

## 8. The Extended Kalman Filter (EKF)

- There exist many versions - we will cover one that appears often, the *hybrid* EKF

$$\dot{x}(t) = q(x(t), u(t), v(t), t) \quad \text{continuous process model}$$

$$z(kT) = h_k(x(kT), w(kT)) \quad \text{discrete measurement model}$$

- We will use  $[\cdot]$  notation in this lecture to denote discrete time indices for continuous time signals:

$$z[k] := z(kT), \quad T = \text{sampling time, a constant}$$

Now we can rewrite the measurement equation:

$$z[k] = h_k(x[k], w[k]) .$$

- Basic idea: linearize the equations about the current operating conditions, then use the standard Kalman Filter equations.

### 8.1 Sensor noise and measurement update (Step 2)

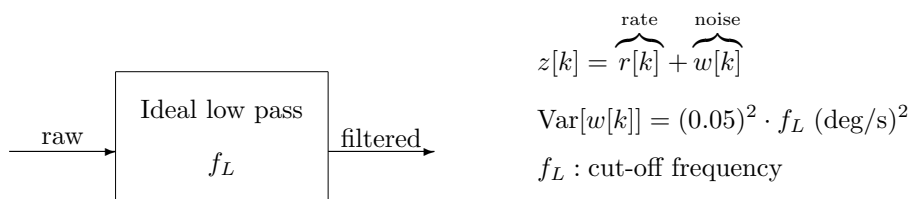
We first consider the measurement update, or what we have called Step 2 in the standard Kalman Filter, with a focus on the following two issues:

1. How to convert sensor noise specifications from continuous time to discrete time.
2. How to deal with the nonlinearity.

We consider Step 2 before Step 1 because it is conceptually easier to deal with Step 2 first.

#### 8.1.1 Sensor noise

What are typical sensor noise specifications, and in particular, how does one convert these specifications to elements of the variance matrix  $R$ ? We will explore this through a specific example. Consider a rate gyroscope, measuring angular rates in deg/s. A typical noise specification for a MEMS-based rate gyroscope, which can measure  $\pm 300$  deg/s, is  $0.05 \text{ deg/s}/\sqrt{\text{Hz}}$ . This is the **noise density**. Notice the units: they are the same as the measurement in question, divided by the square root of frequency.



- To obtain the variance, you simply multiple the square of the noise density by the ideal cut-off frequency. For example, if  $f_L = 100 \text{ Hz}$ , then  $\text{Var}[w[k]] = 0.25 \text{ (deg/s)}^2$ , and the standard deviation (the square root of the variance) is  $0.5 \text{ deg/s}$ . Said another way, to obtain the standard deviation, you simply multiply the noise density by the square root of the ideal cut-off frequency.

- As a reasonable approximation, if you do not filter the signal *before* sampling, it will have a very large variance.
- Many sensors will filter the signal for you.
- Are  $\{w[1], w[2], \dots\}$  independent?
  - If sampling frequency  $1/T \ll 2f_L$  (equivalently, if the Nyquist frequency is significantly smaller than the ideal cut-off frequency), this is a good assumption.
  - If, on the other hand,  $1/T \gg 2f_L$ , this is not a good assumption, and the noise is correlated. One must use more sophisticated Kalman Filter formulas that allow for correlated noise.
- *Trade-off*: the more aggressive the filter (small  $f_L$ ), the smaller the variance, but the bandwidth of the measurement is also reduced, and a phase delay is introduced in the measurement. Furthermore, the noise becomes correlated. You may have to explicitly take these facts into account. Often a good choice is  $f_L \approx 1/2T$ .

### 8.1.2 Measurement update (Step 2)

- Using simplified notation (no time indices)

$$z = h(x, w), \quad \text{where} \quad \begin{array}{l} E[w] = 0, \text{Var}[w] = R \\ \underbrace{E[x] = \hat{x}, \text{Var}[x] = P}_{\text{prior}} \end{array}$$

We are interested in  $E[x|z]$ ,  $\text{Var}[x|z]$ , the posterior.

