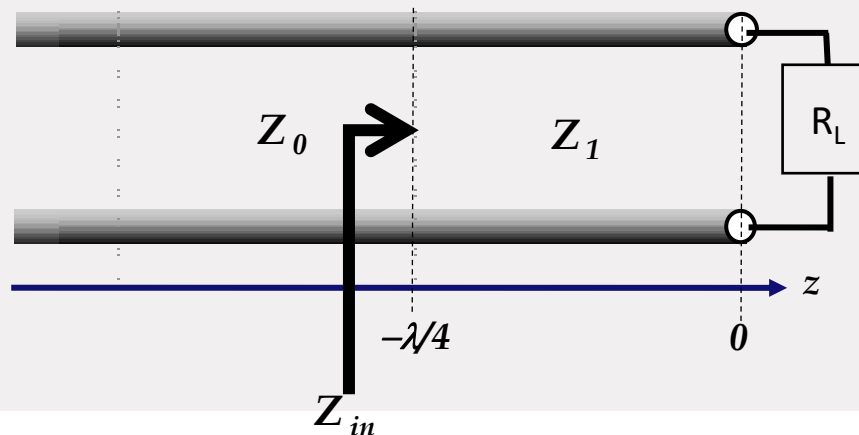
$$Z_{in} = \frac{Z_1^2}{R_L} \quad \longrightarrow \quad \Gamma = 0 \quad \text{if} \quad Z_1 = \sqrt{Z_0 R_L}$$

## Exercise 2 – Quarter-wave transformer

Consider a load resistance of  $Z_L = R_L = 100\ \Omega$  to be matched to a  $50\ \Omega$  line with a quarter-wave transformer. Design a quarter-wave transformer optimized for  $f = 1\ \text{GHz}$ . Assume a lossless transmission line.

Plot the reflection coefficient  $\Gamma$  ( $=S_{11}$ ) (linear and in dB) versus frequency. Frequency range:  $0.5\ \text{GHz} - 10\ \text{GHz}$  with 1051 points.

*Note: length is in meter. If you type “ $l = 100$ ” QUCS assumes  $l = 100\ \text{m}$  and if you type “ $l = 100\ \text{m}$ ” or “ $l = 100\ \text{mm}$ ” QUCS assumes  $l = 100\ \text{mm}$ .*

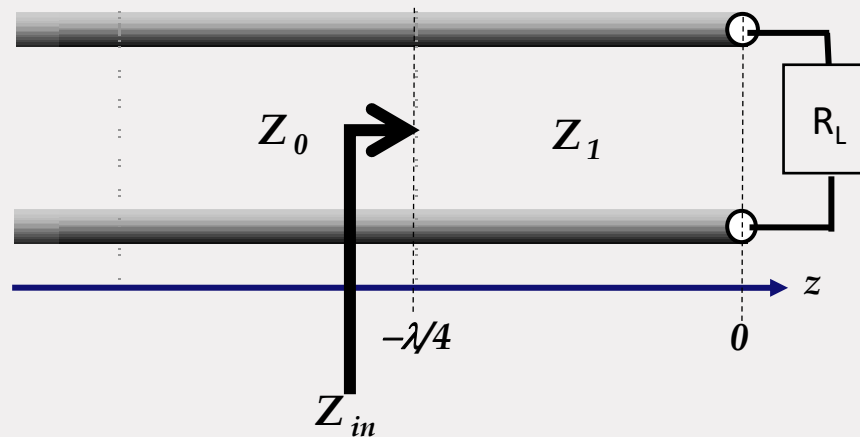


## Exercise 3 – Lossy Quarter-wave transformer

Use the same parameters as found in Exercise 2. In this case the line is lossy. Consider 2 situations:

- attenuation of line = 1 dB/m (in terms of power).
- attenuation of line = 50 dB/m (in terms of power).

Determine  $\Gamma$  ( $=S_{11}$ ) in dB versus frequency. Explain the results



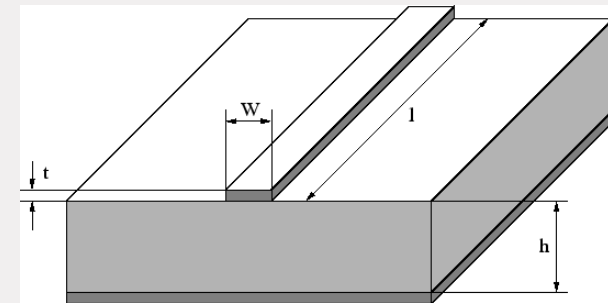
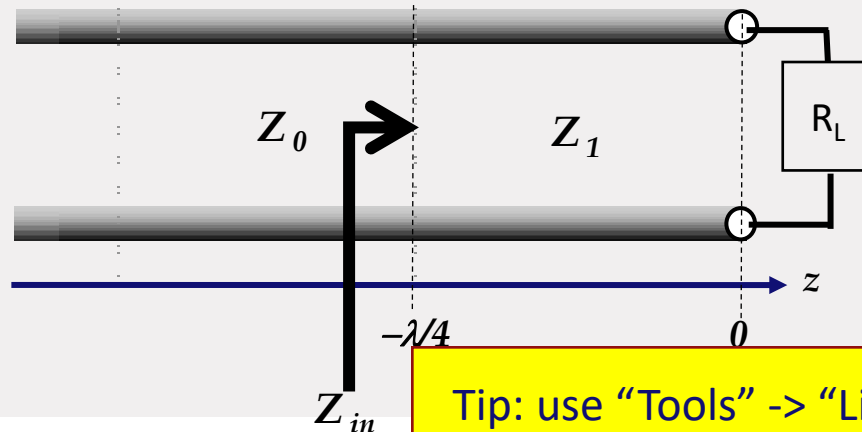


# Exercise 4 – microstrip Quarter-wave transformer

Now use a real microstrip transmission line in your transformer. The microstripline is printed on a PCB with relative permittivity  $\epsilon_r=4.2$ , loss tangent=0.001 and  $h=1.6$  mm,  $t=35$   $\mu\text{m}$ . Consider a load resistance of  $Z_L=R_L=100$  Ohm to be matched to a 50 Ohm line with a quarter-wave transformer.

To do:

1. Determine the width and length of the microstrip transformer optimized for  $f=1$  GHz.
2. Plot the reflection coefficient  $\Gamma$  ( $=S_{11}$ ) versus frequency.
3. Why is L shorter as compared to the ideal Transmission line case?



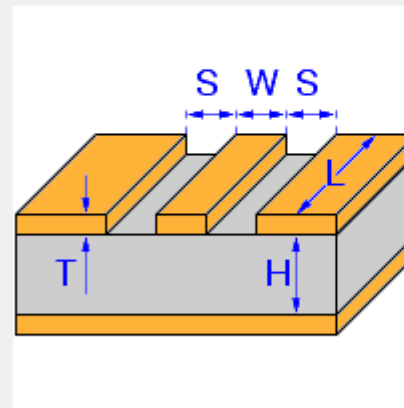
Tip: use “Tools” -> “Line calculation” to determine W,L

# Exercise 5 – Co-planar Quarter-wave transformer

Now use a real grounded co-planar transmission line in your transformer. The coplanar line is printed on a PCB with relative permittivity  $\epsilon_r=4.2$ , loss tangent=0.001 and  $h=1.6\text{mm}$ ,  $t=35\text{ }\mu\text{m}$ . Consider a load resistance of  $Z_L=R_L=100\text{ Ohm}$  to be matched to a ( $Z_0$ ) 50 Ohm line with a quarter-wave transformer.

The grounded coplanar line has a spacing to ground of 0.1 mm.

Grounded Coplanar



Er	4.2	NA
H	1.6	mm
T	35	um
Cond	4.1e+07	NA
Tand	0.001	NA
	0	NA
	0	NA
	0	NA
	0	NA

Component Parameters

Freq 1 GHz

Tip: use “Tools” -> “Line calculation” to determine W,L

# Exercise 5 – Co-planar Quarter-wave transformer

To do:

1. Determine the width and length of the grounded coplanar line transformer optimized for  $f=1$  GHz.
2. Plot the reflection coefficient  $\Gamma$  ( $=S_{11}$ ) versus frequency (linear and dB scale).
3. Which line is less lossy?  
The microstrip line of exercise 4, or this grounded coplanar line?

Tip: use “Tools” -> “Line calculation” to determine W,L