- Let  $\tilde{x} = x - \hat{x}$ . If we *assume* that  $\tilde{x}$  and  $w$  are small (is this a good assumption?), to first order we have

$$z = h(\hat{x}, 0) + \tilde{z},$$

where

$$\begin{aligned} \tilde{z} &:= H\tilde{x} + \tilde{w}, \\ \tilde{w} &:= Mw, \quad H := \frac{\partial h}{\partial x}(\hat{x}, 0), \quad M := \frac{\partial h}{\partial w}(\hat{x}, 0) \\ E[\tilde{w}] &= 0, \text{Var}[\tilde{w}] = MRM^T \\ E[\tilde{x}] &= 0, \text{Var}[\tilde{x}] = P \end{aligned}$$

- We can now compute  $E[\tilde{x}|\tilde{z}]$ : given the measurement  $z$ , we can compute  $\tilde{z} = z - h(\hat{x}, 0)$ , and then use Step 2 of the standard Kalman Filter.
- Once we have computed  $E[\tilde{x}|\tilde{z}]$ , we can readily compute the following:

$$\begin{aligned} E[x|z] &= \hat{x} + E[\tilde{x}|\tilde{z}] \\ \text{Var}[x|z] &= \text{Var}[\tilde{x}|\tilde{z}] \end{aligned}$$

### Summary – Step 2

Let  $\hat{x}_p(k)$  be the prior mean,  $P_p(k)$  the prior variance. Let  $\text{Var}[w[k]] = R(k)$ . Then

$$\begin{aligned} K(k) &:= P_p(k)H^T(k) \left( H(k)P_p(k)H^T(k) + M(k)R(k)M^T(k) \right)^{-1} \\ \hat{x}_m(k) &:= \hat{x}_p(k) + K(k) (z[k] - h_k(\hat{x}_p(k), 0)) \\ P_m(k) &:= (I - K(k)H(k)) P_p(k) (I - K(k)H(k))^T + K(k)M(k)R(k)M^T(k)K^T(k) \\ H(k) &:= \frac{\partial h_k}{\partial x}(\hat{x}_p(k), 0), \quad M(k) := \frac{\partial h_k}{\partial w}(\hat{x}_p(k), 0). \end{aligned}$$

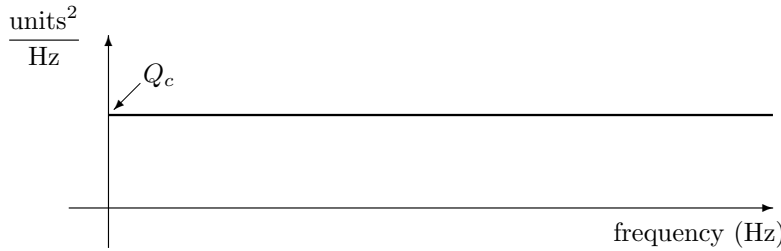
## 8.2 Process noise and prediction step (Step 1)

We must now deal with three issues:

1. How to directly handle continuous time process noise.
2. How to deal with a differential equation instead of a difference equation.
3. How to deal with the nonlinearity.

### 8.2.1 Process noise

We use the same underlying description that we used for continuous time sensor noise; instead of noise density, the power spectral density of the noise is often used, which is the square of the noise density.



- If we filtered  $v(t)$  with a low pass filter at a frequency  $f_L$ , the variance, at any instant of time, would be  $Q_c \cdot f_L$ .
- We don't have the luxury of filtering the process noise, however. In particular, we must deal directly with continuous time white noise, which has infinite variance<sup>1</sup>. One can show that

$$\mathbb{E}[v(t)] = 0, \quad \mathbb{E}[v(t)v^T(t + \tau)] = \delta(\tau)Q_c$$

### 8.2.2 Prediction step (Step 1)

Consider  $0 \leq t < T$  for now, we will generalize this later.

$$\begin{aligned} \dot{x}(t) &= q(x(t), u(t), v(t), t), & \mathbb{E}[x(0)] &= \hat{x}_0, \quad \text{Var}[x(0)] = P_0 \\ \mathbb{E}[v(t)] &= 0, \quad \mathbb{E}[v(t)v^T(t + \tau)] &= \delta(\tau)Q_c \\ v(t), x(0) &\text{ are independent.} \end{aligned}$$

What are  $\mathbb{E}[x(t)]$  and  $\text{Var}[x(t)]$ ?

- Let  $\hat{x}(t)$  solve

$$\dot{\hat{x}}(t) = q(\hat{x}(t), u(t), 0, t), \quad 0 \leq t < T, \quad \hat{x}(0) = \hat{x}_0$$

We will *assume* that  $\hat{x}(t) \approx \mathbb{E}[x(t)]$ . Equivalently, we assume that the expected value operator and function  $q(\cdot)$  commute (this can be a really bad assumption in the case of strong nonlinearities). This approximation is exact if  $q(\cdot)$  is linear.

- We next consider the variance. Let  $\tilde{x}(t) = x(t) - \hat{x}(t)$ . *Assuming* that  $\tilde{x}(t)$  and  $v(t)$  are small (is this a good assumption?)

$$\dot{\tilde{x}}(t) \approx A(t)\tilde{x}(t) + L(t)v(t)$$

$$A(t) := \frac{\partial q}{\partial x}(\hat{x}(t), u(t), 0, t) \quad L(t) := \frac{\partial q}{\partial v}(\hat{x}(t), u(t), 0, t)$$

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<sup>1</sup>To handle this rigorously is well beyond the scope of this class, and one should only view what follows as a rough guide.

Solution for small  $t$ :

$$\begin{aligned}\tilde{x}(t) &\approx \tilde{x}(0) + \int_0^t A(\tau)\tilde{x}(\tau) + L(\tau)v(\tau) d\tau \\ &\approx \tilde{x}(0) + tA(0)\tilde{x}(0) + L(0) \int_0^t v(\tau) d\tau + O(t^2),\end{aligned}$$

Note that  $P(t) := \text{Var}[x(t)] \approx \mathbb{E}[\tilde{x}(t)\tilde{x}^T(t)]$ , since  $\tilde{x}$  and  $x$  have the same variance, and  $\tilde{x}$  is assumed to be zero mean. We therefore have:

$$P(t) \approx P(0) + tA(0)P(0) + tP(0)A^T(0) + L(0) \underbrace{\left( \int_0^t \int_0^t \mathbb{E}[v(\tau)v^T(s)] d\tau ds \right)}_{tQ_c} L^T(0) + O(t^2)$$

Taking the limit as  $t \rightarrow 0$ :

$$\dot{P}(t) = A(t)P(t) + P(t)A^T(t) + L(t)Q_cL^T(t), \quad P(0) = P_0$$

We can thus solve for  $P(t)$  by solving the above matrix differential equation.

### Summary – Step 1

- Let  $\hat{x}_m(k-1)$ ,  $P_m(k-1)$  be given from Step 2.
- Solve

$$\begin{aligned}\dot{\hat{x}}(t) &= q(\hat{x}(t), u(t), 0, t) \quad (k-1)T \leq t \leq kT \\ \hat{x}((k-1)T) &= \hat{x}_m(k-1) \\ u(t) &\text{ is known}\end{aligned}$$

Then  $\hat{x}_p(k) := \hat{x}(kT)$ .

- Solve

$$\begin{aligned}\dot{P}(t) &= A(t)P(t) + P(t)A^T(t) + L(t)Q_cL^T(t) \quad (k-1)T \leq t \leq kT \\ P((k-1)T) &= P_m(k-1)\end{aligned}$$

$$A(t) = \frac{\partial q}{\partial x}(\hat{x}(t), u(t), 0, t), \quad L(t) = \frac{\partial q}{\partial v}(\hat{x}(t), u(t), 0, t)$$

Set  $P_p(k) := P(kT)$ .

- Both the mean and variance calculations require solving a differential equation. One can make various approximations to reduce computation. For example:
  - If  $T$  is small, one can approximate the mean calculation as follows:

$$\hat{x}_p(k) = \hat{x}_m(k-1) + Tq(\hat{x}_m(k-1), u[k-1], 0, T(k-1)).$$

More sophisticated approximations can be used. For example, instead of using  $u[k-1]$ , one could use  $\frac{1}{2}(u[k-1] + u[k])$ ; instead of using  $T(k-1)$ , one could use  $T(k - \frac{1}{2})$ . If more accuracy is required, one could use Runge-Kutta methods.

- A similar approach as above can be used to approximate the variance update.